features

- **Multi-Rate Operation from 155 Mbps Up To 3.3 Gbps**
- **106-mW Power Consumption**
- **Input Offset Cancellation**
- **High Input Dynamic Range**
- **Output Disable**
- **Output Polarity Select**
- **CML Data Outputs**
- **Receive Signal Strength Indicator (RSSI)**
- **Loss of Signal Detection**
- **Single 3.3-V Supply**
- **Surface Mount Small Footprint 3 mm × 3 mm 16-Pin QFN Package**

applications

- **SONET/SDH Transmission Systems at OC3, OC12, OC24, OC48**
- **1.0625-Gbps and 2.125-Gbps Fibre Channel Receivers**
- **Gigabit Ethernet Receivers**

description

The ONET3301PA is a versatile high-speed limiting amplifier for multiple fiber optic applications with data rates up to 3.3 Gbps.

This device provides a gain of about 50 dB, which ensures a fully differential output swing for input signals as low as 3 mV_{p-p}.

The high input signal dynamic range ensures low jitter output signals even when overdriven with input signal swings as high as 1200 mV_{p-p}.

The ONET3301PA includes loss of signal detection, as well as a received signal strength indicator.

The ONET3301PA is available in a small footprint 3 mm × 3 mm 16-pin QFN package and requires a single 3.3-V supply.

This power efficient limiting amplifier typically dissipates less than 106 mW. It is characterized for operation from –40°C to 85°C.



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PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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block diagram

A simplified block diagram of the ONET3301PA is shown in Figure 1.

This compact, low power 3.3-Gbps limiting amplifier consists of a high-speed data path with offset cancellation block, a loss of signal and RSSI detection block, and a bandgap voltage reference and bias current generation block.

The limiting amplifier requires a single 3.3-V supply voltage. All circuit parts are described in detail below.

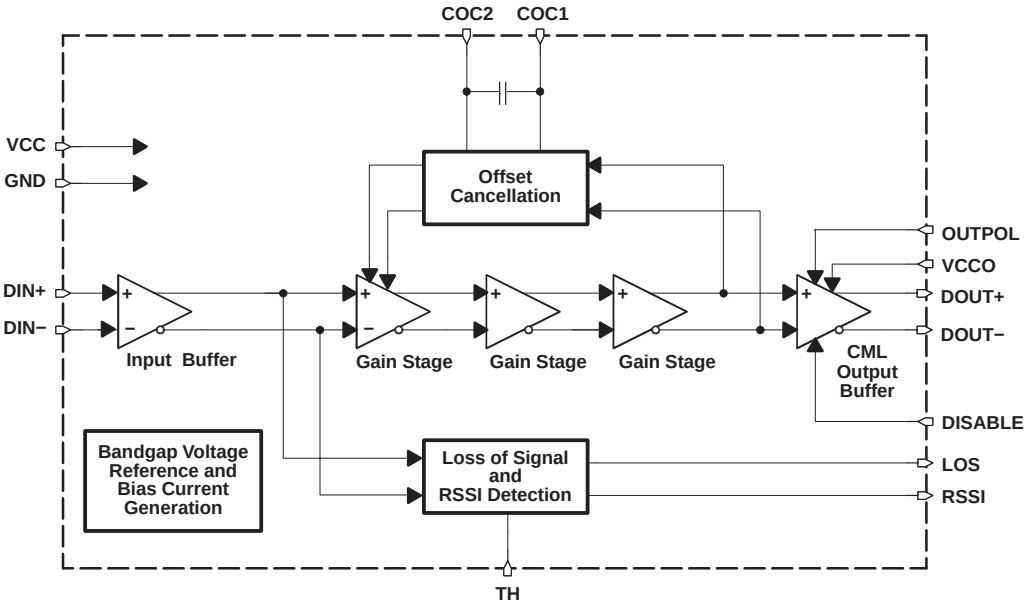


Figure 1. Block Diagram

high-speed data path

The high-speed data signal is applied to the data path by means of the input signal pins DIN+/DIN–. The data path consists of the input stage with 2 × 50-Ω on-chip line termination to VCC, three gain stages, which provide the required typical gain of about 50 dB and a CML output stage. The amplified data output signal is available at the output pins DOUT+/DOUT–, which provide 2 × 50-Ω back-termination to VCCO. The output stage also includes a data polarity switching function, which is controlled by the OUTPOL input and a disable function, controlled by the signal applied to the DISABLE input pin.

An offset cancellation compensates inevitable internal offset voltages and thus ensures proper operation even for small input data signals.

The low frequency cutoff is as low as 45 kHz with the built-in filter capacitor.

For applications, which require even lower cutoff frequencies, an additional external filter capacitor may be connected to the COC1/COC2 pins.

loss of signal and RSSI detection

The output signal of the input buffer is monitored by the loss of signal and RSSI detection circuitry. In this block a signal is generated, which is linearly proportional to the input amplitude over a wide input voltage range. This signal is available at the RSSI output pin.

Furthermore, this circuit block compares the input signal to a threshold, which can be programmed by means of an external resistor connected to the TH pin. If the input signal falls below the specified threshold, a loss of signal is indicated at the LOS pin.

The relation between the LOS assert voltage V_{AST} (in mV_{p-p}) and the external resistor R_{TH} (in kΩ) connected to the TH pin can be approximated as given below:

$$R_{TH} \approx \frac{43\ k\Omega}{V_{AST} / mV_{p-p}} - 600\ \Omega \quad (1)$$

$$V_{AST} \approx \frac{43\ mV_{p-p}}{R_{TH} / k\Omega + 0.6} \quad (2)$$

bandgap voltage and bias generation

The ONET3301PA limiting amplifier is supplied by a single 3.3-V ±10% supply voltage connected to the VCC and VCCO pins. This voltage is referred to ground (GND).

An on-chip bandgap voltage circuitry generates a supply voltage independent reference from which all other internally required voltages and bias currents are derived.

package

For the ONET3301PA, a small footprint 3 mm × 3 mm 16-pin QFN package is used with a lead pitch of 0,5 mm. The pin out is shown in Figure 2.

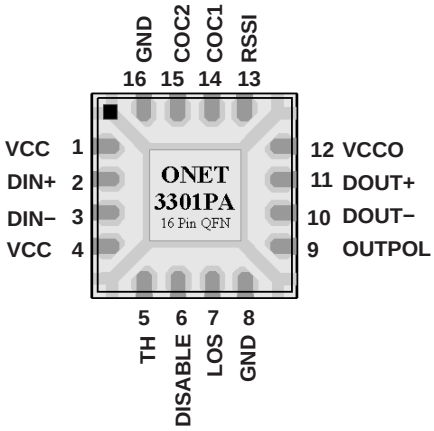


Figure 2. Pinout of ONET3301PA in a 3 mm × 3 mm 16-Pin QFN Package (Top View)

terminal functions

The following table shows a pin description for the ONET3301PA in a 3 mm x 3 mm 16-pin QFN package.

TERMINAL		TYPE	DESCRIPTION
NAME	NO.		
VCC	1, 4	Supply	3.3-V ±10% supply voltage
DIN+	2	Analog in	Noninverted data input. On-chip 50-Ω terminated to VCC
DIN–	3	Analog in	Inverted data input. On-chip 50-Ω terminated to VCC
TH	5	Analog in	LOS threshold adjustment with resistor to GND.
DISABLE	6	CMOS in	Disables CML output stage when set to high level.
LOS	7	CMOS out	High level indicates that the input signal amplitude is below the programmed threshold level.
GND	8, 16, EP	Supply	Circuit ground. Exposed die pad (EP) must be grounded.
OUTPOL	9	CMOS in	Output data signal polarity select (internally pulled up): Setting to high level or leaving pin open selects normal polarity. Low level selects inverted polarity.
DOUT–	10	CML out	Inverted data output. On-chip 50-Ω back-terminated to VCCO
DOUT+	11	CML out	Noninverted data output. On-chip 50-Ω back-terminated to VCCO
VCCO	12	Supply	3.3-V ±10% supply voltage for output stage
RSSI	13	Analog out	Analog output voltage proportional to the input data amplitude. Indicates the strength of the received signal (RSSI).
COC1	14	Analog	Offset cancellation filter capacitor terminal 1. Connect an additional filter capacitor between this pin and COC2 (pin 15). To disable the offset cancellation loop connect COC1 and COC2 (pins 14 and 15).
COC2	15	Analog	Offset cancellation filter capacitor terminal 2. Connect an additional filter capacitor between this pin and COC1 (pin 14). To disable the offset cancellation loop connect COC1 and COC2 (pins 14 and 15).

absolute maximum ratings

over operating free-air temperature range unless otherwise noted†

		VALUE	UNIT
V_{CC}, V_{CCO}	Supply voltage, See Note 1	–0.3 to 4	V
V_{DIN+}, $V_{DIN–}$	Voltage at DIN+, DIN–, See Note 1	0.5 to 4	V
V_{TH}, $V_{DISABLE}$, V_{LOS}, V_{OUTPOL}, V_{DOUT+}, $V_{DOUT–}$, V_{RSSI}, V_{COC1}, V_{COC2}	Voltage at TH, DISABLE, LOS, OUTPOL, DOUT+, DOUT–, RSSI, COC1, and COC2, See Note 1	–0.3 to 4	V
V_{COC_DIFF}	Differential voltage between COC1 and COC2	±1	V
V_{DIN_DIFF}	Differential voltage between DIN+ and DIN–	±2.5	V
I_{LOS}	Current into LOS	–1 to 9	mA
I_{DIN+}, $I_{DIN–}$, I_{DOUT+}, $I_{DOUT–}$	Continuous current at inputs and outputs	–25 to 25	mA
ESD	ESD rating at all pins	3	kV (HBM)
$T_{J(max)}$	Maximum junction temperature	125	°C
T_{stg}	Storage temperature range	−65 to 85	°C
T_A	Characterized free-air operating temperature range	−40 to 85	°C
T_L	Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260	°C

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltage values are with respect to network ground terminal.

recommended operating conditions

	MIN	TYP	MAX	UNIT
Supply voltage, V_{CC}, V_{CCO}	3	3.3	3.6	V
Operating free-air temperature, T_A	−40		85	°C

dc electrical characteristics

over recommended operating conditions (unless otherwise noted), typical operating condition is at V_{CC} = 3.3 V and T_A = 25°C

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{CC},V_{CCO}	Supply voltage		3	3.3	3.6	V
I_{CC}	Supply current	DISABLE = low (excludes CML output current)		32	40	mA
V_{OD}	Differential data output voltage swing	DISABLE = high		0.25	10	mV_{p-p}
		DISABLE = low	600	780	1200	mV_{p-p}
r_{IN}, r_{OUT}	Data input/output resistance	Single ended		50		Ω
	RSSI output voltage	Input = 2 mV_{p-p}, R_{RSSI} ≥ 10 kΩ		100		mV
		Input = 80 mV_{p-p}, R_{RSSI} ≥ 10 kΩ		2800		
	RSSI linearity	20−dB input signal, V_{IN} ≤ 80 mVpp		±3%	±8%	
$V_{(IN_MIN)}$	Data input sensitivity	BER < 10^{-10}		3	5	mV_{p-p}
$V_{(IN_MAX)}$	Data input overload		1200			mV_{p-p}
	CMOS input high voltage		2.1			V
	CMOS input low voltage				0.6	V
	LOS high voltage	I_{SINK} = –30 µA	2.4			V
	LOS low voltage	I_{SOURCE} = 1 mA			0.4	V
	LOS hysteresis	2^{23}–1 PRBS (at 2.5 Gbps and 155 Mbps)	2.5	4.5		dB
V_{TH}	LOS assert threshold range	2^{23}–1 PRBS (at 2.5 Gbps and 155 Mbps)		5−40		mV_{p-p}

ac electrical characteristics

over recommended operating conditions (unless otherwise noted), typical operating condition is at V_{CC} = 3.3 V and T_A = 25°C

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Low frequency −3-dB bandwidth	C_{OC} = open		45	70	kHz
		C_{OC} = 100 nF		0.8		
	Data rate		3.3			Gb/s
v_{NI}	Input referred noise			180		$µV_{RMS}$
DJ	Deterministic jitter, See Note 2	K28.5 pattern at 3.3 Gbps		8.5	25	ps_{p-p}
		2^{23}–1 PRBS equivalent pattern at 2.7 Gbps		9.3	30	
		K28.5 pattern at 2.1 Gbps		7.8	25	
		2^{23}–1 PRBS equivalent pattern at 155 Mbps		25	50	
RJ	Random jitter	Input = 5 mVpp		6.5		ps_{RMS}
		Input = 10 mVpp		3		
t_r	Output rise time	20% to 80%		60	85	ps
t_f	Output fall time	20% to 80%		60	85	ps
PSNR	Power supply noise rejection	f < 2 MHz	26			dB
t_{DIS}	Disable response time			20		ns
t_{LOS}	LOS assert/deassert time		2		100	µs

NOTE 2: Deterministic jitter does not include pulse-width distortion due to residual small output offset voltage.

APPLICATION INFORMATION

Figure 3 shows the ONET3301PA connected with an ac-coupled interface to the data signal source as well as to the output load.

Besides the ac-coupling capacitors, C_1 through C_4 in the input and output data signal lines, the only required external component is the LOS threshold setting resistor R_{TH}. In addition, an optional external filter capacitor (C_{OC}) may be used if a lower cutoff frequency is desired.

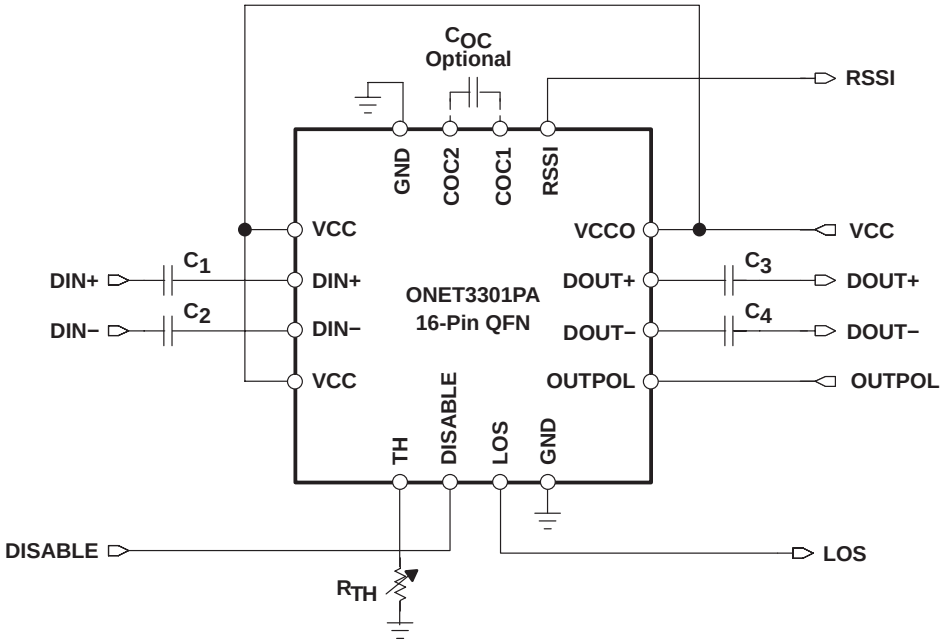


Figure 3. Basic Application Circuit With AC-Coupled I/Os

DIFFERENTIAL OUTPUT VOLTAGE vs DIFFERENTIAL INPUT VOLTAGE

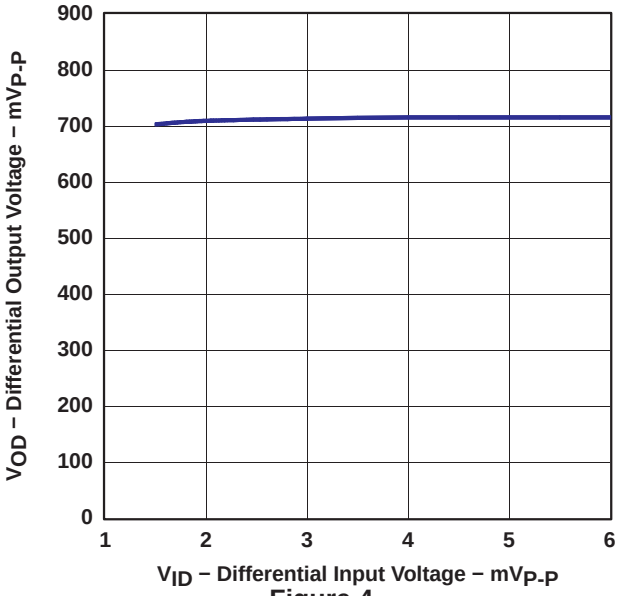


Figure 4

RANDOM JITTER vs DIFFERENTIAL INPUT VOLTAGE

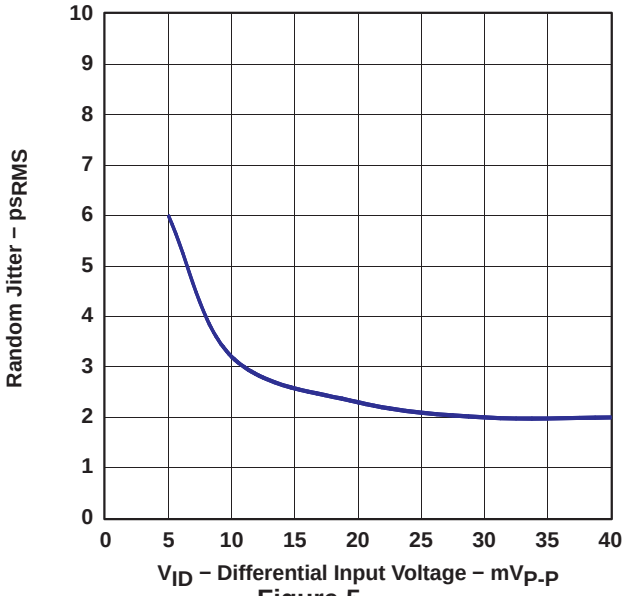


Figure 5

BIT ERROR RATIO vs DIFFERENTIAL INPUT VOLTAGE

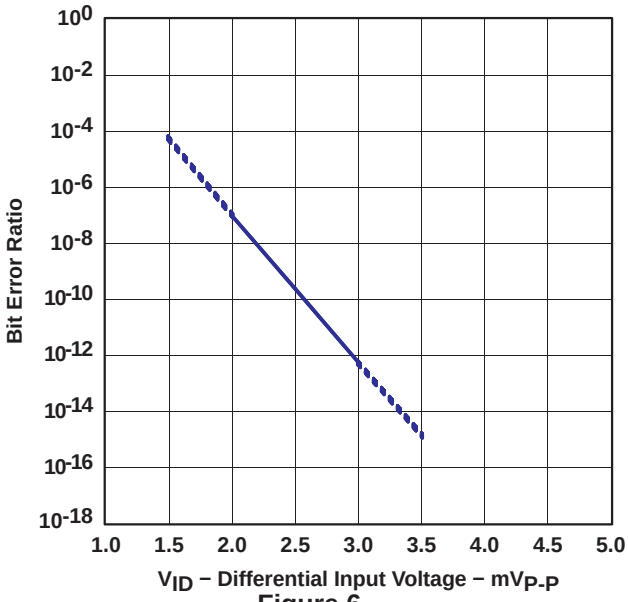


Figure 6

SMALL SIGNAL GAIN vs FREQUENCY

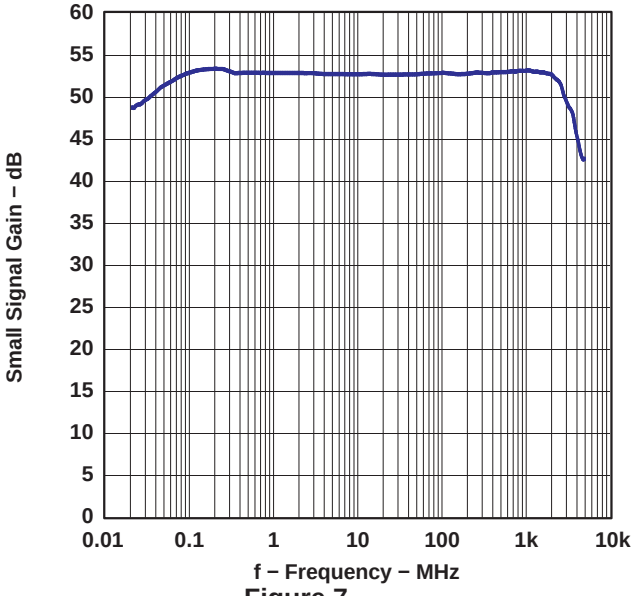


Figure 7

OUTPUT EYE-DIAGRAM AT 3.3 GBPS AND MINIMUM INPUT VOLTAGE (5 mV_{p-p})

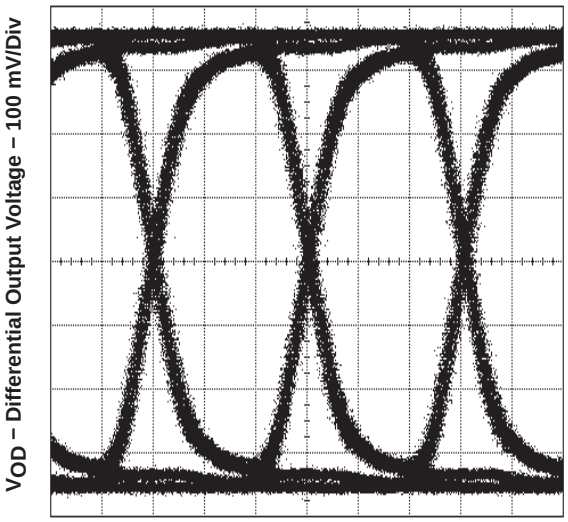


Figure 8

OUTPUT EYE-DIAGRAM AT 3.3 GBPS AND MAXIMUM INPUT VOLTAGE (1200 mV_{p-p})

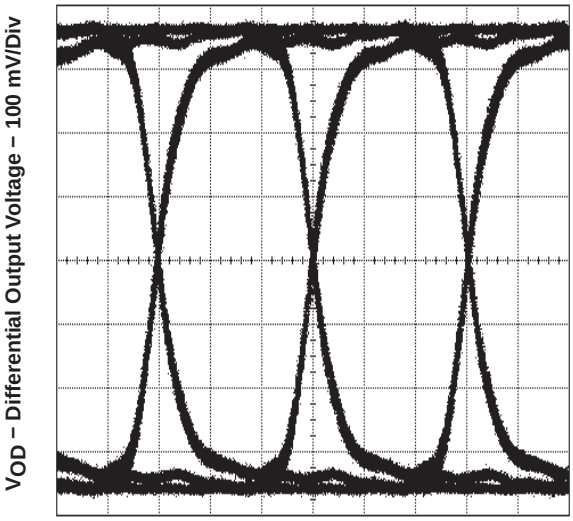


Figure 9

OUTPUT EYE-DIAGRAM AT 3.3 GBPS, 85°C, AND MINIMUM INPUT VOLTAGE (5 mV_{p-p})

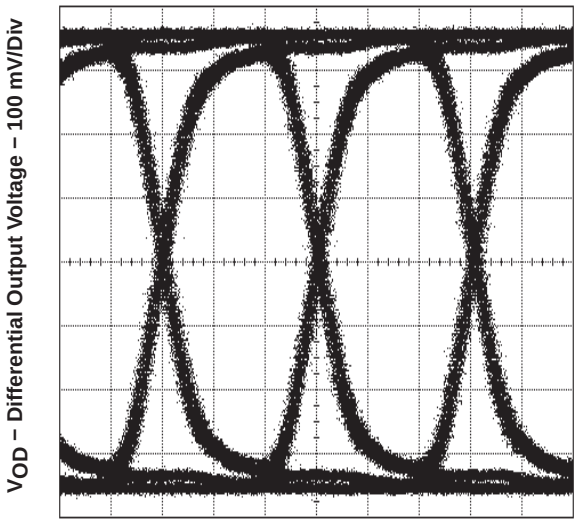


Figure 10

OUTPUT EYE-DIAGRAM AT 2.5 GBPS AND MINIMUM INPUT VOLTAGE (5 mV_{p-p})

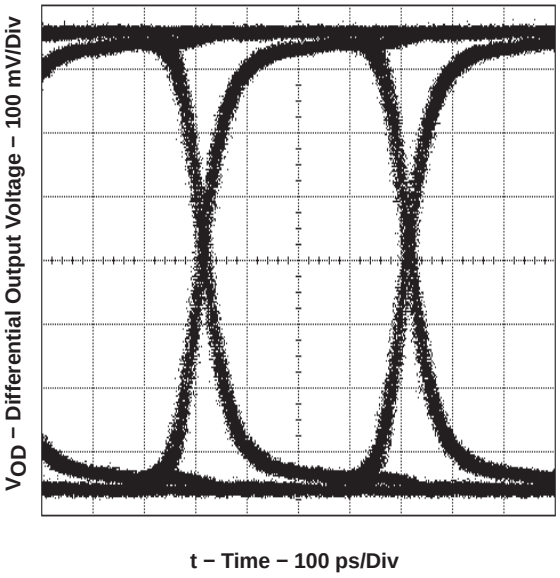


Figure 11

OUTPUT EYE-DIAGRAM AT 2.5 GBPS AND MAXIMUM INPUT VOLTAGE (1200 mV_{p-p})

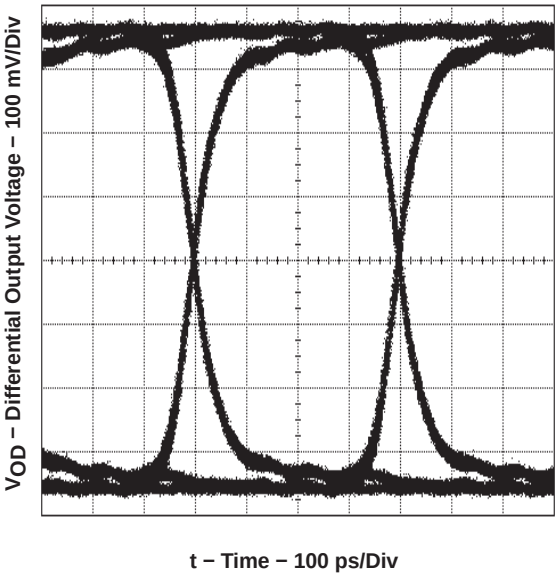


Figure 12

LOS ASSERT/DEASSERT VOLTAGE vs EXTERNAL RESISTANCE R_{TH}

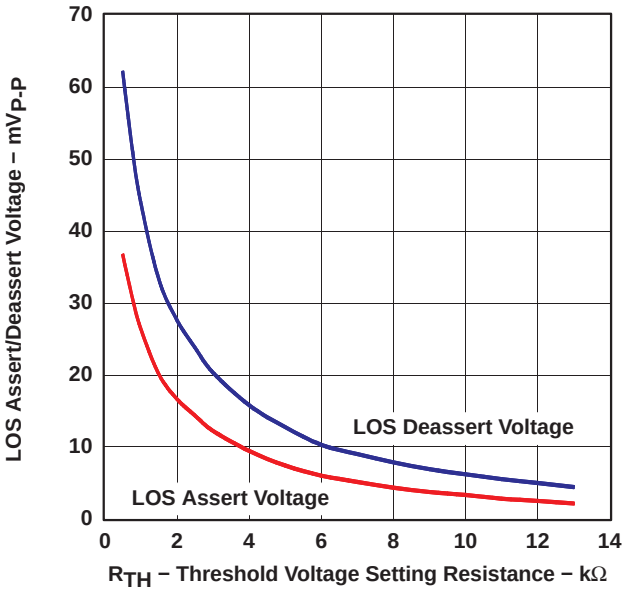


Figure 13

DIFFERENTIAL INPUT RETURN GAIN vs FREQUENCY

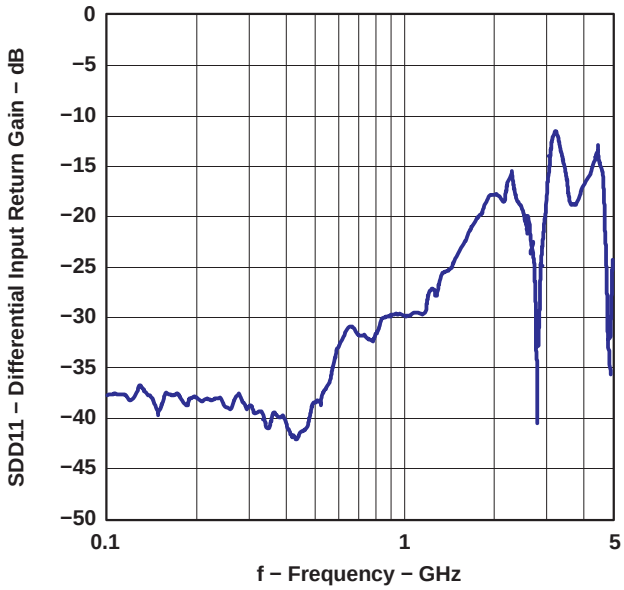


Figure 14

DIFFERENTIAL OUTPUT RETURN GAIN vs FREQUENCY

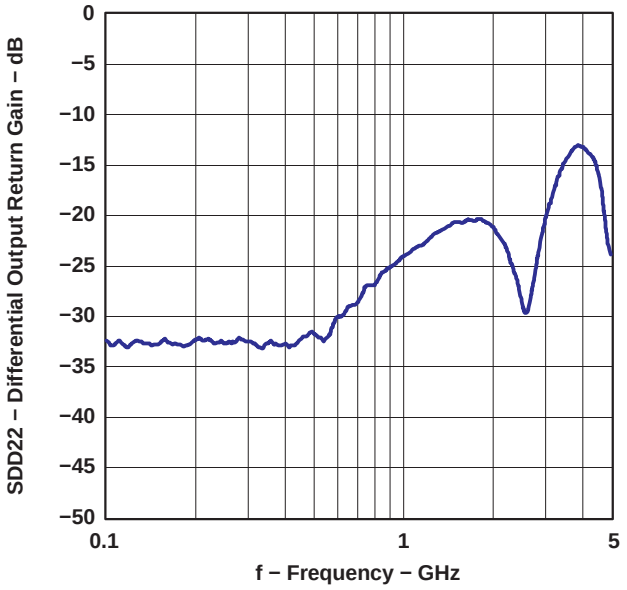


Figure 15

RECEIVE SIGNAL STRENGTH INDICATOR vs DIFFERENTIAL INPUT VOLTAGE

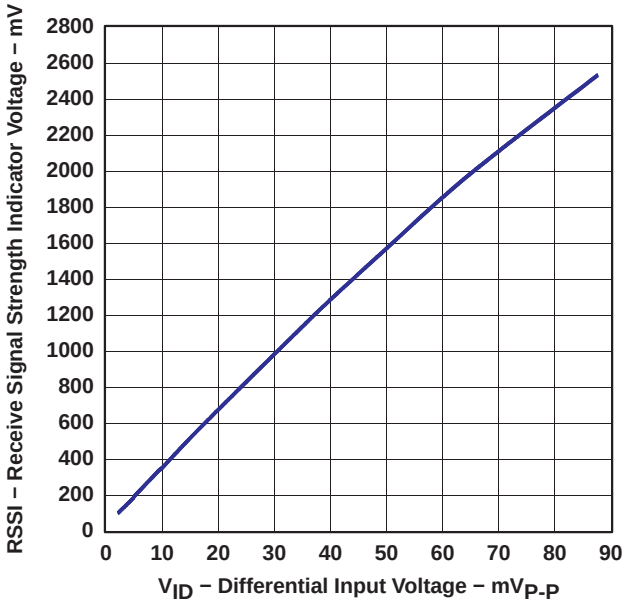


Figure 16

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)
ONET3301PARGTR	NRND	QFN	RGT	16	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
ONET3301PARGTRG4	NRND	QFN	RGT	16	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
ONET3301PARGTT	NRND	QFN	RGT	16	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
ONET3301PARGTTG4	NRND	QFN	RGT	16	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.
TBD: The Pb-Free/Green conversion plan has not been defined.
Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.
Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.
Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

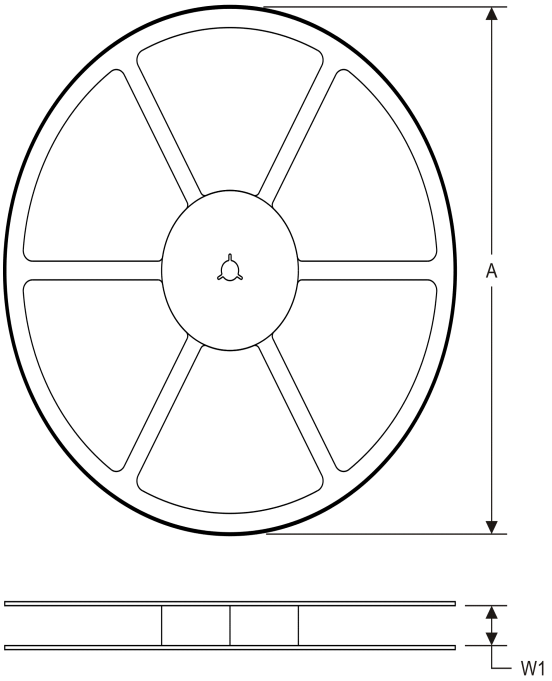
(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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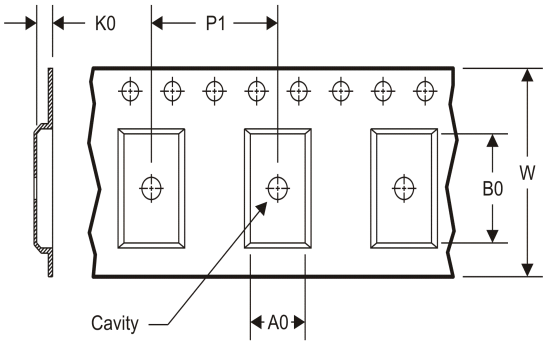
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TAPE AND REEL INFORMATION

REEL DIMENSIONS



TAPE DIMENSIONS



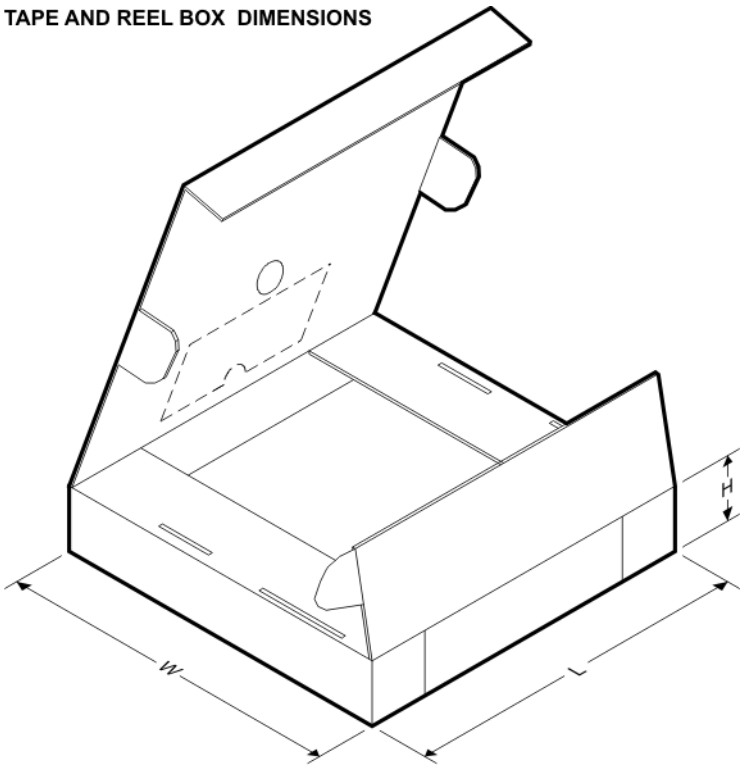
A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

TAPE AND REEL INFORMATION

*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ONET3301PARGTR	QFN	RGT	16	3000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
ONET3301PARGTT	QFN	RGT	16	250	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2

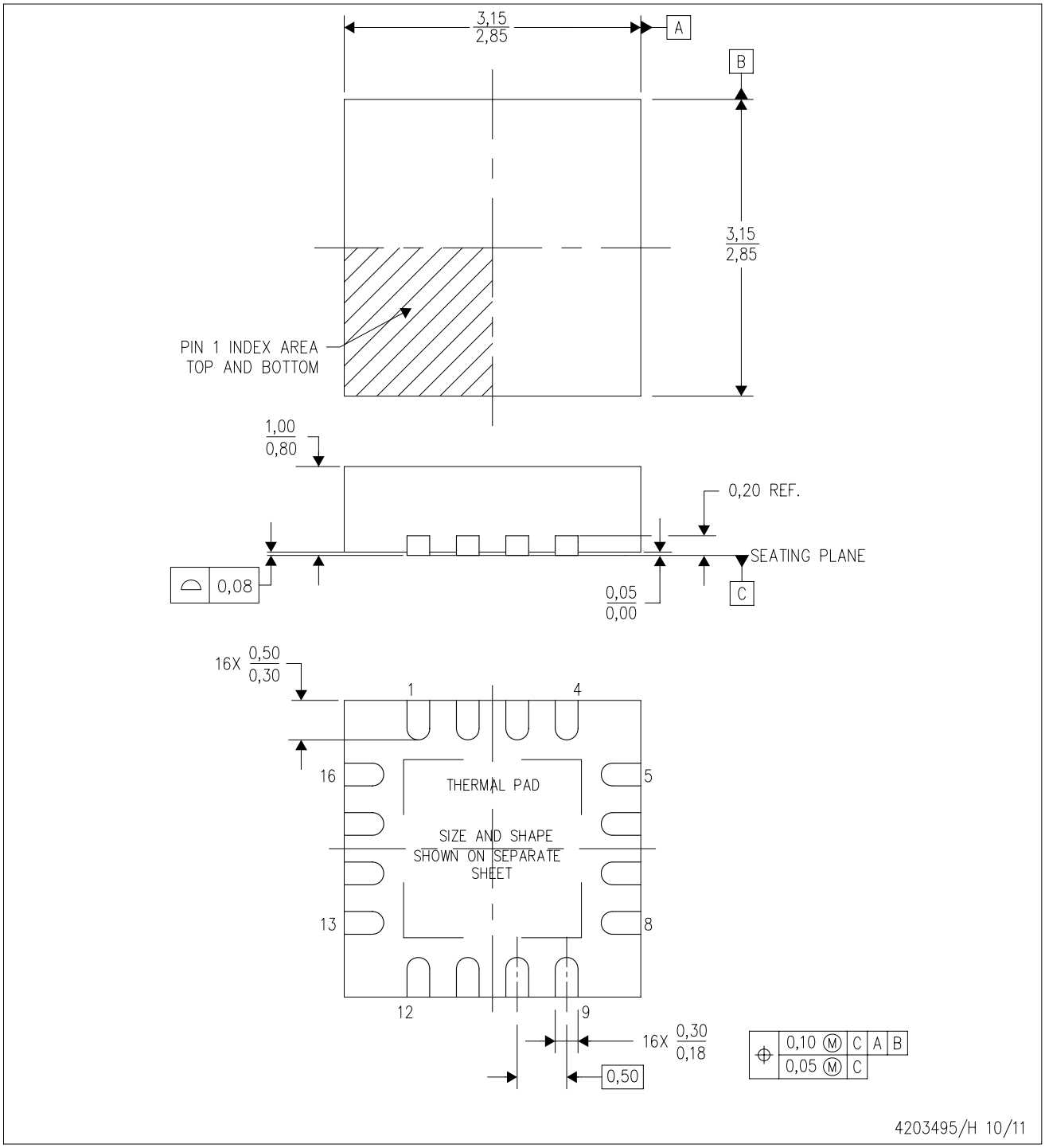
TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ONET3301PARGTR	QFN	RGT	16	3000	338.1	338.1	20.6
ONET3301PARGTT	QFN	RGT	16	250	338.1	338.1	20.6

RGT (S–PVQFN–N16) PLASTIC QUAD FLATPACK NO–LEAD



NOTES:
A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M–1994.
B. This drawing is subject to change without notice.
C. Quad Flatpack, No–leads (QFN) package configuration.
D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
F. Falls within JEDEC MO–220.

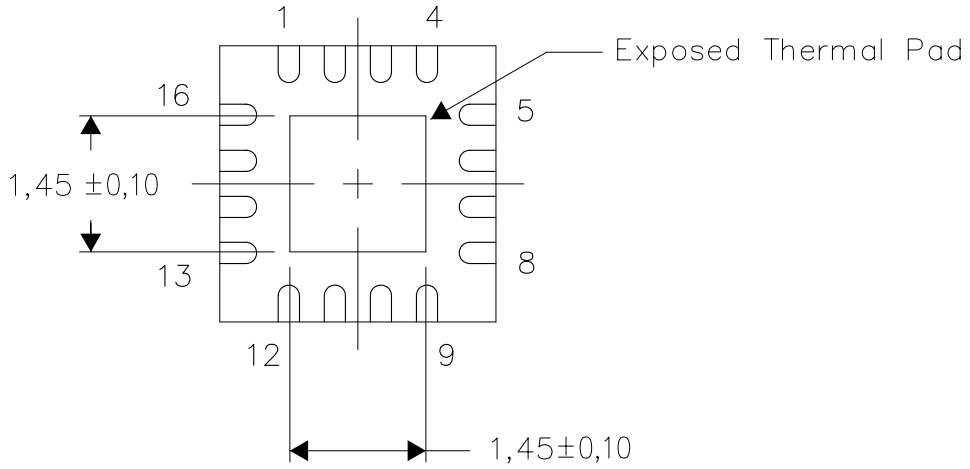
RGT (S–PVQFN–N16) PLASTIC QUAD FLATPACK NO–LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No–Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



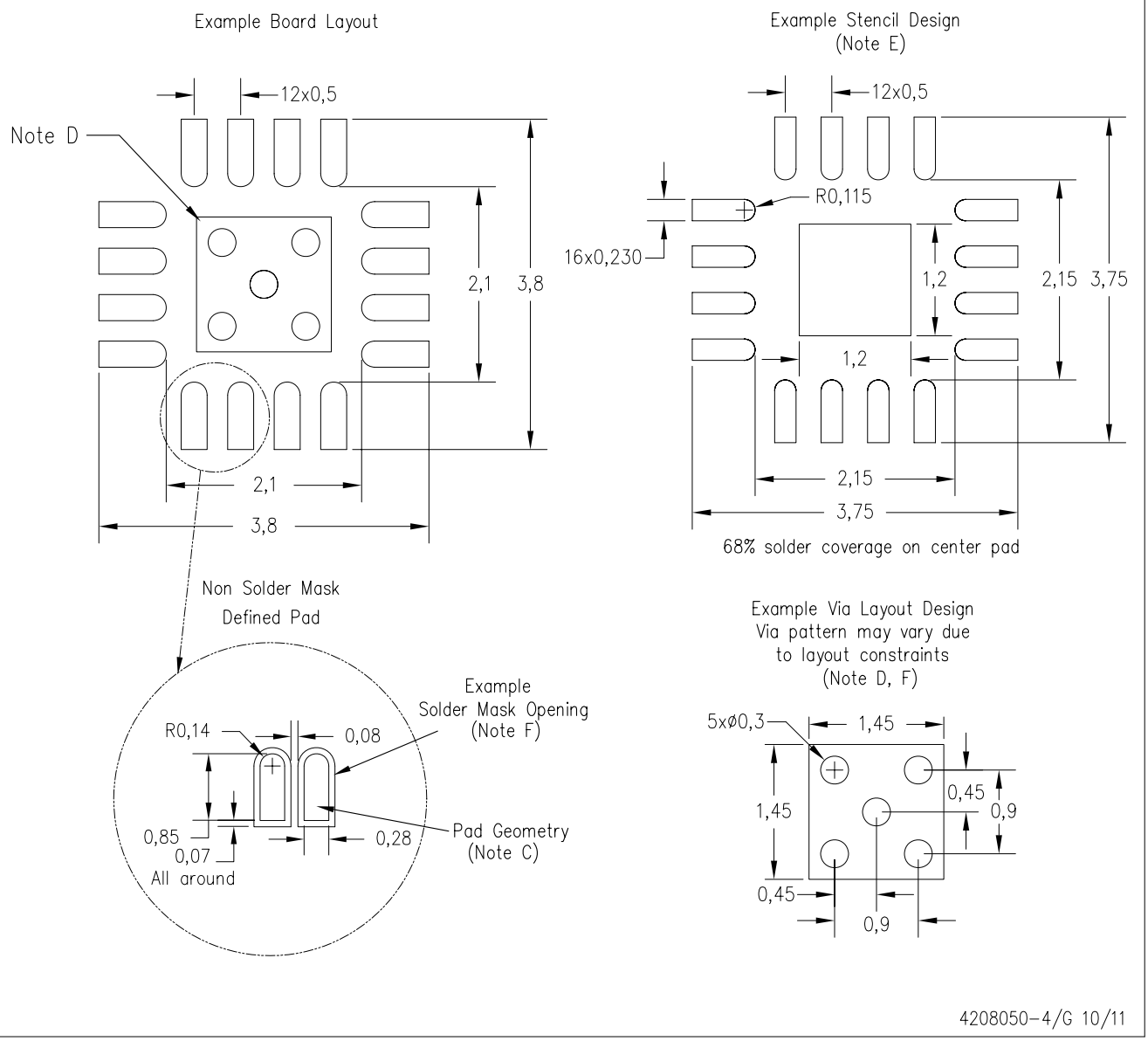
Bottom View

Exposed Thermal Pad Dimensions

4206349–2/Q 10/11

NOTE: All linear dimensions are in millimeters

RGT (S–PVQFN–N16) PLASTIC QUAD FLATPACK NO–LEAD



NOTES:
A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Publication IPC–7351 is recommended for alternate designs.
D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <http://www.ti.com>.
E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
F. Customers should contact their board fabrication site for minimum solder mask web tolerances between signal pads.

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