

## Double channel high-side driver with analog current sense for automotive applications

Datasheet - production data



### Features

Max transient supply voltage	$V_{CC}$	41 V
Operating voltage range	$V_{CC}$	4.5 to 28 V
Typ on-state resistance (per ch.)	$R_{ON}$	5 m $\Omega$
Current limitation (typ)	$I_{LIMH}$	100 A
Off-state supply current	$I_S$	2 $\mu A^{(1)}$

1. Typical value with all loads connected.

- General
  - Very low standby current
  - 3.0 V CMOS compatible inputs
  - Optimized electromagnetic emissions
  - Very low electromagnetic susceptibility
  - Compliance with European directive 2002/95/EC
  - Very low current sense leakage
- Diagnostic functions
  - Proportional load current sense
  - High current sense precision for wide currents range
  - Current sense disable
  - Off-state openload detection
  - Output short to  $V_{CC}$  detection
  - Overload and short to ground (power limitation) indication
  - Thermal shutdown indication
- Protection
  - Undervoltage shutdown
  - Overvoltage clamp
  - Load current limitation
  - Self limiting of fast thermal transients

- Protection against loss of ground and loss of  $V_{CC}$
- Overtemperature shutdown with auto restart (thermal shutdown)
- Inrush current active management by power limitation
- Reverse battery protected with self switch of the PowerMOS
- Electrostatic discharge protection

### Applications

- All types of resistive, inductive and capacitive loads

### Description

The VND5E006ASP-E is a double channel high-side driver manufactured using ST proprietary VIPower™ M0-5 technology and housed in PowerSO-16 package. The device is designed to drive 12 V automotive grounded loads, and to provide protection and diagnostics. They also implement a 3 V and 5 V CMOS compatible interface for the use with any microcontroller.

The device integrates advanced protective functions such as load current limitation, inrush and overload active management by power limitation, overtemperature shut-off with auto restart and overvoltage active clamp. A dedicated analog current sense pin is associated with every output channel providing enhanced diagnostic functions including fast detection of overload and short-circuit to ground through power limitation indication, overtemperature indication, short-circuit to  $V_{CC}$  diagnosis and on-state and off-state open-load detection. The current sensing and diagnostic feedback of the whole device can be disabled by pulling the CS\_DIS pin high to share the external sense resistor with similar devices.

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# 1 Block diagram and pin description

Figure 1. Block diagram

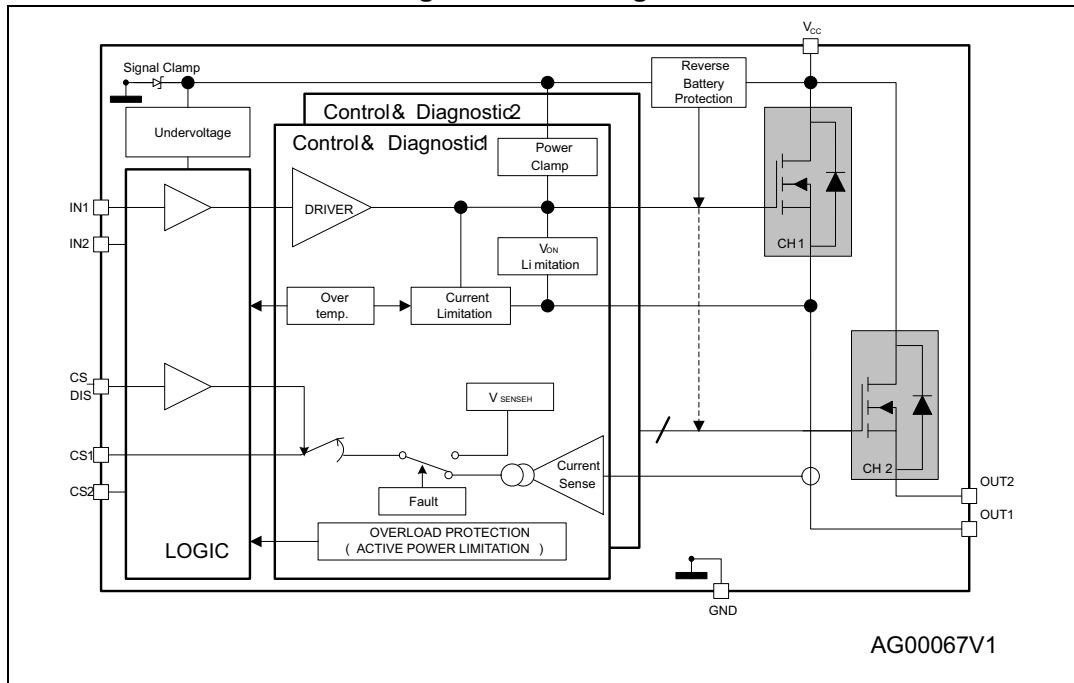


Table 1. Pin function

Name	Function
V <sub>CC</sub>	Battery connection.
OUTn	Power output.
GND	Ground connection.
INn	Voltage controlled input pin with hysteresis, CMOS compatible. Controls output switch state.
CSn	Analog current sense pin, delivers a current proportional to the load current.
CS_DIS	Active high CMOS compatible pin, to disable the current sense pin.

Figure 2. Configuration diagram (top view)

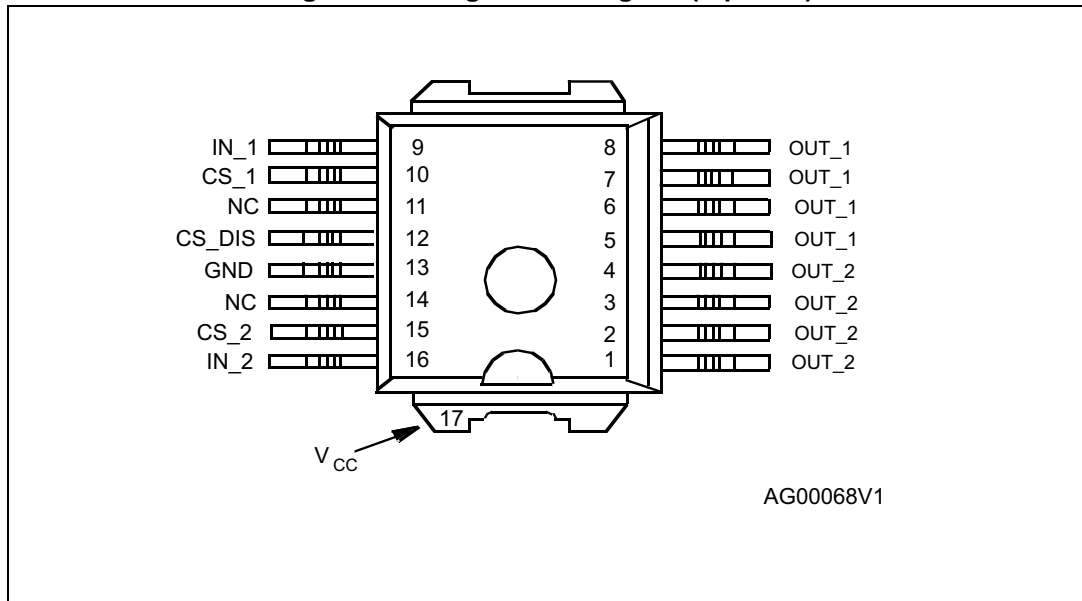
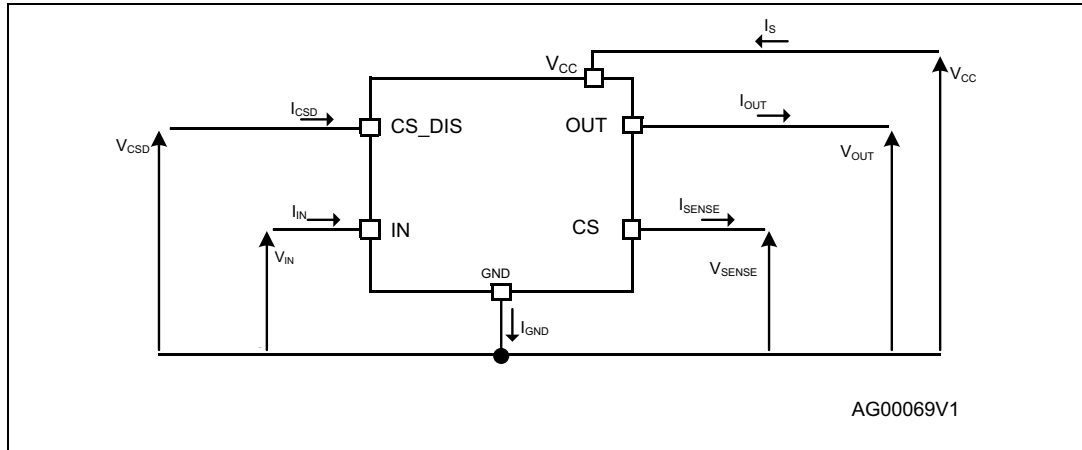


Table 2. Suggested connections for unused and not connected pins

Connection / pin	Current sense	N.C.	Output	Input	CS_DIS
Floating	Not allowed	X	X	X	X
To ground	Through 1 K $\Omega$ resistor	X	Not allowed	Through 10 K $\Omega$ resistor	Through 10 K $\Omega$ resistor

## 2 Electrical specifications

Figure 3. Current and voltage conventions



### 2.1 Absolute maximum ratings

Stressing the device above the ratings listed in the “Absolute maximum ratings” tables may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other conditions above those indicated in the Operating sections of this specification is not implied. Exposure to the conditions in this section for extended periods may affect device reliability. Refer also to the STMicroelectronics SURE Program and other relevant quality documents.

Table 3. Absolute maximum ratings

Symbol	Parameter	Value	Unit
V <sub>CC</sub>	DC supply voltage	28	V
V <sub>CCPK</sub>	Transient supply voltage (T < 400 ms, R <sub>LOAD</sub> > 0.5 Ω)	41	V
-V <sub>CC</sub>	Reverse DC supply voltage	16	V
I <sub>OUT</sub>	DC output current	Internally limited	A
-I <sub>OUT</sub>	Reverse DC output current	60	A
I <sub>IN</sub>	DC input current	-1 to 10	mA
I <sub>CSD</sub>	DC current sense disable input current	-1 to 10	mA
V <sub>CSSENSE</sub>	Current sense maximum voltage	V <sub>CC</sub> -41 +V <sub>CC</sub>	V V
E <sub>MAX</sub>	Maximum switching energy (single pulse) (L = 1.4 mH; R <sub>L</sub> = 0 Ω; V <sub>bat</sub> = 13.5 V; T <sub>jstart</sub> = 150 °C; I <sub>OUT</sub> = I <sub>limL</sub> (Typ.))	600	mJ

Table 3. Absolute maximum ratings (continued)

Symbol	Parameter	Value	Unit
V <sub>ESD</sub>	Electrostatic discharge (Human Body Model: R = 1.5 K $\Omega$ ; C = 100 pF)		
	– Input	4000	V
	– Current sense	2000	
	– CS_DIS	4000	
	– Output	5000	
– V <sub>CC</sub>	5000		
V <sub>ESD</sub>	Charge device model (CDM-AEC-Q100-011)	750	V
T <sub>j</sub>	Junction operating temperature	-40 to 150	°C
T <sub>stg</sub>	Storage temperature	-55 to 150	°C

## 2.2 Thermal data

Table 4. Thermal data

Symbol	Parameter	Maximum value	Unit
R <sub>thj-case</sub>	Thermal resistance junction-case (MAX) (with one channel ON)	0.4	°C/W
R <sub>thj-amb</sub>	Thermal resistance junction-ambient	See <a href="#">Figure 36</a>	°C/W



## 2.3 Electrical characteristics

$8\text{ V} < V_{CC} < 28\text{ V}$ ;  $-40\text{ °C} < T_j < 150\text{ °C}$ , unless otherwise specified.

**Table 5. Power section**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$V_{CC}$	Operating supply voltage		4.5	13	28	V
$V_{USD}$	Undervoltage shutdown			3.5	4.5	V
$V_{USDhyst}$	Undervoltage shutdown hysteresis			0.5		V
$R_{ON}$	On-state resistance	$I_{OUT} = 10\text{ A}$ ; $T_j = 25\text{ °C}$		5		m $\Omega$
		$I_{OUT} = 10\text{ A}$ ; $T_j = 150\text{ °C}$			10	m $\Omega$
		$I_{OUT} = 10\text{ A}$ ; $V_{CC} = 5\text{ V}$ ; $T_j = 25\text{ °C}$			8	m $\Omega$
$R_{ON REV}$	Reverse battery on-state resistance	$V_{CC} = -13\text{ V}$ ; $I_{OUT} = -10\text{ A}$ ; $T_j = 25\text{ °C}$			6	m $\Omega$
$V_{clamp}$	Clamp voltage	$I_S = 20\text{ mA}$	41	46	52	V
$I_S$	Supply current	Off-state; $V_{CC} = 13\text{ V}$ ; $T_j = 25\text{ °C}$ ; $V_{IN} = V_{OUT} = V_{SENSE} = V_{CSD} = 0\text{ V}$		2 <sup>(1)</sup>	5 <sup>(1)</sup>	$\mu\text{A}$
		On-state; $V_{CC} = 13\text{ V}$ ; $V_{IN} = 5\text{ V}$ ; $I_{OUT} = 0\text{ A}$		3.5	6.5	mA
$I_{L(off1)}$	Off-state output current <sup>(2)</sup>	$V_{IN}=V_{OUT}=0\text{V}$ ; $V_{CC}=13\text{V}$ ; $T_j=25\text{°C}$	0	0.01	3	$\mu\text{A}$
		$V_{IN}=V_{OUT}=0\text{V}$ ; $V_{CC}=13\text{V}$ ; $T_j=125\text{°C}$	0		5	$\mu\text{A}$

1. PowerMOS leakage included.

2. For each channel.

**Table 6. Switching ( $V_{CC} = 13\text{ V}$ ;  $T_j = 25\text{ °C}$ )**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$t_{d(on)}$	Turn-on delay time	$R_L = 1.3\ \Omega$ (see <a href="#">Figure 6</a> )	—	35	—	$\mu\text{s}$
$t_{d(off)}$	Turn-off delay time	$R_L = 1.3\ \Omega$ (see <a href="#">Figure 6</a> )	—	20	—	$\mu\text{s}$
$(dV_{OUT}/dt)_{on}$	Turn-on voltage slope	$R_L = 1.3\ \Omega$	—	See <a href="#">Figure 27</a>	—	V/ $\mu\text{s}$
$(dV_{OUT}/dt)_{off}$	Turn-off voltage slope	$R_L = 1.3\ \Omega$	—	See <a href="#">Figure 28</a>	—	V/ $\mu\text{s}$
$W_{ON}$	Switching energy losses during $t_{won}$	$R_L = 1.3\ \Omega$ (see <a href="#">Figure 6</a> )	—	2.5	—	mJ
$W_{OFF}$	Switching energy losses during $t_{woff}$	$R_L = 1.3\ \Omega$ (see <a href="#">Figure 6</a> )	—	1.2	—	mJ

Table 7. Logic inputs

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$V_{IL}$	Input low level voltage				0.9	V
$I_{IL}$	Low level input current	$V_{IN} = 0.9\text{ V}$	1			$\mu\text{A}$
$V_{IH}$	Input high level voltage		2.1			V
$I_{IH}$	High level input current	$V_{IN} = 2.1\text{ V}$			10	$\mu\text{A}$
$V_{I(hyst)}$	Input hysteresis voltage		0.25			V
$V_{ICL}$	Input clamp voltage	$I_{IN} = 1\text{ mA}$	5.5		7	V
		$I_{IN} = -1\text{ mA}$		-0.7		V
$V_{CSDL}$	CS_DIS low level voltage				0.9	V
$I_{CSDL}$	Low level CS_DIS current	$V_{CSD} = 0.9\text{ V}$	1			$\mu\text{A}$
$V_{CSDH}$	CS_DIS high level voltage		2.1			V
$I_{CSDH}$	High level CS_DIS current	$V_{CSD} = 2.1\text{ V}$			10	$\mu\text{A}$
$V_{CSD(hyst)}$	CS_DIS hysteresis voltage		0.25			V
$V_{CSCL}$	CS_DIS clamp voltage	$I_{CSD} = 1\text{ mA}$	5.5		7	V
		$I_{CSD} = -1\text{ mA}$		-0.7		V

Table 8. Protections and diagnostics<sup>(1)</sup>

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$I_{limH}$	DC short circuit current	$V_{CC} = 13\text{ V}$	70	100	140	A
		$5\text{ V} < V_{CC} < 24\text{ V}$			140	A
$I_{limL}$	Short circuit current during thermal cycling	$V_{CC} = 13\text{ V}; T_R < T_j < T_{TSD}$		25		A
$T_{TSD}$	Shutdown temperature		150	175	200	$^{\circ}\text{C}$
$T_R$	Reset temperature		$T_{RS} + 1$	$T_{RS} + 5$		$^{\circ}\text{C}$
$T_{RS}$	Thermal reset of status		135			$^{\circ}\text{C}$
$T_{HYST}$	Thermal hysteresis ( $T_{TSD} - T_R$ )			7		$^{\circ}\text{C}$
$V_{DEMAG}$	Turn-off output voltage clamp	$I_{OUT} = 2\text{ A}; V_{IN} = 0; L = 6\text{ mH}$	$V_{CC} - 28$	$V_{CC} - 31$	$V_{CC} - 35$	V
$V_{ON}$	Output voltage drop limitation	$I_{OUT} = 1\text{ A}; T_j = -40\text{ }^{\circ}\text{C} \dots 150\text{ }^{\circ}\text{C}$ (see <a href="#">Figure 8</a> )		25		mV

1. To ensure long term reliability under heavy overload or short circuit conditions, protection and related diagnostic signals must be used together with a proper software strategy. If the device is subjected to abnormal conditions, this software must limit the duration and number of activation cycles.

Table 9. Current sense (8 V < V<sub>CC</sub> < 18 V)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
K <sub>0</sub>	I <sub>OUT</sub> /I <sub>SENSE</sub>	I <sub>OUT</sub> = 5 A; V <sub>SENSE</sub> = 0.5 V T <sub>j</sub> = -40 °C...150 °C	8300	12640	17600	
K <sub>1</sub>	I <sub>OUT</sub> /I <sub>SENSE</sub>	I <sub>OUT</sub> = 10 A; V <sub>SENSE</sub> = 0.5 V T <sub>j</sub> = -40 °C...150 °C T <sub>j</sub> = 25 °C...150 °C	9200 9602	13220 13220	17300 16703	
dK <sub>1</sub> /K <sub>1</sub> <sup>(1)</sup>	Current sense ratio drift	I <sub>OUT</sub> = 10 A; V <sub>SENSE</sub> = 0.5 V; V <sub>CSD</sub> = 0 V; T <sub>j</sub> = -40 °C...150 °C	-13		13	%
K <sub>2</sub>	I <sub>OUT</sub> /I <sub>SENSE</sub>	I <sub>OUT</sub> = 15 A; V <sub>SENSE</sub> = 4 V T <sub>j</sub> = -40 °C...150 °C T <sub>j</sub> = 25 °C...150 °C	9500 10408	13120 13120	16900 15907	
dK <sub>2</sub> /K <sub>2</sub> <sup>(1)</sup>	Current sense ratio drift	I <sub>OUT</sub> = 15 A; V <sub>SENSE</sub> = 4 V; V <sub>CSD</sub> = 0 V; T <sub>j</sub> = -40 °C...150 °C	-10		10	%
K <sub>3</sub>	I <sub>OUT</sub> /I <sub>SENSE</sub>	I <sub>OUT</sub> = 25 A; V <sub>SENSE</sub> = 4 V T <sub>j</sub> = -40 °C...150 °C T <sub>j</sub> = 25 °C...150 °C	10600 11278	12920 12920	15600 14644	
dK <sub>3</sub> /K <sub>3</sub> <sup>(1)</sup>	Current sense ratio drift	I <sub>OUT</sub> = 25 A; V <sub>SENSE</sub> = 4 V; V <sub>CSD</sub> = 0 V; T <sub>j</sub> = -40 °C...150 °C	-7		7	%
I <sub>SENSE0</sub>	Analog sense leakage current	I <sub>OUT</sub> = 0 A; V <sub>SENSE</sub> = 0 V; V <sub>CSD</sub> = 5 V; V <sub>IN</sub> = 0 V; T <sub>j</sub> = -40 °C...150 °C	0		1	μA
		V <sub>CSD</sub> = 0 V; V <sub>IN</sub> = 5 V; T <sub>j</sub> = -40 °C...150 °C	0		2	μA
		I <sub>OUT</sub> = 10 A; V <sub>SENSE</sub> = 0 V; V <sub>CSD</sub> = V <sub>IN</sub> = 5 V	0		1	μA
V <sub>SENSE</sub>	Max analog sense output voltage	I <sub>OUT</sub> = 25 A; V <sub>CSD</sub> = 0 V	5			V
V <sub>SENSEH</sub> <sup>(2)</sup>	Analog sense output voltage in fault conditions	V <sub>CC</sub> = 13 V; R <sub>SENSE</sub> = 10 KΩ		8		V
I <sub>SENSEH</sub> <sup>(2)</sup>	Analog sense output current in fault conditions	V <sub>CC</sub> = 13 V; V <sub>SENSE</sub> = 5 V		7		mA
t <sub>DSENSE1H</sub>	Delay response time from falling edge of CS_DIS pin	V <sub>SENSE</sub> < 4 V, 5 A < I <sub>out</sub> < 25 A I <sub>SENSE</sub> = 90 % of I <sub>SENSE max</sub> (see <a href="#">Figure 4</a> )		50	100	μs
t <sub>DSENSE1L</sub>	Delay response time from rising edge of CS_DIS pin	V <sub>SENSE</sub> < 4 V, 5 A < I <sub>out</sub> < 25 A I <sub>SENSE</sub> = 10 % of I <sub>SENSE max</sub> (see <a href="#">Figure 4</a> )		5	20	μs

Table 9. Current sense (8 V < V<sub>CC</sub> < 18 V) (continued)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
t <sub>DSENSE2H</sub>	Delay response time from rising edge of INPUT pin	V <sub>SENSE</sub> < 4 V, 5 A < I <sub>out</sub> < 25 A I <sub>SENSE</sub> = 90 % of I <sub>SENSE max</sub> (see <a href="#">Figure 4</a> )		110	600	μs
Δt <sub>DSENSE2H</sub>	Delay response time between rising edge of output current and rising edge of current sense	V <sub>SENSE</sub> < 4V, I <sub>SENSE</sub> = 90 % of I <sub>SENSEMAX</sub> , I <sub>OUT</sub> = 90 % of I <sub>OUTMAX</sub> I <sub>OUTMAX</sub> = 10 (see <a href="#">Figure 7</a> )			300	μs
t <sub>DSENSE2L</sub>	Delay response time from falling edge of INPUT pin	V <sub>SENSE</sub> < 4 V, 5 A < I <sub>out</sub> < 25 A I <sub>SENSE</sub> = 10 % of I <sub>SENSE max</sub> (see <a href="#">Figure 4</a> )		100	250	μs

1. Parameter guaranteed by design; it is not tested.
2. Fault conditions includes: power limitation, overtemperature and open-load off-state detection.

Table 10. Open-load detection (8 V < V<sub>CC</sub> < 18 V)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V <sub>OL</sub>	Open-load off-state voltage detection threshold	V <sub>IN</sub> = 0 V, 8 V < V <sub>CC</sub> < 18 V	2	—	4	V
I <sub>OL</sub>	Open-load on-state current detection threshold	V <sub>IN</sub> = 5 V, 8 V < V <sub>CC</sub> < 18 V I <sub>SENSE</sub> = 5 μA	10	—	100	mA
t <sub>DSTKON</sub>	Output short circuit to V <sub>CC</sub> detection delay at turn off	See <a href="#">Figure 5</a>	180	—	1200	μs
I <sub>L(off2)</sub>	Off-state output current at V <sub>OUT</sub> = 4 V	V <sub>IN</sub> = 0 V; V <sub>SENSE</sub> = 0 V V <sub>OUT</sub> rising from 0 V to 4 V	-120	—	0	μA
td_vol	Delay response from output rising edge to V <sub>SENSE</sub> rising edge in open-load	V <sub>IN</sub> = 0 V; V <sub>OUT</sub> = 4 V V <sub>SENSE</sub> = 90 % of V <sub>SENSEH</sub>		—	20	μs

Figure 4. Current sense delay characteristics

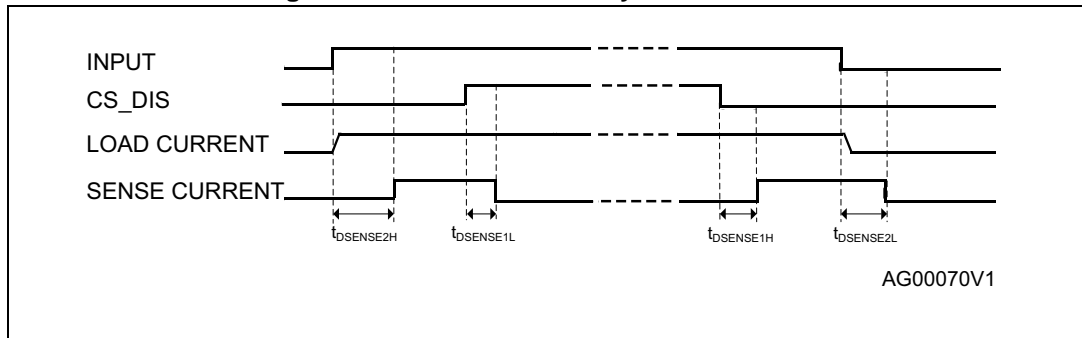


Figure 5. Open-load off-state delay timing

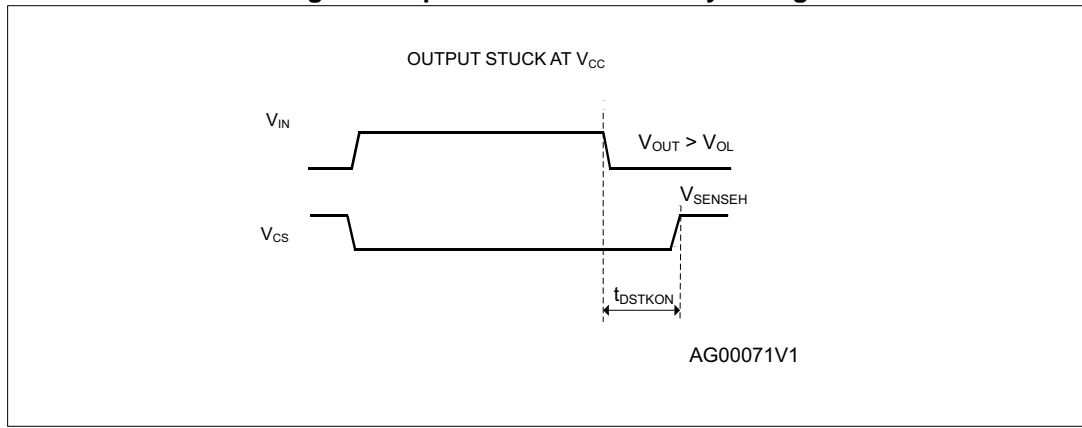


Figure 6. Switching characteristics

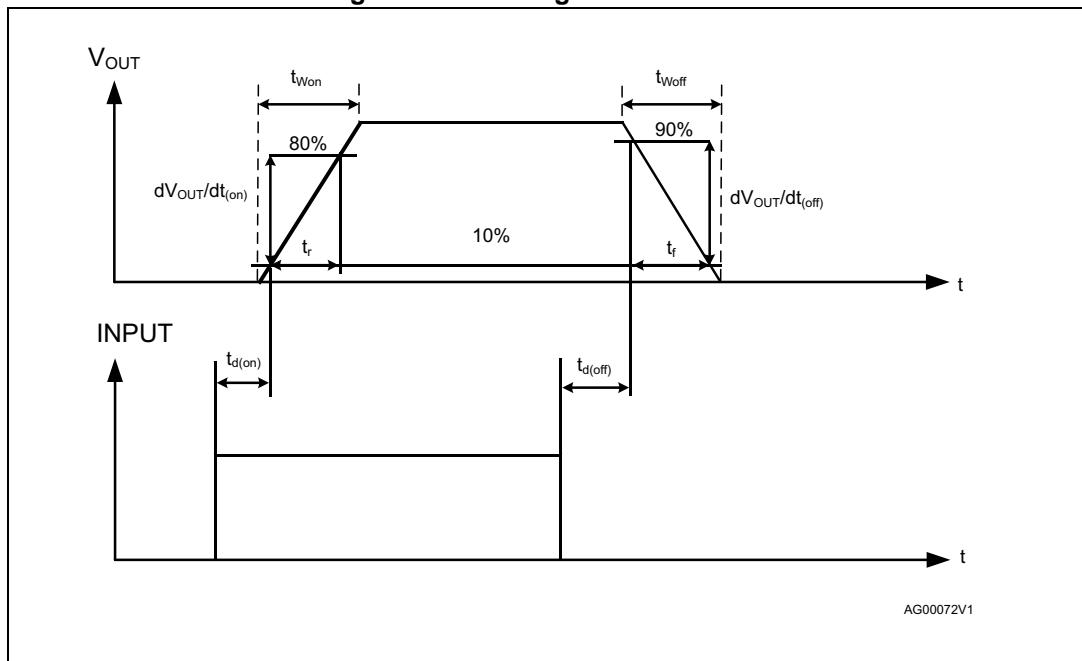


Figure 7. Delay response time between rising edge of output current and rising edge of current sense (CS enabled)

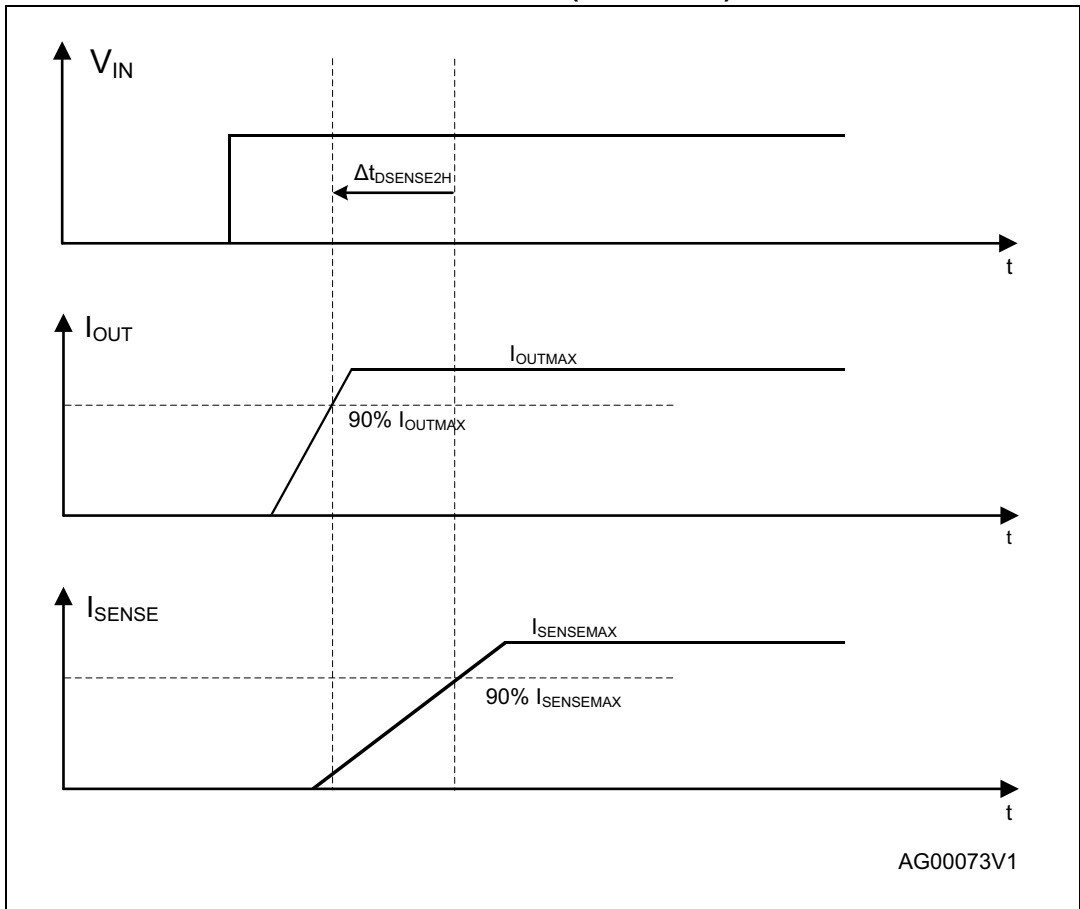


Figure 8. Output voltage drop limitation

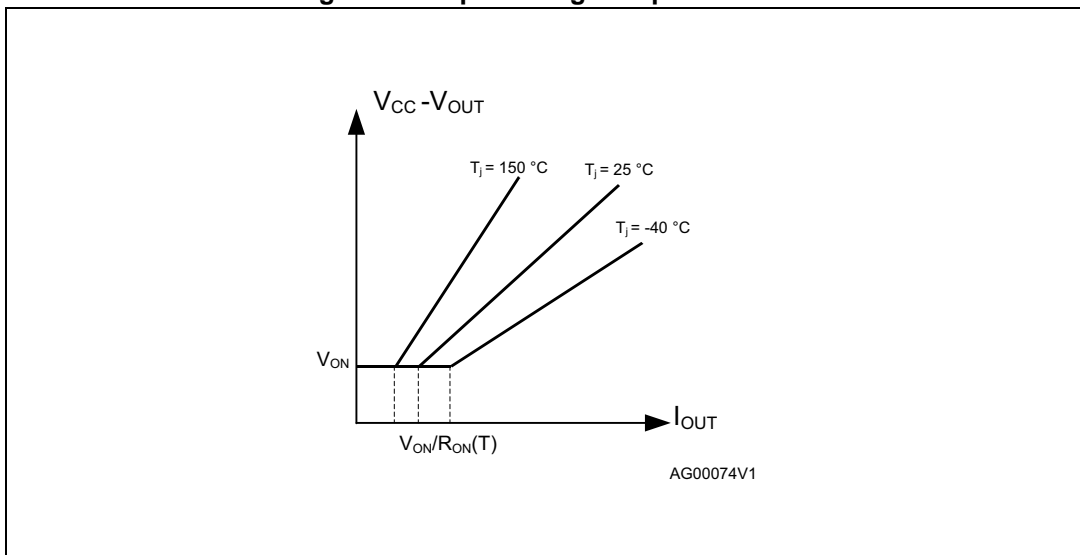


Figure 9.  $I_{OUT}/I_{SENSE}$  vs  $I_{OUT}$

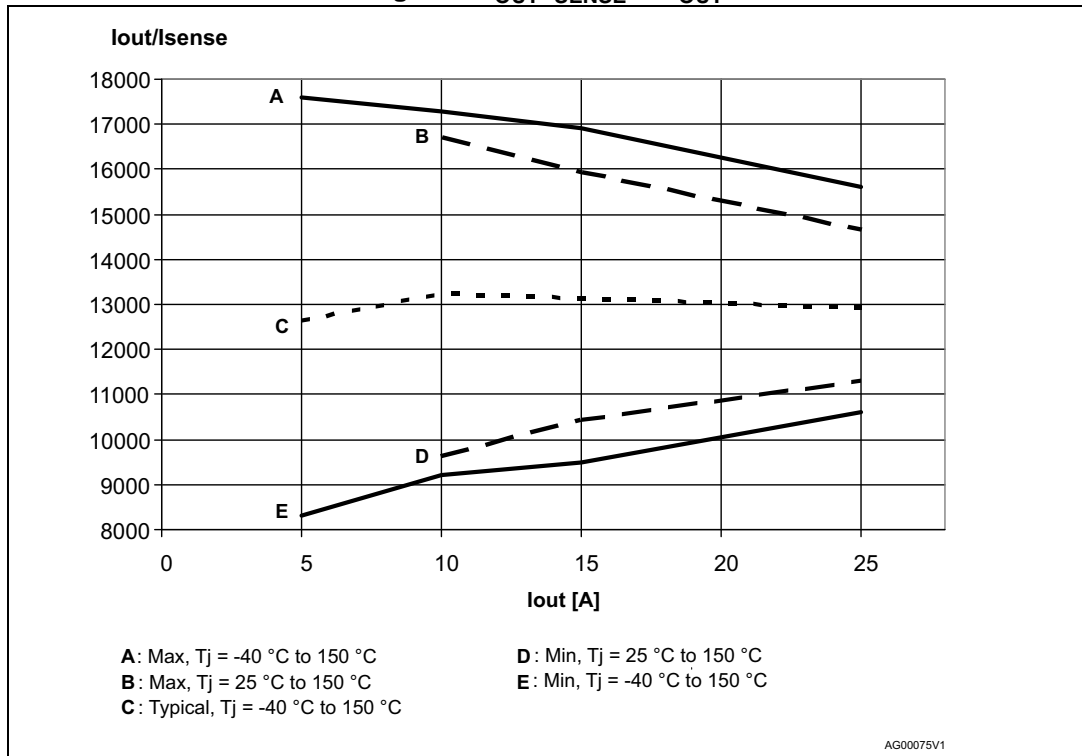
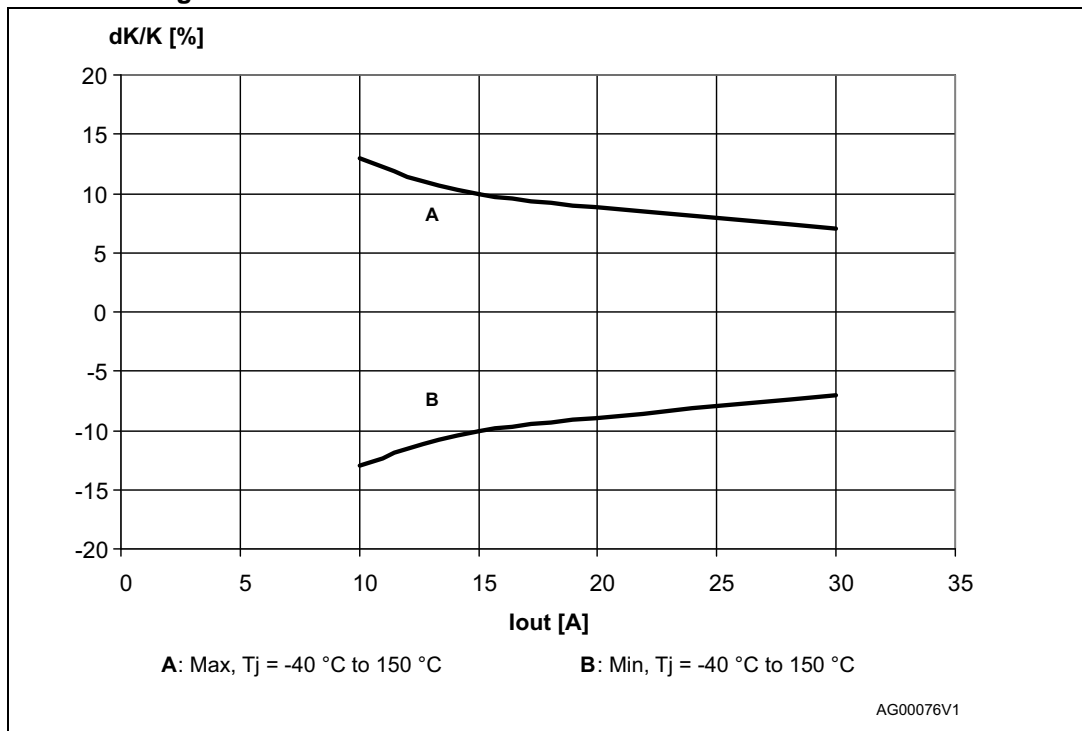


Figure 10. Maximum current sense ratio drift vs load current<sup>(1)</sup>



1. Parameter guaranteed by design; it is not tested.

Table 11. Truth table

Conditions	Input	Output	Sense ( $V_{CSD} = 0\text{ V}$ ) <sup>(1)</sup>
Normal operation	L	L	0
	H	H	Nominal
Overtemperature	L	L	0
	H	L	$V_{SENSEH}$
Undervoltage	L	L	0
	H	L	0
Overload	H	X (no power limitation)	Nominal
	H	Cycling (power limitation)	$V_{SENSEH}$
Short circuit to GND (Power limitation)	L	L	0
	H	L	$V_{SENSEH}$
Open-load off-state (with external pull up)	L	H	$V_{SENSEH}$
Short circuit to $V_{CC}$ (external pull up disconnected)	L	H	$V_{SENSEH}$
	H	H	$V_{SENSEH}$ < Nominal
Negative output voltage clamp	L	L	0

1. If the  $V_{CSD}$  is high, the SENSE output is at a high impedance, its potential depends on leakage currents and external circuit.



**Table 12. Electrical transient requirements (part 1/3)**

ISO 7637-2: 2004(E) Test Pulse	Test levels <sup>(1)</sup>		Number of pulses or test times	Burst cycle/pulse repetition time		Delays and impedance
	III	IV				
1	-75 V	-100 V	5000 pulses	0.5 s	5 s	2 ms, 10 Ω
2a	+37 V	+50 V	5000 pulses	0.2 s	5 s	50 μs, 2 Ω
3a	-100 V	-150 V	1h	90 ms	100 ms	0.1 μs, 50 Ω
3b	+75 V	+100 V	1h	90 ms	100 ms	0.1 μs, 50 Ω
4	-6 V	-7 V	1 pulse			100 ms, 0.01 Ω
5b <sup>(2)</sup>	+65 V	+87 V	1 pulse			400 ms, 2 Ω

1. The above test levels must be considered referred to  $V_{CC} = 13.5$  V except for pulse 5b.
2. Valid in case of external load dump clamp: 40V maximum referred to ground. The protection strategy allows PowerMOS to be cyclically switched on during load dump, so distributing the load dump energy along the time and to transfer a part of it to the load.

**Table 13. Electrical transient requirements (part 2/3)**

ISO 7637-2: 2004(E) Test Pulse	Test level results <sup>(1)</sup>	
	III	IV
1	C	C
2a	C	C
3a	C	C
3b	C	C
4	C	C
5b <sup>(2)(3)</sup>	C	C

1. The above test levels must be considered referred to  $V_{CC} = 13.5$  V except for pulse 5b.
2. Valid in case of external load dump clamp: 40V maximum referred to ground. The protection strategy allows PowerMOS to be cyclically switched on during load dump, so distributing the load dump energy along the time and to transfer a part of it to the load.
3. Suppressed load dump (pulse 5b) is withstood with a minimum load connected as specified in [Table 3: Absolute maximum ratings](#).

**Table 14. Electrical transient requirements (part 3/3)**

Class	Contents
C	All functions of the device are performed as designed after exposure to disturbance.
E	One or more functions of the device are not performed as designed after exposure to disturbance and cannot be returned to proper operation without replacing the

## 2.4 Waveforms

Figure 11. Normal operation

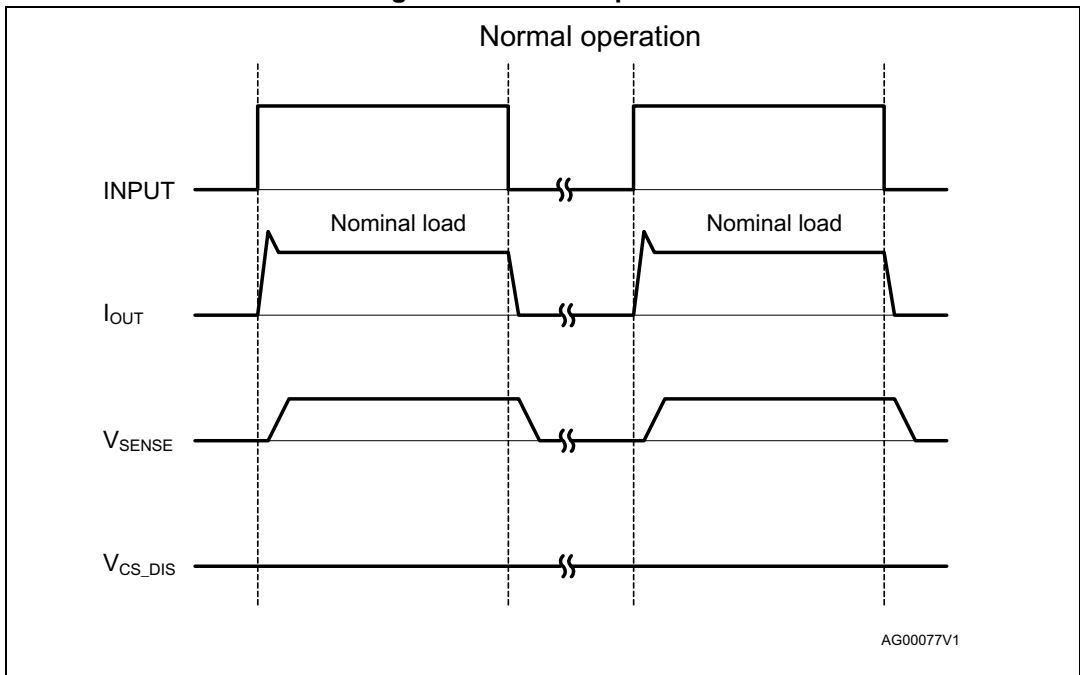


Figure 12. Overload or short to GND

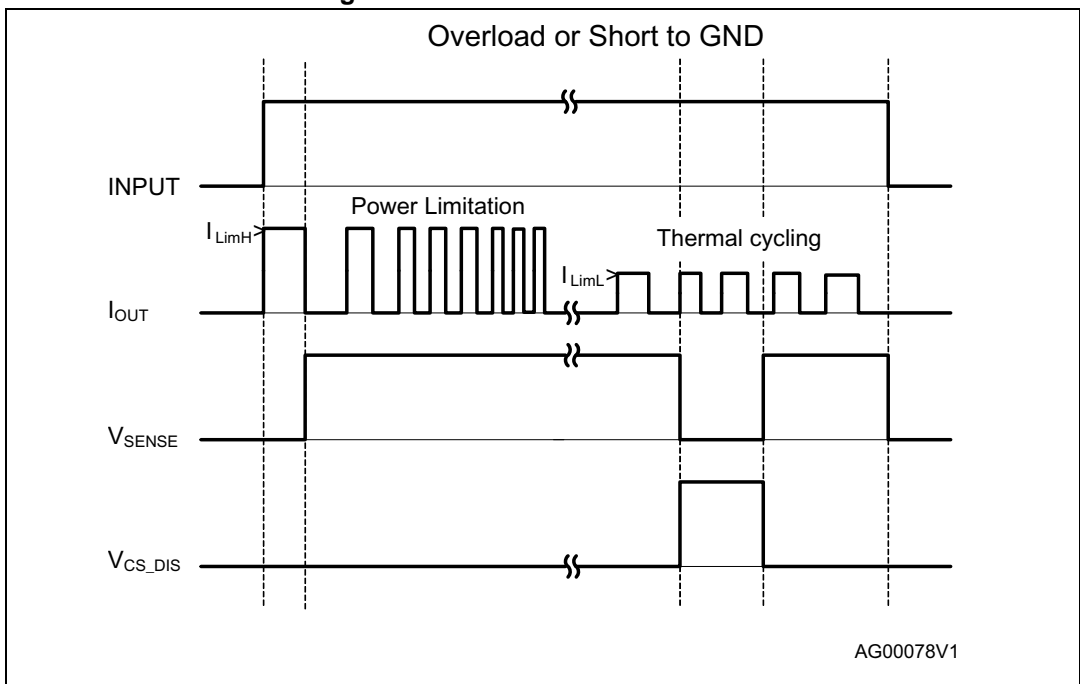


Figure 13. Intermittent overload

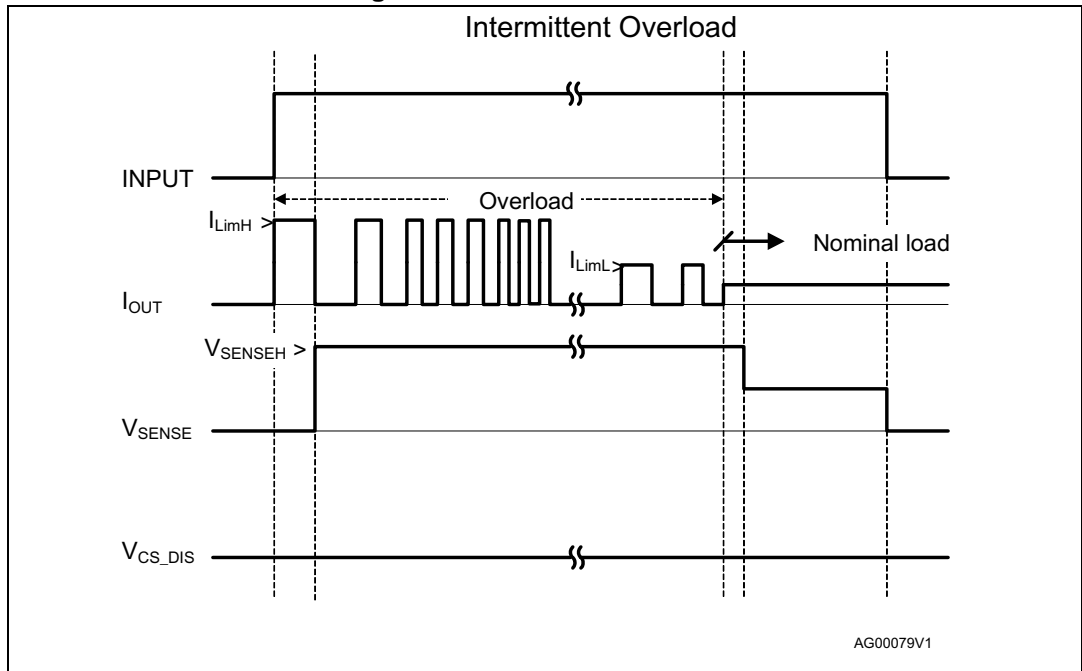


Figure 14. Off-state open-load with external circuitry

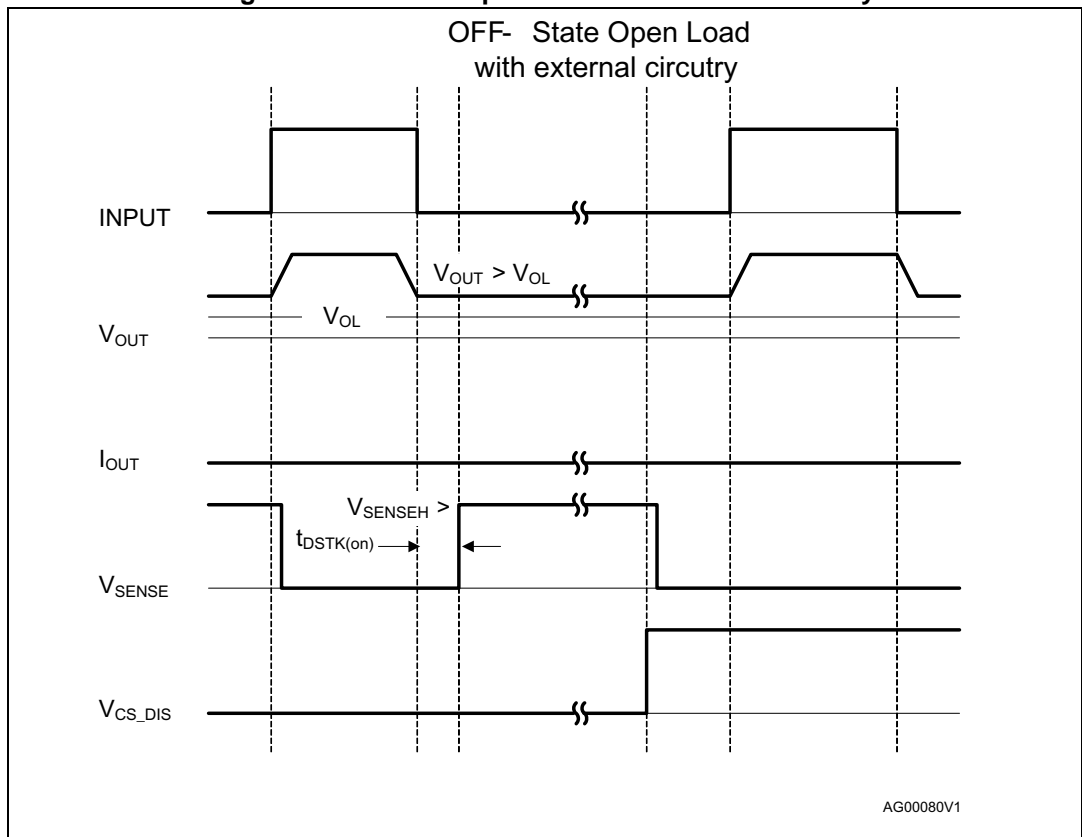


Figure 15. Short to V<sub>CC</sub>

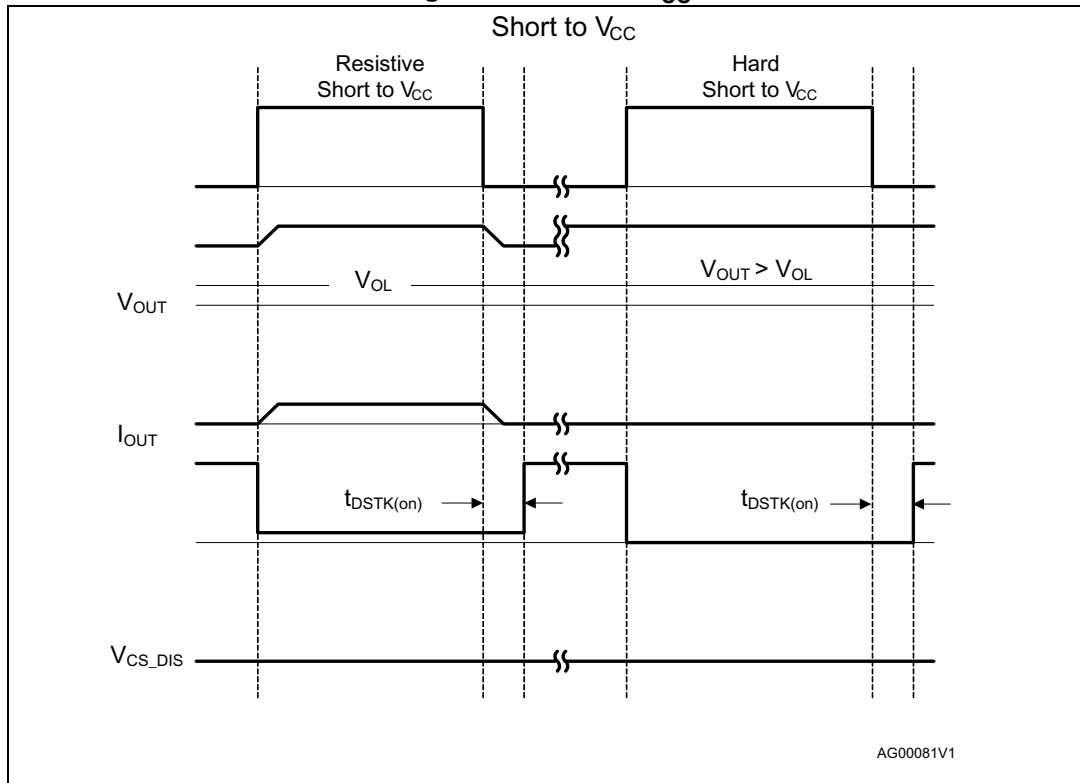
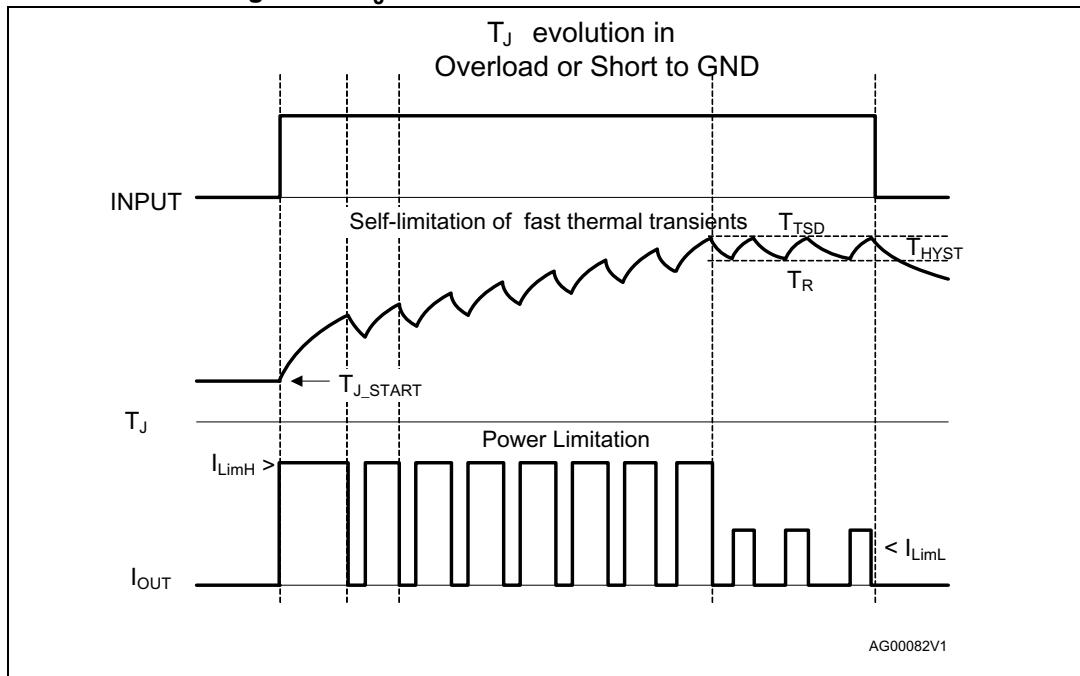


Figure 16. T<sub>J</sub> evolution in overload or short to GND



## 2.5 Electrical characteristics curves

Figure 17. Off-state output current

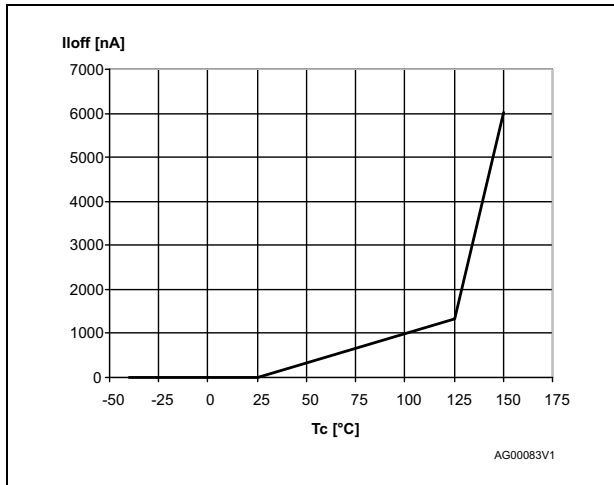


Figure 18. High level input current

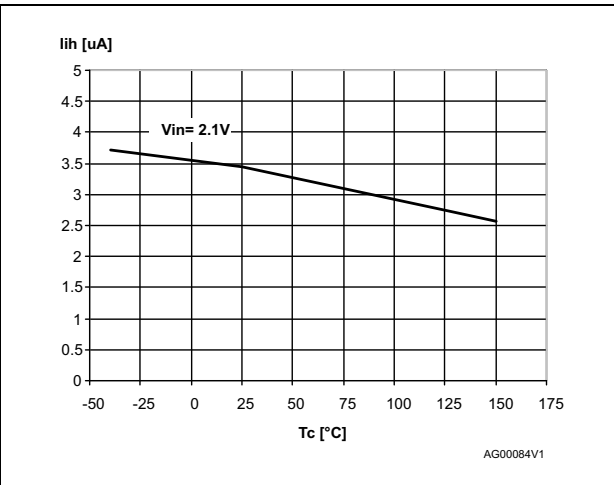


Figure 19. Input clamp voltage

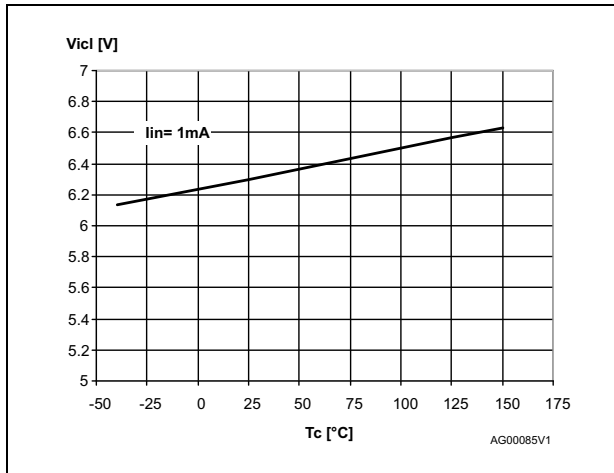


Figure 20. Input high level voltage

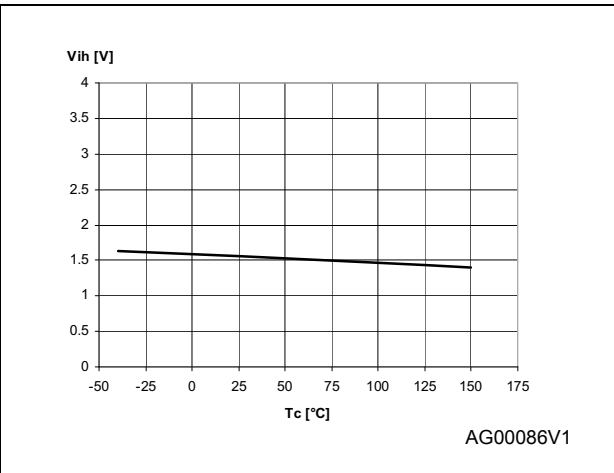


Figure 21. Input low level voltage

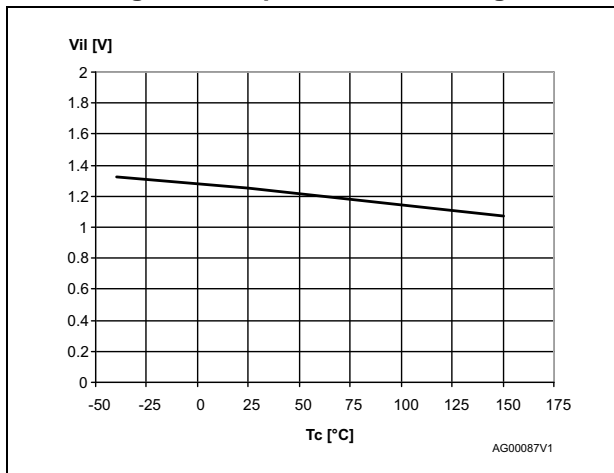


Figure 22. Input hysteresis voltage

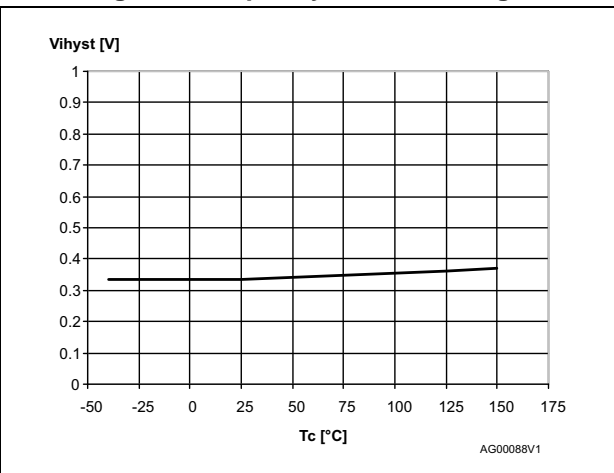


Figure 23. On-state resistance vs  $T_{case}$

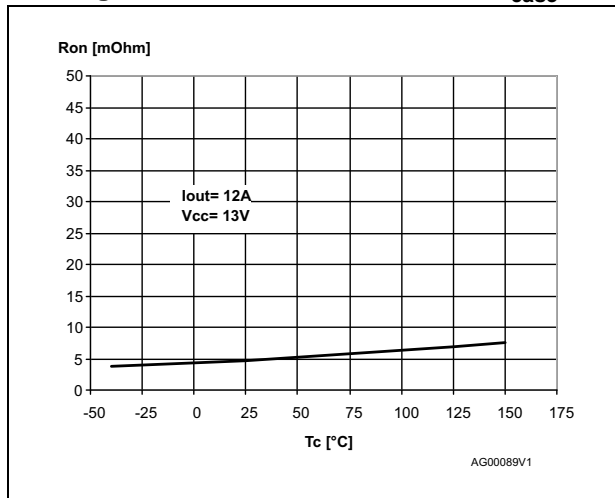


Figure 24. On-state resistance vs  $V_{CC}$

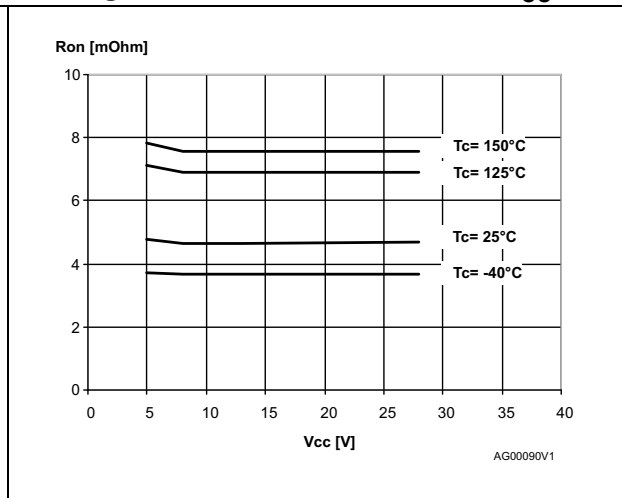


Figure 25. Undervoltage shutdown

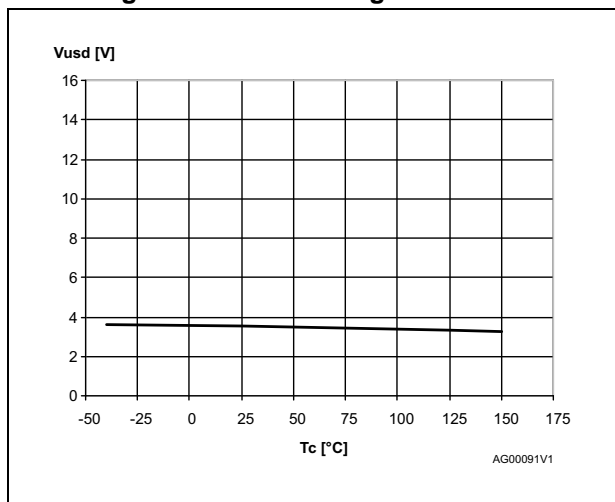


Figure 26.  $I_{LIMH}$  vs  $T_{case}$

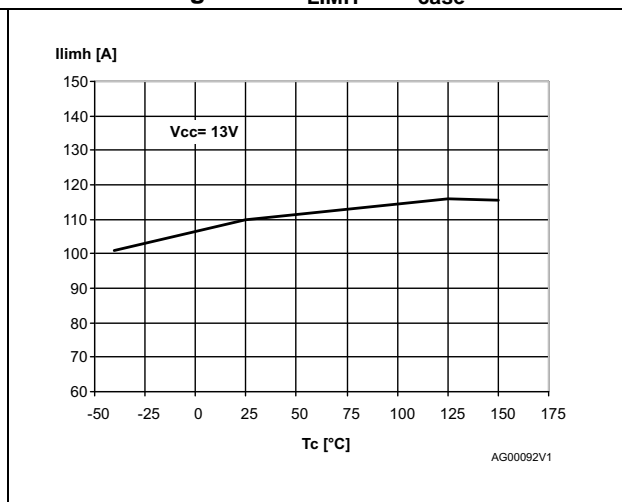


Figure 27. Turn-on voltage slope

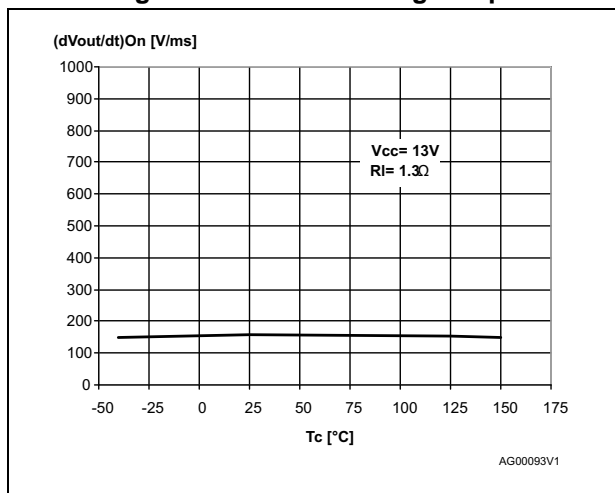


Figure 28. Turn-off voltage slope

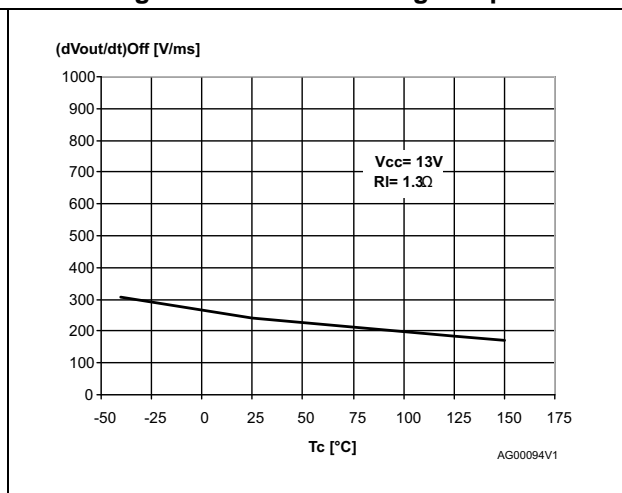


Figure 29. CS\_DIS clamp voltage

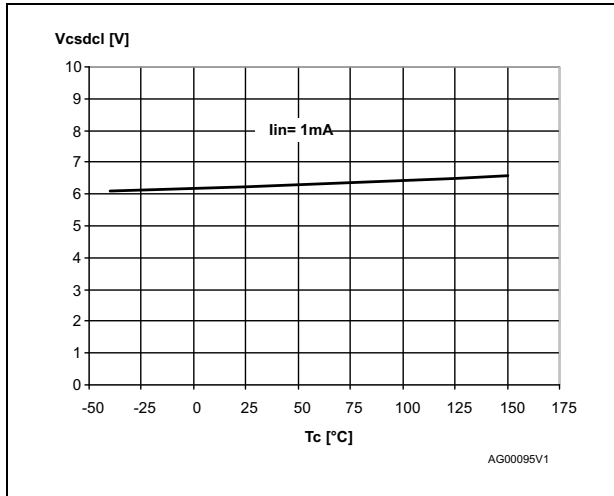


Figure 30. Low level CS\_DIS voltage

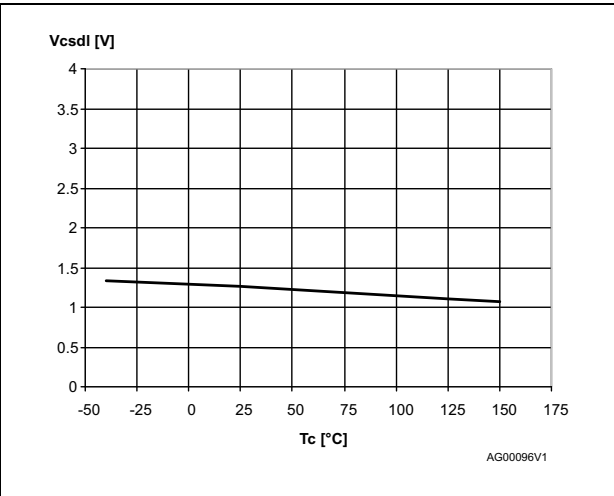
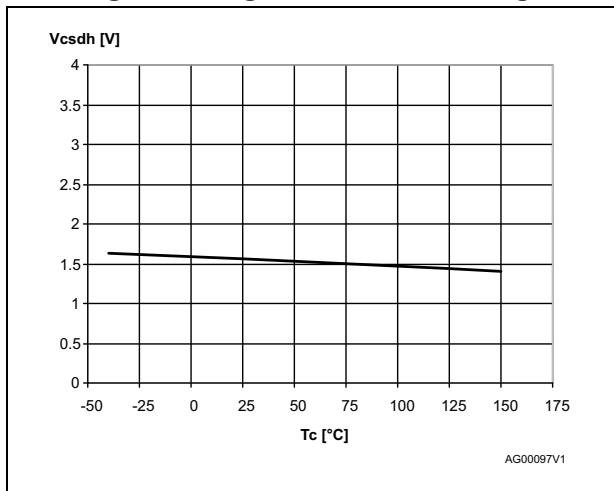
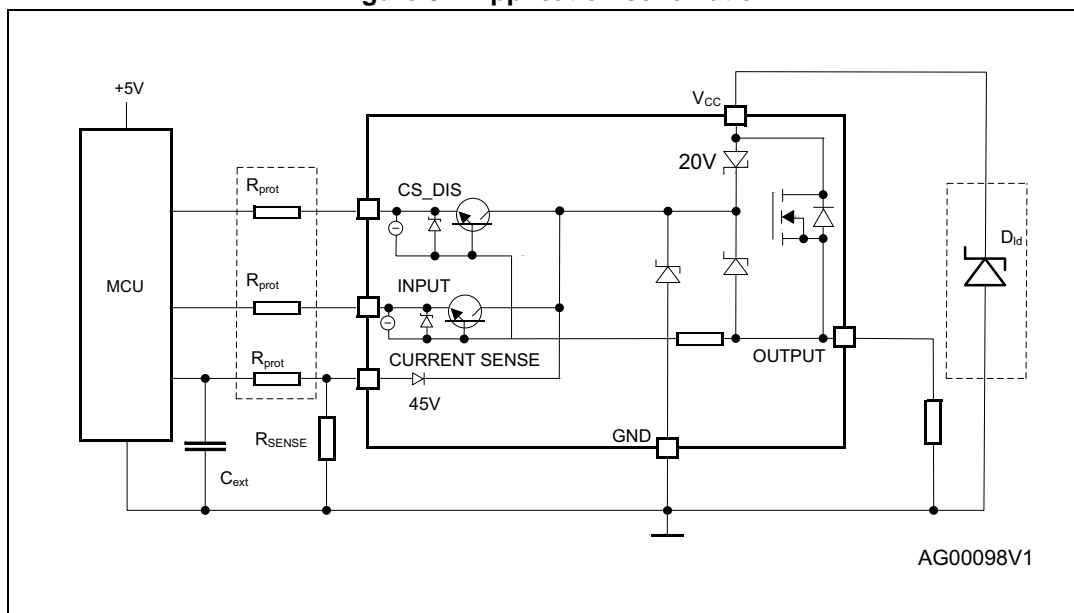


Figure 31. High level CS\_DIS voltage



### 3 Application information

Figure 32. Application schematic



Note: Channel 2 has the same internal circuit as channel 1.

#### 3.1 Load dump protection

$D_{id}$  is necessary (voltage transient suppressor) if the load dump peak voltage exceeds the  $V_{CC}$  max DC rating. The same applies if the device is subject to transients on the  $V_{CC}$  line that are greater than the ones shown in the ISO 7637-2 2004 (E) table.

#### 3.2 MCU I/Os protection

When negative transients are present on the  $V_{CC}$  line, the control pins are pulled negative to approximately -1.5 V. ST suggests to insert a resistor ( $R_{prot}$ ) in line to prevent the microcontroller I/O pins from latching-up.

The value of these resistors is a compromise between the leakage current of microcontroller and the current required by the HSD I/Os (Input levels compatibility) with the latch-up limit of microcontroller I/Os.

##### Equation 1

$$-V_{CCpeak}/I_{latchup} \leq R_{prot} \leq (V_{OH\mu C} - V_{IH}) / I_{IHmax}$$

Calculation example:

For  $V_{CCpeak} = -1.5$  V and  $I_{latchup} \geq 20$  mA;  $V_{OH\mu C} \geq 4.5$  V

$75 \Omega \leq R_{prot} \leq 240$  k $\Omega$ .

Recommended values:  $R_{prot} = 10$  k $\Omega$ ,  $C_{EXT} = 10$  nF.



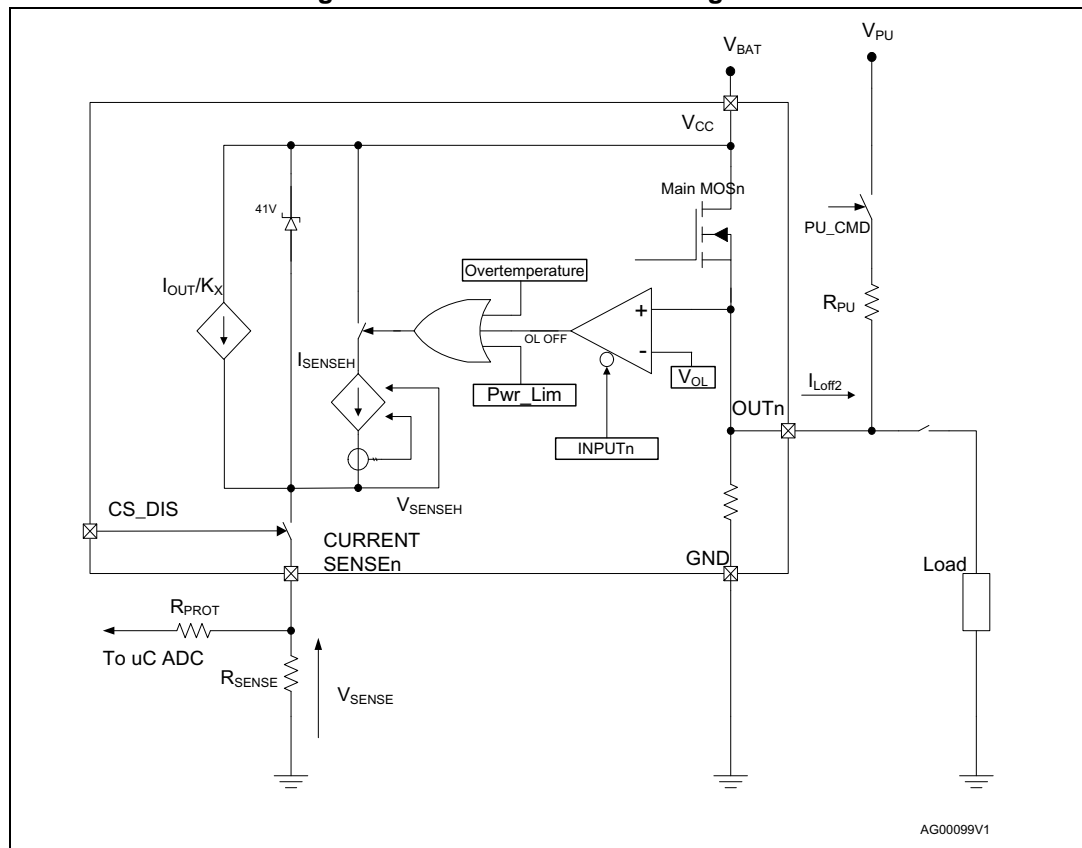
### 3.3 Current sense and diagnostic

The current sense pin performs a double function (see [Figure 33: Current sense and diagnostic](#)):

- Current mirror of the load current in normal operation**, delivering a current proportional to the load one according to a known ratio  $K_X$ .  
 The current  $I_{SENSE}$  can be easily converted to a voltage  $V_{SENSE}$  by means of an external resistor  $R_{SENSE}$ . Linearity between  $I_{OUT}$  and  $V_{SENSE}$  is ensured up to 5V minimum (see parameter  $V_{SENSE}$  in [Table 9: Current sense \(8 V < VCC < 18 V\)](#)). The current sense accuracy depends on the output current (refer to current sense electrical characteristics [Table 9: Current sense \(8 V < VCC < 18 V\)](#)).
- Diagnostic flag in fault conditions**, delivering a fixed voltage  $V_{SENSEH}$  up to a maximum current  $I_{SENSEH}$  in case of the following fault conditions (refer to [Truth table](#)):
  - Power limitation activation
  - Overtemperature
  - Short to  $V_{CC}$  in off-state
  - Open-load in off-state with additional external components.

A logic level high on CS\_DIS pin sets at the same time all the current sense pins of the device in a high impedance state, thus disabling the current monitoring and diagnostic detection. This feature allows multiplexing of the microcontroller analog inputs by sharing of sense resistance and ADC line among different devices.

**Figure 33. Current sense and diagnostic**



### 3.3.1 Short to $V_{CC}$ and off-state open-load detection

Short to  $V_{CC}$

A short circuit between  $V_{CC}$  and output is indicated by the relevant current sense pin set to  $V_{SENSEH}$  during the device off-state. Small or no current is delivered by the current sense during the on-state depending on the nature of the short circuit.

Off-state open-load with external circuitry

Detection of an open-load in off mode requires an external pull-up resistor  $R_{PU}$  connecting the output to a positive supply voltage  $V_{PU}$ .

It is preferable  $V_{PU}$  to be switched off during the module standby mode in order to avoid the overall standby current consumption to increase in normal conditions, i.e. when load is connected.

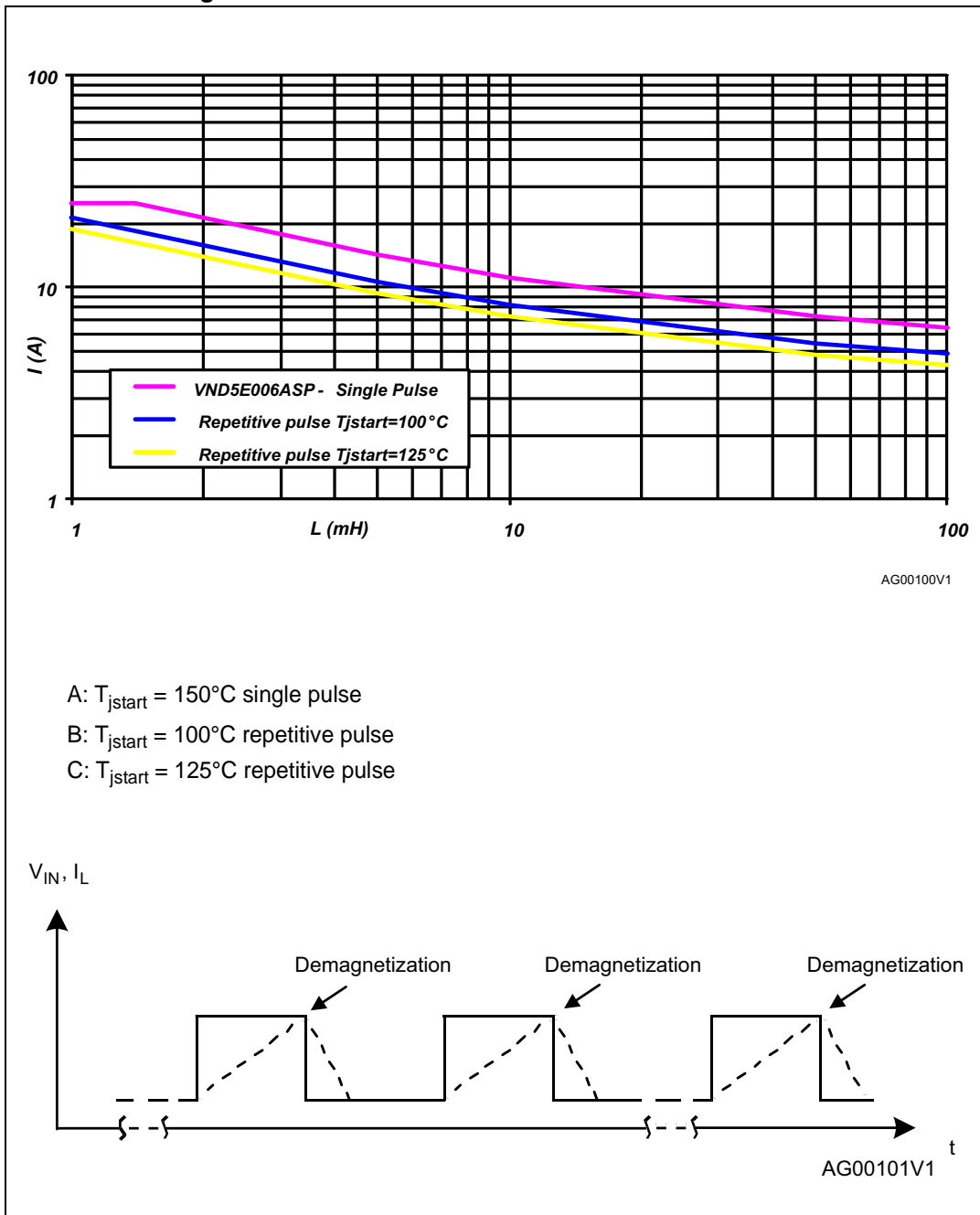
For proper open-load detection in off-state, the external pull-up resistor must be selected according to the following formula:

$$V_{OUT}|_{Pull-up\_ON} = \frac{R_{PD} \cdot V_{PU} - R_{PU} \cdot R_{PD} \cdot I_{L(off2)}}{R_{PU} + R_{PD}} > V_{OLmax} = 4V$$

For the values of  $V_{OLmin}$ ,  $V_{OLmax}$ , and  $I_{L(off2)}$  see [Table 10: Open-load detection \(8 V < VCC < 18 V\)](#).

### 3.4 Maximum demagnetization energy ( $V_{CC} = 13.5\text{ V}$ )

Figure 34. Maximum turn-off current versus inductance

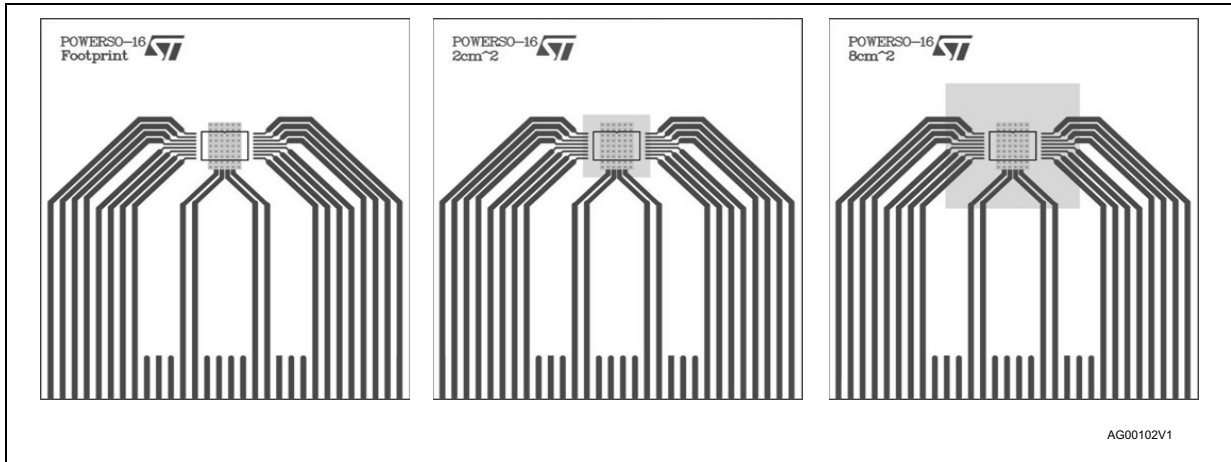


Note: Values are generated with  $R_L = 0\ \Omega$ .  
 In case of repetitive pulses,  $T_{jstart}$  (at beginning of each demagnetization) of every pulse must not exceed the temperature specified above for curves A and B.

## 4 Package and PCB thermal data

### 4.1 PowerSO-16 thermal data

Figure 35. PowerSO-16 PC board<sup>(1)</sup>



1. Layout condition of  $R_{th}$  and  $Z_{th}$  measurements (PCB: Double layer, Thermal Vias, FR4 area= 77mm x 86mm, PCB thickness=1.6mm, Cu thickness=70 $\mu$ m (front and back side), Copper areas: from minimum pad lay-out to 8cm<sup>2</sup>).

Figure 36.  $R_{thj-amb}$  vs PCB copper area in open box free air condition (one channel ON)

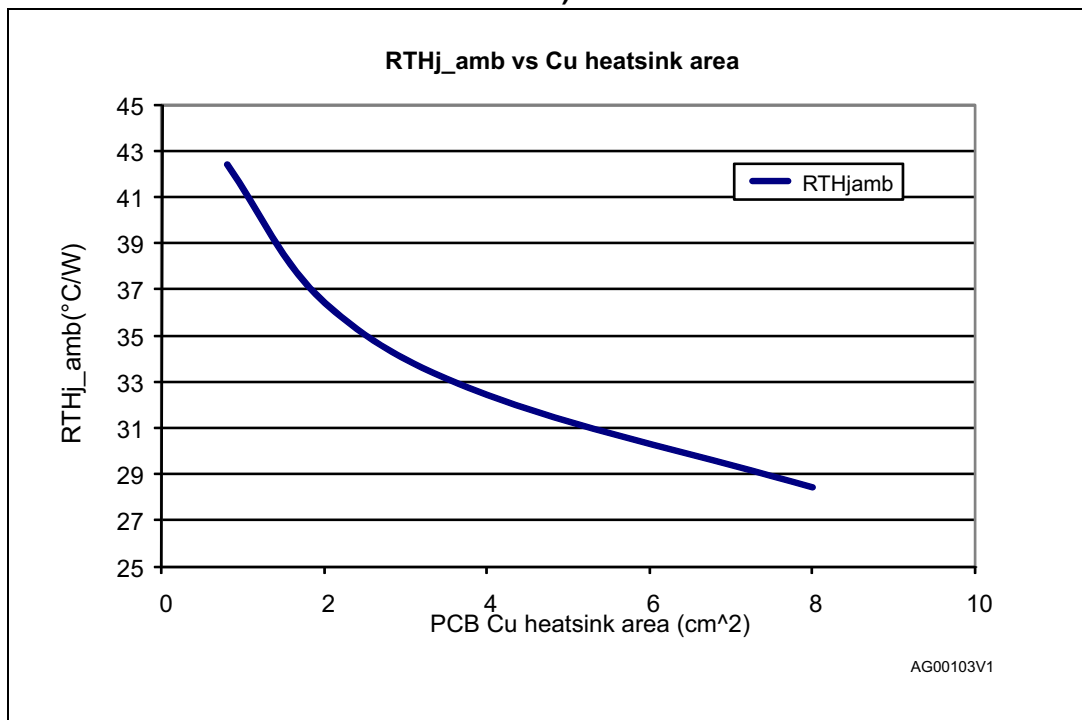


Figure 37. PowerSO-16 thermal impedance junction ambient single pulse (one channel ON)

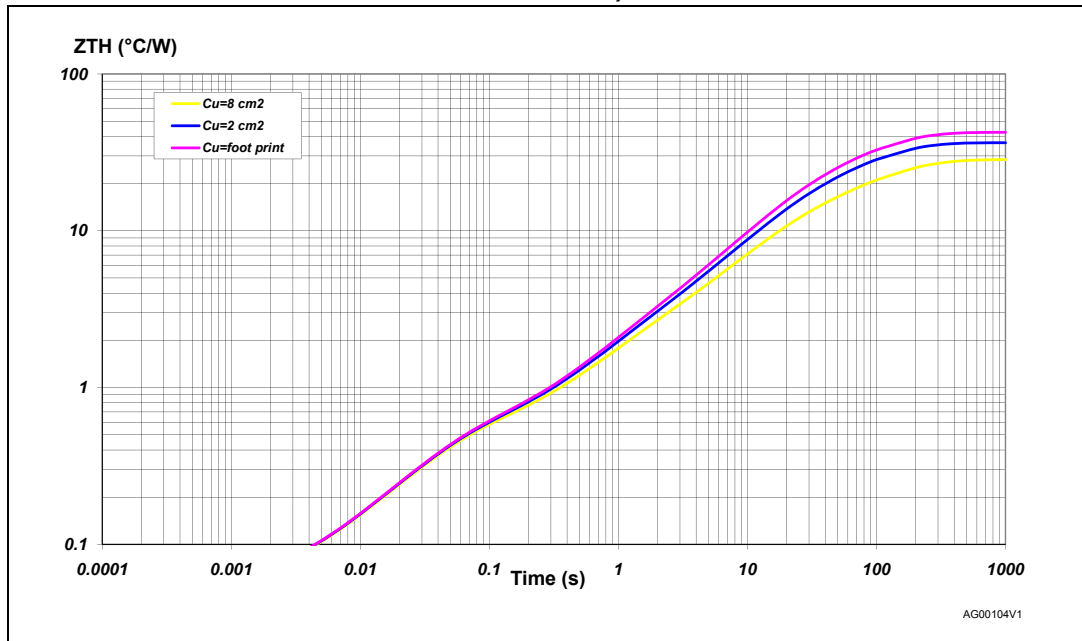
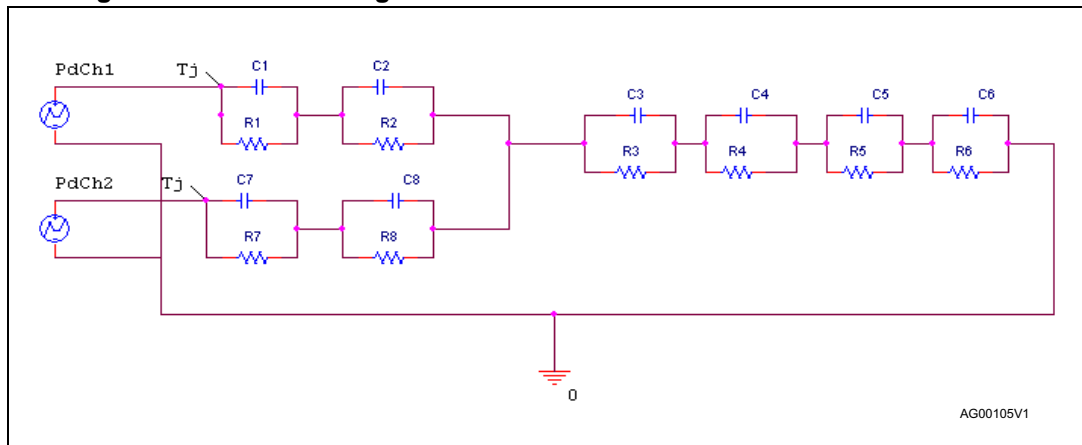


Figure 38. Thermal fitting model of a double channel HSD in PowerSO-16<sup>(1)</sup>



1. The fitting model is a simplified thermal tool and is valid for transient evolutions where the embedded protections (power limitation or thermal cycling during thermal shutdown) are not triggered.

**Equation 2: pulse calculation formula**

$$Z_{TH\delta} = R_{TH} \cdot \delta + Z_{THtp}(1 - \delta)$$

where  $\delta = t_p/T$

**Table 15. Thermal parameters**

Area/island (cm <sup>2</sup> )	Footprint	2	8
R1=R7 (°C/W)	0.05		
R2=R8 (°C/W)	0.4		
R3 (°C/W)	1		
R4 (°C/W)	7		
R5 (°C/W)	12	10	8
R6 (°C/W)	22	18	12
C1=C7 (W.s/°C)	0.01		
C2=C8 (W.s/°C)	0.1		
C3 (W.s/°C)	1		
C4 (W.s/°C)	2		
C5 (W.s/°C)	3	4	7
C6 (W.s/°C)	5	6	12

## 5 Package information

### 5.1 ECOPACK<sup>®</sup> packages

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK<sup>®</sup> packages, depending on their level of environmental compliance. ECOPACK<sup>®</sup> specifications, grade definitions and product status are available at: [www.st.com](http://www.st.com). ECOPACK<sup>®</sup> is an ST trademark.

### 5.2 PowerSO-16 mechanical data

Figure 39. PowerSO-16 package dimensions

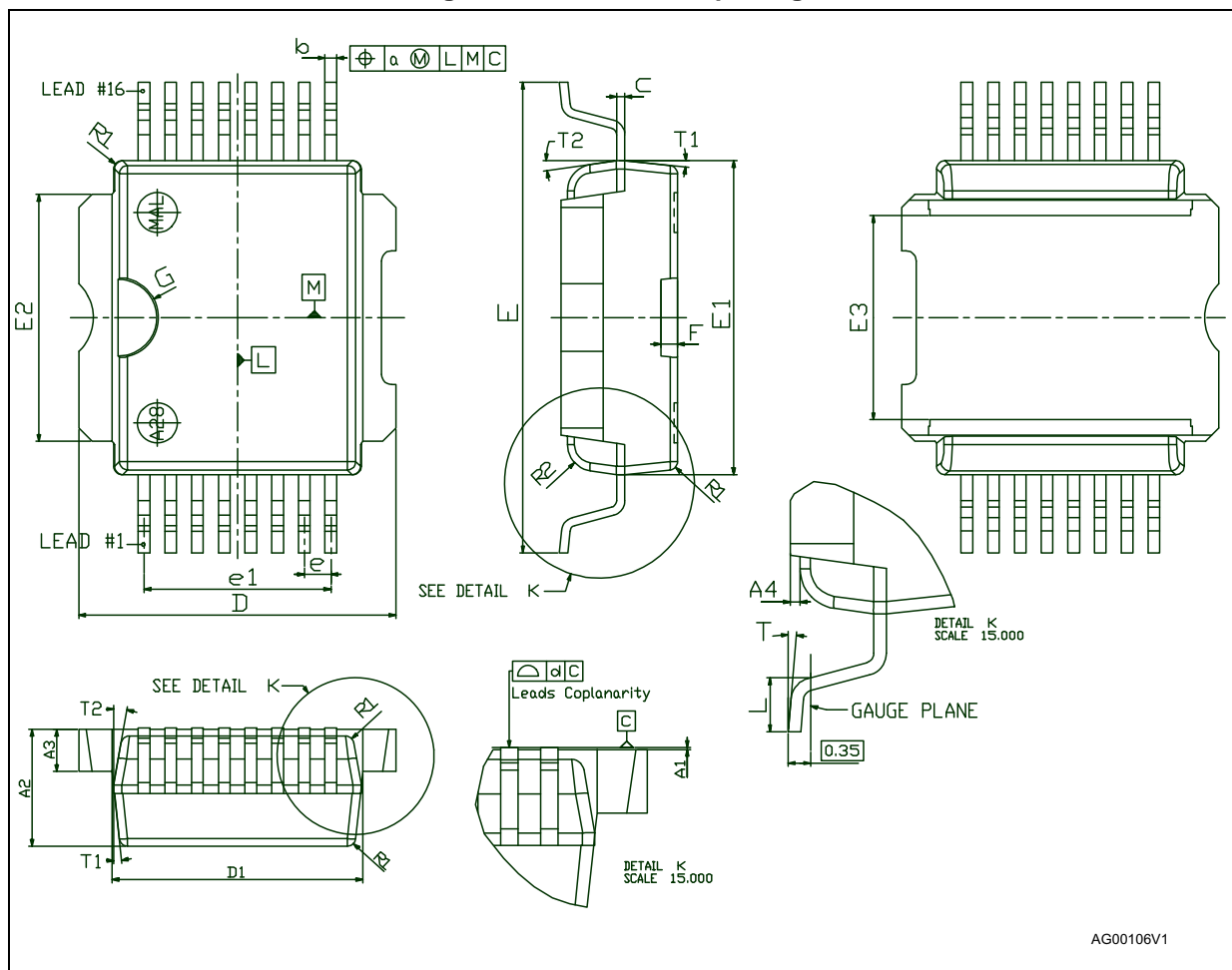


Table 16. PowerSO-16 mechanical data

Dim.	mm		
	Min.	Typ.	Max.
A1	0	0.05	0.1
A2	3.4	3.5	3.6
A3	1.2	1.3	1.4
A4	0.15	0.2	0.25
a		0.2	
b	0.27	0.35	0.43
c	0.23	0.27	0.32
D	9.4	9.5	9.6
D1	7.4	7.5	7.6
d	0	0.05	0.1
E (1)	13.85	14.1	14.35
E1	9.3	9.4	9.5
E2	7.3	7.4	7.5
E3	5.9	6.1	6.3
e		0.8	
e1		5.6	
F		0.5	
G		1.2	
L	0.8	1	1.1
R1			0.25
R2		0.8	
T	2°	5°	8°
T1	6° (typ.)		
T2	10° (typ.)		



### 5.3 Packing information

Figure 40. PowerSO-16 tube shipment (no suffix)

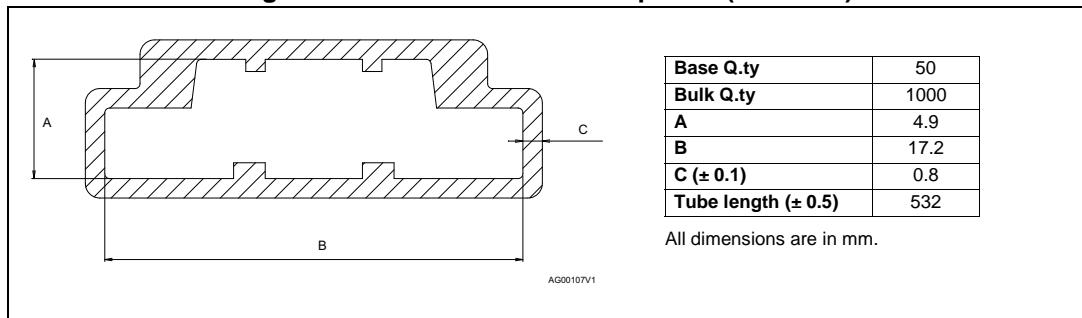


Figure 41. PowerSO-16 tape and reel shipment (suffix "TR")

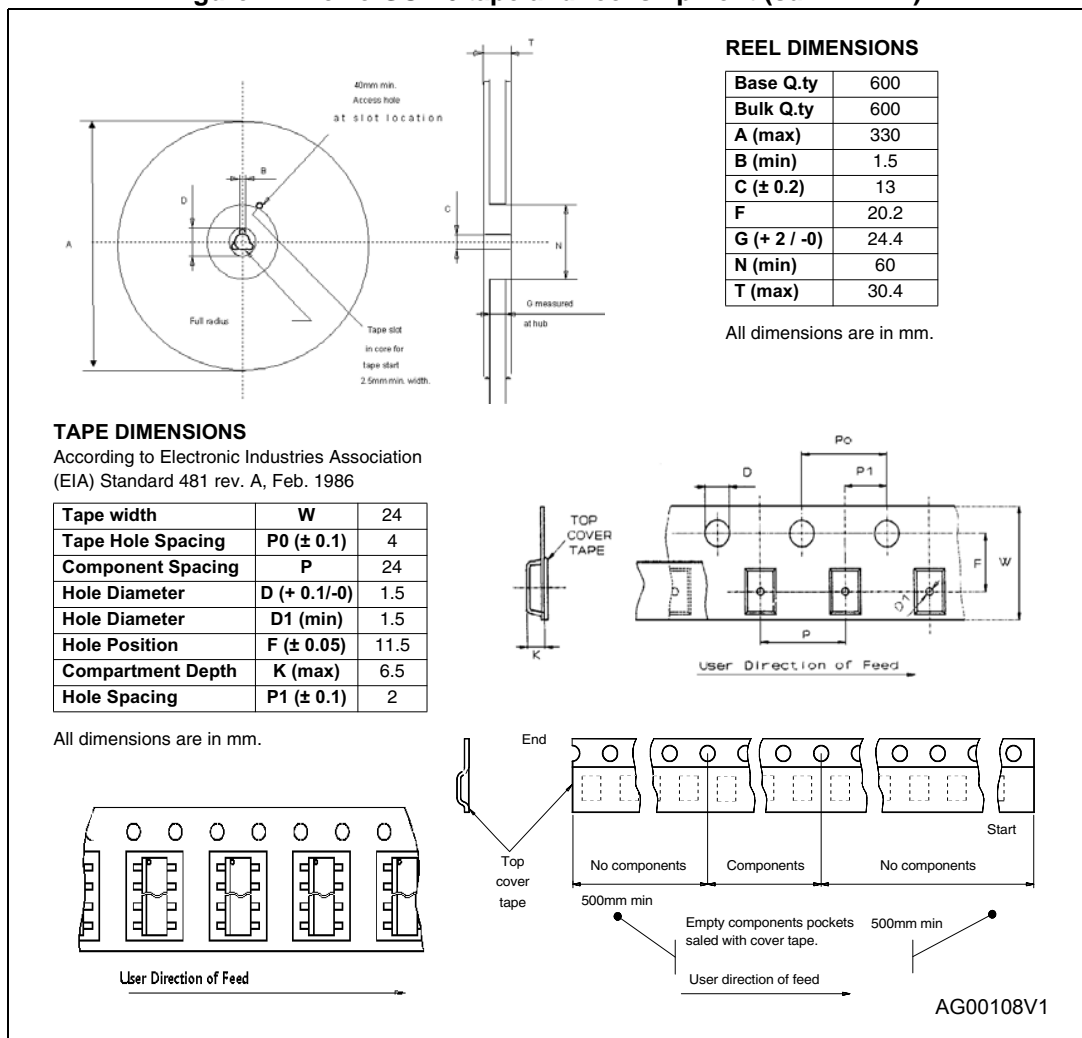
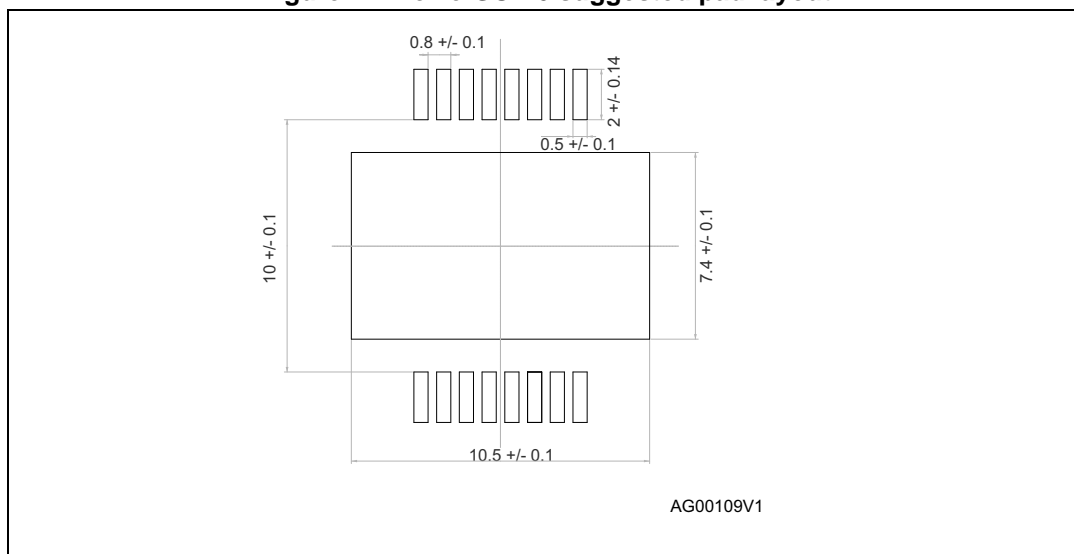


Figure 42. PowerSO-16 suggested pad layout



## 6 Order codes

Table 17. Device summary

Package	Order codes	
	Tube	Tape and reel
PowerSO-16	VND5E006ASP-E	VND5E006ASPTR-E

## 7 Revision history

**Table 18. Document revision history**

Date	Revision	Changes
18-Apr-2010	1	Initial release.
02-Jul-2010	2	Updated <i>Features</i> list.
21-Jul-2010	3	Updated <i>Table 9: Current sense (8 V &lt; VCC &lt; 18 V)</i> .
19-Jan-2011	4	Added <i>Section 3.4: Maximum demagnetization energy (VCC = 13.5 V)</i> <i>Table 3: Absolute maximum ratings:</i> – E <sub>MAX</sub> : updated value and test condition <i>Table 4: Thermal data</i> – Added R <sub>thj-case</sub> row
19-Sep-2013	5	Updated Disclaimer.
28-Oct-2013	6	Updated footnote 2 into the <i>Table 12: Electrical transient requirements (part 1/3)</i> and <i>Table 13: Electrical transient requirements (part 2/3)</i> .

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- Поставка специализированных компонентов военного и аэрокосмического уровня качества (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Actel, Aeroflex, Peregrine, VPT, Syfer, Eurofarad, Texas Instruments, MS Kennedy, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

Компания «Океан Электроники» является официальным дистрибьютором и эксклюзивным представителем в России одного из крупнейших производителей разъемов военного и аэрокосмического назначения «**JONHON**», а так же официальным дистрибьютором и эксклюзивным представителем в России производителя высокотехнологичных и надежных решений для передачи СВЧ сигналов «**FORSTAR**».



**JONHON**

«**JONHON**» (основан в 1970 г.)

Разъемы специального, военного и аэрокосмического назначения:

(Применяются в военной, авиационной, аэрокосмической, морской, железнодорожной, горно- и нефтедобывающей отраслях промышленности)

«**FORSTAR**» (основан в 1998 г.)

ВЧ соединители, коаксиальные кабели, кабельные сборки и микроволновые компоненты:

(Применяются в телекоммуникациях гражданского и специального назначения, в средствах связи, РЛС, а так же военной, авиационной и аэрокосмической отраслях промышленности).



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