

# AS5601 12-bit Programmable Contactless Encoder

## **General Description**

The AS5601 is an easy-to-program magnetic rotary position sensor with incremental quadrature (A/B) and 12-bit digital outputs. Additionally, the PUSH output indicates fast airgap changes between the AS5601 and magnet which can be used to implement a contactless pushbutton function in which the knob can be pressed to move the magnet toward the AS5601.

This AS5601 is designed for contactless encoder applications, and its robust design rejects the influence of any homogenous external stray magnetic fields.

Based on planar Hall sensor technology, this device measures the orthogonal component of the flux density (Bz) from an external magnet.

The industry-standard I<sup>2</sup>C interface supports user programming of non-volatile parameters in the AS5601 without requiring a dedicated programmer.

The AS5601 also provides a smart low-power mode which automatically reduces power consumption

Ordering Information and Content Guide appear at end of datasheet.

### **Key Benefits & Features**

The benefits and features of AS5601, 12-bit Programmable Contactless Encoder are listed below:

Figure 1: Added Value of using AS5601

Benefits	Features
Highest reliability and durability	Contactless angle measurement insensitive to dust and dirt
Simple programming	Simple user-programmable zero position and device configuration
Flexible choice of the number of A/B pulses per revolution	Quadrature output configurable from 8 up to 2048 positions
Contactless pushbutton functionality	Pushbutton output by detecting sudden airgap changes
Low power consumption	Automatic entry into low-power mode
Easy setup	Automatic magnet detection
Small form factor	SOIC-8 package
Robust environmental tolerance	Wide temperature range: -40°C to +125°C



## Applications

The AS5601 is ideally suited for:

- Encoder replacement
- Contactless rotary knobs with push buttons
- Other angular position measurement solutions

## **Block Diagram**

The functional blocks of this device for reference are shown below:

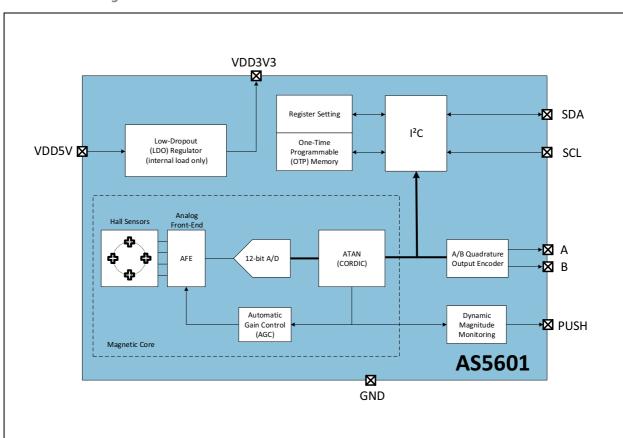
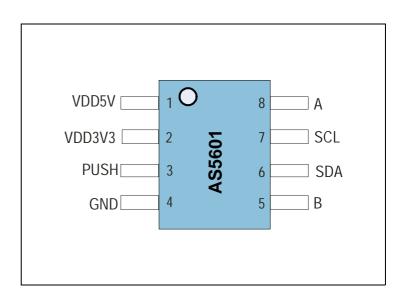


Figure 2: AS5601 Block Diagram



# **Pin Assignments**

Figure 3: SOIC-8 Pin-out



# **Pin Description**

Figure 4: Pin Description

Pin Number	Name	Туре	Description
1	VDD5V	Supply	Positive voltage supply in 5V mode
2	VDD3V3	Supply	Positive voltage supply in 3.3V mode (requires an external 1-µF decoupling capacitor in 5V mode)
3	PUSH	Digital output	Contactless pushbutton function output
4	GND	Supply	Ground
5	В	Digital output	Quadrature incremental signal B
6	SDA	Digital input/output	I <sup>2</sup> C Data
7	SCL	Digital input	l <sup>2</sup> C Clock
8	А	Digital output	Quadrature incremental signal A

## Absolute Maximum Ratings

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated under "Operating Conditions" is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### Figure 5: Absolute Maximum Ratings

Symbol	Parameter	Min	Max	Units	Comments					
	Electrical Parameters									
VDD5V	DC supply voltage at VDD5V pin	-0.3	6.1	V	5.0V operation mode					
VDD3V3	DC supply voltage at VDD3V3 pin	-0.3	4.0	V	3.3V operation mode					
VAIO	Voltage at all digital or analog pins	-0.3	VDD + 0.3	V						
I <sub>SCR</sub>	Input current (latch-up immunity)	-100	100	mA	Norm: JESD78					
	Continuo	us Powe	r Dissipation	(T <sub>A</sub> = +70°C	]					
P <sub>T</sub>	Continuous power dissipation		50	mW						
		Electros	static Dischar	ge						
ESD <sub>HBM</sub>	Electrostatic discharge HBM (human body model)		±1	kV	Norm: MIL 883 E method 3015.7					
	Temperatu	re Rang	es and Storag	ge Conditio	ns					
T <sub>STRG</sub>	Storage temperature range	-55	+125	°C						
T <sub>BODY</sub>	Package body temperature		+260	°C	Norm: ICP/JEDEC J-STD-020 The reflow peak soldering temperature (body temperature) is specified according to IPC/JEDEC J-STD-020 "Moisture/Reflow Sensitivity Classification for Non-hermetic Solid State Surface Mount Devices." The lead finish for Pb-free leaded packages is "Matte Tin" (100% Sn)					
RH <sub>NC</sub>	Relative humidity (non-condensing)	5	85	%						
MSL	Moisture sensitive level		3		Norm: ICP/JEDEC J-STD-033					



## **Electrical Characteristics**

All limits are guaranteed. The parameters with min and max values are guaranteed with production tests or SQC (Statistical Quality Control) methods.

## **Operating Conditions**

Figure 6:

System Electrical Characteristics and Temperature Range

Symbol	Parameter	Conditions	Min	Тур	Max	Units
VDD5V	Positive supply voltage in	5.0V operation mode	4.5	5.0	5.5	V
VUUSV	5.0V mode	During OTP burn procedure <sup>(2)</sup>	4.5	5.0	5.5	v
VDD3V3	Positive supply voltage in	3.3V operation mode	3.0	3.3	3.6	V
VDD3V3	3.3V mode	During OTP burn procedure <sup>(2)</sup>	3.3	3.4	3.5	V
IDD	Supply current in NOM <sup>(1)</sup>	PM = 00 Always on			6.5	mA
IDD_LPM1	Supply current in LPM1 <sup>(1)</sup>	PM = 01 Polling time = 5 ms			3.4	mA
IDD_LPM2	Supply current in LPM2 <sup>(1)</sup>	PM = 10 Polling time = 20 ms			1.8	mA
IDD_LPM3	Supply current in LPM3 <sup>(1)</sup>	PM = 11 Polling time = 100 ms			1.5	mA
IDD_BURN	Supply current per bit for	Initial peak, 1 μs			100	mA
	burn procedure	Steady burning, <30 μs			40	mA
T <sub>A</sub>	Operating temperature		-40		+125	°C
T <sub>P</sub>	Programming temperature		20		30	°C

#### Note(s) and/or Footnote(s):

1. For typical magnetic field (60 mT) excluding current delivered to the external load and tolerance on polling times.

2. For OTP burn procedure the supply line source resistance should not exceed 10hm.



## **Digital Inputs and Outputs**

#### Figure 7: Digital Inputs and Outputs

Symbol	Parameter	Conditions	Min	Тур	Мах	Units
V_IH	High-level input voltage		0.7  imes VDD			V
V_IL	Low-level input voltage				0.3 × VDD	V
V_OH	High-level output voltage		VDD - 0.5			V
V_OL	Low-level output voltage				0.4	V
I_0	Output current for A, B, and PUSH		-2		2	mA
C_L	Capacitive load for A, B, and PUSH				50	pF
I_LKG	Leakage current				±1	μΑ

## **Timing Characteristics**

Figure 8: Timing Conditions

Symbol	Parameter	Conditions	Min	Тур	Мах	Units
T_DETWD	Watchdog detection time	WD = 1	57	60	63	seconds
T_PU	Power-up time				10	ms
F_S	Sampling rate				150	μs
T_SETTL1	Settling time	SF = 00			2.2	ms
T_SETTL2	Settling time	SF = 01			1.1	ms
T_SETTL3	Settling time	SF = 10			0.55	ms
T_SETTL4	Settling time	SF = 11			0.286	ms

# **Magnetic Characteristics**

#### Figure 9: Magnetic Characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Units
Bz	Orthogonal magnetic field strength, regular output noise ON_SLOW and ON_FAST	Required orthogonal component of the magnetic field strength measured at the die's surface along a circle of 1 mm	30	60	90	mT
Bz_ERROR	Minimum required orthogonal magnetic field strength, magnet detection level				8	mT

## System Characteristics

Figure 10: System Characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Units
RES	Core Resolution			12		bit
RES_AB	A/B output resolution		8		2048	positions
VMAX_AB	Maximum rotation speed for incremental output	Continuous Rotation $\geq$ 360deg <sup>(1), (2)</sup>			456	rpm
INL_BL	System INL	Deviation from best line fit; 360° maximum angle, no magnet displacement, no zero-programming performed			±1	degree
ON_SLOW	RMS output noise (1 sigma)	Orthogonal component for the magnetic field within the specified range Bz, after 2.2 ms; SF = 00			0.015	degree
ON_FAST	RMS output noise (1 sigma)	Orthogonal component for the magnetic field within the specified range Bz, after 286 µs; SF = 11			0.043	degree

#### Note(s) and/or Footnote(s):

1. An infinite fast change <180deg results in angle output with maximum configured update frequency.

2. An infinite fast change >= 180deg results in angle output to the shortest next absolute position with maximum configured update frequency. e.g. A change from 0 to 270deg will be indicated as angle output from 0 to -90deg.

## **Detailed Description**

The AS5601 is a Hall-based rotary magnetic position encoder that converts the magnetic field component perpendicular to the surface of the chip into voltages which are used to produce incremental A/B outputs and absolute position indication in registers that can be read over an industry-standard l<sup>2</sup>C bus.

The analog signals from the Hall sensors are first amplified and filtered before being converted by the analog-to-digital converter (ADC) into binary data. The output of the ADC is processed by the hardwired CORDIC block (Coordinate Rotation Digital Computer) to compute the angle and magnitude of the magnetic field vector. The intensity of the magnetic field is used by the automatic gain control (AGC) to adjust the amplification level to compensate for temperature and magnetic field variations.

The angle value provided by the CORDIC algorithm is used by the internal logic to generate the incremental quadrature signals A and B. The magnitude and AGC value is dynamically monitored and generates the PUSH output for fast changes of the airgap between the magnet and the AS5601. Very slow changes are suppressed to provide a robust and reliable pushbutton output that tolerates temperature variation and magnet degradation.

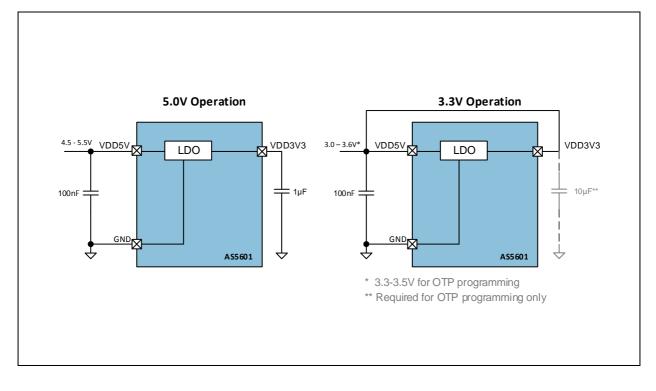
The AS5601 is programmed through an industry-standard I<sup>2</sup>C interface to write an on-chip one-time programmable (OTP) memory. This interface can be used to program a zero angle and to configure the chip.

### **Power Management**

The AS5601 is powered from a 5.0V supply using the on-chip LDO regulator, or it can be powered directly from a 3.3V supply. The internal LDO is not intended to power other external ICs and needs a  $1\mu$ F capacitor to ground, as shown in Figure 11.

In 3.3V operation, the VDD5V and VDD3V3 pins must be tied together.

#### Figure 11: 5.0V and 3.3V Power Supply Options



## I<sup>2</sup>C Interface

The AS5601 supports the 2-wire Fast-mode Plus l<sup>2</sup>C-slave protocol in device mode, in compliance with the NXP Semiconductors (formerly Philips Semiconductors) specification UM10204. A device that sends data onto the bus is a transmitter and a device receiving data is a receiver. The device that controls the message is called a master. The devices that are controlled by the master are called slaves. A master device generates the serial clock (SCL), controls the bus access, and generates the START and STOP conditions that control the bus. The AS5601 always operates as a slave on the l<sup>2</sup>C bus. Connections to the bus are made through the open-drain I/O lines SDA and the input SCL. Clock stretching is not included.

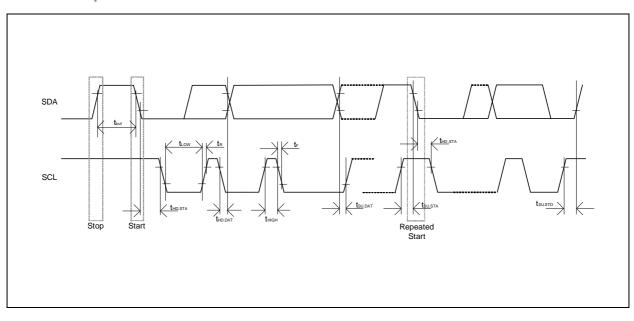
The host MCU (master) initiates data transfers. The 7-bit slave address of the AS5601 is 0x36 (0110110 in binary).

- Supported Modes
- Random/Sequential read
- Byte/Page write
- Automatic increment (ANGLE register)
- Standard-mode
- Fast-mode
- Fast-mode Plus

The SDA signal is the bidirectional data line. The SCL signal is the clock generated by the  $l^2C$  bus master to synchronize sampling data from SDA. The maximum SCL frequency is 1 MHz. Data is sampled on the rising edge of SCL.

## I<sup>2</sup>C Interface Operation





## I<sup>2</sup>C Electrical Specification

#### Figure 13: I<sup>2</sup>C Electrical Specifications

Symbol	Parameter	Conditions	Min	Max	Units
VIL	Logic low input voltage		-0.3	0.3 x VDD	V
VIH	Logic high input voltage		0.7 x VDD	VDD + 0.3	V
VHYS	Hysteresis of Schmitt trigger inputs	VDD > 2.5V	0.05 x VDD		V
VOL	Logic low output voltage (open-drain or open-collector) at 3 mA sink current	VDD > 2.5V		0.4	V
IOL	Logic low output current	VOL = 0.4V	20		mA
t <sub>OF</sub>	Output fall time from VIHmax to VILmax		10	120 <sup>(1)</sup>	ns
t <sub>SP</sub>	Pulse width of spikes that must be suppressed by the input filter			50 <sup>(2)</sup>	ns
I <sub>I</sub>	Input current at each I/O Pin	Input voltage between 0.1 x VDD and 0.9 x VDD	-10	+10 <sup>(3)</sup>	μΑ
C <sub>B</sub>	Total capacitive load for each bus line			550	pF
C <sub>I/O</sub>	l/O capacitance (SDA, SCL) <sup>(4)</sup>			10	pF

#### Note(s) and/or Footnote(s):

1. In Fast-mode Plus, fall time is specified the same for both output stage and bus timing. If series resistors are used this has to be considered for bus timing.

2. Input filters on the SDA and SCL inputs suppress noise spikes of less than 50 ns.

3. I/O pins of Fast-mode and Fast-mode Plus devices must not load or drive the SDA and SCL lines if VDD is switched OFF.

4. Special-purpose devices such as multiplexers and switches may exceed this capacitance because they connect multiple paths together.



## I<sup>2</sup>C Timing

#### Figure 14: I<sup>2</sup>C Timing

Symbol	Parameter	Conditions	Min	Мах	Units
f <sub>SCLK</sub>	SCL clock frequency			1.0	MHz
t <sub>BUF</sub>	Bus free time (time between the STOP and START conditions)		0.5		μs
t <sub>HD;STA</sub>	Hold time; (Repeated) START condition <sup>(1)</sup>		0.26		μs
t <sub>LOW</sub>	Low phase of SCL clock		0.5		μs
t <sub>HIGH</sub>	High phase of SCL clock		0.26		μs
t <sub>su;sta</sub>	Setup time for a Repeated START condition		0.26		μs
t <sub>HD;DAT</sub>	Data hold time <sup>(2)</sup>			0.45	μs
t <sub>SU;DAT</sub>	Data setup time <sup>(3)</sup>		50		ns
t <sub>R</sub>	Rise time of SDA and SCL signals			120	ns
t <sub>F</sub>	Fall time of SDA and SCL signals		10	120 <sup>(4)</sup>	ns
t <sub>SU;STO</sub>	Setup time for STOP condition		0.26		μs

#### Note(s) and/or Footnote(s):

- 1. After this time, the first clock is generated.
- 2. A device must internally provide a minimum hold time of 120 ns (Fast-mode Plus) for the SDA signal (referred to the VIH<sub>min</sub> of SCL) to bridge the undefined region of the falling edge of SCL.
- 3. A Fast-mode device can be used in a standard-mode system, but the requirement  $t_{SU;DAT} = 250$  ns must be met. This is automatic if the device does not stretch the low phase of SCL. If such a device does stretch the low phase of SCL, it must drive the next data bit on SDA ( $t_{Rmax} + t_{SU;DAT} = 1000 + 250 = 1250$  ns ) before SCL is released.
- 4. In Fast-mode Plus, fall time is specified the same for both output stage and bus timing. If series resistors are used, this has to be considered for bus timing.

## I<sup>2</sup>C Modes

#### Invalid Addresses

There are two addresses used to access an AS5601 register. The first is the slave address used to select the AS5601. All I<sup>2</sup>C bus transactions include a slave address. The slave address of the AS5601 is 0x36 (0110110 in binary). The second address is a word address sent in the first byte transferred in a write transaction. The word address selects a register on the AS5601. The word address is loaded into the address pointer on the AS5601. During subsequent read transactions and subsequent bytes in the write transaction, the address pointer provides the address of the selected register. The address pointer is incremented after each byte is transferred, except for certain read transactions to special registers.

If the user sets the address pointer to an invalid word address, the address byte is not acknowledged (the A bit is high). Nevertheless, a read or write cycle is possible. The address pointer is increased after each byte.

#### Reading

When reading from an invalid address, the AS5601 returns all zeros in the data bytes. The address pointer is incremented after each byte. Sequential reads over the whole address range are possible including address overflow.

Automatic increment of the address pointer for ANGLE, RAW ANGLE, and MAGNITUDE registers:

These are special registers which suppress the automatic increment of the address pointer on reads, so a re-read of these registers requires no I<sup>2</sup>C write command to reload the address pointer. This special treatment of the pointer is effective only if the address pointer is set to the higher byte of the register, which holds the least significant bits.

#### Writing

A write to an invalid address is not acknowledged by the AS5601, although the address pointer is incremented. When the address pointer points to a valid address again, a successful write accessed is acknowledged. Page write over the whole address range is possible including address overflow.

#### Supported bus protocol

Data transfer may be initiated only when the bus is not busy. During data transfer, the data line must remain stable whenever SCL is high. Changes in the data line while SCL is high are interpreted as START or STOP conditions.



Accordingly, the following bus conditions have been defined:

#### **Bus Not Busy**

Both SDA and SCL remain high.

#### Start Data Transfer

A change in the state of SDA from high to low while SCL is high defines the START condition.

#### Stop Data Transfer

A change in the state of SDA from low to high while SCL is high defines the STOP condition.

#### Data Valid

The state of the data line represents valid data when, after a START condition, SDA is stable for the duration of the high phase of SCL. The data on SDA must only be changed during the low phase of SCL. There is one clock period per bit of data.

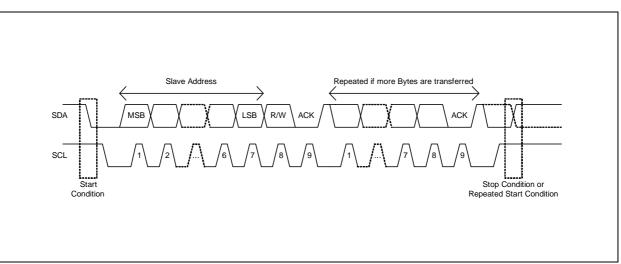
Each I<sup>2</sup>C bus transaction is initiated with a START condition and terminated with a STOP condition. The number of data bytes transferred between START and STOP conditions is not limited, and is determined by the I<sup>2</sup>C bus master. The information is transferred byte-wise and each receiver acknowledges with a ninth bit.

#### Acknowledge

Each  $I^2C$  slave device, when addressed, is obliged to generate an acknowledge after the reception of each byte. The  $I^2C$  bus master device must generate an extra clock period for this acknowledge bit.

A slave that acknowledges must pull down SDA during the acknowledge clock period in such a way that SDA is stable low during the high phase of the acknowledge clock period. Of course, setup and hold times must be taken into account. A master must signal an end of a read transaction by not generating an acknowledge bit on the last byte that has been clocked out of the slave. In this case, the slave must leave SDA high to enable the master to generate the STOP condition.





Depending on the state of the R/W bit, two types of data transfer are possible:

Data transfer from a master transmitter to a slave receiver

The first byte transmitted by the master is the slave address, followed by R/W = 0. Next follows a number of data bytes. The slave returns an acknowledge bit after each received byte. If the slave does not understand the command or data it sends a not acknowledge (NACK). Data is transferred with the most significant bit (MSB) first.

Data transfer from a slave transmitter to a master receiver

The master transmits the first byte (the slave address). The slave then returns an acknowledge bit, followed by the slave transmitting a number of data bytes. The master returns an acknowledge bit after all received bytes other than the last byte. At the end of the last received byte, a NACK is returned. The master generates all of the SCL clock periods and the START and STOP conditions. A transfer is ended with a STOP condition or with a repeated START condition. Because a repeated START condition is also the beginning of the next serial transfer, the bus is not released. Data is transferred with the most significant bit (MSB) first.



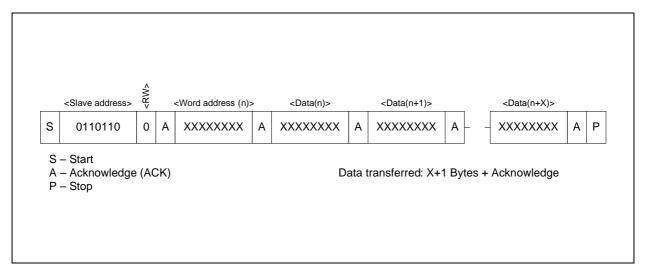
## AS5601 Slave Modes

### Slave Receiver Mode (Write Mode)

Serial data and clock are received through SDA and SCL. Each byte is followed by an acknowledge bit or by a NACK depending on whether the address pointer selects a valid address. START and STOP conditions are recognized as the beginning and end of a bus transaction. The slave address byte is in the first byte received after the START condition. The 7-bit AS5601 address is 0x36 (0110110 in binary).

The 7-bit slave address is followed by the direction bit (R/W), which, for a write, is 0 (low). After receiving and decoding the slave address byte, the slave device drives an acknowledge on SDA. After the AS5601 acknowledges the slave address and write bit, the master transmits a register address (word address) to the AS5601. This is loaded into the address pointer on the AS5601. If the address is a valid readable address, the AS5601 answers by sending an acknowledge (A bit low). If the address pointer selects an invalid address, a NACK is sent (A bit high). The master may then transmit zero or more bytes of data. If the address pointer selects an invalid address, the received data are not stored. The address pointer will increment after each byte transferred whether or not the address is valid. If the address pointer reaches a valid position again, the AS5601 answers with an acknowledge and stores the data. The master generates a STOP condition to terminate the write transaction.

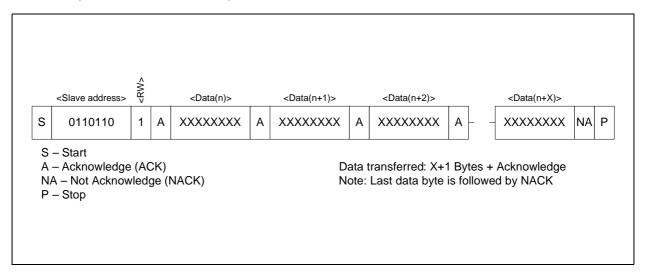
#### Figure 16: Data Write (Slave Receiver Mode)



#### Slave Transmitter Mode (Read Mode)

The first byte is received and handled as in the slave receiver mode. However, in this mode, the direction bit indicates that the AS5601 will drive data on SDA. START and STOP conditions are recognized as the beginning and end of a bus transaction. The slave address byte is the first byte received after the master generates a START condition. The slave address byte contains the 7-bit AS5601 address. The 7-bit slave address is followed by the direction bit (R/W), which, for a read, is 1 (high). After receiving and decoding the slave address byte, the slave device drives an acknowledge on the SDA line. The AS5601 then begins to transmit data starting with the register address pointed to by the address pointer. If the address pointer is not written before the initiation of a read transaction, the first address that is read is the last one stored in the address pointer. The AS5601 must receive a not acknowledge (NACK) to end a read transaction.

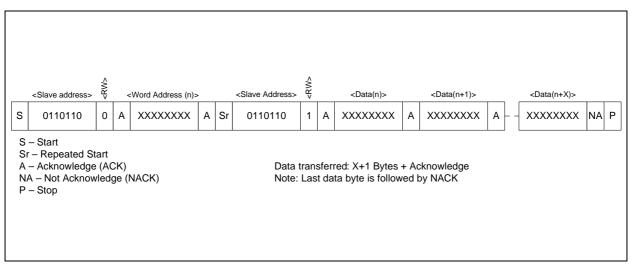
Figure 17: Data Read (Slave Transmitter Mode)





#### Figure 18:

Data Read with Address Pointer Reload (Slave Transmitter Mode)



#### SDA and SCL Input Filters

Input filters for SDA and SCL inputs are included to suppress noise spikes of less than 50 ns.



## **Register Description**

The following registers are accessible over the serial I<sup>2</sup>C interface. The 7-bit device address of the AS5601 is 0x36 (0110110 in binary). To permanently program a configuration, a non-volatile memory (OTP) is provided.

Figure 19: Register Map

Address	Name	R/W	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	Configuration Registers <sup>(1)</sup> , <sup>(2)</sup>									
0x00	ZMCO	R							ZMC	O(1:0)
0x01	ZPOS	R/W/P						ZPOS	(11:8)	
0x02	ZFOJ					ZPO	S(7:0)			
0x07	CONF	R/W/P			WD		FTH(2:0)		SF(	1:0)
0x08	CONF						HYS	Г(1:0)	PM	(1:0)
0x09	ABN	R/W/P						ABN	l(3:0)	
0x0A	PUSHTHR	R/W/P				PUSHT	HR(7:0)			
			Ou	utput Reg	gisters					
0x0C	RAW ANGLE	R						RAW AN	GLE(11:8)	)
0x0D	NAW ANGLE	R				RAW AN	IGLE(7:0)			
0x0E	ANGLE	R						ANGL	E(11:8)	
0x0F	ANGLE	R				ANG	E(7:0)			
			St	atus Reg	jisters					
0x0B	STATUS	R			MD	ML	МН			
0x1A	AGC	R				AGO	[(7:0)			
0x1B	MAGNITUDE	R						MAGNITU	JDE (11:8	)
0x1C		R	MAGNITUDE(7:0)							
Burn Command										
0xFF	BURN	W		Bu	ırn_Angle	e = 0x80;	Burn_Se	tting = 0x	(40	

#### Note(s) and/or Footnote(s):

1. To change a configuration, read out the register, modify only the desired bits and write the new configuration. Blank fields may contain factory settings.

2. During power-up, configuration registers are reset to the permanently programmed value. Not programmed bits are zero.



## **ZPOS Registers**

These registers are used to configure the zero position (ZPOS). This register is used to align the electric grid of the incremental output with the mechanical grid of an encoder switch.

#### **CONF** Register

The CONF register supports customizing the AS5601. Figure 20 shows the mapping of the CONF register.

### **PUSHTHR Register**

This register is used to set-up the contactless pushbutton function. This register must be adjusted according to the airgap and magnet configuration. The swing of the pushbutton function can be found by subtracting the AGC value of the pressed button from the AGC value of the released button. The threshold value for the contactless pushbutton should be half of the swing.

Figure 20: CONF and ABN mapping

Name	Bit Position	Description					
	CONF Mapping						
PM(1:0)	1:0	Power Mode 00 = NOM, 01 = LPM1, 10 = LPM2, 11 = LPM3					
HYST(1:0)	3:2	Hysteresis 00 = OFF, 01 = 1 LSB, 10 = 2 LSBs, 11 = 3 LSBs					
SF(1:0)	9:8	Slow Filter 00 = 16x <sup>(1)</sup> ; 01 = 8x; 10 = 4x; 11 = 2x					
FTH(2:0)	12:10	Fast Filter Threshold 000 = slow filter only, 001 = 6 LSBs, 010 = 7 LSBs, 011 = 9 LSBs, 100 = 18 LSBs, 101 = 21 LSBs, 110 = 24 LSBs, 111 = 10 LSBs					
WD	13	Watchdog Timer 0 = OFF, 1 = ON (automatic entry into LPM3 low-power mode enabled)					
		ABN Mapping					
ABN(3:0)	3:0	Output Positions and Update Rate 0000 : 8 (61 Hz) 0001 : 16 (122 Hz) 0010 : 32 (244 Hz) 0011 : 64 (488 Hz) 0100 : 128 (976 Hz) 0101 : 256 (1.9 kHz) 0110 : 512 (3.9 kHz) 0111 : 1024 (7.8 kHz) others : 2048 (15.6 kHz))					

Note(s) and/or Footnote(s):

1. Forced in Low Power Mode (LPM)

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#### **ANGLE/RAW ANGLE Register**

The RAW ANGLE register contains the unmodified angle. The zero adjusted and filtered output value is available in the ANGLE register.

#### **STATUS Register**

The STATUS register provides bits that indicate the current state of the AS5601.

Figure 21: STATUS Register

Name	State When Bit is High
МН	AGC minimum gain overflow, magnet too strong
ML	AGC maximum gain overflow, magnet too weak
MD	Magnet was detected

#### **AGC Register**

The AS5601 uses automatic gain control (AGC) in a closed loop to compensate for variations of the magnetic field strength due to changes of temperature, airgap between IC and magnet, and magnet degradation. The AGC register indicates the gain. For the most robust performance, the gain value should be in the center of its range. The airgap of the physical system can be adjusted to achieve this value.

#### **MAGNITUDE** Register

The MAGNITUDE register indicates the magnitude value of the internal CORDIC output.

#### Non-Volatile Memory (OTP)

The non-volatile memory is used to permanently program the configuration. To program the non-volatile memory, the  $I^2C$  interface is used. The programming can be either performed in the 5V supply mode or in the 3.3V operation mode but using a minimum supply voltage of 3.3V and a 10 µF capacitor at the VDD3V3 pin to ground. This 10 µF capacitor is needed only during the programming of the device. Two different commands are used to permanently program the device:

## Burn\_Angle Command (ZPOS)

The host microcontroller can perform a permanent programming of ZPOS with a BURN\_ANGLE command. To perform a BURN\_ANGLE command, write the value 0x80 into register 0xFF. The BURN\_ANGLE command can be executed up to 3 times. ZMCO shows how many times ZPOS have been permanently written.

This command may only be executed if the presence of the magnet is detected (MD = 1).

## **Burn\_Setting Command (CONF)**

The host microcontroller can perform a permanent writing of CONFIG with a BURN\_SETTING command. To perform a BURN\_SETTING command, write the value 0x40 into register 0xFF.

The BURN\_SETTING command can be performed only one time.

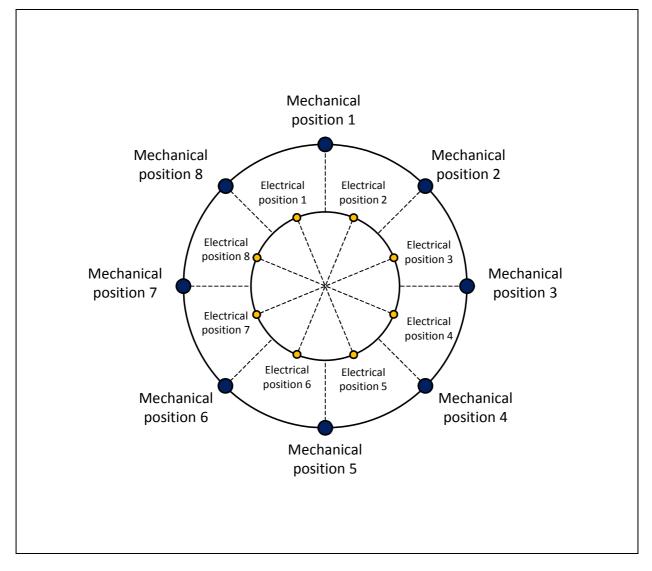
#### Zero Position and Resolution Programming

A fundamental feature is to program the zero position (ZPOS) of the magnetic position encoder. This is required to adjust the A/B outputs to the mechanical pattern (grid) of a contactless encoder by setting the count transitions (transition of A and or B) between two adjacent mechanical positions. An example of a 3-bit contactless encoder is shown in Figure 22.

The electrical positions represent the positions where an A or B transition occurs. The zero position can be placed in correspondence of one of the electrical positions (yellow).

A BURN\_ANGLE command can be executed up to 3 times to permanently program the zero position. It can only be executed if the presence of the magnet is detected (MD = 1).

#### Figure 22: Zero Position Setting of 3-bit Encoder







# The configuration procedure for a rotary encoder is shown below in Figure 23.

Figure 23:

Zero Position and Resolution Programming Procedure

Use the correct hardware configuration as shown in Figure 33				
Step 1	Power up the AS5601.			
Step 2	Configure the desired number of positions using ABN(3:0).			
Step 3	The mechanical configuration snapped into the grid. Read out the actual RAW ANGLE. Calculate the compensation value to adjust the mechanical grid and the encoder angle. Refer to Figure 22 and Figure 24. Write the compensation value into ZPOS. Wait at least 1ms.			
Step 4	Write the required setting into the configuration register CONF and PUSHTHR. Wait at least 1 ms.			
Proceed with Step	Proceed with Step 5 to permanently program the configuration.			
Step 5	Perform a BURN_ANGLE command to permanently program the zero position. Wait at least 1 ms.			
Step 6	Perform a Burn_Setting command to permanently program the configuration. Wait at least 1 ms.			
Step 7	Verify the BURN commands: Write the commands 0x01, 0x11 and 0x10 sequentially into the register 0xFF to load the actual OTP content. Read and verify the permanently programmed registers to verify that the BURN_SETTINGS and BURN_ANGLE command was successful.			
Step 8	Read and verify the permanently programmed registers again after a new power-up cycle.			

#### Note(s) and/or Footnote(s):

1. After each register command, the new setting is effective at the output at least 1 ms later.

2. It is highly recommended to perform a functional test after this procedure.

3. At least 1 ms after each register command the new setting is effective at the output.

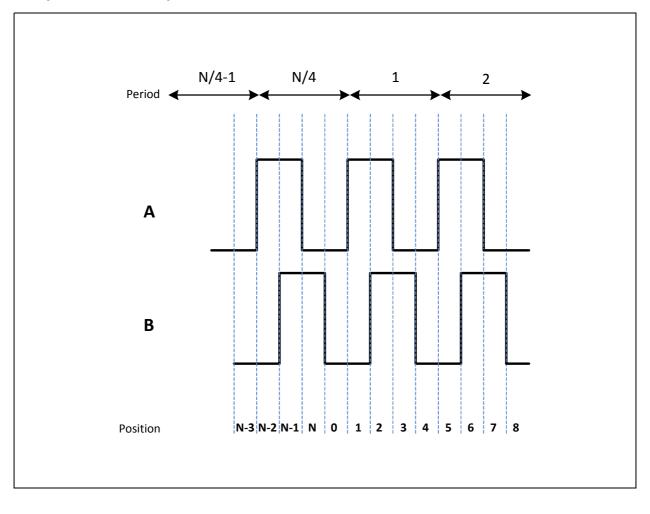
4. The BURN\_ANGLE command can be executed up to 3 times and only if the presence of the magnet is detected (MD = 1).



## **Quadrature Encoder Output**

With the setting ABN(3:0) it is possible to configure the number of positions of the quadrature output. An example for a configuration with 8 positions is shown below.

#### Figure 24: Example Quadrature Output for 8 Positions



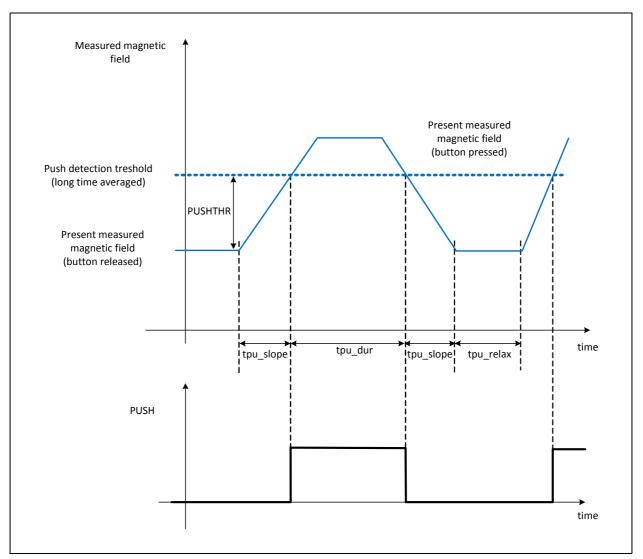
### **Pushbutton Detection**

The AS5601 implements a pushbutton detection function through a dynamic and relative measurement of the orthogonal magnetic field strength. This pushbutton detection function drives the PUSH output pin high when the AS5601 detects a fast increase of the magnetic field (decrease of the airgap between the magnet and the AS5601). After a fast decrease of the magnetic field, the PUSH output is driven low.

#### Figure 25: Pushbutton Detection Function Specifications

Symbol	Parameter	Conditions	Min	Max	Unit
PUSHTHR	Magnetic field threshold		0	255	LSB
t <sub>pu_slope</sub>	Push slope time			500	ms
	Push duration time NOM, LPM1	PM = 0X	10	10000	ms
t <sub>pu_dur</sub>	Push duration time LPM2	PM = 10	40	10000	ms
	Push duration time LPM3	PM = 11	150	10000	ms
t <sub>pu_relax</sub>	Time gap between two consecutive pushes in NOM, LPM1	PM = 0X	40		ms
	Time gap between two consecutive pushes in LPM2	PM = 10	40		ms
	Time gap between two consecutive pushes in LPM3	PM = 11	150		ms
t <sub>pu_min_pulse</sub>	Minimum duration of the PUSH pulse		40	50	ms
t <sub>pu_recovery</sub>	Recovery time after a very long pushbutton event			5000	ms
BTH_VAR	Push amplitude variation	-20		+20	%

#### Figure 26: Pushbutton Detection Function

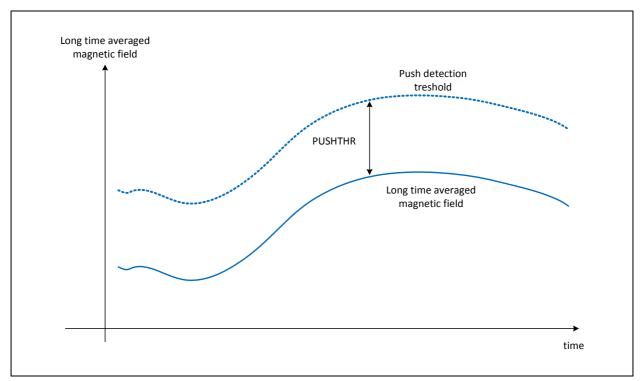


The AS5601 continuously measures the magnetic field intensity. The programmable threshold (PUSHTHR) is applied to a long time average of the magnetic field. A crossing of the current magnetic field and the threshold within a specified time (tpu\_slope) drives the PUSH output high.

Slow changes of the magnetic field, due for example to temperature variations, magnet drift mechanical tolerances, etc. do not generate any pushbutton detection events. The push detection threshold follows the drifts over the time as shown in Figure 27.



#### Figure 27: Magnetic Field Threshold Over Time



## **Step Response and Filter Settings**

The AS5601 has a digital post-processing programmable filter which can be set in fast or slow modes. The fast filter mode can be enabled by setting a fast filter threshold in the FTH bits of the CONF register.

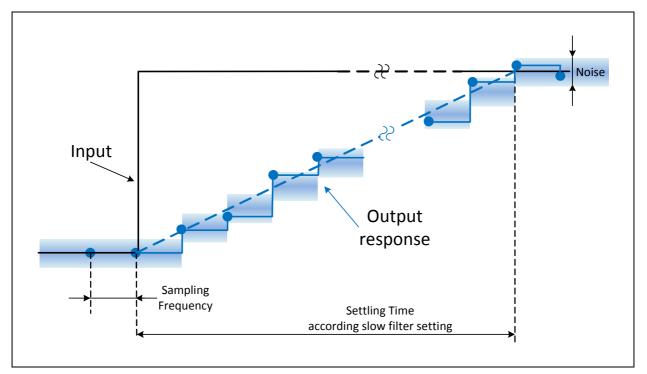
If the fast filter is OFF, the step output response is controlled by the slow linear filter as shown in Figure 29. The step response of the slow filter is programmable with the SF bits in the CONF register. Figure 28 shows the tradeoff between delay and noise for the different SF bit settings.

#### Figure 28:

Step Response Delay vs. Noise Band

SF	Step response delay (ms)	Max. RMS output noise (1 sigma) (degree)
00	2.2	0.015
01	1.1	0.021
10	0.55	0.030
11	0.286	0.043

#### Figure 29: Step Response (fast filter OFF)



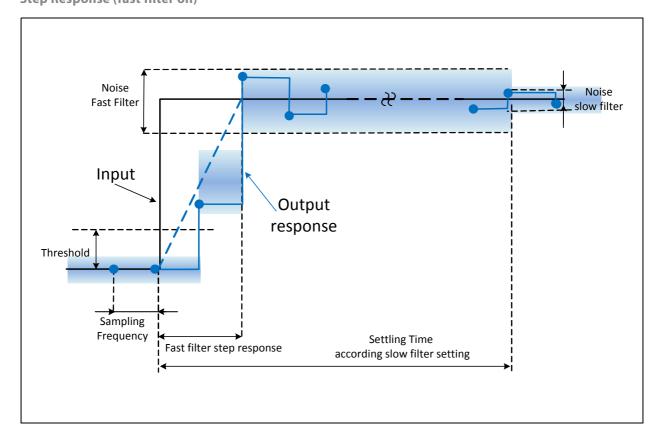
For a fast step response and low noise after settling, the fast filter can be enabled. The fast filter works only if the input variation is greater than the fast filter threshold, otherwise the output response is determined only by the slow filter. The fast filter threshold is programmed with the FTH bits in the CONF register. As shown in Figure 31, the fast filter (corresponds with SF=11) kicks in and takes care of a fast settling. The larger noise band of the fast filter is reduced again after the slow filter (depicted is setting SF=00) has taken over. The different noise bands are shown in Figure 28.

Figure 30: Fast Filter Threshold

FTH	Fast Filter Threshold (LSB)				
	Slow-to-fast filter	Fast-to-slow filter			
000	Slow filter only				
001	6	1			
010	7	1			
011	9	1			
100	18	2			
101	21	2			
110	24	2			
111	10	4			



#### Figure 31: Step Response (fast filter on)



#### Hysteresis

To suppress spurious toggling of the output when the magnet is not moving, a 1 to 3 LSB hysteresis of the 12-bit resolution can be enabled with the HYST bits in the CONF register.

#### **Magnet Detection**

As a safety and diagnostic feature, the AS5601 indicates the absence of the magnet. If the measured magnet field strength goes below the minimum specified level (Bz\_ERROR), quadrature output is not updated and the MD bit in the STATUS register is 0.



#### **Low Power Modes**

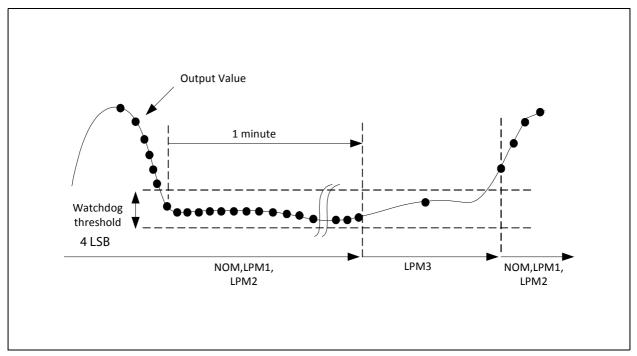
A digital state machine automatically manages the low power modes to reduce the average current consumption. Three low-power modes are available and can be enabled with the PM bits in the CONF register.

In a low-power mode, the fast filter is automatically disabled, because there is no need for a fast settling time if the output refresh is as fast as the polling cycles.

#### Watchdog Timer

The watchdog timer allows saving power by switching into LMP3 if the angle stays within the watchdog threshold of 4 LSB for at least one minute, as shown in Figure 32. The watchdog function can be enabled by setting the WD bit in the CONF register.





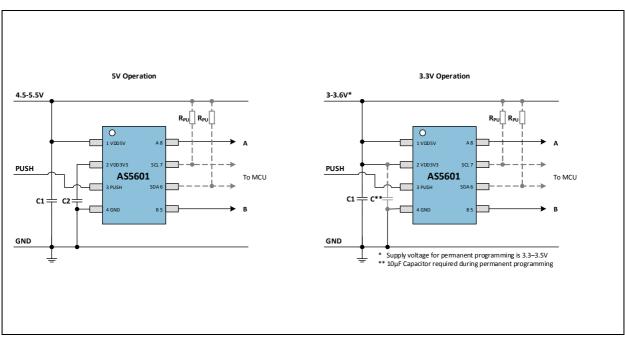


## **Application Information**

### Schematic

All required external components are shown below for the reference application diagram. To improve EMC and for remote applications, consider additional protection circuitry.

Figure 33: Application Diagram for Angle Readout and Programming



#### Figure 34: Recommended External Components

Component	Symbol	Value	Units	Notes	
VDD5V buffer capacitor	C1	100	nF	20%	
LDO regulator capacitor	C2	1	μF	20%; < 100 mΩ; Low ESR ceramic capacitor	
Optional pull-up for I <sup>2</sup> C bus	RPU	4.7	kΩ	refer to UM10204 for pull-up sizing	
To be fulfilled over temperature and lifetime					

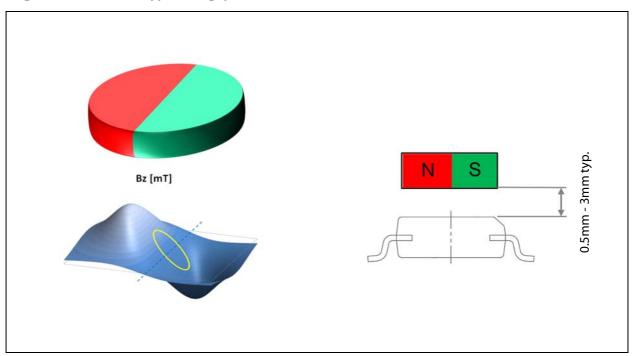


#### **Magnet Requirements**

The AS5601 requires a minimum magnetic field component Bz perpendicular to the sensitive area on the chip. The center of the sensitive area is in the center of the package.

Along the circumference of the Hall element circle the magnetic field Bz should be sine-shaped. The magnetic field gradient of Bz along the radius of the circle should be in the linear range of the magnet to eliminate displacement error by the differential measurement principle.

Figure 35: Magnetic Field Bz and Typical Airgap



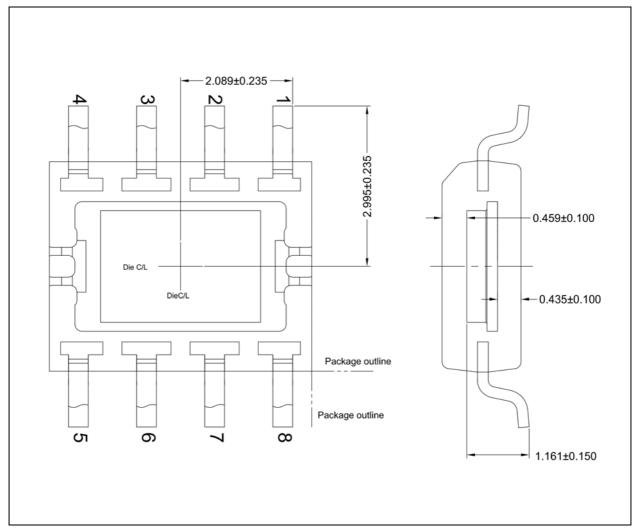
The typical airgap is between 0.5 mm and 3 mm, and it depends on the selected magnet. A larger and stronger magnet allows a larger airgap. Using the AGC value as a guide, the optimal airgap can be found by adjusting the distance between the magnet and the AS5601 so that the AGC value is in the center of its range. The maximum allowed displacement of the rotational axis of the reference magnet from the center of the package is 0.25 mm when using a magnet with a diameter of 6mm.



## **Mechanical Data**

The internal Hall elements are located in the center of the package on a circle with a radius of 1 mm.





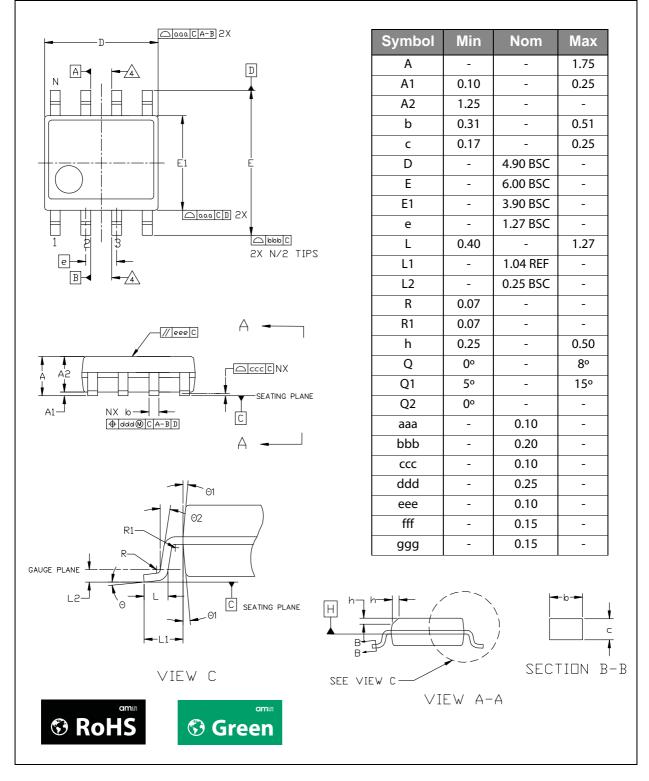
#### Note(s) and/or Footnote(s):

1. All dimensions in mm.

2. Die thickness 356µm nom.

## Package Drawings & Markings

Figure 37: SOIC8 Package Outline Drawing



#### Note(s) and/or Footnote(s):

1. Dimensions & tolerancing confirm to ASME Y14.5M-1994.

2. All dimensions are in millimeters. Angles are in degrees.



Figure 38: Package Marking

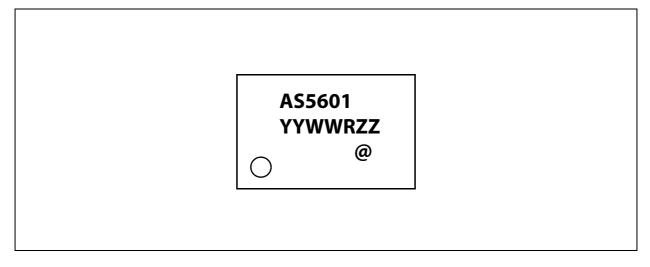


Figure 39: Packaging Code

YY	WW	R	ZZ	@
Last two digits of the current year	Manufacturing week	Plant identifier	Free choice/traceability code	Sublot identifier



## **Ordering & Contact Information**

Figure 40: Ordering Information

Ordering Code	Package	Marking	Delivery Form	Delivery Quantity
AS5601-ASOT	SOIC-8	AS5601	13″ Tape & Reel in dry pack	2500 pcs
AS5601-ASOM	SOIC-8	AS5601	7" Tape & Reel in dry pack	500 pcs

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# **Revision Information**

Changes from 1-01 (2014-Oct-13) to current revision 1-02 (2014-Oct-29)	Page
Updated Figure 5	4

#### Note(s) and/or Footnote(s):

1. Page and figure numbers for the previous version may differ from page and figure numbers in the current revision.

2. Correction of typographical errors is not explicitly mentioned.

am



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