

FEATURES

Conversion gain: 10 dB typical
Image rejection: 30 dBc typical
Noise figure: 6 dB typical
Input power for 1 dB compression (P1dB): -10 dBm typical
Input third-order intercept (IP3): -2 dBm typical
Input second-order intercept (IP2): 25 dBm typical
6x LO leakage at RFIN: -40 dBm typical
Radio frequency (RF) return loss: 10 dB typical
Local oscillator (LO) return loss: 20 dB typical
Die size: 3.599 mm × 2.199 mm × 0.05 mm

APPLICATIONS

E-band communication systems
High capacity wireless backhauls
Test and measurement

GENERAL DESCRIPTION

The HMC7587 is an integrated, E-band gallium arsenide (GaAs), monolithic microwave integrated circuit (MMIC), in-phase/quadrature (I/Q) downconverter chip that operates from 81 GHz to 86 GHz. The HMC7587 provides a small signal conversion gain of 10 dB with 30 dBc of image rejection across the frequency band. The device uses a low noise amplifier followed by an image rejection mixer that is driven by a 6x multiplier.

The image rejection mixer eliminates the need for a filter following the low noise amplifier. Differential I and Q mixer outputs are provided for direct conversion applications. Alternatively, the outputs can be combined using an external 90° hybrid and two external 180° hybrids to allow for single-sideband applications. All data includes the effect of a 3 mil wide ribbon wedge bond on the RF port, and a 1 mil gold wire wedge bond on the intermediate frequency (IF) ports.

FUNCTIONAL BLOCK DIAGRAM

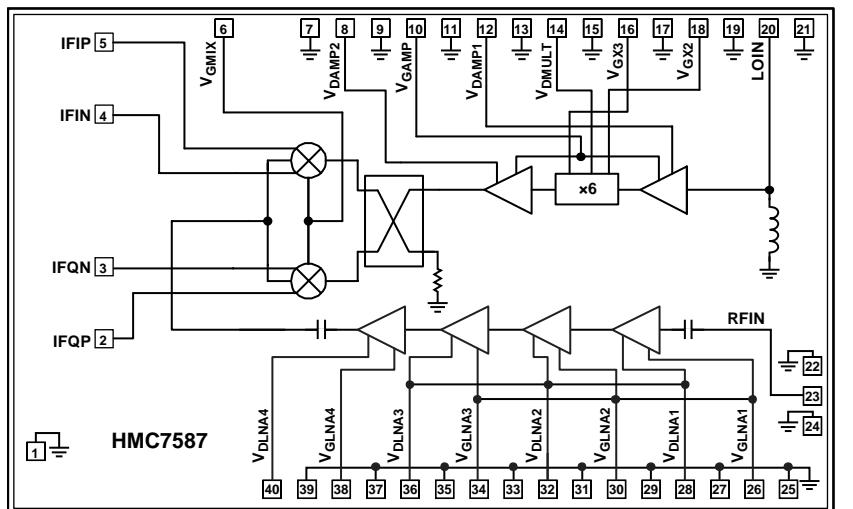


Figure 1.

13141-001

Rev. A

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REVISION HISTORY

3/16—Revision A: Initial Version

SPECIFICATIONS

$T_A = 25^\circ\text{C}$, IF = 500 MHz, $V_{GMIX} = -1$ V, $V_{DAMPx} = 4$ V, $V_{DMULT} = 1.5$ V, voltage on the V_{DLNAx} pins (V_{DLNA}) = 3 V, LO = 2 dBm, upper sideband selected. Measurements performed as a downconverter with external 90° and 180° hybrids at the IF ports, unless otherwise noted.

Table 1.

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
OPERATING CONDITIONS					
RF Frequency Range		81	86		GHz
LO Frequency Range		11.83	14.33		GHz
IF Frequency Range		0	10		GHz
LO Drive Range		2	8		dBm
PERFORMANCE					
Conversion Gain		8	10		dB
Image Rejection		20	30		dBc
Input Third-Order Intercept (IP3)			-2		dBm
Input Second-Order Intercept (IP2)			25		dBm
Input Power for 1 dB Compression (P1dB)			-10		dBm
6× LO Leakage at RF Input (RFIN)			-40		dBm
1× LO Leakage at IF Output (IFOUT)			-50		dBm
Amplitude Balance ¹			-0.5		dB
Phase Balance ¹			±4		Degrees
Noise Figure			6		dB
RF Return Loss	LO = 2 dBm at 12 GHz		10		dB
LO Return Loss			20		dB
IF Return Loss ¹			25		dB
POWER SUPPLY					
Supply Current					
I_{DAMP}^2			175		mA
I_{DMULT}^3	Under LO drive		80		mA
I_{DLNA}^4			50		mA

¹ These measurements were performed without external hybrids at the IF ports.

² Adjust V_{GAMP} between -2 V and 0 V to achieve the total quiescent current, $I_{DAMP} = I_{DAMP1} + I_{DAMP2} = 175$ mA.

³ Adjust V_{GX2} and V_{GX3} between -2 V and 0 V to achieve the quiescent current, $I_{DMULT} = 1$ mA to 2 mA. See the Applications Information section for more information.

⁴ Adjust V_{GLNAX} between -2 V and 0 V to achieve the quiescent current, $I_{DLNA1} + I_{DLNA2} + I_{DLNA3} + I_{DLNA4} = 50$ mA.

ABSOLUTE MAXIMUM RATINGS

Table 2.

Parameter	Rating
Drain Bias Voltage V_{DAMP1}, V_{DAMP2}	4.5 V
V_{DMULT}	3 V
$V_{DLNA1}, V_{DLNA2}, V_{DLNA3}, V_{DLNA4}$	4.5 V
Gate Bias Voltage V_{GAMP}	-3 V to 0 V
V_{GX2}, V_{GX3}	-3 V to 0 V
$V_{GLNA1}, V_{GLNA2}, V_{GLNA3}, V_{GLNA4}$	-3 V to 0 V
V_{GMIX}	-3 V to 0 V
LO Input Power	10 dBm
Maximum Junction Temperature (to Maintain 1 Million Hours Mean Time to Failure (MTTF))	175°C
Storage Temperature Range	-65°C to +150°C
Operating Temperature Range	-55°C to +85°C

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

THERMAL RESISTANCE

Table 3. Thermal Resistance

Package Type	θ_{JC}^1	Unit
40-Pad Bare Die [CHIP]	61.7	°C/W

¹ Based on ABLEBOND® 84-1LMIT as die attach epoxy with thermal conductivity of 3.6 W/mK.

ESD CAUTION



ESD (electrostatic discharge) sensitive device.

Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

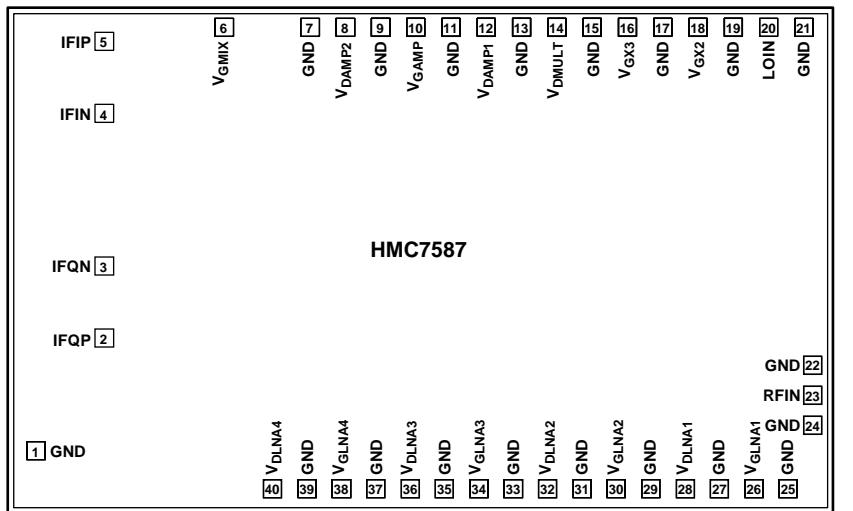


Figure 2. Pad Configuration

Table 4. Pad Function Descriptions

Pad No.	Mnemonic	Description
1, 7, 9, 11, 13, 15, 17, 19, 21, 22, 24, 25, 27, 29, 31, 33, 35, 37, 39	GND	Ground Connect (See Figure 3).
2, 3	IFQP, IFQN	Positive and Negative IF Q Inputs. These pads are dc-coupled. When operation to dc is not required, block these pads externally using a series capacitor with a value chosen to pass the necessary frequency range. For operation to dc, these pads must not source or sink more than 3 mA of current or die malfunction and possible die failure may result (see Figure 4).
4, 5	IFIN, IFIP	Negative and Positive IF I Inputs. These pads are dc-coupled. When operation to dc is not required, block these pads externally using a series capacitor with a value chosen to pass the necessary frequency range. For operation to dc, these pads must not source or sink more than 3 mA of current or die malfunction and possible die failure may result (see Figure 4).
6	VGMIX	Gate Voltage for the FET Mixer (See Figure 5). External bypass capacitors of 120 pF, 0.01 µF, and 4.7 µF are recommended (see Figure 211).
8, 12	VDAMP2, VDAMP1	Power Supply Voltage for the First and the Second Stage LO Amplifier (See Figure 5). External bypass capacitors of 120 pF, 0.01 µF, and 4.7 µF are recommended (see Figure 211).
10	VGAMP	Gate Voltage for the First and the Second Stage LO Amplifier (See Figure 5). External bypass capacitors of 120 pF, 0.01 µF, and 4.7 µF are recommended (see Figure 211).
14	VDMULT	Power Supply Voltage for the LO Multiplier (See Figure 5). External bypass capacitors of 120 pF, 0.01 µF, and 4.7 µF are recommended (see Figure 211).
16, 18	VGX3, VGX2	Gate Voltage for the LO Multiplier (See Figure 5). External bypass capacitors of 120 pF, 0.01 µF, and 4.7 µF are recommended (see Figure 211).
20	LOIN	Local Oscillator Input. This pad is dc-coupled and matched to 50 Ω (see Figure 6).
23	RFIN	RF Input. This pad is ac-coupled and matched to 50 Ω (see Figure 7).
26, 30, 34, 38	VGLNA1, VGLNA2, VGLNA3, VGLNA4	Gate Voltage for the Low Noise Amplifier (See Figure 8). External bypass capacitors of 120 pF, 0.01 µF, and 4.7 µF are recommended (see Figure 211).
28, 32, 36, 40	VDLNA1, VDLNA2, VDLNA3, VDLNA4	Power Supply Voltage for the Low Noise Amplifier (See Figure 8). External bypass capacitors of 120 pF, 0.01 µF, and 4.7 µF are recommended (see Figure 211).
Die Bottom	GND	Ground. The die bottom must be connected to RF/dc ground (see Figure 3).

INTERFACE SCHEMATICS



Figure 3. GND Interface

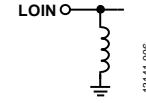


Figure 6. LOIN Interface

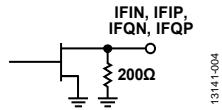


Figure 4. IFIN, IFIP, IFQN, and IFQP Interface

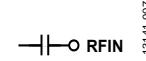
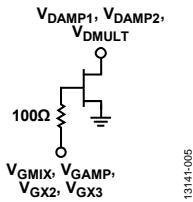
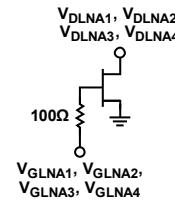


Figure 7. RFIN Interface

Figure 5. V_{GMIX} , V_{DAMP1} , V_{DAMP2} , V_{DMULT} , V_{GAMP} , V_{GX2} , and V_{GX3} InterfaceFigure 8. V_{DLNA1} , V_{DLNA2} , V_{DLNA3} , V_{DLNA4} , V_{GLNA1} , V_{GLNA2} , V_{GLNA3} , and V_{GLNA4} Interface

TYPICAL PERFORMANCE CHARACTERISTICS

UPPER SIDEBAND SELECTED, IF = 500 MHz

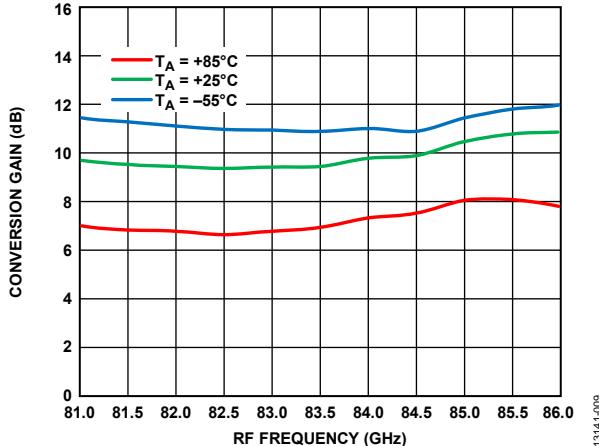


Figure 9. Conversion Gain vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 500 MHz, Voltage on the V_{DLNA} Pins
(V_{DLNA}) = 4 V

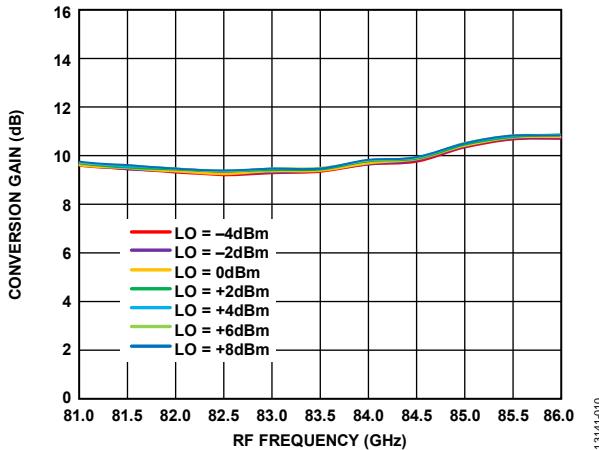


Figure 10. Conversion Gain vs. RF Frequency at Various LO Powers,
RFIN = -20 dBm, IF = 500 MHz, V_{DLNA} = 4 V

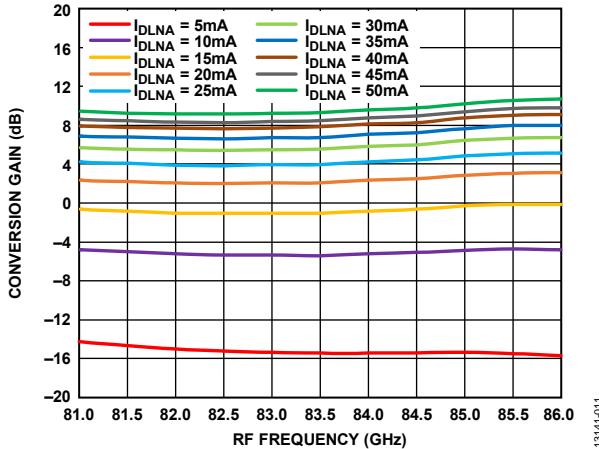


Figure 11. Conversion Gain vs. RF Frequency at Various I_{DLNA} Values,
RFIN = -20 dBm, LO = 2 dBm, IF = 500 MHz, V_{DLNA} = 4 V

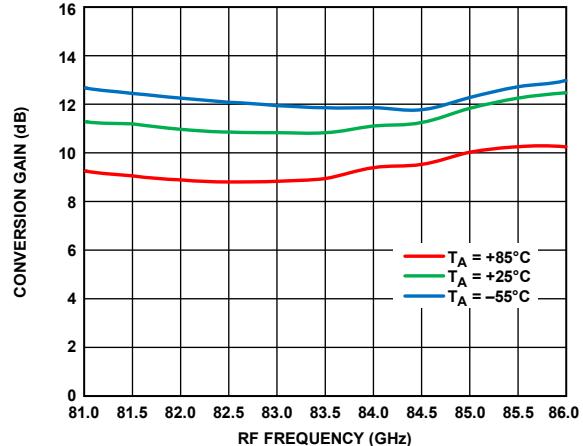


Figure 12. Conversion Gain vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 500 MHz, V_{DLNA} = 3 V

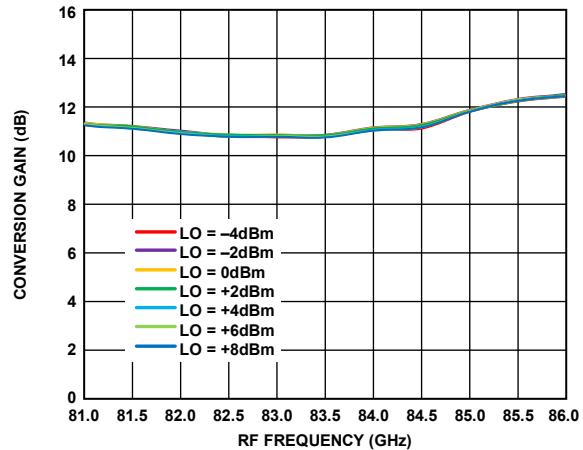


Figure 13. Conversion Gain vs. RF Frequency at Various LO Powers,
RFIN = -20 dBm, IF = 500 MHz, V_{DLNA} = 3 V

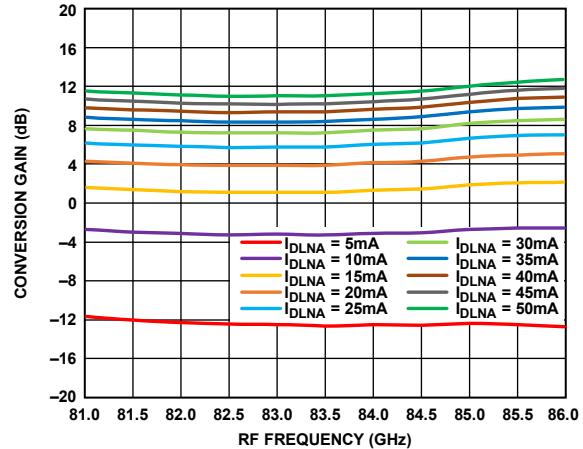
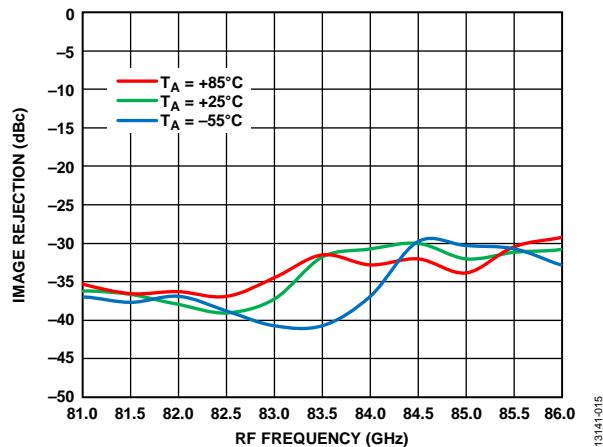
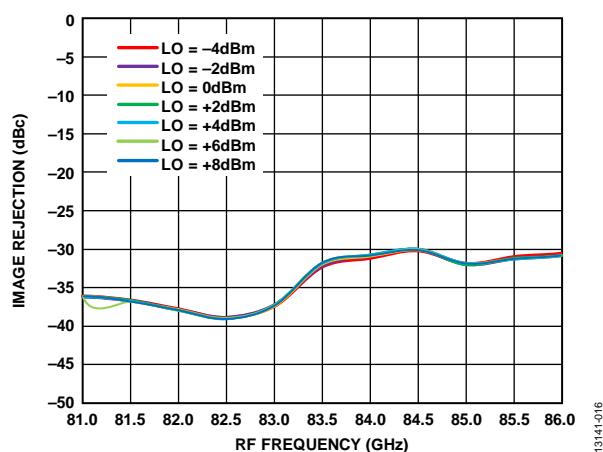


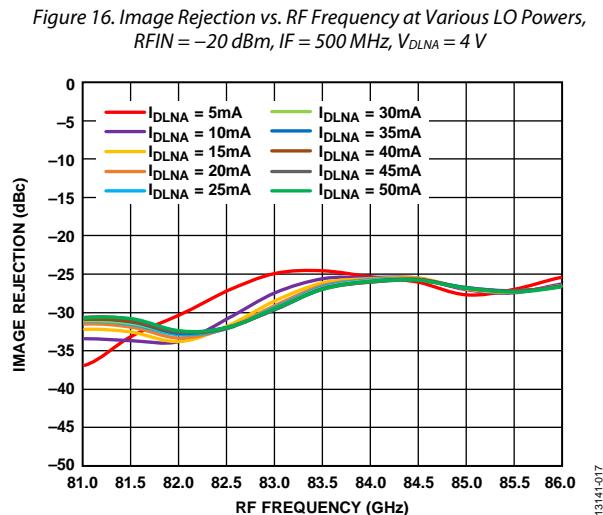
Figure 14. Conversion Gain vs. RF Frequency at Various I_{DLNA} Values,
RFIN = -20 dBm, LO = 2 dBm, IF = 500 MHz, V_{DLNA} = 3 V



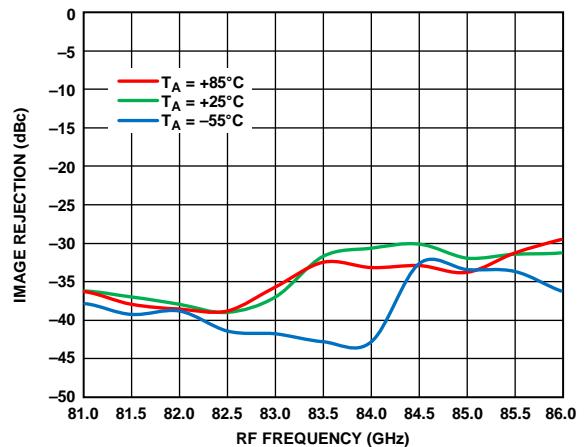
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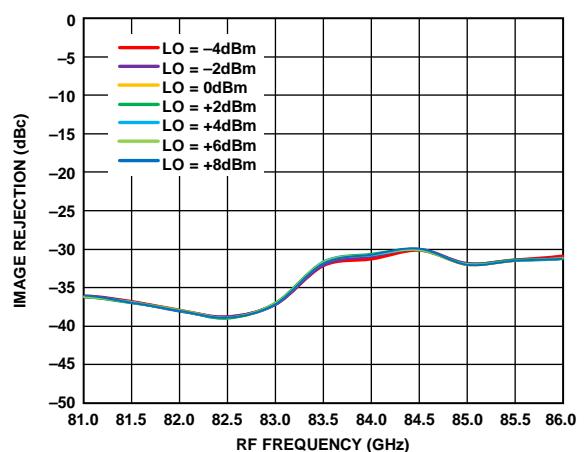
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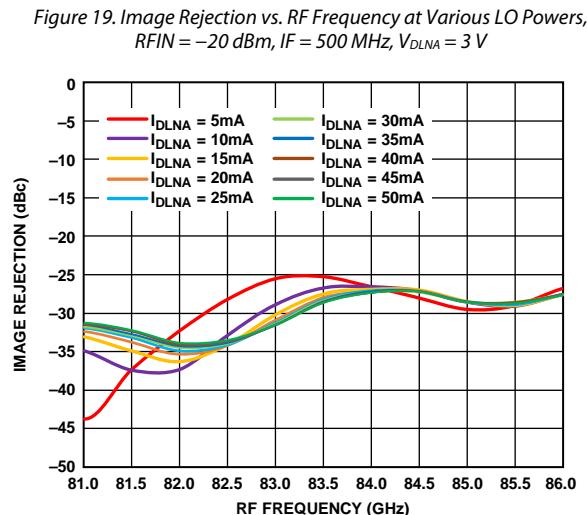
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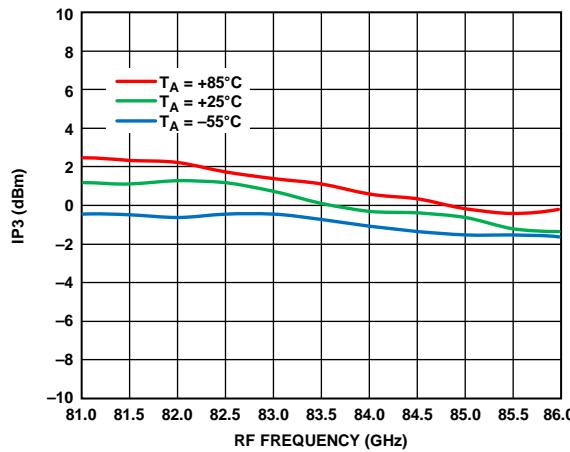
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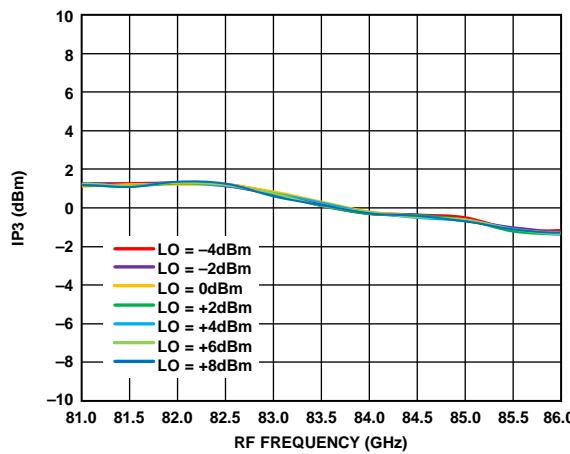
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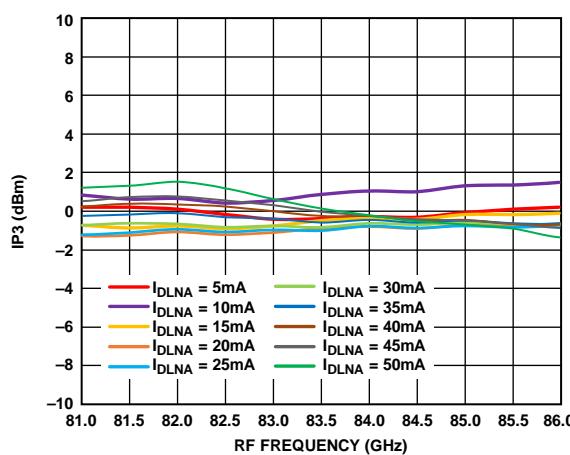
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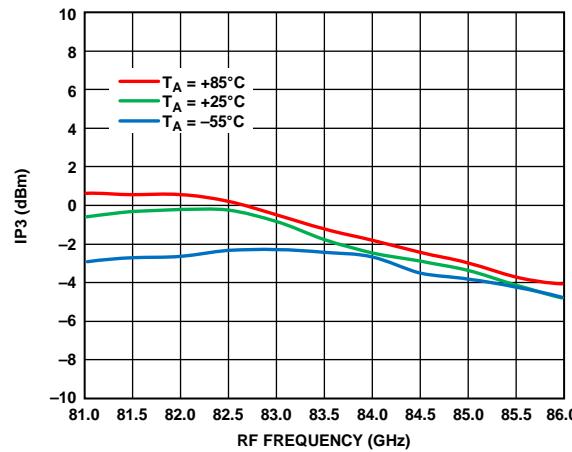
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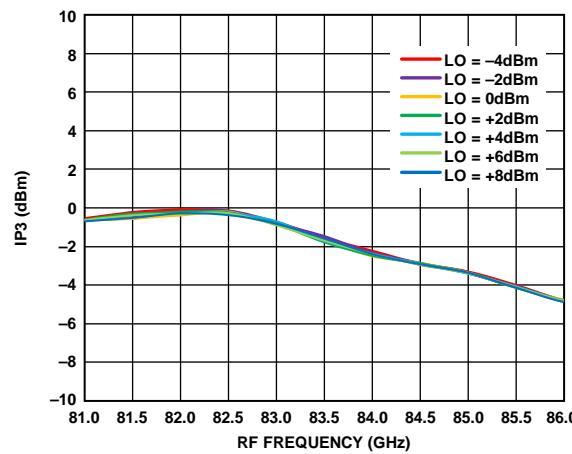
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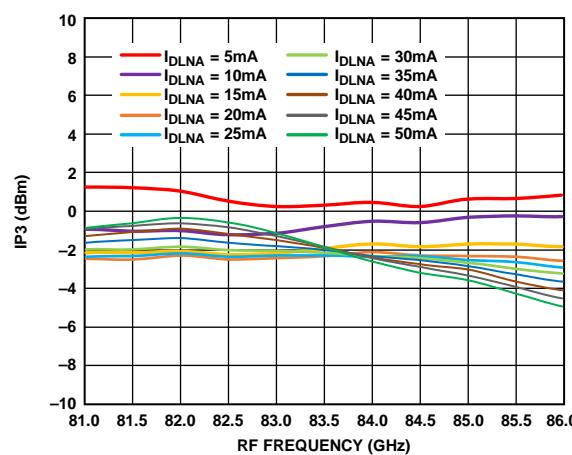
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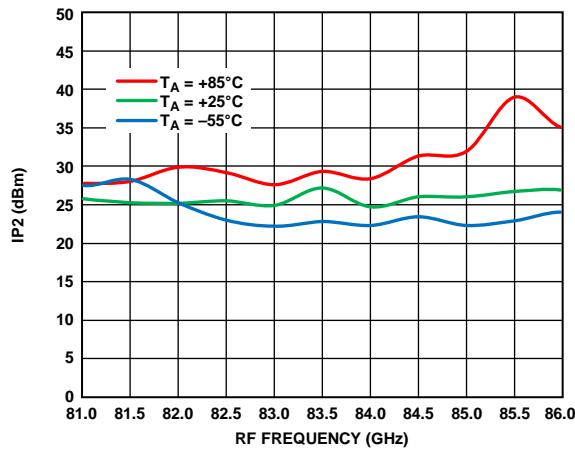
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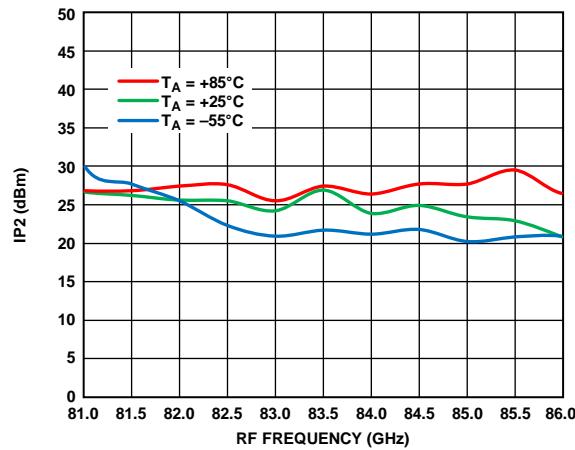
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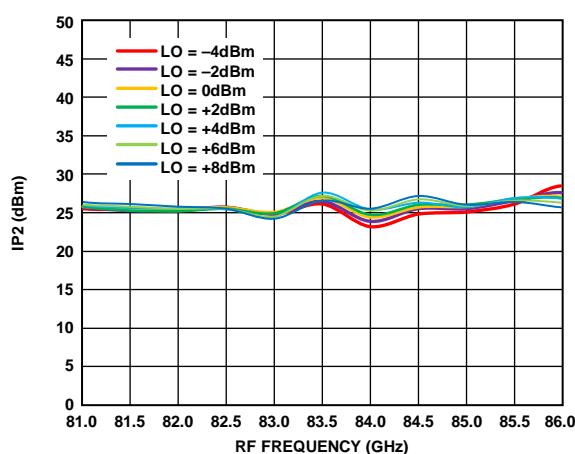
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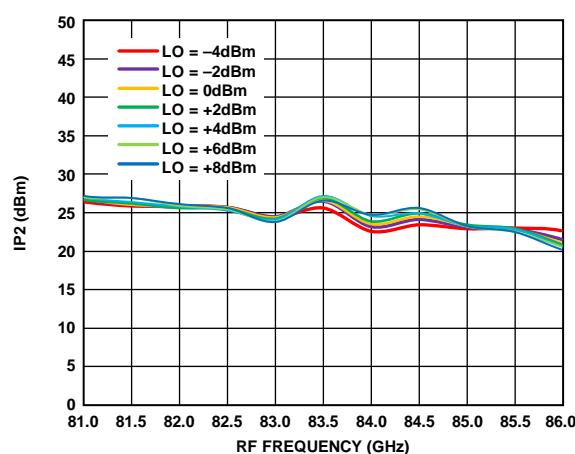
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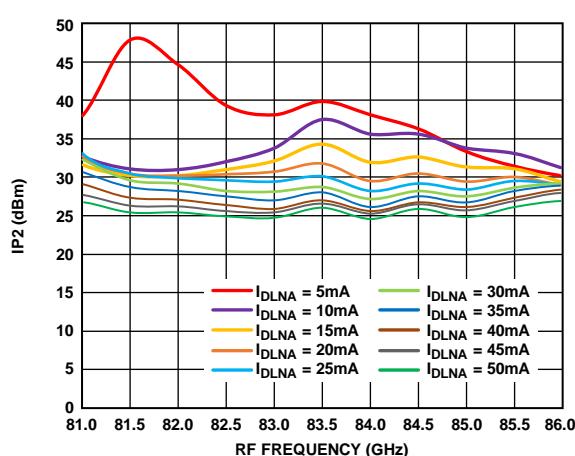
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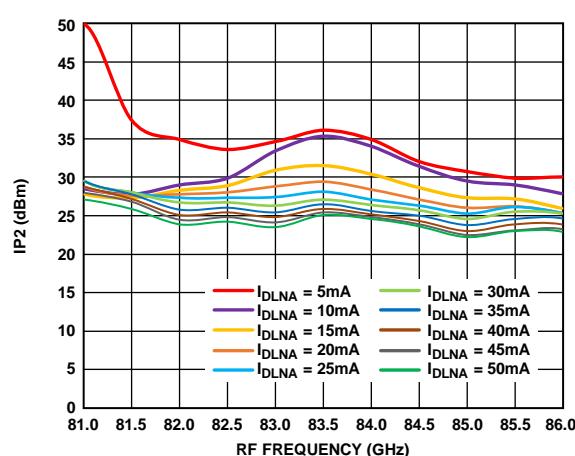
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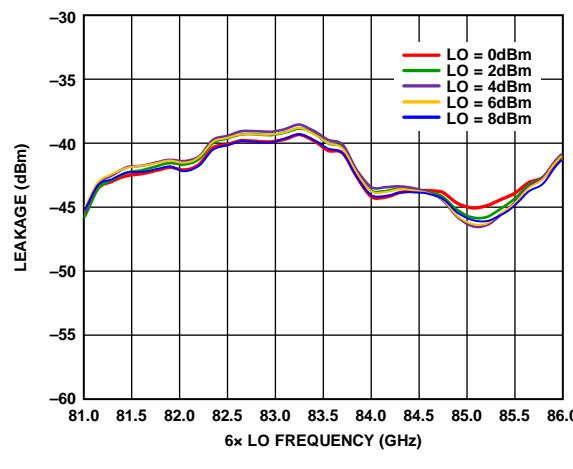
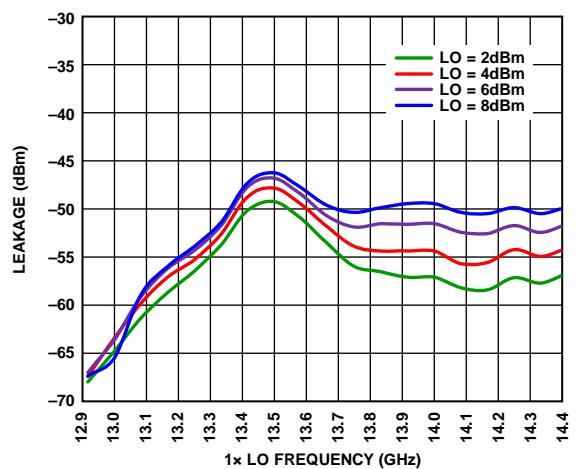
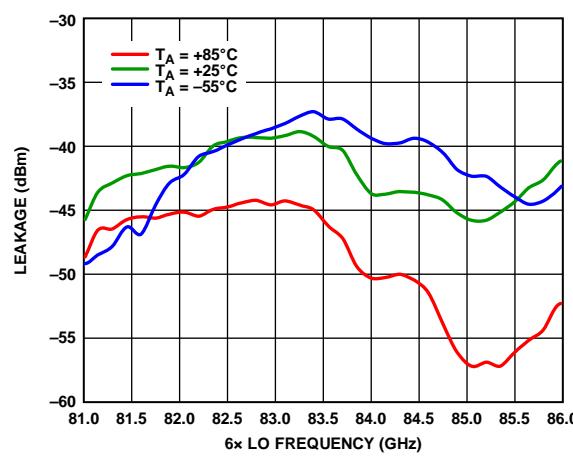
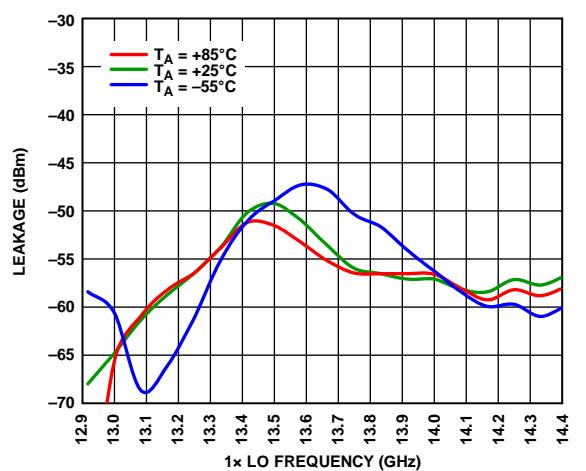
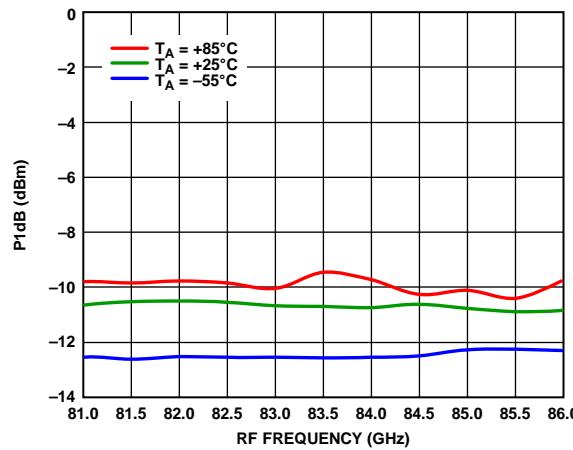
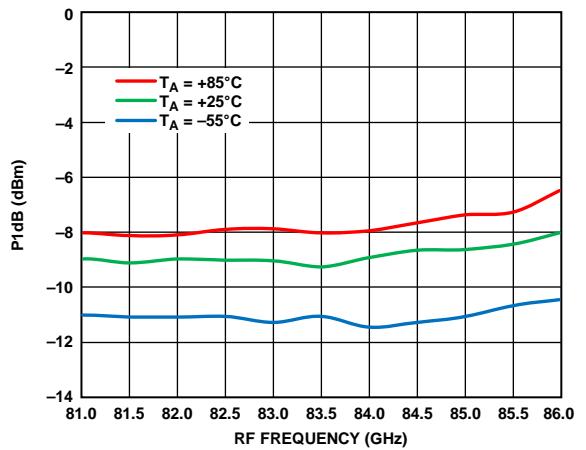
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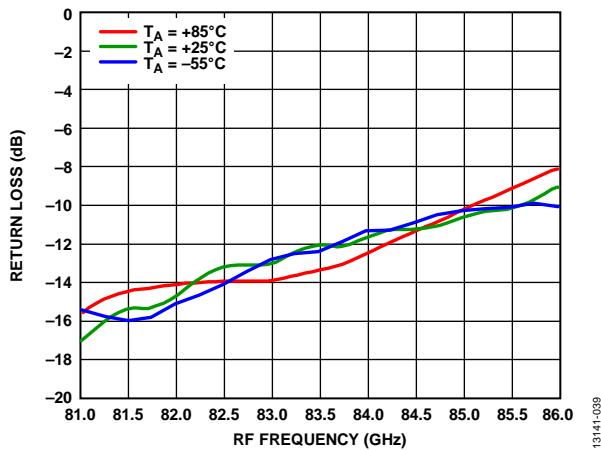


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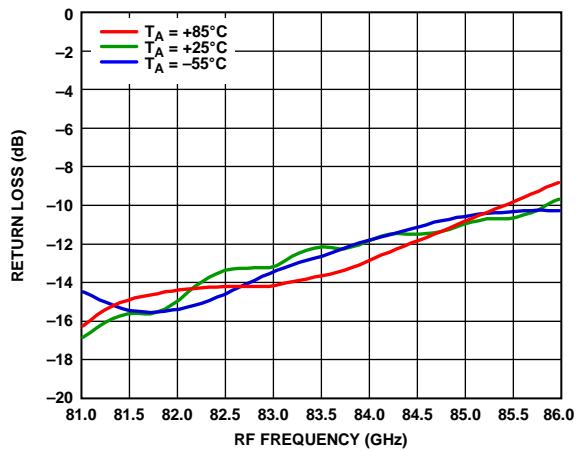


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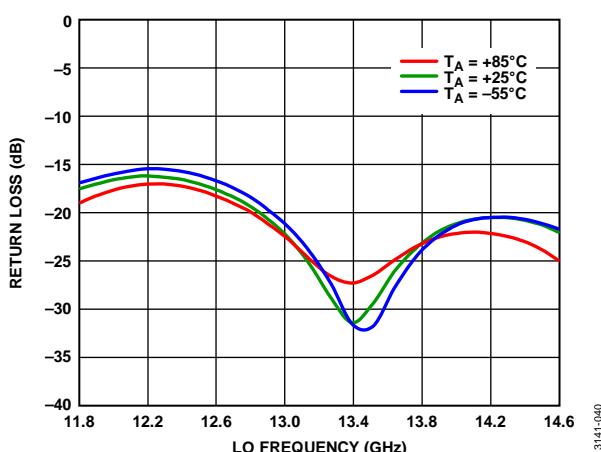




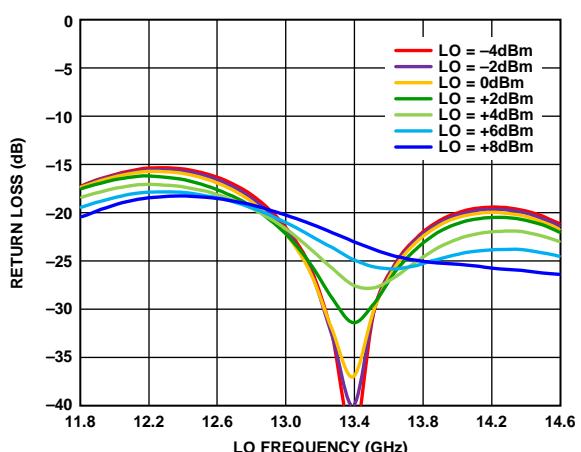
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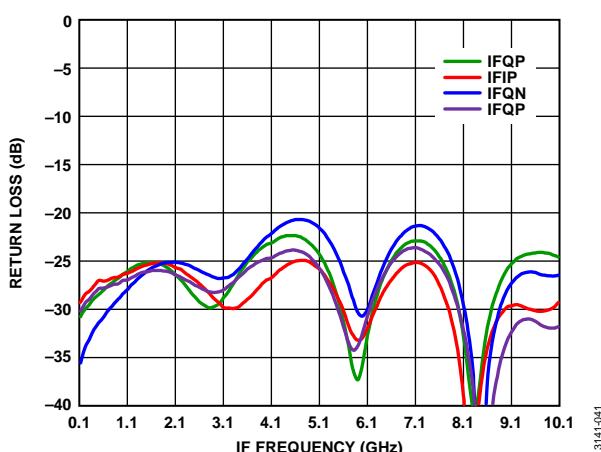
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UPPER SIDEBAND SELECTED, IF = 1000 MHz

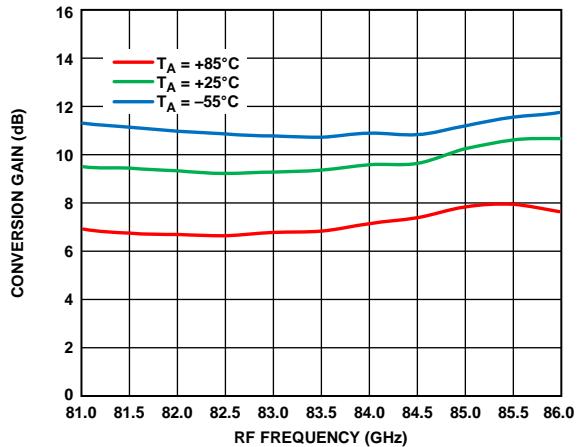


Figure 44. Conversion Gain vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 1000 MHz, V_{DLNA} = 4 V

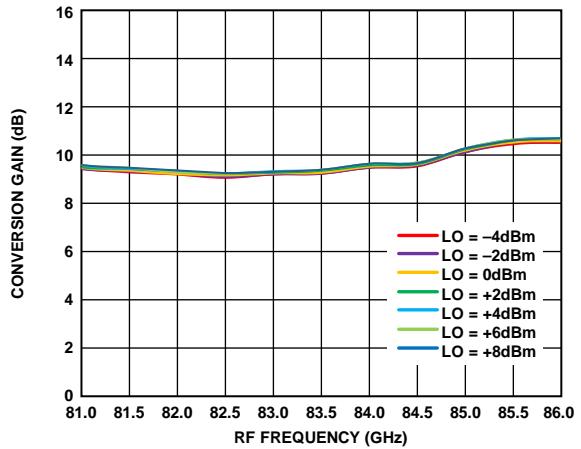


Figure 45. Conversion Gain vs. RF Frequency at Various LO Powers,
RFIN = -20 dBm, IF = 1000 MHz, V_{DLNA} = 4 V

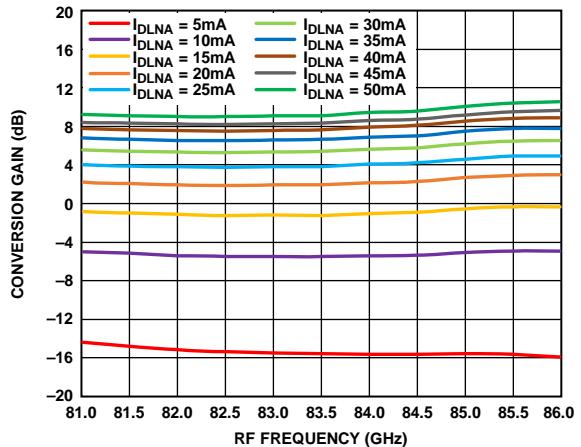


Figure 46. Conversion Gain vs. RF Frequency at Various I_{DLNA} Values,
RFIN = -20 dBm, LO = 2 dBm, IF = 1000 MHz, V_{DLNA} = 4 V

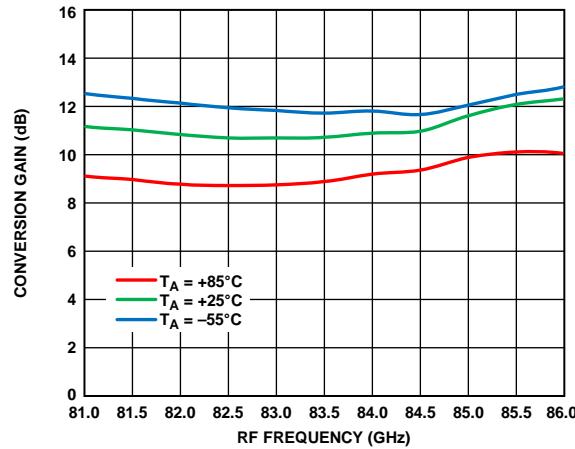


Figure 47. Conversion Gain vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 1000 MHz, V_{DLNA} = 3 V

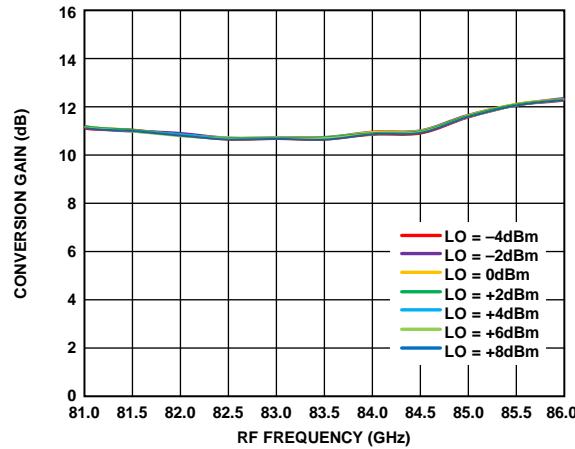


Figure 48. Conversion Gain vs. RF Frequency at Various LO Powers,
RFIN = -20 dBm, IF = 1000 MHz, V_{DLNA} = 3 V

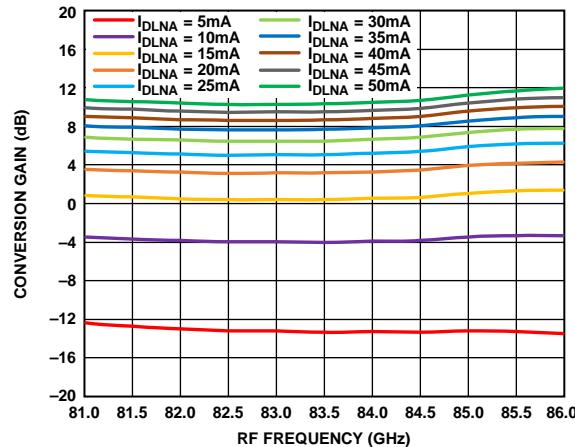
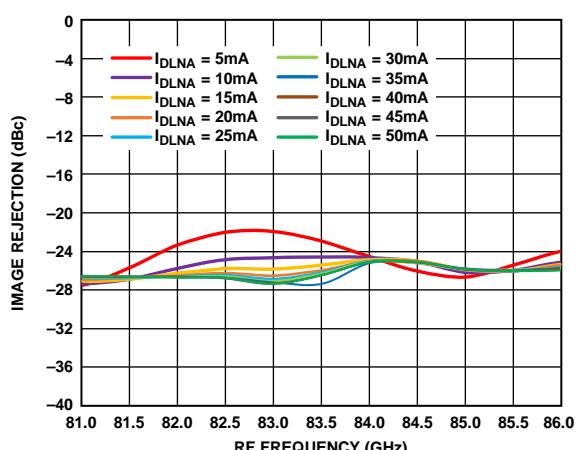
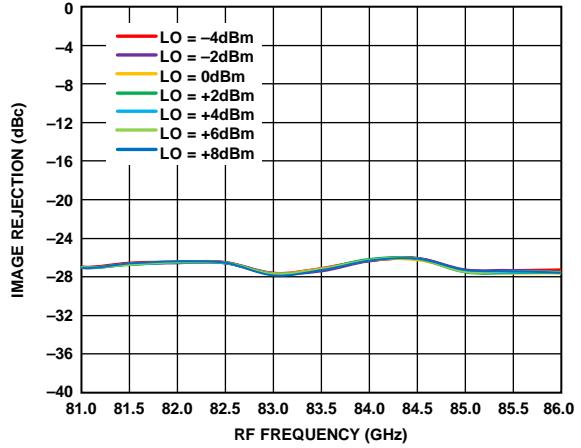
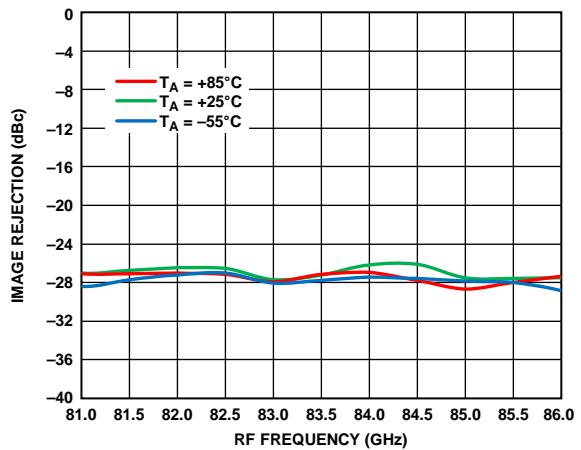
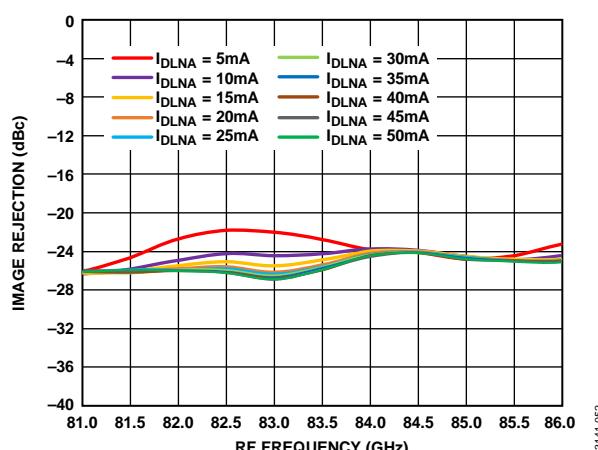
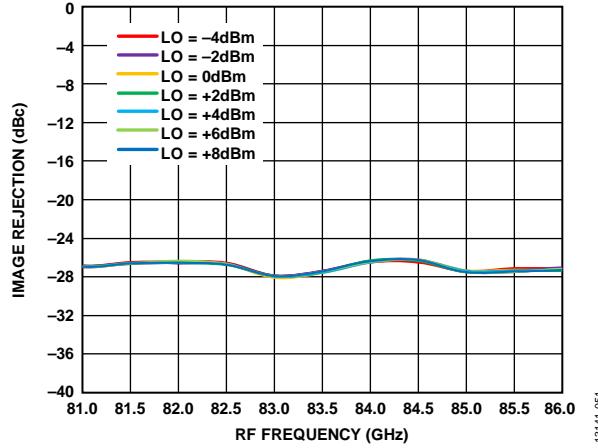
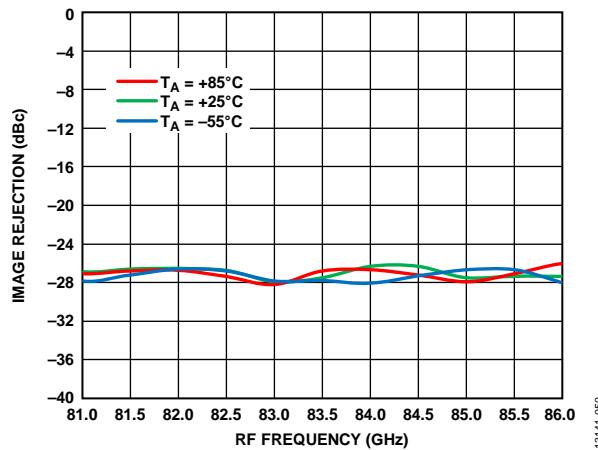
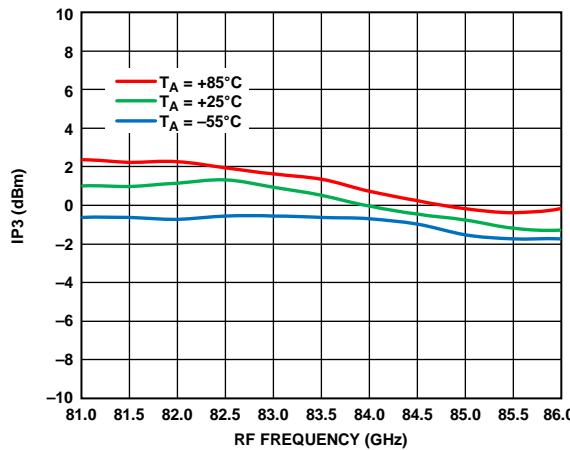
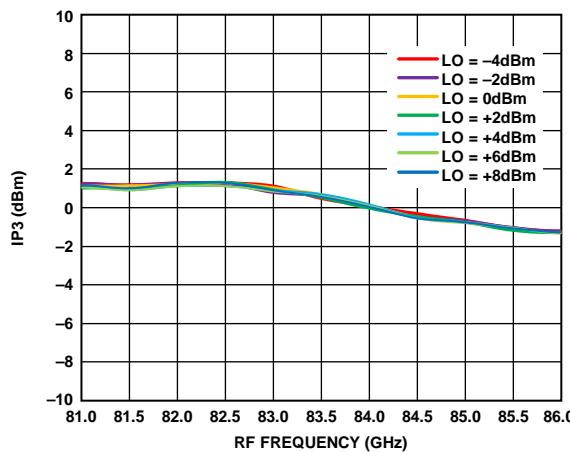


Figure 49. Conversion Gain vs. RF Frequency at Various I_{DLNA} Values,
RFIN = -20 dBm, LO = 2 dBm, IF = 1000 MHz, V_{DLNA} = 3 V

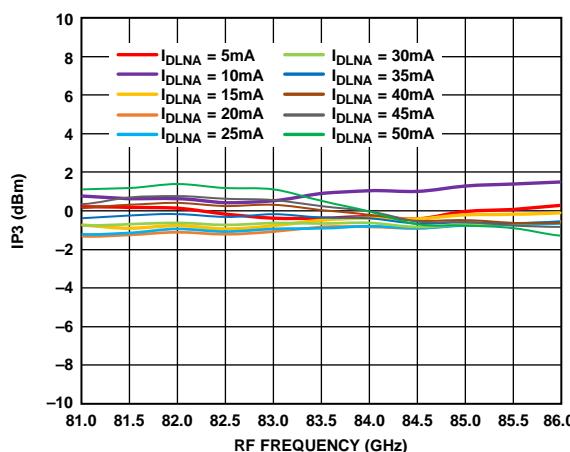




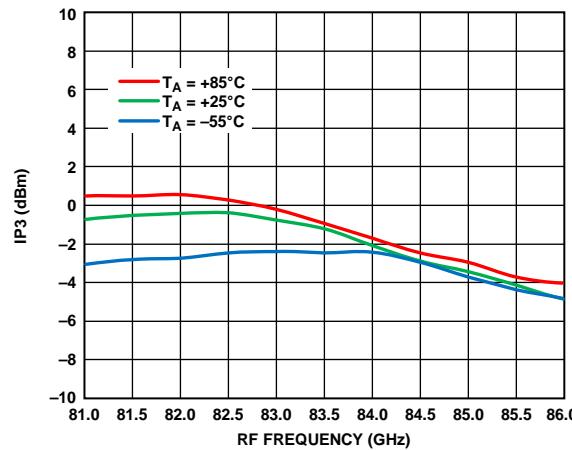
13141-056



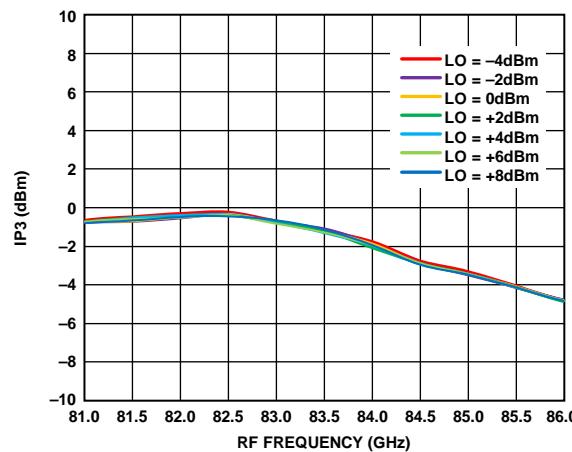
13141-057



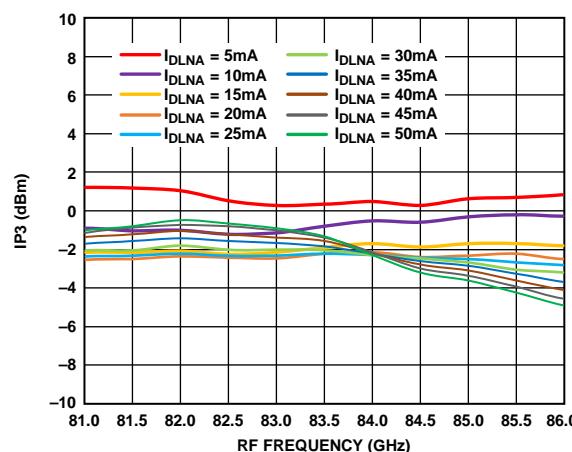
13141-058



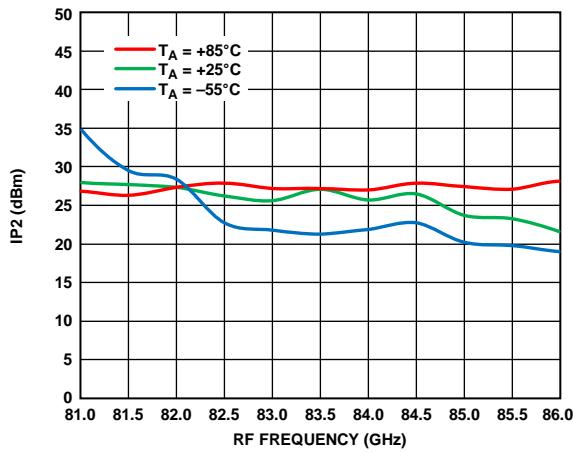
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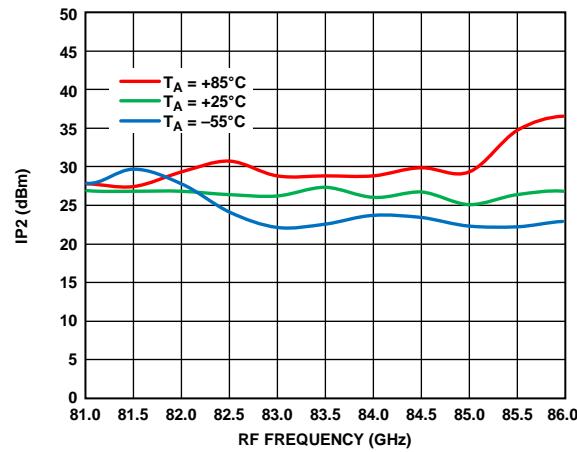
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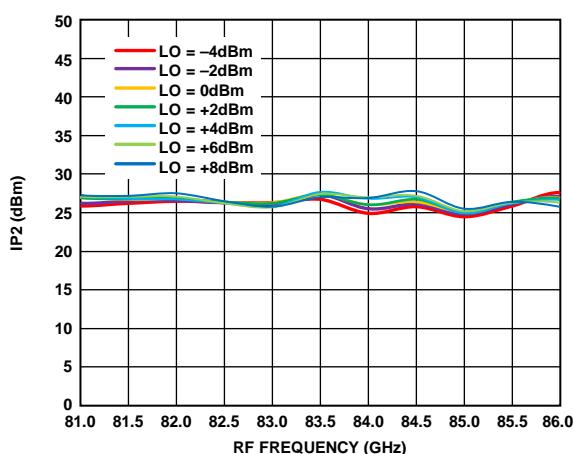
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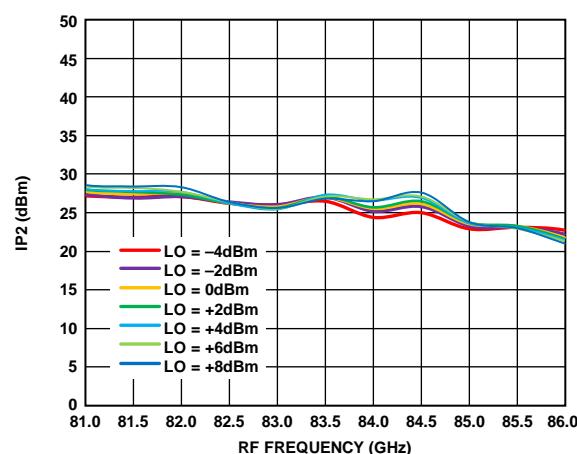
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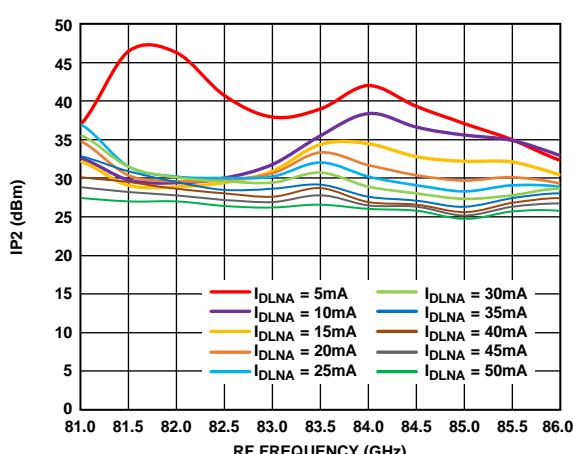
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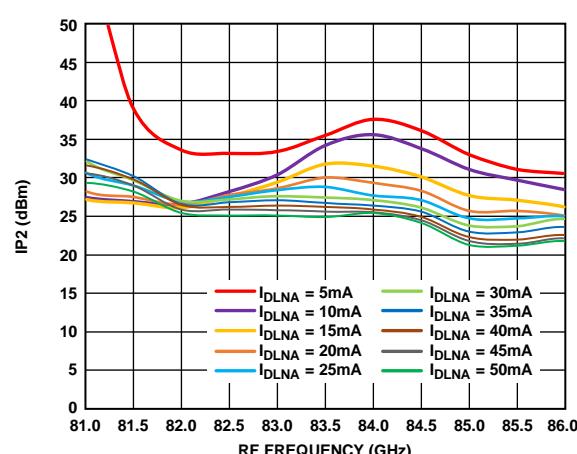
13141-063



13141-066



13141-064



13141-067

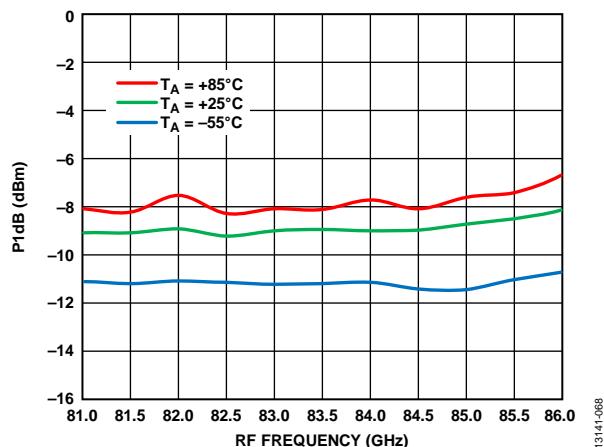


Figure 68. Input P1dB vs. RF Frequency at Various Temperatures,
LO = 2 dBm, IF = 1000 MHz, $V_{DLNA} = 4\text{ V}$

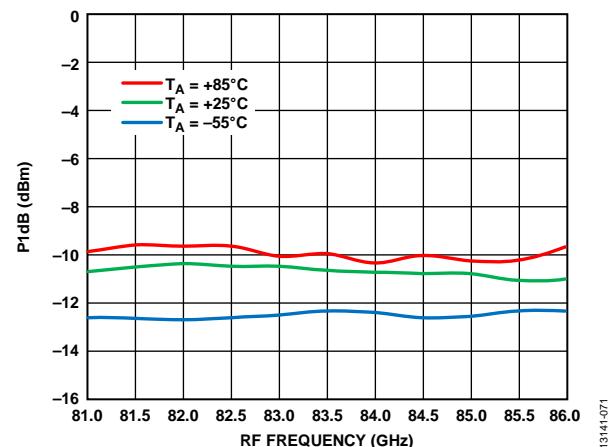


Figure 69. Input P1dB vs. RF Frequency at Various Temperatures,
LO = 2 dBm, IF = 1000 MHz, $V_{DLNA} = 3\text{ V}$

UPPER SIDEBAND SELECTED, IF = 2000 MHz

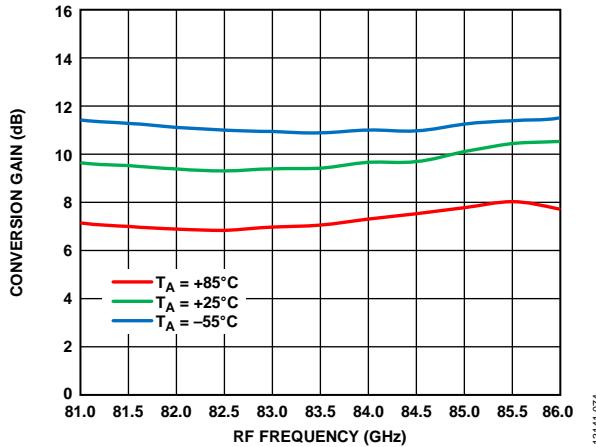


Figure 70. Conversion Gain vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, $V_{DLNA} = 4$ V

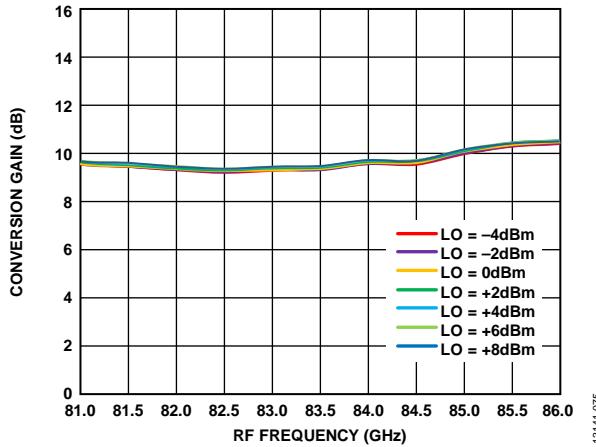


Figure 71. Conversion Gain vs. RF Frequency at Various LO Powers,
RFIN = -20 dBm, IF = 2000 MHz, $V_{DLNA} = 4$ V

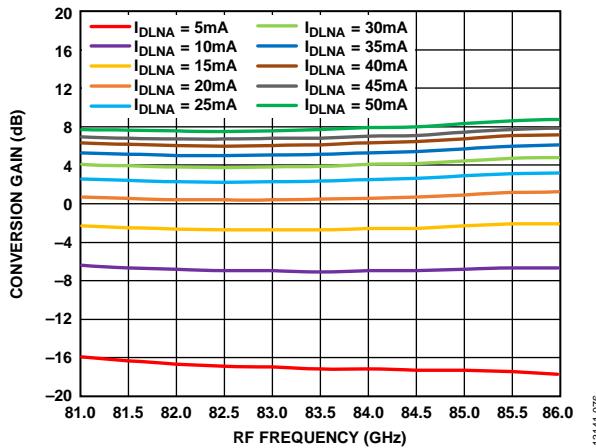


Figure 72. Conversion Gain vs. RF Frequency at Various I_{DLNA} Values,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, $V_{DLNA} = 4$ V

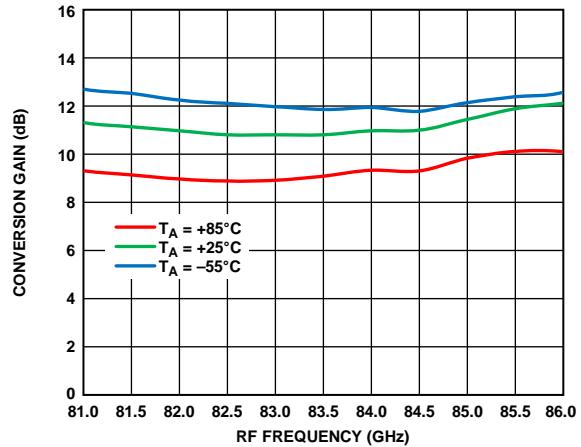


Figure 73. Conversion Gain vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, $V_{DLNA} = 3$ V

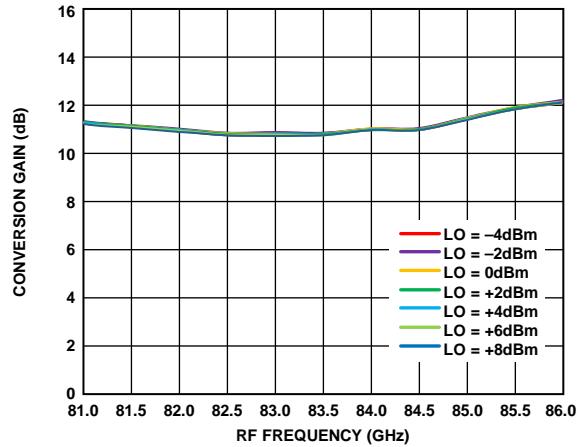


Figure 74. Conversion Gain vs. RF Frequency at Various LO Powers,
RFIN = -20 dBm, IF = 2000 MHz, $V_{DLNA} = 3$ V

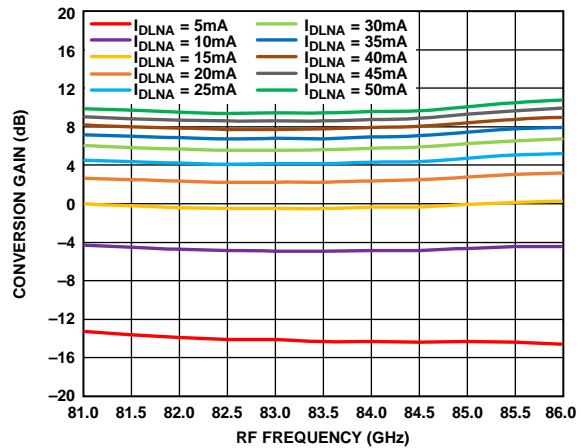
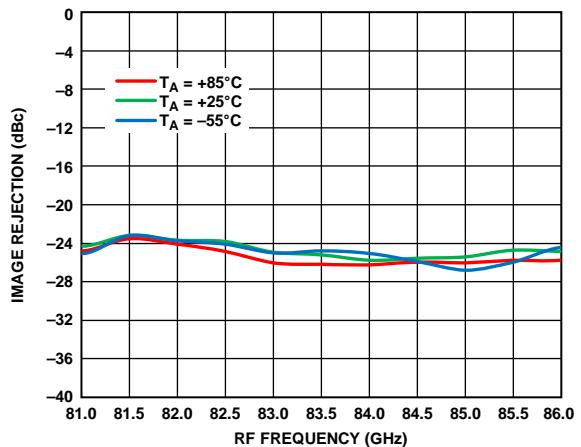
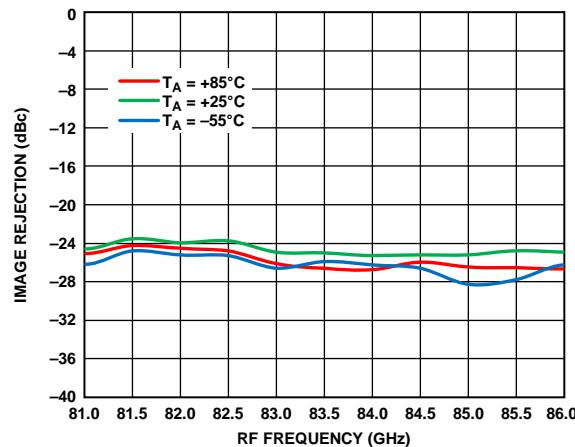


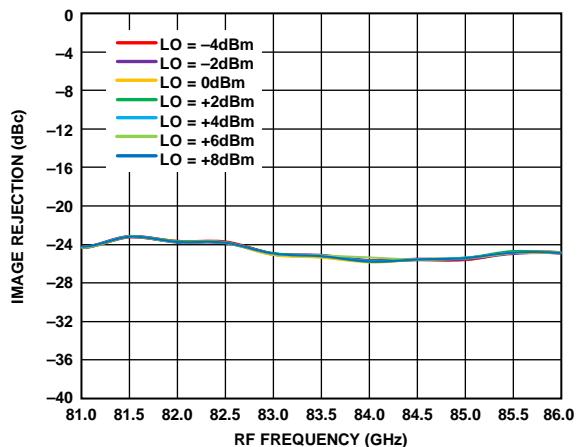
Figure 75. Conversion Gain vs. RF Frequency at Various I_{DLNA} Values,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, $V_{DLNA} = 3$ V



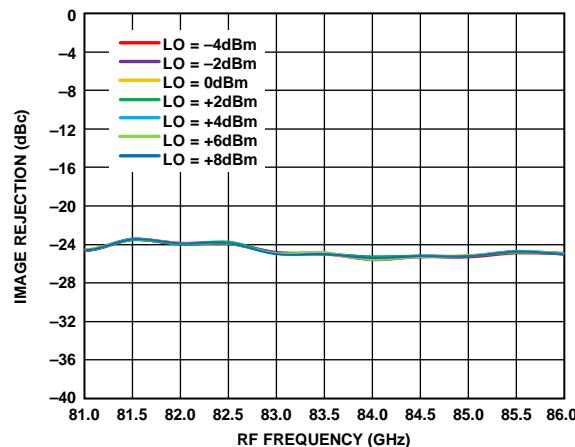
13141-080



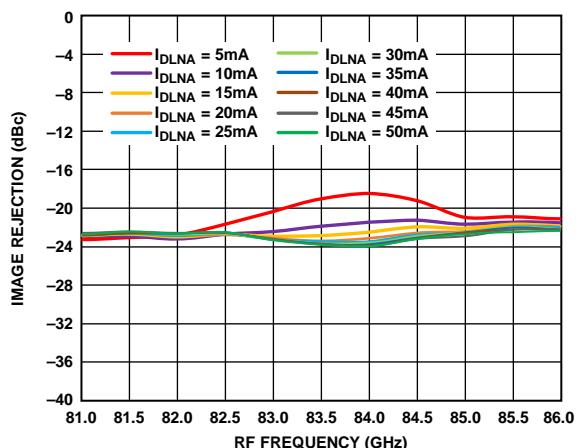
13141-083



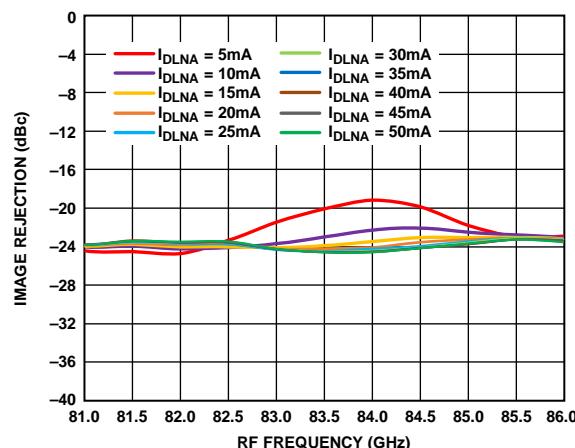
13141-081



13141-084



13141-082



13141-085

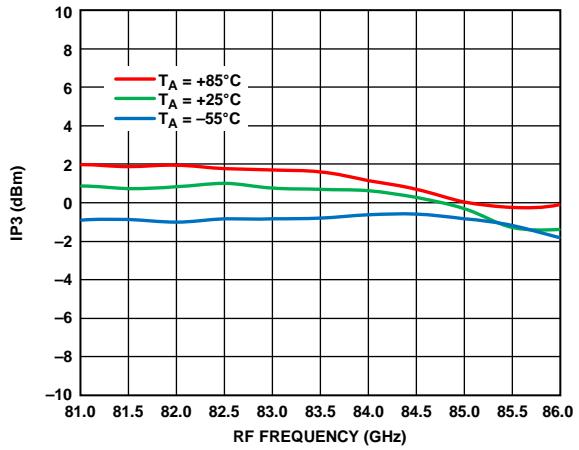


Figure 82. Input IP3 vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, $V_{\text{DLNA}} = 4 \text{ V}$

13141-086

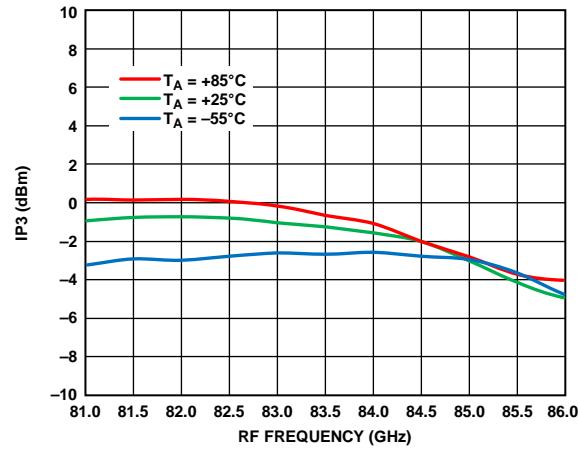


Figure 85. Input IP3 vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, $V_{\text{DLNA}} = 3 \text{ V}$

13141-089

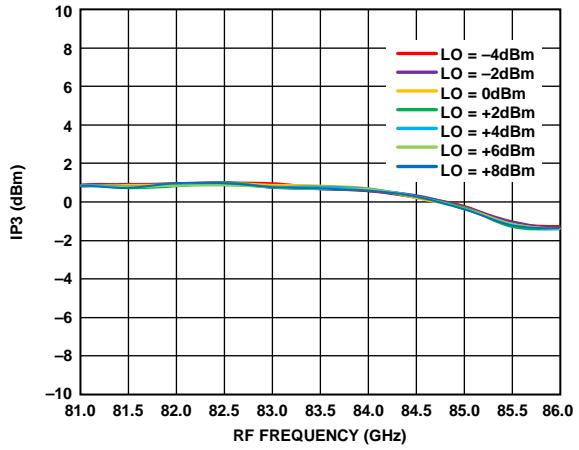


Figure 83. Input IP3 vs. RF Frequency at Various LO Powers,
RFIN = -20 dBm, IF = 2000 MHz, $V_{\text{DLNA}} = 4 \text{ V}$

13141-087

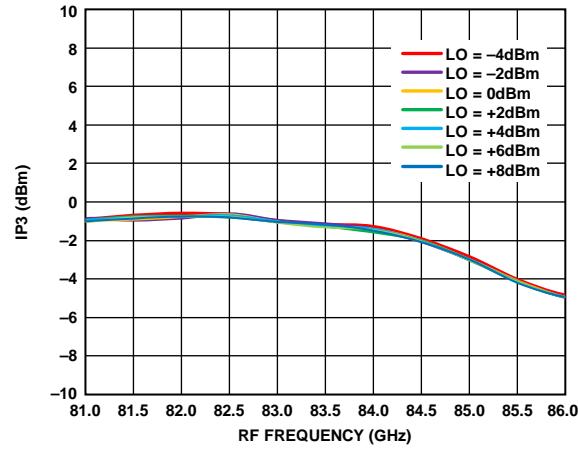


Figure 86. Input IP3 vs. RF Frequency at Various LO Powers,
RFIN = -20 dBm, IF = 2000 MHz, $V_{\text{DLNA}} = 3 \text{ V}$

13141-089

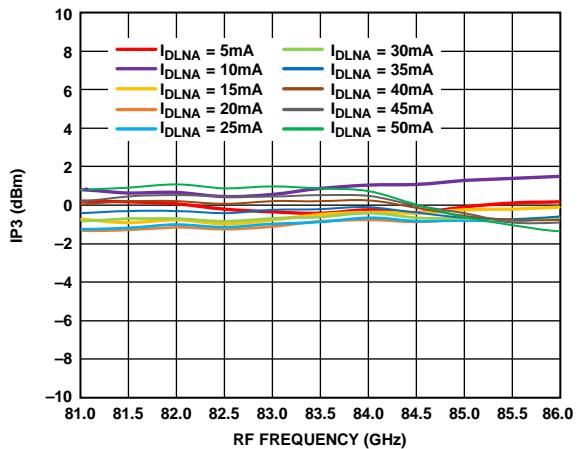


Figure 84. Input IP3 vs. RF Frequency at Various I_{DLNA} Values,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, $V_{\text{DLNA}} = 4 \text{ V}$

13141-088

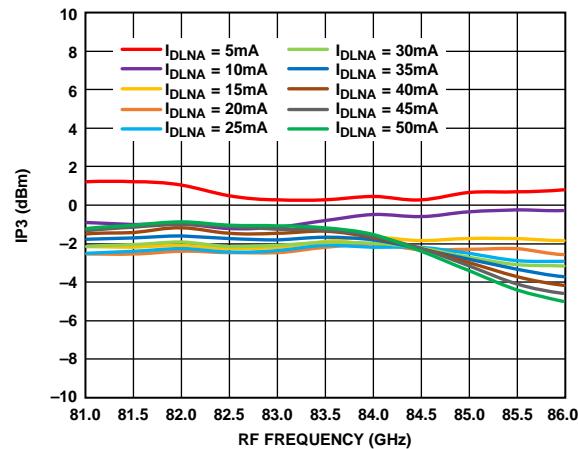
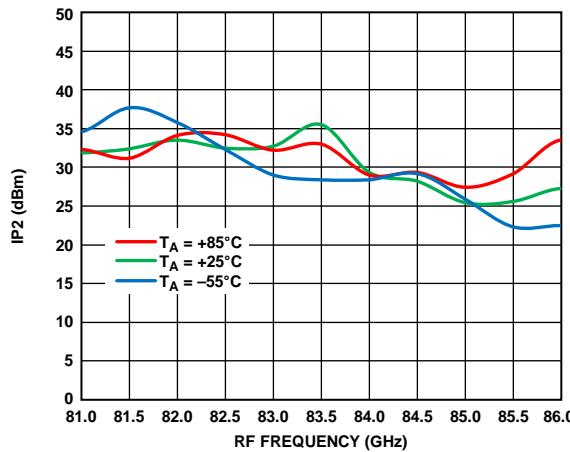
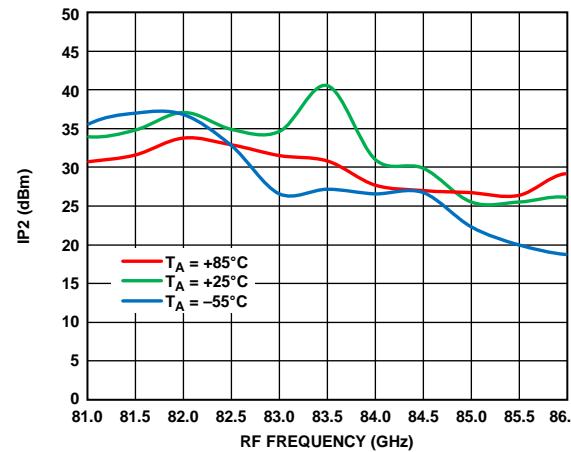


Figure 87. Input IP3 vs. RF Frequency at Various I_{DLNA} Values,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, $V_{\text{DLNA}} = 3 \text{ V}$

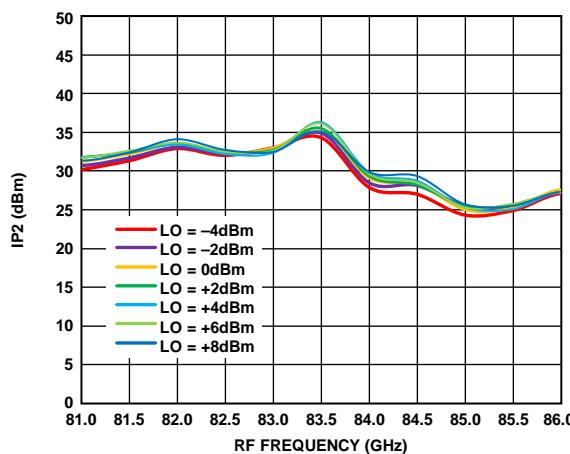
13141-091



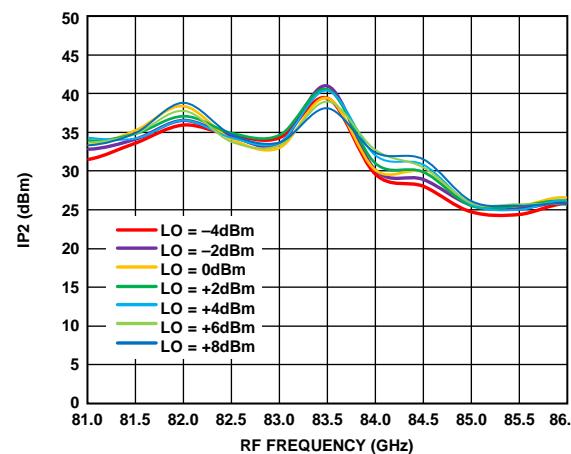
13141-092



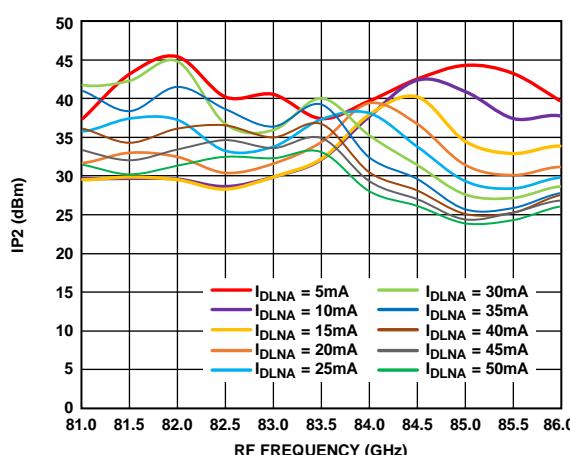
13141-095



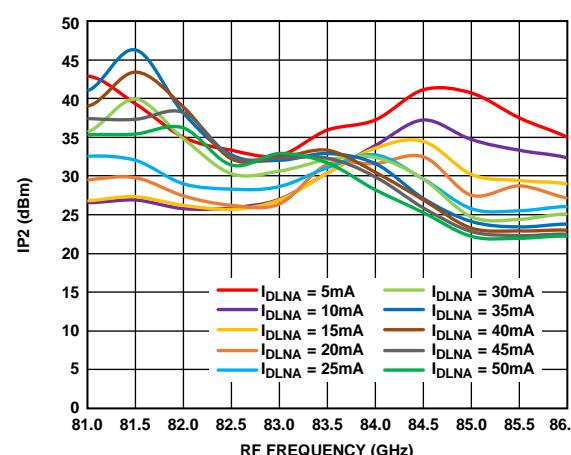
13141-093



13141-096



13141-094



13141-097

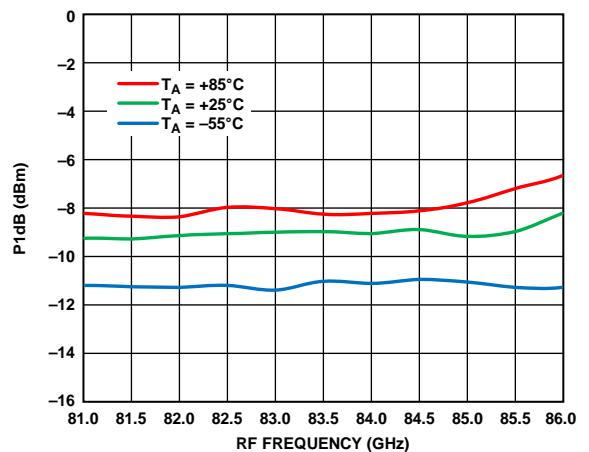


Figure 94. Input P1dB vs. RF Frequency at Various Temperatures,
LO = 2 dBm, IF = 2000 MHz, $V_{DLNA} = 4\text{ V}$

13141-098

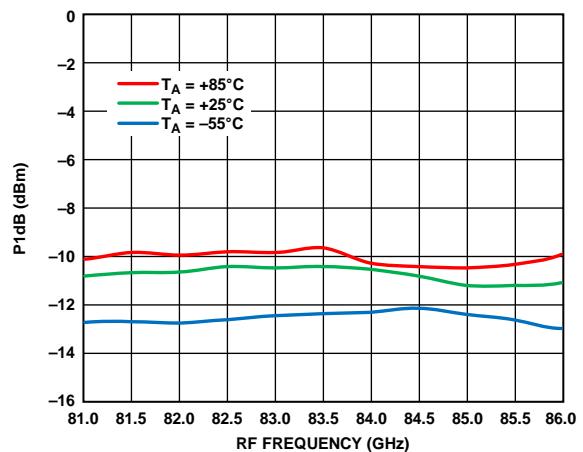


Figure 95. Input P1dB vs. RF Frequency at Various Temperatures,
LO = 2 dBm, IF = 2000 MHz, $V_{DLNA} = 3\text{ V}$

13141-101

NOISE FIGURE PERFORMANCE WITH UPPER SIDEBAND SELECTED

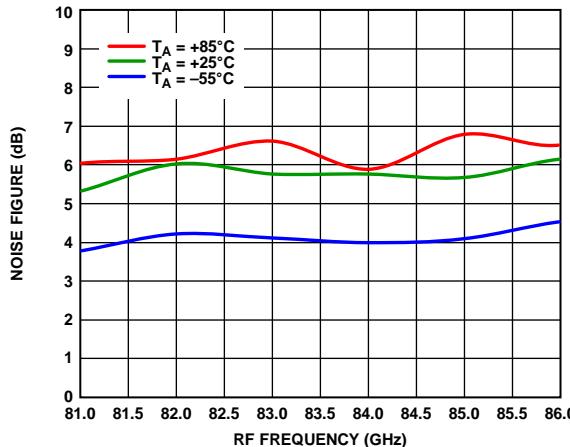


Figure 96. Noise Figure vs. RF Frequency at Various Temperatures,
LO = 2 dBm, IF = 500 MHz, V_{DLNA} = 3 V

13141-105

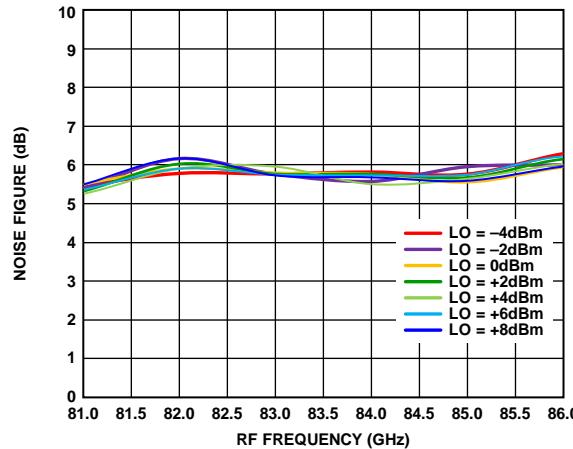


Figure 99. Noise Figure vs. RF Frequency at Various LO Powers,
IF = 500 MHz, V_{DLNA} = 3 V

13141-107

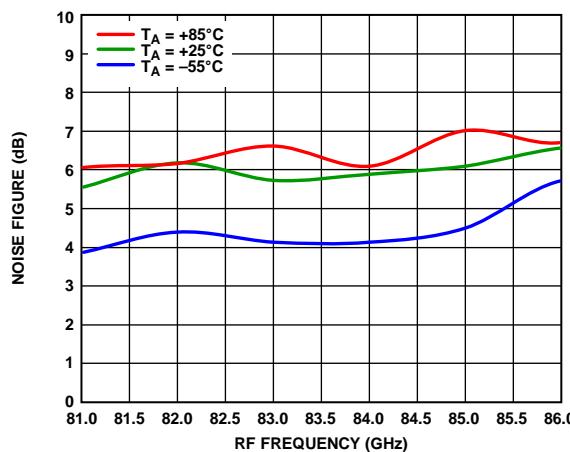


Figure 97. Noise Figure vs. RF Frequency at Various Temperatures,
LO = 2 dBm, IF = 1000 MHz, V_{DLNA} = 3 V

13141-105

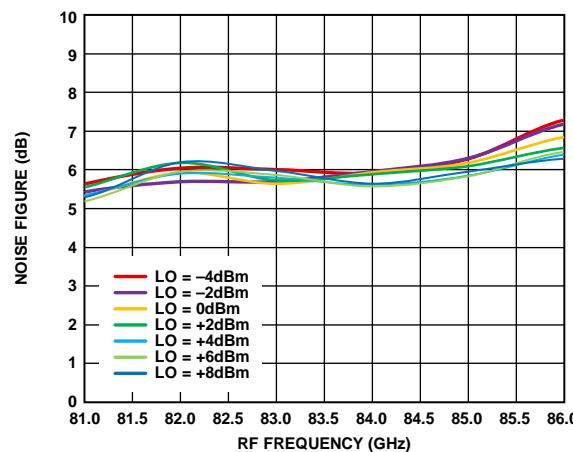


Figure 100. Noise Figure vs. RF Frequency at Various LO Powers,
IF = 1000 MHz, V_{DLNA} = 3 V

13141-108

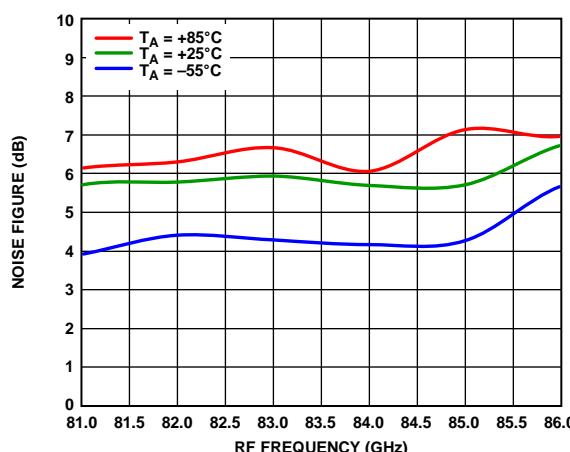


Figure 98. Noise Figure vs. RF Frequency at Various Temperatures,
LO = 2 dBm, IF = 2000 MHz, V_{DLNA} = 3 V

13141-106

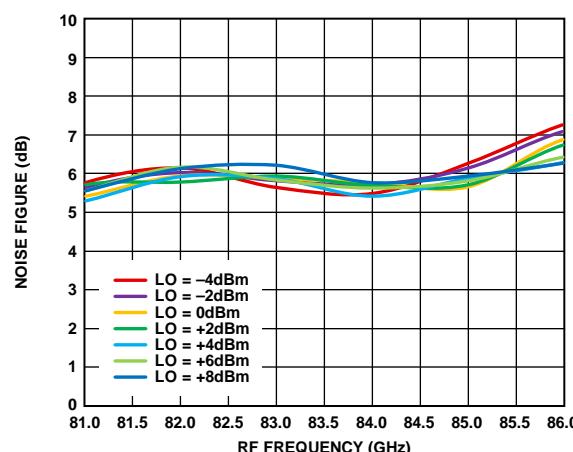


Figure 101. Noise Figure vs. RF Frequency at Various LO Powers,
IF = 2000 MHz, V_{DLNA} = 3 V

13141-108

AMPLITUDE BALANCE PERFORMANCE WITH UPPER SIDEBAND SELECTED

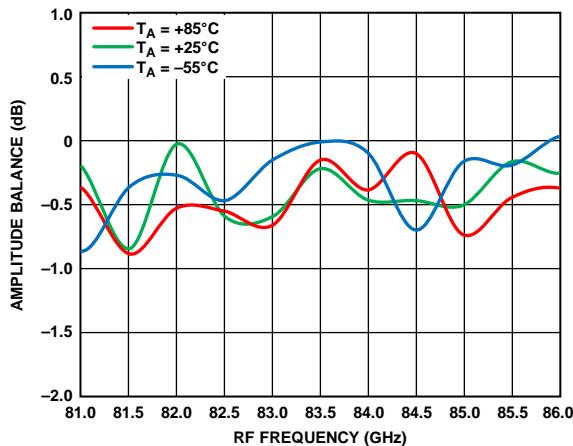


Figure 102. Amplitude Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 500 MHz, $V_{DLNA} = 4\text{ V}$

13141-110

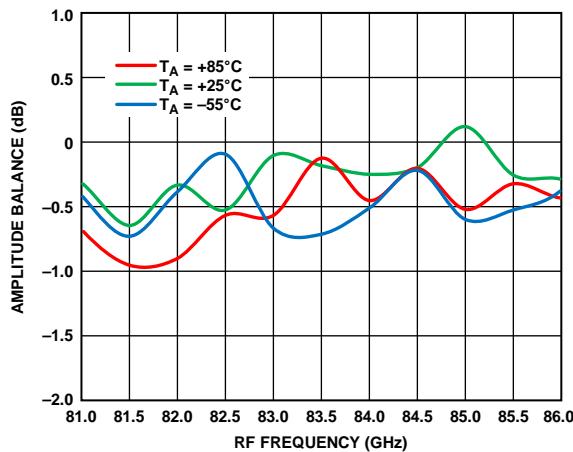


Figure 105. Amplitude Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 500 MHz, $V_{DLNA} = 3\text{ V}$

13141-113

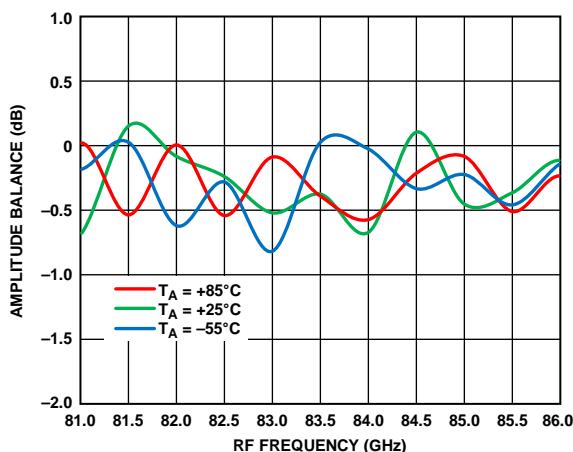


Figure 103. Amplitude Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 1000 MHz, $V_{DLNA} = 4\text{ V}$

13141-111

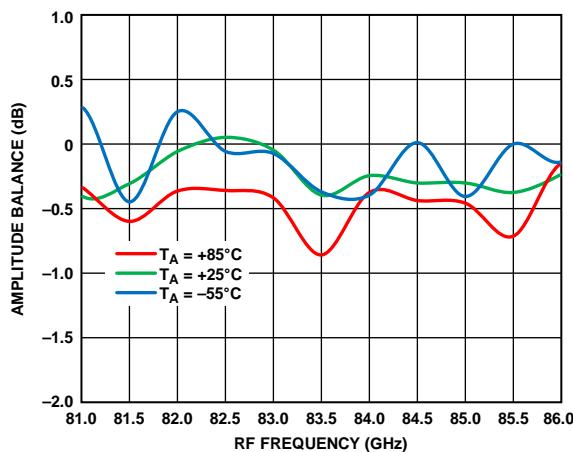


Figure 106. Amplitude Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 1000 MHz, $V_{DLNA} = 3\text{ V}$

13141-114

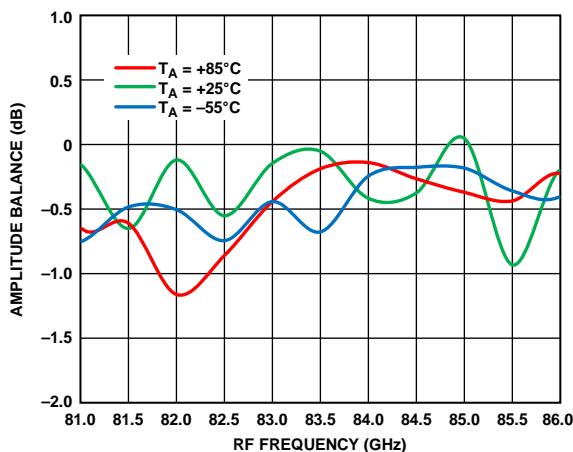


Figure 104. Amplitude Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, $V_{DLNA} = 4\text{ V}$

13141-112

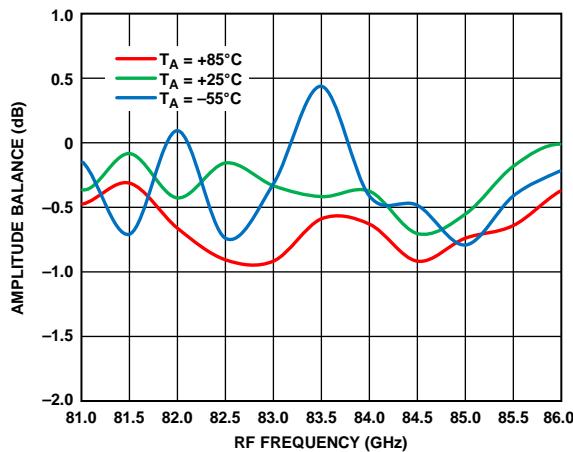
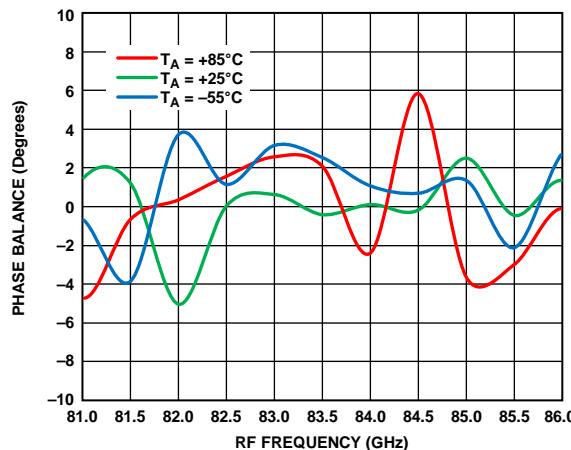


Figure 107. Amplitude Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, $V_{DLNA} = 3\text{ V}$

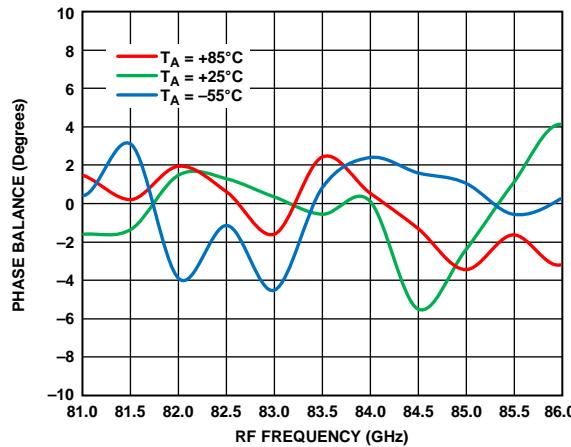
13141-115

PHASE BALANCE PERFORMANCE WITH UPPER SIDEBAND SELECTED



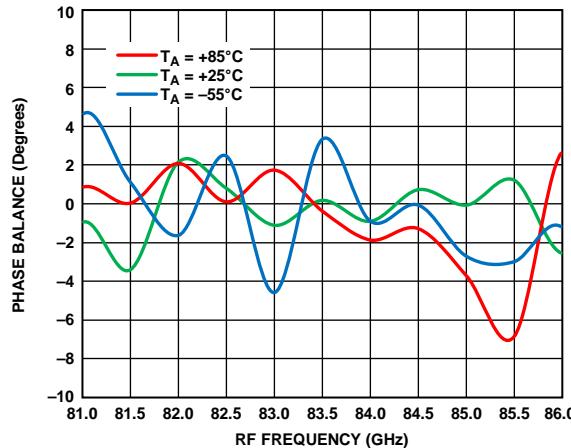
13141-116

Figure 108. Phase Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 500 MHz, V_{DLNA} = 4 V



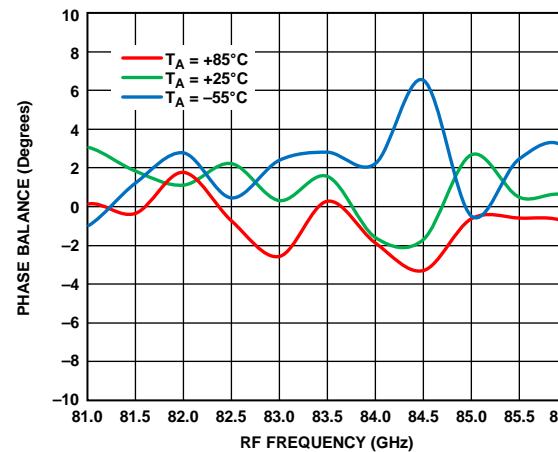
13141-117

Figure 109. Phase Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 1000 MHz, V_{DLNA} = 4 V



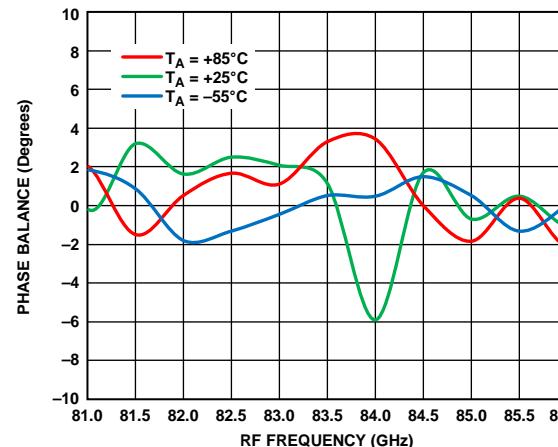
13141-118

Figure 110. Phase Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, V_{DLNA} = 4 V



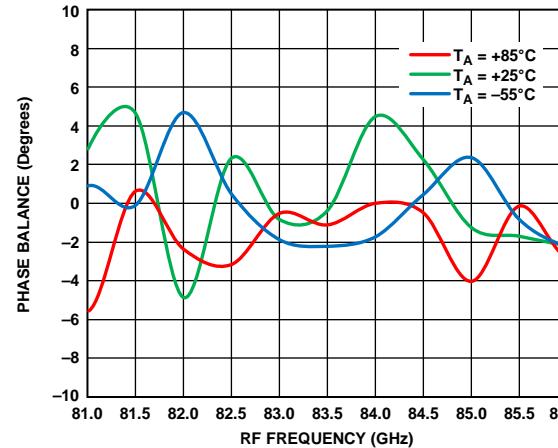
13141-119

Figure 111. Phase Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 500 MHz, V_{DLNA} = 3 V



13141-120

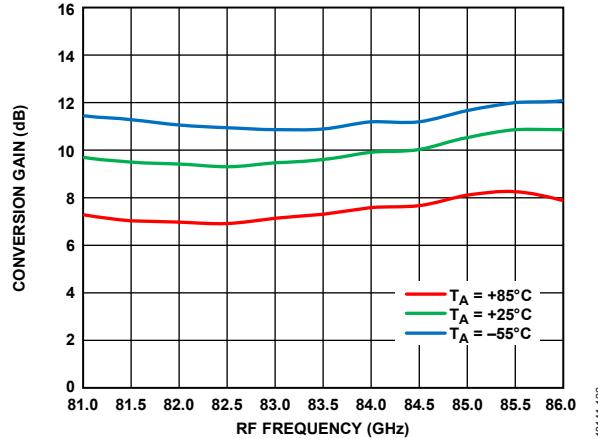
Figure 112. Phase Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 1000 MHz, V_{DLNA} = 3 V



13141-121

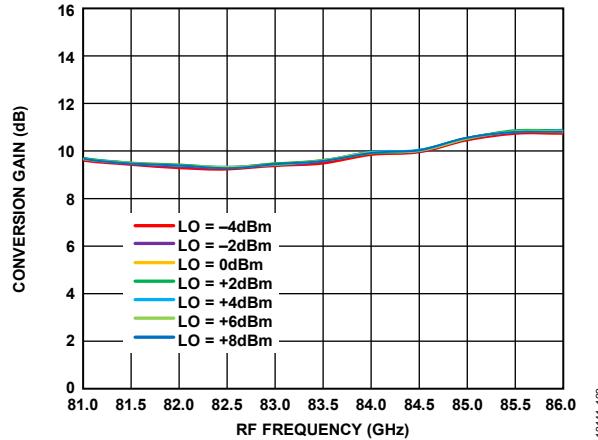
Figure 113. Phase Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, V_{DLNA} = 3 V

LOWER SIDEBAND SELECTED, IF = 500 MHz



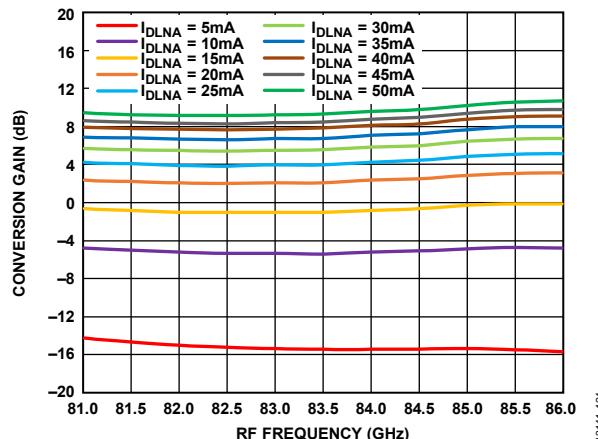
13141-122

Figure 114. Conversion Gain vs. RF Frequency at Various Temperatures, RFIN = -20 dBm, LO = 2 dBm, IF = 500 MHz, V_{DLNA} = 4 V



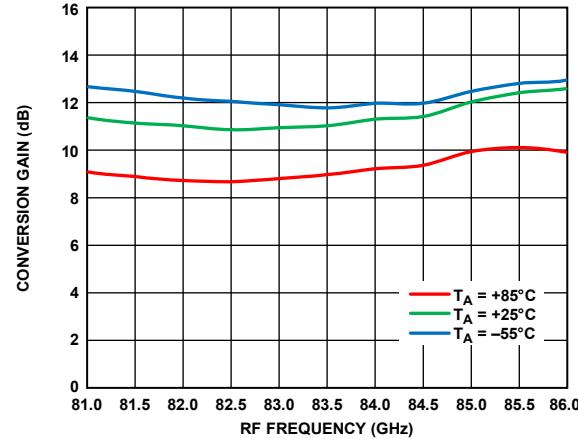
13141-123

Figure 115. Conversion Gain vs. RF Frequency at Various LO Powers, RFIN = -20 dBm, IF = 500 MHz, V_{DLNA} = 4 V



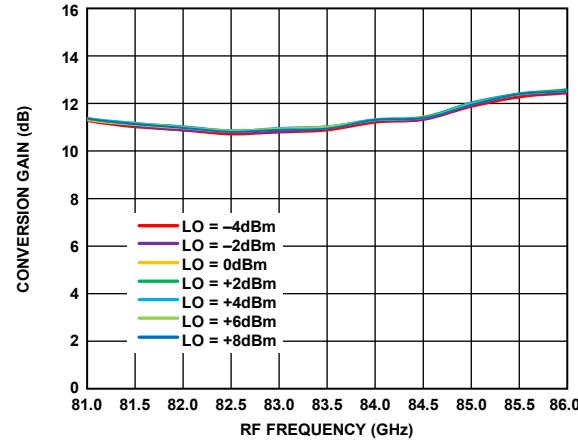
13141-124

Figure 116. Conversion Gain vs. RF Frequency at Various I_{DLNA} Values, RFIN = -20 dBm, LO = 2 dBm, IF = 500 MHz, V_{DLNA} = 4 V



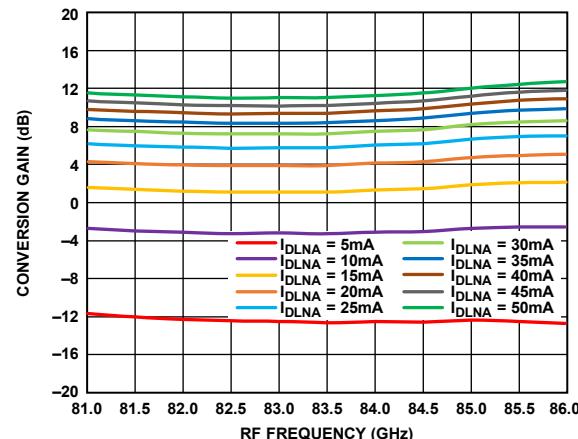
13141-125

Figure 117. Conversion Gain vs. RF Frequency at Various Temperatures, RFIN = -20 dBm, LO = 2 dBm, IF = 500 MHz, V_{DLNA} = 3 V



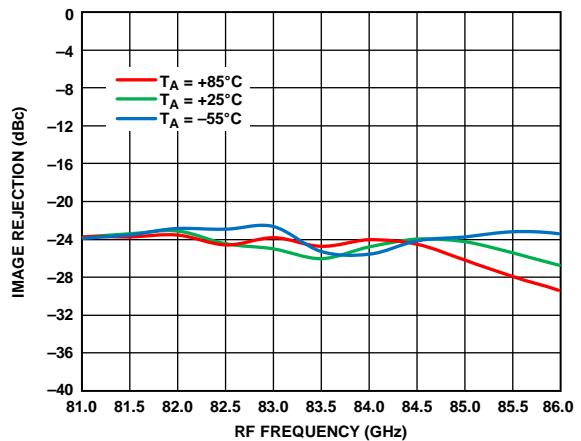
13141-126

Figure 118. Conversion Gain vs. RF Frequency at Various LO Powers, RFIN = -20 dBm, IF = 500 MHz, V_{DLNA} = 3 V

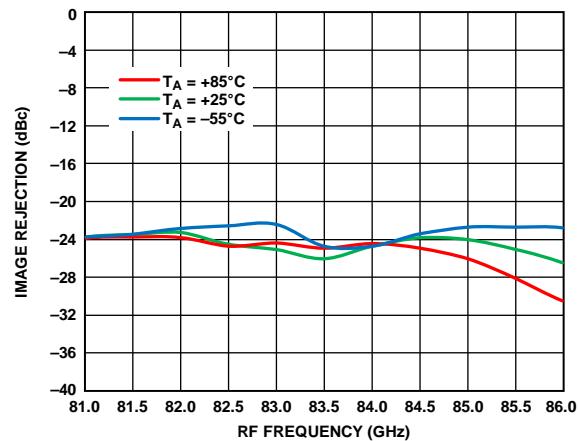


13141-127

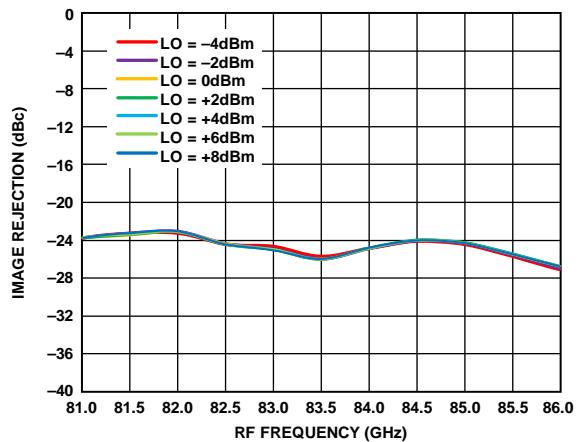
Figure 119. Conversion Gain vs. RF Frequency at Various I_{DLNA} Values, RFIN = -20 dBm, LO = 2 dBm, IF = 500 MHz, V_{DLNA} = 3 V



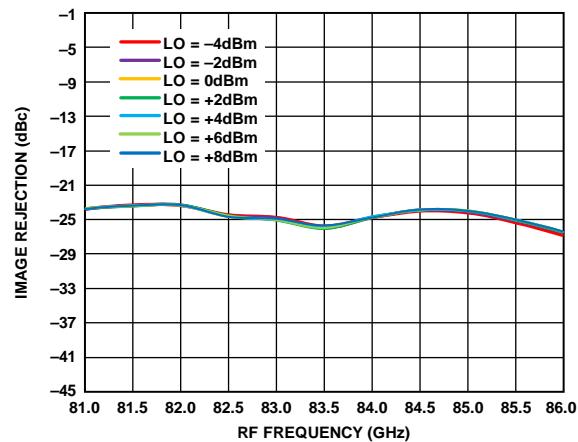
13141-128



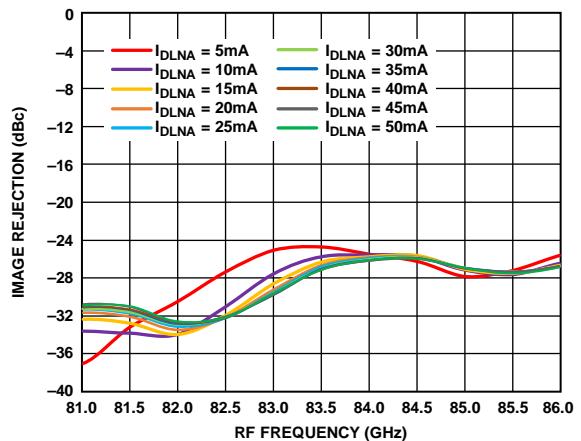
13141-131



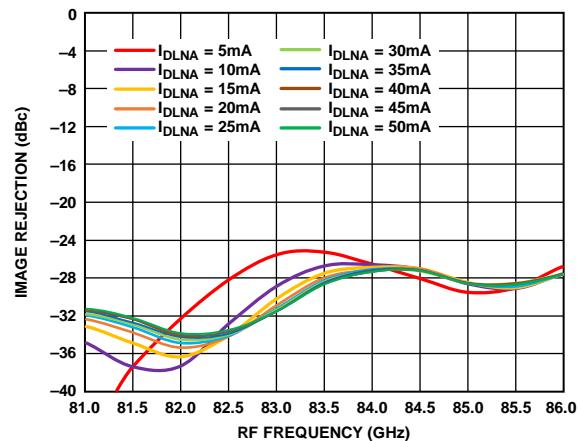
13141-129



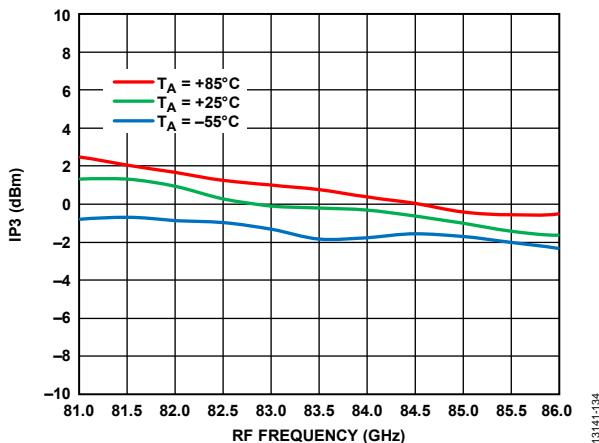
13141-132



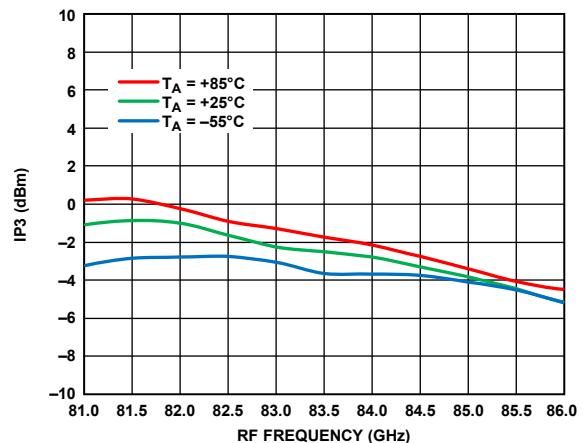
13141-130



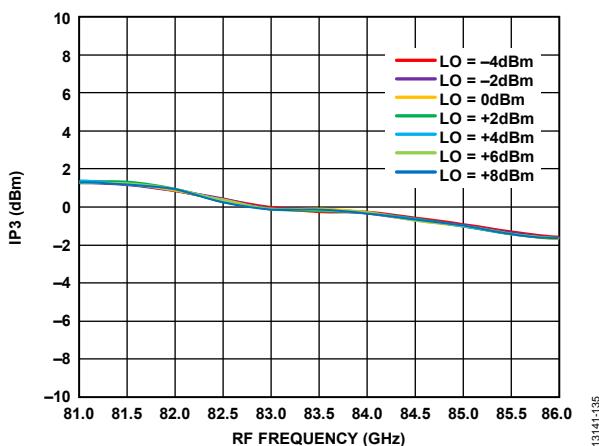
13141-133



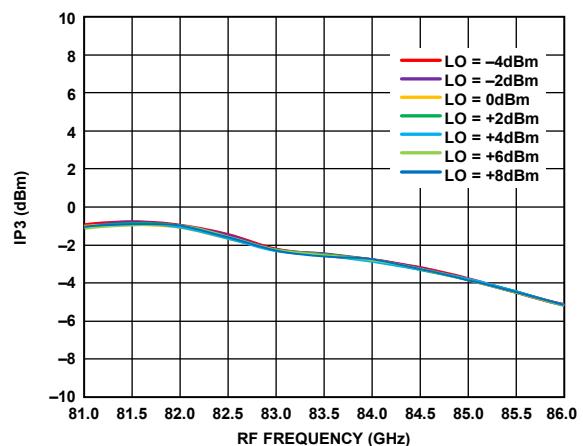
13141-134



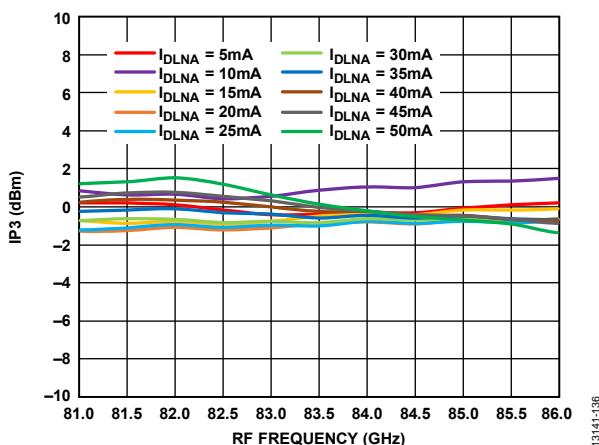
13141-137



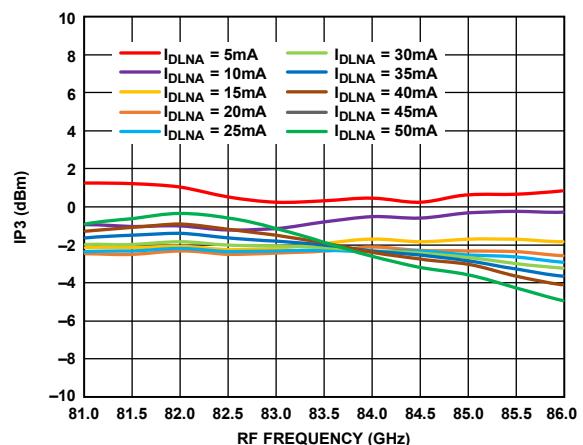
13141-135



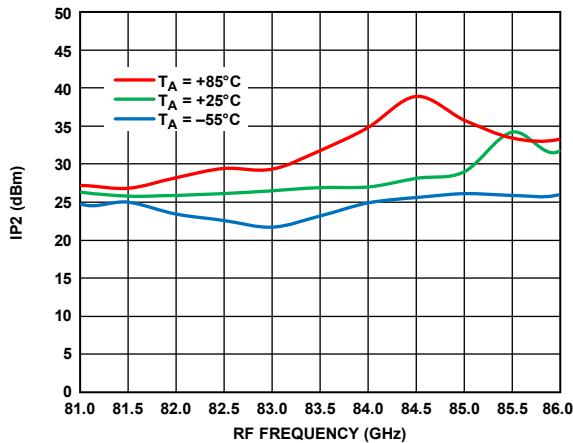
13141-138



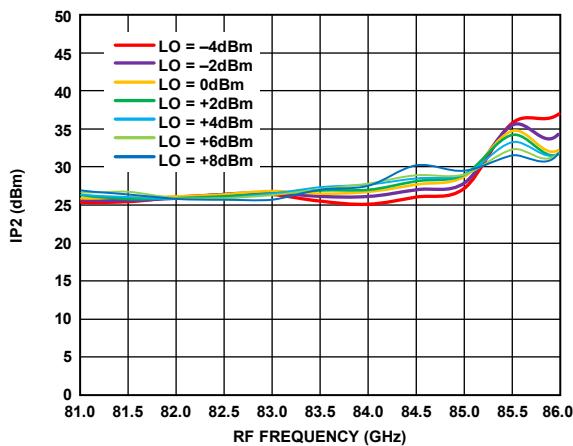
13141-136



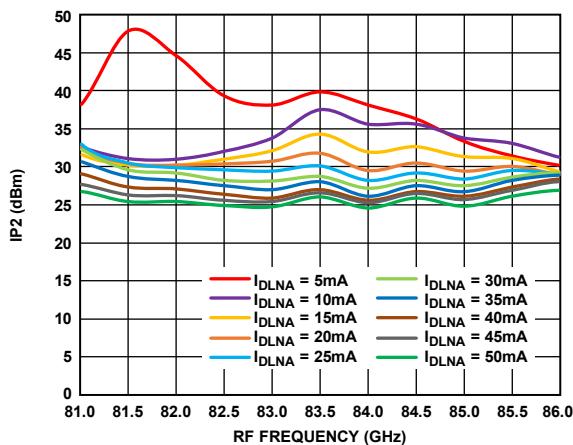
13141-139



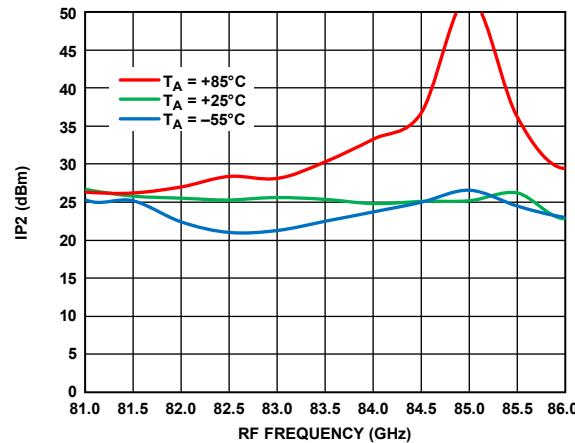
13141-140



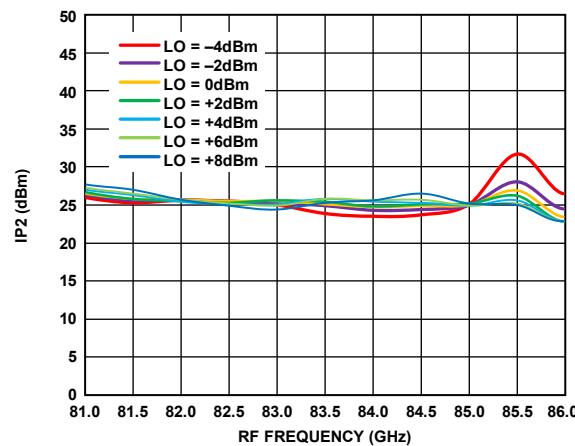
13141-141



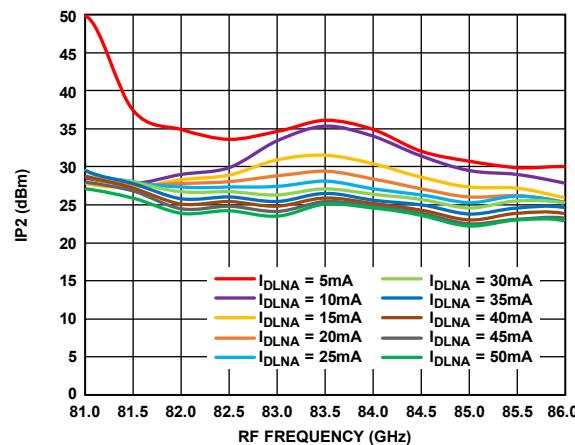
13141-142



13141-143



13141-144



13141-145

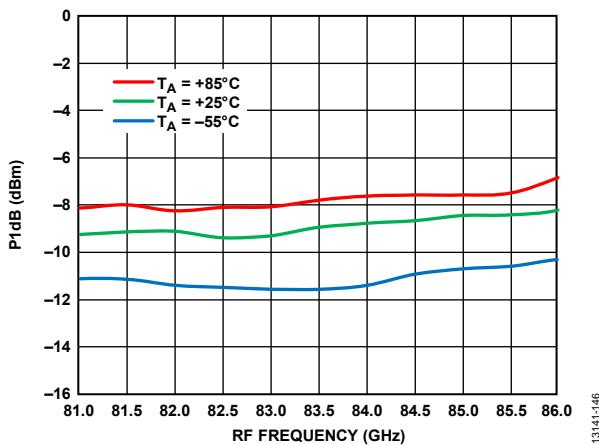


Figure 138. Input P1dB vs. RF Frequency at Various Temperatures,
LO = 2 dBm, IF = 500 MHz, $V_{DLNA} = 4\text{ V}$

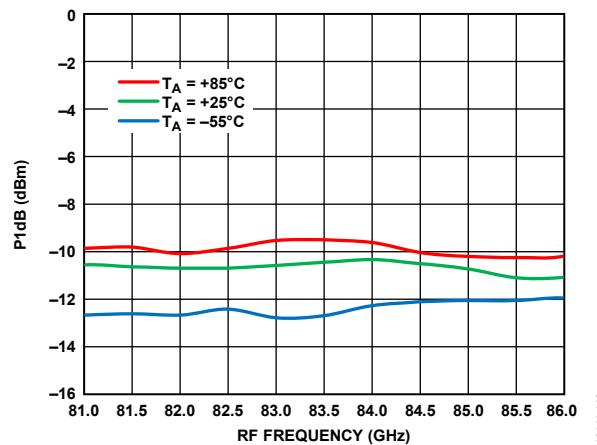


Figure 139. Input P1dB vs. RF Frequency at Various Temperatures,
LO = 2 dBm, IF = 500 MHz, $V_{DLNA} = 3\text{ V}$

LOWER SIDEBAND SELECTED, IF = 1000 MHz

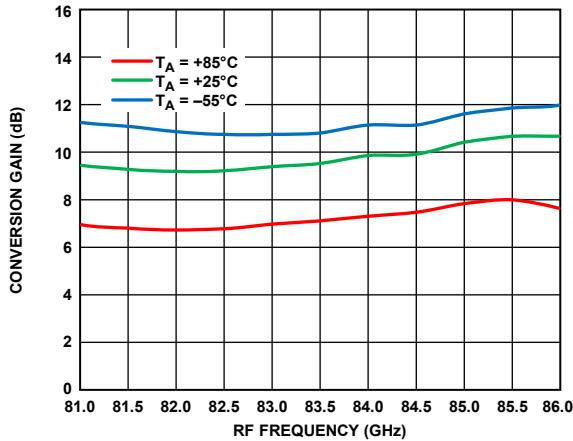


Figure 140. Conversion Gain vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm , LO = 2 dBm , IF = 1000 MHz , $V_{\text{DLNA}} = 4 \text{ V}$

13141-152

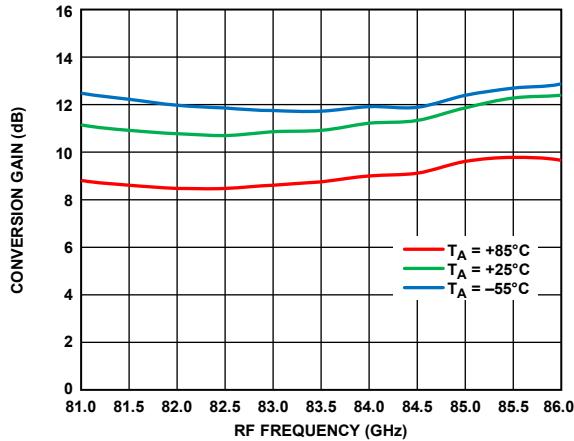


Figure 143. Conversion Gain vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm , LO = 2 dBm , IF = 1000 MHz , $V_{\text{DLNA}} = 3 \text{ V}$

13141-153

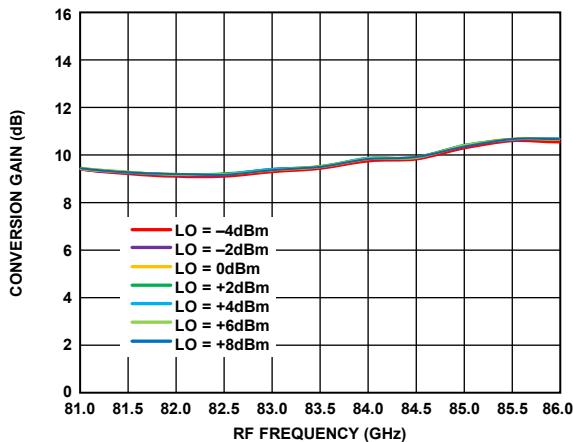


Figure 141. Conversion Gain vs. RF Frequency at Various LO Powers,
RFIN = -20 dBm , IF = 1000 MHz , $V_{\text{DLNA}} = 4 \text{ V}$

13141-153

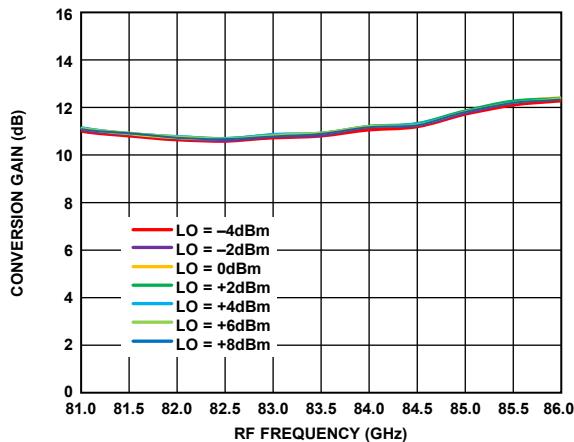


Figure 144. Conversion Gain vs. RF Frequency at Various LO Powers,
RFIN = -20 dBm , IF = 1000 MHz , $V_{\text{DLNA}} = 3 \text{ V}$

13141-156

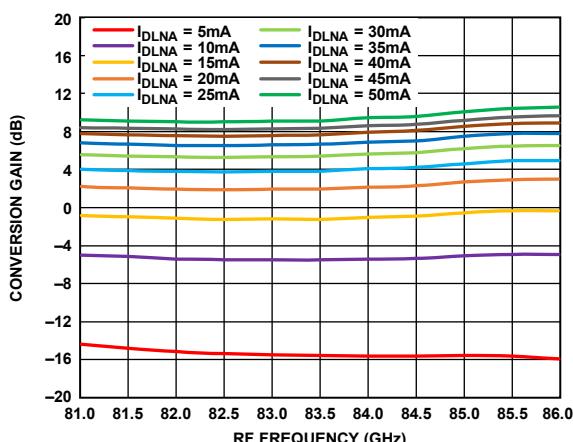


Figure 142. Conversion Gain vs. RF Frequency at Various I_{DLNA} Values,
RFIN = -20 dBm , LO = 2 dBm , IF = 1000 MHz , $V_{\text{DLNA}} = 4 \text{ V}$

13141-154

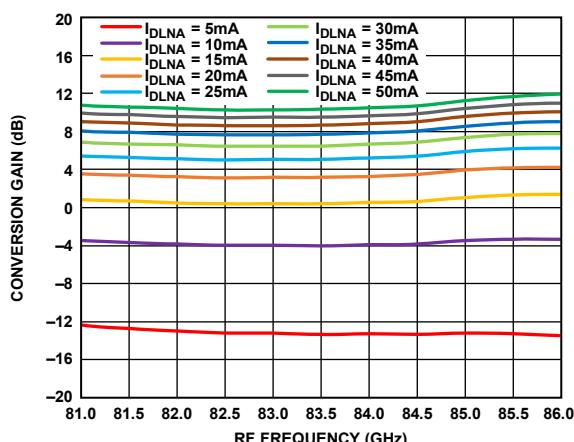


Figure 145. Conversion Gain vs. RF Frequency at Various I_{DLNA} Values,
RFIN = -20 dBm , LO = 2 dBm , IF = 1000 MHz , $V_{\text{DLNA}} = 3 \text{ V}$

13141-157

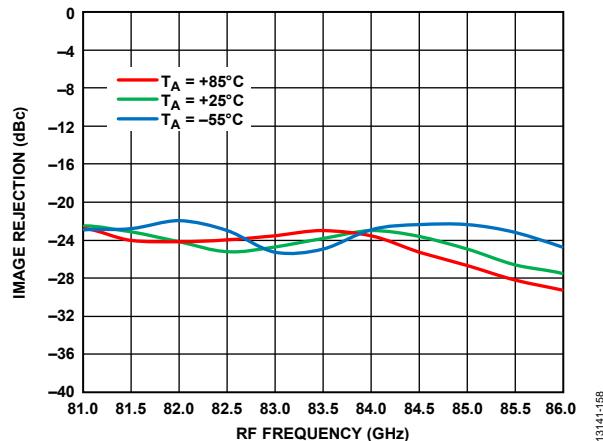


Figure 146. Image Rejection vs. RF Frequency at Various Temperatures, RFIN = -20 dBm, LO = 2 dBm, IF = 1000 MHz, V_{DLNA} = 4 V

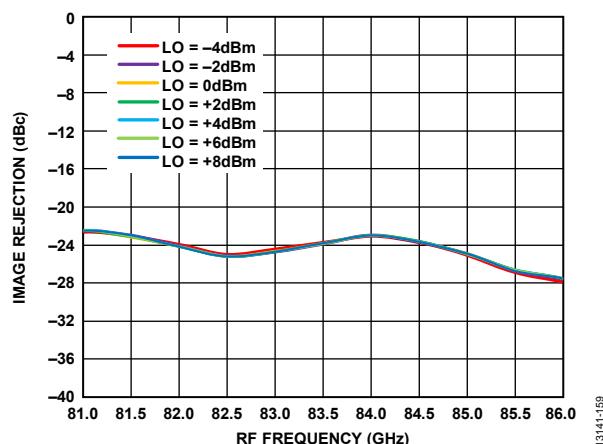


Figure 147. Image Rejection vs. RF Frequency at Various LO Powers, RFIN = -20 dBm, IF = 1000 MHz, V_{DLNA} = 4 V

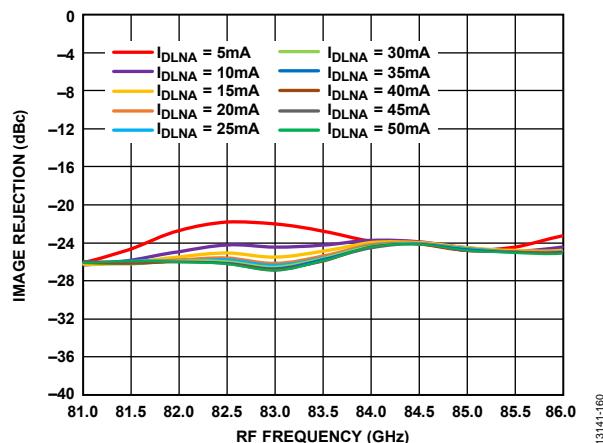


Figure 148. Image Rejection vs. RF Frequency at Various I_{DLNA} Values, RFIN = -20 dBm, LO = 2 dBm, IF = 1000 MHz, V_{DLNA} = 4 V

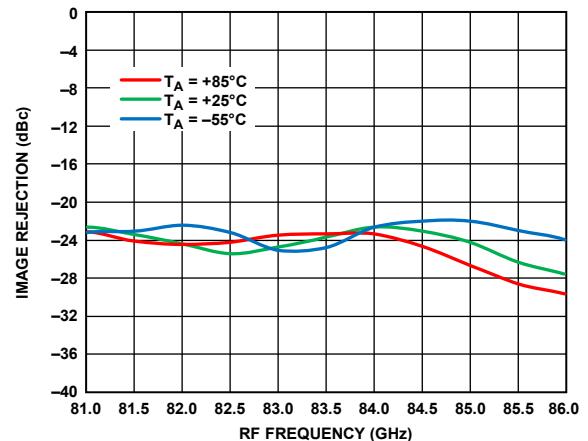


Figure 149. Image Rejection vs. RF Frequency at Various Temperatures, RFIN = -20 dBm, LO = 2 dBm, IF = 1000 MHz, V_{DLNA} = 3 V

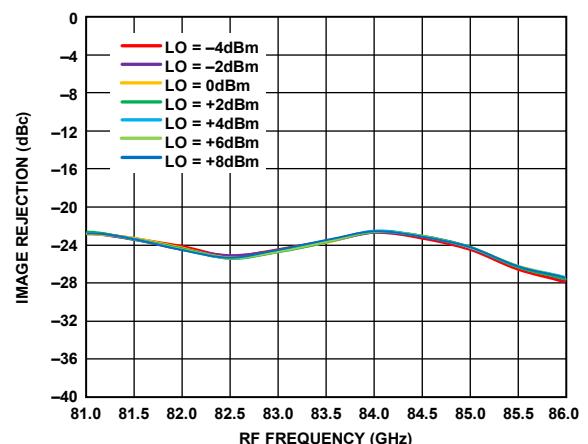


Figure 150. Image Rejection vs. RF Frequency at Various LO Powers, RFIN = -20 dBm, IF = 1000 MHz, V_{DLNA} = 3 V

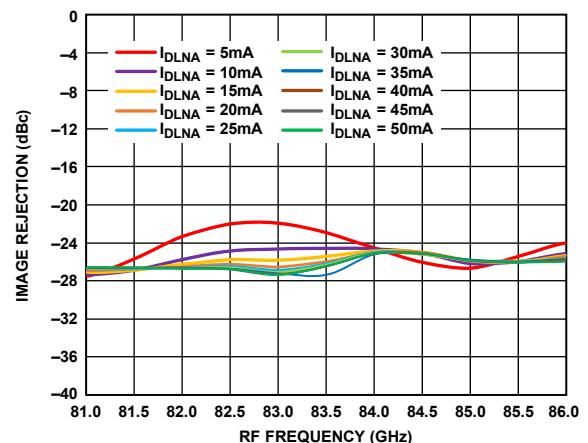
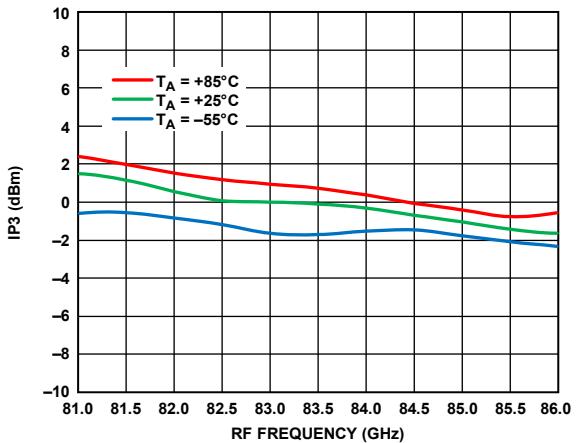
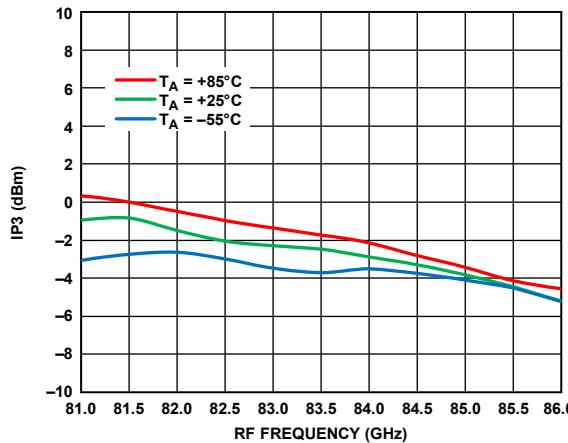


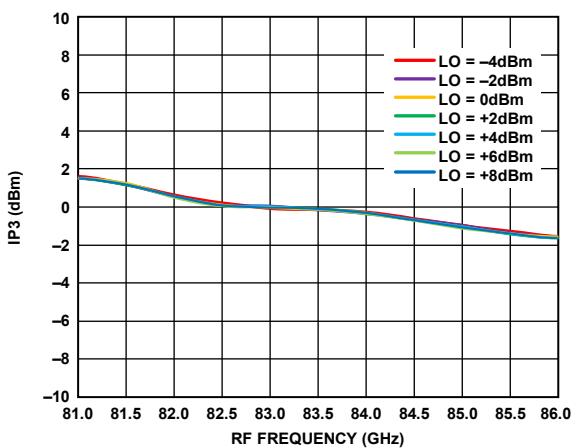
Figure 151. Image Rejection vs. RF Frequency at Various I_{DLNA} Values, RFIN = -20 dBm, LO = 2 dBm, IF = 1000 MHz, V_{DLNA} = 3 V



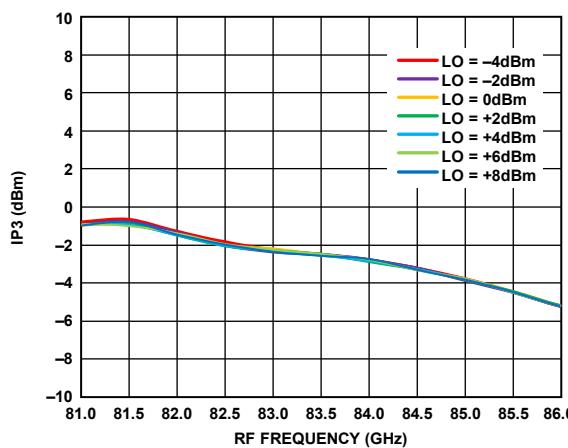
13141-164



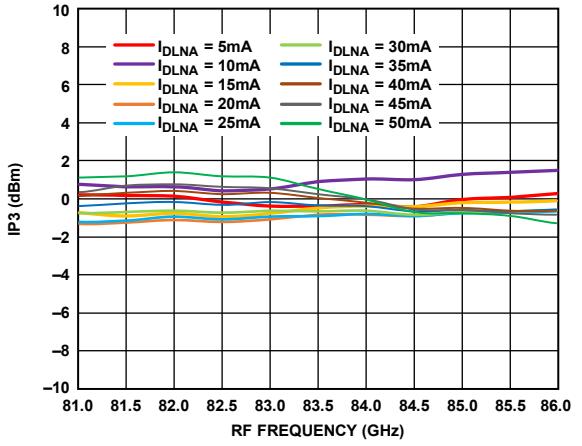
13141-167



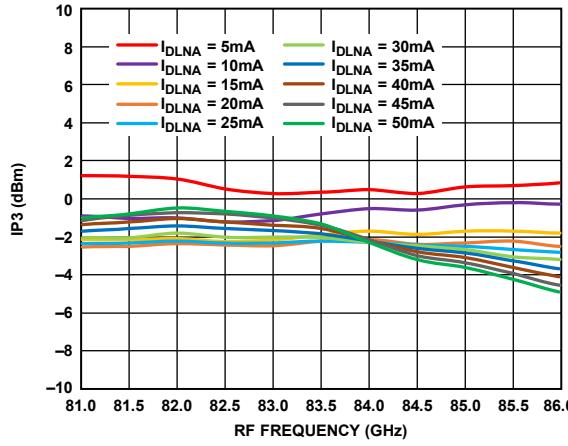
13141-165



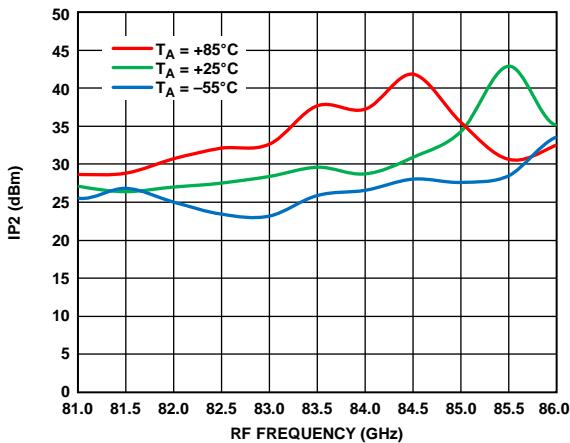
13141-168



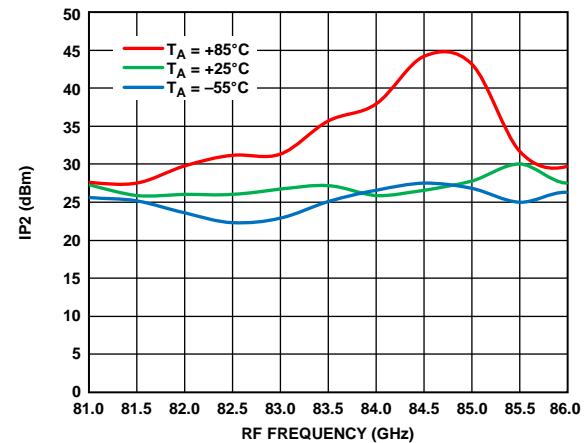
13141-166



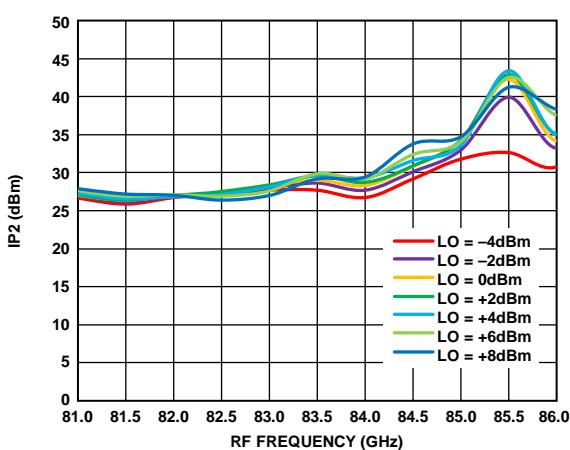
13141-169



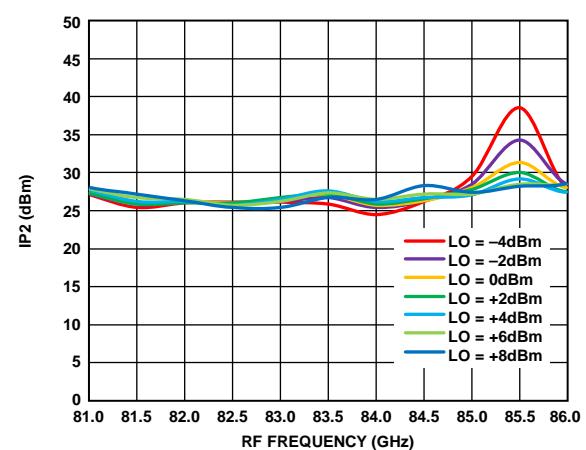
13141-170



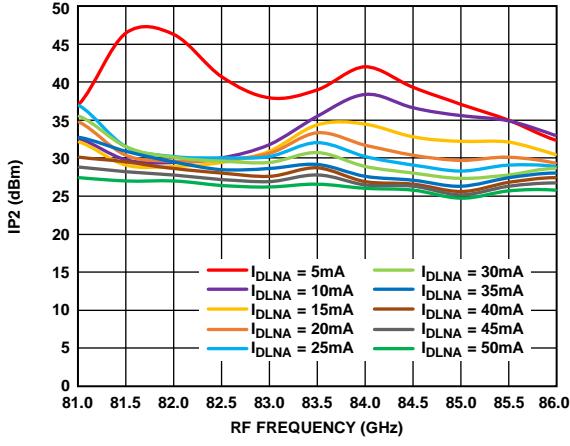
13141-173



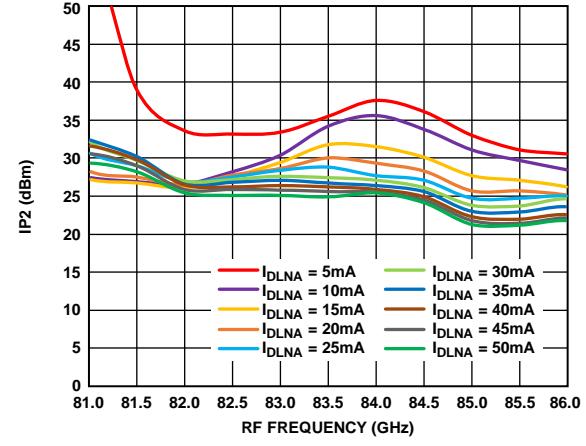
13141-171



13141-174



13141-172



13141-175

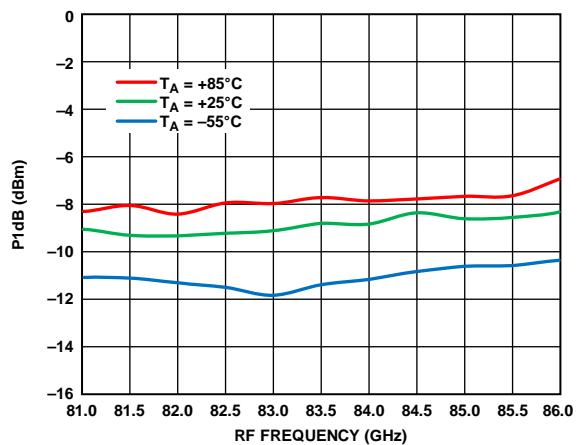


Figure 164. Input P_{1dB} vs. RF Frequency at Various Temperatures,
LO = 2 dBm, IF = 1000 MHz, $V_{DLNA} = 4\text{ V}$

13141-176

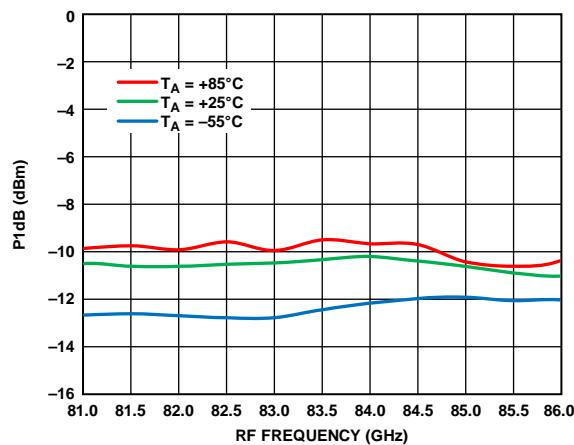


Figure 165. Input P_{1dB} vs. RF Frequency at Various Temperatures,
LO = 2 dBm, IF = 1000 MHz, $V_{DLNA} = 3\text{ V}$

13141-179

LOWER SIDEBAND SELECTED, IF = 2000 MHz

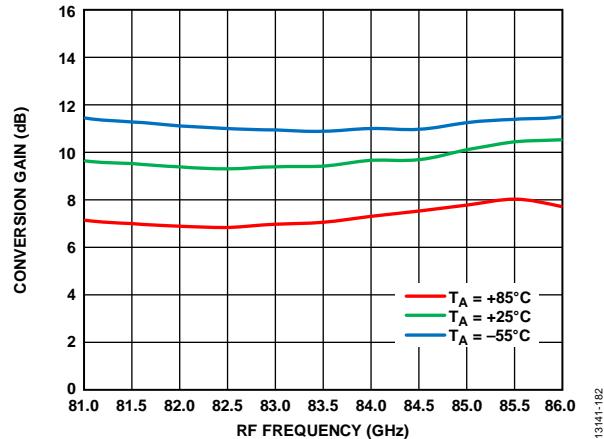


Figure 166. Conversion Gain vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, $V_{DLNA} = 4\text{ V}$

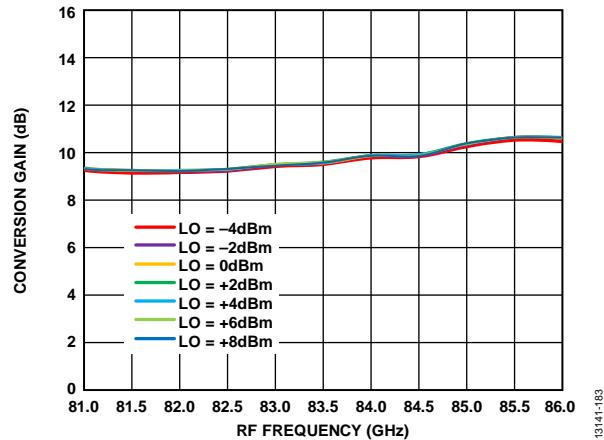


Figure 167. Conversion Gain vs. RF Frequency at Various LO Powers,
RFIN = -20 dBm, IF = 2000 MHz, $V_{DLNA} = 4\text{ V}$

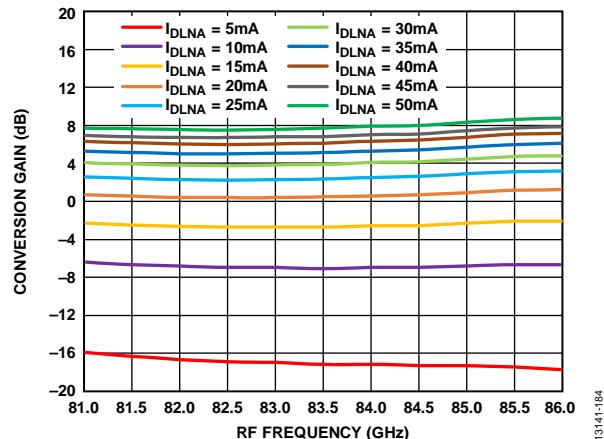


Figure 168. Conversion Gain vs. RF Frequency at Various I_{DLNA} Values,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, $V_{DLNA} = 4\text{ V}$

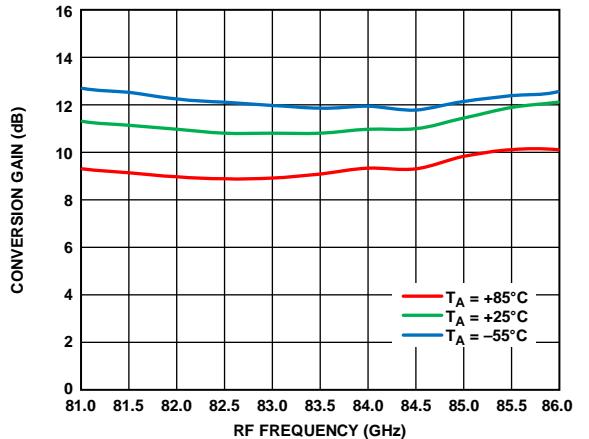


Figure 169. Conversion Gain vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, $V_{DLNA} = 3\text{ V}$

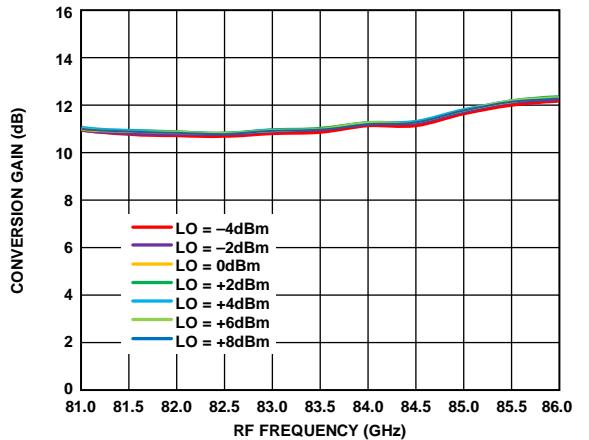


Figure 170. Conversion Gain vs. RF Frequency at Various LO Powers,
RFIN = -20 dBm, IF = 2000 MHz, $V_{DLNA} = 3\text{ V}$

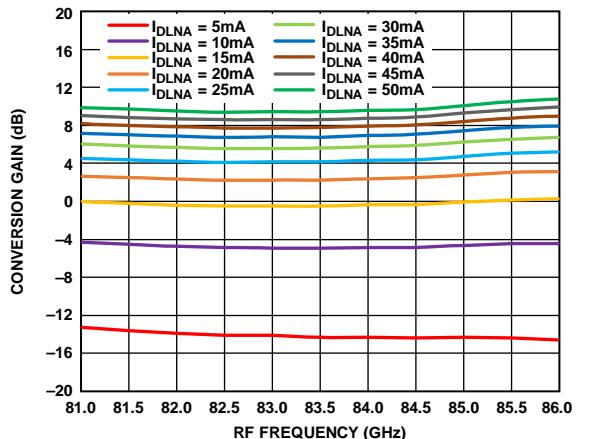
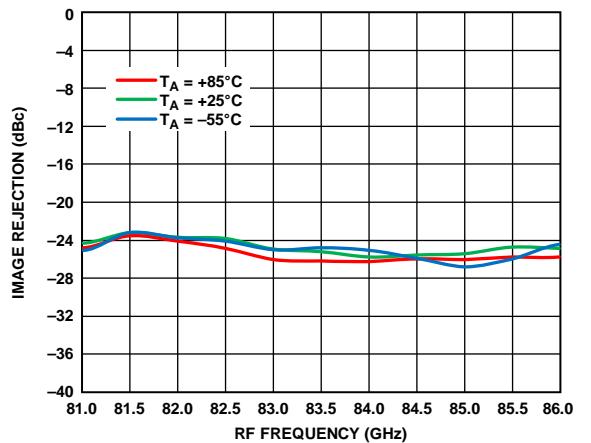
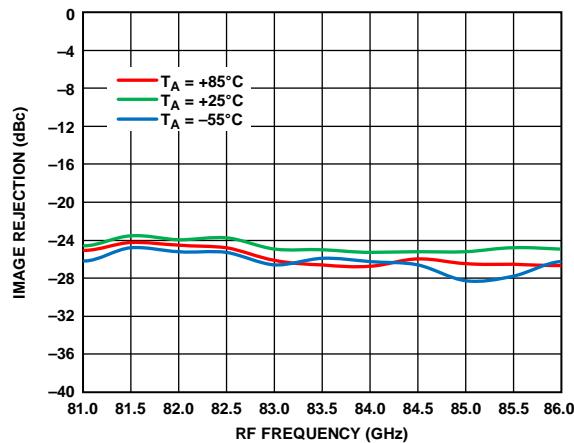


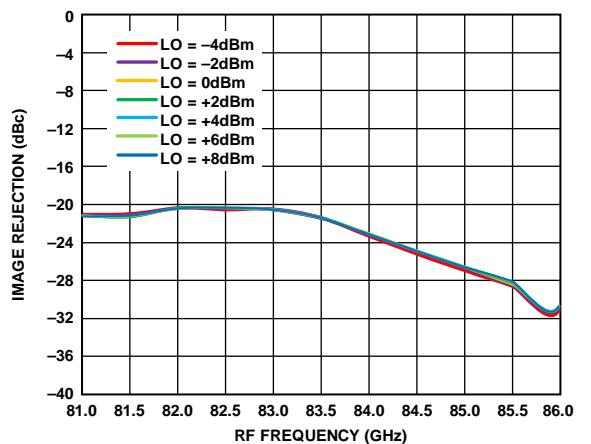
Figure 171. Conversion Gain vs. RF Frequency at Various I_{DLNA} Values,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, $V_{DLNA} = 3\text{ V}$



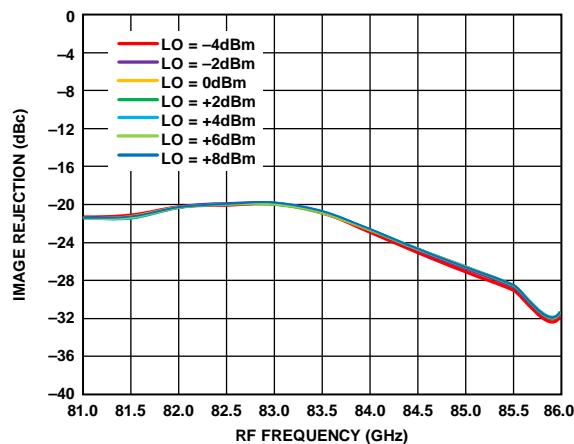
13141-188



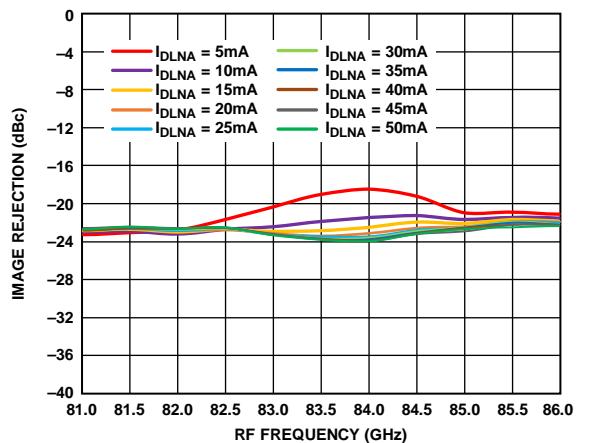
13141-191



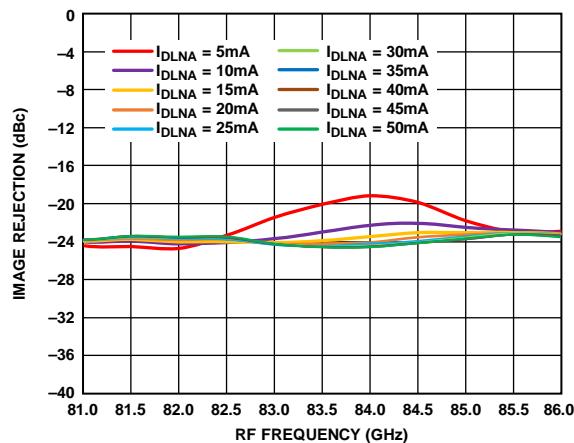
13141-189



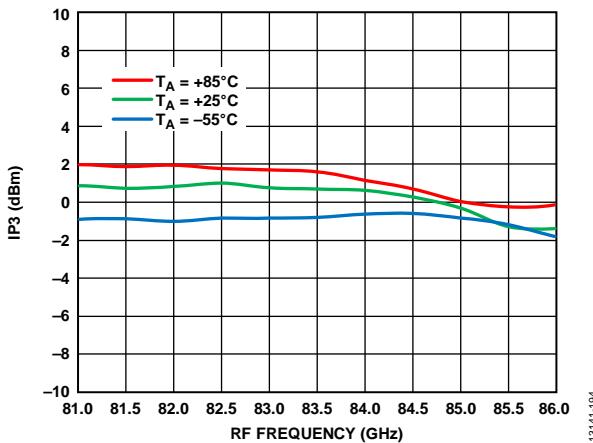
13141-192



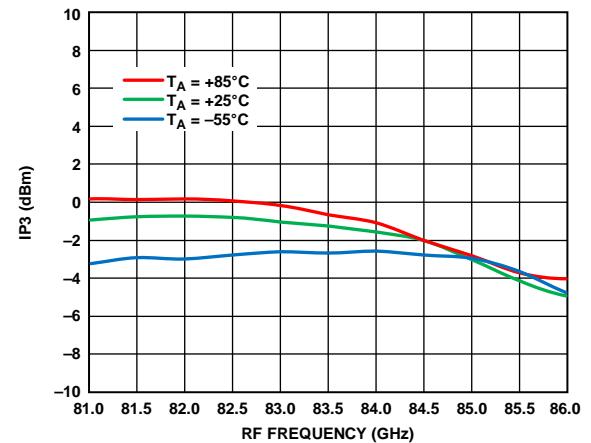
13141-190



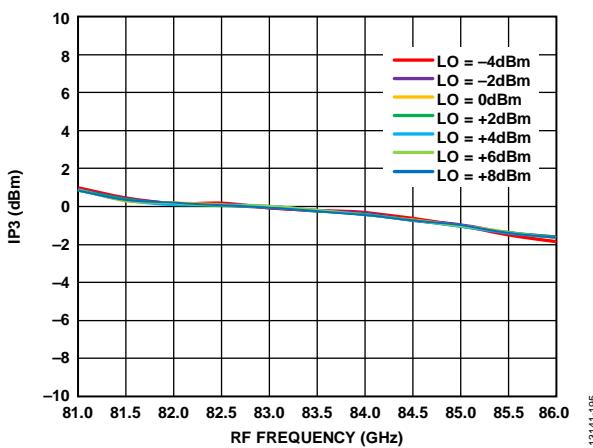
13141-193



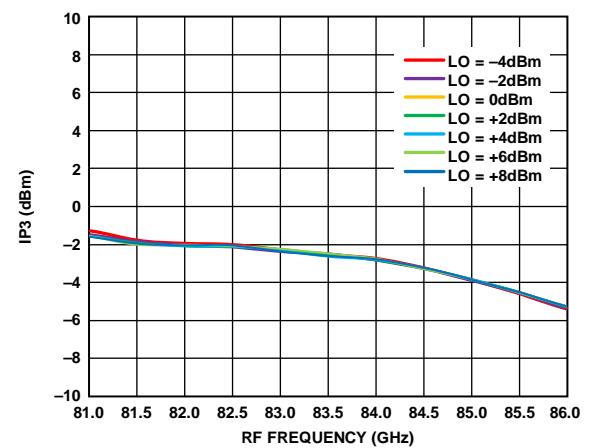
13141-194



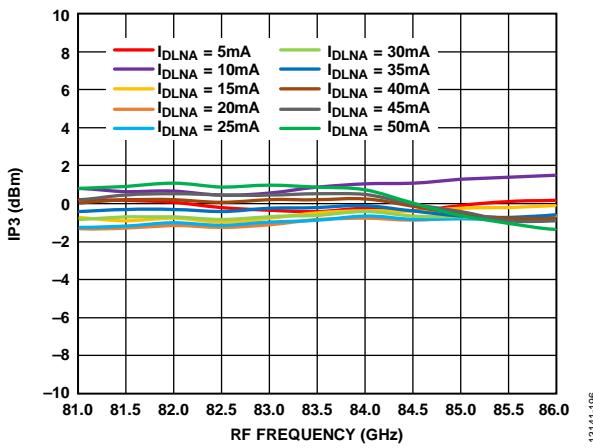
13141-197



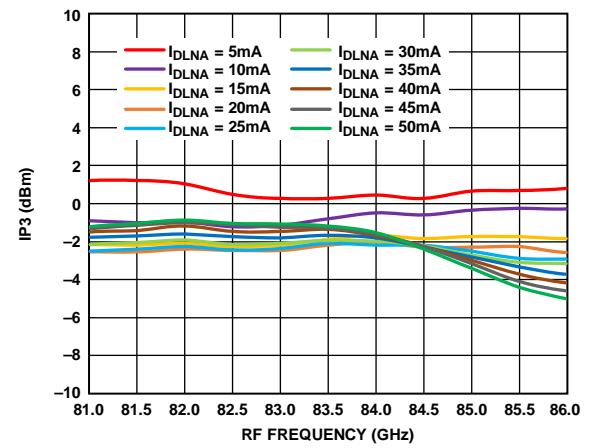
13141-195



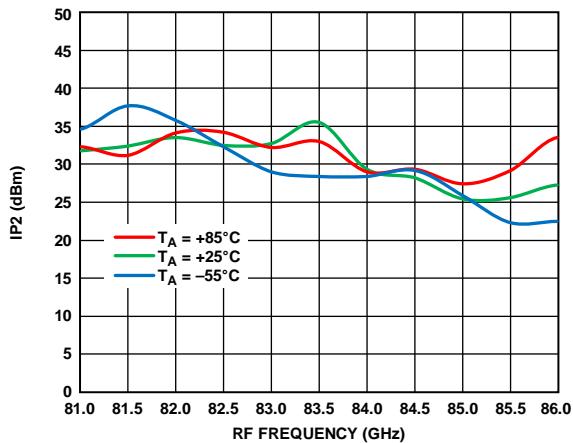
13141-198



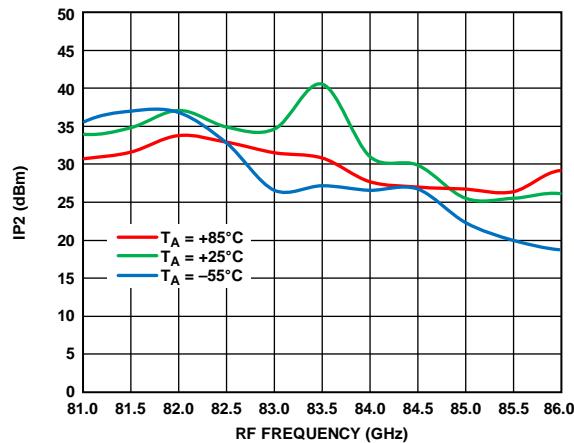
13141-196



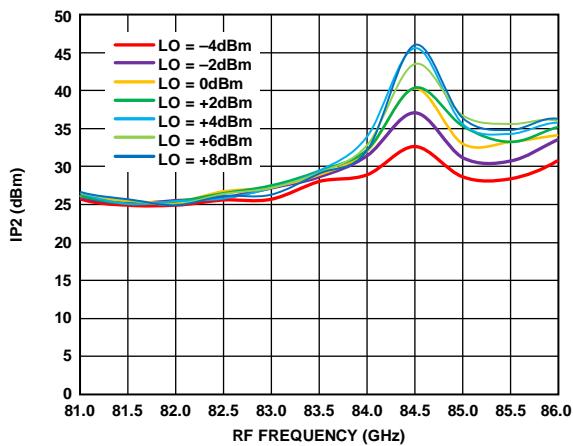
13141-199



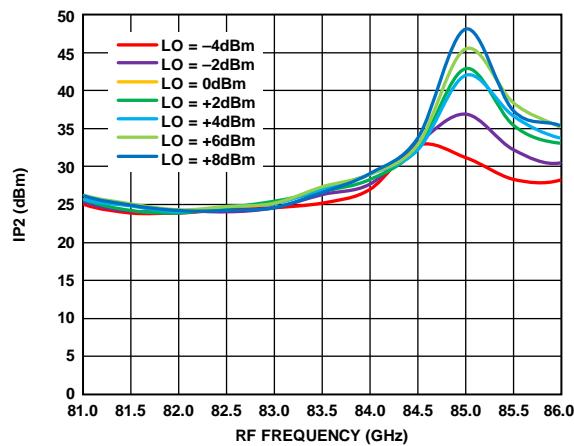
13141-200



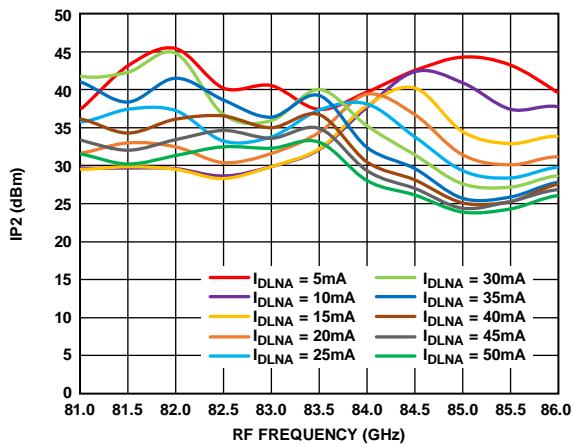
13141-203



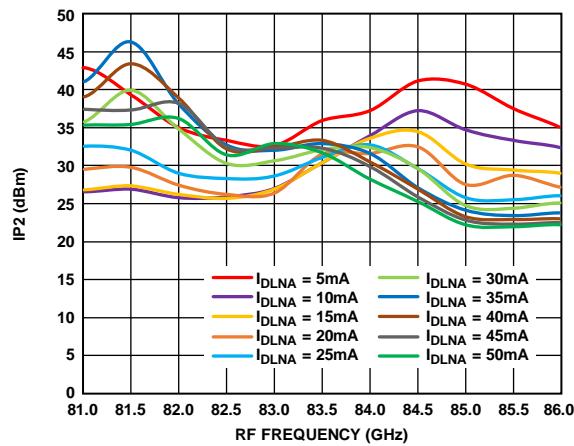
13141-201



13141-204



13141-202



13141-205

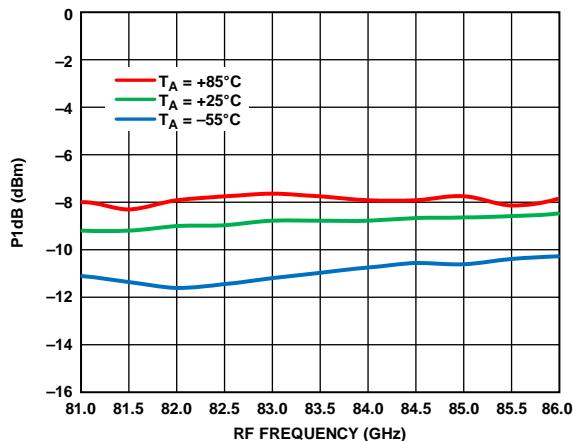


Figure 190. Input P1dB vs. RF Frequency at Various Temperatures,
LO = 2 dBm, IF = 2000 MHz, $V_{DLNA} = 4\text{ V}$

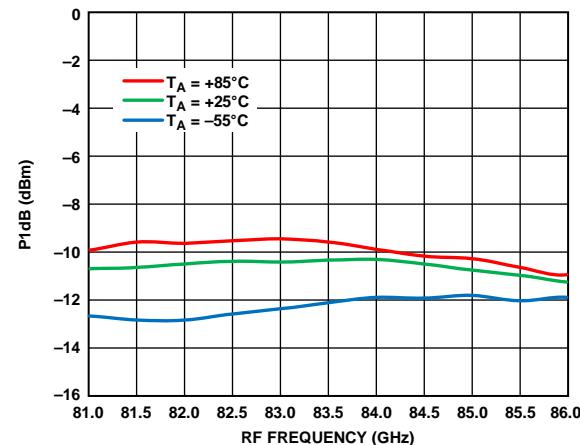


Figure 191. Input P1dB vs. RF Frequency at Various Temperatures,
LO = 2 dBm, IF = 2000 MHz, $V_{DLNA} = 3\text{ V}$

NOISE FIGURE PERFORMANCE WITH LOWER SIDEBAND SELECTED

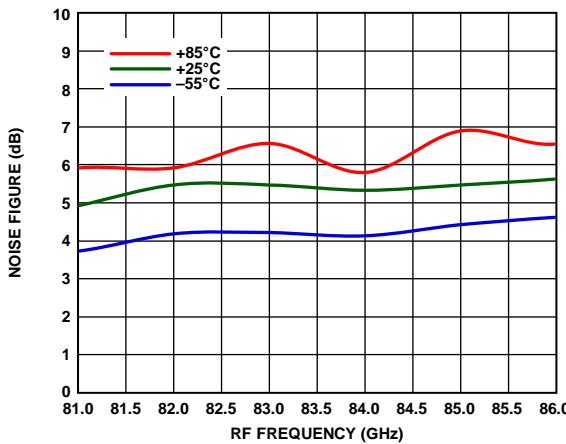


Figure 192. Noise Figure vs. RF Frequency at Various Temperatures,
LO = 2 dBm, IF = 500 MHz, $V_{DLNA} = 3$ V

13141-212

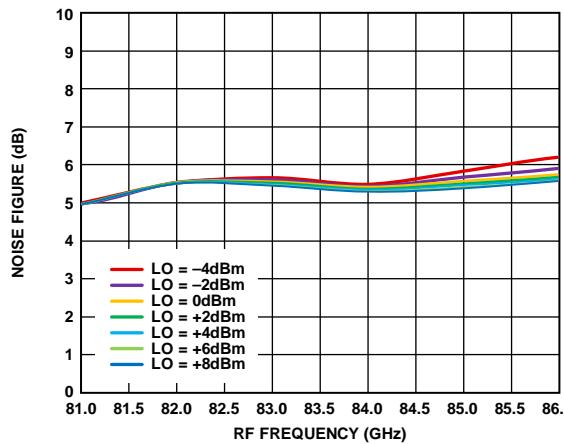


Figure 195. Noise Figure vs. RF Frequency at Various LO Powers,
IF = 500 MHz, $V_{DLNA} = 3$ V

13141-215

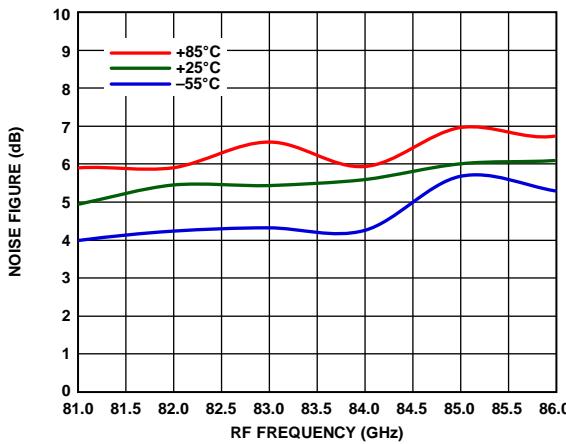


Figure 193. Noise Figure vs. RF Frequency at Various Temperatures,
LO = 2 dBm, IF = 1000 MHz, $V_{DLNA} = 3$ V

13141-213

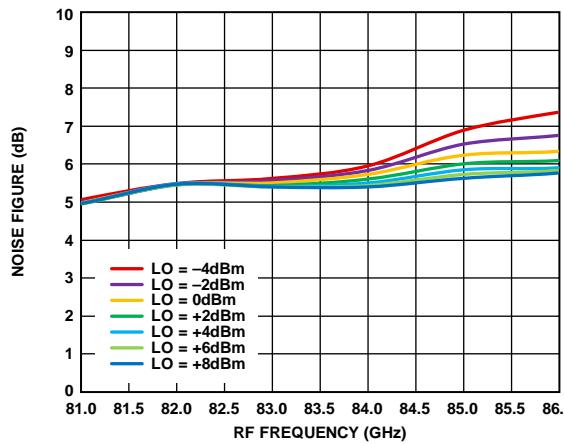


Figure 196. Noise Figure vs. RF Frequency at Various LO Powers,
RFIN = -20 dBm, IF = 500 MHz, $V_{DLNA} = 3$ V

13141-216

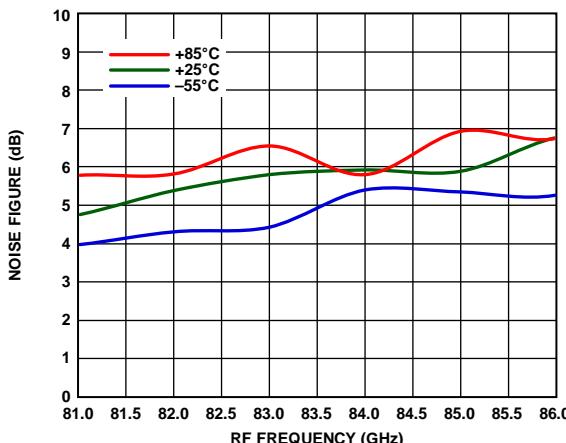


Figure 194. Noise Figure vs. RF Frequency at Various Temperatures,
LO = 2 dBm, IF = 2000 MHz, $V_{DLNA} = 3$ V

13141-214

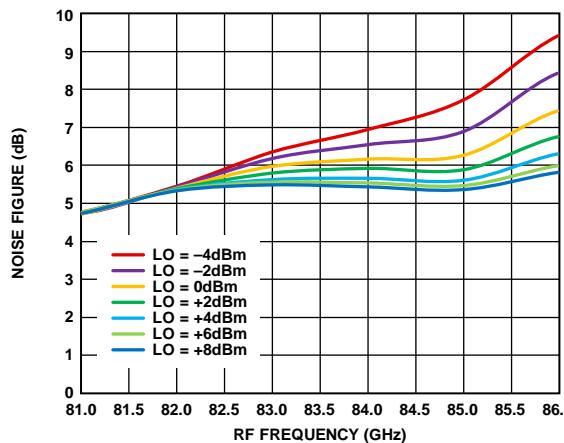


Figure 197. Noise Figure vs. RF Frequency at Various LO Powers,
IF = 2000 MHz, $V_{DLNA} = 3$ V

13141-217

AMPLITUDE BALANCE PERFORMANCE WITH LOWER SIDEBAND SELECTED

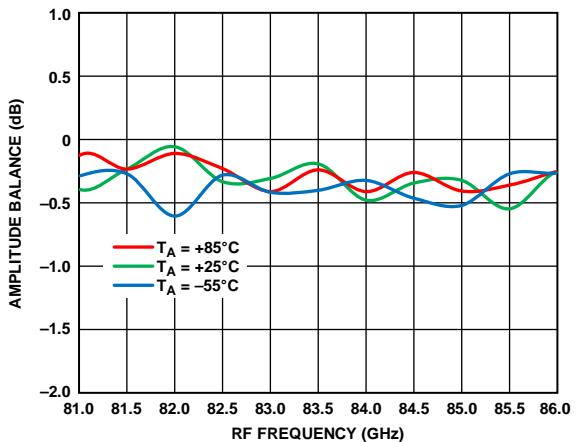


Figure 198. Amplitude Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 500 MHz, V_{DLNA} = 4 V

13141-218

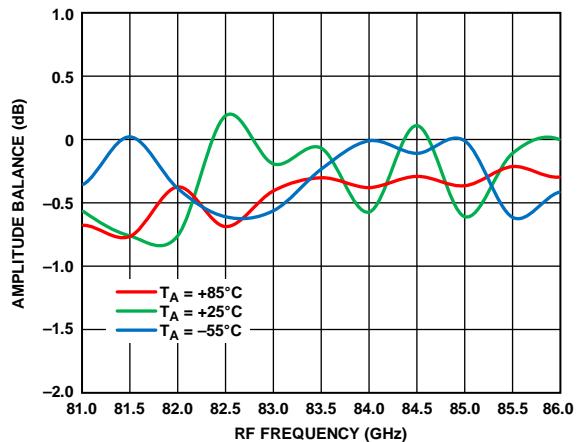


Figure 201. Amplitude Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 500 MHz, V_{DLNA} = 3 V

13141-221

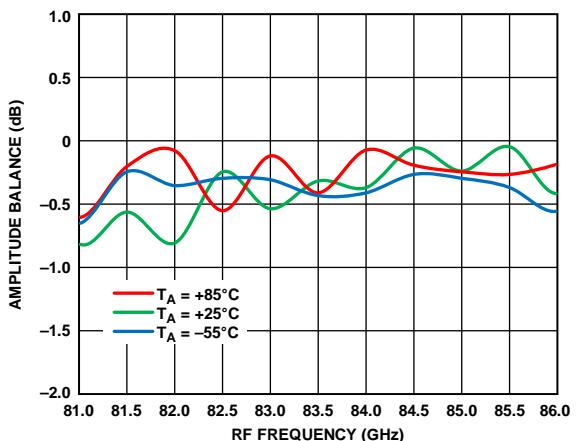


Figure 199. Amplitude Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 1000 MHz, V_{DLNA} = 4 V

13141-219

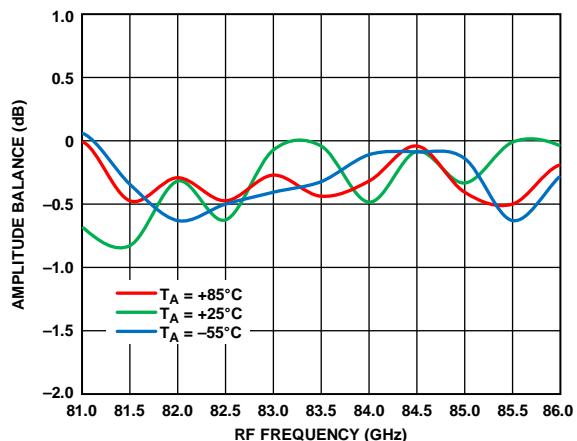


Figure 202. Amplitude Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 1000 MHz, V_{DLNA} = 3 V

13141-222

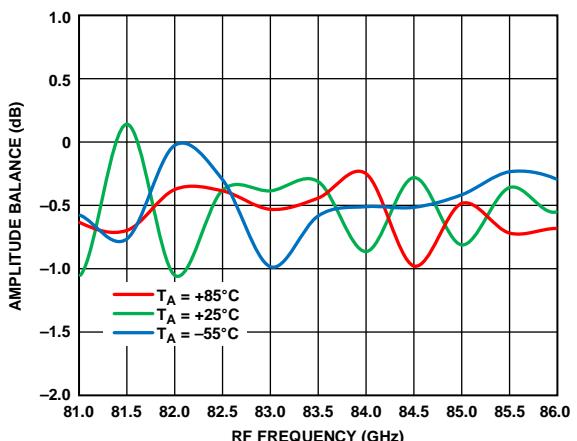


Figure 200. Amplitude Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, V_{DLNA} = 4 V

13141-220

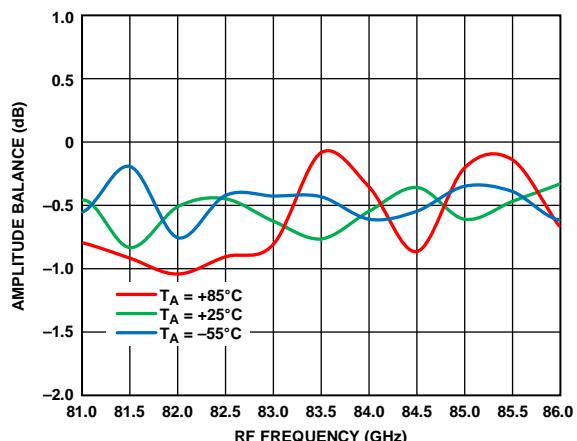


Figure 203. Amplitude Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, V_{DLNA} = 3 V

13141-223

PHASE BALANCE PERFORMANCE WITH LOWER SIDEband SELECTED

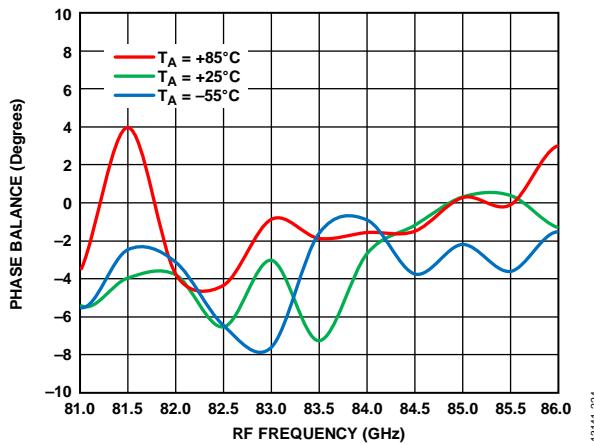


Figure 204. Phase Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 500 MHz, $V_{DLNA} = 4\text{ V}$

13141-224

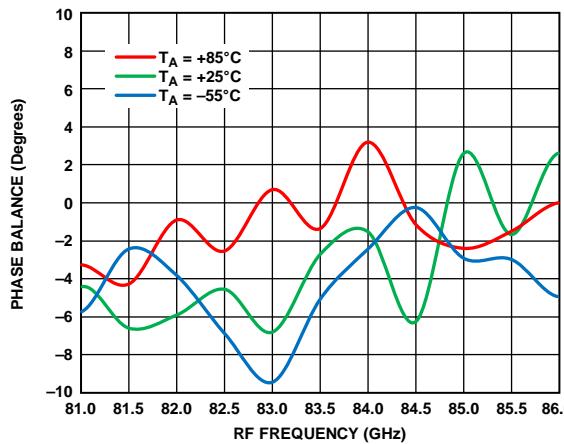


Figure 207. Phase Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 500 MHz, $V_{DLNA} = 3\text{ V}$

13141-227

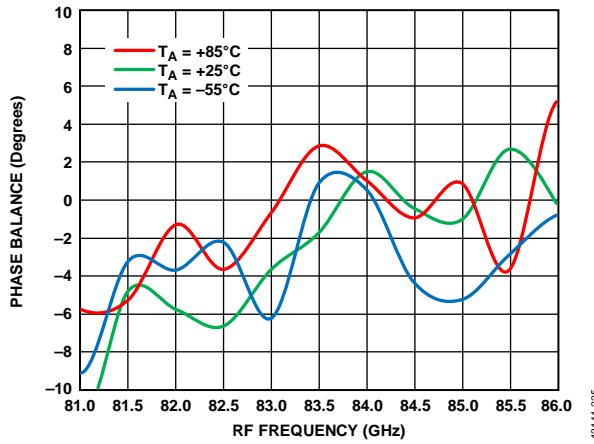


Figure 205. Phase Balance vs. RF Frequency at Various Temperatures
RFIN = -20 dBm, LO = 2 dBm, IF = 1000 MHz, $V_{DLNA} = 4\text{ V}$

13141-225

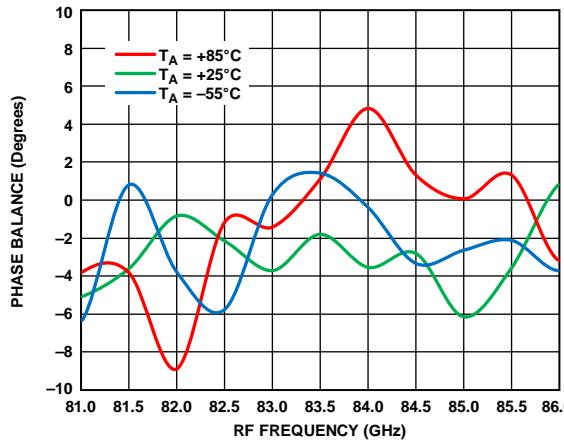


Figure 208. Phase Balance vs. RF Frequency at Various Temperatures
RFIN = -20 dBm, LO = 2 dBm, IF = 1000 MHz, $V_{DLNA} = 3\text{ V}$

13141-228

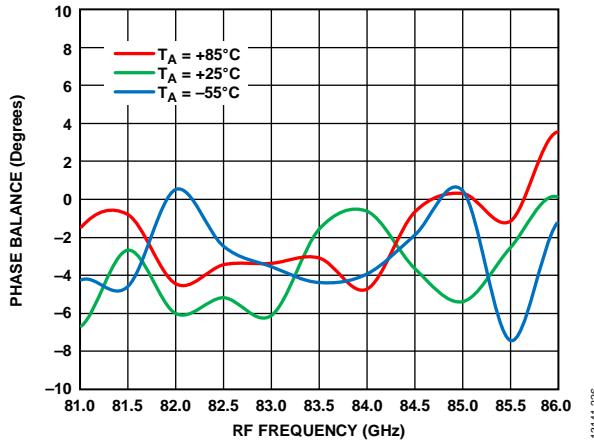


Figure 206. Phase Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, $V_{DLNA} = 4\text{ V}$

13141-226

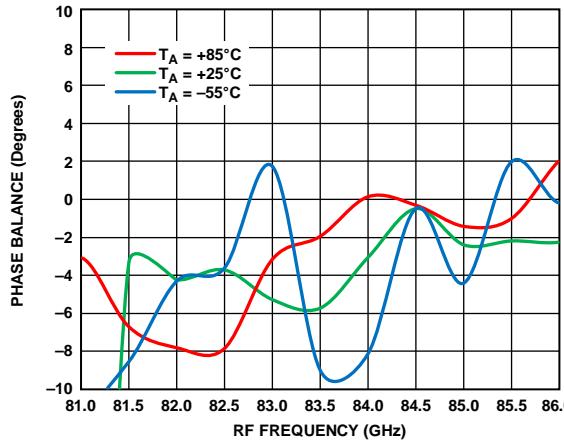


Figure 209. Phase Balance vs. RF Frequency at Various Temperatures,
RFIN = -20 dBm, LO = 2 dBm, IF = 2000 MHz, $V_{DLNA} = 3\text{ V}$

13141-229

**SPURIOUS PERFORMANCE WITH UPPER
SIDEband SELECTED, IF = 500 MHz**

$T_A = 25^\circ\text{C}$, $V_{GMIX} = -1 \text{ V}$, $V_{DAMPx} = 4 \text{ V}$, $V_{DMULT} = 1.5 \text{ V}$,
 $\text{LOIN} = 2 \text{ dBm}$.

Mixer spurious products are measured in dBc from the IF output power level. Spur values are $(M \times RF) - (N \times LO)$. N/A means not applicable.

 $M \times N$ Spurious Outputs, $V_{DLNA} = 4 \text{ V}$

RF = 81 GHz at RFIN = -10 dBm , LO frequency = 13.416 GHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	29.2	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	39	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	57	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	59.7	N/A

RF = 83 GHz at RFIN = -10 dBm , LO frequency = 13.75 GHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	25	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	36.6	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	49.3	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	53.9	N/A

RF = 86 GHz at RFIN = -10 dBm , LO frequency = 14.25 GHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	29.6	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	43.1	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	61.2	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	63.5	N/A

 $M \times N$ Spurious Outputs, $V_{DLNA} = 3 \text{ V}$

RF = 81 GHz at RFIN = -10 dBm , LO frequency = 13.416 GHz at LOIN = 2 dBm.

N × LO							
	0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	30.3	N/A	N/A	N/A
	3	N/A	N/A	N/A	41.5	N/A	N/A
	4	N/A	N/A	N/A	N/A	59.4	N/A
	5	N/A	N/A	N/A	N/A	N/A	64

RF = 83 GHz at RFIN = -10 dBm , LO frequency = 13.75 MHz at LOIN = 2 dBm.

N × LO							
	0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	26	N/A	N/A	N/A
	3	N/A	N/A	N/A	38.7	N/A	N/A
	4	N/A	N/A	N/A	N/A	52.1	N/A
	5	N/A	N/A	N/A	N/A	N/A	57.2

RF = 86 GHz at RFIN = -10 dBm , LO frequency = 14.25 GHz at LOIN = 2 dBm.

N × LO							
	0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	30.1	N/A	N/A	N/A
	3	N/A	N/A	N/A	45.4	N/A	N/A
	4	N/A	N/A	N/A	N/A	62.9	N/A
	5	N/A	N/A	N/A	N/A	N/A	67.3

**SPURIOUS PERFORMANCE WITH UPPER
SIDEBAND SELECTED, IF = 1000 MHz**

$T_A = 25^\circ\text{C}$, $V_{GMIX} = -1 \text{ V}$, $V_{DAMP_x} = 4 \text{ V}$, $V_{DMULT} = 1.5 \text{ V}$, $\text{LOIN} = 2 \text{ dBm}$.

Mixer spurious products are measured in dBc from the IF output power level. Spur values are $(M \times RF) - (N \times LO)$. N/A means not applicable.

 $M \times N$ Spurious Outputs, $V_{DLNA} = 4 \text{ V}$

RF = 81 GHz at RFIN = -10 dBm , LO frequency = 13.333 GHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	28.4	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	38.9	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	56.5	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	59.3	N/A

RF = 83 GHz at RFIN = -10 dBm , LO frequency = 13.666 MHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	26.5	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	37.1	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	52.8	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	56.7	N/A

RF = 86 GHz at RFIN = -10 dBm , LO frequency = 14.166 GHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	27.9	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	42	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	60.2	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	62.7	N/A

 $M \times N$ Spurious Outputs, $V_{DLNA} = 3 \text{ V}$

RF = 81 GHz at RFIN = -10 dBm , LO frequency = 13.333 GHz at LOIN = 2 dBm.

N × LO							
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	29.7	N/A	N/A	N/A
	3	N/A	N/A	N/A	41.7	N/A	N/A
	4	N/A	N/A	N/A	N/A	58.7	N/A
	5	N/A	N/A	N/A	N/A	N/A	63.4

RF = 83 GHz at RFIN = -10 dBm , LO frequency = 13.666 MHz at LOIN = 2 dBm.

N × LO							
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	27.5	N/A	N/A	N/A
	3	N/A	N/A	N/A	39.3	N/A	N/A
	4	N/A	N/A	N/A	N/A	55.1	N/A
	5	N/A	N/A	N/A	N/A	N/A	60.4

RF = 86 GHz at RFIN = -10 dBm , LO frequency = 14.166 GHz at LOIN = 2 dBm.

N × LO							
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	28.8	N/A	N/A	N/A
	3	N/A	N/A	N/A	44.2	N/A	N/A
	4	N/A	N/A	N/A	N/A	62.1	N/A
	5	N/A	N/A	N/A	N/A	N/A	66.6

**SPURIOUS PERFORMANCE WITH UPPER
SIDEband SELECTED, IF = 2000 MHz**

$T_A = 25^\circ\text{C}$, $V_{GMIX} = -1 \text{ V}$, $V_{DAMPx} = 4 \text{ V}$, $V_{DMULT} = 1.5 \text{ V}$, $\text{LOIN} = 2 \text{ dBm}$.

Mixer spurious products are measured in dBc from the IF output power level. Spur values are $(M \times RF) - (N \times LO)$. N/A means not applicable.

 $M \times N$ Spurious Outputs, $V_{DLNA} = 4 \text{ V}$

RF = 81 GHz at RFIN = -10 dBm , LO frequency = 13.166 GHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	28.1	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	42	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	56.3	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	61.1	N/A

RF = 83 GHz at RFIN = -10 dBm , LO frequency = 13.5 MHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	27.8	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	38.5	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	55.6	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	60.8	N/A

RF = 86 GHz at RFIN = -10 dBm , LO frequency = 14 GHz at LOIN = 2 dBm

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	25.1	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	38.5	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	52.9	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	52.3	N/A

 $M \times N$ Spurious Outputs, $V_{DLNA} = 3 \text{ V}$

RF = 81 GHz at RFIN = -10 dBm , LO frequency = 13.166 GHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	29.6	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	45.1	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	58.7	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	65.5	N/A

RF = 83 GHz at RFIN = -10 dBm , LO frequency = 13.5 MHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	28.8	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	41	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	58.6	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	64.7	N/A

RF = 86 GHz at RFIN = -10 dBm , LO frequency = 14 GHz at LOIN = 2 dBm

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	26	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	40.7	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	56.5	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	54.2	N/A

**SPURIOUS PERFORMANCE WITH LOWER
SIDEBAND SELECTED, IF = 500 MHz**

$T_A = 25^\circ\text{C}$, $V_{GMIX} = -1 \text{ V}$, $V_{DAMP_x} = 4 \text{ V}$, $V_{DMULT} = 1.5 \text{ V}$, $\text{LOIN} = 2 \text{ dBm}$.

Mixer spurious products are measured in dBc from the IF output power level. Spur values are $(M \times RF) - (N \times LO)$. N/A means not applicable.

 $M \times N$ Spurious Outputs, $V_{DLNA} = 4 \text{ V}$

RF = 81 GHz at RFIN = -10 dBm , LO frequency = 13.583 GHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	27.6	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	38.5	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	52.8	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	55.9	N/A

RF = 83 GHz at RFIN = -10 dBm , LO frequency = 13.916 GHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	25.4	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	38.3	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	49.2	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	55.1	N/A

RF = 86 GHz at RFIN = -10 dBm , LO frequency = 14.416 GHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	28.9	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	44.8	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	59.1	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	64.8	N/A

 $M \times N$ Spurious Outputs, $V_{DLNA} = 3 \text{ V}$

RF = 81 GHz at RFIN = -10 dBm , LO frequency = 13.583 GHz at LOIN = 2 dBm.

N × LO							
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	28.5	N/A	N/A	N/A
	3	N/A	N/A	N/A	40.6	N/A	N/A
	4	N/A	N/A	N/A	N/A	55.6	N/A
	5	N/A	N/A	N/A	N/A	N/A	59.8

RF = 83 GHz at RFIN = -10 dBm , LO frequency = 13.916 MHz at LOIN = 2 dBm.

N × LO							
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	26.3	N/A	N/A	N/A
	3	N/A	N/A	N/A	40	N/A	N/A
	4	N/A	N/A	N/A	N/A	52.1	N/A
	5	N/A	N/A	N/A	N/A	N/A	58.3

RF = 86 GHz at RFIN = -10 dBm , LO frequency = 14.416 GHz at LOIN = 2 dBm.

N × LO							
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	28.8	N/A	N/A	N/A
	3	N/A	N/A	N/A	46.5	N/A	N/A
	4	N/A	N/A	N/A	N/A	61.6	N/A
	5	N/A	N/A	N/A	N/A	N/A	68.5

**SPURIOUS PERFORMANCE WITH LOWER
SIDEBAND SELECTED, IF = 1000 MHz**

$T_A = 25^\circ\text{C}$, $V_{GMIX} = -1 \text{ V}$, $V_{DAMPx} = 4 \text{ V}$, $V_{DMULT} = 1.5 \text{ V}$, $\text{LOIN} = 2 \text{ dBm}$.

Mixer spurious products are measured in dBc from the IF output power level. Spur values are $(M \times RF) - (N \times LO)$. N/A means not applicable.

 $M \times N$ Spurious Outputs, $V_{DLNA} = 4 \text{ V}$

RF = 81 GHz at RFIN = -10 dBm , LO frequency = 13.666 GHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	26.5	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	37.1	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	51	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	55.6	N/A

RF = 83 GHz at RFIN = -10 dBm , LO frequency = 14 GHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	26.5	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	40	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	50.8	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	59.4	N/A

RF = 86 GHz at RFIN = -10 dBm , LO frequency = 14.5 GHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	27.7	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	42.8	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	56.9	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	68.1	N/A

 $M \times N$ Spurious Outputs, $V_{DLNA} = 3 \text{ V}$

RF = 81 GHz at RFIN = -10 dBm , LO frequency = 13.666 GHz at LOIN = 2 dBm.

N × LO							
	0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	27.6	N/A	N/A	N/A
	3	N/A	N/A	N/A	39.2	N/A	N/A
	4	N/A	N/A	N/A	N/A	53.4	N/A
	5	N/A	N/A	N/A	N/A	N/A	59.1

RF = 83 GHz at RFIN = -10 dBm , LO frequency = 14 GHz at LOIN = 2 dBm.

N × LO							
	0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	27.4	N/A	N/A	N/A
	3	N/A	N/A	N/A	41.8	N/A	N/A
	4	N/A	N/A	N/A	N/A	54.9	N/A
	5	N/A	N/A	N/A	N/A	N/A	62.9

RF = 86 GHz at RFIN = -10 dBm , LO frequency = 14.5 GHz at LOIN = 2 dBm.

N × LO							
	0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	28.1	N/A	N/A	N/A
	3	N/A	N/A	N/A	45.2	N/A	N/A
	4	N/A	N/A	N/A	N/A	60.3	N/A
	5	N/A	N/A	N/A	N/A	N/A	73.2

**SPURIOUS PERFORMANCE WITH LOWER
SIDEBAND SELECTED, IF = 2000 MHz**

$T_A = 25^\circ\text{C}$, $V_{GMIX} = -1 \text{ V}$, $V_{DAMP_x} = 4 \text{ V}$, $V_{DMULT} = 1.5 \text{ V}$, $\text{LOIN} = 2 \text{ dBm}$.

Mixer spurious products are measured in dBc from the IF output power level. Spur values are $(M \times RF) - (N \times LO)$. N/A means not applicable.

 $M \times N$ Spurious Outputs, $V_{DLNA} = 4 \text{ V}$

RF = 81 GHz at RFIN = -10 dBm , LO frequency = 13.833 GHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	25.1	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	38.3	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	48	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	56.5	N/A

RF = 83 GHz at RFIN = -10 dBm , LO frequency = 14.166 GHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	27.9	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	43.2	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	56.2	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	65.4	N/A

RF = 86 GHz at RFIN = -10 dBm , LO frequency = 14.666 GHz at LOIN = 2 dBm.

		N × LO						
		0	1	2	3	4	5	6
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	22.3	N/A	N/A	N/A	N/A
	3	N/A	N/A	N/A	40.1	N/A	N/A	N/A
	4	N/A	N/A	N/A	N/A	48.1	N/A	N/A
	5	N/A	N/A	N/A	N/A	N/A	63	N/A

 $M \times N$ Spurious Outputs, $V_{DLNA} = 3 \text{ V}$

RF = 81 GHz at RFIN = -10 dBm , LO frequency = 13.833 GHz at LOIN = 2 dBm.

N × LO							
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	25.9	N/A	N/A	N/A
	3	N/A	N/A	N/A	39.5	N/A	N/A
	4	N/A	N/A	N/A	N/A	51.8	N/A
	5	N/A	N/A	N/A	N/A	N/A	59.5

RF = 83 GHz at RFIN = -10 dBm , LO frequency = 14.166 GHz at LOIN = 2 dBm.

N × LO							
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	28.9	N/A	N/A	N/A
	3	N/A	N/A	N/A	45.6	N/A	N/A
	4	N/A	N/A	N/A	N/A	60	N/A
	5	N/A	N/A	N/A	N/A	N/A	63.9

RF = 86 GHz at RFIN = -10 dBm , LO frequency = 14.666 GHz at LOIN = 2 dBm.

N × LO							
M × RF	0	N/A	N/A	N/A	N/A	N/A	N/A
	1	N/A	N/A	N/A	N/A	N/A	N/A
	2	N/A	N/A	23.2	N/A	N/A	N/A
	3	N/A	N/A	N/A	42.3	N/A	N/A
	4	N/A	N/A	N/A	N/A	51	N/A
	5	N/A	N/A	N/A	N/A	N/A	67

THEORY OF OPERATION

The HMC7587 is a GaAs low noise I/Q downconverter with an integrated LO buffer and a 6 \times multiplier. See Figure 210 for a functional block diagram of the downconverter circuit architecture.

The RF input is internally ac-coupled and matched to 50 Ω . The input passes through four stages of low noise amplification. The preamplified RF input signal then splits and drives two singly balanced passive mixers.

Quadrature LO signals drive the two I and Q mixer cores. The LO path provides a 6 \times multiplier that allows the use of a lower frequency range LO input signal, typically between 11.83 GHz and 14.33 GHz. The 6 \times multiplier is implemented using a cascade of 3 \times and 2 \times multipliers. The LO buffer amplifiers are included on-chip to allow a typical LO drive level of only 2 dBm for full performance.

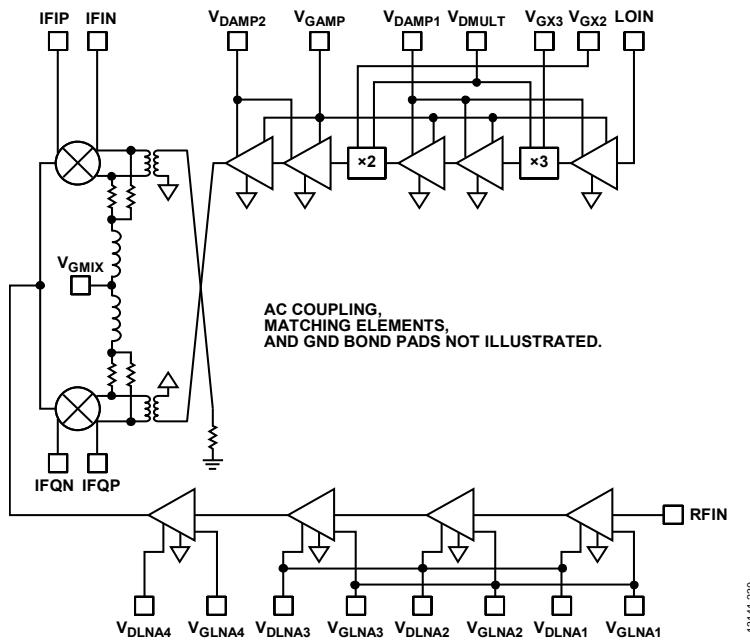


Figure 210. Downconverter Circuit Architecture

13144-230

APPLICATIONS INFORMATION

BIASING SEQUENCE

The HMC7587 uses several amplifier and multiplier stages. The active stages all use depletion mode pseudomorphic high electron mobility transistors (pHEMTs). To ensure transistor damage does not occur, use the following power-up bias sequence:

1. Apply a -2 V bias to V_{GAMP} , V_{GLNA1} , V_{GLNA2} , V_{GLNA3} , V_{GLNA4} , V_{GX2} , and V_{GX3} .
2. Apply a -1 V bias to V_{GMIX} .
3. Apply 4 V to V_{DAMP1} , V_{DAMP2} , V_{DLNA1} , V_{DLNA2} , V_{DLNA3} , and V_{DLNA4} , and apply 1.5 V to V_{DMULT} .
4. Adjust V_{GAMP} between -2 V and 0 V to achieve a total amplifier drain current ($I_{DAMP1} + I_{DAMP2}$) of 175 mA.
5. Adjust V_{GLNA1} , V_{GLNA2} , V_{GLNA3} , and V_{GLNA4} to achieve a total LNA drain current ($I_{DLNA1} + I_{DLNA2} + I_{DLNA3} + I_{DLNA4}$) of 50 mA.
6. Apply the LO input signal with a power level of 2 dBm and adjust V_{GX2} and V_{GX3} between -2 V and 0 V to achieve 80 mA of drain current on V_{DMULT} .

To power down the HMC7587, follow the procedure in reverse.

For additional guidance on general bias sequencing, see the [MMIC Amplifier Biasing Procedure](#) application note.

IMAGE REJECTION DOWNCONVERSION

A typical image rejection downconversion application circuit is shown in Figure 211. For image rejection downconversions, external 180° and 90° hybrid couplers are typically used. The 180° hybrids or baluns convert the differential I and Q output signals to unbalanced waveforms. The 90° hybrid then combines the outputs in quadrature to form a classic Hartley image rejection receiver with a typical image rejection of 30 dBc.

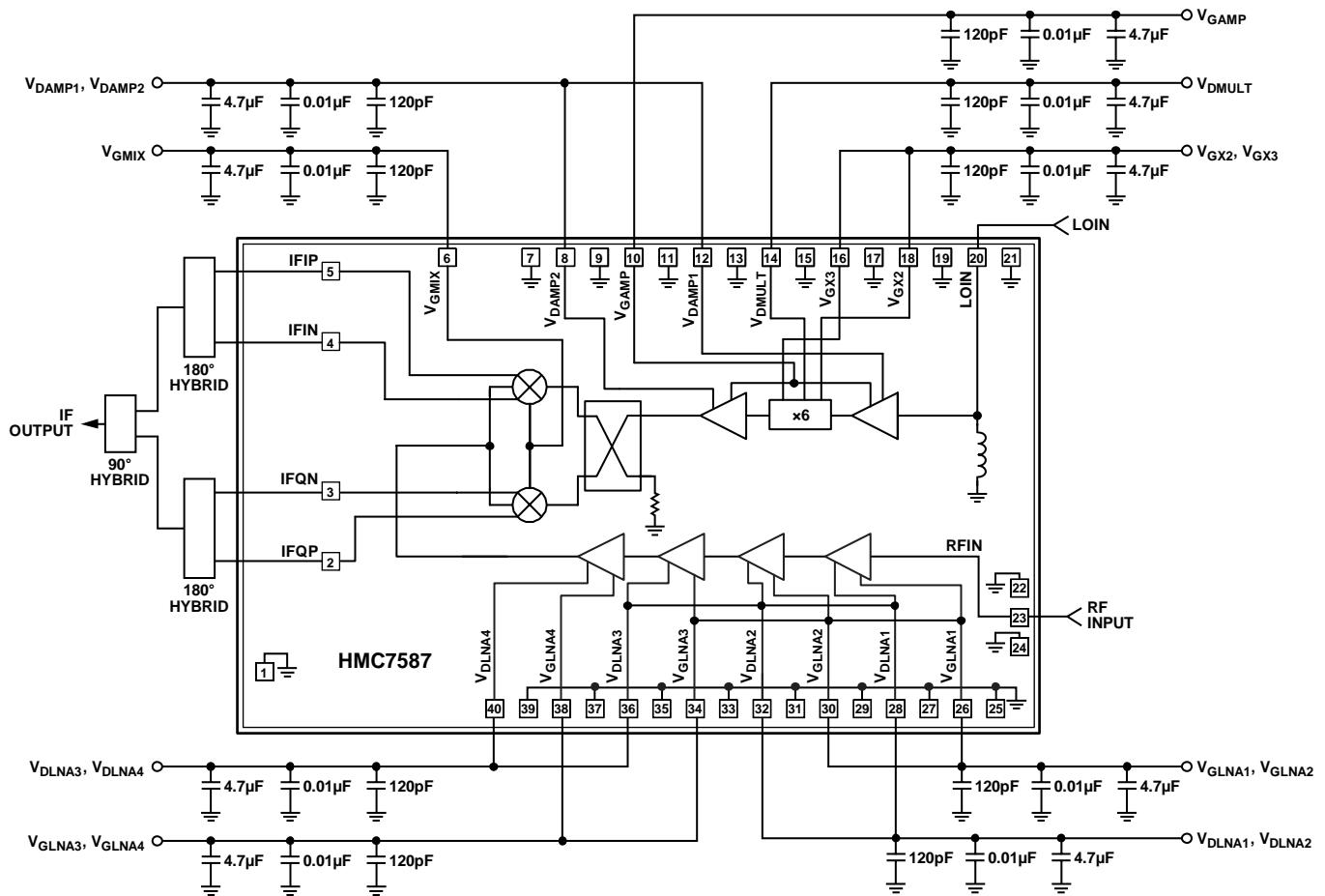


Figure 211. Typical Image Rejection Downconversion Application Circuit

ZERO IF DIRECT CONVERSION

A typical zero IF direct conversion application circuit is shown in Figure 212. It is important to ac couple the IFIP, IFIN, IFQP, and IFQN pads to the ADC inputs. Most ADCs are designed to operate with a common-mode voltage that is above ground.

The HMC7587 I/Q outputs are ground referenced and dc coupling to a differential signal source with a common-mode output voltage other than 0 V can cause degraded RF performance and possible device damage due to electrical overstress.

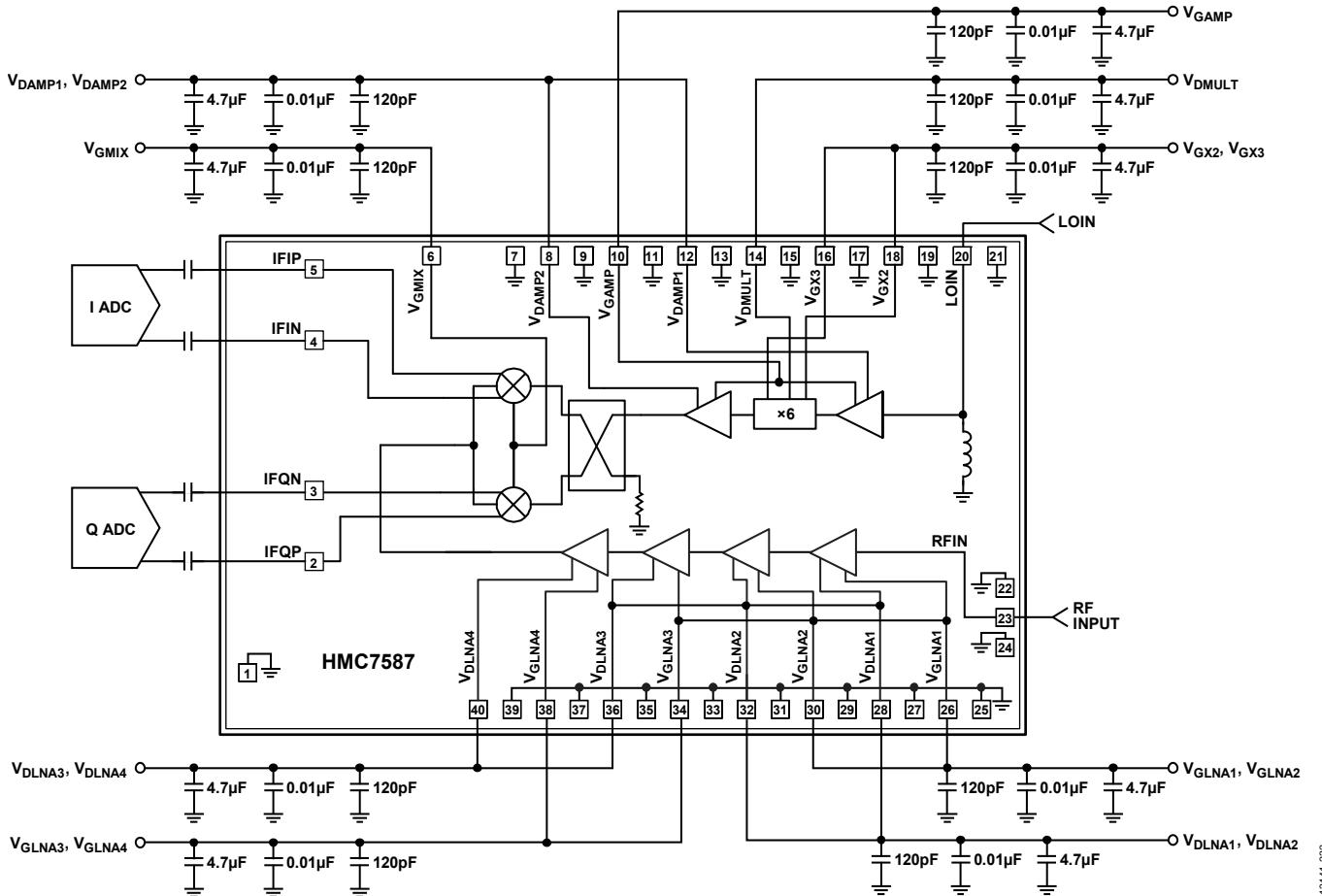
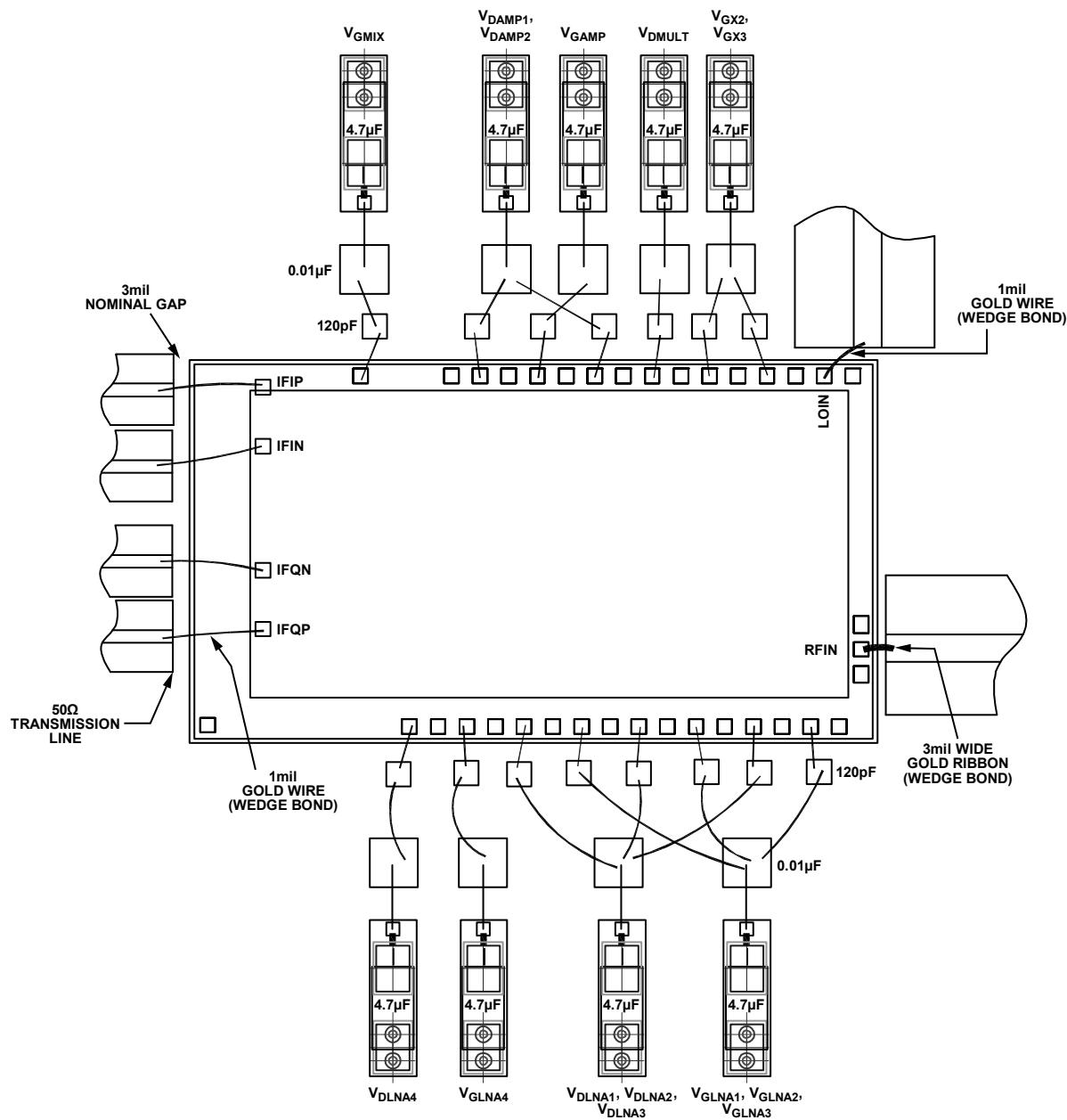


Figure 212. Typical Zero IF Direct Conversion Application Circuit

13141-232

ASSEMBLY DIAGRAM



13141-234

Figure 213. Assembly Diagram

MOUNTING AND BONDING TECHNIQUES FOR MILLIMETERWAVE GaAs MMICS

Attach the die directly to the ground plane eutectically or with conductive epoxy.

To bring RF to and from the chip, use $50\ \Omega$ microstrip transmission lines on 0.127 mm (5 mil) thick alumina thin film substrates (see Figure 214).

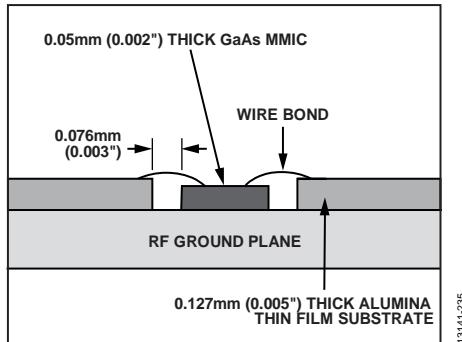


Figure 214. Routing RF Signals

To minimize bond wire length, place microstrip substrates as close to the die as possible. Typical die to substrate spacing is 0.076 mm to 0.152 mm (3 mil to 6 mil).

HANDLING PRECAUTIONS

To avoid permanent damage, adhere to the following precautions.

Storage

All bare die ship in either waffle or gel-based ESD protective containers, sealed in an ESD protective bag. After opening the sealed ESD protective bag, all die must be stored in a dry nitrogen environment.

Cleanliness

Handle the chips in a clean environment. Never use liquid cleaning systems to clean the chip.

Static Sensitivity

Follow ESD precautions to protect against ESD strikes.

Transients

Suppress instrument and bias supply transients while bias is applied. To minimize inductive pickup, use shielded signal and bias cables.

General Handling

Handle the chip on the edges only using a vacuum collet or with a sharp pair of bent tweezers. Because the surface of the chip has fragile air bridges, never touch the surface of the chip with a vacuum collet, tweezers, or fingers.

MOUNTING

The chip is back metallized and can be die mounted with gold/tin (AuSn) eutectic preforms or with electrically conductive epoxy. The mounting surface must be clean and flat.

Eutectic Die Attach

It is best to use an 80%/20% gold/tin preform with a work surface temperature of 255°C and a tool temperature of 265°C. When hot 90%/10% nitrogen/hydrogen gas is applied, maintain the tool tip temperature at 290°C. Do not expose the chip to a temperature greater than 320°C for more than 20 sec. No more than 3 sec of scrubbing is required for attachment.

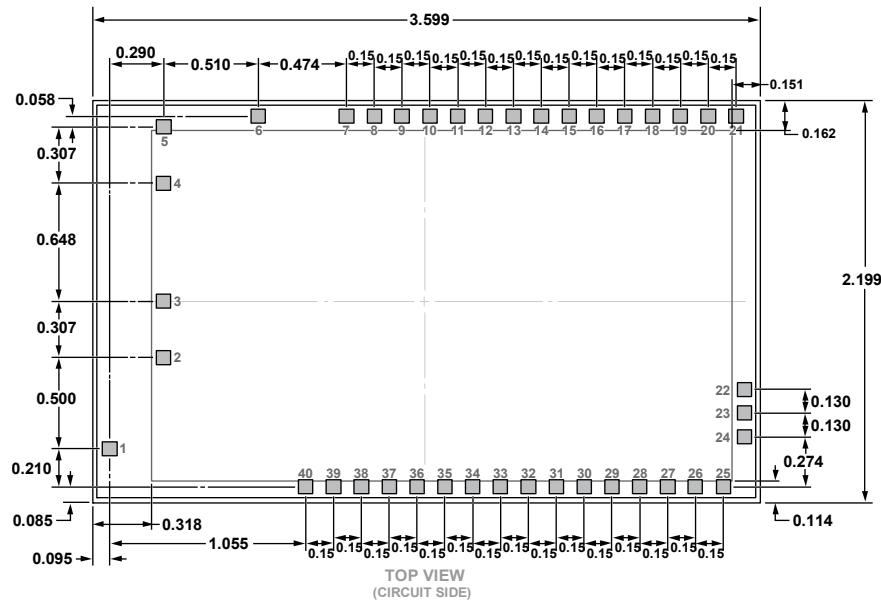
Epoxy Die Attach

ABLEBOND 84-1LIMIT is recommended for die attachment. Apply a minimum amount of epoxy to the mounting surface so that upon placing it into position, a thin epoxy fillet is observed around the perimeter of the chip. Cure epoxy per the schedule provided by the manufacturer.

WIRE BONDING

RF bonds made with (3 mil (0.0762 mm) \times 0.5 mil (0.0127 mm)) gold ribbon are recommended for the RF ports and wedge bonds with 1 mil (0.0254 mm) diameter gold wire are recommended for the IF and LO ports. These bonds must be thermosonically bonded with a force of 40 g to 60 g. DC bonds of 1 mil (0.0254 mm) diameter, thermosonically bonded, are recommended. Create ball bonds with a force of 40 g to 50 g and wedge bonds at 18 g to 22 g. Create all bonds with a nominal stage temperature of 150°C. Apply a minimum amount of ultrasonic energy to achieve reliable bonds. Keep all bonds as possible, less than 12 mils (0.31 mm).

OUTLINE DIMENSIONS



04-10-2015-A

Figure 215. 40-Pad Bare Die [CHIP]
(C-40-1)
Dimensions shown in millimeters

ORDERING GUIDE

Model ¹	Temperature Range	Package Description	Package Option ²
HMC7587	-55°C to +85°C	40-Pad Bare Die [CHIP]	C-40-1
HMC7587-SX	-55°C to +85°C	40-Pad Bare Die [CHIP]	C-40-1

¹ The HMC7587-SX consists of two pairs of the die in a gel pack for sample orders.

² This is a waffle pack option; contact Analog Devices, Inc., for additional packaging options.



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