

# BGX7220

## Dual receiver down mixer

Rev. 1 — 8 August 2012

Product data sheet

## 1. General description

The BGX7220 device combines a pair of high performance, high linearity down-mixers for use in receivers having a common local oscillator, for instance having main and diversity paths. The device covers the frequency range from 700 MHz to 950 MHz. Each mixer provides an input 1 dB compression point ( $ICP_{1dB}$ ) above 13 dBm, with an input third-order intercept point ( $IP3_i$ ) of 26 dBm. The small-signal Noise Figure (NF) is below 10 dB whereas under large signal blocking conditions the Noise Figure is typically 20 dB. Isolation between mixers is typically 55 dB.

## 2. Features and benefits

- 700 MHz to 950 MHz frequency operating range
- Conversion gain 8 dB in the 900 MHz band
- 13 dBm input power at 1 dB input compression point
- 26 dBm input third-order intercept point
- 10 dB typical small signal noise figure
- Integrated active biasing
- 5 V single supply operation
- Independent power-down hardware control pins per mixer
- Low bias current in Power-down mode
- Matched 50 Ω single-ended RF and LO input impedance
- ESD protection at all pins

## 3. Applications

- Mobile network infrastructure
- RF and IF applications
- Communication systems and radars
- Microwave and broadband
- Industrial applications

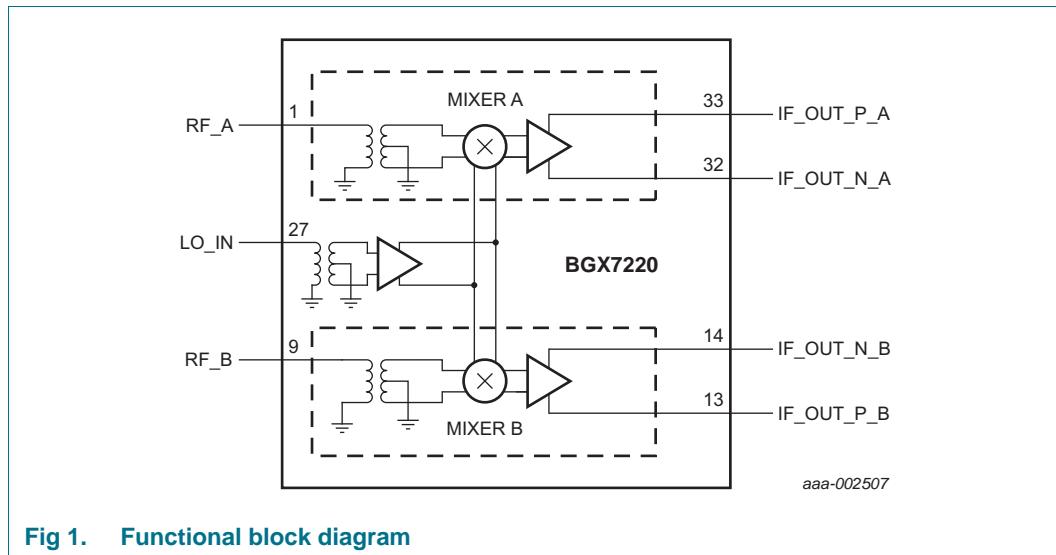
## 4. Ordering information

Table 1. Ordering information

Type number	Package			Version
	Name	Description		
BGX7220HN	HVQFN36	plastic thermal enhanced very thin quad flat package; no leads; 36 terminals; body 6 × 6 × 0.85 mm		SOT1092-2



## 5. Functional diagram



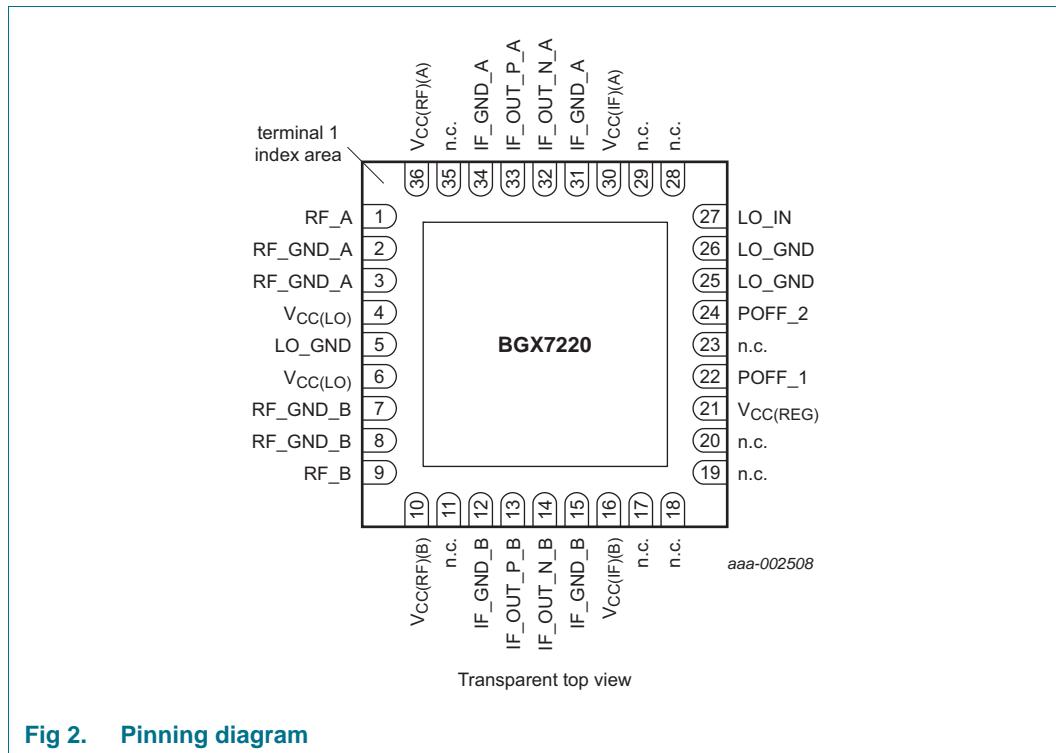
Each mixer, A and B employs a transformer to convert the single-ended RF input into a differential signal to drive the passive MOS mixer. The MOS mixer directly drives the IF amplifier. Its open-collector outputs deliver the differential signal into an external transformer load, referenced to the 5 V supply for maximum signal swing. Each mixer can be independently powered-off by a combination of POFF\_1 and POFF\_2 (see [Table 3](#)). The dual paths allow diversity operation with a common LO path. A transformer at the LO input converts the single-ended RF into a differential signal to drive the LO buffer chain.

The plastic package has an under-side heat-sink paddle which serves as a good RF ground.

## 6. Pinning information

### 6.1 Pinning

Viewing the device from the top (see [Figure 2](#)), the 2 RF input ports are at the left, the common LO input at the right, with IF outputs at the top and bottom. Multiple power and ground pins allow for independent supply domains to improve isolation between blocks.



**Fig 2.** Pinning diagram

## 6.2 Pin description

**Table 2.** Pin description

Symbol	Pin	Type [1]	Description
RF_A	1	I	receiver mixer single-ended RF input; mixer A
RF_GND_A	2	G	RF ground; mixer A
RF_GND_A	3	G	RF ground; mixer A
V <sub>CC(LO)</sub>	4	P	LO power supply
LO_GND	5	G	LO ground
V <sub>CC(LO)</sub>	6	P	LO power supply
RF_GND_B	7	G	RF ground; mixer B
RF_GND_B	8	G	RF ground; mixer B
RF_B	9	I	receiver mixer single-ended RF input; mixer B
V <sub>CC(RF)(B)</sub>	10	P	RF mixer power supply; mixer B
n.c.	11	-	not connected; to be tied to ground
IF_GND_B	12	G	IF ground; mixer B
IF_OUT_P_B	13	O	symmetrical IF output signal; mixer B
IF_OUT_N_B	14	O	symmetrical IF output signal; mixer B
IF_GND_B	15	G	IF amplifier ground; mixer B
V <sub>CC(IF)(B)</sub>	16	P	IF amplifier power supply; mixer B
n.c.	17	-	not connected; to be tied to ground
n.c.	18	-	not connected; to be tied to ground
n.c.	19	-	not connected; to be tied to ground
n.c.	20	-	not connected; to be tied to ground
V <sub>CC(REG)</sub>	21	P	internal regulator power supply

**Table 2.** Pin description ...*continued*

<b>Symbol</b>	<b>Pin</b>	<b>Type</b> <sup>[1]</sup>	<b>Description</b>
P OFF_1	22	I	logic input to power-off mixer
n.c.	23	-	not connected; to be tied to ground
P OFF_2	24	I	logic input to power-off mixer
LO_GND	25	G	LO ground
LO_GND	26	G	LO ground
LO_IN	27	I	single-ended local oscillator positive input
n.c.	28	-	not connected; to be tied to ground
n.c.	29	-	not connected; to be tied to ground
V <sub>CC(IF)(A)</sub>	30	P	IF amplifier power supply; mixer A
IF_GND_A	31	G	IF amplifier mixer; mixer A
IF_OUT_N_A	32	O	symmetrical IF negative output; mixer A
IF_OUT_P_A	33	O	symmetrical IF positive output; mixer A
IF_GND_A	34	G	IF ground; mixer A
n.c.	35	-	not connected; to be tied to ground
V <sub>CC(RF)(A)</sub>	36	P	RF power supply; mixer A
Exposed paddle	-	G	exposed paddle; must be connected to RF and DC ground

[1] G: ground; I: input; O: output; P: power.

## 7. Functional description

### 7.1 Power-up control

**Table 3.** Shutdown control

<b>Mode</b>	<b>Description</b>	<b>Function</b>	<b>P OFF_1</b>	<b>P OFF_2</b>
Active	mixers A and B fully active	shutdown disabled	0	0
Idle	mixers A and B fully off; current supplied to LO buffer	shutdown enabled	1	0
Main	mixer A active; mixer B off	partial shutdown	0	1
Diversity	mixer B active; mixer A off	partial shutdown	1	1

Power-up enable pins to allow each mixer to be placed in Power-down mode. These pins also enable the dedicated LO buffers for individual signal paths. A common LO input stage remains active whatever the state of the power off control inputs, in order to maintain good LO port matching. The time required to pass between active and inactive states is less than 10 µs. If the pins are left open or tied to ground, both mixers will be in active state.

## 8. Limiting values

**Table 4. Limiting values**

In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions	Min	Max	Unit
V <sub>CC</sub>	supply voltage		-0.3	5.5	V
P <sub>i(RF)</sub>	RF input power	continuous	-	20	dBm
P <sub>tot</sub>	total power dissipation		-	1.96	W
T <sub>mb</sub>	mounting base temperature		-40	+85	°C
T <sub>j</sub>	junction temperature		-	150	°C
T <sub>stg</sub>	storage temperature		-65	+150	°C
V <sub>ESD</sub>	electrostatic discharge voltage	EIA/JESD22-A114 (HBM) EIA/JESD22-C101 (FCDM)	-2500 -650	+2500 +650	V

## 9. Thermal characteristics

**Table 5. Thermal characteristics**

Symbol	Parameter	Conditions	Typ	Unit
R <sub>th(j-mb)</sub>	thermal resistance from junction to mounting base	[1]	8	°C/W

[1] Defined according to the conditions described in the Application Note AN11132.

## 10. Static characteristics

**Table 6. Static characteristics**Z<sub>s</sub> = Z<sub>L</sub> = 50 Ω; POFF\_1 = V<sub>IL</sub> and POFF\_2 = V<sub>IH</sub> (shutdown disabled). Typical values at V<sub>CC</sub> = 5 V, T<sub>mb</sub> = 25 °C, unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit	
V <sub>CC</sub>	supply voltage		4.75	5.0	5.25	V	
I <sub>Cq</sub>	quiescent collector current	IF output; per package pin	[1]	-	60	-	mA
<b>Shutdown digital input voltage</b>							
V <sub>IL</sub>	LOW-level input voltage		[1]	0	-	0.5	V
V <sub>IH</sub>	HIGH-level input voltage		[1]	2	-	5	V
<b>All digital inputs current</b>							
I <sub>IL</sub>	LOW-level input current		[1]	-	1	-	μA
I <sub>IH</sub>	HIGH-level input current		[1]	-	50	-	μA

[1] V<sub>CC</sub> = 4.75 V to 5.25 V, T<sub>mb</sub> = -40 °C to +85 °C.

## 11. Dynamic characteristics

**Table 7. Dynamic characteristics**

Typical application values:  $P_{OFF\_1} = V_{IL}$  and  $P_{OFF\_2} = V_{IL}$  (shutdown disabled); RF and LO ports driven by  $50 \Omega$  sources;  $P_{i(RF)} = -5 \text{ dBm}$ ;  $f_{i(RF)} = 850 \text{ MHz}$ ;  $T_{mb} = -40^\circ\text{C}$  to  $+85^\circ\text{C}$ ;  $V_{CC} = 4.75 \text{ V}$  to  $5.25 \text{ V}$ . Typical values at  $V_{CC} = 5 \text{ V}$ ;  $T_{mb} = 25^\circ\text{C}$ ;  $P_{i(RF)} = -5 \text{ dBm}$ ;  $P_{i(Lo)} = 0 \text{ dBm}$ ;  $f_{i(RF)} = 850 \text{ MHz}$ ;  $f_{IF} = 150 \text{ MHz}$ . All parameters are guaranteed by design and characterization, unless otherwise specified.

Symbol	Parameter	Conditions	Min <sup>[1]</sup>	Typ <sup>[2]</sup>	Max <sup>[1]</sup>	Unit	
$I_{CC}$	supply current	both mixers in active mode					
		$f_{lo} = 500 \text{ MHz}$	-	285	330	mA	
		$f_{lo} = 850 \text{ MHz}$	-	300	345	mA	
		$f_{lo} = 1150 \text{ MHz}$	-	310	355	mA	
$f_{i(RF)}$	RF input frequency		[3]	700	-	950	MHz
$f_{lo}$	local oscillator frequency	signal from frequency generator; $f_{IF} = 50 \text{ MHz}$ to $200 \text{ MHz}$	[3]	500	-	1150	MHz
$P_{i(Lo)}$	local oscillator input power	signal from frequency generator	[3]	-3	-	+3	dBm
$G_{conv}$	conversion gain	at $T_{mb} = 25^\circ\text{C}$ ; $f_{i(RF)} = 700 \text{ MHz}$ to $950 \text{ MHz}$	[4]	7.5	8.2	9	dB
$\Delta G/\Delta T$	gain variation with temperature	$T_{mb} = -40^\circ\text{C}$ to $+85^\circ\text{C}$		-	0.007	-	dB/ $^\circ\text{C}$
$\Delta G$	gain deviation	$f_{i(RF)} = 700 \text{ MHz}$ to $715 \text{ MHz}$ ; $f_{lo} = 560 \text{ MHz}$ or $f_{lo} = 855 \text{ MHz}$	[4][7]	-	0.15	-	dB
		$f_{i(RF)} = 750 \text{ MHz}$ to $760 \text{ MHz}$ ; $f_{lo} = 605 \text{ MHz}$ or $f_{lo} = 905 \text{ MHz}$	[4][7]	-	0.4	-	dB
		$f_{i(RF)} = 780 \text{ MHz}$ to $795 \text{ MHz}$ ; $f_{lo} = 640 \text{ MHz}$ or $f_{lo} = 940 \text{ MHz}$	[4][7]	-	0.15	-	dB
		$f_{i(RF)} = 815 \text{ MHz}$ to $860 \text{ MHz}$ ; $f_{lo} = 690 \text{ MHz}$ or $f_{lo} = 985 \text{ MHz}$	[4][7]	-	0.3	-	dB
		$f_{i(RF)} = 880 \text{ MHz}$ to $915 \text{ MHz}$ ; $f_{lo} = 750 \text{ MHz}$ or $f_{lo} = 1045 \text{ MHz}$	[4][7]	-	0.15	-	dB
$ICP_{1dB}$	1 dB input compression point		[6]	10.5	13	-	dBm
$IP3_i$	input third-order intercept point	$f_{i(RF)1}$ to $f_{i(RF)2} = 1 \text{ MHz}$ ; $P_{i(RF)} = -5 \text{ dBm}$ per tone	[4]	24.5	26	-	dBm
2RF-2LO	second-order spurious rejection	2 tone inputs at $P_{i(RF)} = -10 \text{ dBm}$ ; $f_{i(RF)} = 850 \text{ MHz}$ ; $f_{i(Lo)} = 950 \text{ MHz}$ ; $f_{i(SPUR)} = 900 \text{ MHz}$		-57	-63	-	dBc
$NF_{SSB}$	single sideband noise figure			-	9.5	12	dB
$NF_B$	noise figure under blocking conditions	input in-band blocker $+8 \text{ dBm}$ ; $\Delta f_o = 100 \text{ MHz}$	[7]	-	20	-	dB
$\alpha_{L(RF)lo}$	local oscillator RF leakage	at RF input port; LO input power = $0 \text{ dBm}$		-	-	-35	dBm
$\alpha_{L(IF)lo}$	local oscillator IF leakage	at IF output port; LO input power = $0 \text{ dBm}$		-	-	-35	dBm
$\alpha_{isol}$	isolation	between mixer A and B; $P_{i(RF)} = -10 \text{ dBm}$ ; measured at unwanted IF port		45	55	-	dB

**Table 7. Dynamic characteristics ...continued**

Typical application values:  $P_{OFF\_1} = V_{IL}$  and  $P_{OFF\_2} = V_{IL}$  (shutdown disabled); RF and LO ports driven by  $50 \Omega$  sources;  $P_{i(RF)} = -5 \text{ dBm}$ ;  $f_{i(RF)} = 850 \text{ MHz}$ ;  $T_{mb} = -40 \text{ }^\circ\text{C}$  to  $+85 \text{ }^\circ\text{C}$ ;  $V_{CC} = 4.75 \text{ V}$  to  $5.25 \text{ V}$ . Typical values at  $V_{CC} = 5 \text{ V}$ ;  $T_{mb} = 25 \text{ }^\circ\text{C}$ ;  $P_{i(RF)} = -5 \text{ dBm}$ ;  $P_{i(Lo)} = 0 \text{ dBm}$ ;  $f_{i(RF)} = 850 \text{ MHz}$ ;  $f_{IF} = 150 \text{ MHz}$ . All parameters are guaranteed by design and characterization, unless otherwise specified.

Symbol	Parameter	Conditions	Min <sup>[1]</sup>	Typ <sup>[2]</sup>	Max <sup>[1]</sup>	Unit
S11_RF	RF input return loss	$f_{i(RF)} = 700 \text{ MHz}$ to $950 \text{ MHz}$	-	12	-	dB
S11_LO	LO input return loss	$f_{i(Lo)} = 500 \text{ MHz}$ to $1150 \text{ MHz}$	-	12	-	dB
S22_IF	IF output return loss	$f_{IF} = 50 \text{ MHz}$ to $200 \text{ MHz}$	-	14	-	dB

[1] For all minimum and maximum values the conditions are:  $P_{i(RF)} = -5 \text{ dBm}$ ;  $P_{i(Lo)} = 0 \text{ dBm}$ ;  $f_{i(RF)} = 850 \text{ MHz}$ ;  $f_{IF} = 150 \text{ MHz}$ ;  $T_{mb} = -40 \text{ }^\circ\text{C}$  to  $+85 \text{ }^\circ\text{C}$ ;  $V_{CC} = 4.75 \text{ V}$  to  $5.25 \text{ V}$ . Unless otherwise specified in the conditions.

[2] For all typical values, the conditions are:  $P_{i(RF)} = -5 \text{ dBm}$ ;  $P_{i(Lo)} = 0 \text{ dBm}$ ;  $f_{i(RF)} = 850 \text{ MHz}$ ;  $f_{IF} = 150 \text{ MHz}$ ;  $T_{mb} = 25 \text{ }^\circ\text{C}$ ;  $V_{CC} = 5 \text{ V}$ . Unless otherwise specified in the conditions.

[3] Operation outside this range is possible but parameters are not guaranteed.

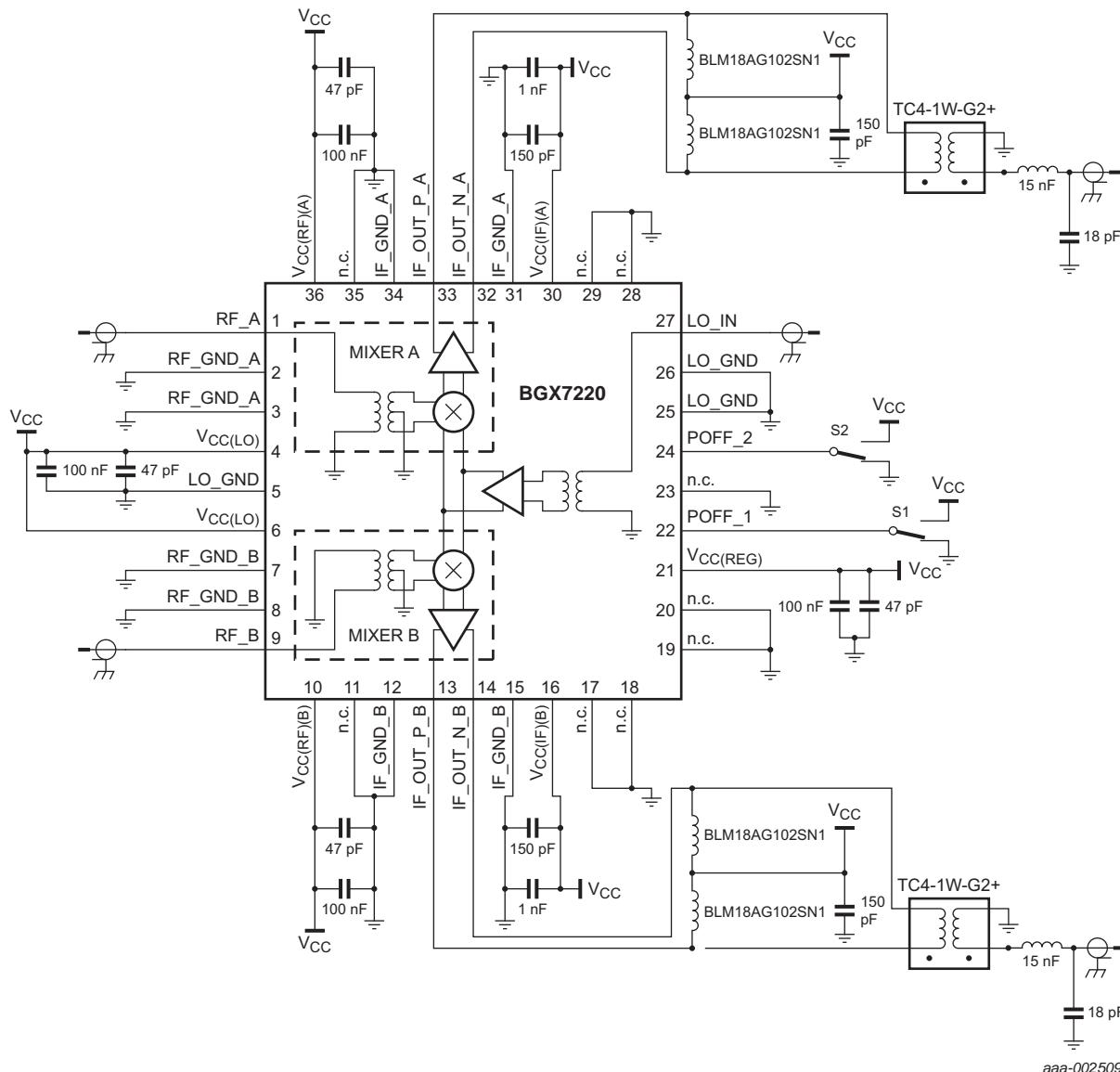
[4] Class A operation.

[5]  $f_{IF}$  is variable.

[6] Maximum reliable continuous input power applied to the RF or IF port of this device is 12 dBm from a  $50 \Omega$  source.

[7]  $NF_B$  can be improved by 1 dB per dB as a function of the  $P_{i(Lo)}$ .

## 12. Application information

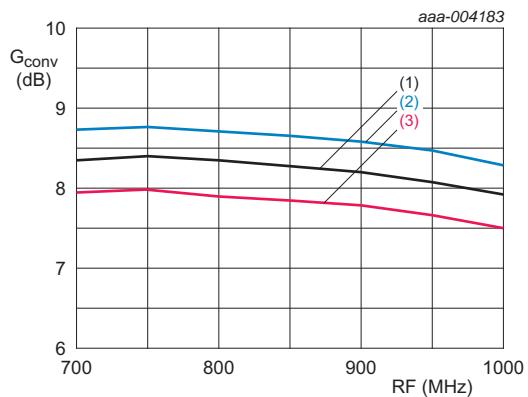


**Fig 3. Application diagram**

[Figure 3](#) shows a typical wideband application circuit. Both RF and RF reference pins need to be AC coupled. The inputs are internally DC biased in order to provide good ESD protection, and to support large input signals without clamping. The output matching requires a transformer to cope with the DC at the output.

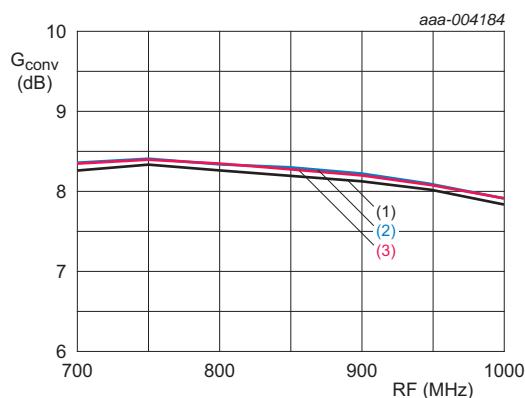
## 13. Test information

Parameters for the following drawings:  $V_{CC} = 5$  V;  $T_{mb} = 25$  °C;  $P_{i(RF)} = -5$  dBm;  $P_{i(LO)} = 0$  dBm;  $f_{IF} = 150$  MHz; unless otherwise specified.



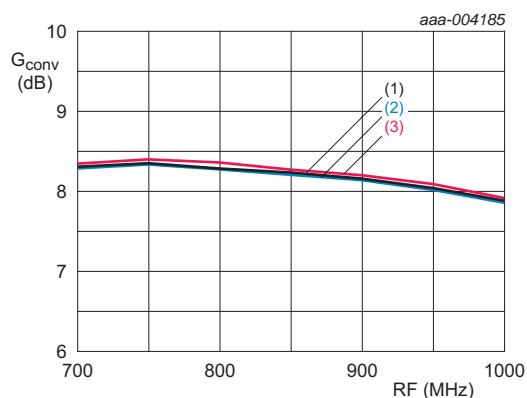
- (1)  $T_{mb} = +25$  °C.
- (2)  $T_{mb} = -40$  °C.
- (3)  $T_{mb} = +85$  °C.

Fig 4.  $G_{conv}$  versus  $f_{RF}$  (high side LO) and  $T_{mb}$



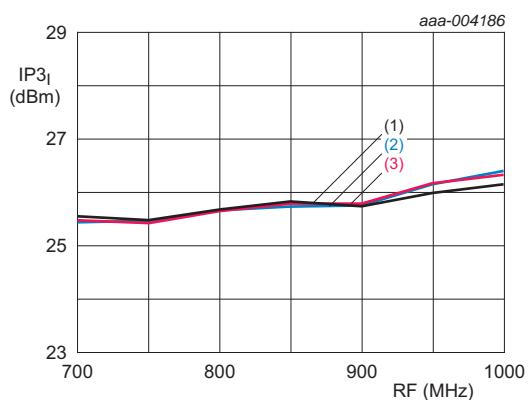
- (1)  $P_{i(LO)} = 0$  dBm.
- (2)  $P_{i(LO)} = -3$  dBm.
- (3)  $P_{i(LO)} = +3$  dBm.

Fig 5.  $G_{conv}$  versus  $f_{RF}$  (high side LO) and  $P_{i(LO)}$



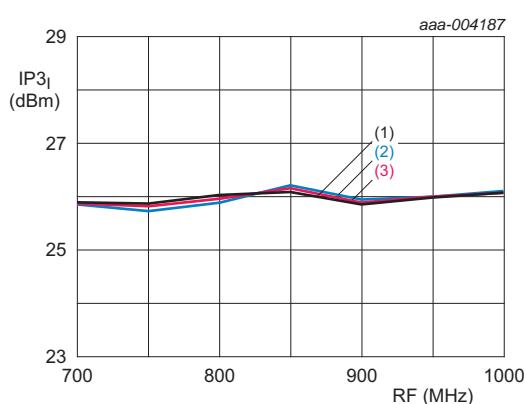
- (1)  $V_{CC} = 5$  V.
- (2)  $V_{CC} = 4.75$  V.
- (3)  $V_{CC} = 5.25$  V.

Fig 6.  $G_{conv}$  versus  $f_{RF}$  (high side LO) and  $V_{CC}$



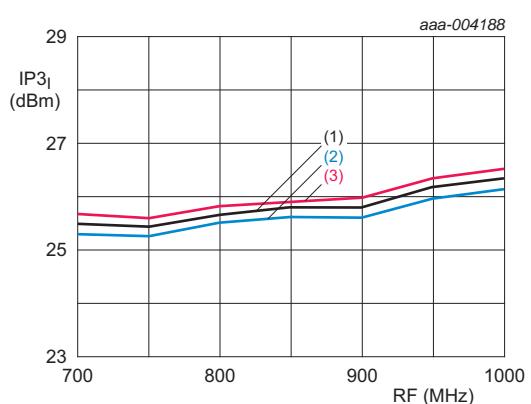
- (1)  $T_{mb} = +25 \text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40 \text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85 \text{ }^{\circ}\text{C}$ .

**Fig 7.** IP3<sub>i</sub> versus f<sub>RF</sub> (high side LO) and T<sub>mb</sub>



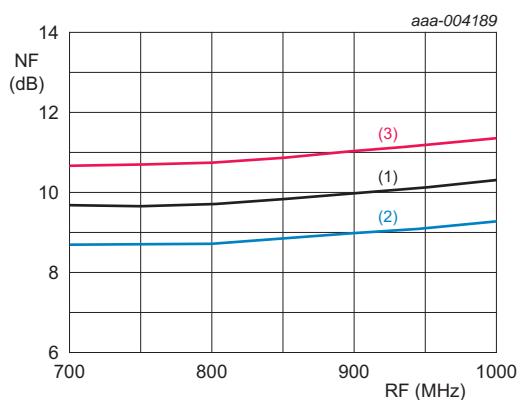
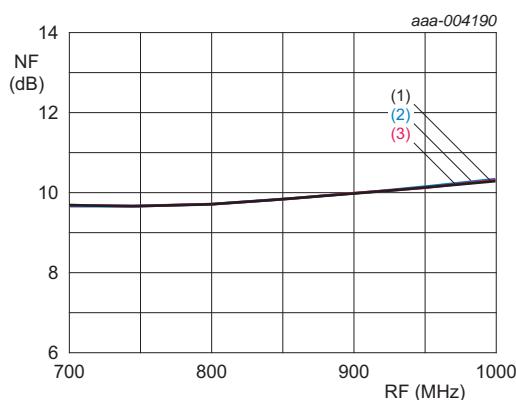
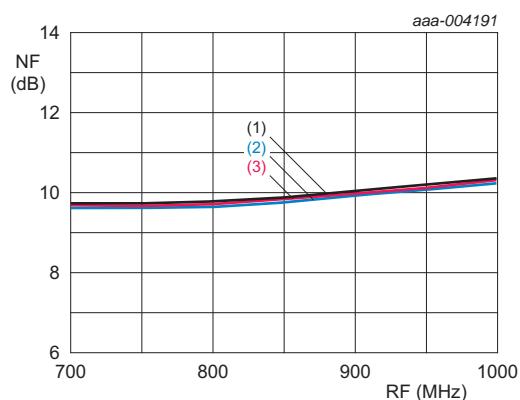
- (1)  $P_{i(\text{LO})} = 0 \text{ dBm}$ .
- (2)  $P_{i(\text{LO})} = -3 \text{ dBm}$ .
- (3)  $P_{i(\text{LO})} = +3 \text{ dBm}$ .

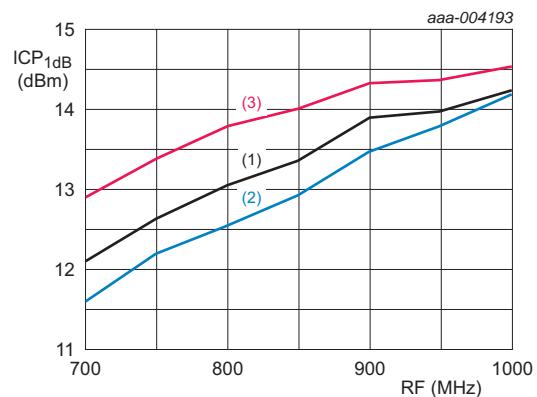
**Fig 8.** IP3<sub>i</sub> versus f<sub>RF</sub> (high side LO) and P<sub>i(LO)</sub>



- (1)  $V_{CC} = 5 \text{ V}$ .
- (2)  $V_{CC} = 4.75 \text{ V}$ .
- (3)  $V_{CC} = 5.25 \text{ V}$ .

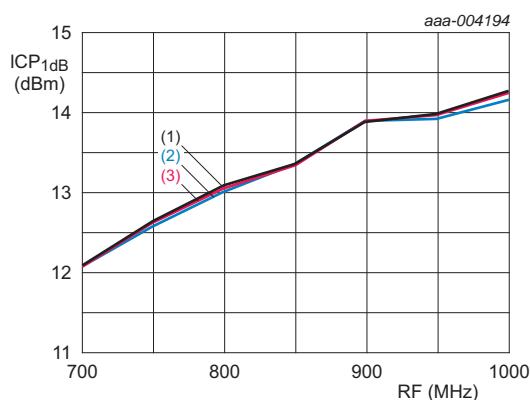
**Fig 9.** IP3<sub>i</sub> versus f<sub>RF</sub> (high side LO) and V<sub>CC</sub>

(1) T<sub>mb</sub> = +25 °C.(2) T<sub>mb</sub> = -40 °C.(3) T<sub>mb</sub> = +85 °C.**Fig 10. NF versus f<sub>RF</sub> (high side LO) and T<sub>mb</sub>**(1) P<sub>i(LO)</sub> = 0 dBm.(2) P<sub>i(LO)</sub> = -3 dBm.(3) P<sub>i(LO)</sub> = +3 dBm.**Fig 11. NF versus f<sub>RF</sub> (high side LO) and P<sub>i(LO)</sub>**(1) V<sub>CC</sub> = 5 V.(2) V<sub>CC</sub> = 4.75 V.(3) V<sub>CC</sub> = 5.25 V.**Fig 12. NF versus f<sub>RF</sub> (high side LO) and V<sub>CC</sub>**



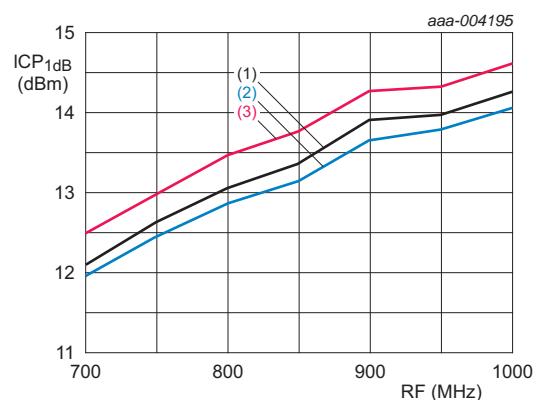
- (1) T<sub>mb</sub> = +25 °C.
- (2) T<sub>mb</sub> = -40 °C.
- (3) T<sub>mb</sub> = +85 °C.

**Fig 13. ICP<sub>1dB</sub> versus f<sub>RF</sub> (high side LO) and T<sub>mb</sub>**



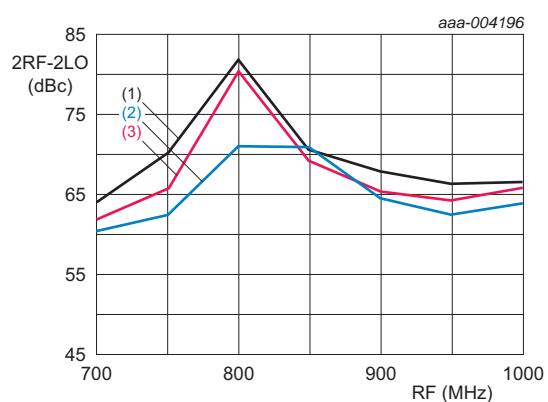
- (1) P<sub>i(LO)</sub> = 0 dBm.
- (2) P<sub>i(LO)</sub> = -3 dBm.
- (3) P<sub>i(LO)</sub> = +3 dBm.

**Fig 14. ICP<sub>1dB</sub> versus f<sub>RF</sub> (high side LO) and P<sub>i(LO)</sub>**

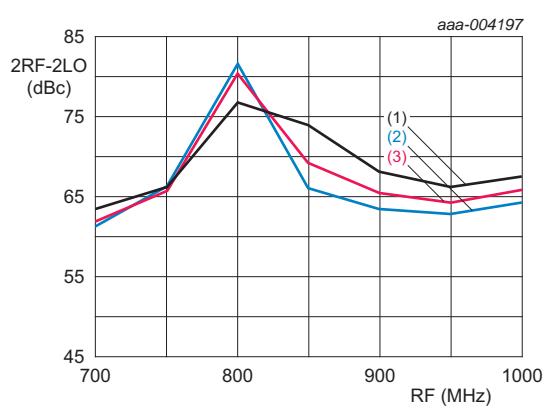


- (1) V<sub>CC</sub> = 5 V.
- (2) V<sub>CC</sub> = 4.75 V.
- (3) V<sub>CC</sub> = 5.25 V.

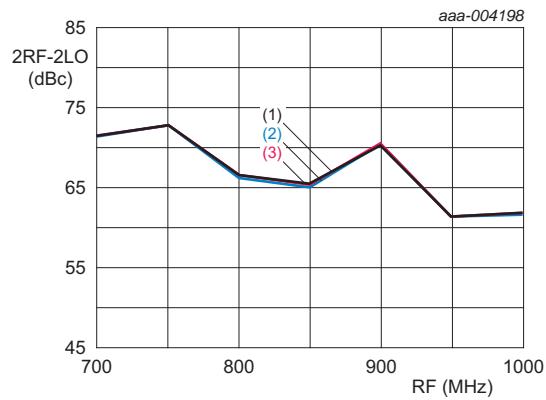
**Fig 15. ICP<sub>1dB</sub> versus f<sub>RF</sub> (high side LO) and V<sub>CC</sub>**



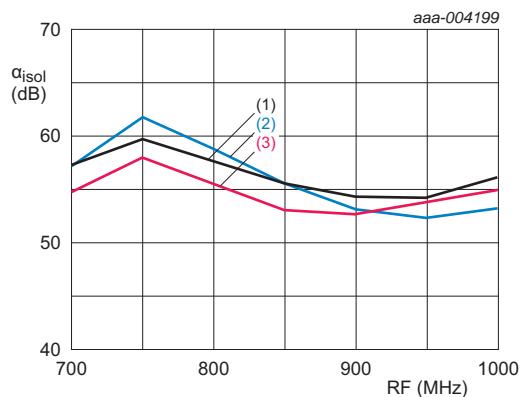
**Fig 16.** 2RF-2LO response versus  $f_{RF}$  (high side LO) and  $T_{mb}$



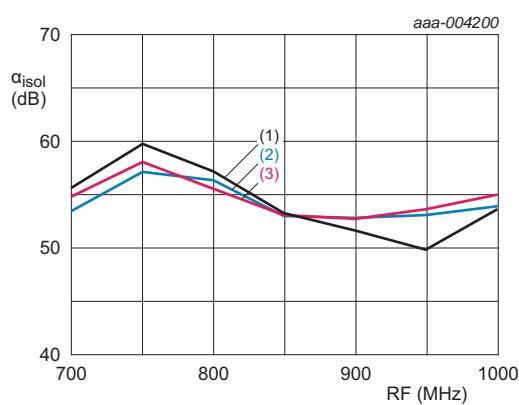
**Fig 17.** 2RF-2LO response versus  $f_{RF}$  (high side LO) and  $P_{i(LO)}$



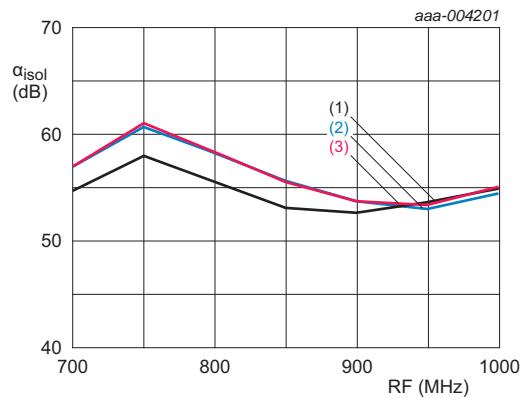
**Fig 18.** 2RF-2LO response versus  $f_{RF}$  (high side LO) and  $V_{CC}$



- (1)  $T_{mb} = +25^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85^{\circ}\text{C}$ .

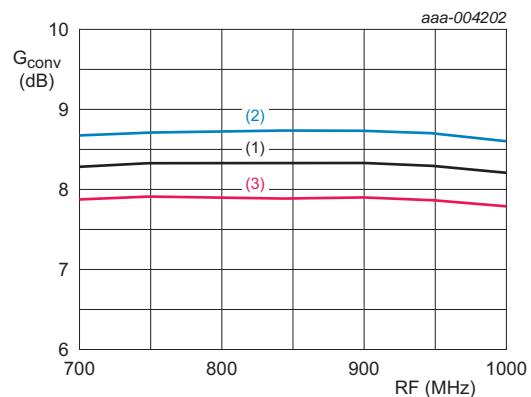
**Fig 19.**  $\alpha_{\text{isol}}$  versus  $f_{\text{RF}}$  (high side LO) and  $T_{mb}$ 

- (1)  $P_{i(\text{LO})} = 0 \text{ dBm}$ .
- (2)  $P_{i(\text{LO})} = -3 \text{ dBm}$ .
- (3)  $P_{i(\text{LO})} = +3 \text{ dBm}$ .

**Fig 20.**  $\alpha_{\text{isol}}$  versus  $f_{\text{RF}}$  (high side LO) and  $P_{i(\text{LO})}$ 

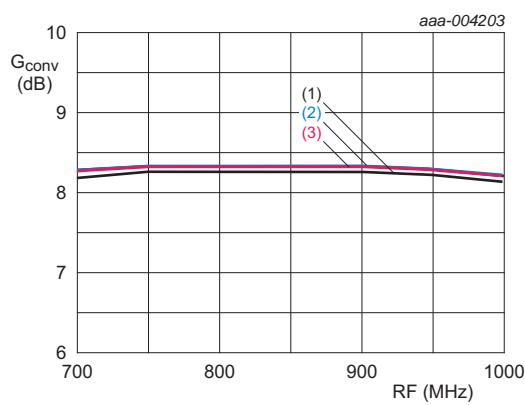
- (1)  $V_{CC} = 5 \text{ V}$ .
- (2)  $V_{CC} = 4.75 \text{ V}$ .
- (3)  $V_{CC} = 5.25 \text{ V}$ .

**Fig 21.**  $\alpha_{\text{isol}}$  versus  $f_{\text{RF}}$  (high side LO) and  $V_{CC}$



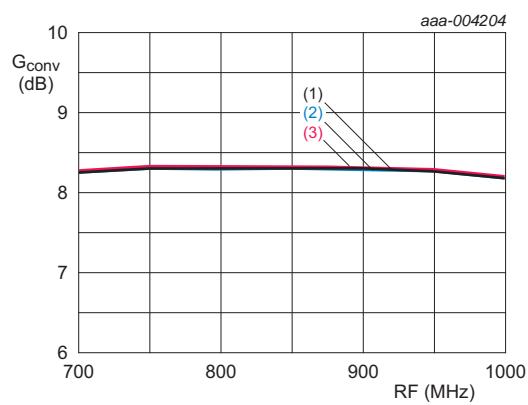
- (1)  $T_{mb} = +25^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85^{\circ}\text{C}$ .

**Fig 22.**  $G_{conv}$  versus  $f_{RF}$  (low side LO) and  $T_{mb}$



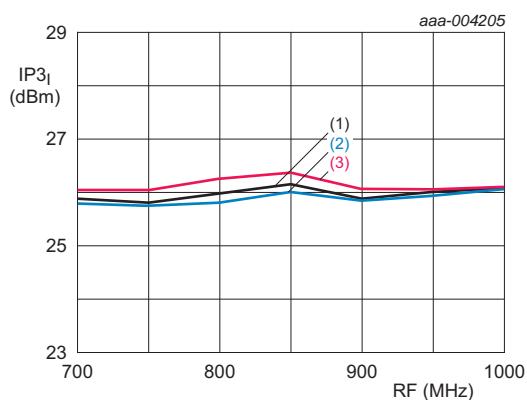
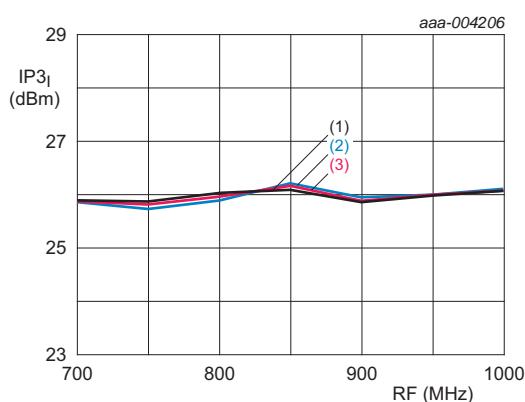
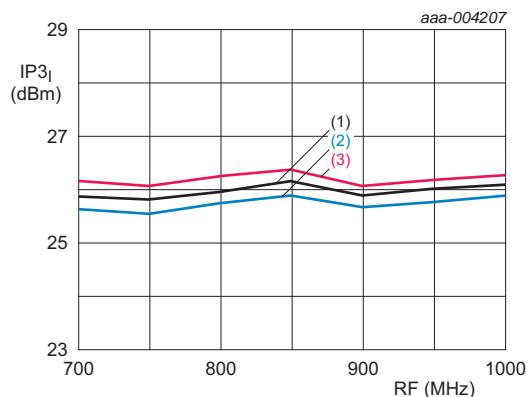
- (1)  $P_{i(LO)} = 0 \text{ dBm}$ .
- (2)  $P_{i(LO)} = -3 \text{ dBm}$ .
- (3)  $P_{i(LO)} = +3 \text{ dBm}$ .

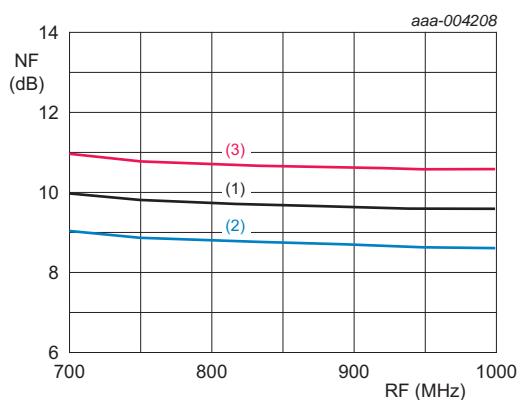
**Fig 23.**  $G_{conv}$  versus  $f_{RF}$  (low side LO) and  $P_{i(LO)}$



- (1)  $V_{CC} = 5 \text{ V}$ .
- (2)  $V_{CC} = 4.75 \text{ V}$ .
- (3)  $V_{CC} = 5.25 \text{ V}$ .

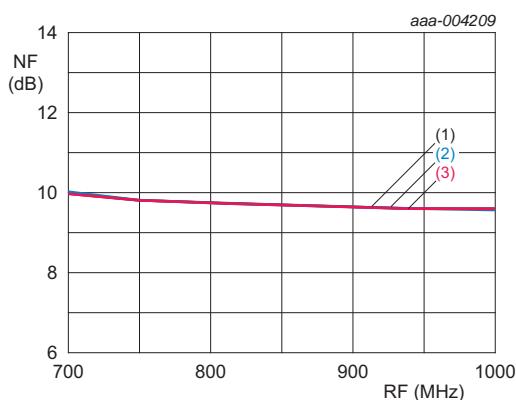
**Fig 24.**  $G_{conv}$  versus  $f_{RF}$  (low side LO) and  $V_{CC}$

**Fig 25.** IP<sub>3I</sub> versus f<sub>RF</sub> (low side LO) and T<sub>mb</sub>**Fig 26.** IP<sub>3I</sub> versus f<sub>RF</sub> (low side LO) and P<sub>i(LO)</sub>**Fig 27.** IP<sub>3I</sub> versus f<sub>RF</sub> (low side LO) and V<sub>CC</sub>



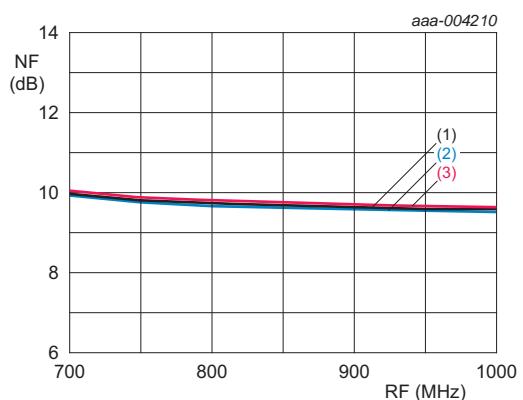
- (1)  $T_{mb} = +25 \text{ }^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40 \text{ }^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85 \text{ }^{\circ}\text{C}$ .

**Fig 28. NF versus  $f_{RF}$  (low side LO) and  $T_{mb}$**



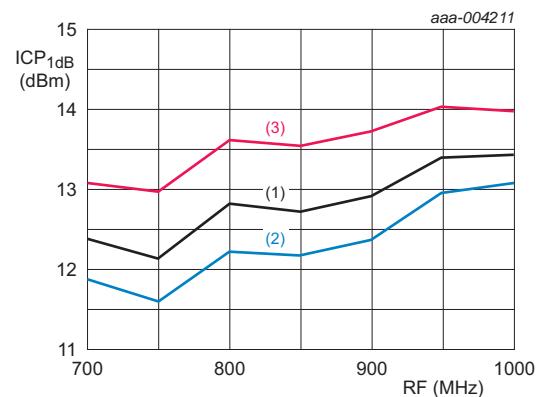
- (1)  $P_{i(LO)} = 0 \text{ dBm}$ .
- (2)  $P_{i(LO)} = -3 \text{ dBm}$ .
- (3)  $P_{i(LO)} = +3 \text{ dBm}$ .

**Fig 29. NF versus  $f_{RF}$  (low side LO) and  $P_{i(LO)}$**

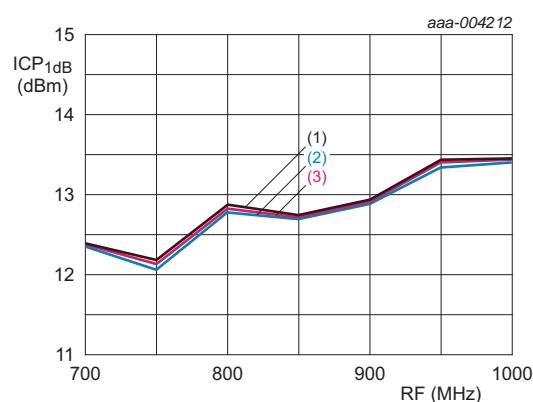


- (1)  $V_{CC} = 5 \text{ V}$ .
- (2)  $V_{CC} = 4.75 \text{ V}$ .
- (3)  $V_{CC} = 5.25 \text{ V}$ .

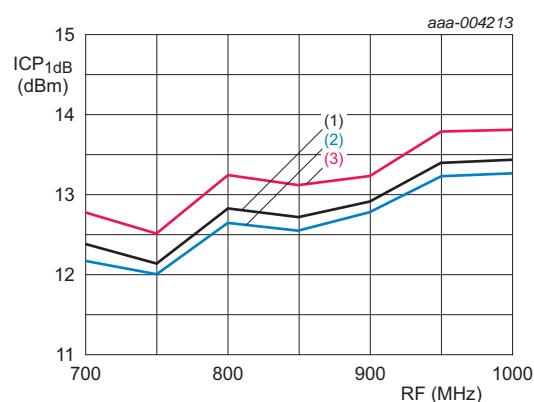
**Fig 30. NF versus  $f_{RF}$  (low side LO) and  $V_{CC}$**



- (1) T<sub>mb</sub> = +25 °C.
- (2) T<sub>mb</sub> = -40 °C.
- (3) T<sub>mb</sub> = +85 °C.

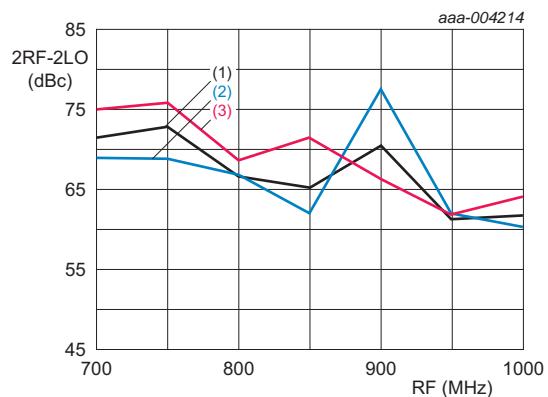
**Fig 31.** ICP<sub>1dB</sub> versus f<sub>RF</sub> (low side LO) and T<sub>mb</sub>

- (1) P<sub>i(LO)</sub> = 0 dBm.
- (2) P<sub>i(LO)</sub> = -3 dBm.
- (3) P<sub>i(LO)</sub> = +3 dBm.

**Fig 32.** ICP<sub>1dB</sub> versus f<sub>RF</sub> (low side LO) and P<sub>i(LO)</sub>

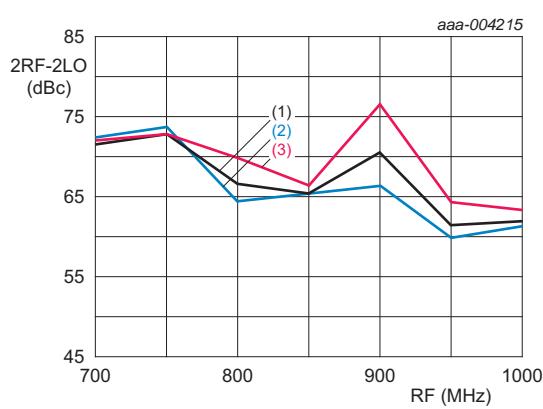
- (1) V<sub>CC</sub> = 5 V.
- (2) V<sub>CC</sub> = 4.75 V.
- (3) V<sub>CC</sub> = 5.25 V.

**Fig 33.** ICP<sub>1dB</sub> versus f<sub>RF</sub> (low side LO) and V<sub>CC</sub>



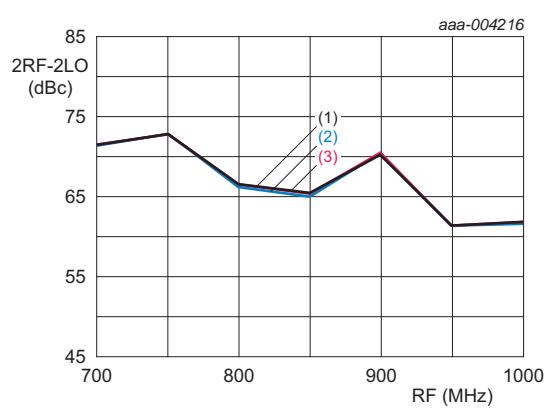
- (1)  $T_{mb} = +25^{\circ}\text{C}$ .
- (2)  $T_{mb} = -40^{\circ}\text{C}$ .
- (3)  $T_{mb} = +85^{\circ}\text{C}$ .

**Fig 34.** 2RF-2LO response versus  $f_{RF}$  (low side LO) and  $T_{mb}$



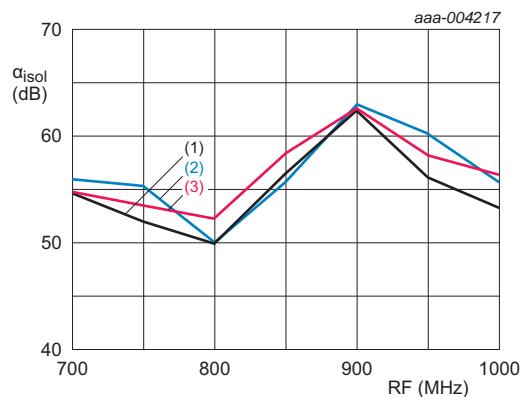
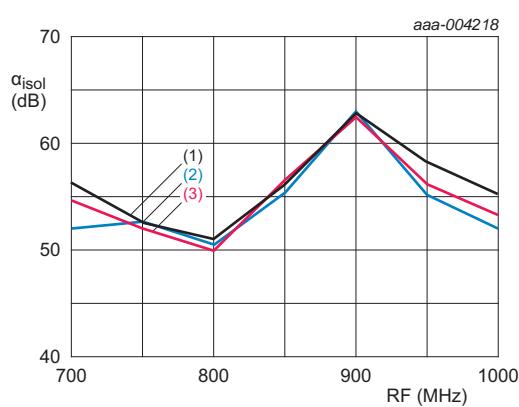
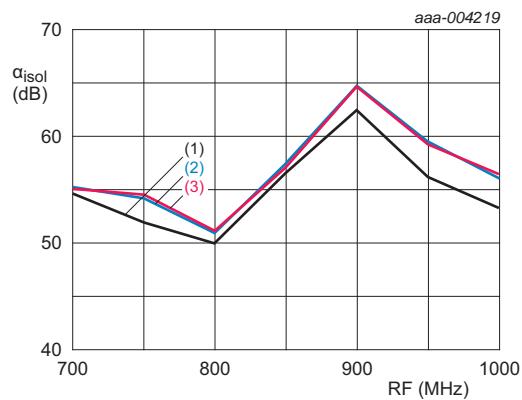
- (1)  $P_{i(LO)} = 0 \text{ dBm}$ .
- (2)  $P_{i(LO)} = -3 \text{ dBm}$ .
- (3)  $P_{i(LO)} = +3 \text{ dBm}$ .

**Fig 35.** 2RF-2LO response versus  $f_{RF}$  (low side LO) and  $P_{i(LO)}$



- (1)  $V_{CC} = 5 \text{ V}$ .
- (2)  $V_{CC} = 4.75 \text{ V}$ .
- (3)  $V_{CC} = 5.25 \text{ V}$ .

**Fig 36.** 2RF-2LO response versus  $f_{RF}$  (low side LO) and  $V_{CC}$

Fig 37.  $\alpha_{\text{isol}}$  versus  $f_{\text{RF}}$  (low side LO) and  $T_{\text{mb}}$ Fig 38.  $\alpha_{\text{isol}}$  versus  $f_{\text{RF}}$  (low side LO) and  $P_{i(\text{lo})}$ Fig 39.  $\alpha_{\text{isol}}$  versus  $f_{\text{RF}}$  (low side LO) and  $V_{\text{CC}}$

## 14. Package outline

HVQFN36: plastic thermal enhanced very thin quad flat package; no leads;  
36 terminals; body 6 x 6 x 0.85 mm

SOT1092-2

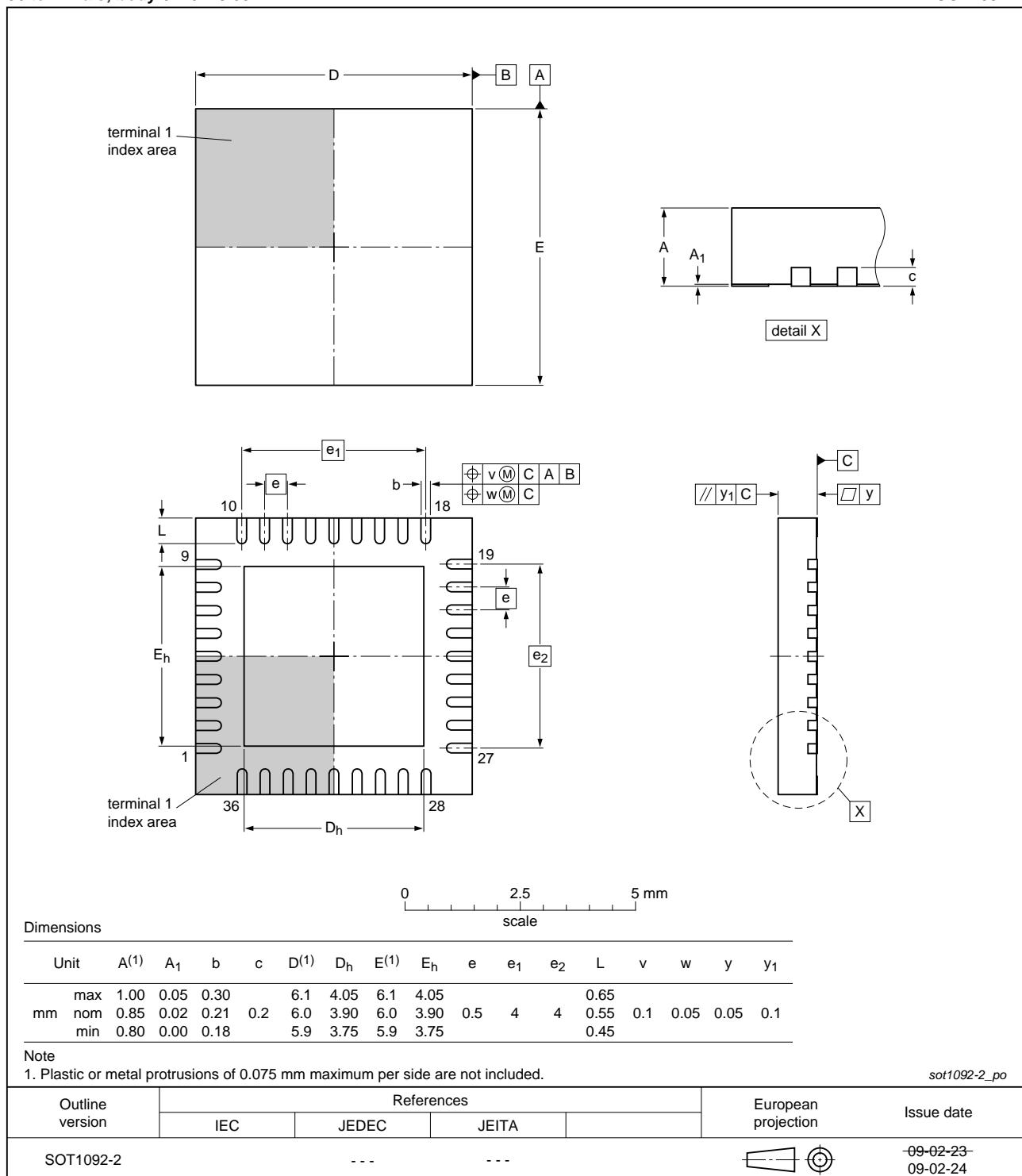


Fig 40. Package outline SOT1092-2 (HVQFN36)

## 15. Abbreviations

**Table 8. Abbreviations**

Acronym	Description
AC	Alternating Current
DC	Direct Current
ESD	ElectroStatic Discharge
FCDM	Field-induced Charged-Device Model
HBM	Human Body Model
IF	Intermediate Frequency
LO	Local Oscillator
MOS	Metal-Oxide Semiconductor
PCB	Printed-Circuit Board
RF	Radio Frequency

## 16. Revision history

**Table 9. Revision history**

Document ID	Release date	Data sheet status	Change notice	Supersedes
BGX7220 v.1	20120808	Product data sheet	-	-

## 17. Legal information

### 17.1 Data sheet status

Document status <sup>[1][2]</sup>	Product status <sup>[3]</sup>	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
Preliminary [short] data sheet	Qualification	This document contains data from the preliminary specification.
Product [short] data sheet	Production	This document contains the product specification.

[1] Please consult the most recently issued document before initiating or completing a design.

[2] The term 'short data sheet' is explained in section "Definitions".

[3] The product status of device(s) described in this document may have changed since this document was published and may differ in case of multiple devices. The latest product status information is available on the Internet at URL <http://www.nxp.com>.

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Date of release: 8 August 2012

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- При необходимости вся продукция военного и аэрокосмического назначения проходит испытания и сертификацию в лаборатории (по согласованию с заказчиком);
- Поставка специализированных компонентов военного и аэрокосмического уровня качества (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Actel, Aeroflex, Peregrine, VPT, Syfer, Eurofarad, Texas Instruments, MS Kennedy, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

Компания «Океан Электроники» является официальным дистрибутором и эксклюзивным представителем в России одного из крупнейших производителей разъемов военного и аэрокосмического назначения «JONHON», а так же официальным дистрибутором и эксклюзивным представителем в России производителя высокотехнологичных и надежных решений для передачи СВЧ сигналов «FORSTAR».



## JONHON

«JONHON» (основан в 1970 г.)

Разъемы специального, военного и аэрокосмического назначения:

(Применяются в военной, авиационной, аэрокосмической, морской, железнодорожной, горно- и нефтедобывающей отраслях промышленности)

«FORSTAR» (основан в 1998 г.)

ВЧ соединители, коаксиальные кабели,  
кабельные сборки и микроволновые компоненты:

(Применяются в телекоммуникациях гражданского и специального назначения, в средствах связи, РЛС, а так же военной, авиационной и аэрокосмической отраслях промышленности).



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