





8V to 36Vin Cool-Power ZVS Buck Regulator Family

Description

The PI33XX is a family of high efficiency, wide input range DC-DC ZVS-Buck regulators integrating controller, power switches, and support components all within a high density System-in-Package (SiP). The integration of a high performance Zero-Voltage Switching (ZVS) topology, within the PI33XX series, increases point of load performance providing best in class power efficiency. The PI33XX requires only an external inductor and minimal capacitors to form a complete DC-DC switching mode buck regulator.

Device	Out	lout Max	
Device	Set	IOUL IVIAX	
PI3311-X0-LGIZ	1.0V	1.0 to 1.4V	10A
PI3318-X0-LGIZ	1.8V	1.4 to 2.0V	10A
PI3312-X0-LGIZ	2.5V	2.0 to 3.1V	10A
PI3301-X0-LGIZ	3.3V	2.3 to 4.1V	10A
PI3302-X0-LGIZ	5.0V	3.3 to 6.5V	10A
PI3303-X0-LGIZ	12V	6.5 to 13.0V	8A
PI3305-X0-LGIZ	15V	10.0 to 16.0V	8A

Table 1 - PI33XX-X0 Portfolio.

The ZVS architecture also enables high frequency operation while minimizing switching losses and maximizing efficiency. The high switching frequency operation reduces the size of the external filtering components, improves power density, and enables very fast dynamic response to line and load transients. The PI33XX series sustains high switching frequency all the way up to the rated input voltage without sacrificing efficiency and, with its 20ns minimum on-time, supports large step down conversions up to 36Vin.

Features

- High Efficiency ZVS-Buck Topology
- Wide input voltage range of 8V to 36V
- Very-Fast transient response
- High accuracy pre-trimmed output voltage
- User adjustable soft-start & tracking
- Power-up into pre-biased load (select versions)
- Parallel capable with single wire current sharing
- Input Over/Under Voltage Lockout (OVLO/UVLO)
- Output Overvoltage Protection (OVP)
- Over Temperature Protection (OTP)
- Fast and slow current limits
- -40°C to 125°C operating range (T_J)
- Optional I²C functionality & programmability:
 - · Vout margining
 - Fault reporting
 - Enable and SYNCI pin polarity
 - Phase delay (interleaving multiple regulators)

Applications

- High efficiency systems
- Computing, Communications, Industrial, Automotive Equipment
- High voltage battery operation

Package Information

• 10mm x 14mm x 2.6mm LGA SiP



I²C is a trademark of NXP Semiconductors

Contents

Order Information3
Pin Description4
Package Pin-Out4
Absolute Maximum Ratings5
PI3311-X0 (1.0 Vout) Electrical Characteristics 6
PI3318-X0 (1.8 Vout) Electrical Characteristics9
PI3312-X0 (2.5 Vout) Electrical Characteristics 12
PI3301-X0 (3.3 Vout) Electrical Characteristics 16
PI3302-X0 (5.0 Vout) Electrical Characteristics 20
PI3303-X0 (12.0 Vout) Electrical Characteristics 24
PI3305-X0 (15.0 Vout) Electrical Characteristics 28
Functional Description32
ENABLE (EN)
Switching Frequency Synchronization32
Output Voltage Trim32
Soft-Start32
Remote Sensing33
Output Current Limit Protection33
Input Under-Voltage Lockout33
Input Over Voltage Lockout33
Output Over Voltage Protection33
Over Temperature Protection33
Pulse Skip Mode (PSM)33
Variable Frequency Operation34
Parallel Operation34
1 ² C Interface Operation 24

Application Description	35
Output Voltage Trim	35
Soft-Start Adjust and Tracking	36
Inductor Pairing	36
Thermal Derating	36
Filter Considerations	38
Layout Guidelines	39
Recommended PCB Footprint and Stencil	10
Package Drawings	11
Warranty	12



Order Information

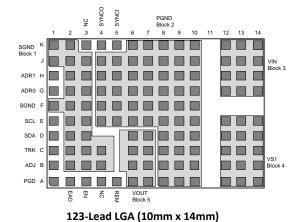
Cool Dower	Output Range		lout	Dankara	T
Cool-Power	Set	Range	Max	Package	Transport Media
PI3311-00-LGIZ	1.0V	1.0 to 1.4V	10A	10mm x 14mm 123-pin LGA	TRAY
PI3318-00-LGIZ	1.8V	1.4 to 2.0V	10A	10mm x 14mm 123-pin LGA	TRAY
PI3312-00-LGIZ	2.5V	2.0V to 3.1V	10A	10mm x 14mm 123-pin LGA	TRAY
PI3301-00-LGIZ	3.3V	2.3 to 4.1V	10A	10mm x 14mm 123-pin LGA	TRAY
PI3302-00-LGIZ	5.0V	3.3 to 6.5V	10A	10mm x 14mm 123-pin LGA	TRAY
PI3303-00-LGIZ	12V	6.5 to 13.0V	8A	10mm x 14mm 123-pin LGA	TRAY
PI3305-00-LGIZ	15V	10.0 to 16.0V	8A	10mm x 14mm 123-pin LGA	TRAY
Higher Current \	/ersions (R	efer to PI33XX-X1-LGIZ	Z Datashee	t)	
PI3311-01-LGIZ	1.0V	1.0 to 1.4V	15A	10mm x 14mm 123-pin LGA	TRAY
PI3318-01-LGIZ	1.8V	1.4 to 2.0V	15A	10mm x 14mm 123-pin LGA	TRAY
PI3312-01-LGIZ	2.5V	2.0 to 3.1V	15A	10mm x 14mm 123-pin LGA	TRAY
PI3301-01-LGIZ	3.3V	2.3 to 4.1V	15A	10mm x 14mm 123-pin LGA	TRAY
I ² C Functionality	& Progran	nmability			
PI3311-20-LGIZ	1.0V	1.0 to 1.4V	10A	10mm x 14mm 123-pin LGA	TRAY
PI3318-20-LGIZ	1.8V	1.0 to 1.4V	10A	10mm x 14mm 123-pin LGA	TRAY
PI3312-20-LGIZ	2.5V	2.0 to 3.1V	10A	10mm x 14mm 123-pin LGA	TRAY
PI3301-20-LGIZ	3.3V	2.3 to 4.1V	10A	10mm x 14mm 123-pin LGA	TRAY
PI3302-20-LGIZ	5.0V	3.30 to 6.5V	10A	10mm x 14mm 123-pin LGA	TRAY
PI3303-20-LGIZ	12V	6.5 to 13.0V	8A	10mm x 14mm 123-pin LGA	TRAY
PI3305-20-LGIZ	15V	10.0 to 16.0V	8A	10mm x 14mm 123-pin LGA	TRAY
PI3311-21-LGIZ	1.0V	1.0 to 1.4V	15A	10mm x 14mm 123-pin LGA	TRAY
PI3318-21-LGIZ	1.8V	1.4 to 2.0V	15A	10mm x 14mm 123-pin LGA	TRAY
PI3312-21-LGIZ	2.5V	2.0 to 3.1V	15A	10mm x 14mm 123-pin LGA	TRAY
PI3301-21-LGIZ	3.3V	2.3 to 4.1V	15A	10mm x 14mm 123-pin LGA	TRAY



Pin Description

Name	Number	Description
SGND	Block 1	Signal ground: Internal logic ground for EA, TRK, SYNCI, SYNCO, ADJ and I ² C (options) communication returns. SGND and PGND are star connected within the regulator package.
PGND	Block 2	Power ground: VIN and VOUT power returns
VIN	Block 3	Input voltage: and sense for UVLO, OVLO and feed forward ramp
VOUT	Block 5	Output voltage: and sense for power switches and feed-forward ramp
VS1	Block 4	Switching node: and ZVS sense for power switches
PGD	A1	Parallel Good: Used for parallel timing management intended for lead regulator.
EAO	A2	Error amp output: External connection for additional compensation and current sharing.
EN	А3	Enable Input: Regulator enable control. Asserted high or left floating – regulator enabled; Asserted low, regulator output disabled. Polarity is programmable via 1 ² C interface.
REM	A5	Remote Sense: High side connection. Connect to output regulation point.
ADJ	B1	Adjust input: An external resistor may be connected between ADJ pin and SGND or VOUT to trim the output voltage up or down.
TRK	C1	Soft-start and track input: An external capacitor may be connected between TRK pin and SGND to decrease the rate of rise during soft-start.
NC	K3, A4	No Connect: Leave pins floating.
SYNCO	K4	Synchronization output: Outputs a low signal for ½ of the minimum period for synchronization of other converters.
SYNCI	K5	Synchronization input: Synchronize to the falling edge of external clock frequency. SYNCl is a high impedance digital input node and should always be connected to SGND when not in use.
SDA	D1	Data Line: Connect to SGND for PI33XX-10 and -11. For use with PI33XX-20 and -21 only.
SCL	E1	Clock Line: Connect to SGND for PI33XX-10 and -11. For use with PI33XX-20 and -21 only.
ADR1	H1	Tri-state Address: No connect for PI33XX-10 and -11. For use with PI33XX-20 and -21 only.
ADR0	G1	Tri-state Address: No connect for PI33XX-10 and -11. For use with PI33XX-20 and -21 only.

Package Pin-Out



Top view

Block 1: B2-4, C2-4, D2-3, E2-3, F1-3, G2-3, H2-3, J1-3, K1-2

Block 2: A8-10, B8-10, C8-10, D8-10, E4-10, F4-10, G4-10, H4-10, J4-10, K6-10

Block 3: G12-14, H12-14, J12-14, K12-14

Block 4: A12-14, B12-14, C12-14, D12-14, E12-14,

Block 5: A6-7, B6-7, C6-7, D6-7

Absolute Maximum Ratings

Notes: At 25°C ambient temperature. All voltage nodes are referenced to PGND.

VIN	-0.7V to 36V
VS1	-0.7 to 36V, -4V for 5ns
VOUT	See relevant product section
SGND	100mA
PGD, SYNCO, SYNCI, EN, EAO, ADJ, TRK, ADR1, ADR2, SCL, SDA, REM	-0.3V to 5.5V / 5mA
Storage Temperature	-65°C to 150°C
Operating Junction Temperature	-40°C to 125°C
Soldering Temperature for 20 seconds	260°C
ESD Rating	2kV HBM

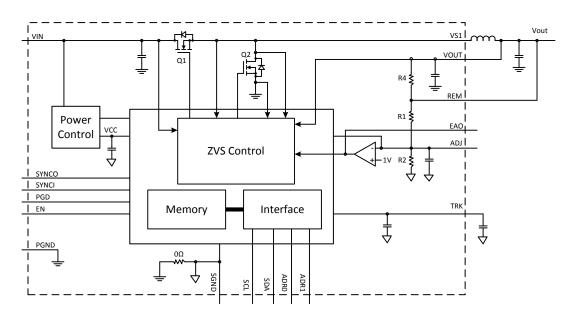


Figure 1: Simplified Block Diagram

(I²C pins SCL, SDA, ADRO, and ADR1 only active for Pl33XX-20 and Pl33XX-21 versions)



Rev 1.4

11/2012

Specifications apply for $-40^{\circ}\text{C} < \text{T}_{\text{J}} < 125^{\circ}\text{C}$, Vin =24V, L1=125nH (Note 1) unless other conditions are noted.

Parameter	Symbol	Min	Тур	Max	Units	Conditions
Input Specifications		1			•	
Input Voltage	V _{IN_DC}	8	24	36	V	Minimum 1mA load required
Input Current	I _{IN_DC}		476		mA	Vin = 24V, T _C = 25°C, lout=10A
Input Current At Output Short (fault condition duty cycle)	I _{IN_Short}			20	mA	
Input Quiescent Current	I _{Q_VIN}		2 2.5		mA	Disabled Enabled (no load)
Input Voltage Slew Rate	V_{IN_SR}			1	V/µs	
Output Specifications						
Output Voltage Total Regulation	V _{OUT_DC}	0.987	1.0	1.013	V	Note 2.
Output Voltage Trim Range	V_{OUT_DC}	1.0		1.4	V	Note 3.
Line Regulation	$\Delta V_{OUT}(\Delta V_{IN})$		0.1		%	@25°C, 8V <vin<36v< td=""></vin<36v<>
Load Regulation	$\Delta V_{OUT}(\Delta I_{OUT})$		0.1		%	@25°C, 0.5A <lout<10a< td=""></lout<10a<>
Output Voltage Ripple	V _{OUT_AC}		20		mVp-p	lout=5A, Cout=8x100μF, 20MHz BW Note 4.
Continuous Output Current Range	I _{OUT_DC}			10	А	Note 5. Min 1mA load required.
Current Limit	I _{OUT_CL}		12		Α	
Protection		•	•	•	•	
UVLO Start Threshold	V _{UVLO_START}	7.20	7.60	8.00	V	
UVLO Stop Threshold	V_{UVLO_STOP}	6.90	7.25	7.60	V	
UVLO Hysteresis	V _{UVLO_HYS}	4	5	6	%	
OVLO Threshold	V _{OVLO}	37	38.4		V	
OVLO Hysteresis	V _{OVLO_HYS}		0.77		V	
UVLO/OVLO Fault Delay Time	t _{f_DLY}		128		Cycles	Number of the switching frequency cycles
UVLO/OVLO Response Time	t _f		500		ns	+1% overdrive
Output Over Voltage Protection	V _{OVP}		20		%	Above V _{OUT}
Over-Temperature Fault Threshold	T _{OTP}	130	135	140	°C	
Over-Temperature Restart Hysteresis	T _{OTP_HYS}		30		°C	
Note 1: All parameters reflect regulator and ind		Inte 3: Outn	ut current ca	nahility may	v be limited and other	

Note 1: All parameters reflect regulator and inductor system performance. Measurements were made using a standard PI33XX evaluation board with 3x4" dimensions and 4 layer, 2oz copper. Refer to inductor pairing table within Application Description section for specific inductor manufacturer and value.

Note 2: Regulator is assured to meet performance specifications by design, test correlation, characterization, and/or statistical process control.

Note 3: Output current capability may be limited and other performance may vary from noted electrical characteristics when switching frequency or Vout is modified.

Note 4: Refer to Output Ripple plots.

Note 5: Refer to Load Current vs. Ambient Temperature curves. **Note 6:** Refer to Switching Frequency vs. Load current curves.



Specifications apply for $-40^{\circ}\text{C} < \text{T}_{J} < 125^{\circ}\text{C}$, Vin =24V, L1=125nH (Note 1) unless other conditions are noted.

Parameter	Symbol	Min	Тур	Max	Units	Conditions
Timing					-	
Switching Frequency	f_S		500		kHz	Note 6.
Fault Restart Delay	t _{FR_DLY}		30		ms	
Sync In (SYNCI)						
Synchronization Frequency Range	Δf_{SYNCI}	50		110	%	Relative to set switching frequency. Note 3.
SYNCI Threshold	V_{SYNCI}		2.5		V	
Sync Out (SYNCO)						
SYNCO High	V _{SYNCO_HI}	4.5			V	Source 1mA
SYNCO Low	V _{SYNCO_LO}			0.5	V	Sink 1mA
SYNCO Rise Time	t _{SYNCO_RT}		10	20	ns	20pF load
SYNCO Fall Time	t _{SYNCO_FT}		10	20	ns	20pF load
Soft Start And Tracking		1	1		1	•
TRK Active Range (Nominal)	V_{TRK}	0		1	V	
TRK Enable Threshold	V_{TRK_OV}	20	40	60	mV	
Charge Current (Soft – Start)	I _{TRK}	-70	-50	-30	μΑ	
Discharge Current (Fault)	I _{TRK_DIS}		6.8		mA	
Soft-Start Time	t _{ss}		2.2	2.6	ms	C _{TRK} = 0
Enable						•
High Threshold	V _{EN_HI}	0.9	1	1.1	V	
Low Threshold	V_{EN_LO}	0.7	0.8	0.9	V	
Threshold Hysteresis	V_{EN_HYS}	100	200	300	mV	
Enable Pull-Up Voltage (floating, unfaulted)	V _{EN_PU}		2		V	
Enable Pull-Down Voltage (floating, faulted)	V _{EN_PD}		0		V	
Source Current	I _{EN_SO}		-50		uA	
Sink Current	I _{EN_SK}		50		uA	
Note 1: All parameters reflect regulator and ind	uctor system		Note 3: Outp	ut current ca	apability ma	y be limited and other

Note 1: All parameters reflect regulator and inductor system performance. Measurements were made using a standard PI33XX evaluation board with 3x4" dimensions and 4 layer, 2oz copper. Refer to inductor pairing table within Application Description section for specific inductor manufacturer and value.

Note 2: Regulator is assured to meet performance specifications by design, test correlation, characterization, and/or statistical process control.

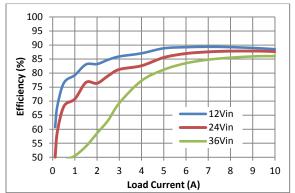
Note 3: Output current capability may be limited and other performance may vary from noted electrical characteristics when switching frequency or Vout is modified.

Note 4: Refer to Output Ripple plots.

Note 5: Refer to Load Current vs. Ambient Temperature curves. Note 6: Refer to Switching Frequency vs. Load current curves.

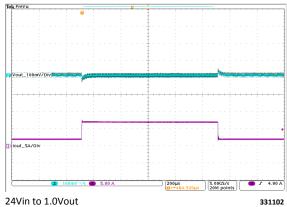


Efficiency at 25°C



Regulator and inductor performance

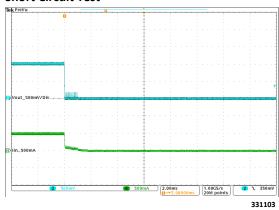
Transient Response: 2A to 7A, at 5A/μs



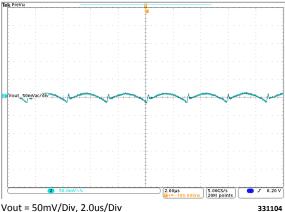
Cout = 8X 100µF Ceramic

331101

Short Circuit Test

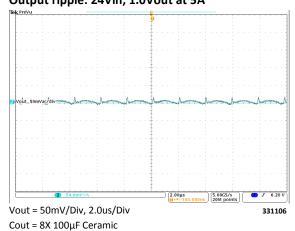


Output Ripple: 24Vin, 1.0Vout at 10A

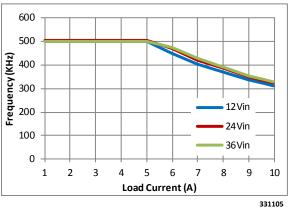


Cout = 8X 100µF Ceramic

Output ripple: 24Vin, 1.0Vout at 5A



Switching Frequency vs. Load Current



Specifications apply for -40°C < T_J < 125°C, Vin =24V, L1=155nH (Note 1) unless other conditions are noted.

Parameter	Symbol	Min	Тур	Max	Units	Conditions
Input Specifications					•	
Input Voltage	V_{IN_DC}	8	24	36	V	
Input Current	I _{IN_DC}		835		mA	Vin = 24V, T _C = 25°C, lout=10A
Input Current At Output Short (fault condition duty cycle)	I _{IN_Short}			20	mA	
Input Quiescent Current	I _{Q_VIN}		2 2.5		mA	Disabled Enabled (no load)
Input Voltage Slew Rate	V _{IN_SR}			1	V/µs	
Output Specifications				•	•	
Output Voltage Total Regulation	V _{OUT_DC}	1.77	1.8	1.82	V	Note 2.
Output Voltage Trim Range	V _{OUT_DC}	1.4		2.0	V	Note 3.
Line Regulation	$\Delta V_{OUT}(\Delta V_{IN})$		0.1		%	@25°C, 8V <vin<36v< td=""></vin<36v<>
Load Regulation	$\Delta V_{OUT}(\Delta I_{OUT})$		0.1		%	@25°C, 0.5A <lout<10a< td=""></lout<10a<>
Output Voltage Ripple	V_{OUT_AC}		25		mVp-p	lout=5A, Cout=6x100μF, 20MHz BW Note 4.
Continuous Output Current Range	I _{OUT_DC}			10	Α	Note 5.
Current Limit	I _{OUT_CL}		12		Α	
Protection						
UVLO Start Threshold	$V_{\text{UVLO_START}}$	7.20	7.60	8.00	V	
UVLO Stop Threshold	V_{UVLO_STOP}	6.90	7.25	7.60	V	
UVLO Hysteresis	V_{UVLO_HYS}	4	5	6	%	
OVLO Threshold	V _{OVLO}	37	38.4		V	
OVLO Hysteresis	V_{OVLO_HYS}		0.77		V	
UVLO/OVLO Fault Delay Time	t_{f_DLY}		128		Cycles	Number of the switching frequency cycles
UVLO/OVLO Response Time	t _f		500		ns	+1% overdrive
Output Over Voltage Protection	V _{OVP}		20		%	Above V _{OUT}
Over-Temperature Fault Threshold	T _{OTP}	130	135	140	°C	
Over-Temperature Restart Hysteresis	T _{OTP_HYS}		30		°C	

Note 1: All parameters reflect regulator and inductor system performance. Measurements were made using a standard PI33XX evaluation board with 3x4" dimensions and 4 layer, 2oz copper. Refer to inductor pairing table within Application Description section for specific inductor manufacturer and value.

Note 2: Regulator is assured to meet performance specifications by design, test correlation, characterization, and/or statistical process control.

Note 3: Output current capability may be limited and other performance may vary from noted electrical characteristics when switching frequency or Vout is modified.

Note 4: Refer to Output Ripple plots.

Note 5: Refer to Load Current vs. Ambient Temperature curves. **Note 6:** Refer to Switching Frequency vs. Load current curves.



Specifications apply for $-40^{\circ}\text{C} < \text{T}_{J} < 125^{\circ}\text{C}$, Vin =24V, L1=155nH (Note 1) unless other conditions are noted.

Parameter	Symbol	Min	Тур	Max	Units	Conditions
Timing						
Switching Frequency	f _S		600		kHz	Note 6.
Fault Restart Delay	t _{FR_DLY}		30		ms	
Sync In (SYNCI)						
Synchronization Frequency Range	Δf_{SYNCI}	50		110	%	Relative to set switching frequency. Note 3.
SYNCI Threshold	V _{SYNCI}		2.5		V	
Sync Out (SYNCO)						
SYNCO High	V _{SYNCO_HI}	4.5			V	Source 1mA
SYNCO Low	V _{SYNCO_LO}			0.5	V	Sink 1mA
SYNCO Rise Time	t _{SYNCO_RT}		10	20	ns	20pF load
SYNCO Fall Time	t _{SYNCO_FT}		10	20	ns	20pF load
Soft Start And Tracking						1
TRK Active Range (Nominal)	V_{TRK}	0		1	V	
TRK Enable Threshold	V_{TRK_OV}	20	40	60	mV	
Charge Current (Soft – Start)	I _{TRK}	-70	-50	-30	μΑ	
Discharge Current (Fault)	I _{TRK_DIS}		6.8		mA	
Soft-Start Time	t _{ss}		2.2	2.6	ms	C _{TRK} = 0
Enable						
High Threshold	V _{EN_HI}	0.9	1	1.1	V	
Low Threshold	V_{EN_LO}	0.7	0.8	0.9	V	
Threshold Hysteresis	V _{EN_HYS}	100	200	300	mV	
Enable Pull-Up Voltage (floating, unfaulted)	V _{EN_PU}		2		V	
Enable Pull-Down Voltage (floating, faulted)	V _{EN_PD}		0		V	
Source Current	I _{EN_SO}		-50		uA	
Sink Current	I _{EN_SK}		50		uA	

Note 1: All parameters reflect regulator and inductor system performance. Measurements were made using a standard PI33XX evaluation board with 3x4" dimensions and 4 layer, 2oz copper. Refer to inductor pairing table within Application Description section for specific inductor manufacturer and value.

Note 2: Regulator is assured to meet performance specifications by design, test correlation, characterization, and/or statistical process control.

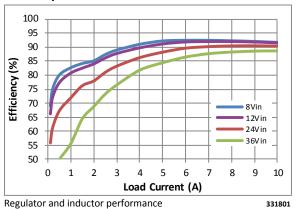
Note 3: Output current capability may be limited and other performance may vary from noted electrical characteristics when switching frequency or Vout is modified.

Note 4: Refer to Output Ripple plots.

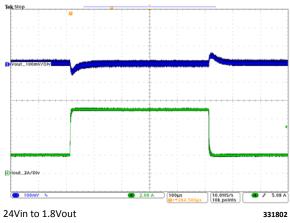
Note 5: Refer to Load Current vs. Ambient Temperature curves. **Note 6:** Refer to Switching Frequency vs. Load current curves.



Efficiency at 25°C

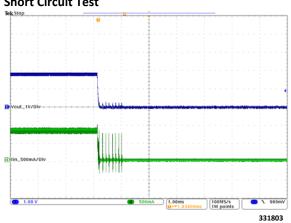


Transient Response: 2A to 7A, at 5A/μs

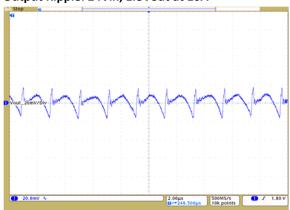


Cout = 6X 100µF Ceramic

Short Circuit Test



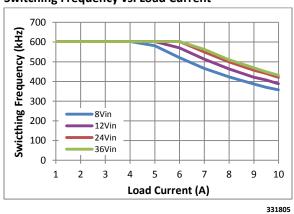
Output Ripple: 24Vin, 1.8Vout at 10A



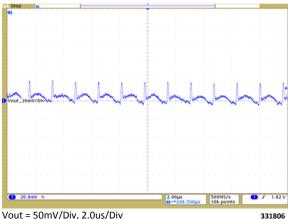
Vout = 50mV/Div, 2.0us/Div Cout = 6X 100μF Ceramic

331804

Switching Frequency vs. Load Current



Output ripple: 24Vin, 1.8Vout at 5A



Cout = 6X 100µF Ceramic



Specifications apply for $-40^{\circ}\text{C} < \text{T}_1 < 125^{\circ}\text{C}$, Vin =24V, L1=200nH (Note 1) unless other conditions are noted.

Parameter	Symbol	Min	Тур	Max	Units	Conditions
Input Specifications						
Input Voltage	V _{IN_DC}	8	24	36	V	Note 7.
Input Current	I _{IN_DC}		1.14		А	Vin = 24V, T _C = 25°C, lout=10A
Input Current At Output Short (fault condition duty cycle)	I _{IN_Short}			20	mA	
Input Quiescent Current	I _{Q_VIN}		2 2.5		mA	Disabled Enabled (no load)
Input Voltage Slew Rate	V_{IN_SR}			1	V/µs	
Output Specifications	-					
Output Voltage Total Regulation	V _{OUT_DC}	2.465	2.5	2.535	V	Note 2.
Output Voltage Trim Range	V _{OUT_DC}	2.0	2.5	3.1	V	Note 3. Note 7.
Line Regulation	$\Delta V_{OUT} (\Delta V_{IN})$		0.1		%	@25°C, 8V <vin<36v< td=""></vin<36v<>
Load Regulation	$\Delta V_{OUT} (\Delta I_{OUT})$		0.1		%	@25°C, 0.5A <lout<10a< td=""></lout<10a<>
Output Voltage Ripple	V _{OUT_AC}		28		mVp-p	lout=5A, Cout=4x100μF, 20MHz BW
Continuous Output Current Range	I _{OUT_DC}			10	Α	Note 5. Note 7.
Current Limit	I _{OUT_CL}		12		Α	
Protection						
UVLO Start Threshold	V _{UVLO_START}	7.20	7.60	8.00	V	
UVLO Stop Threshold	V _{UVLO_STOP}	6.90	7.25	7.60	V	
UVLO Hysteresis	V _{UVLO_HYS}	4	5	6	%	
OVLO Threshold	V _{OVLO}	37	38.4		V	
OVLO Hysteresis	V_{OVLO_HYS}		0.77		V	
UVLO/OVLO Fault Delay Time	t _{f_DLY}		128		Cycles	Number of the switching frequency cycles
UVLO/OVLO Response Time	t _f		500		ns	+1% overdrive
Output Over Voltage Protection	V _{OVP}		20		%	Above Vout
Over-Temperature Fault Threshold	T _{OTP}	130	135	140	°C	
Over-Temperature Restart Hysteresis	T _{OTP_HYS}		30		°C	
Note 1: All parameters reflect regulator and inductor system Note 3: Output current capability may be limited and other						

Note 1: All parameters reflect regulator and inductor system performance. Measurements were made using a standard PI33XX evaluation board with 3x4" dimensions and 4 layer, 2oz copper. Refer to inductor pairing table within Application Description section for specific inductor manufacturer and value.

Note 2: Regulator is assured to meet performance specifications by design, test correlation, characterization, and/or statistical process control.

Note 3: Output current capability may be limited and other performance may vary from noted electrical characteristics when switching frequency or Vout is modified.

Note 4: Refer to Output Ripple plots.

Rev 1.4

11/2012

Note 5: Refer to Load Current vs. Ambient Temperature curves.

Note 6: Refer to Switching Frequency vs. Load current curves.

Note 7: Minimum 5V between Vin-Vout must be maintained or a minimum load of 1mA required.



Specifications apply for $-40^{\circ}\text{C} < \text{T}_{J} < 125^{\circ}\text{C}$, Vin =24V, L1=200nH (Note 1) unless other conditions are noted.

Parameter	Symbol	Min	Тур	Max	Units	Conditions
Timing			•			
Switching Frequency	f _s		500		kHz	Note 6.
Fault Restart Delay	t _{FR_DLY}		30		ms	
Sync In (SYNCI)						•
Synchronization Frequency Range	Δf_{SYNCI}	50		110	%	Relative to set switching frequency. Note 3.
SYNCI Threshold	V_{SYNCI}		2.5		V	
Sync Out (SYNCO)						
SYNCO High	V _{SYNCO_HI}	4.5			V	Source 1mA
SYNCO Low	V _{SYNCO_LO}			0.5	V	Sink 1mA
SYNCO Rise Time	t _{SYNCO_RT}		10	20	ns	20pF load
SYNCO Fall Time	t _{SYNCO_FT}		10	20	ns	20pF load
Soft Start And Tracking	.	1	1		l	1
TRK Active Range (Nominal)	V_{TRK}	0		1	V	
TRK Enable Threshold	V _{TRK_OV}	20	40	60	mV	
Charge Current (Soft – Start)	I _{TRK}	-70	-50	-30	μΑ	
Discharge Current (Fault)	I _{TRK_DIS}		6.8		mA	
Soft-Start Time	t _{ss}		2.2	2.6	ms	C _{TRK} = 0
Enable						
High Threshold	$V_{EN_{HI}}$	0.9	1	1.1	V	
Low Threshold	V_{EN_LO}	0.7	0.8	0.9	V	
Threshold Hysteresis	V _{EN_HYS}	100	200	300	mV	
Enable Pull-Up Voltage (floating, unfaulted)	V _{EN_PU}		2		V	
Enable Pull-Down Voltage (floating, faulted)	V _{EN_PD}		0		V	
Source Current	I _{EN_SO}		-50		uA	
Sink Current Note 1: All parameters reflect regulator and in	I _{EN_SK}		50		uA	y be limited and other

Note 1: All parameters reflect regulator and inductor system performance. Measurements were made using a standard Pl33XX evaluation board with 3x4" dimensions and 4 layer, 2oz copper. Refer to inductor pairing table within Application Description section for specific inductor manufacturer and value.

Note 2: Regulator is assured to meet performance specifications by design, test correlation, characterization, and/or statistical process control.

Note 3: Output current capability may be limited and other performance may vary from noted electrical characteristics when switching frequency or Vout is modified.

Note 4: Refer to Output Ripple plots.

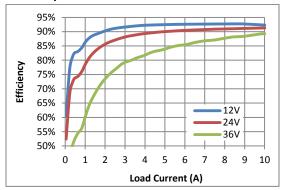
Note 5: Refer to Load Current vs. Ambient Temperature curves.

Note 6: Refer to Switching Frequency vs. Load current curves.

Note 7: Minimum 5V between Vin-Vout must be maintained or a minimum load of 1mA required.



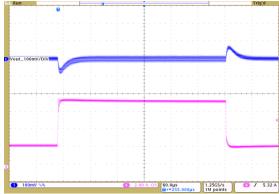
Efficiency at 25°C



Regulator and inductor performance

331201

Transient Response: 5A to 10A, at 5A/µs

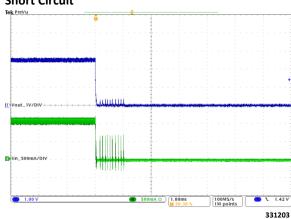


24Vin to 2.5Vout, Cout = 4 x 100μF Ceramic

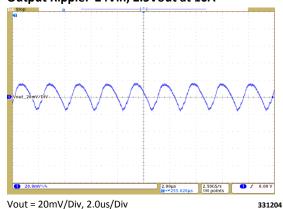
331202

Vout (Ch1) = 100mV/Div, lout (Ch4) = 2A/Div, 80us/Div

Short Circuit



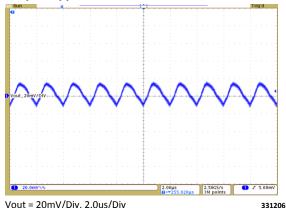
Output Ripple: 24Vin, 2.5Vout at 10A



Vout = 20mV/Div, 2.0us/Div

Cout = $4 \times 100 \mu F$ Ceramic

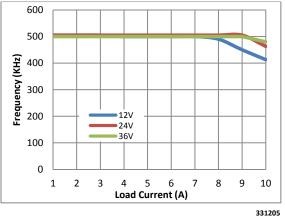
Output Ripple: 24Vin, 2.5Vout at 5A



Vout = 20mV/Div, 2.0us/Div

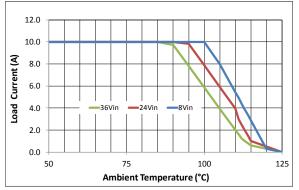
Cout = 4 x 100µF Ceramic

Switching Frequency vs. Load Current



PICOR

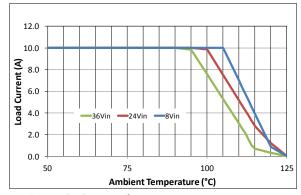
Load Current vs. Ambient Temperature, 0 LFM



Regulator and inductor performance

331207

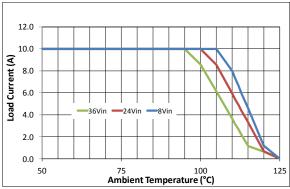
Load Current vs. Ambient Temperature, 200 LFM



Regulator and inductor performance

331208

Load Current vs. Ambient Temperature, 400 LFM



Regulator and inductor performance

331209



Specifications apply for $-40^{\circ}\text{C} < \text{T}_{1} < 125^{\circ}\text{C}$, Vin =24V, L1=200nH (Note 1) unless other conditions are noted.

Parameter	Symbol	Min	Тур	Max	Units	Conditions
Input Specifications			1			-
Input Voltage	V_{IN_DC}	8	24	36	V	Note 7.
Input Current	I _{IN_DC}		1.49		А	Vin = 24V, T _C = 25°C, lout=10A
Input Current At Output Short (fault condition duty cycle)	I _{IN_Short}			20	mA	
Input Quiescent Current	I _{Q_VIN}		2 2.5		mA	Disabled Enabled (no load)
Input Voltage Slew Rate	V_{IN_SR}			1	V/µs	
Output Specifications			•	•	•	
Output Voltage Total Regulation	V _{OUT_DC}	3.25	3.30	3.36	V	Note 2.
Output Voltage Trim Range	V _{OUT_DC}	2.3	3.3	4.1	V	Note 3. Note 7.
Line Regulation	$\Delta V_{OUT}(\Delta V_{IN})$		0.10		%	@25°C, 8V <vin<36v< td=""></vin<36v<>
Load Regulation	$\Delta V_{OUT}(\Delta I_{OUT})$		0.10		%	@25°C, 0.5A <lout<10a< td=""></lout<10a<>
Output Voltage Ripple	V _{OUT_AC}		37.5		mVp-p	Iout=5A, Cout=4x100μF, 20MHz BW Note 4.
Continuous Output Current Range	I _{OUT_DC}			10	Α	
Current Limit	I _{OUT_CL}		12		Α	
Protection						
UVLO Start Threshold	V_{UVLO_START}	7.20	7.60	8.00	V	
UVLO Stop Threshold	V _{UVLO_STOP}	6.90	7.25	7.60	V	
UVLO Hysteresis	V_{UVLO_HYS}	4	5	6	%	
OVLO Threshold	V _{OVLO}	37	38.4		V	
OVLO Hysteresis	V _{OVLO_HYS}		0.77		V	
UVLO/OVLO Fault Delay Time	t _{f_DLY}		128		Cycles	Number of the switching frequency cycles
UVLO/OVLO Response Time	t _f		500		ns	+1% overdrive
Output Over Voltage Protection	V _{OVP}		20		%	Above V _{OUT}
Over-Temperature Fault Threshold	T _{OTP}	130	135	140	°C	
Over-Temperature Restart Hysteresis Note 1: All parameters reflect regulator and ind	T _{OTP_HYS}		30		°C	y be limited and other

Note 1: All parameters reflect regulator and inductor system performance. Measurements were made using a standard PI33XX evaluation board with 3x4" dimensions and 4 layer, 2oz copper. Refer to inductor pairing table within Application Description section for specific inductor manufacturer and value.

Note 2: Regulator is assured to meet performance specifications by design, test correlation, characterization, and/or statistical process control.

Note 3: Output current capability may be limited and other performance may vary from noted electrical characteristics when switching frequency or Vout is modified.

Note 4: Refer to Output Ripple plots.

Note 5: Refer to Load Current vs. Ambient Temperature curves. Note 6: Refer to Switching Frequency vs. Load current curves. Note 7: Minimum 5V between Vin-Vout must be maintained or a minimum load of 1mA required.



Specifications apply for $-40^{\circ}\text{C} < \text{T}_{J} < 125^{\circ}\text{C}$, Vin =24V, L1=200nH (Note 1) unless other conditions are noted.

Parameter	Symbol	Min	Тур	Max	Units	Conditions
Timing			•	•		
Switching Frequency	f_S		650		kHz	Note 6.
Fault Restart Delay	t _{FR_DLY}		30		ms	
Sync In (SYNCI)					•	•
Synchronization Frequency Range	Δf_{SYNCI}	50		110	%	Relative to set switching frequency. Note 3.
SYNCI Threshold	V_{SYNCI}		2.5		V	
Sync Out (SYNCO)						
SYNCO High	V _{SYNCO_HI}	4.5			V	Source 1mA
SYNCO Low	V _{SYNCO_LO}			0.5	V	Sink 1mA
SYNCO Rise Time	t _{SYNCO_RT}		10	20	ns	20pF load
SYNCO Fall Time	t _{SYNCO_FT}		10	20	ns	20pF load
Soft Start And Tracking					1	1
TRK Active Range (Nominal)	V_{TRK}	0		1	V	
TRK Enable Threshold	V_{TRK_OV}	20	40	60	mV	
Charge Current (Soft – Start)	I _{TRK}	-70	-50	-30	μΑ	
Discharge Current (Fault)	I _{TRK_DIS}		6.8		mA	
Soft-Start Time	t _{ss}		2.2	2.6	ms	C _{TRK} = 0
Enable						
High Threshold	V _{EN_HI}	0.9	1	1.1	V	
Low Threshold	V_{EN_LO}	0.7	0.8	0.9	V	
Threshold Hysteresis	V _{EN_HYS}	100	200	300	mV	
Enable Pull-Up Voltage (floating, unfaulted)	V _{EN_PU}		2		V	
Enable Pull-Down Voltage (floating, faulted)	V _{EN_PD}		0		V	
Source Current	I _{EN_SO}		-50		uA	
Sink Current	I _{EN_SK}		50		uA	
Note 1: All parameters reflect regulator and ind	uctor system	N	lote 3: Outp	ut current ca	apability ma	y be limited and other

Note 1: All parameters reflect regulator and inductor system performance. Measurements were made using a standard PI33XX evaluation board with 3x4" dimensions and 4 layer, 2oz copper. Refer to inductor pairing table within Application Description section for specific inductor manufacturer and value.

Note 2: Regulator is assured to meet performance specifications by design, test correlation, characterization, and/or statistical process control.

Note 3: Output current capability may be limited and other performance may vary from noted electrical characteristics when switching frequency or Vout is modified.

Note 4: Refer to Output Ripple plots.

Rev 1.4

11/2012

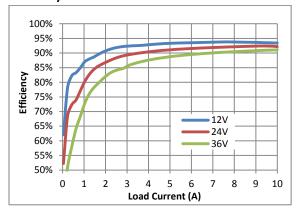
Note 5: Refer to Load Current vs. Ambient Temperature curves.

Note 6: Refer to Switching Frequency vs. Load current curves.

Note 7: Minimum 5V between Vin-Vout must be maintained or a minimum load of 1mA required.



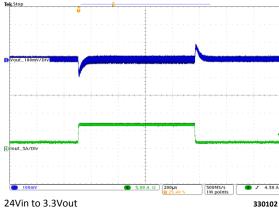
Efficiency at 25°C



Regulator and inductor performance

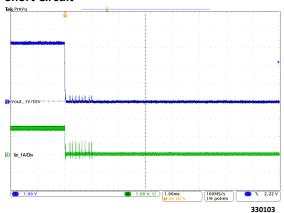
330101

Transient Response: 2A to 7A, at 5A/μs

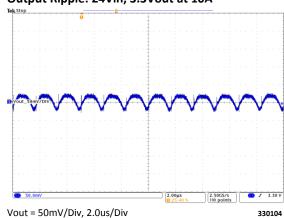


Cout = 4 x 100µF Ceramic

Short Circuit

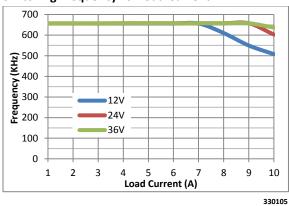


Output Ripple: 24Vin, 3.3Vout at 10A

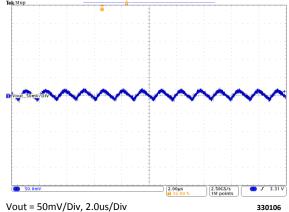


Cout = 4 x 100µF Ceramic

Switching Frequency vs. Load Current



Output Ripple: 24Vin, 3.3Vout at 5A

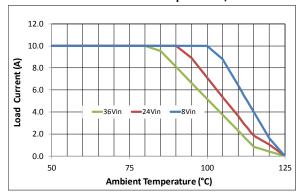


Vout = 50mV/Div, 2.0us/Div

Cout = $4 \times 100 \mu F$ Ceramic



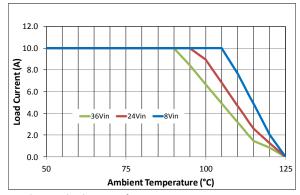
Load Current vs. Ambient Temperature, 0 LFM



Regulator and inductor performance

330107

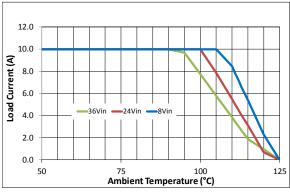
Load Current vs. Ambient Temperature, 200 LFM



Regulator and inductor performance

330108

Load Current vs. Ambient Temperature, 400 LFM



Regulator and inductor performance

330109



Specifications apply for $-40^{\circ}\text{C} < \text{T}_{1} < 125^{\circ}\text{C}$, Vin =24V, L1=200nH (Note 1) unless other conditions are noted.

Parameter	Symbol	Min	Тур	Max	Units	Conditions
Input Specifications			•	•	•	
Input Voltage	V _{IN_DC}	8	24	36	V	Note 7.
Input Current	I _{IN_DC}		2.23		А	Vin = 24V, T _C = 25°C, lout=10A
Input Voltage Slew Rate	V_{IN_SR}			1	V/µs	
Input Current At Output Short (fault condition duty cycle)	I _{IN_Short}			20	mA	
Input Quiescent Current	I _{Q_VIN}		2 2.5		mA	Disabled Enabled (no load)
Input Voltage Slew Rate	V _{IN_SR}			1	V/µs	
Output Specifications			•	•	•	
Output Voltage Total Regulation	V _{OUT_DC}	4.93	5.00	5.07	V	Note 2.
Output Voltage Trim Range		3.3		6.5	V	Note 3. Note 7.
Line Regulation	$\Delta V_{OUT}(\Delta V_{IN})$		0.1	0.15	%	@25°C, 8V <vin<36v< td=""></vin<36v<>
Load Regulation	$\Delta V_{OUT}(\Delta I_{OUT})$		0.1	0.15	%	@25°C, 0.5A <lout<10a< td=""></lout<10a<>
Output Voltage Ripple	V _{OUT_AC}		30		mVp-p	Iout=5A, Cout=4x47μF 20MHz BW Note 4.
Continuous Output Current Range	I _{OUT_DC}			10	Α	Note 5. Note 7.
Current Limit	I _{OUT_CL}		12		Α	
Protection						
UVLO Start Threshold	V_{UVLO_START}	7.20	7.60	8.00	V	
UVLO Stop Threshold	V_{UVLO_STOP}	6.90	7.25	7.60	V	
UVLO Hysteresis	$V_{\text{UVLO_HYS}}$	4	5	6	%	
OVLO Threshold	V _{OVLO}	37	38.4		V	
OVLO Hysteresis	V _{OVLO_HYS}		0.77		V	
UVLO/OVLO Fault Delay Time	t _{f_DLY}		128		Cycles	Number of the switching frequency cycles
UVLO/OVLO Response Time	t _f		500		ns	+1% overdrive
Output Over Voltage Protection	V _{OVP}		20		%	Above Vout
Over-Temperature Fault Threshold	T _{OTP}	130	135	140	°C	
Over-Temperature Restart Hysteresis	T _{OTP_HYS}		30		°C	
Note 1: All parameters reflect regulator and inc	reflect regulator and inductor system Note 3: Output current capability may be limited and					av be limited and other

Note 1: All parameters reflect regulator and inductor system performance. Measurements were made using a standard PI33XX evaluation board with 3x4" dimensions and 4 layer, 2oz copper. Refer to inductor pairing table within Application Description section for specific inductor manufacturer and value.

Note 2: Regulator is assured to meet performance specifications by design, test correlation, characterization, and/or statistical process control.

Note 3: Output current capability may be limited and other performance may vary from noted electrical characteristics when switching frequency or Vout is modified.

Note 4: Refer to Output Ripple plots.

Note 5: Refer to Load Current vs. Ambient Temperature curves.

Note 6: Refer to Switching Frequency vs. Load current curves.

Note 7: Minimum 5V between Vin-Vout must be maintained or a minimum load of 3mA required.



Specifications apply for $-40^{\circ}\text{C} < \text{T}_{J} < 125^{\circ}\text{C}$, Vin =24V, L1=200nH (Note 1) unless other conditions are noted.

Parameter	Symbol	Min	Тур	Max	Units	Conditions
Timing						
Switching Frequency	f _S		1.0		MHz	Note 6.
Fault Restart Delay	t _{FR_DLY}		30		ms	
Sync In (SYNCI)						
Synchronization Frequency Range	Δf_{SYNCI}	50		110	%	Relative to set switching frequency. Note 3.
SYNCI Threshold	V _{SYNCI}		2.5		٧	
Sync Out (SYNCO)						
SYNCO High	V _{SYNCO_HI}	4.5			V	Source 1mA
SYNCO Low	V _{SYNCO_LO}			0.5	٧	Sink 1mA
SYNCO Rise Time	t _{SYNCO_RT}		10	20	ns	20pF load
SYNCO Fall Time	t _{SYNCO_FT}		10	20	ns	20pF load
Soft Start And Tracking	l	l			l	1
TRK Active Range (Nominal)	V_{TRK}	0		1	V	
TRK Enable Threshold	V_{TRK_OV}	20	40	60	mV	
Charge Current (Soft – Start)	I _{TRK}	-70	-50	-30	μΑ	
Discharge Current (Fault)	I _{TRK_DIS}		6.8		mA	
Soft-Start Time	t _{ss}		2.2	2.6	ms	C _{TRK} = 0
Enable						
High Threshold	V _{EN_HI}	0.9	1	1.1	V	
Low Threshold	V_{EN_LO}	0.7	0.8	0.9	V	
Threshold Hysteresis	V _{EN_HYS}	100	200	300	mV	
Enable Pull-Up Voltage (floating, unfaulted)	V _{EN_PU}		2		V	
Enable Pull-Down Voltage (floating, faulted)	V _{EN_PD}		0		V	
Source Current	I _{EN_SO}		-50		uA	
Sink Current Note 1: All parameters reflect regulator and indu	I _{EN_SK}		50		uA	

Note 1: All parameters reflect regulator and inductor system performance. Measurements were made using a standard PI33XX evaluation board with 3x4" dimensions and 4 layer, 2oz copper. Refer to inductor pairing table within Application Description section for specific inductor manufacturer and value.

Note 2: Regulator is assured to meet performance specifications by design, test correlation, characterization, and/or statistical process control.

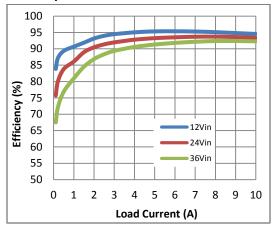
Note 3: Output current capability may be limited and other performance may vary from noted electrical characteristics when switching frequency or Vout is modified.

Note 4: Refer to Output Ripple plots.

Note 5: Refer to Load Current vs. Ambient Temperature curves. Note 6: Refer to Switching Frequency vs. Load current curves. Note 7: Minimum 5V between Vin-Vout must be maintained or a minimum load of 3mA required.



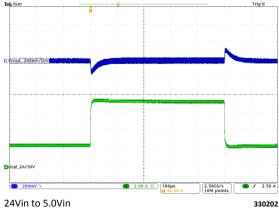
Efficiency at 25°C



Regulator and inductor performance

330201

Transient Response: 2A to 7A, at 5A/μs

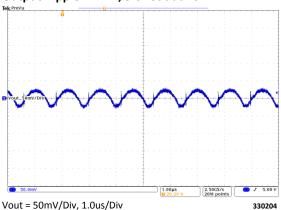


Cout = 4 X 47uF Ceramic

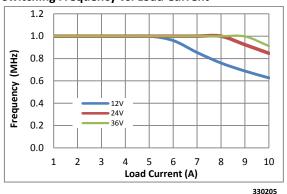
Short Circuit Test



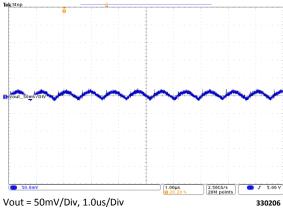
Output Ripple: 24Vin, 5.0Vout at 10A



Switching Frequency vs. Load Current



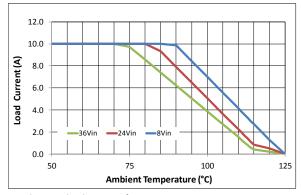
Output Ripple: 24Vin, 5.0Vout at 5A



Cout = 4 X 47uF Ceramic

Cout = $4 \times 47 \mu F$ Ceramic

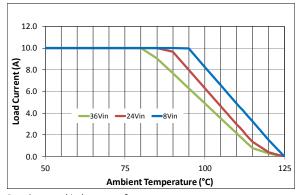
Load Current vs. Ambient Temperature, 0 LFM



Regulator and inductor performance

330207

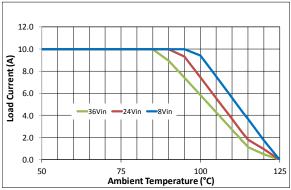
Load Current vs. Ambient Temperature, 200 LFM



Regulator and inductor performance

330208

Load Current vs. Ambient Temperature, 400 LFM



Regulator and inductor performance

330209



Specifications apply for $-40^{\circ}\text{C} < \text{T}_{1} < 125^{\circ}\text{C}$, Vin =24V, L1=230nH (Note 1) unless other conditions are noted.

Parameter	Symbol	Min	Тур	Max	Units	Conditions
Input Specifications						
Input Voltage	V _{IN_DC}	17.4	24	36	V	Note 7.
Input Current	I _{IN_DC}		4.15		А	Vin = 24V, T _C = 25°C, lout=8A
Input Current At Output Short (fault condition duty cycle)	I _{IN_Short}			20	mA	
Input Quiescent Current	I _{Q_VIN}		2 2.5		mA	Disabled Enabled (no load)
Input Voltage Slew Rate	V _{IN_SR}			1	V/µs	
Output Specifications		•		•	•	
Output Voltage Total Regulation	V _{OUT_DC}	11.82	12.0	12.18	V	Note 2.
Output Voltage Trim Range	V _{OUT_DC}	6.5	12	13.0	V	Note 3. Note 7.
Line Regulation	$\Delta V_{OUT}(\Delta V_{IN})$		0.1		%	@25°C, 17.4V <vin<36v< td=""></vin<36v<>
Load Regulation	$\Delta V_{OUT}(\Delta I_{OUT})$		0.1		%	@25°C, 0.5A <lout<8a< td=""></lout<8a<>
Output Voltage Ripple	V _{OUT_AC}		60		mVp-p	lout=4A, Cout=4x22μF, 20MHz BW Note 4.
Continuous Output Current Range	I _{OUT_DC}			8	Α	Note 5.
Current Limit	I _{OUT_CL}		9		Α	
Protection						
UVLO Start Threshold	V_{UVLO_START}	15.80	16.60	17.40	V	
UVLO Stop Threshold	V_{UVLO_STOP}	15.00	15.80	16.60	V	
UVLO Hysteresis	$V_{\text{UVLO_HYS}}$	4	5	6	%	
OVLO Threshold	V _{OVLO}	37	38.4		V	
OVLO Hysteresis	V _{OVLO_HYS}		0.77		V	
UVLO/OVLO Fault Delay Time	t _{f_DLY}		128		Cycles	Number of the switching frequency cycles
UVLO/OVLO Response Time	t _f		500		ns	+1% overdrive
Output Over Voltage Protection	V _{OVP}		20		%	Above Vout
Over-Temperature Fault Threshold	T _{OTP}	130	135	140	°C	
Over-Temperature Restart Hysteresis Note 1: All parameters reflect regulator and ind	T _{OTP_HYS}		30		°C	y be limited and other

Note 1: All parameters reflect regulator and inductor system performance. Measurements were made using a standard PI33XX evaluation board with 3x4" dimensions and 4 layer, 2oz copper. Refer to inductor pairing table within Application Description section for specific inductor manufacturer and value.

Note 2: Regulator is assured to meet performance specifications by design, test correlation, characterization, and/or statistical process control.

Note 3: Output current capability may be limited and other performance may vary from noted electrical characteristics when switching frequency or Vout is modified.

Note 4: Refer to Output Ripple plots.

Note 5: Refer to Load Current vs. Ambient Temperature curves.

Note 6: Refer to Switching Frequency vs. Load current curves.

Note 7: Minimum 5V between Vin-Vout must be maintained or a minimum load of 1mA required.



Specifications apply for $-40^{\circ}\text{C} < \text{T}_{J} < 125^{\circ}\text{C}$, Vin =24V, L1=230nH (Note 1) unless other conditions are noted.

Parameter	Symbol	Min	Тур	Max	Units	Conditions
Timing			I		1	
Switching Frequency	f _S		1.4		MHz	Note 6.
Fault Restart Delay	t _{FR_DLY}		30		ms	
Sync In (SYNCI)						
Synchronization Frequency Range	Δf_{SYNCI}	50		110	%	Relative to set switching frequency. Note 3.
SYNCI Threshold	V_{SYNCI}		2.5		V	
Sync Out (SYNCO)						
SYNCO High	V _{SYNCO_HI}	4.5			V	Source 1mA
SYNCO Low	V _{SYNCO_LO}			0.5	V	Sink 1mA
SYNCO Rise Time	t _{SYNCO_RT}		10	20	ns	20pF load
SYNCO Fall Time	t _{SYNCO_FT}		10	20	ns	20pF load
Soft Start And Tracking	l		I		I	
TRK Active Range (Nominal)	V_{TRK}	0		1	V	
TRK Enable Threshold	V_{TRK_OV}	20	40	60	mV	
Charge Current (Soft – Start)	I _{TRK}	-70	-50	-30	μΑ	
Discharge Current (Fault)	I _{TRK_DIS}		6.8		mA	
Soft-Start Time	t _{ss}		2.2	2.6	ms	C _{TRK} = 0
Enable						
High Threshold	V_{EN_HI}	0.9	1	1.1	V	
Low Threshold	V _{EN_LO}	0.7	0.8	0.9	V	
Threshold Hysteresis	V _{EN_HYS}	100	200	300	mV	
Enable Pull-Up Voltage (floating, unfaulted)	V _{EN_PU}		2		V	
Enable Pull-Down Voltage (floating, faulted)	V _{EN_PD}		0		V	
Source Current	I _{EN_SO}		-50		uA	
Sink Current	I _{EN_SK}		50		uA	

Note 1: All parameters reflect regulator and inductor system performance. Measurements were made using a standard PI33XX evaluation board with 3x4" dimensions and 4 layer, 2oz copper. Refer to inductor pairing table within Application Description section for specific inductor manufacturer and value.

Note 2: Regulator is assured to meet performance specifications by design, test correlation, characterization, and/or statistical process control.

Note 3: Output current capability may be limited and other performance may vary from noted electrical characteristics when switching frequency or Vout is modified.

Note 4: Refer to Output Ripple plots.

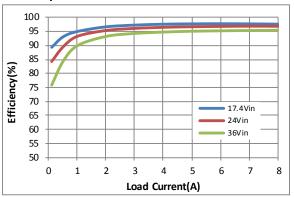
Note 5: Refer to Load Current vs. Ambient Temperature curves.

Note 6: Refer to Switching Frequency vs. Load current curves.

Note 7: Minimum 5V between Vin-Vout must be maintained or a minimum load of 1mA required.



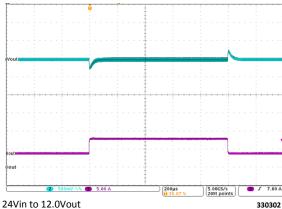
Efficiency at 25°



Regulator and inductor performance

330301

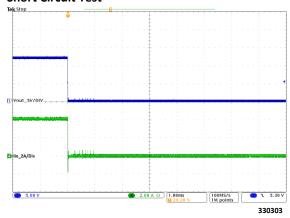
Transient Response: 4A to 8A, at 5A/μs



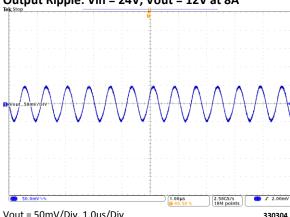
24Vin to 12.0Vout

Cout = 4 X 22uF Ceramic

Short Circuit Test



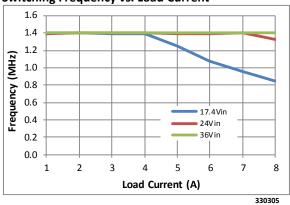
Output Ripple: Vin = 24V, Vout = 12V at 8A



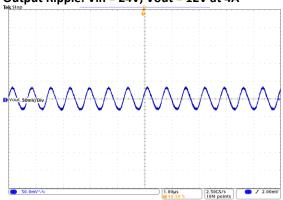
Vout = 50mV/Div, 1.0us/Div

Cout = 4 X 22uF Ceramic

Switching Frequency vs. Load Current



Output Ripple: Vin = 24V, Vout = 12V at 4A



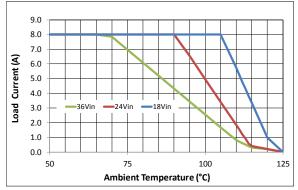
Vout = 50mV/Div, 1.0us/Div

Cout = 4 X 22uF Ceramic



330306

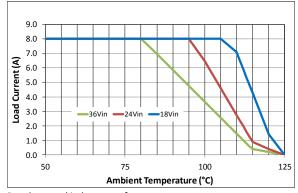
Load Current vs. Ambient Temperature, 0 LFM



Regulator and inductor performance

330307

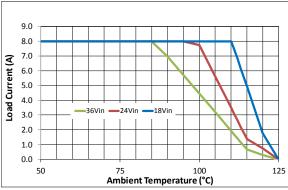
Load Current vs. Ambient Temperature, 200 LFM



Regulator and inductor performance

330308

Load Current vs. Ambient Temperature, 400 LFM



Regulator and inductor performance

330309



Specifications apply for $-40^{\circ}\text{C} < \text{T}_{1} < 125^{\circ}\text{C}$, Vin =24V, L1=230nH (Note 1) unless other conditions are noted.

Parameter	Symbol	Min	Тур	Max	Units	Conditions
Input Specifications				•		
Input Voltage	V _{IN_DC}	20.4	24	36	V	Note 7.
Input Current	I _{IN_DC}		5.15		А	Vin = 24V, T _C = 25°C, lout=8A
Input Current At Output Short (fault condition duty cycle)	I _{IN_Short}			20	mA	
Input Quiescent Current	I _{Q_VIN}		2 2.5		mA	Disabled Enabled (no load)
Input Voltage Slew Rate	V _{IN_SR}			1	V/µs	
Output Specifications					•	-
Output Voltage Total Regulation	V_{OUT_DC}	14.78	15.0	15.23	V	Note 2.
Output Voltage Trim Range		10.0	15	16	V	Note 3. Note 7.
Line Regulation	$\Delta V_{OUT}(\Delta V_{IN})$		0.1		%	@25°C, 20.4V <vin<36v< td=""></vin<36v<>
Load Regulation	$\Delta V_{OUT}(\Delta I_{OUT})$		0.1		%	@25°C, 0.5A <lout<8a< td=""></lout<8a<>
Output Voltage Ripple	V _{OUT_AC}		60		mVp-p	lout=4A, Cout=4x22μF, 20MHz BW Note 4.
Continuous Output Current Range	I _{OUT_DC}			8	Α	Note 5. Note 7.
Current Limit	I _{OUT_CL}		9		Α	
Protection						
UVLO Start Threshold	V _{UVLO_START}	18.4	19.4	20.4	V	
UVLO Stop Threshold	V _{UVLO_STOP}	17.4	18.4	19.4	V	
UVLO Hysteresis	V _{UVLO_HYS}	4	5	6	%	
OVLO Threshold	V _{OVLO}	37	38.4		V	
OVLO Hysteresis	V _{OVLO_HYS}		0.77		V	
UVLO/OVLO Fault Delay Time	t _{f_DLY}		128		Cycles	Number of the switching frequency cycles
UVLO/OVLO Response Time	t _f		500		ns	+1% overdrive
Output Over Voltage Protection	V _{OVP}		20		%	Above Vout
Over-Temperature Fault Threshold	T _{OTP}	130	135	140	°C	
Over-Temperature Restart Hysteresis	T _{OTP_HYS}		30		°C	

Note 1: All parameters reflect regulator and inductor system performance. Measurements were made using a standard Pl33XX evaluation board with 3x4" dimensions and 4 layer, 2oz copper. Refer to inductor pairing table within Application Description section for specific inductor manufacturer and value.

Note 2: Regulator is assured to meet performance specifications by design, test correlation, characterization, and/or statistical process control.

Note 3: Output current capability may be limited and other performance may vary from noted electrical characteristics when switching frequency or Vout is modified.

Note 4: Refer to Output Ripple plots.

Note 5: Refer to Load Current vs. Ambient Temperature curves.

Note 6: Refer to Switching Frequency vs. Load current curves.

Note 7: Minimum 5V between Vin-Vout must be maintained or a minimum load of 1mA required.



Specifications apply for -40° C < T_1 < 125 $^{\circ}$ C, Vin =24V, L1=230nH (Note 1) unless other conditions are noted.

Parameter	Symbol	Min	Тур	Max	Units	Conditions
Timing						
Switching Frequency	f _S		1.5		MHz	Note 6.
Fault Restart Delay	t _{FR_DLY}		30		ms	
Sync In (SYNCI)						
Synchronization Frequency Range	Δf_{SYNCI}	50		110	%	Relative to set switching frequency. Note 3.
SYNCI Threshold	V _{SYNCI}		2.5		٧	
Sync Out (SYNCO)						
SYNCO High	V _{SYNCO_HI}	4.5			V	Source 1mA
SYNCO Low	V _{SYNCO_LO}			0.5	V	Sink 1mA
SYNCO Rise Time	t _{SYNCO_RT}		10	20	ns	20pF load
SYNCO Fall Time	t _{SYNCO_FT}		10	20	ns	20pF load
Soft Start And Tracking	I			l	l	1
TRK Active Range (Nominal)	V_{TRK}	0		1	V	
TRK Enable Threshold	V_{TRK_OV}	20	40	60	mV	
Charge Current (Soft – Start)	I _{TRK}	-70	-50	-30	μΑ	
Discharge Current (Fault)	I _{TRK_DIS}		6.8		mA	
Soft-Start Time	t _{ss}		2.2	2.6	ms	C _{TRK} = 0
Enable						
High Threshold	V _{EN_HI}	0.9	1	1.1	V	
Low Threshold	V_{EN_LO}	0.7	0.8	0.9	V	
Threshold Hysteresis	V _{EN_HYS}	100	200	300	mV	
Enable Pull-Up Voltage (floating, unfaulted)	V _{EN_PU}		2		V	
Enable Pull-Down Voltage (floating, faulted)	V _{EN_PD}		0		V	
Source Current	I _{EN_SO}		-50		uA	
Sink Current Note 1: All parameters reflect regulator and indi	I _{EN_SK}		50		uA	y be limited and other

Note 1: All parameters reflect regulator and inductor system performance. Measurements were made using a standard PI33XX evaluation board with 3x4" dimensions and 4 layer, 2oz copper. Refer to inductor pairing table within Application Description section for specific inductor manufacturer and value.

Note 2: Regulator is assured to meet performance specifications by design, test correlation, characterization, and/or statistical process control.

Note 3: Output current capability may be limited and other performance may vary from noted electrical characteristics when switching frequency or Vout is modified.

Note 4: Refer to Output Ripple plots.

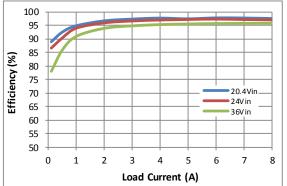
Note 5: Refer to Load Current vs. Ambient Temperature curves.

Note 6: Refer to Switching Frequency vs. Load current curves.

Note 7: Minimum 5V between Vin-Vout must be maintained or a minimum load of 1mA required.



Efficiency at 25°C



Regulator and inductor performance

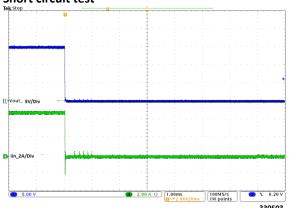
Transient Response: 2A to 6A, at 5A/μs



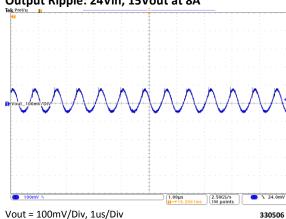
330501

Cout = $4x22\mu$ F Ceramic

Short circuit test

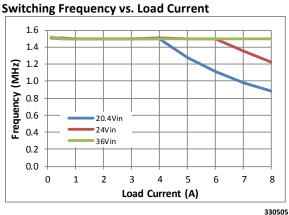


Output Ripple: 24Vin, 15Vout at 8A

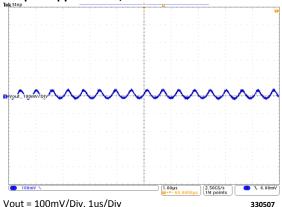


Vout = 100mV/Div, 1us/Div

Cout = 4x22µF Ceramic



Output Ripple: 24Vin, 15Vout at 4A

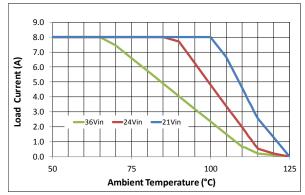


Vout = 100mV/Div, 1us/Div

Cout = $4x22\mu$ F Ceramic



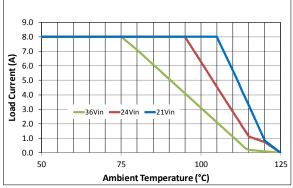
Load Current vs. Ambient Temperature, 0 LFM



Regulator and inductor performance

330507

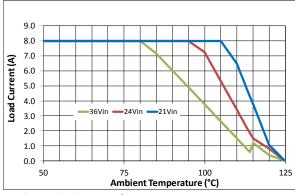
Load Current vs. Ambient Temperature, 200 LFM



Regulator and inductor performance

330508

Load Current vs. Ambient Temperature, 400 LFM



Regulator and inductor performance

330509



Functional Description

The PI33XX is a family of highly integrated ZVS-Buck regulators. The PI33XX has a set output voltage that is trimmable within a prescribed range shown in Table 1. Performance and maximum output current are characterized with a specific external power inductor (see Table 5).

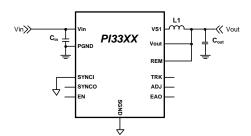


Figure 2 - ZVS-Buck with required components

For basic operation, Figure 2 shows the connections and components required. No additional design or settings are required.

ENABLE (EN)

EN is the enable pin of the converter. The EN Pin is referenced to SGND and permits the user to turn the converter on or off. The EN default polarity is a positive logic assertion. If the EN pin is left floating or asserted high, the converter output is enabled. Pulling EN pin below 0.8 Vdc with respect to SGND will disable the regulator output.

The EN input polarity can be programmed (PI33XX-20 and PI33XX-21 versions only) via the I²C data bus. When the EN pin polarity is programmed for negative logic assertion; and if the EN pin is left floating, the regulator output is enabled. Pulling the EN pin above 1.0 Vdc with respect to SGND, will disable the regulator output.

Switching Frequency Synchronization

The SYNCI input allows the user to synchronize the controller switching frequency by an external clock referenced to SGND. The external clock can synchronize the unit between 50% and 110% of the preset switching frequency (f_s). For PI33XX-20 and PI33XX-21 versions only, the phase delay can be programmed via I^2C bus with respect to the clock

applied at SYNCI pin. Phase delay allows PI33XX regulators to be paralleled and operate in an interleaving mode.

The PI33XX default for SYNCI is to sync with respect to the falling edge of the applied clock providing 180° phase shift from SYNCO. This allows for the paralleling of two PI33XX devices without the need for further user programming or external sync clock circuitry. The user can change the SYNCI polarity to sync with the external clock rising edge via the I²C data bus (PI33XX-20 and PI33XX-21 versions only).

When using the internal oscillator, the SYNCO pin provides a 5V clock that can be used to sync other regulators. Therefore, one PI33XX can act as the lead regulator and have additional PI33XXs running in parallel and interleaved.

Output Voltage Trim

The PI33XX output voltage can be trimmed up from the preset output by connecting a resistor from ADJ pin to SGND and can be trimmed down by connecting a resistor from ADJ pin to VOUT. The Table 2 defines the voltage ranges for the PI33XX family.

Device	Οι	itput Voltage
	Set	Range
PI3311-X0-LGIZ	1.0V	1.0 to 1.4V
PI3318-X0-LGIZ	1.8V	1.4 to 2.0V
PI3312-X0-LGIZ	2.5V	2.0 to 3.1V
PI3301-X0-LGIZ	3.3V	2.3 to 4.1V
PI3302-X0-LGIZ	5.0V	3.3 to 6.5V
PI3303-X0-LGIZ	12V	6.5 to 13.0V
PI3305-X0-LGIZ	15V	10.0 to 16.0V

Table 2 - PI33XX family output voltage ranges.

Soft-Start

The PI33XX includes an internal soft-start capacitor to ramp the output voltage in 2ms from 0V to full output voltage. Connecting an external capacitor from the TRK pin to SGND will increase the start-up ramp period. See, "Soft Start Adjustment and Track,"



in the Applications Description section for more details.

Remote Sensing

An internal 100Ω resistor is connected between REM pin and VOUT pin to provide regulation when the REM is left open. With REM open, the converter will regulate 100mV above its set point. Connect REM to the desired reference node to be regulated.

Output Current Limit Protection

PI33XX has two methods implemented to protect from output short or over current condition.

Slow Current Limit protection: prevents the output load from sourcing current higher than the regulator's maximum rated current. If the output current exceeds the Current Limit (I_{OUT_CL}) for 1024us, a slow current limit fault is initiated and the regulator is shutdown which eliminates output current flow. After Fault Restart Delay (t_{FR_DLY}), a soft-start cycle is initiated. This restart cycle will be repeated indefinitely until the excessive load is removed.

Fast Current Limit protection: PI33XX monitors the regulator inductor current pulse-by-pulse to prevent the output from supplying very high current due to a sudden low impedance short. If the regulator senses a high inductor current pulse, it will initiate a fault and stop switching until Fault Restart Delay ends and then initiate a soft-start cycle.

Both the Fast and Slow current limit faults are stored in a Fault Register and can be read and cleared (PI33XX-20 and PI33XX-21 versions only) via I²C data bus.

Input Under-Voltage Lockout

If VIN falls below the input Under Voltage Lockout (UVLO) threshold, but remains high enough to power the internal bias supply, the PI33XX will complete the current cycle and stop switching. If VIN recovers within 128 switching cycles, the PI33XX will resume normal operation. If this time limit is exceeded, the system will enter a low power state and initiate a fault. The system will restart once the input voltage is reestablished and after the Fault Restart Delay. A UVLO fault is stored in a Fault Register and can be

read and cleared (PI33XX-20 and PI33XX-21 versions only) via I²C data bus.

Input Over Voltage Lockout

If VIN exceeds the input Over Voltage Lockout (OVLO) threshold (V_{OVLO}), while the controller is running, the PI33XX will complete the current cycle and stop switching. If VIN recovers within 128 switching cycles, the PI33XX will resume normal operation. Otherwise, the system will enter a low power state and sets an OVLO fault. The system will resume operation when the input voltage falls below 98% of the OVLO threshold and after the Fault Restart Delay. The OVLO fault is stored in a Fault Register and can be read and cleared (PI33XX-20 and PI33XX-21 versions only) via I²C data bus.

Output Over Voltage Protection

The PI33XX family is equipped with output Over Voltage Protection (OVP) to prevent damage to input voltage sensitive devices. If the output voltage exceeds 20% of its set regulated value, the regulator will complete the current cycle, stop switching and issue an OVP fault. The system will resume operation once the output voltage falls below the OVP threshold and after Fault Restart Delay. The OVP fault is stored in a Fault Register and can be read and cleared (PI33XX-20 and PI33XX-21 versions only) via I²C data bus.

Over Temperature Protection

The internal package temperature is monitored to prevent internal components from reaching their thermal maximum. If the Over Temperature Protection Threshold (OTP) is exceeded (T_{OTP}), the regulator will complete the current switching cycle, enter a low power mode, set a fault flag, and will soft-start when the internal temperature falls below Over-Temperature Restart Hysteresis (T_{OTP_HYS}). The OTP fault is stored in a Fault Register and can be read and cleared (PI33XX-20 and PI33XX-21 versions only) via I²C data bus.

Pulse Skip Mode (PSM)

PI33XX features a PSM to achieve high efficiency at light loads. The regulators are setup to skip pulses if



EAO falls below a PSM threshold. Depending on conditions and component values, this may result in single pulses or several consecutive pulses followed by skipped pulses. Skipping cycles significantly reduces gate drive power and improves light load efficiency. The regulator will leave PSM once the EAO rises above the Skip Mode threshold.

Variable Frequency Operation

Each PI33XX is preprogrammed to a base operating frequency, with respect to the power stage inductor (see Table 5), to operate at peak efficiency across line and load variations. At low line and high load applications, the base frequency will decrease to accommodate these extreme operating ranges. By stretching the frequency, the ZVS operation is preserved throughout the total input line voltage range therefore maintaining optimum efficiency.

Parallel Operation

Paralleling multiple PI33XX modules can be used to increase the output current capability of a single power rail and reduce output voltage ripple.

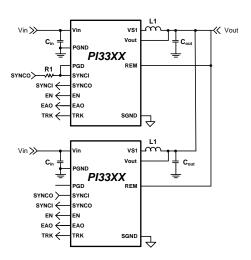


Figure 3 - PI33XX parallel operation

By connecting the EAO pins and SGND pins of each module together the units will share the current equally. When the TRK pins of each unit are connected together, the units will track each other during soft-start and all unit EN pins have to be released to allow the units to start (See Figure 3). Also, any fault event in any regulator will disable the

other regulators. The two regulators will be out of phase with each other reducing output ripple (refer to Switching Frequency Synchronization).

To provide synchronization between regulators over the entire operational frequency range, the Parallel Good (PGD) pin must be connected to the lead regulator's SYNCI pin and a $2.5k\Omega$ Resistor, R1, must be placed between SYNCO return and the lead regulator's SYNCI pin, as shown in Figure 3. In this configuration, at system soft-start, the PGD pin pulls SYNCI low forcing the lead regulator to initialize the open-loop startup synchronization. Once the regulators reach regulation, SYNCI is released and the system is now synchronized in a closed-loop configuration which allows the system to adjust, on the fly, when any of the individual regulators begin to enter variable frequency mode in the loop.

Multi-phasing three regulators is possible (PI33XX-20 and PI33XX-21 only) with no change to the basic single-phase design. For more information about how to program phase delays within the regulator, please refer to Picor application note PI33XX-2X Multi-Phase Design Guide.

I²C Interface Operation

PI33XX-20 and PI33XX-21 provide an I²C digital interface that enables the user to program the EN pin polarity (from high to low assertion) and switching frequency synchronization phase/delay. These are one time programmable options to the device.

Also, the PI33XX-20 and PI33XX-21 allow for dynamic Vout margining via I²C that is useful during development (settings stored in volatile memory only and not retained by the device). The PI33XX-20 and PI33XX-21 also have the option for fault telemetry including:

- Over temperature protection
- Fast/Slow current limit
- Output voltage high
- Input overvoltage
- Input undervoltage



For more information about how to utilize the I²C interface please refer to Picor application note *PI33XX-2X I*²C *Digital Interface Guide*.

Application Description

Output Voltage Trim

The PI33XX family of Buck Regulators provides seven common output voltages: 1.0V, 1.8V, 2.5V, 3.3V, 5.0V, 12V and 15V. A post-package trim step is implemented to offset any resistor divider network errors ensuring maximum output accuracy. With a single resistor connected from the ADJ pin to SGND or REM, each device's output can be varied above or below the nominal set voltage (with the exception of the PI3311-X0 which can only be above the set voltage of 1V).

Device	Out	put Voltage
Device	Set	Range
PI3311-X0-LGIZ	1.0V	1.0 to 1.4V
PI3318-X0-LGIZ	1.8V	1.4 to 2.0V
PI3312-X0-LGIZ	2.5V	2.0 to 3.1V
PI3301-X0-LGIZ	3.3V	2.3 to 4.1V
PI3302-X0-LGIZ	5.0V	3.3 to 6.5V
PI3303-X0-LGIZ	12V	6.5 to 13.0V
PI3305-X0-LGIZ	15V	10.0 to 16.0V

Table 3 - PI33XX family output voltage ranges

The remote pin (REM) should always be connected to the VOUT pin, if not used, to prevent an output voltage offset. Figure 4 shows the internal feedback voltage divider network.

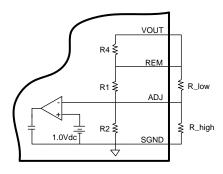


Figure 4 - Internal resistor divider network

R1, R2, and R4 are all internal 1.0 % resistors and R_low and R_high are external resistors for which the designer can add to modify VOUT to a desired output. The internal resistor value for each regulator is listed below in Table 4.

Device	R1	R2	R4
PI3311-X0-LGIZ	1k	Open	100
PI3318-X0-LGIZ	0.806K	1.0k	100
PI3312-X0-LGIZ	1.5k	1.0k	100
PI3301-X0-LGIZ	2.61k	1.13k	100
PI3302-X0-LGIZ	4.53k	1.13k	100
PI3303-X0-LGIZ	11.0k	1.0k	100
PI3305-X0-LGIZ	14.0k	1.0k	100

Table 4 - PI33XX Internal divider values

By choosing an output voltage value within the ranges stated in Table 3, VOUT can simply be adjusted up or down by selecting the proper R_high or R_low value, respectively. The following equations can be used to calculate R_high and R_low values:

$$R_{high} = \frac{1}{\frac{(Vout - 1)}{R1} - \left(\frac{1}{R2}\right)} \tag{1}$$

$$R_{low} = \frac{1}{\frac{1}{R2(Vout - 1)} - \left(\frac{1}{R1}\right)}$$
 (2)

If, for example, a 4.0V output is needed, the user should choose the regulator with a trim range covering 4.0V from Table 3. For this example, the PI3301 is selected (3.3V set voltage). First step would be to use Equation (1) to calculate R_high since the required output voltage is higher than the regulator set voltage. The resistor-divider network values for the PI3301 are can be found in Table 4 and are R1=2.61k Ω and R2=1.13k Ω . Inserting these values in to Equation (1), R_high is calculated as follows:

$$3.78k = \frac{1}{\frac{(4.0-1)}{2.61k} - \left(\frac{1}{1.13k}\right)}$$



Resistor R-high would be connected as in Figure 4 to achieve the 4.0V regulator output. No R_low resistor would be used since in this example the trim is above the regulator set voltage.

Soft-Start Adjust and Tracking

The TRK pin offers a means to increase the regulator's soft-start time or to track with additional regulators. The soft-start slope is controlled by an internal 100nF and a fixed charge current to provide a minimum startup time of 2ms (typical) for all PI33XX regulators. By adding an additional external capacitor to the TRK pin, the soft-start time can be increased further. The following equation can be used to calculate the proper capacitor for a desired soft-start times:

$$C_{TRK} = (t_{TRK} \times I_{TRK}) - 100 \times 10^{-9},$$

Where, t_{TRK} is the soft-start time and I_{TRK} is a 50uA internal charge current (see Electrical Characteristics for limits).

There is typically either proportional or direct tracking implemented within a design. For proportional tracking between several regulators at startup, simply connect all devices TRK pins together. This type of tracking will force all connected regulators to startup and reach regulation at the same time (see Figure 5 (a)).

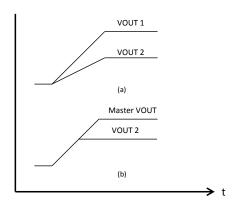


Figure 5 - PI33XX tracking methods

For Direct Tracking, choose the regulator with the highest output voltage as the master and connect the master to the TRK pin of the other regulators through a divider (Figure 6) with the same ratio as the slave's feedback divider (see Table 4 for values).

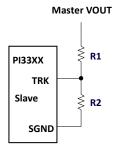


Figure 6 - Voltage divider connections for direct tracking

All connected regulators' soft-start slopes will track with this method. Direct tracking timing is demonstrated in Figure 5 (b). All tracking regulators should have their Enable (EN) pins connected together to work properly.

Inductor Pairing

The PI33XX utilizes an external inductor. This inductor has been optimized for maximum efficiency performance. Table 5 details the specific inductor value and part number utilized for each PI33XX device and are manufactured by either Picor or Cooper Bussman Coiltronics accordingly.

Device	Inductor [nH]	Inductor Part Number	Manufacturer
PI3311-X0	125	PI6125-00-FPIZ	Picor
PI3318-X0	155	PI6155-00-FPIZ	Picor
PI3312-X0	200	FPT705-200-R	Coiltronics
PI3301-X0	200	FPT705-200-R	Coiltronics
PI3302-X0	200	FPT705-200-R	Coiltronics
PI3303-X0	230	FPT705-230-R	Coiltronics
PI3305-X0	230	FPT705-230-R	Coiltronics

Table 5 - PI33XX Inductor pairing

Thermal Derating

Thermal de-rating curves are provided that are based on component temperature changes versus load current, input voltage and air flow. It is recommended to use these curves as a guideline for proper thermal de-rating. These curves represent the entire system and are inclusive to both the Picor regulator and the external inductor. Maximum



thermal operation is limited by either the MOSFETs or inductor depending upon line and load conditions.

Thermal measurements were made using a standard PI33XX Evaluation board which is 3x4 inches in area and uses 4-layer, 2oz copper. Thermal measurements were made on the three main power devices, the two internal MOSFETs and the external inductor, with air flows of 0, 200, and 400 LFM.



Filter Considerations

The PI33XX requires input bulk storage capacitance as well as low impedance ceramic X5R input capacitors to ensure proper start up and high frequency decoupling for the power stage. The PI33XX will draw nearly all of the high frequency current from the low impedance ceramic capacitors when the main high side MOSFET is conducting. During the time the high side MOSFET is off, they are replenished from the bulk capacitor. If the input impedance is high at the switching frequency of the converter, the bulk capacitor must supply all of the average current into the converter, including replenishing the ceramic capacitors. This value has been chosen to be $100\mu F$ so that the PI33XX can

start up into a full resistive load and supply the output capacitive load with the default minimum soft start capacitor when the input source impedance is 50 Ohms at 1MHz. The ESR for this capacitor should be approximately $20m\Omega$. The RMS ripple current in this capacitor is small, so it should not be a concern if the input recommended ceramic capacitors are used. Table 6 shows the recommended input and output capacitors to be used for the various models as well as expected transient response, RMS ripple currents per capacitor, and input and output ripple voltages. Table 7 includes the recommended input and output ceramic capacitors.

Device	V _{IN} (V)	I _{LOAD} (A)	C _{INPUT} Bulk Elec.	C _{INPUT} Ceramic X5R	С _{оитрит} Ceramic X5R	C _{INPUT} Ripple Current (I _{RMS})	C _{OUTPUT} Ripple Current (I _{RMS)}	Input Ripple (mVpp)	Output Ripple (mVpp)	Transient Deviation (mVpk)	Recovery Time (μs)	Load Step (A) (Slew/µs)
PI3311	24	10	100μF 50V	4X4.7μF	8X100μF 2X1 μF	0.5	0.8	120	20	-/+40	40	5
	5		50V	1X0.1 μF			100	15			(5A/μs)	
PI3318	24	10	100μF 50V	4X4.7μF	6X100μF 2X1 μF	0.5 0.8	0.8	120	20	-/+40	40	5
		5		50V	1X0.1 uF			100	15			(5A/μs)
PI3312	24	10	100μF 50V	4X4.7μF	4X100μF 2X1 μF	1	1.75	150	50	-/+80	25	5
		5	301		1X0.1 μF			100	24	•		(10A/µs)
PI3301	24	10	100μF	4X4.7μF	4X100μF 2X1 μF	1.05	1.625	200	40	-/+100	20	5
		5	50V		1X0.1 μF			125	33			(10A/μs)
PI3302	24	10	100μF 50V 4X4.7μ	4X4.7μF	4X47μF 2X1 μF	1.2 1.5	1.5	220	50	-/+170	30	5 (5A/μs)
		5			1X0.1 μF			140	30			
PI3303	24	8	100μF 50V	· 4Χ4./μΓ	4X22μF 2X1 μF	1.3	1.36	275	100	-/+300	30	4
		4	331		1X0.1 μF			150	60	,		(10A/µs)
PI3305	24	8	100μF 50V	4X4.7μF	4X22μF 2X1 μF	1.38	1.2	280	150	-/+400	30	4
		4			1X0.1 μF			160	75			(10A/µs)

Table 6 - Recommended input and output capacitance

MURATA PART NUMBER	DESCRIPTION	MURATA PART NUMBER	DESCRIPTION
GRM188R71C105KA12D	1uF 16V 0603 X7R	GRM31CR71H475KA12K	4.7uF 50V 1206 X7R
GRM319R71H104KA01D	0.1uF 50V 1206 X7R	GRM31CR61A476ME15L	47uF 10V 1206 X5R
GRM31CR60J107ME39L	100uF 6.3V 1206 X5R	GRM31CR61E226KE15L	22uF 25V 1206 X5R

Table 7 - Capacitor manufacturer part numbers



Layout Guidelines

To optimize maximum efficiency and low noise performance from a PI33XX design, layout considerations are necessary. Reducing trace resistance and minimizing high current loop returns along with proper component placement will contribute to optimized performance.

A typical buck converter circuit is shown in Figure 9. The potential areas of high parasitic inductance and resistance are the circuit return paths, shown as LR below.

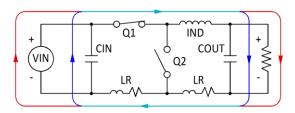


Figure 9 - Typical Buck Converter

The path between the COUT and CIN capacitors is of particular importance since the AC currents are flowing through both of them when Q1 is turned on. Figure 10, schematically, shows the reduced trace length between input and output capacitors. The shorter path lessens the effects that copper trace parasitics can have on the PI33XX performance.

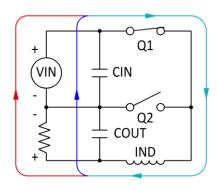


Figure 10 - Current flow: Q1 closed

When Q1 is on and Q2 is off, the majority of CIN's current is used to satisfy the output load and to recharge the COUT capacitors. When Q1 is off and Q2 is on, the load current is supplied by the inductor and the COUT capacitor as shown in Figure 11. During this period CIN is also being recharged by the VIN. Minimizing CIN loop inductance is important to reduce peak voltage excursions when Q1 turns off. Also, the difference in area between the CIN loop and COUT loop is vital to minimize switching and GND noise.

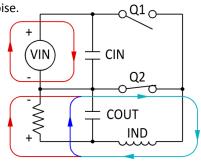


Figure 11 - Current flow: Q2 closed

The recommended component placement, shown in Figure 12, illustrates the tight path between CIN and COUT (and VIN and VOUT) for the high AC return current. This optimized layout is used on the PI33XX evaluation board.

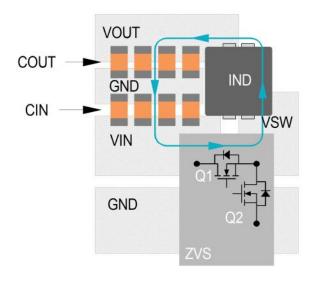


Figure 12 - Recommended component placement and metal routing



Recommended PCB Footprint and Stencil

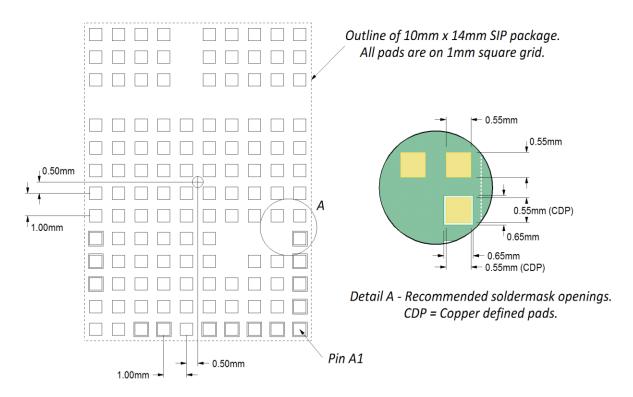
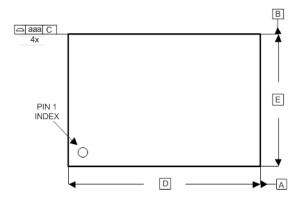


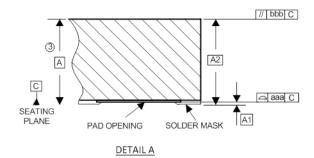
Figure 13 - Recommended Receiving PCB footprint.

Figure 13 details the recommended receiving footprint for PI33XX 10mm x 14mm package. All pads should have a final copper size of 0.55mm x 0.55mm, whether they are solder-mask defined or copper defined, on a 1mm x 1mm grid. All stencil openings are 0.55mm when using a 6mil stencil.

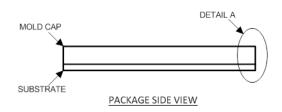


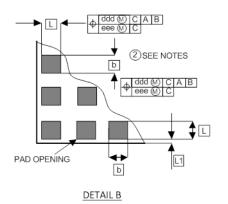
Package Drawings





PACKAGE TOP VIEW





NOM

2.56

0.55

MAX

2.62

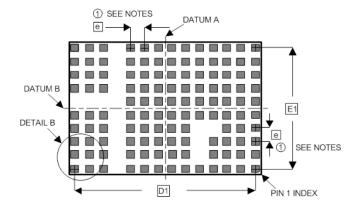
0.05

2.57

0.60

0.08

0.10 0.08



PACKAGE BOTTOM VIEW

L	0.50	0.55	0,60		
D		14.00 BSC			
E		10.00 BSC			
D1		13.00 BSC			
E1		9.00 BSC			
е	1.00BSC				
L1	0.10	0.15	0.20		
aaa			0.10		
bbb			0.10		

MIN

2.50

0.50

SYMBOL

A1

Α2

b

ССС

ddd

eee

NOTES

- $\mbox{\rm `e'}$ REPRESENTS THE BASIC TERMINAL PITCH, SPECIFIES THE TRUE GEOMETRIC POSITION OF THE TERMINAL AXIS.
- 2 DIMENSION 'b' APPLIES TO METALLIZED PAD OPENING.
- ③ DIMENSION 'A' INCLUDES PACKAGE WARPAGE.
- EXPOSED METALLIZED PADS ARE CU PADS WITH SURFACE FINISH PROTECTION.
- 5 ALL DIMENSIONS IN MILLIMETERS.



Warranty

Information furnished by Vicor is believed to be accurate and reliable. However, no responsibility is assumed by Vicor for its use. Vicor makes no representations or warranties with respect to the accuracy or completeness of the contents of this publication. Vicor reserves the right to make changes to any products, specifications, and product descriptions at any time without notice. Information published by Vicor has been checked and is believed to be accurate at the time it was printed; however, Vicor assumes no responsibility for inaccuracies. Testing and other quality controls are used to the extent Vicor deems necessary to support Vicor's product warranty. Except where mandated by government requirements, testing of all parameters of each product is not necessarily

Specifications are subject to change without notice.

Vicor's Standard Terms and Conditions

All sales are subject to Vicor's Standard Terms and Conditions of Sale, which are available on Vicor's webpage or upon request.

Product Warranty

In Vicor's standard terms and conditions of sale, Vicor warrants that its products are free from non-conformity to its Standard Specifications (the "Express Limited Warranty"). This warranty is extended only to the original Buyer for the period expiring two (2) years after the date of shipment and is not transferable.

UNLESS OTHERWISE EXPRESSLY STATED IN A WRITTEN SALES AGREEMENT SIGNED BY A DULY AUTHORIZED VICOR SIGNATORY, VICOR DISCLAIMS ALL REPRESENTATIONS, LIABILITIES, AND WARRANTIES OF ANY KIND (WHETHER ARISING BY IMPLICATION OR BY OPERATION OF LAW) WITH RESPECT TO THE PRODUCTS, INCLUDING, WITHOUT LIMITATION, ANY WARRANTIES OR REPRESENTATIONS AS TO MERCHANTABILITY, FITNESS FOR PARTICULAR PURPOSE, INFRINGEMENT OF ANY PATENT, COPYRIGHT, OR OTHER INTELLECTUAL PROPERTY RIGHT, OR ANY OTHER MATTER.

This warranty does not extend to products subjected to misuse, accident, or improper application, maintenance, or storage. Vicor shall not be liable for collateral or consequential damage. Vicor disclaims any and all liability arising out of the application or use of any product or circuit and assumes no liability for applications assistance or buyer product design. Buyers are responsible for their products and applications using Vicor products and components. Prior to using or distributing any products that include Vicor components, buyers should provide adequate design, testing and operating safeguards.

Vicor will repair or replace defective products in accordance with its own best judgment. For service under this warranty, the buyer must contact Vicor to obtain a Return Material Authorization (RMA) number and shipping instructions. Products returned without prior authorization will be returned to the buyer. The buyer will pay all charges incurred in returning the product to the factory. Vicor will pay all reshipment charges if the product was defective within the terms of this warranty.

Life Support Policy

VICOR'S PRODUCTS ARE NOT AUTHORIZED FOR USE AS CRITICAL COMPONENTS IN LIFE SUPPORT DEVICES OR SYSTEMS WITHOUT THE EXPRESS PRIOR WRITTEN APPROVAL OF THE CHIEF EXECUTIVE OFFICER AND GENERAL COUNSEL OF VICOR CORPORATION. As used herein, life support devices or systems are devices which (a) are intended for surgical implant into the body, or (b) support or sustain life and whose failure to perform when properly used in accordance with instructions for use provided in the labeling can be reasonably expected to result in a significant injury to the user. A critical component is any component in a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system or to affect its safety or effectiveness. Per Vicor Terms and Conditions of Sale, the user of Vicor products and components in life support applications assumes all risks of such use and indemnifies Vicor against all liability and damages.

Intellectual Property Notice

Vicor and its subsidiaries own Intellectual Property (including issued U.S. and Foreign Patents and pending patent applications) relating to the products described in this data sheet. No license, whether express, implied, or arising by estoppel or otherwise, to any intellectual property rights is granted by this document. Interested parties should contact Vicor's Intellectual Property

The products described on this data sheet are protected by the following U.S. Patents Numbers: RE 40,072.

Vicor Corporation 25 Frontage Road Andover, MA, USA 01810 USA

Picor Corporation 51 Industrial Drive North Smithfield, RI 02896 USA

Customer Service: custserv@vicorpower.com Technical Support: apps@vicorpower.com





Компания «Океан Электроники» предлагает заключение долгосрочных отношений при поставках импортных электронных компонентов на взаимовыгодных условиях!

Наши преимущества:

- Поставка оригинальных импортных электронных компонентов напрямую с производств Америки, Европы и Азии, а так же с крупнейших складов мира;
- Широкая линейка поставок активных и пассивных импортных электронных компонентов (более 30 млн. наименований);
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
- Экспресс доставка в любую точку России;
- Помощь Конструкторского Отдела и консультации квалифицированных инженеров;
- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
- Поставка электронных компонентов под контролем ВП;
- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001;
- При необходимости вся продукция военного и аэрокосмического назначения проходит испытания и сертификацию в лаборатории (по согласованию с заказчиком);
- Поставка специализированных компонентов военного и аэрокосмического уровня качества (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Actel, Aeroflex, Peregrine, VPT, Syfer, Eurofarad, Texas Instruments, MS Kennedy, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

Компания «Океан Электроники» является официальным дистрибьютором и эксклюзивным представителем в России одного из крупнейших производителей разъемов военного и аэрокосмического назначения «JONHON», а так же официальным дистрибьютором и эксклюзивным представителем в России производителя высокотехнологичных и надежных решений для передачи СВЧ сигналов «FORSTAR».



«JONHON» (основан в 1970 г.)

Разъемы специального, военного и аэрокосмического назначения:

(Применяются в военной, авиационной, аэрокосмической, морской, железнодорожной, горно- и нефтедобывающей отраслях промышленности)

«**FORSTAR**» (основан в 1998 г.)

ВЧ соединители, коаксиальные кабели, кабельные сборки и микроволновые компоненты:

(Применяются в телекоммуникациях гражданского и специального назначения, в средствах связи, РЛС, а так же военной, авиационной и аэрокосмической отраслях промышленности).



Телефон: 8 (812) 309-75-97 (многоканальный)

Факс: 8 (812) 320-03-32

Электронная почта: ocean@oceanchips.ru

Web: http://oceanchips.ru/

Адрес: 198099, г. Санкт-Петербург, ул. Калинина, д. 2, корп. 4, лит. А