

High-side driver with CurrentSense analog feedback for automotive applications

Datasheet - production data



Features

Max transient supply voltage	V_{CC}	40 V
Operating voltage range	V_{CC}	4 to 28 V
Typ. on-state resistance (per Ch)	R_{ON}	8.5 m Ω
Current limitation (typ)	I_{LIMH}	96 A
Standby current (max)	I_{STBY}	0.5 μ A

- Automotive qualified
- General
 - Single channel smart high side driver with CS analog feedback
 - Very low standby current
 - Compatible with 3 V and 5 V CMOS outputs
- CS diagnostic functions
 - Analog feedback of: load current with high precision proportional current mirror
 - Overload and short to ground (power limitation) indication
 - Thermal shutdown indication
 - OFF-state open-load detection
 - Output short to V_{CC} detection
 - Sense enable/ disable
- Protections
 - Undervoltage shutdown
 - Overvoltage clamp

- Load current limitation
- Self limiting of fast thermal transients
- Configurable latch-off on overtemperature or power limitation with dedicated fault reset pin
- Loss of ground and loss of V_{CC}
- Reverse battery with external components
- Electrostatic discharge protection

Applications

- All types of Automotive resistive, inductive and capacitive loads
- Specially intended for Automotive Headlamps

Description

The device is a single channel high-side driver manufactured using ST proprietary VIPower[®] technology and housed in PowerSSO-16 package. The device is designed to drive 12 V automotive grounded loads through a 3 V and 5 V CMOS-compatible interface, and to provide protection and diagnostics.

The device integrates advanced protective functions such as load current limitation, overload active management by power limitation and overtemperature shutdown with configurable latch-off.

A FaultRST pin unlatches the output in case of fault or disables the latch-off functionality.

A sense enable pin allows OFF-state diagnosis to be disabled during the module low-power mode as well as external sense resistor sharing among similar devices.

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1 Block diagram and pin description

Figure 1: Block diagram

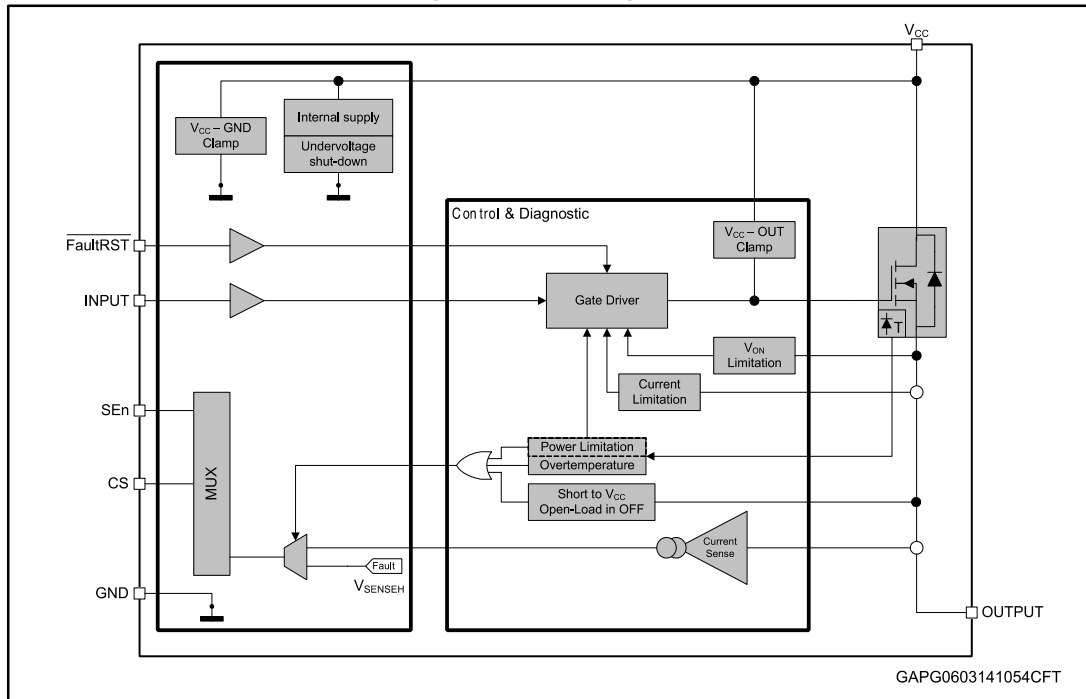
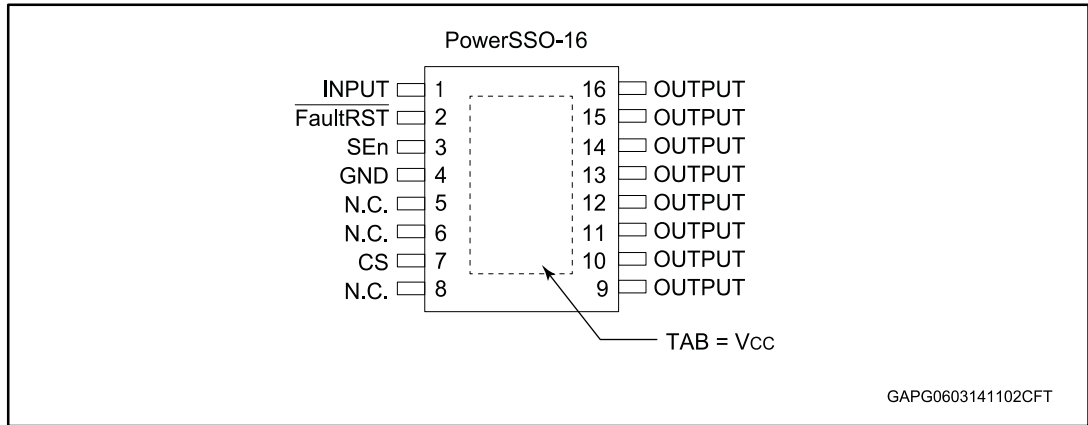


Table 1: Pin functions

Name	Function
V _{CC}	Battery connection.
OUTPUT	Power outputs. All the pins must be connected together.
GND	Ground connection. Must be reverse battery protected by an external diode / resistor network.
INPUT	Voltage controlled input pin with hysteresis, compatible with 3 V and 5 V CMOS outputs. It controls output switch state.
CS	Analog current sense output pin delivers a current proportional to the load current.
SEn	Active high compatible with 3 V and 5 V CMOS outputs pin; it enables the CS diagnostic pin.
FaultRST	Active low compatible with 3 V and 5 V CMOS outputs pin; it unlatches the output in case of fault; If kept low, sets the outputs in auto-restart mode.

Figure 2: Configuration diagram (top view)



Pins 9, 10, 11 and 12 are internally connected; Pins 13, 14, 15 and 16 are internally connected; All output pins must be connected together on PCB.

Table 2: Suggested connections for unused and not connected pins

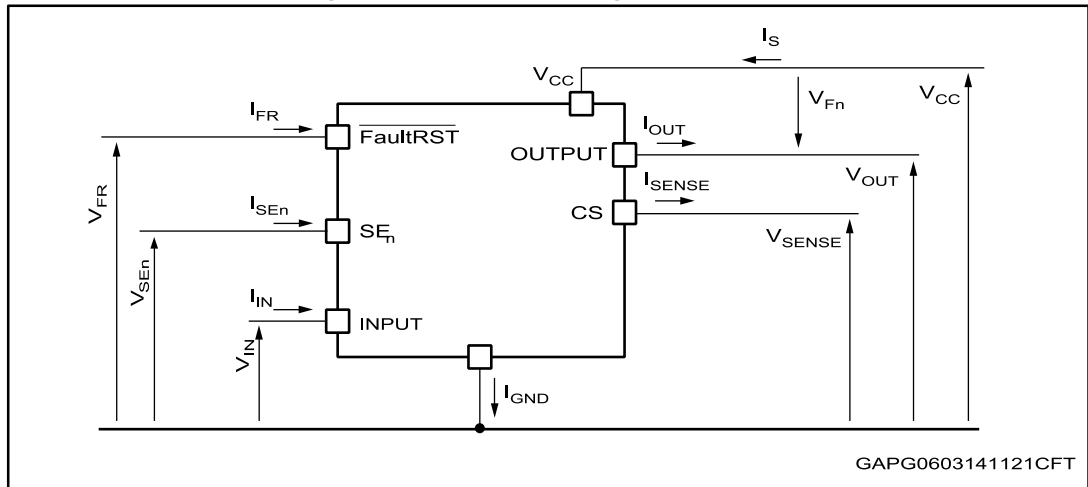
Connection / pin	CS	N.C.	Output	Input	SEn, FaultRST
Floating	Not allowed	X ⁽¹⁾	X	X	X
To ground	Through 1 kΩ resistor	X	Not allowed	Through 15 kΩ resistor	Through 15 kΩ resistor

Notes:

⁽¹⁾X: do not care.

2 Electrical specification

Figure 3: Current and voltage conventions



$V_F = V_{OUT} - V_{CC}$ during reverse battery condition.

2.1 Absolute maximum ratings

Stressing the device above the rating listed in [Table 3: "Absolute maximum ratings"](#) may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other conditions above those indicated in the operating sections of this specification is not implied. Exposure to the conditions in table below for extended periods may affect device reliability.

Table 3: Absolute maximum ratings

Symbol	Parameter	Value	Unit
V_{CC}	DC supply voltage	38	V
$-V_{CC}$	Reverse DC supply voltage	0.3	
V_{CCPK}	Maximum transient supply voltage (ISO 16750-2:2010 Test B clamped to 40V; $R_L = 4 \Omega$)	40	
V_{CCJS}	Maximum jump start voltage for single pulse short circuit protection	28	
$-I_{GND}$	DC reverse ground pin current	200	mA
I_{OUT}	OUTPUT DC output current	Internally limited	A
$-I_{OUT}$	Reverse DC output current	35	
I_{IN}	INPUT DC input current	-1 to 10	mA
I_{SEn}	SEn DC input current		
I_{FR}	FaultRST DC input current		
V_{FR}	FaultRST DC input voltage	7.5	V
I_{SENSE}	CS pin DC output current ($V_{GND} = V_{CC}$ and $V_{SENSE} < 0 V$)	10	mA
	CS pin DC output current in reverse ($V_{CC} < 0V$)	-20	

Symbol	Parameter	Value	Unit
E_{MAX}	Maximum switching energy (single pulse) ($T_{DEMAG} = 0.4 \text{ ms}$; $T_{jstart} = 150 \text{ }^\circ\text{C}$)	170	mJ
V_{ESD}	Electrostatic discharge (JEDEC 22A-114F)		
	• INPUT	4000	V
	• CS	2000	V
	• SEn, FaultRST	4000	V
	• OUTPUT	4000	V
	• V_{CC}	4000	V
V_{ESD}	Charge device model (CDM-AEC-Q100-011)	750	V
T_j	Junction operating temperature	-40 to 150	$^\circ\text{C}$
T_{stg}	Storage temperature	-55 to 150	

2.2 Thermal data

Table 4: Thermal data

Symbol	Parameter	Typ. value	Unit
$R_{thj-board}$	Thermal resistance junction-board (JEDEC JESD 51-5 / 51-8) ⁽¹⁾	3.85	$^\circ\text{C/W}$
$R_{thj-amb}$	Thermal resistance junction-ambient (JEDEC JESD 51-5) ⁽²⁾	54.8	
$R_{thj-amb}$	Thermal resistance junction-ambient (JEDEC JESD 51-7) ⁽¹⁾	21	

Notes:

⁽¹⁾Device mounted on four-layers 2s2p PCB.

⁽²⁾Device mounted on two-layers 2s0p PCB with 2 cm² heatsink copper trace.

2.3 Main electrical characteristics

$7 \text{ V} < V_{CC} < 18 \text{ V}$; $-40 \text{ }^\circ\text{C} < T_j < 150 \text{ }^\circ\text{C}$, unless otherwise specified.

All typical values refer to $V_{CC} = 13 \text{ V}$; $T_j = 25 \text{ }^\circ\text{C}$, unless otherwise specified.

Table 5: Power section

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{CC}	Operating supply voltage		4	13	28	V
V_{USD}	Undervoltage shutdown				4	
$V_{USDReset}$	Undervoltage shutdown reset				5	
$V_{USDhyst}$	Undervoltage shutdown hysteresis			0.3		
R_{ON}	On-state resistance	$I_{OUT} = 5 \text{ A}$; $T_j = 25 \text{ }^\circ\text{C}$		8.5		m Ω
		$I_{OUT} = 5 \text{ A}$; $T_j = 150 \text{ }^\circ\text{C}$			17	
		$I_{OUT} = 5 \text{ A}$; $V_{CC} = 4 \text{ V}$; $T_j = 25 \text{ }^\circ\text{C}$			12.75	
V_{clamp}	Clamp voltage	$I_s = 20 \text{ mA}$; $T_j = -40 \text{ }^\circ\text{C}$	38			V
		$I_s = 20 \text{ mA}$; $25 \text{ }^\circ\text{C} < T_j < 150 \text{ }^\circ\text{C}$	41	46	52	

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
I _{STBY}	Supply current in standby at V _{CC} = 13 V ⁽¹⁾	V _{CC} = 13 V; V _{IN} = V _{OUT} = V _{FR} = V _{SEn} = 0 V; T _j = 25 °C			0.5	μA
		V _{CC} = 13 V; V _{IN} = V _{OUT} = V _{FR} = V _{SEn} = 0 V; T _j = 85 °C ⁽²⁾			0.5	μA
		V _{CC} = 13 V; V _{IN} = V _{OUT} = V _{FR} = V _{SEn} = 0 V; T _j = 125 °C			3	μA
t _{D_STBY}	Standby mode blanking time	V _{CC} = 13 V; V _{IN} = 5 V; V _{SEn} = V _{FR} = 0 V; I _{OUT} = 0 A	60	300	550	μs
I _{S(ON)}	Supply current	V _{CC} = 13 V; V _{SEn} = V _{FR} = 0 V; V _{IN} = 5 V; I _{OUT} = 0 A		3	5	mA
I _{GND(ON)}	Control stage current consumption in ON state. All channels active.	V _{CC} = 13 V; V _{SEn} = 5 V; V _{FR} = 0 V; V _{IN} = 5 V; I _{OUT} = 5 A			6	mA
I _{L(off)}	Off-state output current at V _{CC} = 13 V	V _{IN} = V _{OUT} = 0 V; V _{CC} = 13 V; T _j = 25 °C	0	0.01	0.5	μA
		V _{IN} = V _{OUT} = 0 V; V _{CC} = 13 V; T _j = 125 °C	0		3	
V _F	Output - V _{CC} diode voltage	I _{OUT} = -5 A; T _j = 150 °C			0.7	V

Notes:

(1)PowerMOS leakage included.

(2)Parameter specified by design; not subject to production test.

Table 6: Switching

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{CC} = 13 V; -40°C < T_j < 150°C, unless otherwise specified						
t _{d(on)} ⁽¹⁾	Turn-on delay time at T _j = 25°C	R _L = 2.6 Ω	10	45	120	μs
t _{d(off)} ⁽¹⁾	Turn-off delay time at T _j = 25°C		10	48	100	
(dV _{OUT} /dt) _{on} ⁽¹⁾	Turn-on voltage slope at T _j = 25°C	R _L = 2.6 Ω	0.1	0.2	0.7	V/μs
(dV _{OUT} /dt) _{off} ⁽¹⁾	Turn-off voltage slope at T _j = 25°C		0.1	0.3	0.7	
W _{ON}	Switching energy losses at turn-on (t _{won})	R _L = 2.6 Ω	—	0.8	1.2 ⁽²⁾	mJ
W _{OFF}	Switching energy losses at turn-off (t _{woff})	R _L = 2.6 Ω	—	0.6	1 ⁽²⁾	mJ
t _{SKW} ⁽¹⁾	Differential Pulse skew (t _{PHL} - t _{PLH})	R _L = 2.6 Ω	-65	-15	35	μs

Notes:(1)See [Figure 6: "Switching times and Pulse skew"](#).

(2)Parameter guaranteed by design and characterization; not subject to production test.

Table 7: Logic Inputs

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
7 V < V_{CC} < 28 V; -40°C < T_j < 150°C						
INPUT characteristics						
V _{IL}	Input low level voltage				0.9	V
I _{IL}	Low level input current	V _{IN} = 0.9 V	1			μA
V _{IH}	Input high level voltage		2.1			V
I _{IH}	High level input current	V _{IN} = 2.1 V			10	μA
V _{I(hyst)}	Input hysteresis voltage		0.2			V
V _{ICL}	Input clamp voltage	I _{IN} = 1 mA	5.3		7.5	V
		I _{IN} = -1 mA		-0.7		
FaultRST characteristics						
V _{FRL}	Input low level voltage				0.9	V
I _{FRL}	Low level input current	V _{IN} = 0.9 V	1			μA
V _{FRH}	Input high level voltage		2.1			V
I _{FRH}	High level input current	V _{IN} = 2.1 V			10	μA
V _{FR(hyst)}	Input hysteresis voltage		0.2			V
V _{FRCL}	Input clamp voltage	I _{IN} = 1 mA	5.3		7.2	V
		I _{IN} = -1 mA		-0.7		
SEn characteristics (7 V < V_{CC} < 18 V)						
V _{SEnL}	Input low level voltage				0.9	V
I _{SEnL}	Low level input current	V _{IN} = 0.9 V	1			μA
V _{SEnH}	Input high level voltage		2.1			V
I _{SEnH}	High level input current	V _{IN} = 2.1 V			10	μA
V _{SEn(hyst)}	Input hysteresis voltage		0.2			V
V _{SEnCL}	Input clamp voltage	I _{IN} = 1 mA	5.3		7.5	V
		I _{IN} = -1 mA		-0.7		

Table 8: Protections

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
7 V < V_{CC} < 18 V; -40°C < T_j < 150°C						
I _{LIMH}	DC short circuit current	V _{CC} = 13 V	70	98	140	A
		4 V < V _{CC} < 18 V ⁽¹⁾			140	
I _{LIML}	Short circuit current during thermal cycling	V _{CC} = 13 V; T _R < T _j < T _{TSD}		33		
T _{TSD}	Shutdown temperature		150	175	200	°C
T _R	Reset temperature ⁽¹⁾		T _{RS} + 1	T _{RS} + 7		
T _{RS}	Thermal reset of fault diagnostic indication	V _{FR} = 0 V; V _{SEn} = 5 V;	135			
T _{HYST}	Thermal hysteresis (T _{TSD} - T _R) ⁽¹⁾			7		

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
7 V < V_{CC} < 18 V; -40°C < T_j < 150°C						
ΔT_{J_SD}	Dynamic temperature	T _j = -40 °C; V _{CC} = 13 V		60		K
t _{LATCH_RST}	Fault reset time for output unlatch ⁽¹⁾	V _{FR} = 5 V to 0 V; V _{SEN} = 5 V; V _{IN} = 5 V	3	10	20	μs
V _{DEMAG}	Turn-off output voltage clamp	I _{OUT} = 2 A; L = 6 mH; T _j = -40 °C	V _{CC} - 38			V
		I _{OUT} = 2 A; L = 6 mH; T _j = 25 °C to 150 °C	V _{CC} - 41	V _{CC} - 46	V _{CC} - 52	V
V _{ON}	Output voltage drop limitation	I _{OUT} = 0.25 A		20		mV

Notes:

⁽¹⁾Parameter guaranteed by design and characterization; not subject to production test.

Table 9: CurrentSense

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
7 V < V_{CC} < 18 V; -40°C < T_j < 150°C						
V _{SENSE_CL}	Current sense clamp voltage	V _{SEN} = 0 V; I _{SENSE} = 1 mA	-17		-12	V
		V _{SEN} = 0 V; I _{SENSE} = -1 mA		7		V
Current Sense characteristics						
K ₀	I _{OUT} /I _{SENSE}	I _{OUT} = 0.9 A; V _{SENSE} = 0.5 V; V _{SEN} = 5 V	3465	6150	9135	
dK ₀ /K ₀ ⁽¹⁾⁽²⁾	Current sense ratio drift	I _{OUT} = 0.9 A; V _{SENSE} = 0.5 V; V _{SEN} = 5 V	-20		20	%
K ₁	I _{OUT} /I _{SENSE}	I _{OUT} = 1.5 A; V _{SENSE} = 4 V; V _{SEN} = 5 V	3735	5990	8725	
dK ₁ /K ₁ ⁽²⁾⁽¹⁾	Current sense ratio drift	I _{OUT} = 1.5 A; V _{SENSE} = 4 V; V _{SEN} = 5 V	-15		15	%
K ₂	I _{OUT} /I _{SENSE}	I _{OUT} = 6 A; V _{SENSE} = 4 V; V _{SEN} = 5 V	4410	5890	7360	
dK ₂ /K ₂ ⁽²⁾⁽¹⁾	Current sense ratio drift	I _{OUT} = 6 A; V _{SENSE} = 4 V; V _{SEN} = 5 V	-10		+10	%
K ₃	I _{OUT} /I _{SENSE}	I _{OUT} = 18 A; V _{SENSE} = 4 V; V _{SEN} = 5 V	5290	5880	6470	
dK ₃ /K ₃ ⁽²⁾⁽¹⁾	Current sense ratio drift	I _{OUT} = 18 A; V _{SENSE} = 4 V; V _{SEN} = 5 V; T _j = -40 °C to 150 °C	-5		5	%
V _{OUT_MSD} ⁽²⁾	Output Voltage for current sense shutdown	V _{IN} = 5 V; V _{SEN} = 5 V; R _{SENSE} = 2.7 kΩ; I _{OUT} = 5 A		5		V
V _{SENSE_SAT}	CS saturation voltage	V _{CC} = 7 V; R _{SENSE} = 2.7 kΩ; V _{SEN} = 5 V; V _{IN} = 5 V; I _{OUT} = 18 A; T _j = 150°C	5			V
I _{SENSE_SAT} ⁽²⁾	CS saturation current	V _{CC} = 7 V; V _{SENSE} = 4 V; V _{IN} = 5 V; V _{SEN} = 5 V; T _j = 150°C	4			mA

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
7 V < V_{CC} < 18 V; -40°C < T_j < 150°C						
I _{OUT_SAT} ⁽²⁾	Output saturation current	V _{CC} = 7 V; V _{SENSE} = 4 V; V _{IN} = 5 V; V _{SEn} = 5 V; T _j = 150°C	24			A
I _{SENSE0}	Current sense leakage current	CS disabled: V _{SEn} = 0 V;	0		0.5	μA
		CS disabled: -1 V < V _{SENSE} < 5 V ⁽²⁾	-0.5		0.5	
		CS enabled: V _{SEn} = 5 V; Channel ON; I _{OUT} = 0 A; Diagnostic selected; V _{IN} = 5 V; I _{OUT} = 0 A;	0		2	
		CS enabled: V _{SEn} = 5 V; Channel OFF; Diagnostic selected: V _{IN} = 0 V	0		2	
OFF-state diagnostic						
V _{OL}	OFF-state open-load voltage detection threshold	V _{IN} = 0 V; V _{SEn} = 5 V	2	3	4	V
I _{L(off2)}	OFF-state output sink current	V _{IN} = 0 V; V _{OUT} = V _{OL} ; T _j = -40°C to 125°C	-100		-15	μA
t _{DSTKON}	OFF-state diagnostic delay time from falling edge of INPUT (see Figure 8: "TDSTKON")	V _{IN} = 5 V to 0 V; V _{SEn} = 5 V; I _{OUT} = 0 A; V _{OUT} = 4 V	100	350	700	μs
t _{D_OL_V}	Settling time for valid OFF-state open load diagnostic indication from rising edge of SEN	V _{IN} = 0 V; V _{FR} = 0 V; V _{OUT} = 4 V; V _{SEn} = 0 V to 5 V			60	μs
t _{D_VOL}	OFF-state diagnostic delay time from rising edge of V _{OUT}	V _{IN} = 0 V; V _{SEn} = 5 V; V _{OUT} = 0 V to 4 V		5	30	μs
Fault diagnostic feedback (see Table 10: "Truth table")						
V _{SENSEH}	Current sense output voltage in fault condition	V _{CC} = 13 V; V _{IN} = 0 V; V _{SEn} = 5 V; I _{OUT} = 0 A; V _{OUT} = 4 V; R _{SENSE} = 1 kΩ	5		6.6	V
I _{SENSEH}	Current sense output current in fault condition	V _{CC} = 13 V; V _{SENSE} = 5 V	7	20	30	mA
Current sense timings (current sense mode - see Figure 7: "CurrentSense timings (current sense mode)")						
t _{DSENSE1H}	Current sense settling time from rising edge of SEN	V _{IN} = 5 V; V _{SEn} = 0 V to 5 V; R _{SENSE} = 1 kΩ; R _L = 2.6 Ω			60	μs
t _{DSENSE1L}	Current sense disable delay time from falling edge of SEN	V _{IN} = 5 V; V _{SEn} = 5 V to 0 V; R _{SENSE} = 1 kΩ; R _L = 2.6 Ω		5	20	μs
t _{DSENSE2H}	Current sense settling time from rising edge of INPUT	V _{IN} = 0 V to 5 V; V _{SEn} = 5 V; R _{SENSE} = 1 kΩ; R _L = 2.6 Ω		100	350	μs
Δt _{DSENSE2H}	Current sense settling time from rising edge of I _{OUT} (dynamic response to a step change of I _{OUT})	V _{IN} = 5 V; V _{SEn} = 5 V; R _{SENSE} = 1 kΩ; I _{SENSE} = 90 % of I _{SENSEMAX} ; R _L = 2.6 Ω			150	μs

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
7 V < V_{CC} < 18 V; -40°C < T_j < 150°C						
t _{DSENSE2L}	Current sense turn-off delay time from falling edge of INPUT	V _{IN} = 5 V to 0 V; V _{SEn} = 5 V; R _{SENSE} = 1 kΩ; R _L = 2.6 Ω		50	250	μs

Notes:

- (1) All values refer to V_{CC} = 13 V; T_j = 25°C, unless otherwise specified.
- (2) Parameter specified by design; not subject to production test.

Figure 4: IO_{UT}/I_{SENSE} versus IO_{UT}

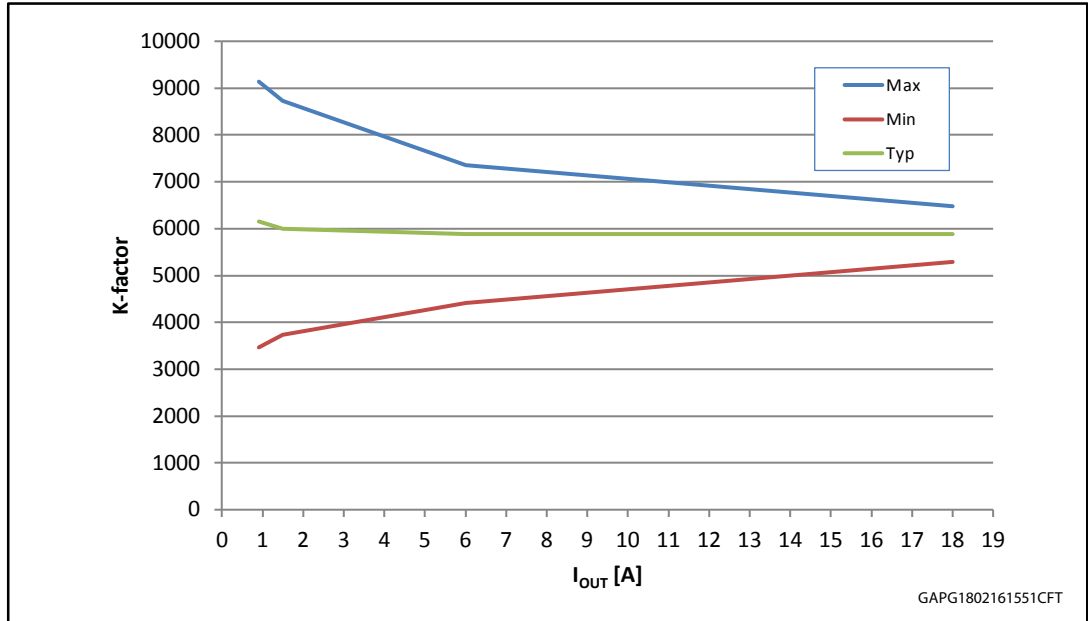


Figure 5: Current sense accuracy versus IO_{UT}

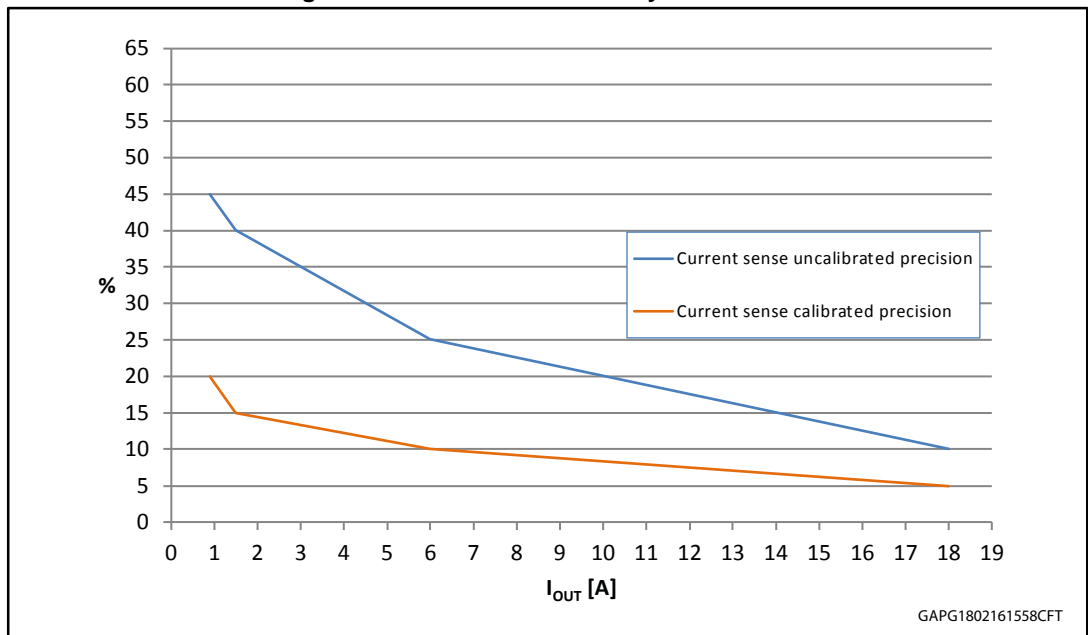


Figure 6: Switching times and Pulse skew

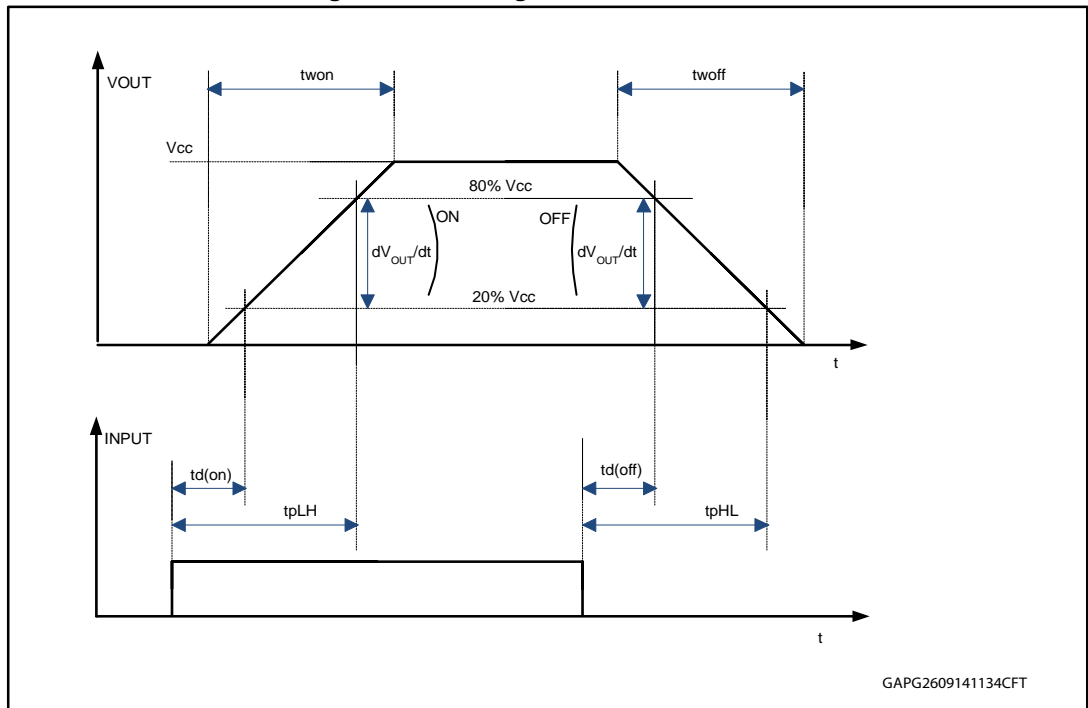


Figure 7: CurrentSense timings (current sense mode)

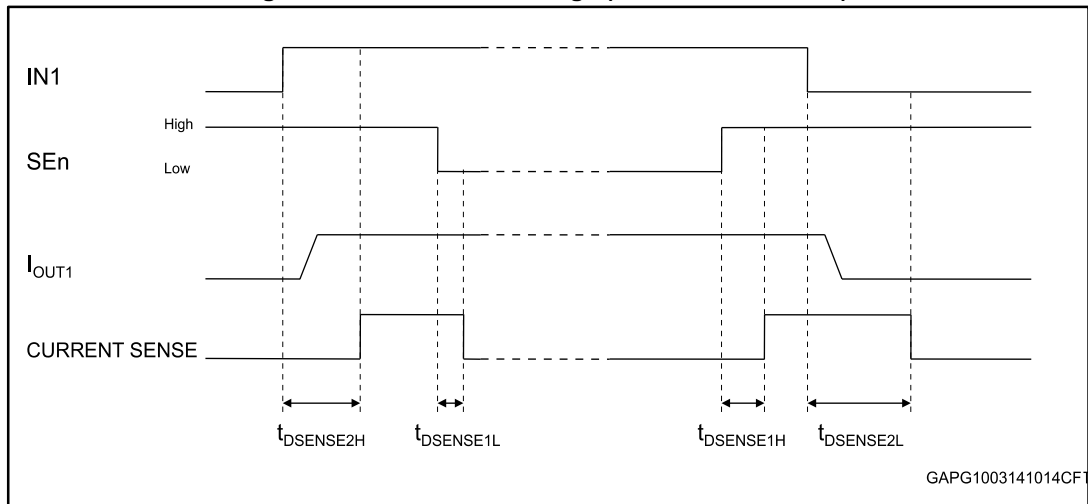


Figure 8: TDSTKON

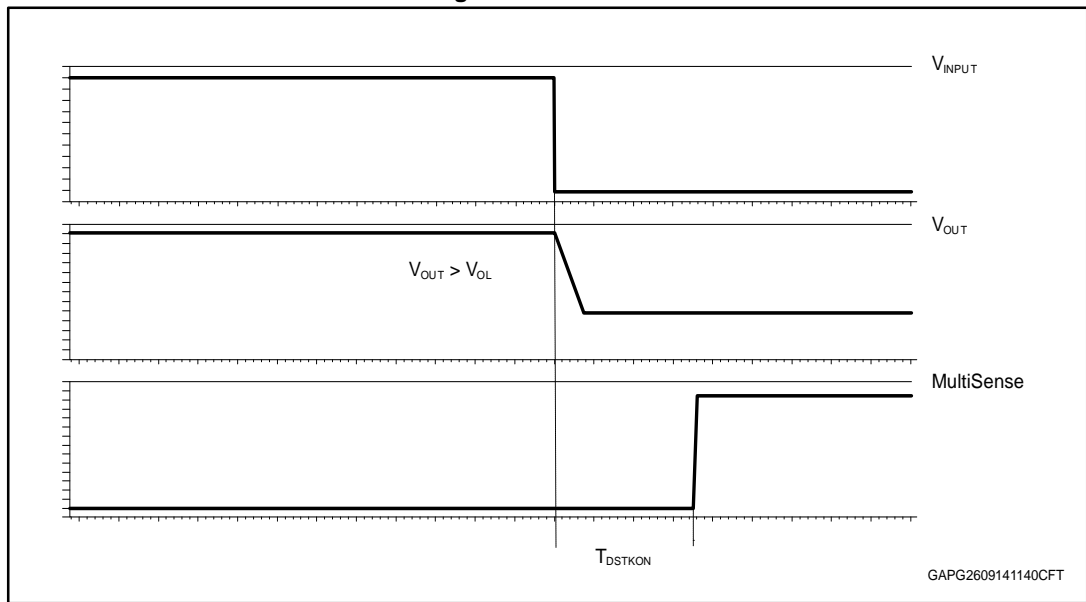


Table 10: Truth table

Mode	Conditions	IN _x	FR	SEn	OUT _x	CurrentSense	Comments
Stand by	All logic inputs low	L	L	L	L	Hi-Z	Low quiescent current consumption
Normal	Nominal load connected; $T_j < 150\text{ }^\circ\text{C}$	L	X	See (1)	L	See (1)	
		H	L		H		Outputs configured for auto-restart
		H	H		H		Outputs configured for Latch-off
Overload	Overload or short to GND causing: $T_j > T_{TSD}$ or $\Delta T_j > \Delta T_{j_SD}$	L	X	See (1)	L	See (1)	
		H	L		H		Output cycles with temperature hysteresis
		H	H		L		Output latches-off
Undervoltage	$V_{CC} < V_{USD}$ (falling)	X	X	X	L L	Hi-Z Hi-Z	Re-start when $V_{CC} > V_{USD} + V_{USDhyst}$ (rising)
OFF-state diagnostics	Short to V_{CC}	L	X	See (1)	H	See (1)	
	Open-load	L	X		H		External pull-up
Negative output voltage	Inductive loads turn-off	L	X	See (1)	< 0 V	See (1)	

Notes:

(1) Refer to [Table 11: "CurrentSense multiplexer addressing"](#)

Table 11: CurrentSense multiplexer addressing

SEn	MUX channel	CS output			
		Nomal mode	Overload	OFF-state diag. <i>(1)(2)(3)</i>	Negative output
L		Hi-Z			
H	Output diagnostic	$I_{SENSE} = 1/K * I_{OUT}$	$V_{SENSE} = V_{SENSEH}$	$V_{SENSE} = V_{SENSEH}$	Hi-Z

Notes:

- (1)Example 2: FR = 1; IN = 0; OUT = latched, $V_{OUT} > V_{OL}$; MUX channel = channel 0 diagnostic; CS = V_{SENSEH}
- (2)Example 1: FR = 1; IN = 0; OUT = L (latched); MUX channel = channel 0 diagnostic; CS = 0
- (3)In case the output channel corresponding to the selected MUX channel is latched off while the relevant input is low, CS pin delivers feedback according to OFF-State diagnostic.

2.4 Waveforms

Figure 9: Latch functionality - behavior in hard short circuit condition ($T_{AMB} \ll T_{TSD}$)

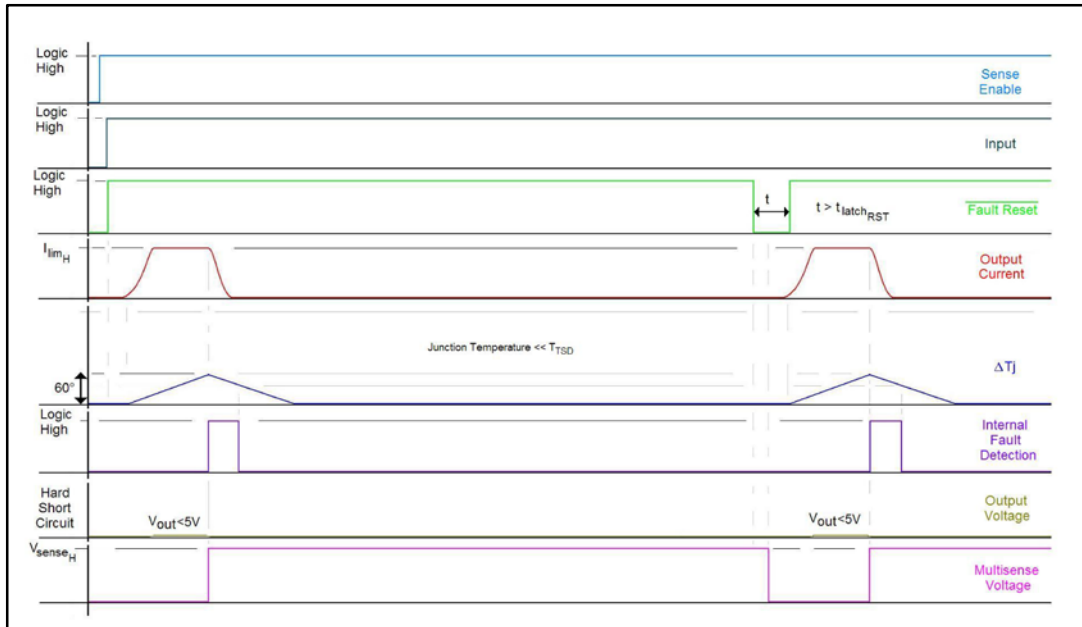


Figure 10: Latch functionality - behavior in hard short circuit condition

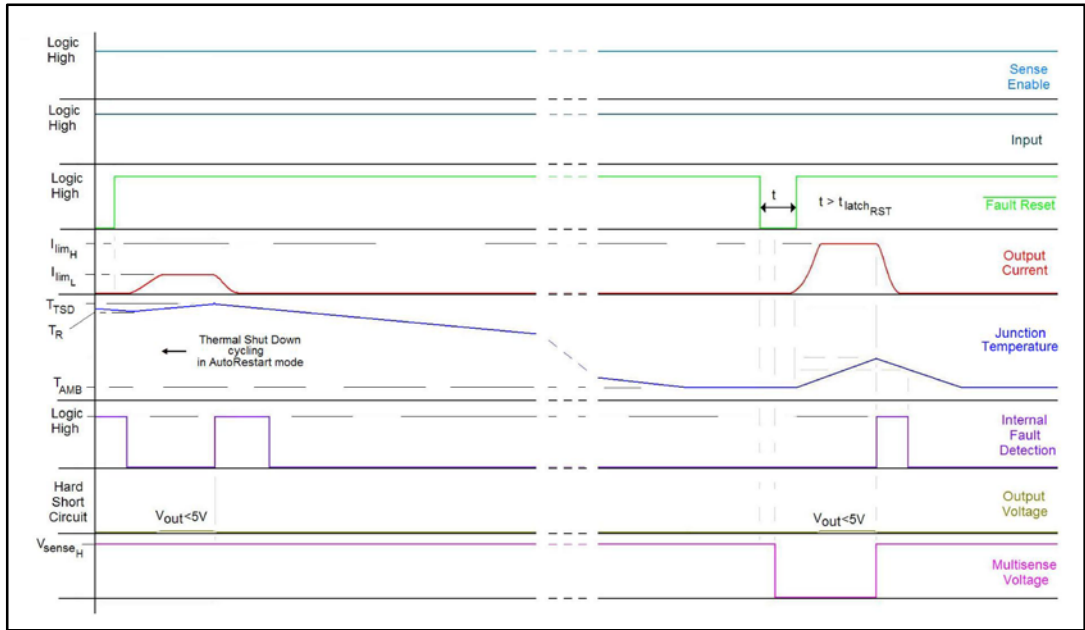


Figure 11: Latch functionality - behavior in hard short circuit condition (autorestart mode + latch off)

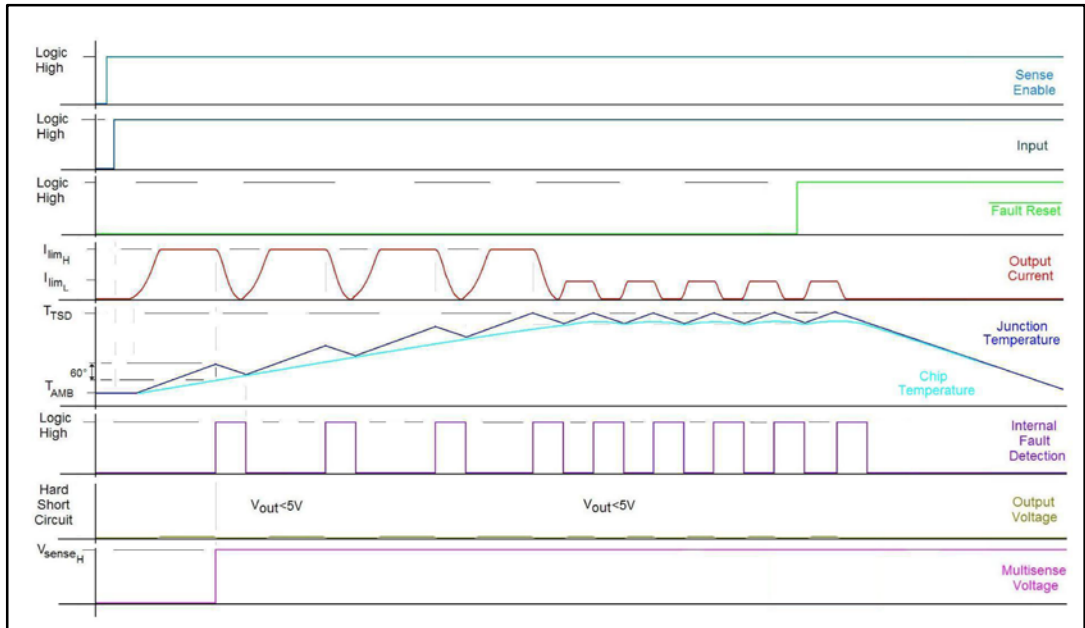


Figure 12: Standby mode activation

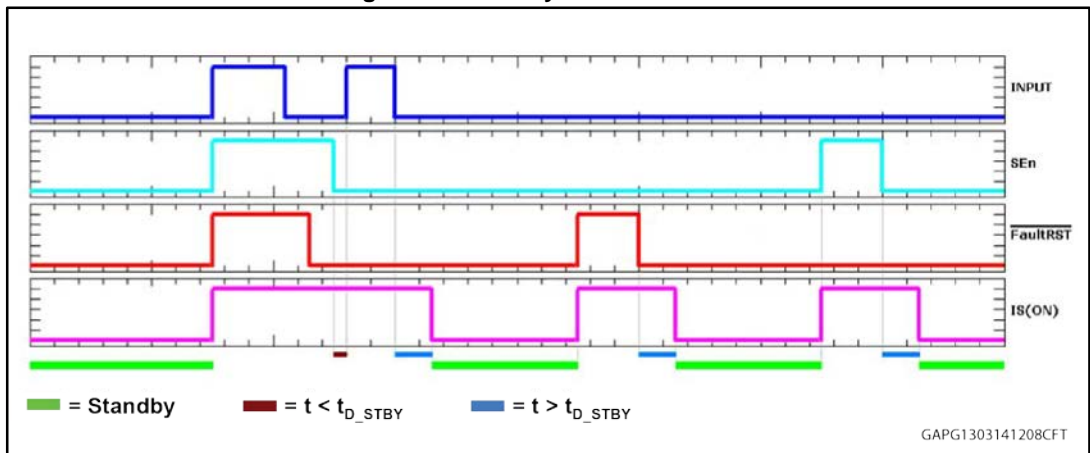
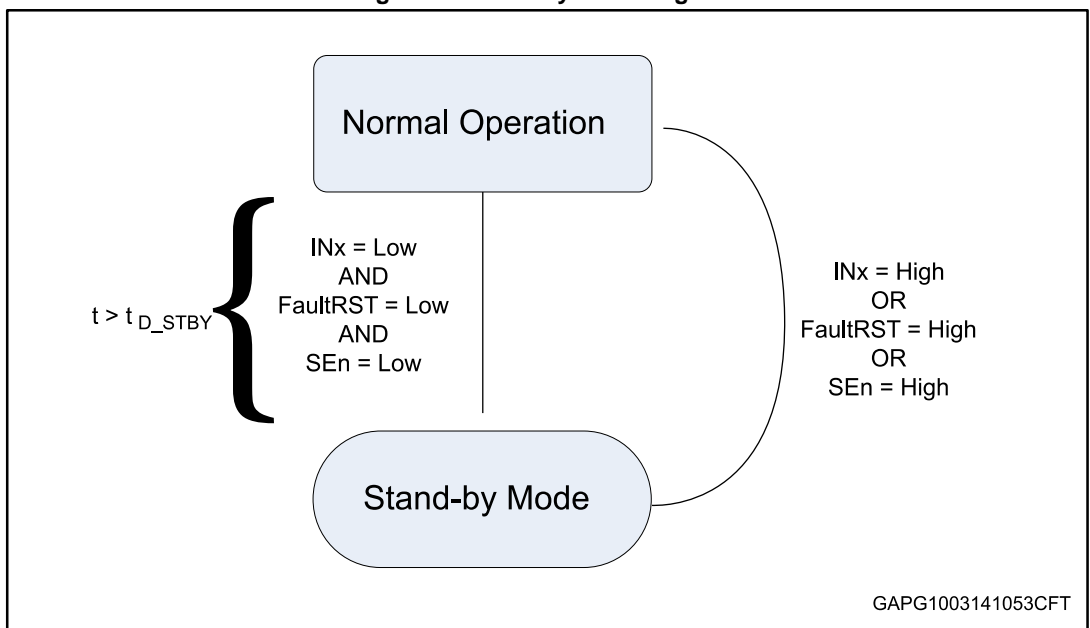
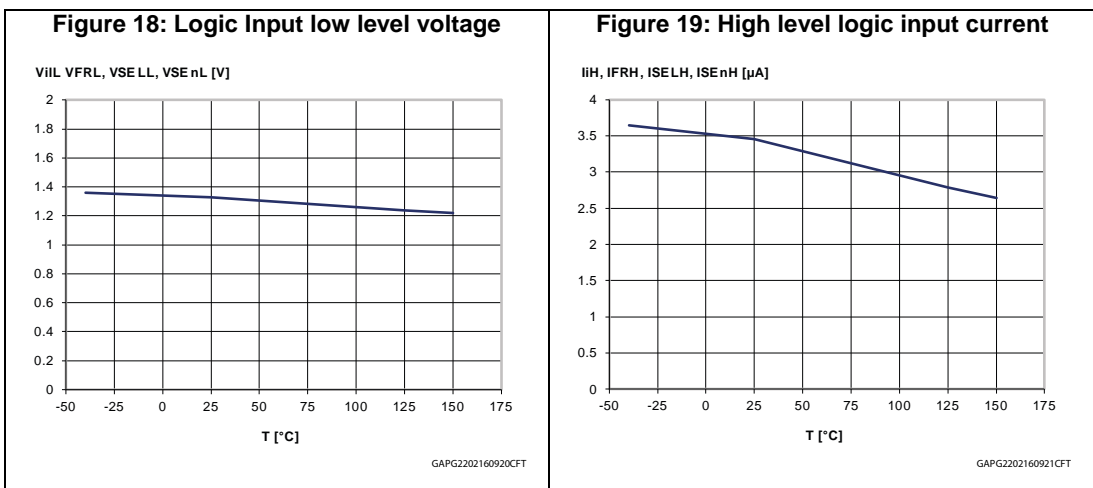
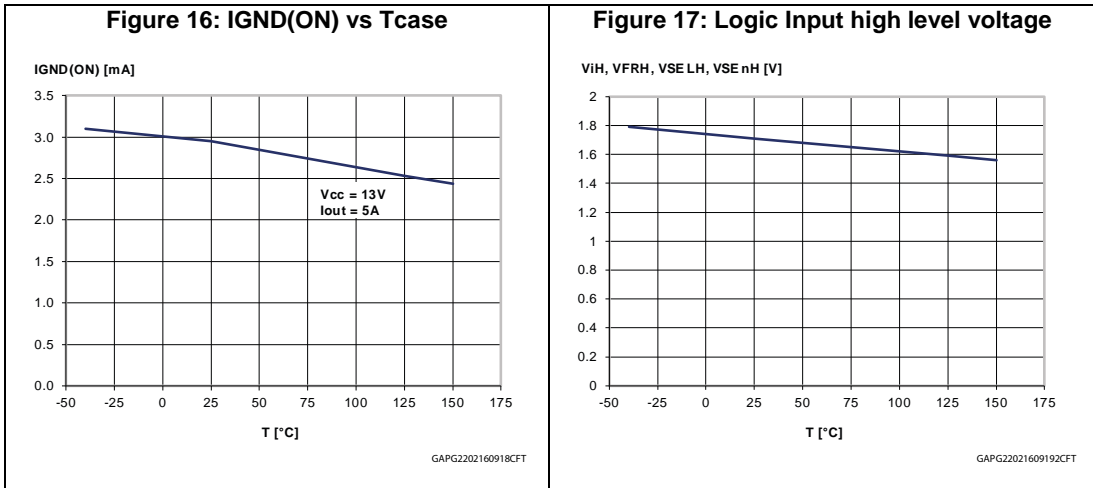
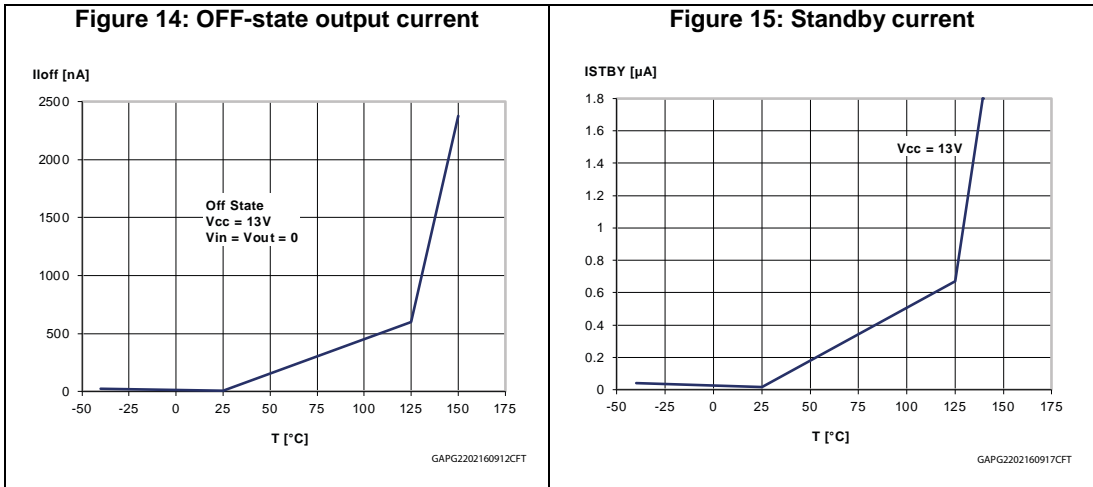


Figure 13: Standby state diagram



2.5 Electrical characteristics curves



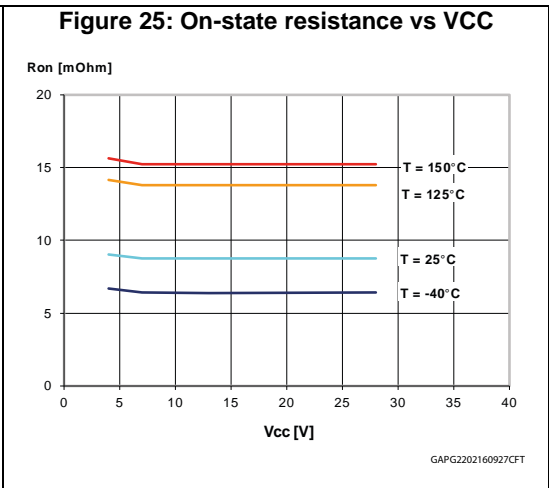
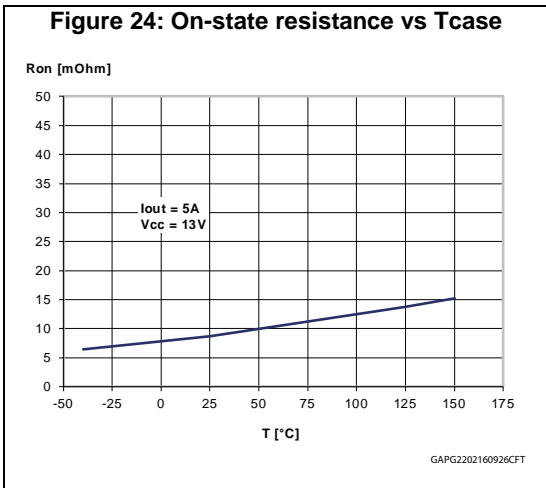
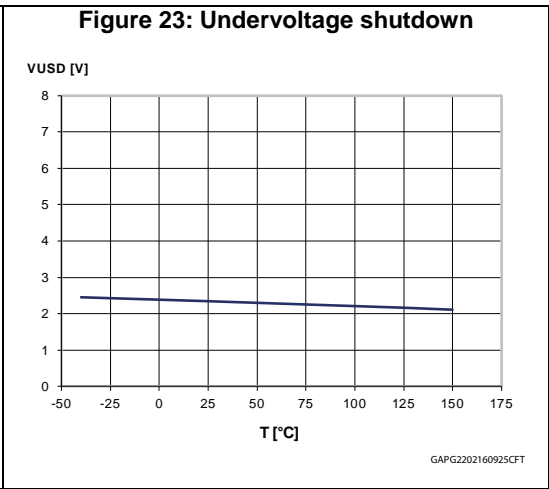
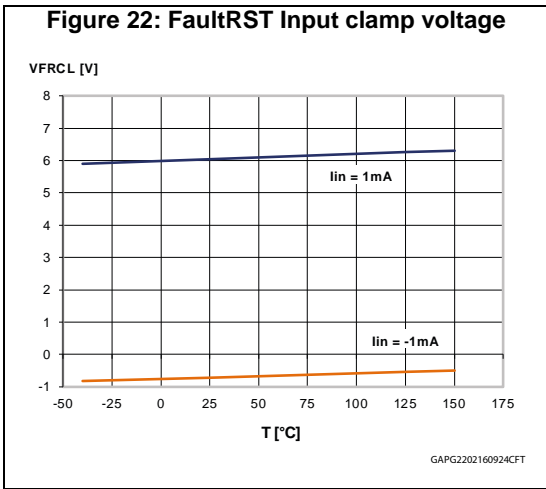
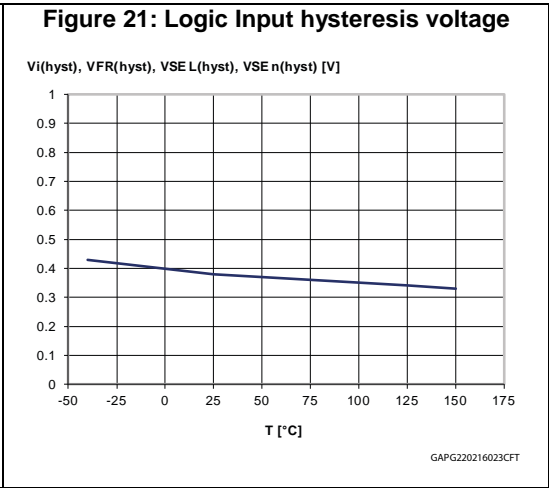
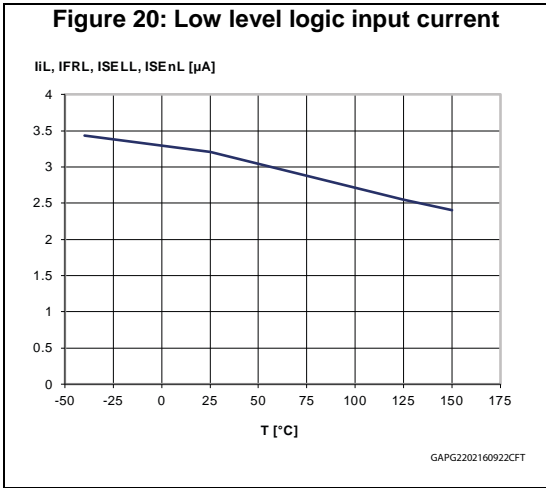
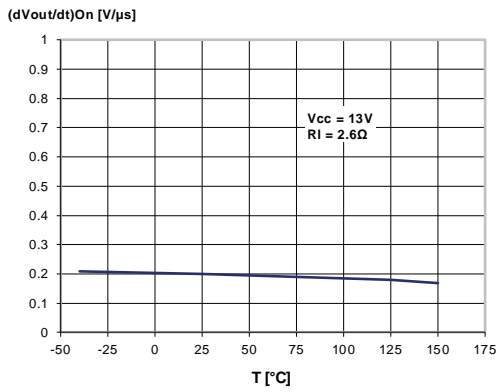
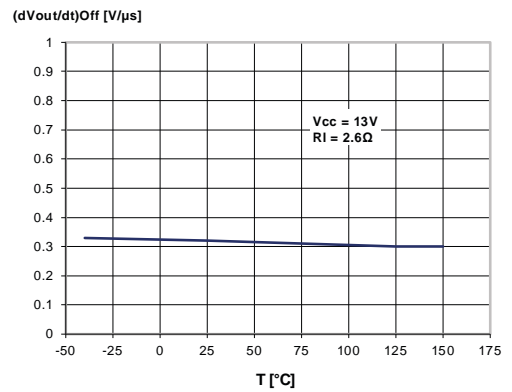


Figure 26: Turn-on voltage slope



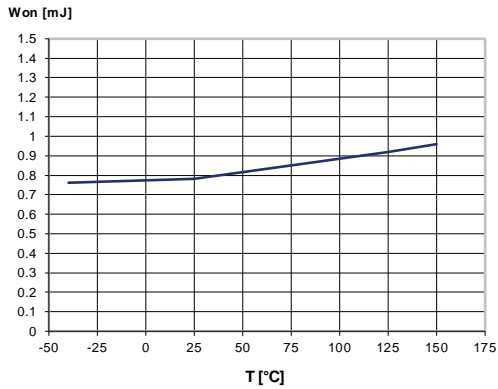
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Figure 27: Turn-off voltage slope



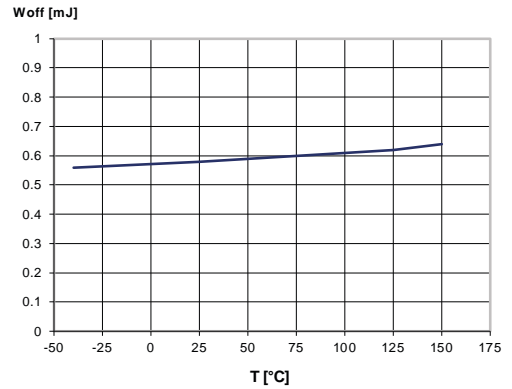
GAPG2202160929CFT

Figure 28: Won vs Tcase



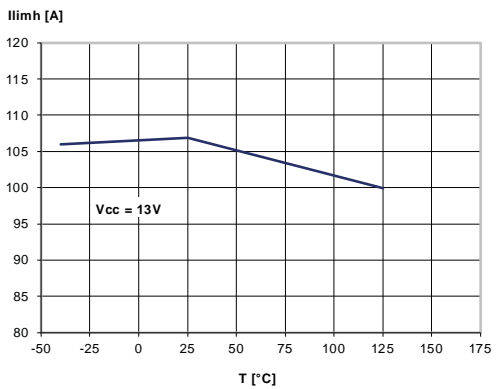
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Figure 29: Woff vs Tcase



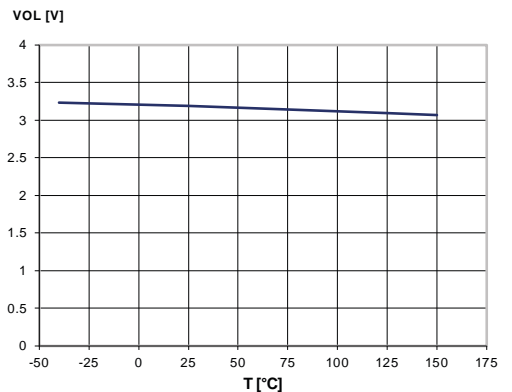
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Figure 30: Ilimh vs. Tcase



GAPG2202160932CFT

Figure 31: OFF-state open-load voltage detection threshold



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Figure 32: Vsense clamp vs. Tcase

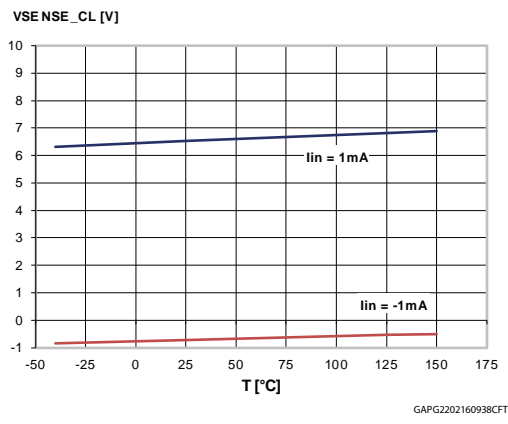
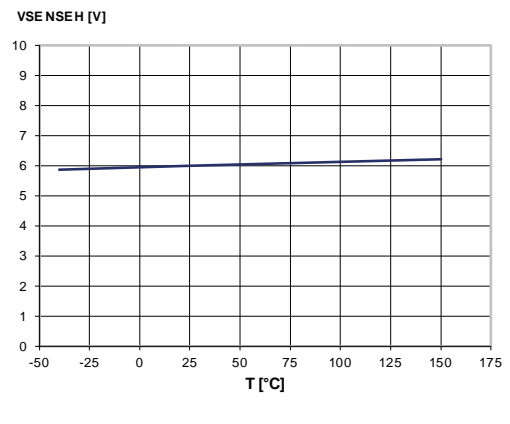


Figure 33: Vsenseh vs. Tcase



3 Protections

3.1 Power limitation

The basic working principle of this protection consists of an indirect measurement of the junction temperature swing ΔT_j through the direct measurement of the spatial temperature gradient on the device surface in order to automatically shut off the output MOSFET as soon as ΔT_j exceeds the safety level of ΔT_{j_SD} . According to the voltage level on the FaultRST pin, the output MOSFET switches on and cycles with a thermal hysteresis according to the maximum instantaneous power which can be handled (FaultRST = Low) or remains off (FaultRST = High). The protection prevents fast thermal transient effects and, consequently, reduces thermo-mechanical fatigue.

3.2 Thermal shutdown

In case the junction temperature of the device exceeds the maximum allowed threshold (typically 175°C), it automatically switches off and the diagnostic indication is triggered. According to the voltage level on the FaultRST pin, the device switches on again as soon as its junction temperature drops to T_R (FaultRST = Low) or remains off (FaultRST = High).

3.3 Current limitation

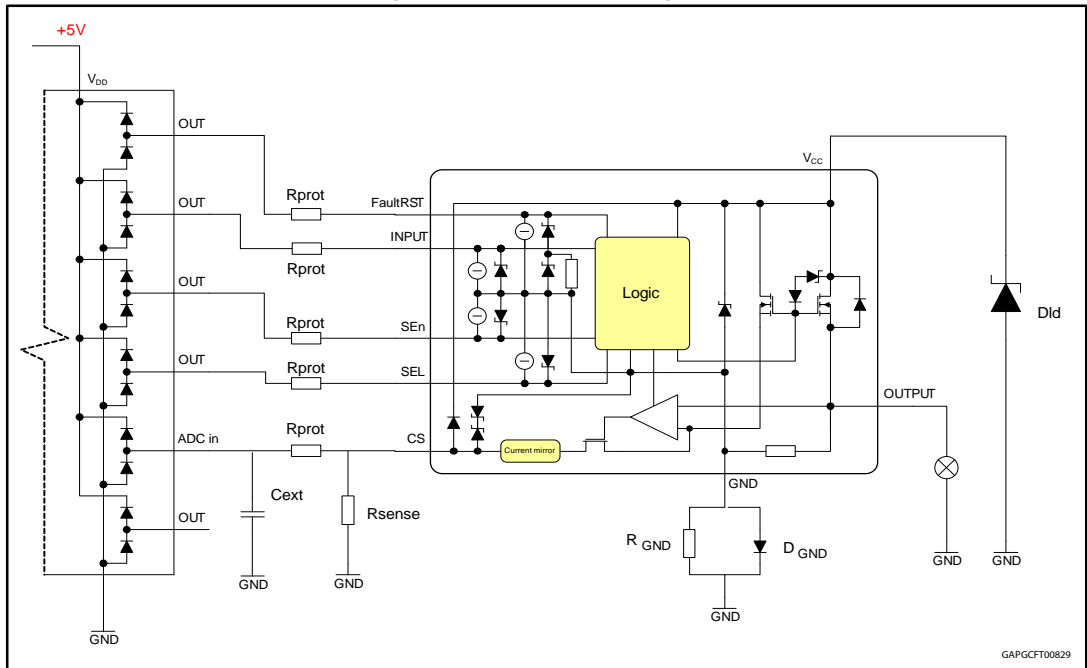
The device is equipped with an output current limiter in order to protect the silicon as well as the other components of the system (e.g. bonding wires, wiring harness, connectors, loads, etc.) from excessive current flow. Consequently, in case of short circuit, overload or during load power-up, the output current is clamped to a safety level, I_{LIMH} , by operating the output power MOSFET in the active region.

3.4 Negative voltage clamp

In case the device drives inductive load, the output voltage reaches a negative value during turn off. A negative voltage clamp structure limits the maximum negative voltage to a certain value, V_{DEMAG} , allowing the inductor energy to be dissipated without damaging the device.

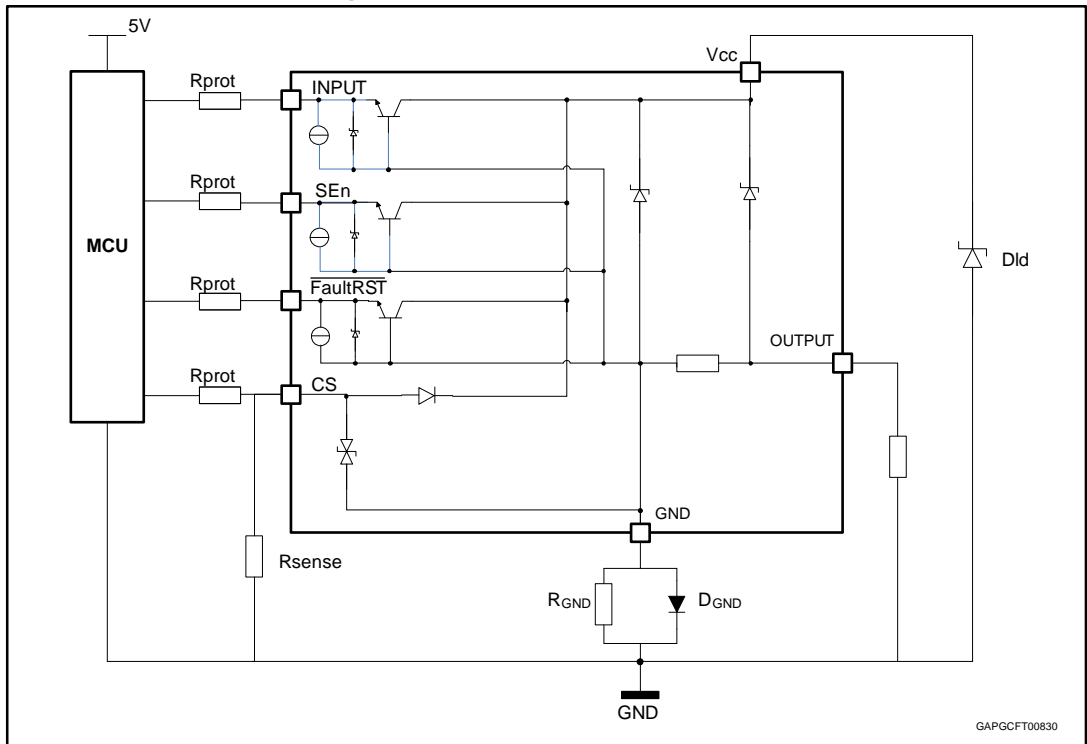
4 Application information

Figure 34: Application diagram



4.1 GND protection network against reverse battery

Figure 35: Simplified internal structure



4.1.1 Diode (DGND) in the ground line

A resistor (typ. $R_{\text{GND}} = 4.7 \text{ k}\Omega$) should be inserted in parallel to D_{GND} if the device drives an inductive load.

This small signal diode can be safely shared amongst several different HSDs. Also in this case, the presence of the ground network produces a shift ($\approx 600 \text{ mV}$) in the input threshold and in the status output values if the microprocessor ground is not common to the device ground. This shift does not vary if more than one HSD shares the same diode/resistor network.

4.2 Immunity against transient electrical disturbances

The immunity of the device against transient electrical emissions, conducted along the supply lines and injected into the V_{CC} pin, is tested in accordance with ISO 7637-2:2011 (E) and ISO 16750-2:2010.

The related function performance status classification is shown in [Table 12: "ISO 7637-2 - electrical transient conduction along supply line"](#).

Test pulses are applied directly to DUT (Device Under Test) both in ON and OFF-state and in accordance to ISO 7637-2:2011(E), chapter 4. The DUT is intended as the present device only, without components and accessed through V_{CC} and GND terminals.

Status II is defined in ISO 7637-1 Function Performance Status Classification (FPSC) as follows: "The function does not perform as designed during the test but returns automatically to normal operation after the test".

Table 12: ISO 7637-2 - electrical transient conduction along supply line

Test Pulse 2011(E)	Test pulse severity level with Status II functional performance status		Minimum number of pulses or test time	Burst cycle / pulse repetition time		Pulse duration and pulse generator internal impedance
	Level	$U_s^{(1)}$		min	max	
1	III	-112V	500 pulses	0,5 s		2ms, 10 Ω
2a	III	+55V	500 pulses	0,2 s	5 s	50 μ s, 2 Ω
3a	IV	-220V	1h	90 ms	100 ms	0.1 μ s, 50 Ω
3b	IV	+150V	1h	90 ms	100 ms	0.1 μ s, 50 Ω
4 ⁽²⁾	IV	-7V	1 pulse			100ms, 0.01 Ω
Load dump according to ISO 16750-2:2010						
Test B ⁽³⁾		40V	5 pulse	1 min		400ms, 2 Ω

Notes:

⁽¹⁾ U_s is the peak amplitude as defined for each test pulse in ISO 7637-2:2011(E), chapter 5.6.

⁽²⁾Test pulse from ISO 7637-2:2004(E).

⁽³⁾With 40 V external suppressor referred to ground ($-40^\circ\text{C} < T_j < 150^\circ\text{C}$).

4.3 MCU I/Os protection

If a ground protection network is used and negative transients are present on the V_{CC} line, the control pins will be pulled negative. ST suggests to insert a resistor (R_{prot}) in line both to prevent the microcontroller I/O pins from latching-up and to protect the HSD inputs.

The value of these resistors is a compromise between the leakage current of microcontroller and the current required by the HSD I/Os (Input levels compatibility) with the latch-up limit of microcontroller I/Os.

Equation

$$V_{CCpeak}/I_{latchup} \leq R_{prot} \leq (V_{OH\mu C} - V_{IH} - V_{GND}) / I_{IHmax}$$

Calculation example:

For $V_{CCpeak} = -150\text{ V}$; $I_{latchup} \geq 20\text{ mA}$; $V_{OH\mu C} \geq 4.5\text{ V}$

$$7.5\text{ k}\Omega \leq R_{prot} \leq 140\text{ k}\Omega.$$

Recommended values: $R_{prot} = 15\text{ k}\Omega$

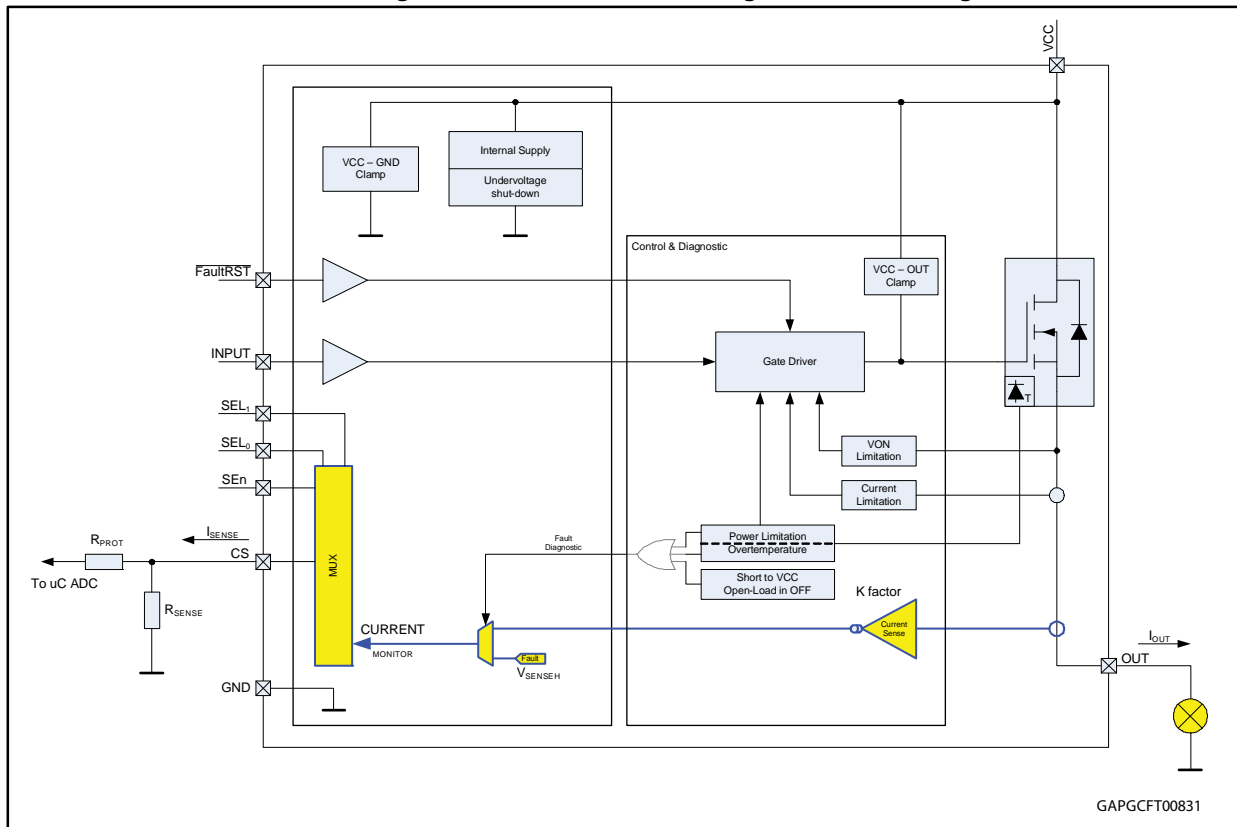
4.4 CS - analog current sense

Diagnostic information on device and load status are provided by an analog output pin (CS) delivering the following signals:

- Current monitor: current mirror of channel output current

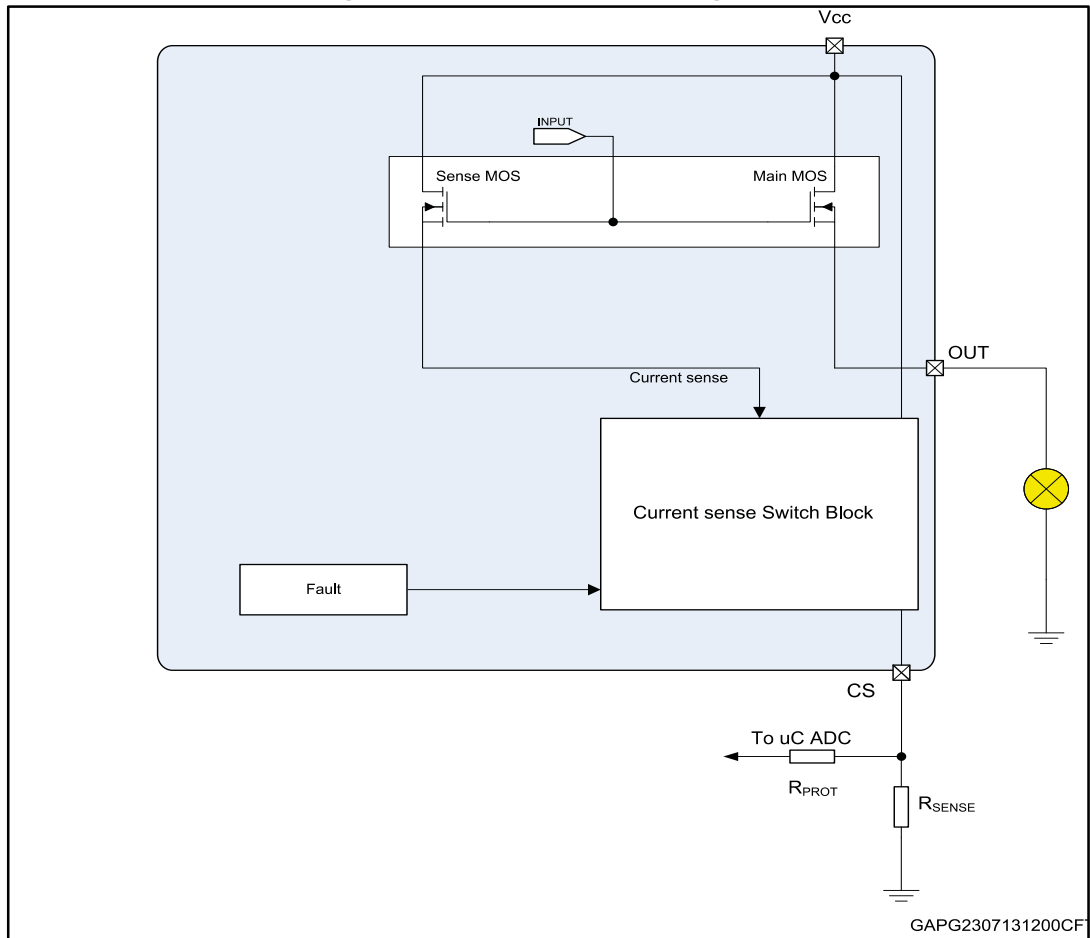
Those signals are routed through an analog multiplexer which is configured and controlled by means of SELx and SEn pins, according to the address map in *MultiSense multiplexer addressing Table*.

Figure 36: CurrentSense and diagnostic – block diagram



4.4.1 Principle of CurrentSense signal generation

Figure 37: CurrentSense block diagram



Current sense

This current mode is selected in the MultiSense, this output is capable to provide:

- Current mirror proportional to the load current in normal operation, delivering current proportional to the load according to known ratio named **K**
- Diagnostics flag in fault conditions delivering fixed voltage V_{SENSEH}

The current delivered by the current sense circuit, I_{SENSE} , can be easily converted to a voltage V_{SENSE} by using an external sense resistor, R_{SENSE} , allowing continuous load monitoring and abnormal condition detection.

Normal operation (channel ON, no fault, SEn active)

While device is operating in normal conditions (no fault intervention), V_{SENSE} calculation can be done using simple equations

Current provided by CS output: $I_{SENSE} = I_{OUT}/K$

Voltage on R_{SENSE} : $V_{SENSE} = R_{SENSE} \cdot I_{SENSE} = R_{SENSE} \cdot I_{OUT}/K$

Where:

- V_{SENSE} is voltage measurable on R_{SENSE} resistor
- I_{SENSE} is current provided from CS pin in current output mode

- I_{OUT} is current flowing through output
- K factor represents the ratio between PowerMOS cells and SenseMOS cells; its spread includes geometric factor spread, current sense amplifier offset and process parameters spread of the overall circuitry the specifying ratio between I_{OUT} and I_{SENSE} .

Failure flag indication

In case of power limitation/overtemperature, the fault is indicated by the CS pin which is switched to a “current limited” voltage source, V_{SENSEH} .

In any case, the current sourced by the CS in this condition is limited to I_{SENSEH} .

The typical behavior in case of overload or hard short circuit is shown in *Waveforms section*.

Figure 38: Analogue HSD – open-load detection in off-state

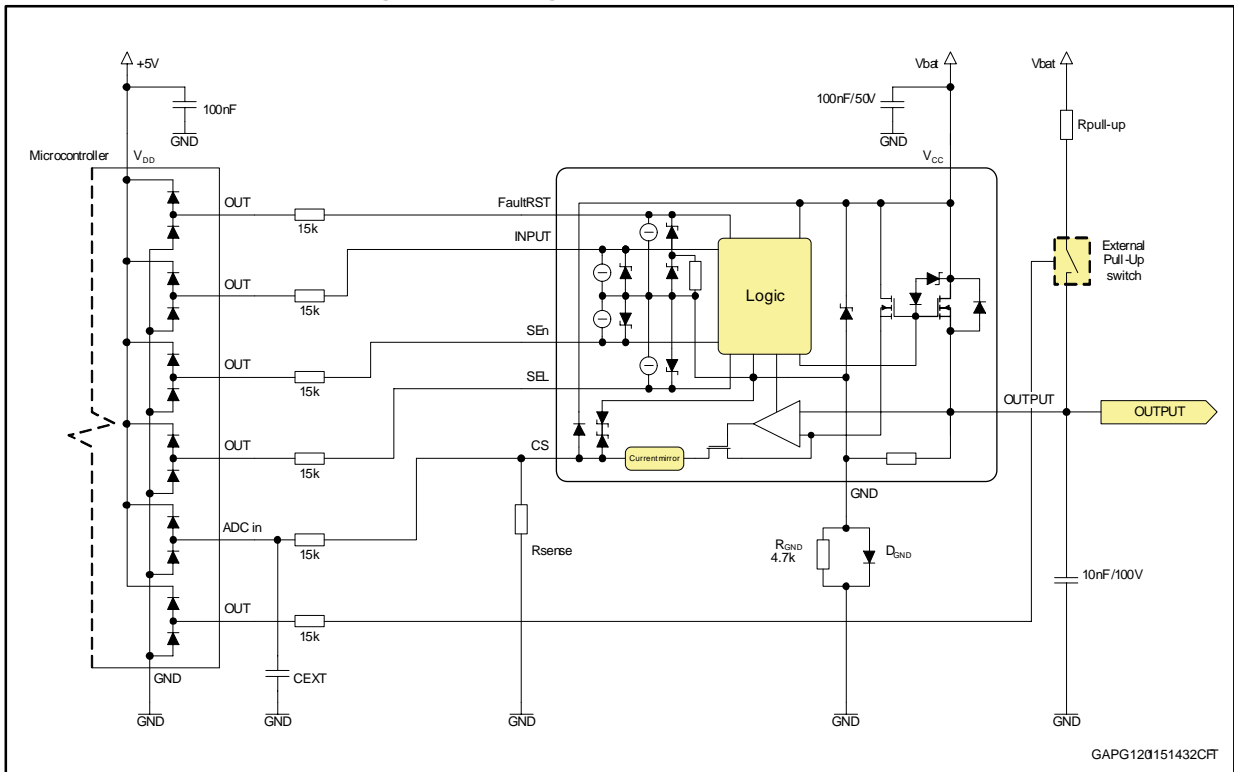
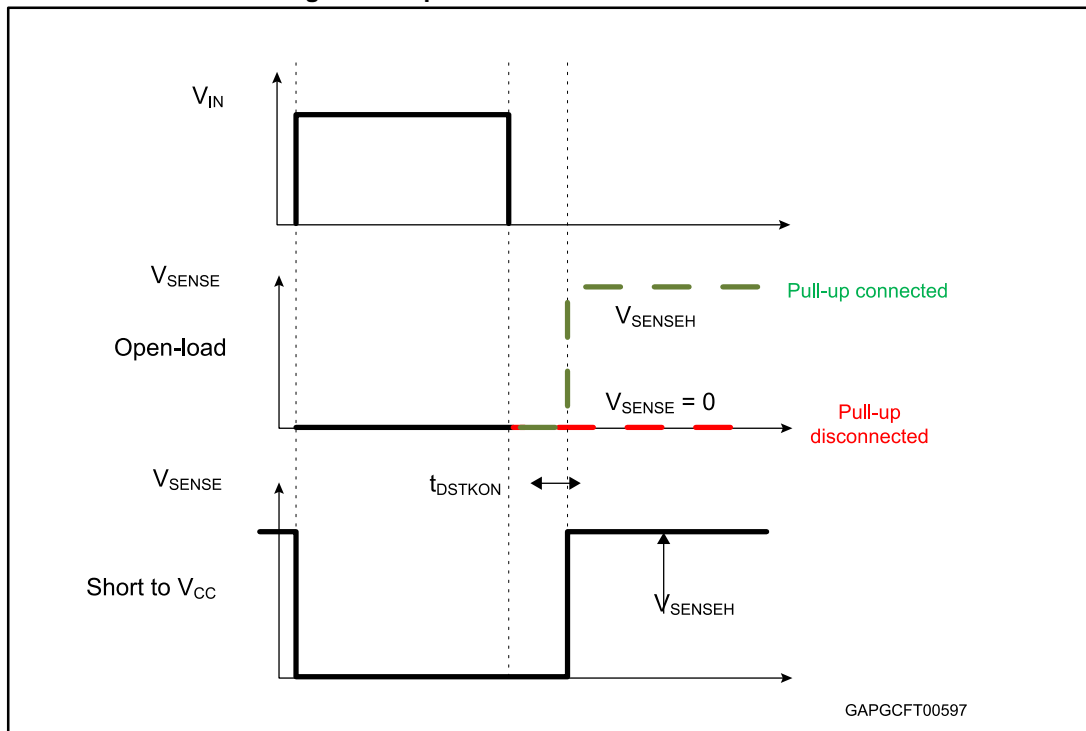


Figure 39: Open-load / short to VCC condition



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Table 13: CurrentSense pin levels in off-state

Condition	Output	CS	SEn
Open-load	$V_{OUT} > V_{OL}$	Hi-Z	L
	$V_{OUT} < V_{OL}$	V_{SENSEH}	H
Short to V_{CC}	$V_{OUT} > V_{OL}$	Hi-Z	L
	$V_{OUT} < V_{OL}$	0	H
Nominal	$V_{OUT} > V_{OL}$	Hi-Z	L
	$V_{OUT} < V_{OL}$	0	H

4.4.2 Short to VCC and OFF-state open-load detection

Short to V_{CC}

A short circuit between V_{CC} and output is indicated by the relevant current sense pin set to V_{SENSEH} during the device off-state. Small or no current is delivered by the current sense during the on-state depending on the nature of the short circuit.

OFF-state open-load with external circuitry

Detection of an open-load in off mode requires an external pull-up resistor R_{PU} connecting the output to a positive supply voltage V_{PU} .

It is preferable that V_{PU} is switched off during the module standby mode in order to avoid the overall standby current consumption to increase in normal conditions, i.e. when load is connected.

R_{PU} must be selected in order to ensure V_{OUT} > V_{OLmax} in accordance with the following equation:

Equation

$$R_{PU} < \frac{V_{PU} - 4}{I_{L(off2)min @ 4V}}$$

5 Package and PCB thermal data

5.1 PowerSSO-16 thermal data

Figure 40: PowerSSO-16 on two-layers PCB (2s0p to JEDEC JESD 51-5)

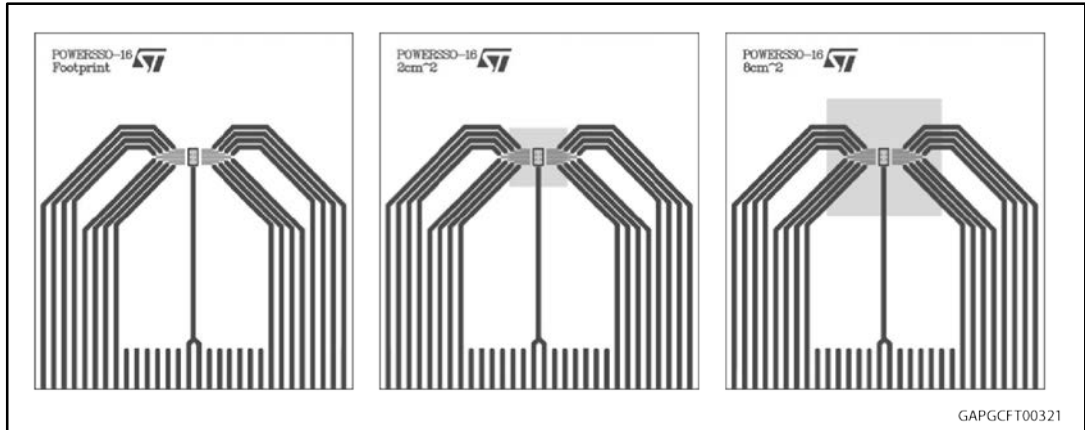


Figure 41: PowerSSO-16 on four-layers PCB (2s2p to JEDEC JESD 51-7)

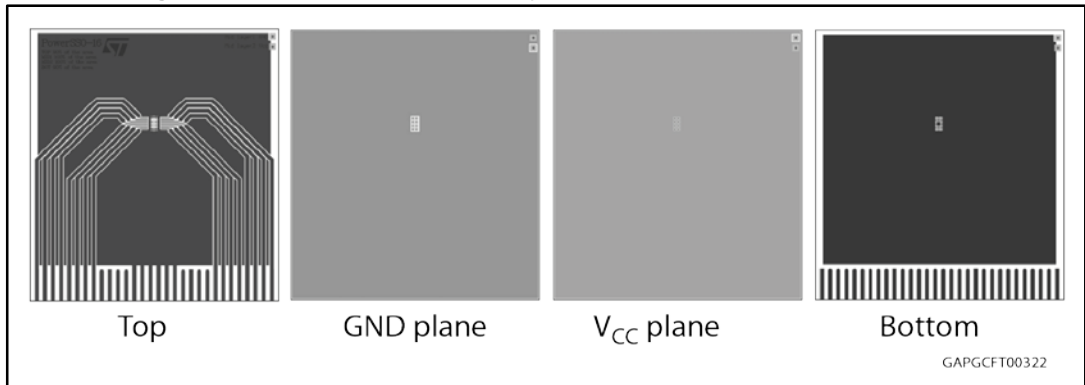


Table 14: PCB properties

Dimension	Value
Board finish thickness	1.6 mm +/- 10%
Board dimension	77 mm x 86 mm
Board Material	FR4
Copper thickness (top and bottom layers)	0.070 mm
Copper thickness (inner layers)	0.035 mm
Thermal via separation	1.2 mm
Thermal via diameter	0.3 mm +/- 0.08 mm
Copper thickness on vias	0.025 mm
Footprint dimension (top layer)	2.2 mm x 3.9 mm
Heatsink copper area dimension (bottom layer)	Footprint, 2 cm ² or 8 cm ²

Figure 42: Rthj-amb vs PCB copper area in open box free air conditions

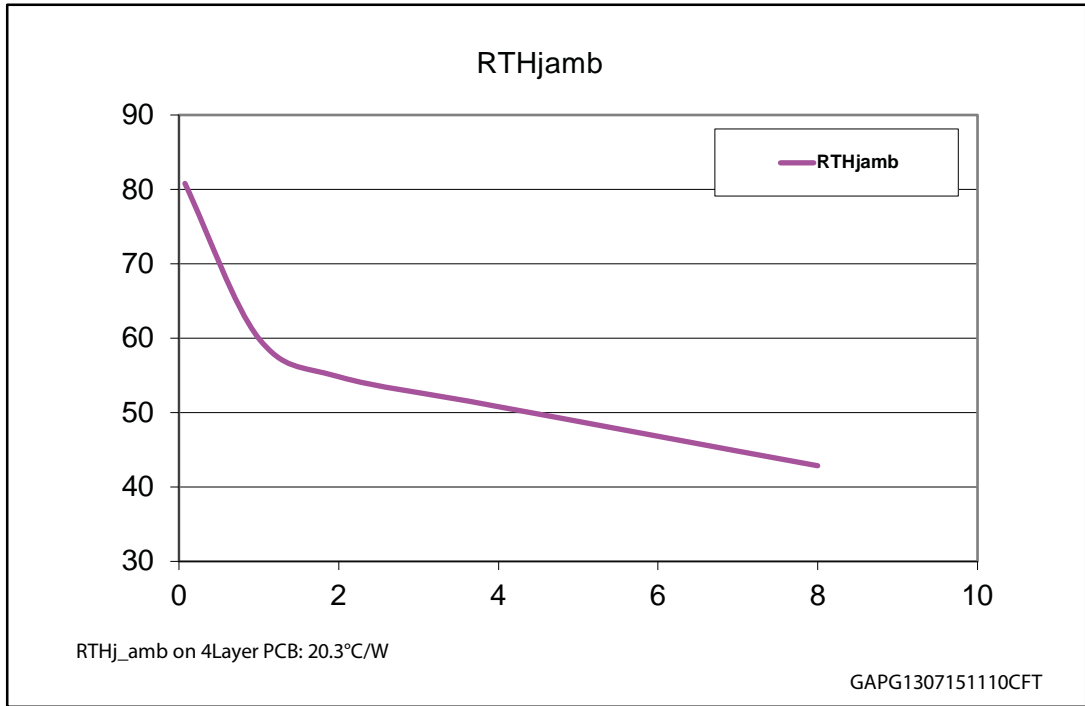
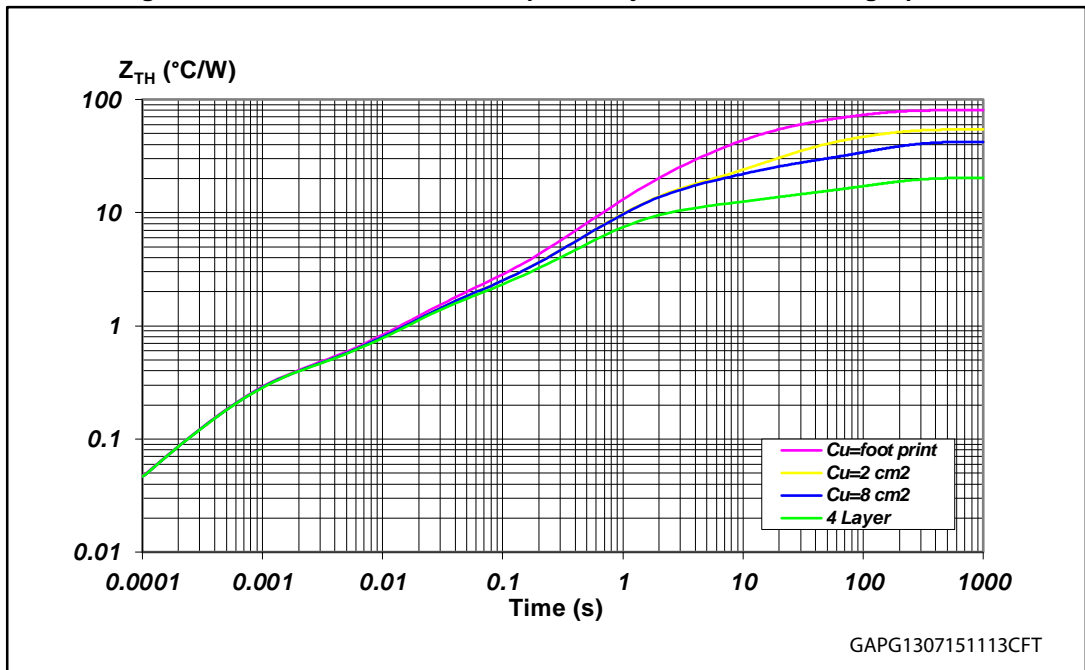


Figure 43: PowerSSO-16 thermal impedance junction ambient single pulse

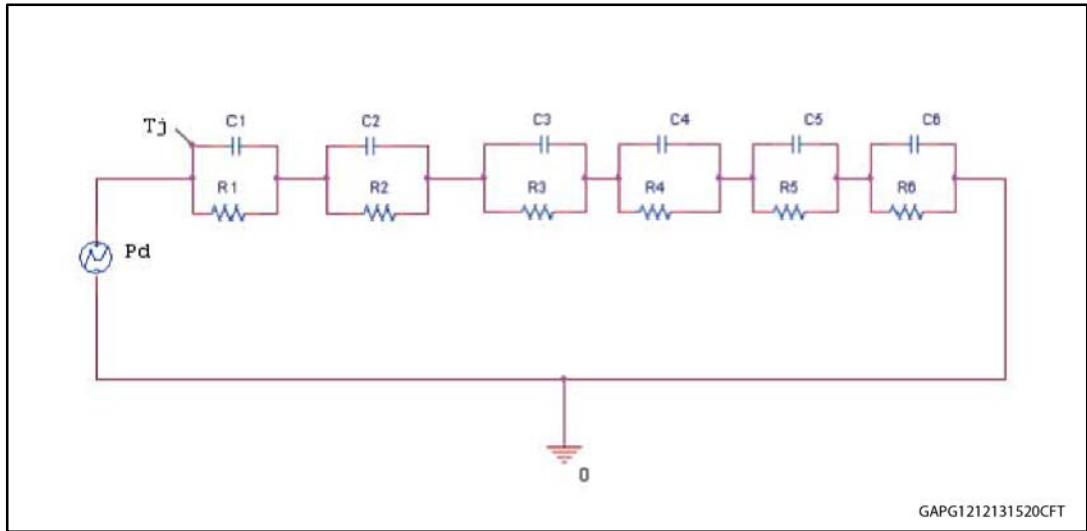


Equation: pulse calculation formula

$$Z_{TH\delta} = R_{TH} \cdot \delta + Z_{THtp} (1 - \delta)$$

where $\delta = t_p/T$

Figure 44: Thermal fitting model for PowerSSO-16



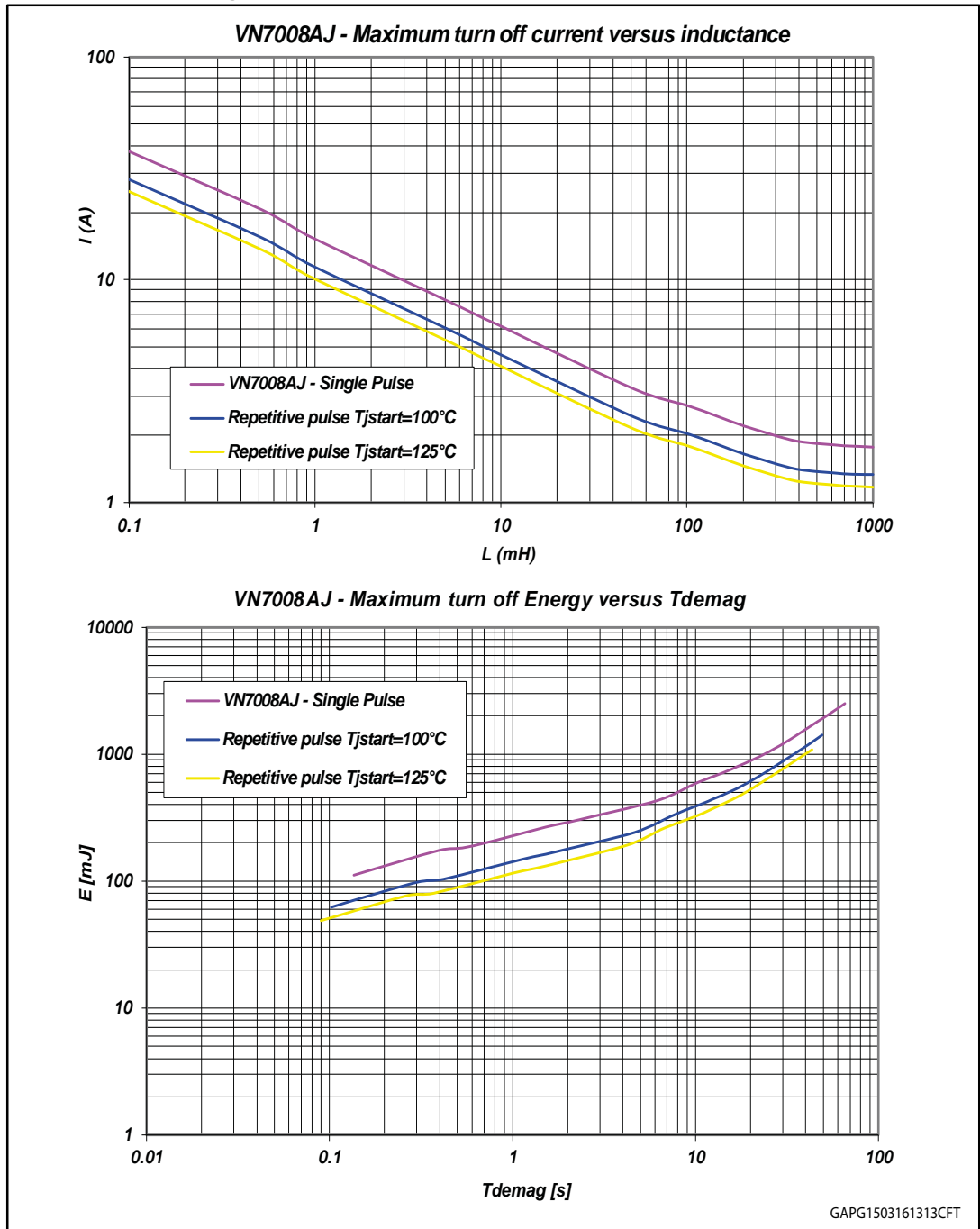
The fitting model is a simplified thermal tool and is valid for transient evolutions where the embedded protections (power limitation or thermal cycling during thermal shutdown) are not triggered.

Table 15: Thermal parameters

Area/island (cm ²)	FP	2	8	4L
R1 (°C/W)	0.3			
R2 (°C/W)	1			
R3 (°C/W)	7	7	7	5
R4 (°C/W)	16	6	6	4
R5 (°C/W)	30	20	10	3
R6 (°C/W)	26	20	18	7
C1 (W·s/°C)	0.0023			
C2 (W·s/°C)	0.021			
C3 (W·s/°C)	0.12			
C4 (W·s/°C)	0.2	0.3	0.3	0.4
C5 (W·s/°C)	0.4	1	1	4
C6 (W·s/°C)	3	5	7	18

6 Maximum demagnetization energy (VCC = 16 V)

Figure 45: Maximum turn off current versus inductance



Values are generated with $R_L = 0 \Omega$.

In case of repetitive pulses, T_{jstart} (at the beginning of each demagnetization) of every pulse must not exceed the temperature specified above for curves A and B.

7 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK® packages, depending on their level of environmental compliance. ECOPACK® specifications, grade definitions and product status are available at: www.st.com. ECOPACK® is an ST trademark.

7.1 PowerSSO-16 package information

Figure 46: PowerSSO-16 package dimensions

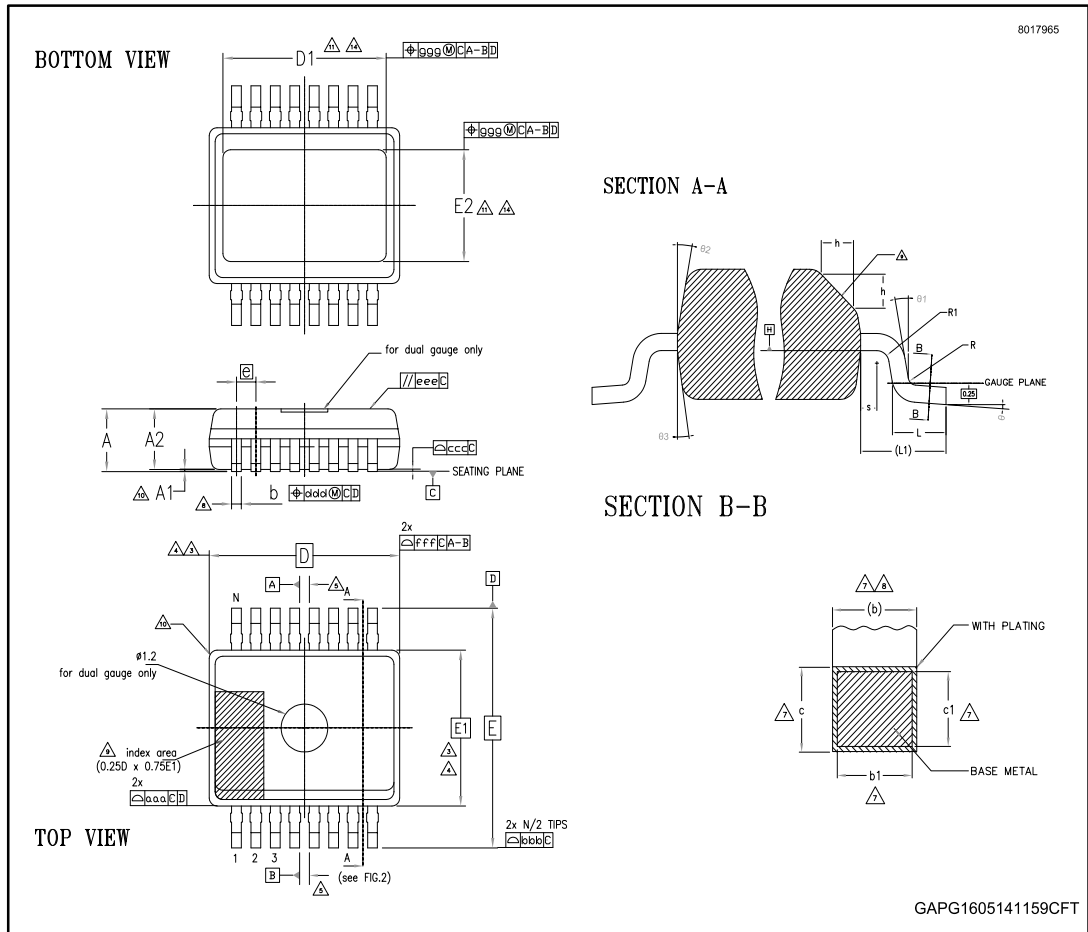


Table 16: PowerSSO-16 mechanical data

Symbol	Millimeters		
	Min.	Typ.	Max.
Θ	0°		8°
Θ1	0°		
Θ2	5°		15°
Θ3	5°		15°
A			1.70
A1	0.00		0.10
A2	1.10		1.60

Symbol	Millimeters		
	Min.	Typ.	Max.
b	0.20		0.30
b1	0.20	0.25	0.28
c	0.19		0.25
c1	0.19	0.20	0.23
D	4.9 BSC		
D1	3.60		4.20
e	0.50 BSC		
E	6.00 BSC		
E1	3.90 BSC		
E2	1.90		2.50
h	0.25		0.50
L	0.40	0.60	0.85
L1	1.00 REF		
N	16		
R	0.07		
R1	0.07		
S	0.20		
Tolerance of form and position			
aaa	0.10		
bbb	0.10		
ccc	0.08		
ddd	0.08		
eee	0.10		
fff	0.10		
ggg	0.15		

7.2 PowerSSO-16 packing information

Figure 47: PowerSSO-16 reel 13"

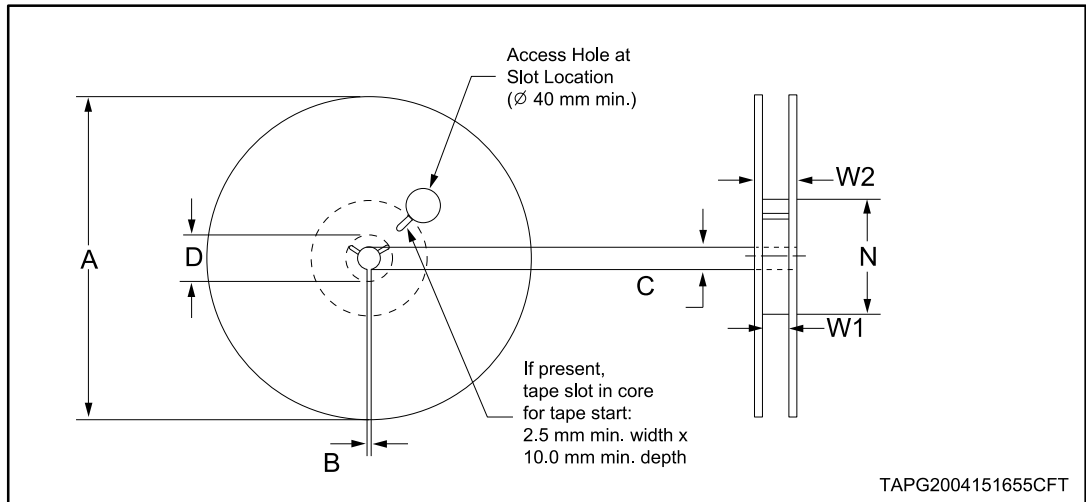


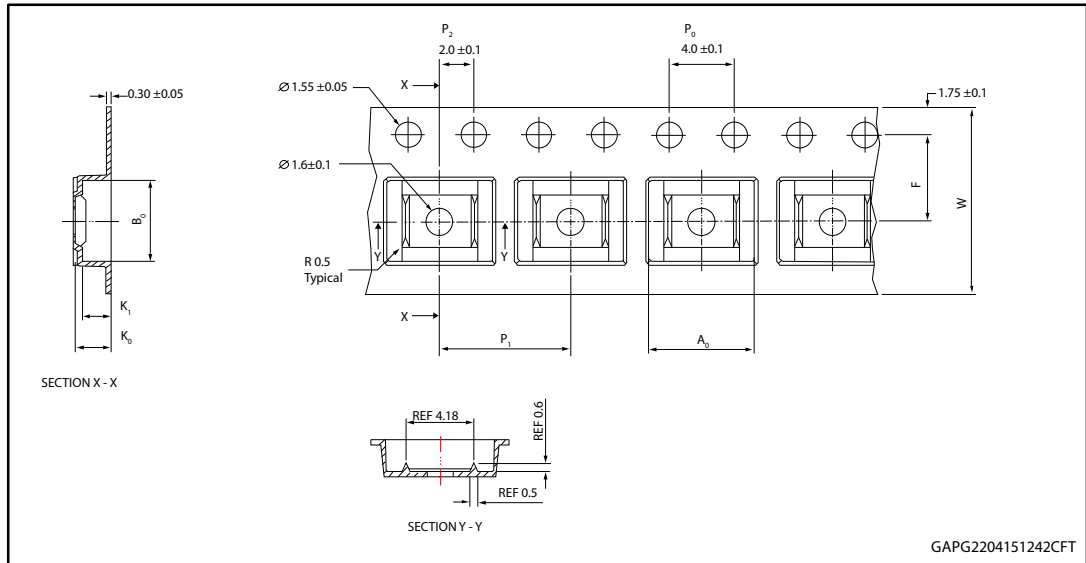
Table 17: Reel dimensions

Description	Value ⁽¹⁾
Base quantity	2500
Bulk quantity	2500
A (max)	330
B (min)	1.5
C (+0.5, -0.2)	13
D (min)	20.2
N	100
W1 (+2 /-0)	12.4
W2 (max)	18.4

Notes:

⁽¹⁾All dimensions are in mm.

Figure 48: PowerSSO-16 carrier tape



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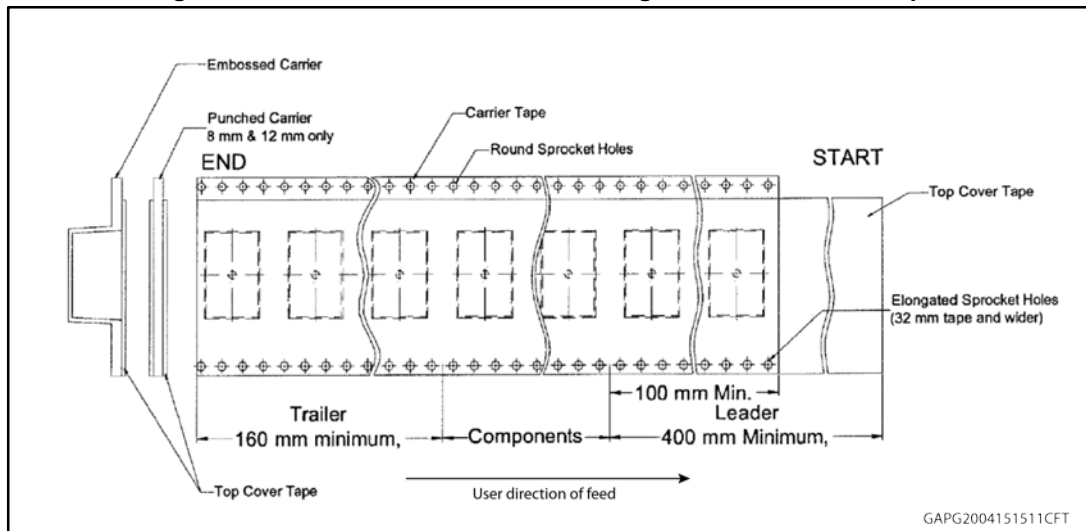
Table 18: PowerSSO-16 carrier tape dimensions

Description	Value ⁽¹⁾
A ₀	6.50 ± 0.1
B ₀	5.25 ± 0.1
K ₀	2.10 ± 0.1
K ₁	1.80 ± 0.1
F	5.50 ± 0.1
P ₁	8.00 ± 0.1
W	12.00 ± 0.3

Notes:

⁽¹⁾All dimensions are in mm.

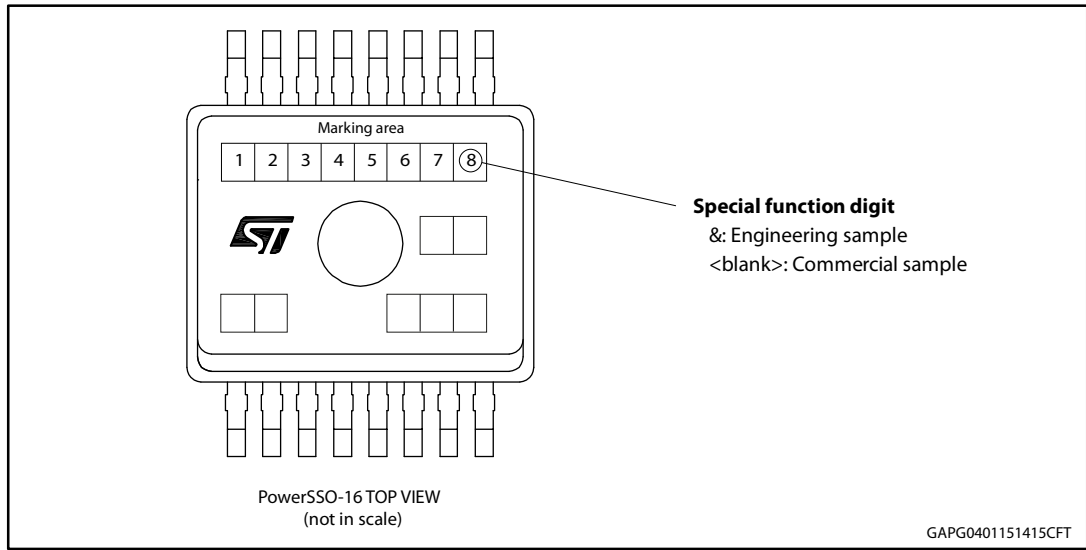
Figure 49: PowerSSO-16 schematic drawing of leader and trailer tape



GAPG2004151511CFT

7.3 PowerSSO-16 marking information

Figure 50: PowerSSO-16 marking information



Engineering Samples: these samples can be clearly identified by a dedicated special symbol in the marking of each unit. These samples are intended to be used for electrical compatibility evaluation only; usage for any other purpose may be agreed only upon written authorization by ST. ST is not liable for any customer usage in production and/or in reliability qualification trials.

Commercial Samples: fully qualified parts from ST standard production with no usage restrictions.

8 Order codes

Table 19: Device summary

Package	Order codes
	Tape and reel
PowerSSO-16	VN7008AJTR

9 Revision history

Table 20: Document revision history

Date	Revision	Changes
23-Apr-2015	1	Initial release
13-Jul-2015	2	Updated Table 4: "Thermal data" Table 6: "Switching" : <ul style="list-style-type: none"> • W_{ON}, W_{OFF}, t_{SKEW}: updated values Table 7: "Logic Inputs" : <ul style="list-style-type: none"> • V_{ICL}, V_{FRCL}, V_{SENCL}: updated maximum value Table 9: "CurrentSense" : <ul style="list-style-type: none"> • K_x, $t_{DSENSE2H}$, $\Delta t_{DSENSE2H}$: updated values Updated Section 5.1: "PowerSSO-16 thermal data"
22-Feb-2016	3	Table 3: "Absolute maximum ratings" : <ul style="list-style-type: none"> • E_{MAX}: updated value Table 6: "Switching" : <ul style="list-style-type: none"> • $t_{d(on)}$, $t_{d(off)}$, W_{ON}, W_{OFF}, t_{SKEW}: updated values Table 9: "CurrentSense" : <ul style="list-style-type: none"> • K_0, K_1: updated values Added Figure 4: "IOUT/ISENSE versus IOUT" and Figure 5: "Current sense accuracy versus IOUT" Added Section 2.5: "Electrical characteristics curves" and Section 4: "Application information" Updated Section 5.1: "PowerSSO-16 thermal data"
15-Mar-2016	4	Added Section 6: "Maximum demagnetization energy (VCC = 16 V)"

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