

### **EVALUATION KIT AVAILABLE Gigabit Multimedia Serial Link** Serializer with LVDS System Interface

### **General Description**

The MAX9249 serializer with LVDS system interface utilizes Maxim's Gigabit multimedia serial link (GMSL) technology. The MAX9249 serializer pairs with any GMSL deserializer to form a complete digital serial link for joint transmission of high-speed video, audio, and control data.

The MAX9249 allows a maximum serial payload data rate of 2.5Gbps for a 15m shielded twisted-pair (STP) cable. The serializer operates up to a maximum clock rate of 104MHz (3-channel LVDS) or 78MHz (4-channel LVDS). This serial link supports display panels from QVGA (320 x 240) to WXGA (1280 x 800) and higher with 24-bit color.

The 3-channel mode handles three lanes of LVDS data (21 bits), UART control signals, and three audio signals. The 4-channel mode handles four lanes of LVDS data (28 bits), UART control signals, three audio signals, and/or up to three auxiliary parallel inputs. The three audio inputs form a standard I2S interface, supporting sample rates from 8kHz to 192kHz and audio word lengths of 4 to 32 bits. The embedded control channel forms a full-duplex, differential, 100kbps to 1Mbps UART link between the serializer and deserializer. The electronic control unit (ECU), or microcontroller (µC), can be located on the MAX9249 side of the link (typical for video display), on the deserializer side of the link (typical for image sensing), or on both sides. In addition, the control channel enables ECU/µC control of peripherals on the remote side, such as backlight control, grayscale Gamma correction, camera module, and touch screen. Base-mode communication with peripherals uses either I2C or the GMSL UART format. A bypass mode enables full-duplex communication using custom UART formats.

The MAX9249 serializer driver preemphasis, along with the channel equalizer on the GMSL deserializer, extends the link length and enhances the link reliability. Spread spectrum is available on the MAX9249 to reduce EMI on the serial link and the parallel output of the GMSL deserializer. The serial output complies with ISO 10605 and IEC 61000-4-2 ESD protection standards.

The core supply for the MAX9249 is 1.8V. The I/O supply ranges from 1.8V to 3.3V. The MAX9249 is available in a 48-pin TQFP package (7mm x 7mm) with an exposed pad. Electrical performance is guaranteed over the -40°C to +105°C automotive temperature range.

#### **Features**

- Pairs with Any GMSL Deserializer
- ♦ 2.5Gbps Payload Rate AC-Coupled Serial Link with 8B/10B Line Coding
- ♦ Supports Up to WXGA (1280 x 800) with 24-Bit
- ♦ 8.33MHz to 104MHz (3-Channel LVDS) or 6.25MHz to 78MHz (4-Channel LVDS) Input Clock
- ♦ 4-Bit to 32-Bit Word Length, 8kHz to 192kHz I2S **Audio Channel Supports High-Definition Audio**
- **♦** Embedded Half-/Full-Duplex Bidirectional Control Channel (100kbps to 1Mbps)
- ◆ Interrupt Supports Touch-Screen Functions for **Display Panels**
- ♦ Remote-End I<sup>2</sup>C Master for Peripherals
- Preemphasis Line Driver
- Programmable Spread Spectrum on the Serial **Outputs for Reduced EMI**
- ◆ Automatic Data-Rate Detection Allows "On-the-Fly" Data-Rate Change
- ◆ Input Clock PLL Jitter Attenuator
- ♦ Built-In PRBS Generator for BER Testing of the **Serial Link**
- ◆ Line-Fault Detector Detects Serial Link Shorts to Ground, Battery, or Open Link
- ◆ ISO 10605 and IEC 61000-4-2 ESD Protection
- **◆** -40°C to +105°C Operating Temperature Range
- 1.8V to 3.3V I/O, 1.8V Core, and 3.3V LVDS **Supplies**
- Patent Pending

### **Applications**

High-Resolution Automotive Navigation Rear-Seat Infotainment Megapixel Camera Systems

### **Ordering Information**

PART	TEMP RANGE	PIN-PACKAGE		
MAX9249GCM/V+	-40°C to +105°C	48 TQFP-EP*		
MAX9249GCM/V+T	-40°C to +105°C	48 TQFP-EP*		

N denotes an automotive qualified part.

- +Denotes a lead(Pb)-free/RoHS-compliant package.
- \*EP = Exposed pad.

T = Tape and reel.

#### **ABSOLUTE MAXIMUM RATINGS**

AVDD to AGND	0.5V to +1.9V
LVDSVDD to AGND	0.5V to +3.9V
DVDD to GND	0.5V to +1.9V
IOVDD to GND	0.5V to +3.9V
Any Ground to Any Ground	0.5V to +0.5V
RXIN, RXCLKIN_ to AGND	0.5V to +3.9V
OUT+, OUT- to AGND	0.5V to +1.9V
LMN_ to AGND (15mA current limit)	0.5V to +3.9V
All Other Pins to Any Ground	$0.5V$ to $(V_{IOVDD} + 0.5V)$
OUT+, OUT- Short Circuit to Ground or	Supply Continuous
Continuous Power Dissipation ( $TA = +70$	O°C)
48-Pin TQFP (derate 36.2mW/°C above	ve +70°C)2898.6mW
ESD Protection	
Human Body Model (R <sub>D</sub> = 1.5k $\Omega$ , Cs	= 100pF)
(RXIN, RXCLKIN_, OUT+, OUT-) to	AGND±8kV
All Other Pins to GND	±3kV

IEC 61000-4-2 (RD = 330 $\Omega$ , Cs = 150pF) Contact Discharge	
(RXIN, RXCLKIN_) to AGND	±4kV
(OUT+, OUT-) to AGND	
Air Discharge	
(RXIN, RXCLKIN_) to AGND	±8kV
(OUT+, OUT-) to AGND	±12kV
ISO 10605 (RD = $2k\Omega$ , Cs = 330pF)	
Contact Discharge	
(RXIN, RXCLKIN_) to AGND	
(OUT+, OUT-) to AGND	±10kV
Air Discharge	
(RXIN, RXCLKIN_) to AGND	±20kV
(OUT+, OUT-) to AGND	
Operating Temperature Range	
Junction Temperature	+150°C
Storage Temperature Range	
Lead Temperature (soldering, 10s)	
Soldering Temperature (reflow)	
= : , , ,	

### PACKAGE THERMAL CHARACTERISTICS (Note 1)

48 TQFP-EP

Junction-to-Ambient Thermal Resistance ( $\theta$ JA)......27.6°C/W Junction-to-Case Thermal Resistance ( $\theta$ JC).............2°C/W

**Note 1:** Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to <a href="https://www.maxim-ic.com/thermal-tutorial">www.maxim-ic.com/thermal-tutorial</a>.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### DC ELECTRICAL CHARACTERISTICS

 $(V_{DVDD} = V_{AVDD} = 1.7V \text{ to } 1.9V, V_{LVDSVDD} = 3.0V \text{ to } 3.6V, V_{IOVDD} = 1.7V \text{ to } 3.6V, R_L = 100\Omega \pm 1\%$  (differential),  $T_A = -40^{\circ}\text{C}$  to  $+105^{\circ}\text{C}$ , unless otherwise noted. Differential input voltage  $|V_{ID}| = 0.1V \text{ to } 1.2V$ , input common-mode voltage  $V_{CM} = |V_{ID}/2|$  to  $2.4V - |V_{ID}/2|$ . Typical values are at  $V_{DVDD} = V_{AVDD} = V_{IOVDD} = 1.8V$ ,  $V_{LVDSVDD} = 3.3V$ ,  $V_{LVDSVDD} = 1.8V$ .

PARAMETER	SYMBOL		CONDITIONS	MIN	TYP	MAX	UNITS	
SINGLE-ENDED INPUTS (PWDN,	SSEN, BWS	, DRS, MS, CD	S, AUTOS, SD/CNTLO, SCK,	WS, CNTL	_)			
High-Level Input Voltage	Viita	PWDN, SSEN, I	BWS, DRS, MS, CDS, AUTOS	0.65 x VIOVDD			V	
Inigri-Level Iriput voltage	VIH1	SD/CNTL0, SCK, WS, CNTL_		0.7 x VIOVDD			V	
Low-Level Input Voltage	VIL1					0.35 x VIOVDD	V	
Input Current	l <sub>IN1</sub>	$V_{IN} = 0$ to $V_{IOV}$	'DD	-10		+10	μΑ	
Input Clamp Voltage	VCL	ICL = -18mA				-1.5	V	
SINGLE-ENDED OUTPUT (INT)								
High-Level Output Voltage	VOH1	I <sub>OH</sub> = -2mA		V <sub>IOVDD</sub> - 0.2			٧	
Low-Level Output Voltage	VOL1	IOL = 2mA				0.2	V	
Outout Chart Circuit Current	lon	\\c_ 0\\	V <sub>IOVDD</sub> = 3.0V to 3.6V	16	35	64	mΛ	
Output Short-Circuit Current	los	VO = 0V	$V_{IOVDD} = 1.7V \text{ to } 1.9V$	3	12	21	mA	

### DC ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DVDD} = V_{AVDD} = 1.7V \text{ to } 1.9V, V_{LVDSVDD} = 3.0V \text{ to } 3.6V, V_{IOVDD} = 1.7V \text{ to } 3.6V, R_L = 100\Omega \pm 1\%$  (differential),  $T_A = -40^{\circ}\text{C}$  to  $+105^{\circ}\text{C}$ , unless otherwise noted. Differential input voltage  $|V_{ID}| = 0.1V \text{ to } 1.2V$ , input common-mode voltage  $V_{CM} = |V_{ID}/2|$  to  $2.4V - |V_{ID}/2|$ . Typical values are at  $V_{DVDD} = V_{AVDD} = V_{IOVDD} = 1.8V$ ,  $V_{LVDSVDD} = 3.3V$ ,  $T_A = +25^{\circ}\text{C}$ .)

PARAMETER	SYMBOL		MIN	TYP	MAX	UNITS	
I <sup>2</sup> C AND UART I/O, OPEN-DRAIN	OUTPUT (R	X/SDA, TX/SC	L, <b>LFLT)</b>				
High-Level Input Voltage	V <sub>IH2</sub>			0.7 x VIOVDD			V
Low-Level Input Voltage	V <sub>IL2</sub>					0.3 x VIOVDD	V
Input Current	liN2	VIN = 0 to VIO	VDD (Note 2)	-110		+5	μΑ
Low-Level Open-Drain Output	1/01.0	101 2mm A	$V_{IOVDD} = 1.7V \text{ to } 1.9V$			0.4	V
Voltage	VOL2	IOL = 3mA	$V_{IOVDD} = 3.0V \text{ to } 3.6V$			0.3	V
DIFFERENTIAL OUTPUT (OUT+,	OUT-)						
		Preemphasis of	off (Figure 1)	300	400	500	
Differential Output Voltage	Vod	3.3dB preemp (Figure 2)	hasis setting, V <sub>OD(P)</sub>	350		610	mV
		3.3dB deemph (Figure 2)	nasis setting, V <sub>OD(D)</sub>	240		425	
Change in VOD Between Complementary Output States	ΔV <sub>OD</sub>					15	mV
Output Offset Voltage (VOUT+ + VOUT-)/2 = VOS	Vos	Preemphasis of	off	1.1	1.4	1.56	V
Change in Vos Between Complementary Output States	ΔV <sub>OS</sub>					15	mV
		Vout+ or Vou	T- = 0V	-60			
Output Short-Circuit Current	los	Vout+ or Vou	T- = 1.9V			25	mA
Magnitude of Differential Output Short-Circuit Current	losp	V <sub>OD</sub> = 0V				25	mA
Output Termination Resistance (Internal)	Ro	From OUT+, C	OUT- to VAVDD	45	54	63	Ω
REVERSE CONTROL-CHANNEL	RECEIVER (	OUT+, OUT-)					
High Switching Threshold	VCHR					27	mV
Low Switching Threshold	VCLR			-27			mV
LINE-FAULT DETECTION INPUT	(LMN_)						
Short-to-GND Threshold	VTG	Figure 3				0.3	V
Normal Thresholds	VTN	Figure 3		0.57		1.07	V
Open Thresholds	V <sub>TO</sub>	Figure 3		1.45		VIO + 60mV	V
Open Input Voltage	VIO	Figure 3		1.47		1.75	V
Short-to-Battery Threshold	VTE	Figure 3		2.47			
LVDS INPUTS (RXIN, RXCLKI							
Differential Input High Threshold	VTH					50	mV
Differential Input Low Threshold	V <sub>T</sub> L			-50			mV

### DC ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DVDD} = V_{AVDD} = 1.7V \text{ to } 1.9V, V_{LVDSVDD} = 3.0V \text{ to } 3.6V, V_{IOVDD} = 1.7V \text{ to } 3.6V, R_L = 100\Omega \pm 1\% \text{ (differential)}, T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}$ , unless otherwise noted. Differential input voltage  $|V_{ID}| = 0.1V \text{ to } 1.2V$ , input common-mode voltage  $V_{CM} = |V_{ID}/2| \text{ to } 2.4V - |V_{ID}/2|$ . Typical values are at  $V_{DVDD} = V_{AVDD} = V_{IOVDD} = 1.8V$ ,  $V_{LVDSVDD} = 3.3V$ ,  $V_{LVDSVDD} = 1.8V$ ,  $V_{LVDSVDD} =$ 

PARAMETER	SYMBOL		CONDITIONS	MIN	TYP	MAX	UNITS
Input Differential Termination Resistance	RTERM			85	110	135	Ω
Input Current	IIN+, IIN-	PWDN = high	or low, IN+ and IN- are shorted	-25		+25	μΑ
Power-Off Input Current	I <sub>INO+</sub> , I <sub>INO-</sub>	VAVDD = VDV	DD = VIOVDD = 0V	-40		+40	μΑ
POWER SUPPLY							
			fRXCLKIN_ = 16.6MHz		125	165	
Worst-Case Supply Current	Iwcs	BWS = GND	frxclkin_ = 33.3MHz		135	175	mA
(Figure 4)			fRXCLKIN_ = 66.6MHz		150	190	IIIA
			fRXCLKIN_ = 104MHz		175	220	
Sleep-Mode Supply Current	Iccs	LVDS inputs are not driven			45	125	μΑ
Power-Down Supply Current	Iccz	PWDN = GND	, LVDS inputs are not driven		5	80	μΑ

#### **AC ELECTRICAL CHARACTERISTICS**

 $(V_{DVDD} = V_{AVDD} = 1.7V \text{ to } 1.9V, V_{IOVDD} = 1.7V \text{ to } 3.6V, R_L = 100\Omega \pm 1\% \text{ (differential)}, T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}, \text{ unless otherwise noted.}$  Differential input voltage  $|V_{ID}| = 0.15V \text{ to } 1.2V, \text{ input common-mode voltage } V_{CM} = |V_{ID}/2| \text{ to } 2.4V - |V_{ID}/2|.$  Typical values are at  $V_{DVDD} = V_{AVDD} = V_{IOVDD} = 1.8V, V_{LVDSVDD} = 3.3V, T_A = +25^{\circ}\text{C.})$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
CLOCK INPUT (RXCLKIN_)						
		BWS = GND, VDRS = VIOVDD	8.33	3 16.66		
Clock Frequency	fpyoutth	BWS = GND, DRS = GND	16.66		104	MHz
	fRXCLKIN_	V <sub>BWS</sub> = V <sub>IOVDD</sub> , V <sub>DRS</sub> = V <sub>IOVDD</sub>	6.25		12.5	IVITZ
		VBWS = VIOVDD, DRS = GND	12.5		78	
I <sup>2</sup> C/UART PORT TIMING (Note 3)						
Output Rise Time	t <sub>R</sub>	30% to 70%, $C_L$ = 10pF to 100pF, 1k $\Omega$ pullup to IOVDD	20		150	ns
Output Fall Time	tF	70% to 30%, $C_L$ = 10pF to 100pF, $1k\Omega$ pullup to IOVDD	20		150	ns
Input Setup Time	tset	I <sup>2</sup> C only (Figure 5)	100			ns
Input Hold Time	tHOLD	I <sup>2</sup> C only (Figure 5)	0			ns
SWITCHING CHARACTERISTICS	(Note 3)					
Differential Output Rise/Fall Time	t <sub>R</sub> , t <sub>F</sub>	20% to 80%, $V_{OD} \ge 400$ mV, $R_L = 100\Omega$ , serial-bit rate = 3.125Gbps (Note 3)		90	150	ps
Total Serial Output Jitter	tTSOJ1	3.125Gbps PRBS signal, measured at VoD = 0V differential, preemphasis disabled (Figure 6)		0.25		UI
Deterministic Serial Output Jitter	tDSOJ2	3.125Gbps PRBS signal		0.15		UI
CNTL_ Input Setup Time	tset	CNTL_ (Figure 7)	3			ns
CNTL_ Input Hold Time	tHOLD	CNTL_ (Figure 7)	1.5			ns
RXIN Skew Margin	trskm	Figure 8	0.3	·		UI

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### **AC ELECTRICAL CHARACTERISTICS (continued)**

 $(VDVDD = VAVDD = 1.7V \text{ to } 1.9V, VIOVDD = 1.7V \text{ to } 3.6V, R_L = 100\Omega \pm 1\% \text{ (differential)}, T_A = -40^{\circ}\text{C to } +105^{\circ}\text{C}, unless otherwise noted.}$  Differential input voltage  $|V_{ID}| = 0.15V \text{ to } 1.2V$ , input common-mode voltage  $|V_{CM}| = |V_{ID}/2| \text{ to } 2.4V - |V_{ID}/2|$ . Typical values are at  $|V_{DVDD}| = |V_{AVDD}| = |V_{AVDD}| = 1.8V$ ,  $|V_{LVDSVDD}| = 3.3V$ ,  $|V_{LVDSVDD}| = 1.8V$ 

PARAMETER	SYMBOL		CONDITIONS	MIN	TYP MAX	UNITS
Carializar Dalay (Nata 4)	ton	Figure 0	Spread spectrum enabled		2950	Dito
Serializer Delay (Note 4)	tsd	Figure 9	Spread spectrum disabled		390	Bits
Link Start Time	tLOCK	Figure 10			3.5	ms
Power-Up Time	tpu	Figure 11			3.5	ms
I <sup>2</sup> S INPUT TIMING						
WS Frequency	fws	Table 3		8	192	kHz
Sample Word Length	nws	Table 3		4	32	Bits
SCK Frequency	fsck	fsck = fw	s x nws x 2	(8 x 4) x 2	(192 x 32) x 2	kHz
SCK Clock High Time (Note 3)	tHC	VSCK ≥ VII	H, tSCK = 1/fSCK	0.35 x tsck		ns
SCK Clock Low Time (Note 3)	tLC	VSCK ≤ VI	L, tSCK = 1/fSCK	0.35 x tsck		ns
SD/CNTL0, WS Setup Time	tset	Figure 12	(Note 3)	2		ns
SD/CNTL0, WS Hold Time	tHOLD	Figure 12	(Note 3)	2		ns

Note 2: Minimum I<sub>IN</sub> due to voltage drop across the internal pullup resistor.

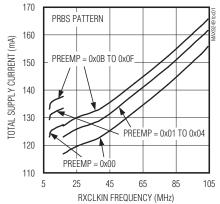
**Note 3:** Not production tested.

Note 4: Bit time = 
$$\frac{1}{30 \times f_{RXCLKIN}}$$
 (BWS = 0), =  $\frac{1}{40 \times f_{RXCLKIN}}$  (BWS =  $V_{IOVDD}$ )

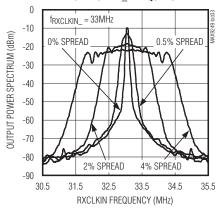
### **Typical Operating Characteristics**

(VDVDD = VAVDD = VIOVDD = 1.8V, VLVDSVDD = 3.3V, TA = +25°C, unless otherwise noted.)

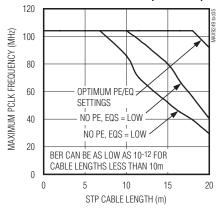




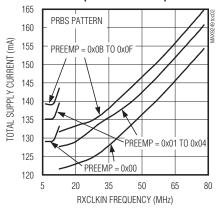
### OUTPUT POWER SPECTRUM vs. RXCLKIN\_ FREQUENCY



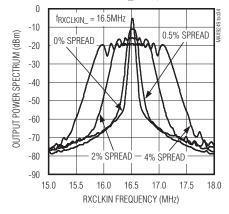
MAXIMUM PCLK FREQUENCY vs. STP CABLE LENGTH (BER < 10-9)



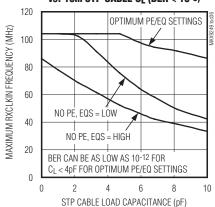
#### TOTAL SUPPLY CURRENT vs. RXCLKIN\_ FREQUENCY (4-CHANNEL MODE)



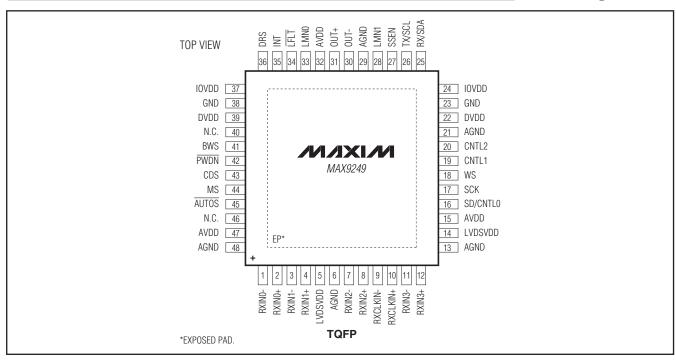
### OUTPUT POWER SPECTRUM vs. RXCLKIN\_ FREQUENCY



### MAXIMUM RXCLKIN\_ FREQUENCY vs. 10m STP CABLE C<sub>L</sub> (BER < 10-9)



### Pin Configuration



### **Pin Description**

PIN	NAME	FUNCTION
1–4, 7, 8, 11, 12	RXIN, RXIN_+	Differential LVDS Data Inputs. Set BWS = low (3-channel mode) to use RXIN0_ to RXIN2 Set BWS = high (4-channel mode) to use RXIN0_ to RXIN3
5, 14	LVDSVDD	3.3V LVDS Power Supply. Bypass LVDSVDD to AGND with 0.1µF and 0.001µF capacitors as close as possible to the device with the smaller value capacitor closest to LVDSVDD.
6, 13, 21, 29, 48	AGND	Analog Ground
9, 10	RXCLKIN-, RXCLKIN+	LVDS Input for the LVDS Clock
15, 32, 47	AVDD	1.8V Analog Power Supply. Bypass AVDD to AGND with 0.1µF and 0.001µF capacitors as close as possible to the device with the smaller value capacitor closest to AVDD.
16	SD/CNTL0	I <sup>2</sup> S Serial-Data Input with Internal Pulldown to GND. Disable I <sup>2</sup> S to use SD/CNTL0 as an additional input.
17	SCK	I <sup>2</sup> S Serial-Clock Input with Internal Pulldown to GND
18	WS	I <sup>2</sup> S Word-Select Input with Internal Pulldown to GND
19	CNTL1	Control Input 1 with Internal Pulldown to GND. Data is latched every RXCLKIN_ cycle (Figure 7). CNTL1 is not available in 3-channel mode. Drive BWS high (4-channel mode) to use this input. CNTL1 or RES (RES from VESA Standard Panel Specification) is mapped to DIN27 (see the Reserved Bit (RES) section).

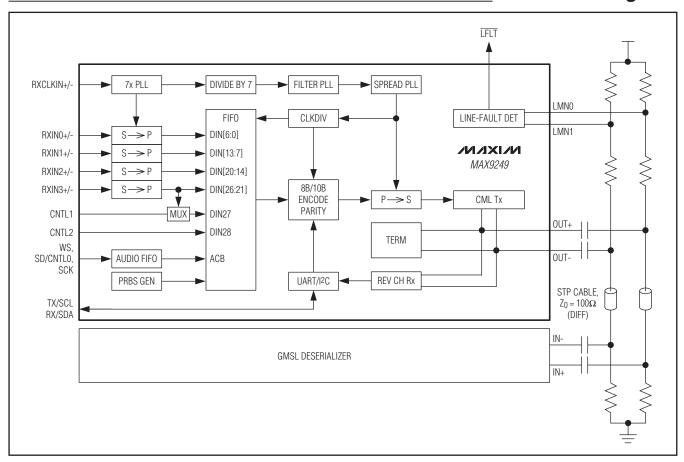
### Pin Description (continued)

PIN	NAME	FUNCTION
20	CNTL2	Control Input 2 with Internal Pulldown to GND. Data is latched every RXCLKIN_ cycle (Figure 7). CNTL2 is not available in 3-channel mode. Drive BWS high (4-channel mode) to use this input. CNTL2 is mapped to DIN28.
22, 39	DVDD	1.8V Digital Power Supply. Bypass DVDD to GND with 0.1µF and 0.001µF capacitors as close as possible to the device with the smaller value capacitor closest to DVDD.
23, 38	GND	Digital and I/O Ground
24, 37	IOVDD	I/O Supply Voltage. 1.8V to 3.3V logic I/O power supply. Bypass IOVDD to GND with 0.1µF and 0.001µF capacitors as close as possible to the device with the smallest value capacitor closest to IOVDD.
25	RX/SDA	Receive/Serial Data. UART receive or I <sup>2</sup> C serial-data input/output with internal $30k\Omega$ pullup to IOVDD. In UART mode, RX/SDA is the Rx input of the MAX9249's UART. In I <sup>2</sup> C mode, RX/SDA is the SDA input/output of the MAX9249's I <sup>2</sup> C master.
26	TX/SCL	Transmit/Serial Clock. UART transmit or I <sup>2</sup> C serial-clock output with internal $30k\Omega$ pullup to IOVDD. In UART mode, TX/SCL is the Tx output of the MAX9249's UART. In I <sup>2</sup> C mode, TX/SCL is the SCL output of the MAX9249's I <sup>2</sup> C master.
27	SSEN	Spread-Spectrum Enable. Serial link spread-spectrum enable input requires external pulldown or pullup resistors. The state of SSEN latches upon power-up or when resuming from power-down mode (PWDN = low). Set SSEN = high for ±0.5% spread spectrum on the serial link. Set SSEN = low to use the serial link without spread spectrum.
28	LMN1	Line-Fault Monitor Input 1 (see Figure 3 for details)
30, 31	OUT-, OUT+	Differential CML Output+/ Differential outputs of the serial link.
33	LMN0	Line-Fault Monitor Input 0 (see Figure 3 for details)
34	<u>LFLT</u>	Line Fault. Active-low, open-drain line-fault output with a $60k\Omega$ internal pullup resistor. $\overline{LFLT} = low$ indicates a line fault. $\overline{LFLT}$ is high impedance when $\overline{PWDN} = low$ .
35	INT	Interrupt Output to Indicate Remote Side Requests. INT = low upon power-up and when $\overline{PWDN}$ = low. A transition on the INT input of the GMSL deserializer toggles the MAX9249's INT output.
36	DRS	Data-Rate Select. Data-rate range-selection input requires external pulldown or pullup resistors. Set DRS = high for RXCLKIN_ frequencies of 8.33MHz to 16.66MHz (3-channel mode) or 6.25MHz to 12.5MHz (4-channel mode). Set DRS = low for RXCLKIN_ frequencies of 16.66MHz to 104MHz (3-channel mode) or 12.5MHz to 78MHz (4-channel mode).
40, 46	N.C.	Internally Not Connected. Connect to GND or leave unconnected.
41	BWS	Bus-Width Select. Input width selection requires external pulldown or pullup resistors. Set BWS = low for 3-channel mode. Set BWS = high for 4-channel mode.
42	PWDN	Power-Down. Active-low power-down input requires external pulldown or pullup resistors.
43	CDS	Control Direction Selection. Control link direction selection input requires external pulldown or pullup resistors. Set CDS = low for $\mu$ C use on the MAX9249 side of the serial link. Set CDS = high for $\mu$ C use on the GMSL deserializer side of the serial link.
44	MS	Mode Select. Control link mode-selection input requires external pulldown or pullup resistors. Set MS = low to select base mode. Set MS = high to select the bypass mode.

### Pin Description (continued)

PIN	NAME	FUNCTION
45	AUTOS	Autostart Setting. Active-low power-up mode-selection input requires external pulldown or pullup resistors. Set AUTOS = high to power up the device with no link active. Set AUTOS = low to have the MAX9249 power up the serial link with autorange detection (see Tables 8 and 9).
_	EP	Exposed Pad. EP internally connected to AGND. <b>MUST</b> externally connect EP to the AGND plane for proper thermal and electrical performance.

### **Functional Diagram**



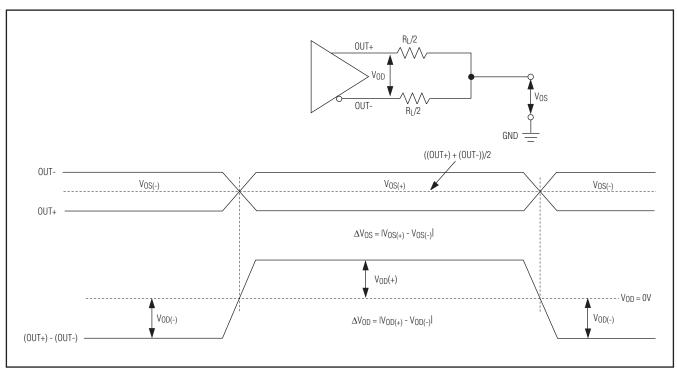


Figure 1. Serial-Output Parameters

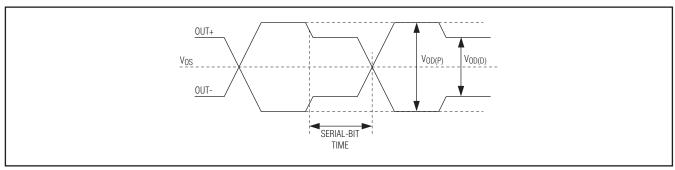


Figure 2. Output Waveforms at OUT+ and OUT-

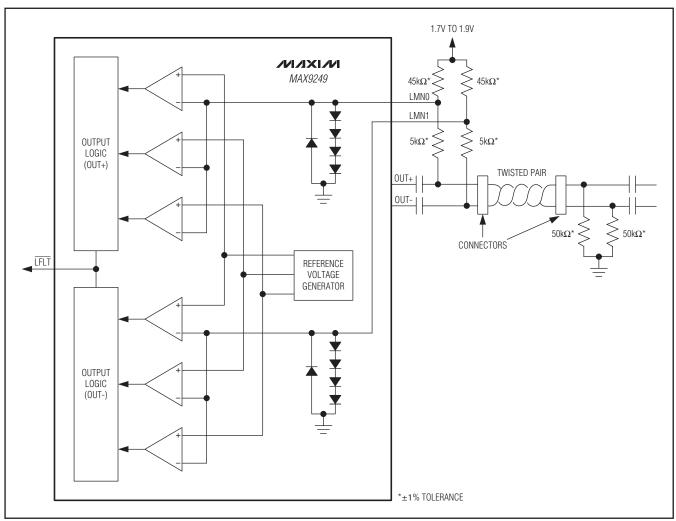


Figure 3. Line-Fault Detector Circuit

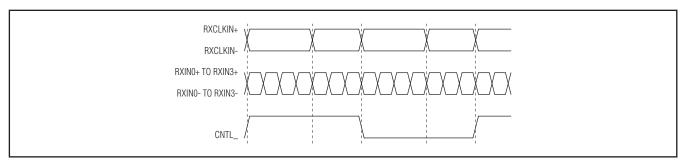


Figure 4. Worst-Case Pattern Input

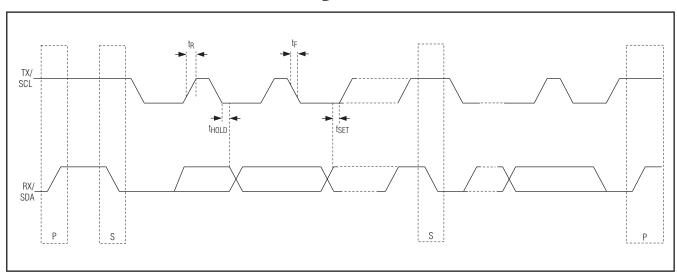


Figure 5. I<sup>2</sup>C Timing Parameters

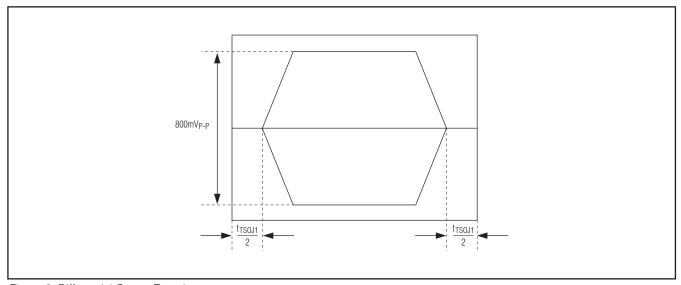


Figure 6. Differential Output Template

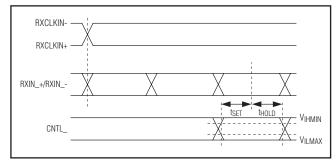


Figure 7. Input Setup-and-Hold Times

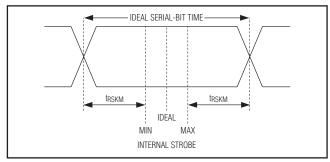


Figure 8. LVDS Receiver Input Skew Margin

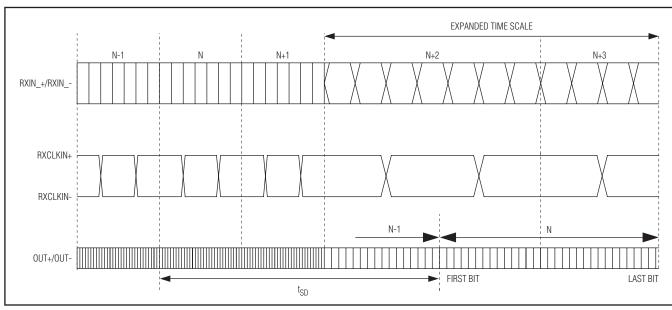


Figure 9. Serializer Delay

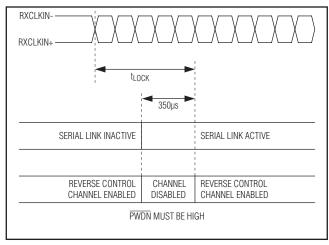


Figure 10. Link Startup Time

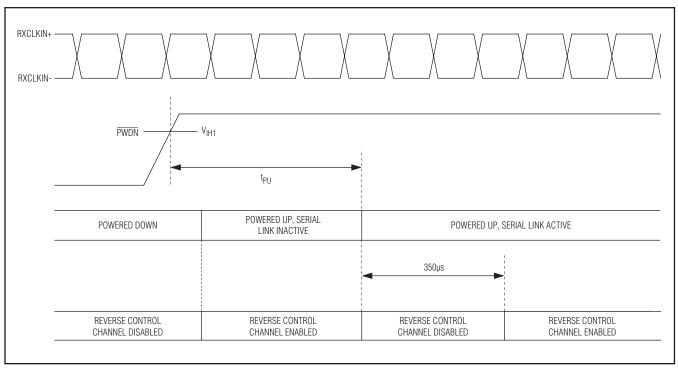


Figure 11. Power-Up Delay

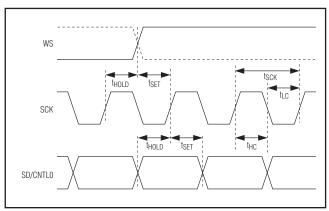


Figure 12. Input I<sup>2</sup>S Timing Parameters

### **Detailed Description**

The MAX9249 serializer with LVDS system interface utilizes Maxim's GMSL technology. The MAX9249 serializer pairs with any GMSL deserializer to form a complete digital serial link for joint transmission of high-speed video, audio, and control data.

The MAX9249 allows a maximum serial payload data rate of 2.5Gbps for a greater than 15m STP cable. The serializer operates up to a maximum clock of 104MHz for a 3-channel LVDS input or 78MHz for a 4-channel LVDS input. This serial link supports display panels from QVGA  $(320 \times 240)$  up to WXGA  $(1280 \times 800)$  with 24-bit color.

The 3-channel mode handles three lanes of LVDS data (21 bits), UART control signals, and three audio signals. The 4-channel mode handles four lanes of LVDS data (28 bits), UART control signals, three audio signals, and/or up to three auxiliary parallel inputs. The three audio inputs form a standard I²S interface, supporting sample rates from 8kHz to 192kHz and audio word lengths of 4 to 32 bits. The embedded control channel forms a full-duplex, differential, 100kbps to 1Mbps UART link between the serializer and deserializer. The ECU, or  $\mu\text{C}$ , can be located on the MAX9249 side of the link (typical

for video display), on the deserializer side of the link (typical for image sensing), or on both sides. In addition, the control channel enables ECU/µC control of peripherals in the remote side, such as backlight control, grayscale Gamma correction, camera module, and touch screen. Base-mode communication with peripherals uses either I2C or the GMSL UART format. A bypass mode enables full-duplex communication using custom UART formats.

The MAX9249 serializer driver preemphasis, along with the channel equalizer on the GMSL deserializer, extends the link length and enhances the link reliability. Spread spectrum is available on the MAX9249 to reduce EMI on the serial link and the parallel output of the GMSL deserializer. The serial output complies with ISO 10605 and IEC 61000-4-2 ESD protection standards.

#### **Register Mapping**

The  $\mu$ C configures various operating conditions of the MAX9249 and GMSL deserializer through internal registers. The default device addresses stored in the R0 and R1 registers of both the MAX9249 and GSML deserializer are 0x80 and 0x90, respectively. Write to the R0/R1 registers in both devices to change the device address of the MAX9249 or GMSL deserializer.

Table 1. Power-Up Default Register Map (see Table 12)

REGISTER ADDRESS (HEX)	POWER-UP DEFAULT (HEX)	POWER-UP DEFAULT SETTINGS (MSB FIRST)
0x00	0x80	SERID =1000000, serializer device address is 1000 000 RESERVED = 0
0x01	0x90	DESID =1001000, deserializer device address is 1001 000 RESERVED = 0
0x02	0x1F, 0x3F	SS = 000 (SSEN = low), SS = 001 (SSEN = high), spread-spectrum settings depend on SSEN pin state at power-up AUDIOEN = 1, I <sup>2</sup> S channel enabled PRNG = 11, automatically detect the pixel clock range SRNG = 11, automatically detect serial-data rate
0x03	0x00	AUTOFM = 00, calibrate spread-modulation rate only once after locking SDIV = 000000, autocalibrate sawtooth divider

Table 1. Power-Up Default Register Map (see Table 12) (continued)

REGISTER ADDRESS (HEX)	POWER-UP DEFAULT (HEX)	POWER-UP DEFAULT SETTINGS (MSB FIRST)
0x04	0x03, 0x13, 0x83 or 0x93	SEREN = 0 (AUTOS = high), SEREN = 1 (AUTOS = low), serial link enable default depends on AUTOS pin state at power-up CLINKEN = 0, configuration link disabled PRBSEN = 0, PRBS test disabled SLEEP = 0 or 1, sleep-mode state depends on CDS and AUTOS pin state at power-up (see the <i>Link Startup Procedure</i> section) INTTYPE = 00, base mode uses I <sup>2</sup> C REVCCEN = 1, reverse control channel active (receiving) FWDCCEN = 1, forward control channel active (sending)
0x05	0x70	I2CMETHOD = 0, I2C packets include register address DISFPLL = 1, filter PLL disabled CMLLVL = 11, 400mV CML signal level PREEMP = 0000, preemphasis off
0x06	0x40	RESERVED = 01000000
0x07	0x22	RESERVED = 00100010
0x08	0x0A (read only)	RESERVED = 0000 LFNEG = 10, no faults detected LFPOS = 10, no faults detected
0x0C	0x70	RESERVED = 01110000
0x0D	0x0F	SETINT = 0, interrupt output set to low RESERVED = 00 DISRES = 0, RES mapped to DIN27 SKEWADJ = 1111, no X7PLL clock skew adjustment
0x1E	0x03 (read only)	ID = 00000011, device ID is 0x03
0x1F	0x0X (read only)	RESERVED = 0000 REVISION = XXXX, revision number

## **VESA Standard Panel Bitmapping and Bus-Width Selection**

The LVDS input has two selectable widths, 3-channel and 4-channel. The MAX9249 accepts the VESA standard panel 3- or 4-channel LVDS (Table 2). Inputs on the MAX9249 are mapped internally, according to Figures 13 and 14. In 3-channel mode, RXIN3\_ and CNTL1/CNTL2 are not available. For both modes, the SD/CNTL0, SCK, and WS pins are for I2S audio. The MAX9249 accepts clock rates from 8.33MHz to 104MHz for 3-channel mode and 6.25MHz to 78MHz for 4-channel mode.

### **Serial Link Signaling and Data Format**

The MAX9249 high-speed data serial output uses CML signaling with programmable preemphasis and AC-coupling. The GMSL deserializer uses AC-coupling and programmable channel equalization. When using both the preemphasis and equalization, the MAX9249/GMSL deserializer can operate up to 3.125Gbps over STP cable lengths to 15m or more.

The MAX9249 serializer scrambles and encodes the LVDS input data and sends the 8B/10B coded signal through the serial link. The GMSL deserializer recovers

Table 2. Bus-Width Selection Using BWS

INPUT BITS		EL MODE = LOW)	4-CHANNEL MODE (BWS = HIGH)		
	VESA STANDARD PANEL MAPPING	AUXILIARY SIGNALS MAPPING	VESA STANDARD PANEL MAPPING	AUXILIARY SIGNALS MAPPING	
DIN[0:5]	R[0:5]	_	R[0:5]	_	
DIN[6:11]	G[0:5]	_	G[0:5]	_	
DIN[12:17]	B[0:5]	_	B[0:5]	_	
DIN[18:20]	HS, VS, DE	_	HS, VS, DE	_	
DIN[21:22]	Not used	Not used	R6, R7	_	
DIN[23:24]	Not used	Not used	G6, G7	_	
DIN[25:26]	Not used	Not used	B6, B7	_	
DIN27	Not used	Not used	RES*	CNTL1	
DIN28	Not used	Not used	_	CNTL2	
SD/CNTL0	_	SD/CNTL0	_	SD/CNTL0	

<sup>\*</sup>RES = Reserved (see the Reserved Bit (RES) section for details).

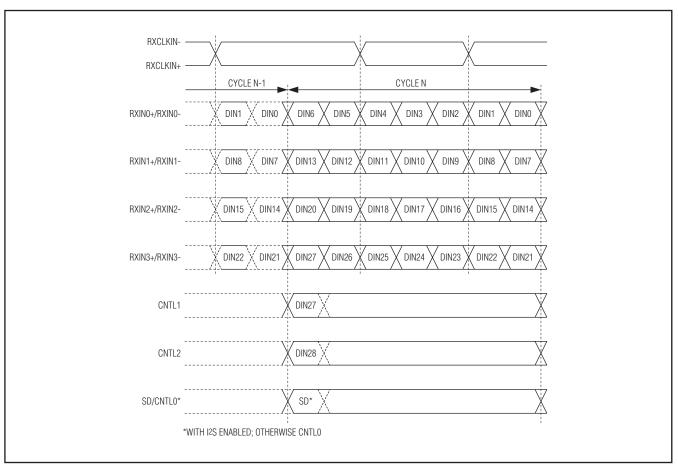


Figure 13. LVDS Input Timing

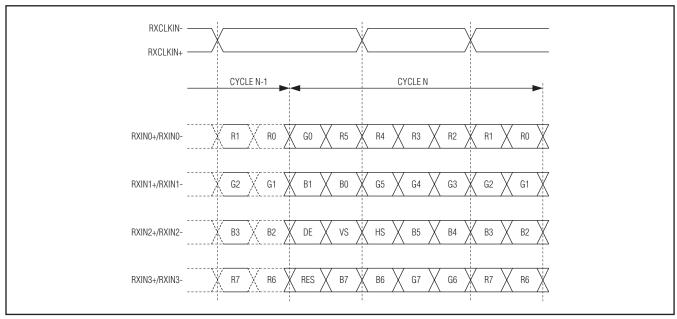


Figure 14. VESA Standard Panel Clock and Bit Assignment

the embedded serial clock and then samples, decodes, and descrambles before outputting the data. Figures 15 and 16 show the serial-data packet format before scrambling and 8B/10B coding. In 3-channel or 4-channel mode, 21 or 28 bits come from the RXIN\_ \_ LVDS inputs. Control bits can be mapped to DIN27 and DIN28 in 4-channel mode. The audio channel bit (ACB) contains an encoded audio signal derived from the three I2S inputs (SD/CNTL0, SCK, and WS). The forward control-channel (FCC) bit carries the forward control data. The last bit (PCB) is the parity bit of the previous 23 or 31 bits.

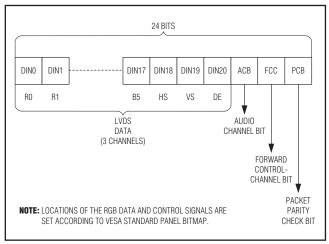


Figure 15. 3-Channel Mode Serial Link Data Format

#### Reserved Bit (RES)

In 4-channel mode, the MAX9249 serializes all bits of all four lanes including RES by default. Set DISRES (D4 of Register 0x0D) to 1 to map CNTL1 to DIN27 instead of RES.

#### **Reverse Control Channel**

The MAX9249 uses the reverse control channel to receive I<sup>2</sup>C/UART and interrupt signals from the GMSL deserializer in the opposite direction of the video stream. The reverse control channel and forward video data coexist on the same twisted pair forming a bidirectional link. The reverse control channel operates independently from the forward control channel. The reverse control channel is available 500µs after power-up. The MAX9249 temporarily disables the reverse control channel for 350µs after starting/stopping the forward serial link.

#### **Data-Rate Selection**

The MAX9249 uses the DRS input to set the RXCLKIN\_frequency. Set DRS high for an RXCLKIN\_frequency of 6.25MHz to 12.5MHz (4-channel mode) or 8.33MHz to 16.66MHz (3-channel mode). Set DRS low for normal operation with an RXCLKIN\_frequency of 12.5MHz to 78MHz (4-channel mode) or 16.66MHz to 104MHz (3-channel mode).

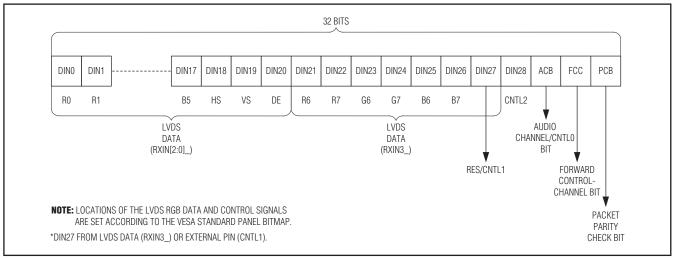


Figure 16. 4-Channel Mode Serial Link Data Format

Table 3. Maximum Audio WS Frequency (kHz) for Various RXCLKIN\_ Frequencies

WORD LENGTH (BITS)		RXCLKIN_ FREQUENCY (DRS = LOW) (MHz)			RXCLKIN_ FREQUENCY (DRS = HIGH) (MHz)			
	12.5 15 16.6 > 20			> 20	6.25	7.5	8.33	> 10
8	> 192	> 192	> 192	> 192	> 192	> 192	> 192	> 192
16	> 192	> 192	> 192	> 192	> 192	> 192	> 192	> 192
18	185.5	> 192	> 192	> 192	185.5	> 192	> 192	> 192
20	174.6	> 192	> 192	> 192	174.6	> 192	> 192	> 192
24	152.2	182.7	> 192	> 192	152.2	182.7	> 192	> 192
32	123.7	148.4	164.3	> 192	123.7	148.4	164.3	> 192

#### **Audio Channel**

The I<sup>2</sup>S audio channel supports audio sampling rates from 8kHz to 192kHz and audio word lengths from 4 bits to 32 bits. The audio bit clock (SCK) does not have to be synchronized with RXCLKIN\_. The MAX9249 automatically encodes audio data into a single bit stream synchronous with RXCLKIN\_. The GMSL deserializer decodes the audio stream and stores audio words in a FIFO. Audio rate detection uses an internal oscillator to continuously determine the audio data rate and output the audio in I<sup>2</sup>S format. The audio channel is enabled by default. When the audio channel is disabled, the audio data on the MAX9249 and GMSL deserializer is treated as a control pin (CNTL0).

Low RXCLKIN\_ frequencies limit the maximum audio sampling rate. Table 3 lists the maximum audio sampling rate for various RXCLKIN\_ frequencies. Spread-

spectrum settings do not affect the I<sup>2</sup>S data rate or WS clock frequency.

#### **Control Channel and Register Programming**

The control channel is available for the  $\mu C$  to send and receive control data over the serial link simultaneously with the high-speed data. Configuring the CDS pin allows the  $\mu C$  to control the link from either the MAX9249 or the GMSL deserializer side to support video-display or image-sensing applications.

The control channel between the  $\mu$ C and MAX9249 or GMSL deserializer runs in base mode or bypass mode according to the mode selection (MS) input of the device connected to the  $\mu$ C. Base mode is a half-duplex control channel and the bypass mode is a full-duplex control channel. In base mode, the  $\mu$ C is the host and can access the registers of both the MAX9249 and GMSL deserializer from either side of the link by using the GMSL

UART protocol. The  $\mu C$  can also program the peripherals on the remote side by sending the UART packets to the MAX9249 or GMSL deserializer, with the UART packets converted to I²C by the device on the remote side of the link (GMSL deserializer for LCD or MAX9249 for image-sensing applications). The  $\mu C$  communicates with a UART peripheral in base mode (through INTTYPE register settings), using the half-duplex default GMSL UART protocol of the MAX9249/GMSL deserializer. The device addresses of the MAX9249 and GMSL deserializer in base mode are programmable. The default values are 0x80 for the MAX9249 and 0x90 for the GMSL deserializer.

In base mode, when the peripheral interface uses I<sup>2</sup>C (default), the MAX9249/GMSL deserializer convert packets to I<sup>2</sup>C that have device addresses different from those of the MAX9249 or GMSL deserializer. The converted I<sup>2</sup>C bit rate is the same as the original UART bit rate.

In bypass mode, the MAX9249/GMSL deserializer ignore UART commands from the  $\mu C$  and the  $\mu C$  communicates with the peripherals directly using its own defined UART protocol. The  $\mu C$  cannot access the MAX9249/GMSL deserializer's registers in this mode. Peripherals accessed through the forward control channel using the UART interface need to handle at least one RXCLKIN\_period of jitter due to the asynchronous sampling of the UART signal by RXCLKIN\_.

The MAX9249 embeds control signals going to the GMSL deserializer in the high-speed forward link. Do not send a logic-low value longer than 100 $\mu$ s in either base or bypass mode. The GMSL deserializer uses a proprietary differential line coding to send signals back towards the MAX9249. The speed of the control channel ranges from 100kbps to 1Mbps in both directions. The MAX9249/GMSL deserializer automatically detect the control channel bit rate in base mode. Packet bit rates can vary up to 3.5x from the previous bit rate (see the *Changing the Clock Frequency* section). Figure 17 shows the UART protocol for writing and reading in base mode between the  $\mu$ C and the MAX9249/GMSL deserializer.

Figure 18 shows the UART data format. Even parity is used. Figures 19 and 20 detail the formats of the SYNC byte (0x79) and the ACK byte (0xC3). The  $\mu$ C and the connected slave chip generate the SYNC byte and ACK byte, respectively. Events such as device wake-up and interrupt generate transitions on the control channel that should be ignored by the  $\mu$ C. Data written to the MAX9249/GMSL deserializer registers does not take

effect until after the acknowledge byte is sent. This allows the  $\mu C$  to verify write commands received without error, even if the result of the write command directly affects the serial link. The slave uses the SYNC byte to synchronize with the host UART data rate automatically. If the INT or MS inputs of the GMSL deserializer toggles while there is control-channel communication, the control-channel communication may be corrupted. In the event of a missed acknowledge, the  $\mu C$  should assume there was an error in the packet when the slave device receives it, or that an error occurred during the response from the slave device. In base mode, the  $\mu C$  must keep the UART Tx/Rx lines high for 16 bit times before starting to send a new packet.

As shown in Figure 21, the remote-side device converts the packets going to or coming from the peripherals from the UART format to the I<sup>2</sup>C format and vice versa. The remote device removes the byte number count and adds or receives the ACK between the data bytes of I<sup>2</sup>C. The I<sup>2</sup>C's data rate is the same as the UART data rate.

# Interfacing Command-Byte-Only I<sup>2</sup>C Devices

The MAX9249 and GMSL deserializer UART-to-I<sup>2</sup>C conversion interfaces with devices that do not require register addresses, such as the MAX7324 GPIO expander. In this mode, the I<sup>2</sup>C master ignores the register address byte and directly reads/writes the subsequent data bytes (Figure 22). Change the communication method of the I<sup>2</sup>C master using the I<sup>2</sup>CMETHOD bit. I<sup>2</sup>CMETHOD = 1 sets command-byte-only mode, while I<sup>2</sup>CMETHOD = 0 sets normal mode where the first byte in the data stream is the register address.

#### **Interrupt Control**

The INT pin of the MAX9249 is the interrupt output and the INT pin of the GMSL deserializer is the interrupt input. The interrupt output on the MAX9249 follows the transitions at the interrupt input. This interrupt function supports remote-side functions such as touch-screen peripherals, remote power-up, or remote monitoring. Interrupts that occur during periods where the reverse control channel is disabled, such as link startup/shutdown, are automatically resent once the reverse control channel becomes available again. Bit D4 of register 0x06 in the GMSL deserializer also stores the interrupt input state. The INT output of the MAX9249 is low after power-up. In addition, the µC can set the INT output of MAX9249 by writing to the SETINT register bit. In normal operation, the state of the interrupt output changes when the interrupt input on the GMSL deserializer toggles.

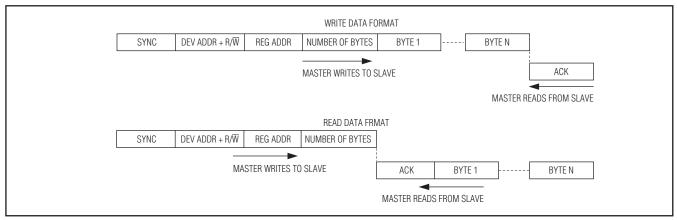


Figure 17. GMSL UART Protocol for Base Mode

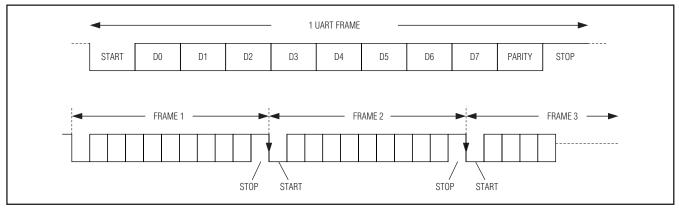


Figure 18. GMSL UART Data Format for Base Mode

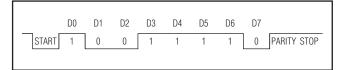


Figure 19. SYNC Byte (0x79)

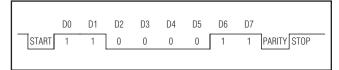


Figure 20. ACK Byte (0xC3)

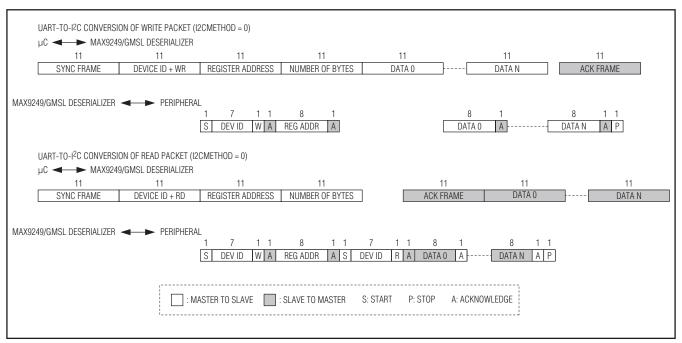


Figure 21. Format Conversion Between GMSL UART and I<sup>2</sup>C with Register Address (I2CMETHOD = 0)

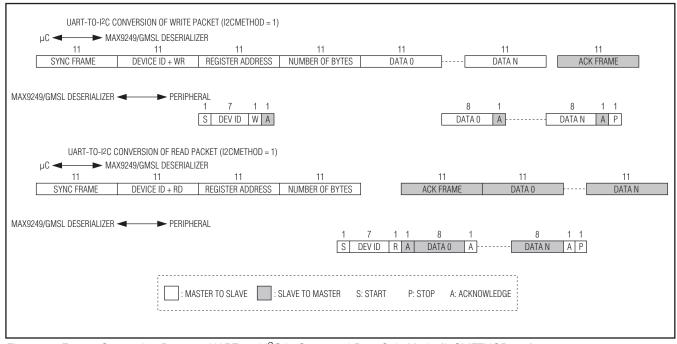


Figure 22. Format Conversion Between UART and I<sup>2</sup>C in Command-Byte-Only Mode (I2CMETHOD = 1)

Table 4. CML Driver Strength (Default Level, CMLLVL = 11)

DDEEMBLIA OLO LEVEL	PREEMPHASIS			SINGLE-ENDED	VOLTAGE SWING
PREEMPHASIS LEVEL (dB)*	SETTING (0x05, D[3:0])	(mA)	IPRE (mA)	MAX (mV)	MIN (mV)
-6.0	0100	12	4	400	200
-4.1	0011	13	3	400	250
-2.5	0010	14	2	400	300
-1.2	0001	15	1	400	350
0	0000	16	0	400	400
1.1	1000	16	1	425	375
2.2	1001	16	2	450	350
3.3	1010	16	3	475	325
4.4	1011	16	4	500	300
6.0	1100	15	5	500	250
8.0	1101	14	6	500	200
10.5	1110	13	7	500	150
14.0	1111	12	8	500	100

<sup>\*</sup>Negative preemphasis levels denote deemphasis.

**Table 5. Serial Output Spread** 

SS	SPREAD (%)
000	No spread spectrum.  Power-up default when SSEN = low.
001	±0.5% spread spectrum.  Power-up default when SSEN = high.
010	±1.5% spread spectrum
011	±2% spread spectrum
100	No spread spectrum
101	±1% spread spectrum
110	±3% spread spectrum
111	±4% spread spectrum

#### **Preemphasis Driver**

The serial line driver in the MAX9249 employs current-mode logic (CML) signaling. The driver can be programmed to generate a preemphasized waveform according to the cable length and characteristics. There are 13 preemphasis settings, as shown in Table 4. Negative preemphasis levels are deemphasis levels in which the swing is the same as normal, but the no-transition data is deemphasized. Program the preemphasis levels through register 0x05 D[3:0] of the MAX9249. This preemphasis function compensates the high-frequency loss of the cable and enables reliable transmission over longer link distances. Additionally, a lower power-drive mode can be entered by programming CMLLVL bits

(0x05 D[5:4]) to reduce the driver strength down to 75% (CMLLVL = 10) or 50% (CMLLVL = 01) from 100% (CMLLVL = 11, default).

#### **Spread Spectrum**

To reduce the EMI generated by the transitions on the serial link and outputs of the GMSL deserializer, both the MAX9249 and GMSL deserializer support spread spectrum. Turning on spread spectrum on the MAX9249 spreads the serial data and the GMSL deserializer outputs. Do not enable spread for both the MAX9249 and GMSL deserializer. The six selectable spread-spectrum rates at the MAX9249 serial output are  $\pm 0.5\%$ ,  $\pm 1\%$ ,  $\pm 1.5\%$ ,  $\pm 2\%$ ,  $\pm 3\%$ , and  $\pm 4\%$  (Table 5). Some spread-spectrum rates can only be used at lower RXCLKIN\_frequencies (Table 6). There is no RXCLKIN\_ frequency limit for the 0.5% spread rate.

Set the MAX9249 SSEN input high to select 0.5% spread at power-up and SSEN input low to select no spread at power-up. The state of SSEN is latched upon power-up or when resuming from power-down mode. Whenever the MAX9249 spread spectrum is turned on or off, the serial link automatically restarts and remains unavailable while the GMSL deserializer relocks to the serial data.

Turning on spread spectrum on the MAX9249 or GMSL deserializer does not affect the audio data stream. Changes in the MAX9249 spread settings only affect the GMSL deserializer MCLK output if it is derived from RXCLKIN\_ (MCLKSRC = 0).

**Table 6. Spread-Spectrum Rate Limitations** 

3-CHANNEL MODE RXCLKIN_ FREQUENCY (MHz)	4-CHANNEL MODE RXCLKIN_ FREQUENCY (MHz)	SERIAL LINK BIT RATE (Mbps)	AVAILABLE SPREAD RATES	
< 33.3	< 25	< 1000	All rates available	
33.3 to < 66.7	20 to < 50	1000 to < 2000	1.5%, 1.0%, 0.5%	
≥ 66.7	≥ 50	≥ 2000	0.5%	

**Table 7. Modulation Coefficients and Maximum SDIV Settings** 

BUS-WIDTH MODE	SPREAD-SPECTRUM SETTING (%)	MODULATION COEFFICIENT (DECIMAL)	SDIV UPPER LIMIT (DECIMAL)
	0.5	104	63
	1	104	40
4-Channel	1.5	152	54
4-Onamilei	2	204	30
	3	152	27
	4	204	15
	0.5	80	63
	1	80	52
2 Channal	1.5	112	63
3-Channel	2	152	42
	3	112	37
	4	152	21

Both devices include a sawtooth divider to control the spread-modulation rate. Autodetection or manual programming of the RXCLKIN\_ operation range guarantees a spread-spectrum modulation frequency within 20kHz to 40kHz. Additionally, manual configuration of the sawtooth divider (SDIV, 0x03 D[5:0]) allows the user to set a modulation frequency according to the RXCLKIN\_frequency. Always keep the modulation frequency between 20kHz to 40kHz to ensure proper operation.

#### Manual Programming of the Spread-Spectrum Divider

The modulation rate for the MAX9249 relates to the RXCLKIN\_ frequency as follows:

$$f_{M} = (1 + DRS) \frac{f_{RXCLKIN}}{MOD \times SDIV}$$

where:

f<sub>M</sub> = Modulation frequency

DRS = DRS pin input value (0 or 1)

fraction = LVDS clock frequency

MOD = Modulation coefficient given in Table 7

SDIV = 6-bit SDIV setting, manually programmed by the  $\mu$ C

To program the SDIV setting, first look up the modulation coefficient according to the part number and desired bus-width and spread-spectrum settings. Solve the above equation for SDIV using the desired pixel clock and modulation frequencies. If the calculated SDIV value is larger than the maximum allowed SDIV value in Table 7, set SDIV to the maximum value.

#### Sleep Mode

The MAX9249/GMSL deserializer include low-power sleep mode to reduce power consumption on the device not attached to the  $\mu$ C (the GMSL deserializer in LCD applications and the MAX9249 in camera applications). Set the corresponding remote IC's SLEEP bit to 1 to initiate sleep mode. The MAX9249 sleeps immediately after

Table 8. Startup Selection for Video-Display Applications (CDS = Low)

CASE	AUTOS (MAX9249)	MAX9249 POWER-UP STATE	MS (GMSL DESERIALIZER)	GMSL DESERIALIZER POWER-UP STATE	LINK STARTUP MODE
1	Low	Serialization enabled	Low	Normal (SLEEP = 0)	Both devices power up with serial link active (autostart)
2	High	Serialization disabled	High	Sleep mode (SLEEP = 1)	Serial link is disabled and the GMSL deserializer powers up in sleep mode. Set SEREN = 1 or CLINKEN = 1 in the MAX9249 to start the serial link and wake up the GMSL deserializer.
3	High	Serialization disabled	Low	Normal (SLEEP = 0)	Both devices power up in normal mode with the serial link disabled. Set SEREN = 1 or CLINKEN = 1 in the MAX9249 to start the serial link.
4	Low	Serialization enabled	High	Sleep mode (SLEEP = 1)	GMSL deserializer starts in sleep mode. Link autostarts upon MAX9249 power-up. Use this case when the GMSL deserializer powers up before the MAX9249.

setting its SLEEP = 1. The GMSL deserializer sleeps after serial link inactivity or 8ms (whichever arrives first) after setting its SLEEP = 1. See the *Link Startup Procedure* section for details on waking up the device for different  $\mu$ C and starting conditions.

The  $\mu C$  side device cannot enter into sleep mode. If an attempt is made to program the  $\mu C$  side device for sleep, the SLEEP bit remains 0. Use the  $\overline{PWDN}$  input pin to bring the  $\mu C$  side device into a low-power state.

#### **Configuration Link Mode**

The MAX9249 includes a low-speed configuration link to allow control-data connection between the two devices in the absence of a valid clock input. In either display or camera applications, the configuration link can be used to program equalizer/preemphasis or other registers before establishing the video link. An internal oscillator provides RXCLKIN\_ for establishing the serial configuration link between the MAX9249 and GMSL deserializer. Set CLINKEN = 1 on the MAX9249 to turn on the configuration link. The configuration link remains active as long as the video link has not been enabled. The video link overrides the configuration link and attempts to lock when SEREN = 1.

### **Link Startup Procedure**

Table 8 lists four startup cases for video-display applications. Table 9 lists two startup cases for image-sensing applications. In either video-display or image-sensing applications, the control link is always available after the high-speed data link or the configuration link is established and the MAX9249/GMSL deserializer registers or the peripherals are ready for programming.

#### **Video-Display Applications**

For the video-display application, with a remote display unit, connect the  $\mu C$  to the serializer (MAX9249) and set CDS = low for both the MAX9249 and GMSL deserializer. Table 8 summarizes the four startup cases based on the settings of  $\overline{AUTOS}$  and MS.

#### Case 1: Autostart Mode

After power-up or when PWDN transitions from low to high for both the serializer and deserializer, the serial link establishes if a stable RXCLKIN\_ is present. The MAX9249 locks to RXCLKIN\_ and sends the serial data to the GMSL deserializer. The GMSL deserializer then detects activity on the serial link and locks to the input serial data.

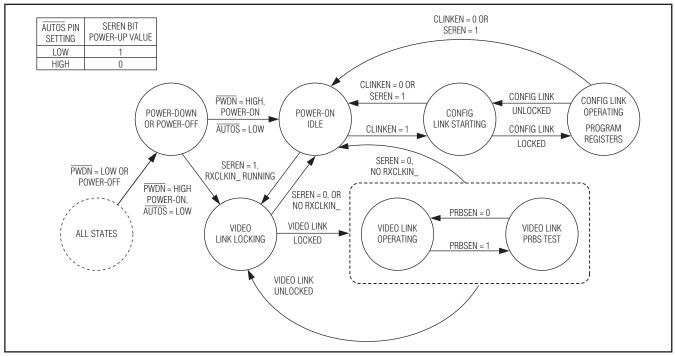


Figure 23. State Diagram, CDS = Low (LCD Application)

### Table 9. Startup Selection for Image-Sensing Applications (CDS = High)

CASE	AUTOS (MAX9249)	MAX9249 POWER-UP STATE	GMSL DESERIALIZER POWER-UP STATE	LINK STARTUP MODE
1	Low	Serialization enabled	Normal (SLEEP = 0)	Autostart
2	High	Sleep mode (SLEEP = 1)	Normal (SLEEP = 0)	MAX9249 is in sleep mode. Wake up the MAX9249 through the control channel (μC attached to the GMSL deserializer).

#### Case 2: Standby Start Mode

After power-up or when  $\overline{\text{PWDN}}$  transitions from low to high for both the serializer and deserializer, the GMSL deserializer starts up in sleep mode, and the MAX9249 stays in standby mode (does not send serial data). Use the  $\mu\text{C}$  and program the MAX9249 to set SEREN = 1 to establish a video link or CLINKEN = 1 to establish the configuration link. After locking to a stable RXCLKIN\_ (for SEREN = 1) or the internal oscillator (for CLINKEN = 1), the MAX9249 sends a wake-up signal to the deserializer. The GMSL deserializer exits sleep mode after locking to the serial data and sets SLEEP = 0. If after 8ms the deserializer does not lock to the input serial data, the GMSL deserializer goes back to sleep, and the internal sleep bit remains set (SLEEP = 1).

#### Case 3: Remote Side Autostart Mode

After power-up or when PWDN transitions from low to high, the remote device (GMSL deserializer) starts up and tries to lock to an incoming serial signal with sufficient power. The host side (MAX9249) is in standby mode and does not try to establish a link. Use the µC and program the MAX9249 to set SEREN = 1 (and apply a stable RXCLKIN\_) to establish a video link or CLINKEN = 1 to establish the configuration link. In this case, the GMSL deserializer ignores the short wake-up signal sent from MAX9249.

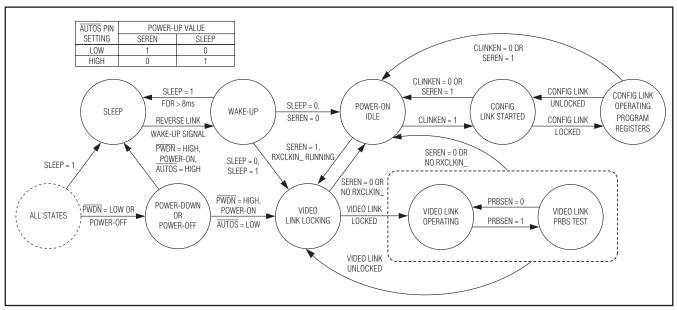


Figure 24. State Diagram, CDS = High (Camera Application)

#### Case 4: Remote Side in Sleep Mode

After power-up or when \$\overline{PWDN}\$ transitions from low to high, the remote device (GMSL deserializer) starts up in sleep mode. The high-speed link establishes automatically after MAX9249 powers up with a stable RXCLKIN\_ and sends a wake-up signal to the GMSL deserializer. Use this mode in applications where the GMSL deserializer powers up before the MAX9249.

#### **Image-Sensing Applications**

For image-sensing applications, connect the  $\mu C$  to the GMSL deserializer and set CDS = high for both the MAX9249 and GMSL deserializer. The GMSL deserializer powers up normally (SLEEP = 0) and continuously tries to lock to a valid serial input. Table 9 summarizes both startup cases, based on the state of the MAX9249  $\overline{\text{AUTOS}}$  pin.

#### Case 1: Autostart Mode

After power-up, or when PWDN transitions from low to high, the MAX9249 locks to a stable RXCLKIN\_ and sends the high-speed data to the GMSL deserializer. The GMSL deserializer locks to the serial data and outputs the video data and clock.

#### Case 2: Sleep Mode

After power-up or when  $\overline{PWDN}$  transitions from low to high, the MAX9249 starts up in sleep mode. To wake up the MAX9249, use the  $\mu C$  to send a GMSL protocol UART

frame containing at least three rising edges (e.g., 0x66), at a bit rate no greater than 1Mbps. The low-power wake-up receiver of the MAX9249 detects the wake-up frame over the reverse control channel and powers up. Reset the sleep bit (SLEEP = 0) of the MAX9249 using a regular control channel write packet to power up the device fully. Send the sleep bit write packet at least 500µs after the wake-up frame. The MAX9249 goes back to sleep mode if its sleep bit is not cleared within 5ms (min) after detecting a wake-up frame.

# Applications Information Self-PRBS Test

The MAX9249/GMSL deserializer link includes a PRBS pattern generator and bit-error verification function. Set PRBSEN =1 (0x04 D5) first in the MAX9249 and then the GMSL deserializer to start the PRBS test. Set PRBSEN =0 (0x04 D5) first in the GMSL deserializer and then the MAX9249 to exit the PRBS self-test. The GMSL deserializer uses an 8-bit register (0x0E) to count the number of detected errors. The control link also controls the start and stop of the error counting. During PRBS mode, the device does not count decoding errors and the GMSL deserializer  $\overline{\text{ERR}}$  output reflects PRBS errors only. Refer to the respective GMSL deserializer data sheet for more details.

# Microcontrollers on Both Sides of the GMSL Link (Dual μC Control)

Usually the microcontroller is either on the serializer (MAX9249) side for video-display applications or on the deserializer side for image-sensing applications. For the former case, both the CDS pins of the MAX9249/GMSL deserializer are set to low, and for the latter case, the CDS pins are set to high. However, if the CDS pin of the MAX9249 is low and the same pin of the GMSL deserializer is high, then the MAX9249/GMSL deserializer connect to both  $\mu\text{Cs}$  simultaneously. In such a case, the  $\mu\text{Cs}$  on either side can communicate with the MAX9249/GMSL deserializer.

Contentions of the control link can happen if the µCs on both sides are using the link at the same time. The MAX9249/GMSL deserializer do not provide the solution for contention avoidance. The serializer/deserializer do not send an acknowledge frame when communication fails due to contention. Users can always implement a higher layer protocol to avoid the contention. In addition, if UART communication across the serial link is not required, the µCs can disable the forward and reverse control channel through the REVCCEN and FWDCCEN bits (0x04 D[1:0]) in the MAX9249/GMSL deserializer. UART communication across the serial link is stopped and contention between µCs no longer occurs. During dual µCs operation, if one of the CDS pins on either side changes state, the link resumes the corresponding state described in the Link Startup Procedure section.

As an example of dual  $\mu C$  use in an image-sensing application, the MAX9249 can be in sleep mode and waiting for wake-up by the GMSL deserializer. After wake-up, the serializer-side  $\mu C$  sets the MAX9249 CDS pin low and assumes master control of the MAX9249 registers.

#### **Jitter-Filtering PLL**

In some applications, the input clock to the MAX9249 (RXCLKIN\_) includes jitter that reduces link reliability. The MAX9249 has a programmable narrow-band jitter-filtering PLL to attenuate frequency components outside the PLL's bandwidth (< 100kHz, typ). Enable the jitter-filtering PLL by setting DISFPLL = 0 (0x05 D6).

#### **Changing the Clock Frequency**

Both the video clock rate (frxclkin) and the control-channel clock rate (fuart) can be changed on-the-fly to support applications with multiple clock speeds. It is recommended to enable the serial link after RXCLKIN\_stabilizes. Stop RXCLKIN\_ for 5µs and restart the serial link or toggle SEREN after each change in the RXCLKIN\_frequency to recalibrate any automatic settings if a clean frequency change cannot be guaranteed. The reverse control channel remains unavailable for 350µs after serial link start or stop. Limit on-the-fly changes in fuart to factors of less than 3.5 at a time to ensure that the device recognizes the UART sync pattern. For example, when lowering the UART frequency from 1Mbps to 100kbps, first send data at 333kbps and then at 100kbps to have reduction ratios of 3 and 3.333, respectively.

#### **LOCK Output Loopback**

For quick loss-of-lock notification, the GMSL deserializer can loop back its LOCK output to the MAX9249 using the INT signal. Connect the LOCK output to the INT input of the GMSL deserializer. The interrupt output on the MAX9249 follows the transitions at the LOCK output of the GMSL deserializer. Reverse control-channel communication does not require an active forward link to operate and accurately tracks the LOCK status of the video link. LOCK asserts for video link only and not for the configuration link.

#### **Line-Fault Detection**

The line-fault detector in the MAX9249 monitors for line failures such as short to ground, short to power supply, and open link for system fault diagnosis. Figure 3 shows the required external resistor connections.  $\overline{LFLT} = low$ when a line fault is detected and  $\overline{LFLT}$  = high when the line returns to normal. The line-fault type is stored in 0x08 D[3:0] of the MAX9249. The fault-detector threshold voltages are referenced to the MAX9249 ground. Additional passive components set the DC level of the cable (Figure 3). If the MAX9249 and GMSL deserializer grounds are different, the link DC voltage during normal operation can vary and cross one of the fault-detection thresholds. For the fault-detection circuit, select the resistor's power rating to handle a short to the battery and use surface-mount resistors with small case size to minimize parasitic effects to the high-speed signal. Table 10 lists the mapping for line-fault types.

**Table 10. Line-Fault Mapping** 

REGISTER ADDRESS	BITS	NAME	VALUE	LINE-FAULT TYPE	
			00	Negative cable wire shorted to battery	
	ומיטו	LFNEG	01	Negative cable wire shorted to ground	
	D[3:2]		10	Normal operation	
0x08					11
UXUO		1:0] LFPOS	00	Positive cable wire shorted to battery	
	D[1.0]		01	Positive cable wire shorted to ground	
וון טוו:נ	D[1:0]		10	Normal operation	
			11	Positive cable wire open	

#### **Choosing I<sup>2</sup>C/UART Pullup Resistors**

Both I2C/UART open-drain lines require pullup resistors to provide a logic-high level. There are trade-offs between power dissipation and speed, and a compromise made in choosing pullup resistor values. Every device connected to the bus introduces some capacitance even when the device is not in operation. I2C specifies 300ns rise times to go from low to high (30% to 70%) for fast mode, which is defined for data rates up to 400kbps (see the I2C specifications in the *AC Electrical Characteristics* section for details). To meet the fast-mode rise-time requirement, choose the pullup resistors so that rise time  $t_{\rm R}=0.85 \times R_{\rm PULLUP} \times C_{\rm BUS} < 300ns$ . The waveforms are not recognized if the transition time becomes too slow. The MAX9249 supports I2C/UART rates up to 1Mbps.

#### **AC-Coupling**

AC-coupling isolates the receiver from DC voltages up to the voltage rating of the capacitor. Four capacitors—two at the serializer output and two at the deserializer input—are needed for proper link operation and to provide protection if either end of the cable is shorted to a high voltage. AC-coupling blocks low-frequency ground shifts and low-frequency common-mode noise.

#### **Selection of AC-Coupling Capacitors**

Voltage droop and the digital sum variation (DSV) of transmitted symbols cause signal transitions to start from different voltage levels. Because the transition time is finite, starting the signal transition from different voltage levels causes timing jitter. The time constant for an AC-coupled link needs to be chosen to reduce droop

and jitter to an acceptable level. The RC network for an AC-coupled link consists of the CML receiver termination resistor (RTR), the CML driver termination resistor (RTD), and the series AC-coupling capacitors (C). The RC time constant for four equal-value series capacitors is (C x (RTD + RTR))/4. RTD and RTR are required to match the transmission line impedance (usually  $100\Omega$ ). This leaves the capacitor selection to change the system time constant. Use at least  $0.2\mu\text{F}$  high-frequency surface-mount ceramic capacitors, with sufficient voltage rating to withstand a short to battery, to pass the lower speed reverse control-channel signal. Use capacitors with a case size less than  $3.2\text{mm} \times 1.6\text{mm}$  to have lower parasitic effects to the high-speed signal.

#### **Power-Supply Circuits and Bypassing**

The MAX9249 uses a VAVDD and VDVDD of 1.7V to 1.9V, and a VLVDSVDD of 3.0V to 3.6V. All single-ended inputs and outputs on the MAX9249 derive power from a VIOVDD of 1.7V to 3.6V, which scale with IOVDD. Proper voltage-supply bypassing is essential for high-frequency circuit stability.

#### **Cables and Connectors**

Interconnect for CML typically has a differential impedance of  $100\Omega$ . Use cables and connectors that have matched differential impedance to minimize impedance discontinuities. Twisted-pair and shielded twisted-pair cables tend to generate less EMI due to magnetic-field canceling effects. Balanced cables pick up noise as common mode rejected by the CML receiver. Table 11 lists the suggested cables and connectors used in the GMSL link.

### Table 11. Suggested Connectors and Cables for GMSL

VENDOR	CONNECTOR	CABLE	
JAE Electronics, Inc.	MX38-FF	A-BW-Lxxxx	
Nissei Electric Co., Ltd.	GT11L-2S	F-2WME AWG28	
Rosenberger Hochfrequenztechnik GmbH	D4S10A-40ML5-Z	Dacar 538	

#### **Board Layout**

Separate the digital signals and CML/LVDS high-speed signals to prevent crosstalk. Use a four-layer PCB with separate layers for power, ground, CML/LVDS, and digital signals. Layout PCB traces close to each other for a  $100\Omega$  differential characteristic impedance. The trace dimensions depend on the type of trace used (microstrip or stripline). Note that two  $50\Omega$  PCB traces do not have  $100\Omega$  differential impedance when brought close together—the impedance goes down when the traces are brought closer.

Route the PCB traces for a CML/LVDS channel (there are two conductors per CML/LVDS channel) in parallel to maintain the differential characteristic impedance. Avoid vias. Keep PCB traces that make up a differential pair equal length to avoid skew within the differential pair.

#### **ESD Protection**

The MAX9249 ESD tolerance is rated for Human Body Model, IEC 61000-4-2, and ISO 10605. The ISO 10605 and IEC 61000-4-2 standards specify ESD tolerance for electronic systems. CML/LVDS I/O are tested for ISO 10605 ESD protection and IEC 61000-4-2 ESD protection. All pins are tested for the Human Body Model. The Human Body Model discharge components are Cs = 100pF and RD = 1.5k $\Omega$  (Figure 25). The IEC 61000-4-2 discharge components are Cs = 150pF and RD = 330 $\Omega$  (Figure 26). The ISO 10605 discharge components are Cs = 330pF and RD = 2k $\Omega$  (Figure 27).

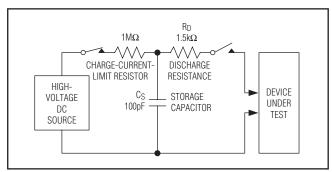


Figure 25. Human Body Model ESD Test Circuit

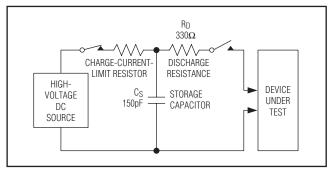


Figure 26. IEC 61000-4-2 Contact Discharge ESD Test Circuit

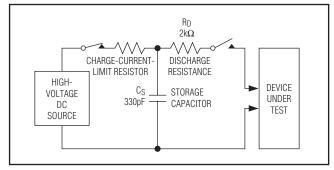


Figure 27. ISO 10605 Contact Discharge ESD Test Circuit

Table 12. Register Table (See Table 1 for Default Value Details)

REGISTER ADDRESS	BITS	NAME	VALUE	FUNCTION	DEFAULT VALUE							
0,400	D[7:1]	SERID	XXXXXXX	Serializer device address	1000000							
0x00	D0	_	0	Reserved	0							
0x01	D[7:1]	DESID	XXXXXXX	Deserializer device address	1001000							
UXUT	D0	_	0	Reserved	0							
			000	No spread spectrum. Power-up default when SSEN = low.								
			001	$\pm 0.5\%$ spread spectrum. <b>Power-up default when SSEN = high.</b>								
	D[7:5]	SS	010	±1.5% spread spectrum	000, 001							
	[6.7]	33	011	±2% spread spectrum	000, 001							
			100	No spread spectrum								
			101	±1% spread spectrum								
			110	±3% spread spectrum								
0,400			111	±4% spread spectrum								
0x02	D4	ALIDIOEN	0	Disable I <sup>2</sup> S channel	4							
	D4	AUDIOEN	1	Enable I <sup>2</sup> S channel	1							
		3:2] PRNG	00	12.5MHz to 25MHz pixel clock								
	D[0 0]		01	25MHz to 50MHz pixel clock	1-1							
	D[3:2]		PRNG	10	50MHz to 104MHz pixel clock	11						
			11	Automatically detect the pixel clock range								
			00	0.5 to 1Gbps serial-bit rate								
	D[1:0]	00010	01	1 to 2Gps serial-bit rate								
		D[1:0]	D[1:0] 	D[1:0]	D[1:0]	D[1:0]	D[1:0]	D[1:0]	SRNG	10	2 to 3.125Gbps serial-bit rate	11
			11	Automatically detect serial-bit rate								
			00	Calibrate spread-modulation rate only once after locking								
	D[7.0]		01	Calibrate spread-modulation rate every 2ms after locking	00							
0x03	D[7:6] AUTOFM -	10	Calibrate spread-modulation rate every 16ms after locking	00								
			11	Calibrate spread-modulation rate every 256ms after locking								
			000000	Autocalibrate sawtooth divider								
	D[5:0]	SDIV	XXXXXX	Manual SDIV setting. See the Manual Programming of the Spread-Spectrum Divider section.	000000							

Table 12. Register Table (See Table 1 for Default Value Details) (continued)

REGISTER ADDRESS	BITS	NAME	VALUE	FUNCTION	DEFAULT VALUE
	D7	SEREN	0	Disable serial link. Power-up default when AUTOS = high. Reverse control-channel communication remains unavailable for 350µs after the MAX9249 starts/stops the serial link.	0, 1
			1	Enable serial link. Power-up default when  AUTOS = low. Reverse control-channel communication remains unavailable for 350µs after the MAX9249 starts/stops the serial link.	
	D6	CLINKEN	0	Disable configuration link	0
			1	Enable configuration link	0
	D5	PRBSEN	0	Disable PRBS test	0
	D5		1	Enable PRBS test	U
0x04		OL EED	0	Normal mode. Default value depends on CDS and AUTOS pin values at power-up.	0, 1
	D4	SLEEP	1	Activate sleep mode. Default value depends on CDS and AUTOS pin values at power-up.	
	D[3:2]	INTTYPE	00	Base mode uses I <sup>2</sup> C peripheral interface	00
			01	Base mode uses UART peripheral interface	
			10, 11	Base mode peripheral interface disabled	
	Di	DEVICOEN	0	Disable reverse control channel from deserializer (receiving)	1
	D1	REVCCEN	1	Enable reverse control channel from deserializer (receiving)	
	D0	D0 FWDCCEN	0	Disable forward control channel to deserializer (sending)	-1
			1	Enable forward control channel to deserializer (sending)	1

Table 12. Register Table (See Table 1 for Default Value Details) (continued)

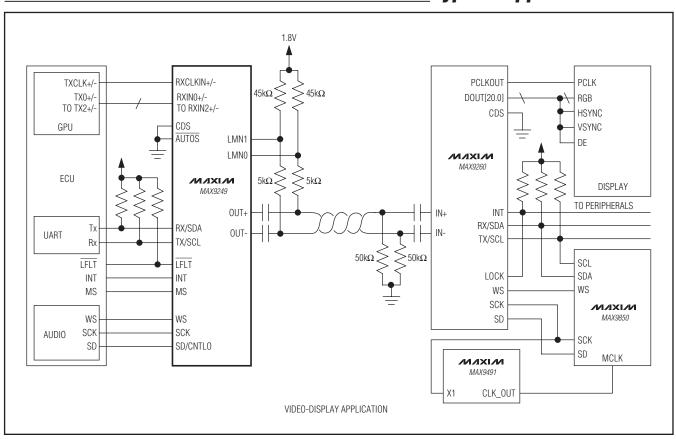
REGISTER ADDRESS	BITS	NAME	VALUE	FUNCTION	DEFAULT VALUE
	D7		0	I <sup>2</sup> C conversion sends the register address	
		I2CMETHOD	1	Disable sending of I <sup>2</sup> C register address (command-byte-only mode)	0
	DC	DICEDIA	0	Filter PLL active	4
	D6	DISFPLL	1	Filter PLL disabled	1
	D[5 4]	OMILIVII	00	Do not use	
			01	200mV CML signal level	
	D[5:4]	CMLLVL	10	300mV CML signal level	11
			11	400mV CML signal level	
			0000	Preemphasis off	
			0001	-1.2dB preemphasis	
			0010	-2.5dB preemphasis	
0x05			0011	-4.1dB preemphasis	
			0100	-6.0dB preemphasis	
			0101	Do not use	
			0110	Do not use	
	D[3:0]	PREEMP	0111	Do not use	0000
			1000	1.1dB preemphasis	0000
			1001	2.2dB preemphasis	
			1010	3.3dB preemphasis	
			1011	4.4dB preemphasis	
			1100	6.0dB preemphasis	
			1101	8.0dB preemphasis	
			1110	10.5dB preemphasis	
			1111	14.0dB preemphasis	
0x06	D[7:0]	_	01000000	Reserved	01000000
0x07	D[7:0]	_	00100010	Reserved	00100010
	D[7:4]	_	0000	Reserved	0000 (read only)
0x08	D[3:2]	3:2] LFNEG	00	Negative cable wire shorted to battery	
			01	Negative cable wire shorted to ground	10
			10	Normal operation	(read only)
			11	Negative cable wire open	
			00	Positive cable wire shorted to battery	
	D[/ 63	LEBGG	01	Positive cable wire shorted to ground	10
	D[1:0]	[1:0] LFPOS	10	Normal operation	(read only)
			11	Positive cable wire open	
0x0C	D[7:0]	_	01110000	Reserved	01110000

Table 12. Register Table (See Table 1 for Default Value Details) (continued)

REGISTER ADDRESS	BITS	NAME	VALUE	FUNCTION	DEFAULT VALUE				
	D7	OFTINIT	0	Set INT low when SETINT transitions from 1 to 0	0				
	D7	SETINT	1	Set INT high when SETINT transitions from 0 to 1					
	D[6:5]	_	00	Reserved	00				
	D4	DISRES 0	0	RES (LVDS interface) mapped to DIN27	- 0				
	D4	DISHES	1	CNTL1 mapped to DIN27	U				
			0000	Adjust x7 PLL clock skew + 50ps					
			0001	Adjust x7 PLL clock skew + 100ps					
			0010	Adjust x7 PLL clock skew + 200ps					
	D[3:0]		0011	Adjust x7 PLL clock skew + 250ps	1111				
			0100	Adjust x7 PLL clock skew + 300ps					
0x0D			0101	Adjust x7 PLL clock skew + 350ps					
		)] SKEWADJ	0110	Adjust x7 PLL clock skew + 400ps					
			0111	Do not use					
			1000	Adjust x7 PLL clock skew - 50ps					
							1001	Adjust x7 PLL clock skew - 100ps	
			1010	Adjust x7 PLL clock skew - 200ps	]				
				1011	Adjust x7 PLL clock skew - 250ps	]			
			1100	Adjust x7 PLL clock skew - 300ps					
			1101	Adjust x7 PLL clock skew - 350ps					
			1110	Adjust x7 PLL clock skew - 400ps					
			1111	No x7PLL clock skew adjustment					
0x1E	D[7:0]	ID	00000011	Device identifier (MAX9249 = 0x03)	00000011 (read only)				
0x1F	D[7:4]	_	0000	Reserved	0000 (read only)				
	D[3:0]	REVISION	XXXX	Device revision	(read only)				

X = Don't care.

### **Typical Application Circuit**



#### **Chip Information**

### PROCESS: CMOS

### Package Information

For the latest package outline information and land patterns (footprints), go to <a href="www.maxim-ic.com/packages">www.maxim-ic.com/packages</a>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE	PACKAGE	OUTLINE	LAND	
TYPE	CODE	NO.	PATTERN NO.	
48 TQFP-EP	C48E+8	21-0065	<u>90-0138</u>	

### **Revision History**

REVISION NUMBER	REVISION DATE	DESCRIPTION	
0	1/10	Initial release	_
1	3/10	Improved yield	2, 3
2	5/10	Added soldering temperature (reflow) to the <i>Absolute Maximum Ratings</i> section and corrected spread-spectrum modulation settings in Table 7	
3	1/11	Added Patent Pending to Features	1
4	1/12	Corrected GND to AGND in Absolute Maximum Ratings	2



Компания «Океан Электроники» предлагает заключение долгосрочных отношений при поставках импортных электронных компонентов на взаимовыгодных условиях!

#### Наши преимущества:

- Поставка оригинальных импортных электронных компонентов напрямую с производств Америки, Европы и Азии, а так же с крупнейших складов мира;
- Широкая линейка поставок активных и пассивных импортных электронных компонентов (более 30 млн. наименований);
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
- Экспресс доставка в любую точку России;
- Помощь Конструкторского Отдела и консультации квалифицированных инженеров;
- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
- Поставка электронных компонентов под контролем ВП;
- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001;
- При необходимости вся продукция военного и аэрокосмического назначения проходит испытания и сертификацию в лаборатории (по согласованию с заказчиком);
- Поставка специализированных компонентов военного и аэрокосмического уровня качества (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Actel, Aeroflex, Peregrine, VPT, Syfer, Eurofarad, Texas Instruments, MS Kennedy, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

Компания «Океан Электроники» является официальным дистрибьютором и эксклюзивным представителем в России одного из крупнейших производителей разъемов военного и аэрокосмического назначения «JONHON», а так же официальным дистрибьютором и эксклюзивным представителем в России производителя высокотехнологичных и надежных решений для передачи СВЧ сигналов «FORSTAR».



**«JONHON»** (основан в 1970 г.)

Разъемы специального, военного и аэрокосмического назначения:

(Применяются в военной, авиационной, аэрокосмической, морской, железнодорожной, горно- и нефтедобывающей отраслях промышленности)

«**FORSTAR**» (основан в 1998 г.)

ВЧ соединители, коаксиальные кабели, кабельные сборки и микроволновые компоненты:

(Применяются в телекоммуникациях гражданского и специального назначения, в средствах связи, РЛС, а так же военной, авиационной и аэрокосмической отраслях промышленности).



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