19-1694; Rev 2; 12/07

EVALUATION KIT AVAILABLE

3V to 12V Current-Limiting Hot-Swap Controllers with Autoretry, DualSpeed/BiLevel Fault Protection

General Description

The MAX4271/MAX4272/MAX4273 comprise a complete family of integrated 3V to 12V hot-swap controllers. They allow the safe insertion and removal of circuit cards into live backplanes.

The discharged filter capacitors of the circuit card provide a low impedance to the live backplane. High inrush currents from the backplane to the circuit card can burn up connectors and components, or momentarily collapse the backplane power supply leading to a system reset. This family of hot-swap controllers prevents such problems by regulating the current to a preset limit when the board is plugged in, allowing the system to stabilize safely. After the startup cycle is completed, two on-chip comparators provide DualSpeed/BiLevel™ protection against short circuits, load glitches, and overcurrent conditions. In the event of a fault condition, the load is disconnected. Fault recovery is handled by unlatching (MAX4271), autoretry (MAX4272), or programmed (MAX4273) methods.

The MAX4271 family includes many integrated features that reduce component count and design time. An onboard charge pump provides the gate drive for a low-cost, external n-FET. Integrated features like startup current regulation and current glitch protection eliminate external timing resistors and capacitors. Also featured are an open-drain status output to indicate a fault condition, and an adjustable overcurrent response time.

The MAX4271 (latched fault protection) and MAX4272 (autoretry fault protection) come in 8-pin SO packages. The MAX4273 (full function) comes in the space-saving 16-pin QSOP package and 16-pin SO package. All parts are specified across the extended temperature range, and have an absolute maximum rating of 15V to provide extra protection against inductive kickback during board removal.

DualSpeed/BiLevel is a trademark of Maxim Integrated Products.

Features

- ♦ **Provide Safe Hot Swap for +3V to +12V Power Supplies with Few External Components**
- ♦ **Unique Current Regulation Architecture Minimizes n-FET Linear Mode Duration**
- ♦ **Autoretry Feature (MAX4272/MAX4273)**
- ♦ **DualSpeed/BiLevel Current Limit Protects Against Current Glitches and Short Circuits**
- ♦ **Power-On RESET (MAX4273)**
- ♦ **15V Absolute Maximum Rating Protects Against Inductive Kickbacks During Board Removal**
- ♦ **Internal Charge Pump Generates Gate Drive for External n-MOSFET**
- ♦ **Status Output Pin Indicates Fault/Safe Condition**
- ♦ **Space-Saving 8-Pin SO, 16-Pin QSOP Packages**

Applications

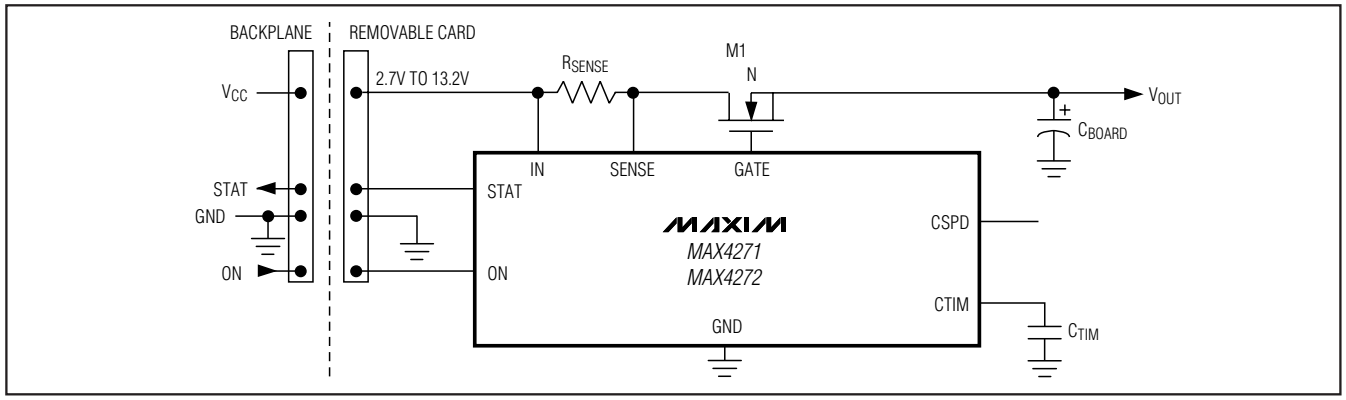
Base Stations
RAID
Remote-Access Servers
Network Routers and Switches
ISDN

Ordering Information

PART	TEMP RANGE	PIN-PACKAGE	PKG CODE
MAX4271ESA	-40°C to +85°C	8 SO	S8-1
MAX4272ESA	-40°C to +85°C	8 SO	S8-1
MAX4273EEE	-40°C to +85°C	16 QSOP	E16-1
MAX4273ESE	-40°C to +85°C	16 SO	E16-1

Pin Configurations appear at end of data sheet.

Typical Operating Circuit



For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

3V to 12V Current-Limiting Hot-Swap Controllers with Autoretry, DualSpeed/BiLevel Fault Protection

ABSOLUTE MAXIMUM RATINGS

IN to GND .....+15V
STAT, OUTC, LLMON, AUXVCC to GND .....-0.3V to +14V
GATE to GND .....-0.3V to (V_{IN} + 8.5V)
GATE to LLMON (Note 1).....-1V to +8.5V
INC, ON to GND (Note 2) .....-1V to +14V
CEXT to GND.....-8.5V to (V_{IN} + 0.3V)
CSPD, CTON,
REF to GND .....-0.3V to the lower of (V_{IN} + 0.3V) or +12V
VSENSE, RTH, CTIM to GND.....-0.3V to (V_{IN} + 0.3V)
Current into INC, ON (Note 2) .....±2mA
Current into Any Other Pin .....±50mA
Continuous Power Dissipation (T_A = +70°C)
8-Pin SO (derate 5.9mW/°C above +70°C).....471mW
16-Pin QSOP (derate 8.3mW/°C above +70°C).....667mW
16-Pin SO (derate 8.7mW/°C above +70°C).....696mW
Operating Temperature Range .....-40°C to +85°C
Storage Temperature Range.....-65°C to +150°C
Lead Temperature (soldering, 10s) .....+300°C

Note 1: GATE can be pulled below LLMON, but current must be limited to 2mA.
Note 2: INC and ON can be pulled below ground. Limiting the current to 2mA ensures that these pins are never lower than about -0.8V.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(V_{IN} = +2.7V to +13.2V, T_A = -40°C to +85°C, unless otherwise noted. Typical values are at V_{IN} = +5V and T_A = +25°C.) (Note 3)

PARAMETER	SYMBOL	CONDITIONS			MIN	TYP	MAX	UNITS
POWER SUPPLIES								
Input Voltage Range	V_{IN}				2.7		13.2	V
Supply Current	I_Q	V_{ON} = V_{IN}				0.6	1	mA
CURRENT CONTROL								
Slow Comparator Threshold	$V_{SC,TH}$	V_{IN} - V_{SENSE}		T_A = +25°C	45	50	55	mV
				T_A = T_{MIN} to T_{MAX}	43.5		56	
Slow Comparator Response Time	t_{CSPD}	C_{SPD} = floating			10	20	40	µs
		C_{SPD} = 100nF to GND			10	20	40	ms
Fast Comparator Threshold	$V_{FC,TH}$	MAX4273 only	5kΩ on RTH to V_{IN}	T_A = +25°C	45	50	55	mV
				T_A = T_{MIN} to T_{MAX}	43.5		56	
			75kΩ on RTH to V_{IN}	T_A = +25°C	675	750	825	
				T_A = T_{MIN} to T_{MAX}	650		840	
			RTH = GND		180	200	220	
		MAX4271/MAX4272						
Fast Comparator Response Time	t_{FCD}	10mV overdrive, from overload condition to gate discharging				350		ns
SENSE Input Bias Current	$I_{B,SENSE}$	V_{SENSE} = V_{IN}				0.2	10	µA
MOSFET DRIVER								
Startup Period (Notes 4, 5)	t_{START}	MAX4271/MAX4272 C_{TIM} = 100nF			21	31	41	ms
		MAX4273 C_{TON} = 100nF			21	31	41	
		No capacitor				5.5		µs
Gate Charge Current	I_{GATE}	GATE = IN (Note 6)				100		µA

ELECTRICAL CHARACTERISTICS (continued)

(V_{IN} = +2.7V to +13.2V, T_A = -40°C to +85°C, unless otherwise noted. Typical values are at V_{IN} = +5V and T_A = +25°C.) (Note 3)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Slow Turn-Off Time	$t_{SLOW,OFF}$	Time from trigger to V_{GATE} < 0.1V (triggered by either the ON input or the slow comparator), C_{GATE} = 1000pF to GND			60		µs
Fast Turn-Off Time	$t_{FAST,OFF}$	Time from current overload to V_{GATE} < 0.1V (triggered by the fast comparator, during normal operation), C_{GATE} = 1000pF to GND			15		µs
Maximum Gate Protection Voltage		Voltage at which internal zener clamp circuitry is triggered, measured with respect to V_{IN} (MAX4271/MAX4272), measured with respect to V_{LLMON} (MAX4273)			6.7	7.5	V
Minimum Gate Drive Voltage		Measured with respect to V_{IN}, I_{GATE} = 8.5µA	$V_{IN} \geq 5V$	5			V
			$V_{IN} \geq 2.7V$	2.7			
Gate Discharge Current	$I_{GATE,DIS}$	During fast discharge, due to a fast comparator fault in normal operation		0.4	1	2.75	mA
		During startup (current regulation provided by fast comparator)		25	70	195	µA
		During normal discharge, due to a slow comparator fault in normal mode, or by ON going low		75	200	550	
LLMON Overvoltage Threshold		Startup is initiated only after V_{LLMON} is less than this voltage (MAX4273)		0.1			V
GATE Overvoltage Threshold		Startup is initiated only after V_{GATE} is less than this voltage		0.1		0.6	V
LLMON Impedance		Impedance to GND, after a fault (MAX4273)			1		kΩ
REFERENCE (MAX4273)							
Output Voltage	V_{REF}	No load, V_{IN} = 5V		1.164	1.2	1.236	V
Line Regulation	$\Delta V_{REF, LINE}$	2.7V ≤ V_{IN} ≤ 13.2V, no load			1	8	mV
Load Regulation	$\Delta V_{REF, LOAD}$	I_{REF} = 0 to 100µA, V_{IN} = 5V			0.6	3	mV
ON AND RESET COMPARATORS							
Threshold Voltage		V_{IN} = 5V, rising threshold at ON or INC		0.575	0.6	0.625	V
Hysteresis	V_{HYST}				3		mV
Power-Supply Rejection Ratio	PSRR	2.7V ≤ V_{IN} ≤ 13.2V				±1	mV/V
Propagation Delay	$t_{D,COMP}$	10mV overdrive, ON going positive or negative, INC going negative			10		µs
		INC going positive		100	150	200	ms

ELECTRICAL CHARACTERISTICS (continued)

(V_{IN} = +2.7V to +13.2V, T_A = -40°C to +85°C, unless otherwise noted. Typical values are at V_{IN} = +5V and T_A = +25°C.) (Note 3)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Input Voltage Range		Input can be driven to the absolute maximum limit without false output inversion	-0.1		13.2	V
Input Bias Current	$I_{B,COMP}$			0.001	1	µA
ON Pulse Width Low (Note 7)	$t_{RESTART}$	To unlatch a fault MAX4271, MAX4273 with C_{TIM} = IN	20			µs
DIGITAL OUTPUTS (STAT, OUTC)						
Output Leakage Current		$V_{STAT} \le 13.2$, $V_{OUTC} \le 13.2$			1	µA
Output Voltage Low	V_{OL}	I_{SINK} = 1mA			0.4	V
RETRY TIMING (MAX4272, MAX4273)						
Retry Timeout Period	t_{RETRY}	100nF capacitor on C_{TIM} (Note 5)	0.5	1	2	s
Default Retry Timeout Period	t_{RETRY} (default)	C_{TIM} = no connection		176		µs
UNDERVOLTAGE LOCKOUT (UVLO)						
Threshold	V_{UVLO}	Startup is initiated when this threshold is exceeded at IN	2.25		2.67	V
Hysteresis	$V_{UVLO,HYST}$			100		mV
Delay	$t_{D,UVLO}$	Time the input voltage must exceed undervoltage lockout before startup is initiated	100	150	200	ms

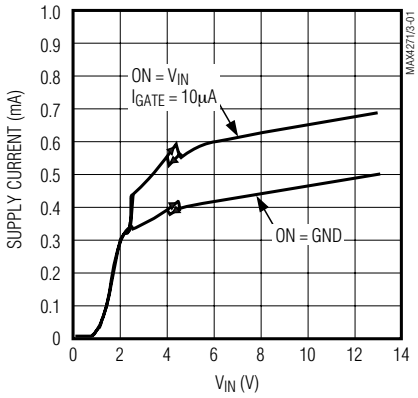
Note 3: All devices are 100% tested at T_A = +25°C. All temperature limits are guaranteed by design.
Note 4: Startup period is the time during which the slow comparator is ignored and the fast comparator regulates the sense current. It is measured from the time ON is brought high.
Note 5: Inferred from test with C_{TON} = 10nF (MAX4273) and C_{TIM} = 1nF.
Note 6: The current available at GATE is a function of V_{GATE} (see *Typical Operating Characteristics*).
Note 7: Guaranteed by design.

3V to 12V Current-Limiting Hot-Swap Controllers with Autoretry, DualSpeed/BiLevel Fault Protection

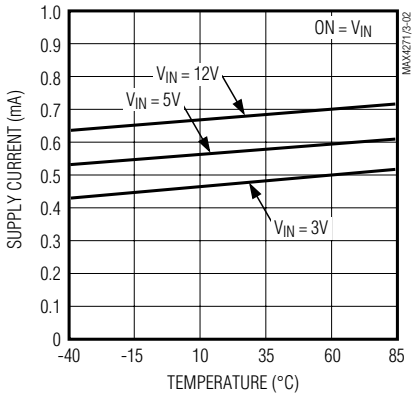
Typical Operating Characteristics

(V_{IN} = 5V, R_{SENSE} = 100mΩ, C_{BOARD} = 47µF, T_A = +25°C, unless otherwise noted.)

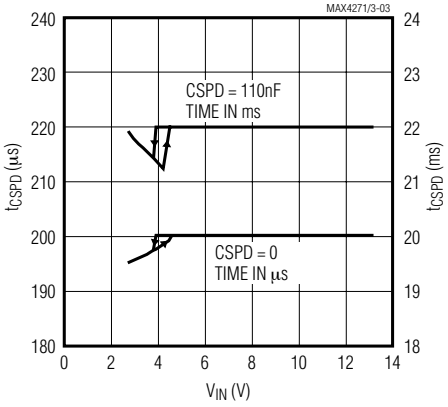
SUPPLY CURRENT vs. INPUT VOLTAGE



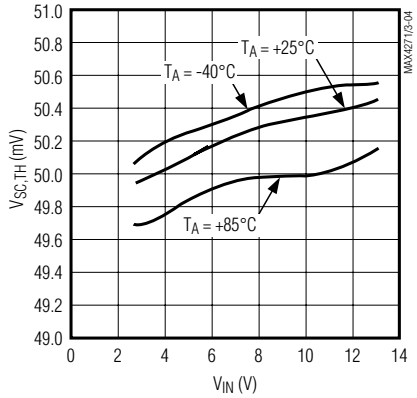
SUPPLY CURRENT vs. TEMPERATURE



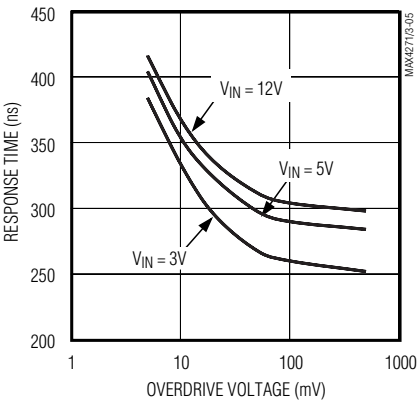
SLOW COMPARATOR RESPONSE TIME vs. INPUT VOLTAGE



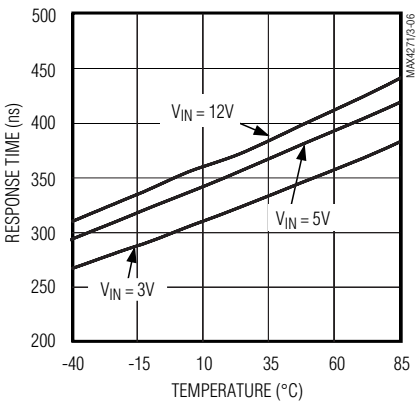
SLOW COMPARATOR THRESHOLD vs. INPUT VOLTAGE



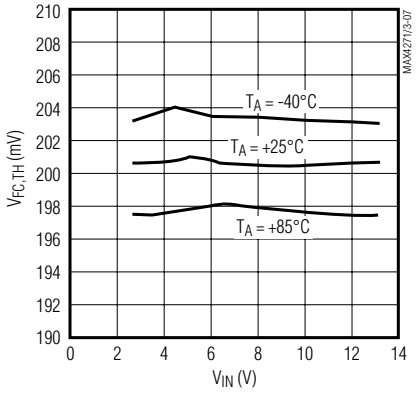
FAST COMPARATOR RESPONSE TIME vs. OVERDRIVE VOLTAGE



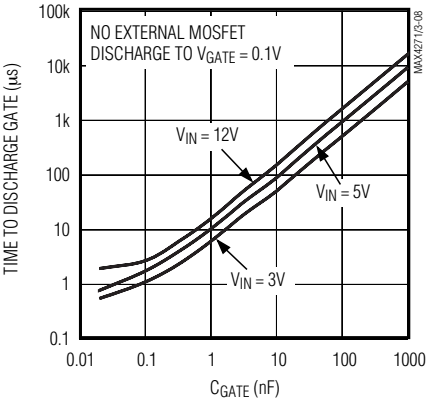
FAST COMPARATOR RESPONSE TIME vs. TEMPERATURE



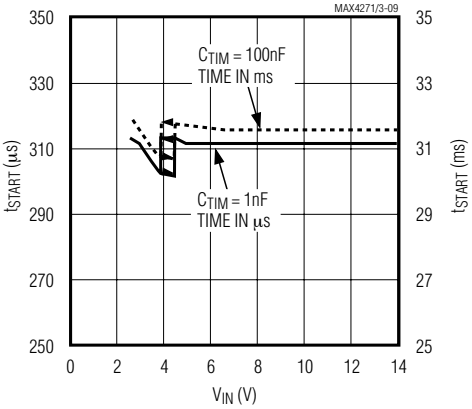
FAST COMPARATOR THRESHOLD vs. INPUT VOLTAGE



TIME TO FAST DISCHARGE GATE vs. C_{GATE}



STARTUP PERIOD vs. INPUT VOLTAGE

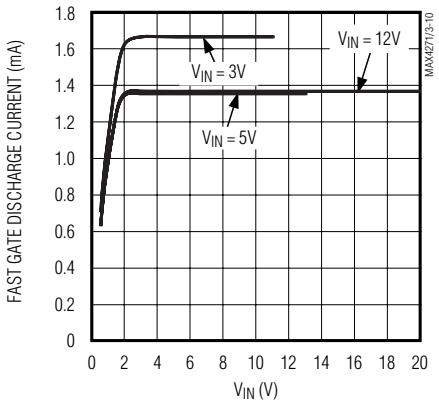


3V to 12V Current-Limiting Hot-Swap Controllers with Autoretry, DualSpeed/BiLevel Fault Protection

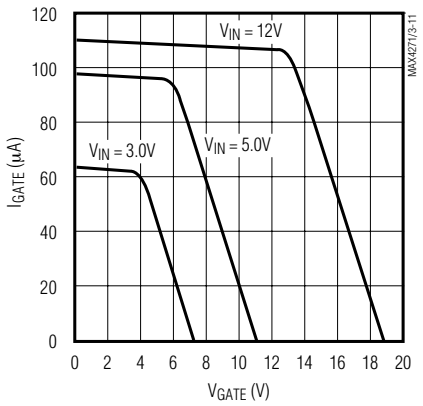
Typical Operating Characteristics (continued)

(V_{IN} = 5V, R_{SENSE} = 100mΩ, C_{BOARD} = 47µF, T_A = +25°C, unless otherwise noted.)

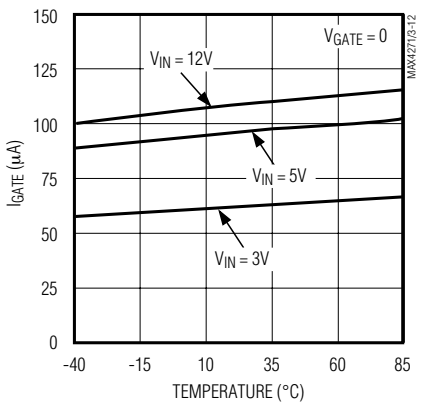
FAST GATE DISCHARGE CURRENT vs. V_{GATE}



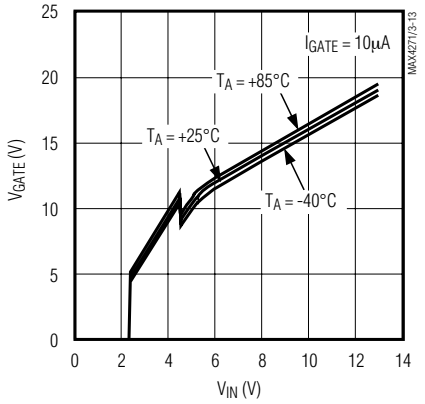
GATE CHARGE CURRENT vs. GATE VOLTAGE



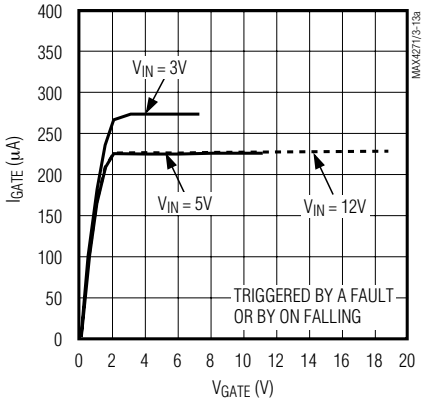
GATE CHARGE CURRENT vs. TEMPERATURE



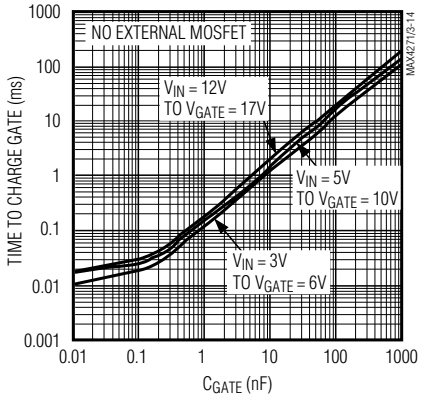
GATE VOLTAGE vs. INPUT VOLTAGE



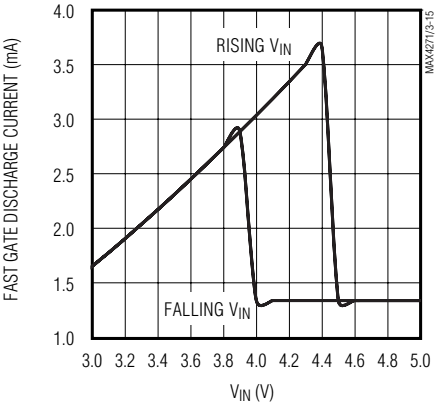
SLOW GATE DISCHARGE CURRENT vs. GATE VOLTAGE



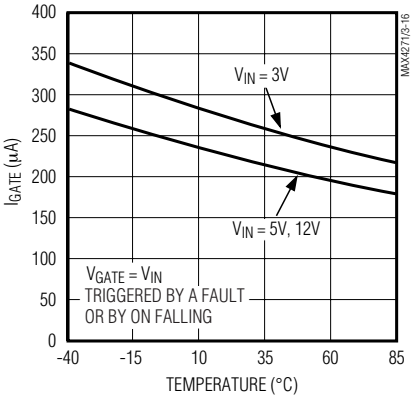
TIME TO CHARGE GATE vs. C_{GATE}



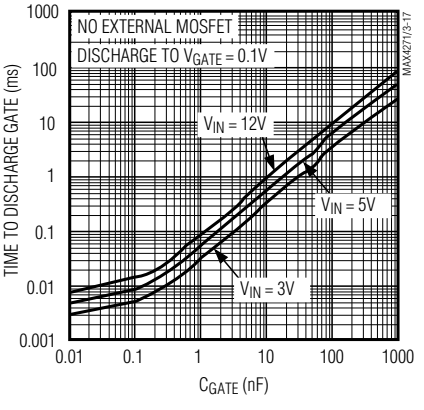
FAST GATE DISCHARGE CURRENT vs. V_{IN}



SLOW GATE DISCHARGE CURRENT vs. TEMPERATURE



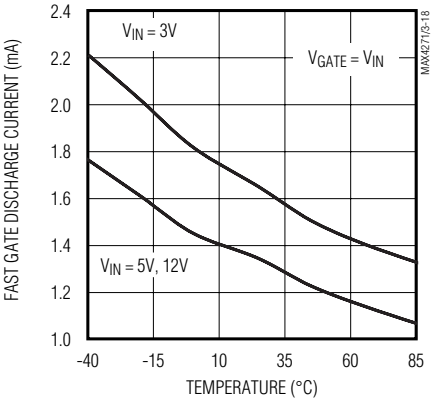
TIME TO SLOW DISCHARGE GATE vs. C_{GATE}



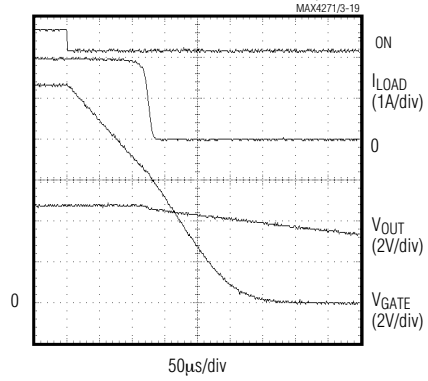
Typical Operating Characteristics (continued)

(V_{IN} = 5V, R_{SENSE} = 100mΩ, C_{BOARD} = 47µF, T_A = +25°C, unless otherwise noted.)

FAST GATE DISCHARGE CURRENT vs. TEMPERATURE

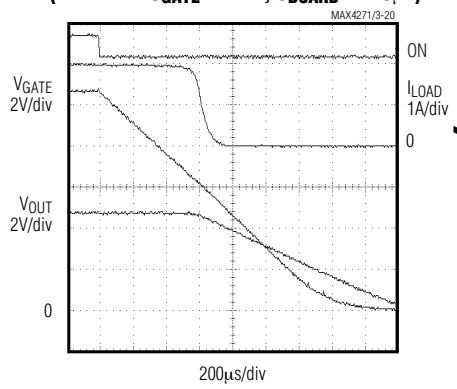


MAX4271/MAX4272 TURN-OFF TIME (C_{BOARD} = 470µF)



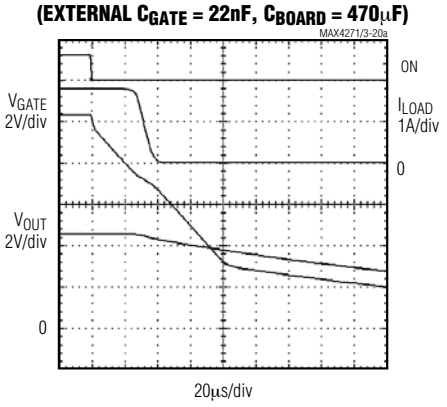
C_{BOARD} = 470µF, R_{SENSE} = 100mΩ, C_{GATE} = 0

MAX4271/MAX4272 TURN-OFF TIME (EXTERNAL C_{GATE} = 22nF, C_{BOARD} = 470µF)

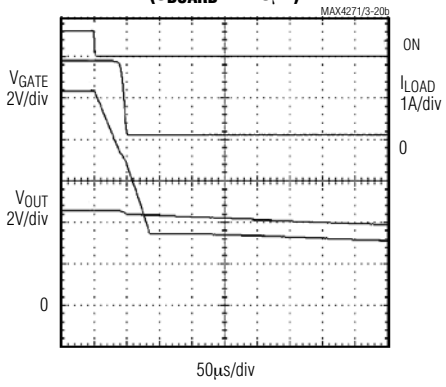


C_{BOARD} = 470µF, R_{SENSE} = 100mΩ, C_{GATE} = 22nF, R_S = 0

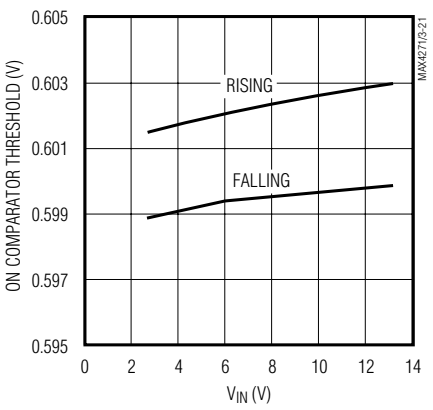
MAX4273 TURN-OFF TIME (EXTERNAL C_{GATE} = 22nF, C_{BOARD} = 470µF)



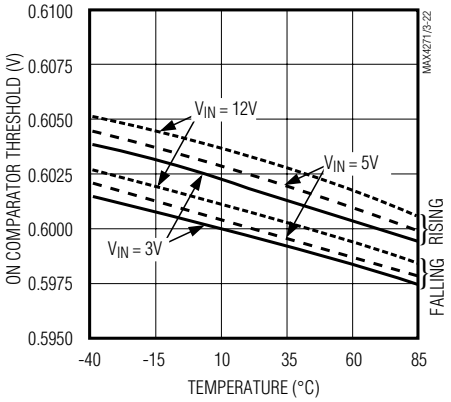
MAX4273 TURN-OFF TIME (C_{BOARD} = 470µF)



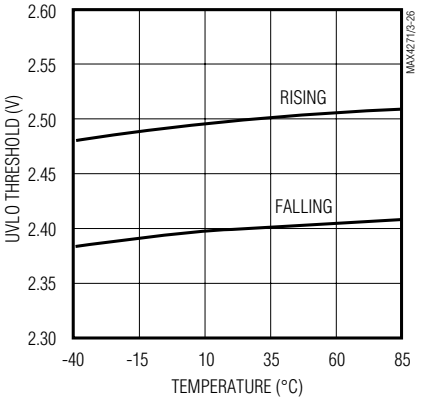
ON COMPARATOR THRESHOLD vs. INPUT VOLTAGE



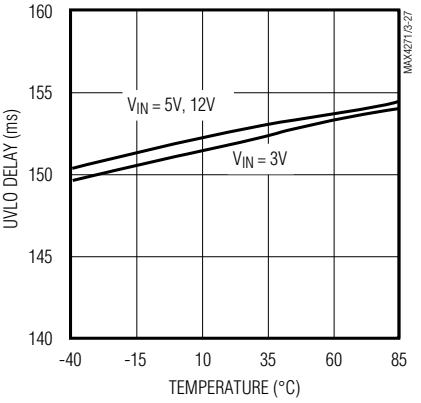
ON-COMPARATOR THRESHOLD vs. TEMPERATURE



UVLO THRESHOLD VOLTAGE vs. TEMPERATURE



UVLO DELAY vs. TEMPERATURE

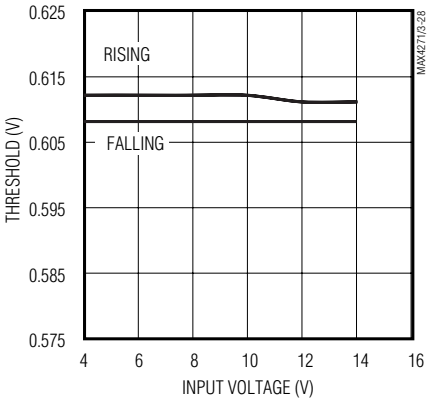


3V to 12V Current-Limiting Hot-Swap Controllers with Autoretry, DualSpeed/BiLevel Fault Protection

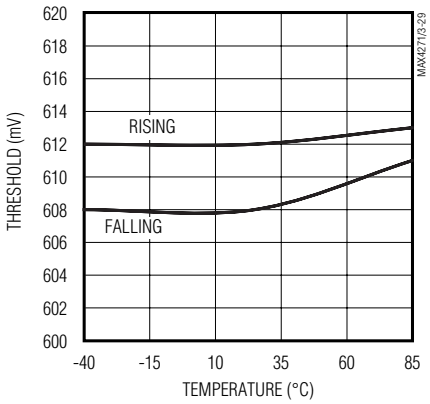
Typical Operating Characteristics (continued)

(V_{IN} = 5V, R_{SENSE} = 100mΩ, C_{BOARD} = 47µF, T_A = +25°C, unless otherwise noted.)

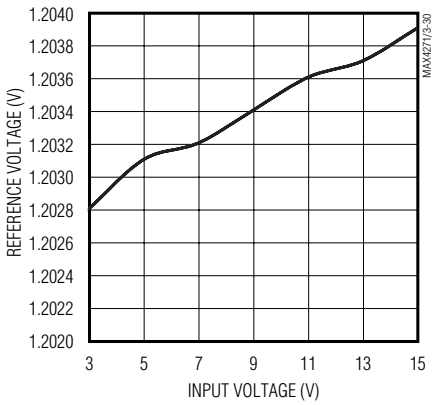
INC THRESHOLD vs. INPUT VOLTAGE



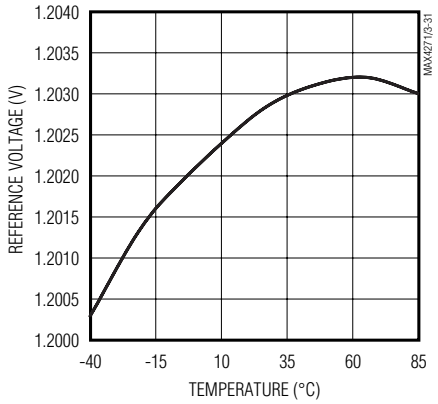
INC THRESHOLD vs. TEMPERATURE



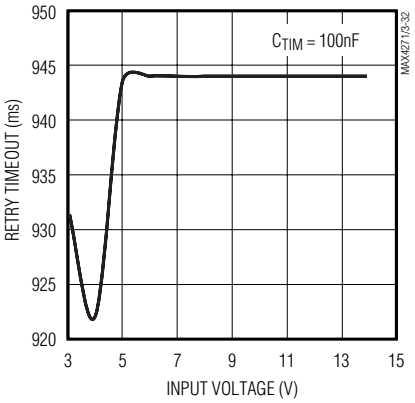
MAX4273 REFERENCE VOLTAGE vs. INPUT VOLTAGE



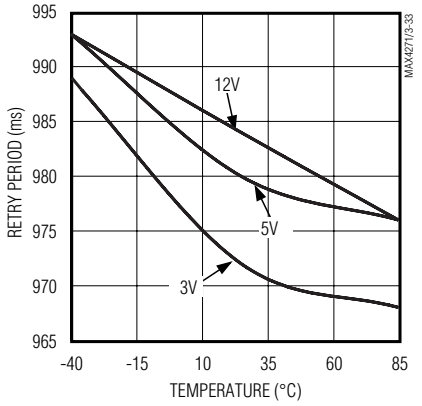
REFERENCE VOLTAGE vs. TEMPERATURE



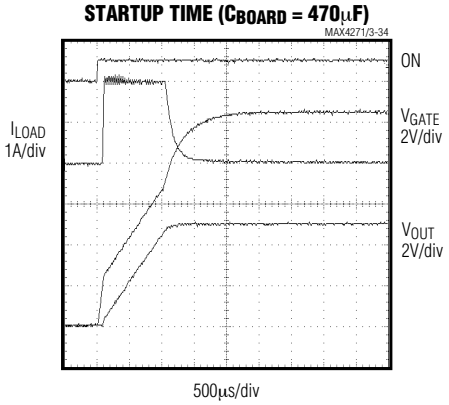
RETRY TIMEOUT PERIOD vs. INPUT VOLTAGE



RETRY TIMEOUT PERIOD vs. TEMPERATURE



STARTUP TIME (C_{BOARD} = 470µF)

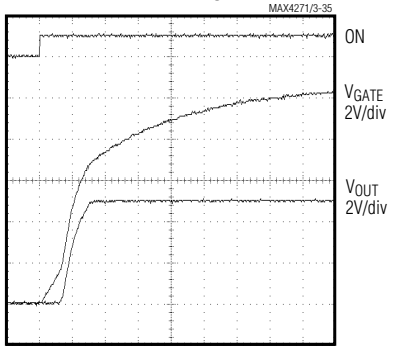


C_{BOARD} = 470µF, R_{SENSE} = 100mΩ,
C_{TIM} = 10nF, C_{GATE} = 0

Typical Operating Characteristics (continued)

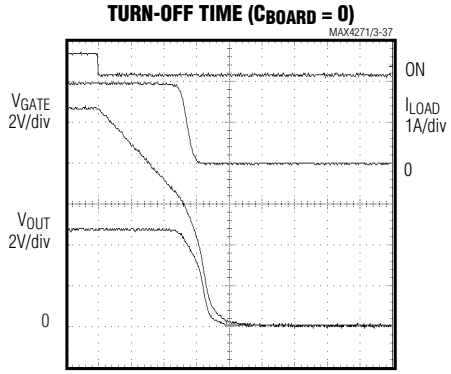
(V_{IN} = 5V, R_{SENSE} = 100mΩ, C_{BOARD} = 47µF, T_A = +25°C, unless otherwise noted.)

STARTUP TIME (C_{BOARD} = 0)



C_{BOARD} = 0, R_{SENSE} = 100mΩ,
C_{TIM} = 10nF, C_{GATE} = 0

TURN-OFF TIME (C_{BOARD} = 0)



C_{BOARD} = 0, R_{SENSE} = 100mΩ,
C_{GATE} = 0, R_S = 0

Pin Description

PIN			FUNCTION
MAX4271 MAX4272	MAX4273	NAME	
—	1	RTH	Current-Sense Threshold Setting Input. Connect a resistor from RTH to V_{IN} to set the fast comparator threshold. Bypass to V_{IN} with 0.1µF. Connect RTH to V_{IN} to disable both the fast comparator and current regulation at startup. Short RTH to GND for a 200mV threshold. See *Fast Comparator Threshold* section.
—	2	AUXVCC	Auxiliary Supply Input. Supply input for short-circuit switchover. To use this, connect a 1µF capacitor from AUXVCC to GND; otherwise, leave floating. See *Auxiliary V_{CC}* section.
1	3	IN	Input Voltage. Connect to +2.7V to +13.2V supply; 15V absolute maximum voltage rating.
2	4	SENSE	Current-Sense Resistor Voltage Input. R_{SENSE} is connected from IN to SENSE.
3	5	GATE	Gate Drive Output. Connect to gate of external N-channel pass transistor.
—	6	CEXT	External Gate Capacitance Connection. Connect a capacitor from CEXT to GATE to increase the gate charging time. This pin goes high impedance during a fast comparator fault for fast discharge.
—	7	LLMON	Load Line Monitor. Connect to the source of the external N-channel MOSFET. The external FET is turned on only when the load voltage is less than 100mV.
4	8	GND	Ground
—	9	CTON	Startup Timer Input. Leave floating or connect the timing capacitor from CTON to GND. See *Startup and Retry Timers* section.
5	10	CSPD	Slow Comparator Speed Setting. Leave floating or connect the timing capacitor from CSPD to GND. See *Slow Comparator Response Time* section.

3V to 12V Current-Limiting Hot-Swap Controllers with Autoretry, DualSpeed/BiLevel Fault Protection

Pin Description (continued)

PIN			FUNCTION
MAX4271 MAX4272	MAX4273	NAME	
6	11	CTIM	Startup and Retry Timers Input. Controls the startup time and the autoretry time (32x startup time) in the MAX4272, only the autoretry time in the MAX4273, and only the startup time in the MAX4271. Leave floating or connect the timing capacitor from CTIM to GND. Connect to IN for latched mode (this prevents autoretry in MAX4273). See *Startup and Retry Timers* section.
7	12	STAT	Status Output. High indicates startup completed with no fault (Table 1). STAT is an open-drain output.
8	13	ON	ON Comparator Input. Connect high for normal operation; connect low to force the MOSFET off. Comparator threshold $V_{TH,ON}$ = 0.6V allows for precise control over shutdown feature. Pulse ON low for 20µs min to unlatch after a fault (MAX4273 in latched mode, or MAX4271). Negative pulses are ignored during autoretry (MAX4273 in autoretry mode, or MAX4272). See *ON and Reset Comparators* section.
—	14	REF	1.2V Reference Output. **Do not bypass with a capacitor to GND.**
—	15	INC	Uncommitted Comparator Input. Controls OUTC.
—	16	OUTC	Uncommitted Comparator Output. OUTC goes high 150ms after INC goes high. OUTC goes low immediately after INC goes low. OUTC is an open-drain output.

Detailed Description

The MAX4271/MAX4272/MAX4273 are circuit breaker ICs designed for hot-swap applications where a line card is inserted into a live backplane. Normally, when a line card is plugged into a live backplane, the card's discharged filter capacitors provide a low impedance that can momentarily cause the main power supply to collapse. The MAX4271/MAX4272/MAX4273 are designed to reside either in the backplane or in the removable card to provide inrush current limiting and short-circuit protection. This is achieved by using an external N-channel MOSFET, an external current-sense resistor, and two on-chip comparators. Figures 1 and 2 show the functional diagrams.

The timing and voltage levels for several critical parameters can be adjusted with external resistors, external capacitors, or by pin strapping. The timing components are optional; without them, the part is set to its nominal values, as shown in the *Electrical Characteristics*. The parameters that can be adjusted are:

- Current-limit threshold
- Slow comparator response time
- Startup timer
- Fast comparator threshold
- Autoretry timeout period (time the part is shut down after an overcurrent event)
- Fault management (latched/autoretry)
- Current overload threshold

Startup Mode

CTIM (MAX4271/MAX4272) or CTON (MAX4273) sets the startup period (see *Startup and Retry Timers*). The startup period begins after three conditions are met:

1) 150ms after V_{IN} exceeds the UVLO threshold (see *Over/Undervoltage Lockouts*)

AND

2) 10µs after V_{ON} >0.6V

AND

3) The device is no longer in retry mode.

During startup, the slow comparator is disabled and the inrush current can be limited in two ways:

1) Slow ramping of the current to the load by controlling the external MOSFET gate voltage

OR

2) Limiting the current to the load by regulating the voltage across the external current-sense resistor

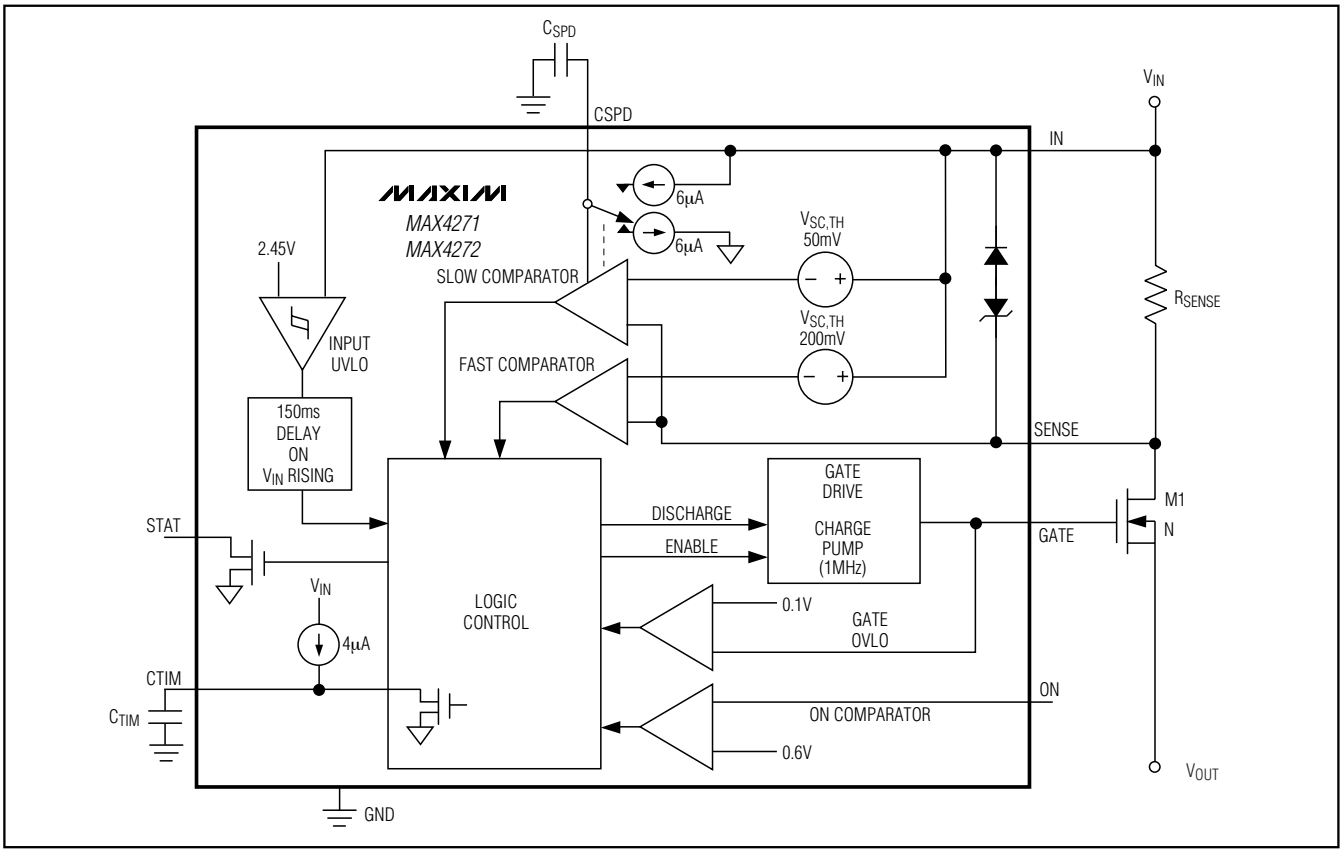


Figure 1. MAX4271/MAX4272 Functional Diagram

Unlike other circuit breaker ICs, the MAX4271/MAX4272/MAX4273 regulate the current to a preset level instead of completely turning off the external N-FET if an overcurrent condition occurs during startup.

In startup mode, the gate drive current is limited to 100µA and decreases with the increase of the gate voltage (see Gate Charge Current vs. Gate Voltage in the *Typical Operating Characteristics*). This allows the controller to slowly enhance the MOSFET. If the fast comparator detects an overcurrent, the gate voltage is momentarily discharged with a fixed 70µA current until the load current through the sense resistor (R_{SENSE}) decreases below its threshold point. This effectively regulates the inrush current during startup. Figure 3 shows the startup waveforms. STAT goes high at the end of the startup period if no fault condition is present.

Normal Operation (DualSpeed/BiLevel)

In normal operation (after the startup timer has expired), protection is provided by turning off the external MOSFET when a fault condition is encountered. DualSpeed/BiLevel fault protection incorporates two comparators with different thresholds and response times to monitor the current:

1) **Slow comparator**. This comparator has an externally set response time (from 20µs to seconds) and a fixed 50mV threshold voltage. The slow comparator ignores low-amplitude momentary current glitches. After an extended overcurrent condition, a fault is generated and the MOSFET gate is slowly discharged.

2) **Fast comparator**. This comparator has a fixed 350ns response time and a 200mV threshold voltage (adjustable from 50mV to 750mV in MAX4273). The fast comparator turns off the MOSFET immediately after it detects a large amplitude event such as a short circuit.

3V to 12V Current-Limiting Hot-Swap Controllers with Autoretry, DualSpeed/BiLevel Fault Protection

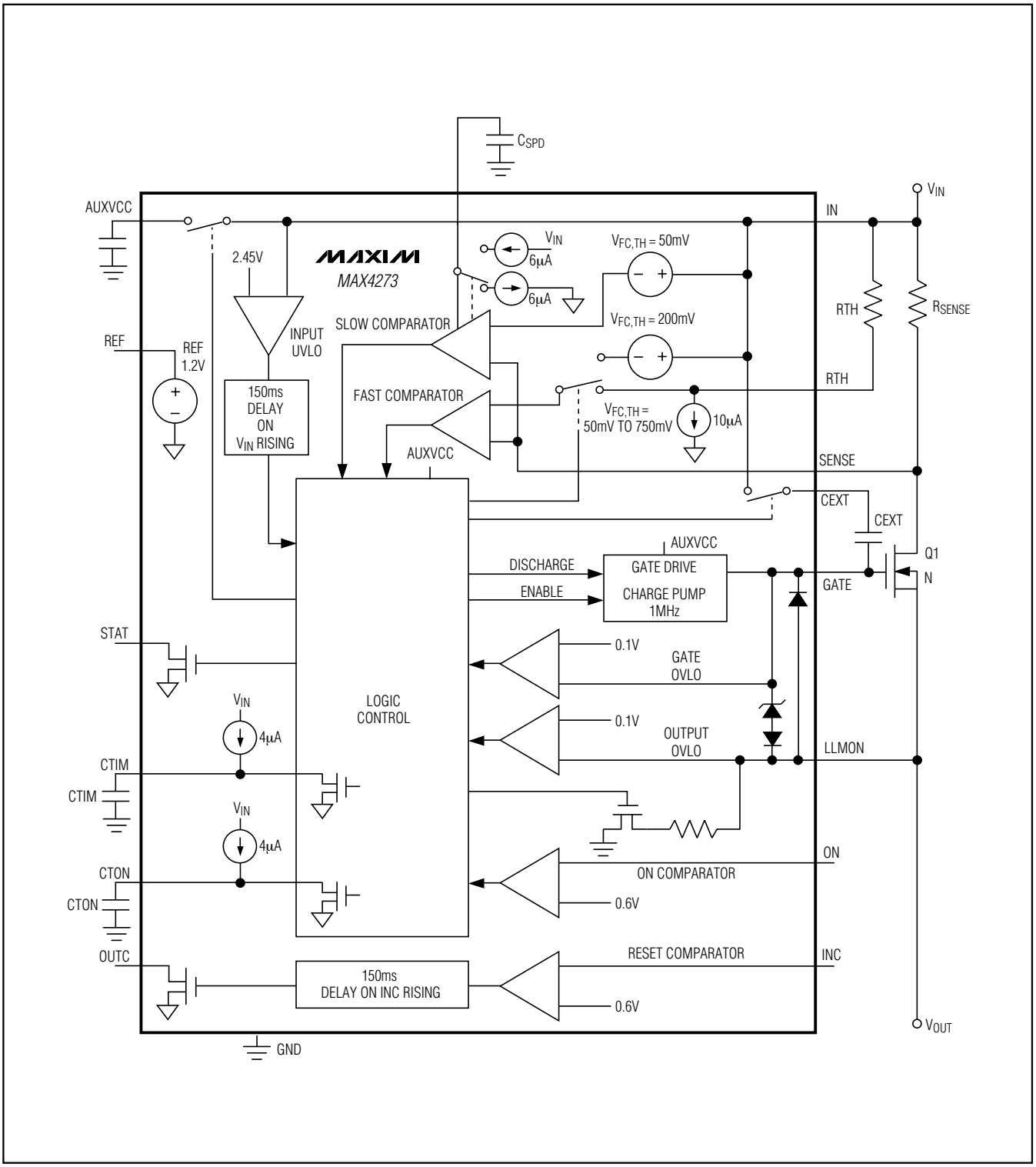


Figure 2. MAX4273 Functional Diagram

In each case, when a fault is encountered, the status pin (STAT) goes low, and for the MAX4273, the device discharges the output voltage through a 1kΩ resistor from LLMON to GND. After a fault, the MAX4271 stays latched off and the MAX4272 enters retry mode, while the MAX4273 has selectable latched or retry mode. Figure 4 shows the waveforms of a fault condition.

BiLevel Fault Protection

Slow Comparator

The slow comparator is disabled at startup while the external MOSFET is turning on. This allows the part to ignore the higher-than-normal inrush current charging the board capacitors when a card is first plugged in.

If the slow comparator detects an overload condition while in normal operation (after startup is completed), it turns off the external MOSFET by discharging the gate capacitance with a 200µA current. The slow comparator threshold is set at 50mV and has a default delay of 20µs (CSPD floating), allowing it to ignore power-supply glitches and noise. The response time can be lengthened with an external capacitor at CSPD (Figure 10).

If the overcurrent condition is not continuous, then the duration above the threshold minus the duration below must be greater than 20µs (or the external programmed value) for the device to trip. When the current is above the threshold, CSPD is charged with a 6µA current source. A fault is detected when CSPD is charged to the trip point of 1.2V. Therefore, a pulsing current with a duty cycle of 50% or greater (i.e., the current is above the threshold level > 50% of the time) is considered a fault condition even if it is never higher than the threshold for longer than the slow comparator's set response time.

The discharge rate depends on the N-FET gate capacitance and the external capacitance at GATE. In the MAX4273, CEXT remains connected and capacitance to this point has to be discharged by the same current. This increases the discharge time. Once the fault condition is detected, the STAT pin goes low and the device goes into retry or latched mode.

Fast Comparator

The fast comparator behaves differently according to the operating mode.

During startup, the fast comparator is part of a simple current regulator. When the sensed current is above the fast comparator threshold, the gate is discharged with a 70µA current source. When the sensed current drops below the threshold, the charge pump turns on again. The sensed current will rise and fall near the threshold due to the fast comparator and charge-pump propagation delay. The gate voltage will be roughly sawtooth shaped, and the load current will present a 20% ripple. The ripple can be reduced by adding a capacitor from GATE to GND.

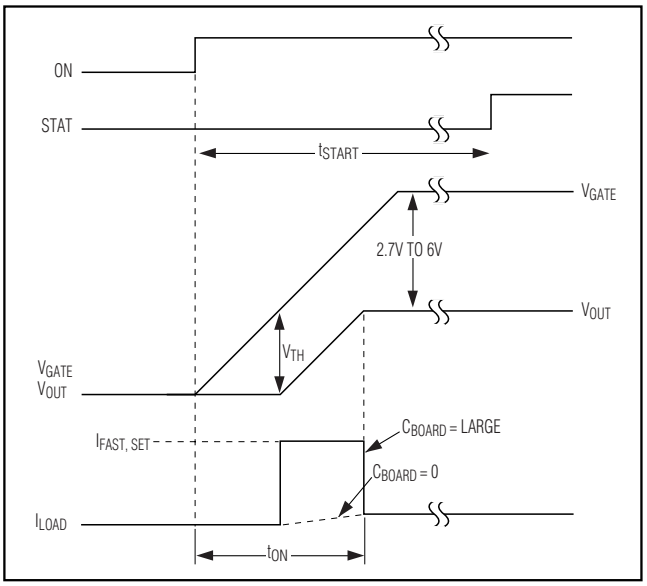


Figure 3. Startup Waveforms

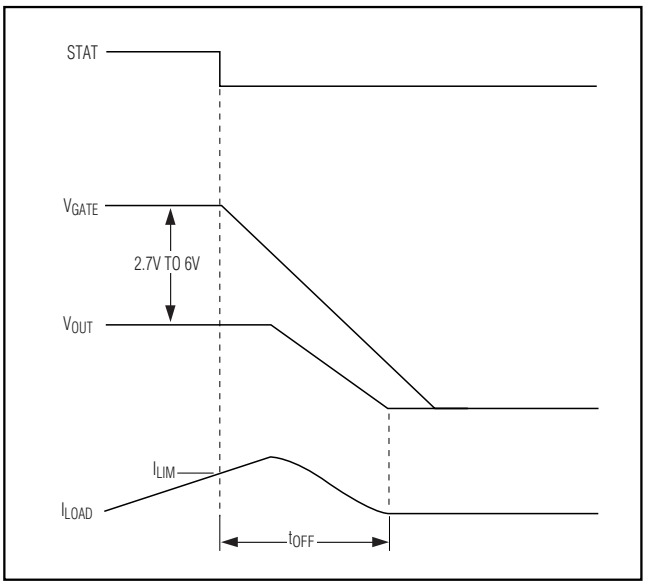


Figure 4. Response to a Fault Condition

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If the sensed current is still high after the startup timer expires, the MOSFET gate is discharged completely as described below.

In normal operation (after startup), the fast comparator is used as an emergency off switch. If the load current reaches the fast comparator threshold, the device quickly forces the MOSFET off completely. This could happen in the event of a serious current overload or a dead short. The fast comparator has a 350ns response time and discharges GATE with a 1mA current. Given a 1000pF gate capacitance and 12V gate voltage, the MOSFET will be off in about 12µs. Any additional capacitance connected between GATE and GND to slow down the startup time also increases the turn-off time.

In the MAX4273, CEXT goes high impedance during the fast discharge. This reduces the effective capacitance on GATE if a capacitor is used between GATE and CEXT, and allows the MOSFET to quickly turn off. In turn, this allows adjustment of the MOSFET charging time without affecting the fast discharge rate, although it does affect the slow discharge rate.

The MAX4271/MAX4272 fast comparator threshold is set to four times the slow comparator threshold (i.e., 200mV). The MAX4273 fast comparator threshold is set to 200mV by connecting RTH to GND, is disabled by connecting RTH to IN, or is adjustable by an external resistor connected to IN (see *Fast Comparator Threshold (RTH)*).

Latched/Autoretry

The MAX4271 MOSFET driver stays latched off after a fault condition until it is reset by a negative-going pulse on the ON pin. The MAX4272 is periodically turned on after a fault condition with a timeout duration set by an external timing capacitor on CTIM. The MAX4273 has a selectable latched mode or retry mode. Connect CTIM to IN to set the device in latched mode, or use an external capacitor at CTIM to set the retry timeout.

Pulse ON low for 20µs (min) to restart after a fault (MAX4271/MAX4273 in latched mode). Negative pulses are ignored during autoretry (MAX4273 in autoretry mode, or MAX4272).

The capacitor on CTIM affects the MAX4272's retry timeout period (time the part is shut down after an overcurrent event) and the startup time (see the *Electrical Characteristics*). The retry timeout period is fixed at 32 times the startup time in order to minimize power dissipation in the external MOSFET in case of a short-circuit condition (see *MOSFET Thermal Considerations*). This is not an issue for parts latched off during a fault condition since they stay off until commanded on. The MAX4273 configured in retry mode has a separate startup timer capacitor (CTON) and retry timeout capacitor (CTIM). This allows the user to change the ratio between startup time and retry timeout period.

Status Output

The status output is an open-drain output that goes low under the following conditions:

- During the UVLO delay period
- In startup
- Forced off (ON <0.6V)
- In an overcurrent condition
- In the retry timeout period (or latched off, for the latched parts)

STAT is high only if the part is in normal mode and no faults are present (Table 1). Figure 5 shows the status (STAT) output timing diagram.

Over/Undervoltage Lockouts

The UVLO prevents the MAX4271/MAX4272/MAX4273 from turning on the external MOSFET until V_{IN} exceeds the lockout threshold (2.25V min) for 150ms. The UVLO protects the external MOSFET from insufficient gate drive voltage. The 150ms timeout ensures that the board is fully plugged into the backplane and that V_{IN}

Table 1. Status Output Truth Table

IN UVLO DELAY PERIOD	PART IN STARTUP	ON PIN	OVERCURRENT CONDITION	PART IN RETRY-TIMEOUT PERIOD (OR LATCHED OFF DUE TO OVERCURRENT CONDITION)	STAT PIN (STATUS)
Yes	X	X	X	X	Low
X	Yes	X	X	X	Low
X	X	Low	X	X	Low
X	X	X	Yes	X	Low
X	X	X	X	Yes	Low
No	No	High	No	No	High

is stable. Any input voltage transient at IN below the UVLO threshold will reset the device and initiate a start-up sequence.

These devices also have an overvoltage lockout (OVLO) feature that prevents the device from restarting after a fault condition if the discharge has not been completed. V_{GATE} has to be discharged to below 0.1V. Additionally, the MAX4273 LLMON pin discharges the load line with a 1kΩ pulldown and prevents startup until the load voltage is below 0.1V.

Since the MAX4271/MAX4272 do not monitor the output voltage, a startup sequence can be initiated while the board capacitance is still charged.

A large board capacitance or a short startup period may prevent the MAX4272 from charging completely in one startup period. The MAX4272 responds to these conditions by charging the capacitor with bursts defined by a t_{ON} duty cycle and a period of $t_{ON} + t_{RETRY}$. The charging will be complete after several retries unless the resistive load or current load excessively discharges the board capacitance during the retry timeout. This feature applies to the MAX4273 if LLMON is left floating or is connected to GND. To prevent multiple charging bursts, ensure that the t_{ON} timer exceeds the minimum time required to complete the charge of the board capacitance (see *Component Selection*).

Gate Overvoltage Protection

New-generation MOSFETs have an absolute maximum rating of ±8V for the gate-to-source voltage (V_{GS}). To protect these MOSFETs, the MAX4271/MAX4272 limit the gate-to-drain voltage (the MAX4273 limits the gate-to-source voltage) to +7.5V with an internal zener diode. No protection is provided for negative V_{GS} (MAX4271/MAX4272). If GATE can be discharged to ground faster than the output voltage, an external small-signal protection diode (D1) can be used, as shown in Figure 6. The MAX4273 has the protection diode internal.

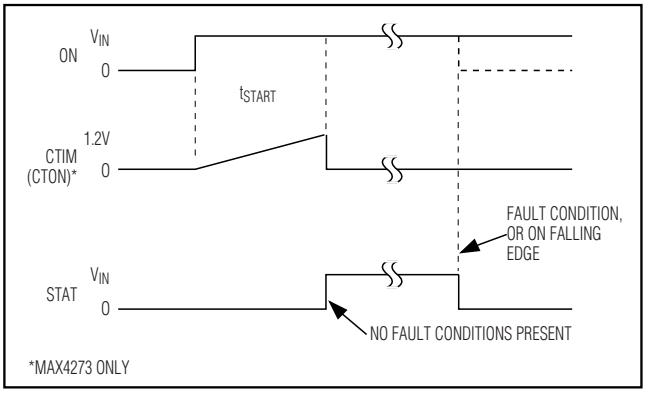

Figure 5. Status (STAT) Output Timing Diagram

Applications Information

Component Selection

N-Channel MOSFET

Select the external N-channel MOSFET according to the application's current level. Table 2 lists some recommended components. The MOSFET's on-resistance ($R_{DS(ON)}$) should be chosen low enough to have a minimum voltage drop at full load to limit the MOSFET power dissipation. High $R_{DS(ON)}$ can cause output ripple if the board has pulsing loads or triggers an external undervoltage reset monitor at full load. Determine the device power-rating requirement to accommodate a short circuit on the board at startup with the device

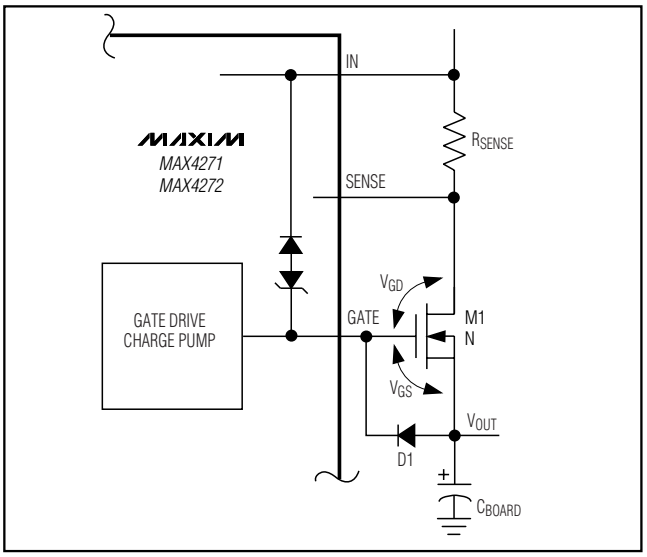

Figure 6. External Gate-Source Protection

Table 2. Recommended N-Channel MOSFETs

PART NUMBER	MANUFACTURER	DESCRIPTION
IRF7413	International Rectifier	11mΩ, 8 SO, 30V
IRF7401		22mΩ, 8 SO, 20V
IRL3502S		6mΩ, D2PAK, 20V
MMSF3300	Motorola	20mΩ, 8 SO, 30V
MMSF5N02H		30mΩ, 8 SO, 20V
MTB60N05H		14mΩ, D2PAK, 50V
FDS6670A	Fairchild	10mΩ, 8 SO, 30V
NDS8426A		13.5mΩ, 8 SO, 20V
FDB8030L		4.5mΩ, D2PAK, 30V

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configured in automatic retry mode (see *MOSFET Thermal Considerations*).

Using the MAX4271/MAX4273 in latched mode allows the consideration of MOSFETs with higher $R_{DS(ON)}$ and lower power ratings. A MOSFET can typically withstand single-shot pulses with higher dissipation than the specified package rating. Low MOSFET gate capacitance is not necessary since the inrush current limiting is achieved by limiting the gate dv/dt. Table 3 lists some recommended manufacturers and components.

Sense Resistor

The slow comparator threshold voltage is set at 50mV. Select a sense resistor that causes a 50mV voltage drop at a current level above the maximum normal operating current; typically, set the overload current at 1.2 to 1.5 times the nominal load current. The fast comparator threshold is typically set at 200mV. This will set the fault current limit at four times the overload current limit. The MAX4273 fast comparator threshold can be set between 50mV and 750mV; see Table 4 for a detailed listing.

Choose the sense resistor power rating according to the device configuration. If no retry mode is selected, $P_{RSENSE} = (I_{OVERLOAD})^2 \times R_{SENSE}$; if retry is selected, then $P_{RSENSE} = (I_{FAULT})^2 \times R_{SENSE} \times (t_{ON}/t_{RETRY})$.

Fast Comparator Threshold (RTH) (MAX4273)

The fast comparator threshold is determined by the external resistor connected at RTH. To select threshold voltages between 50mV and 750mV, use resistor values between 5kΩ and 75kΩ according to Figure 7.

Resistor values between 200Ω and 5kΩ are not recommended. Setting the threshold voltage of the fast comparator below 50mV will effectively override the slow comparator operation. The MAX4273 fast comparator can be disabled by shorting the RTH pin to V_{IN} (V_{IN} - 25mV or less). Ground RTH to set the threshold to 200mV internally.

Startup and Retry Timers (CTIM, CTON)

The startup (t_{START}) and retry (t_{RETRY}) timers are determined by the capacitors connected at CTIM and CTON. The capacitor connected to CTIM has two functions for the three devices as follows:

DEVICE	CTIM FUNCTION
MAX4271	Startup time
MAX4272	Startup time and sets retry timer
MAX4273	Sets retry timer

Table 3. Component Manufacturers

COMPONENT	MANUFACTURER	PHONE	WEB
Sense Resistors	Dale-Vishay	402-564-3131	www.vishay.com
	IRC	704-264-8861	www.irctt.com
MOSFETS	Fairchild	888-522-5372	www.fairchildsemi.com
	International Rectifier	310-322-3331	www.irf.com
	Motorola	602-244-3576	www.mot-sps.com/ppd/

Table 4. Current Levels vs. R_{SENSE}

R_{SENSE} (mΩ)	PART NUMBER	OVERLOAD THRESHOLD SET BY SLOW COMPARATOR (A)	FAULT CURRENT THRESHOLD SET BY FAST COMPARATOR (A)
10	MAX4271/MAX4272	5	20
	MAX4273	5	5 to 75
50	MAX4271/MAX4272	1	4
	MAX4273	1	1 to 15
100	MAX4271/MAX4272	0.5	2
	MAX4273	0.5	0.5 to 7.5

CTON determines the maximum time allowed to complete turn-on for the MAX4273. The default values for turn-on time (t_{ON}) and t_{RETRY} are chosen by leaving these pins floating; they are 10µs and 320µs, respectively. These are also the minimum values (not controlled and dependent on stray capacitance). Longer timings are determined by the size of the capacitor according to Figure 8, and can be determined in Table 5, which lists the startup and retry timing parameters.

Set the t_{ON} timer long enough to allow for the MOSFET to be enhanced and the load capacitor to be charged completely .

There are two ways of completing the startup sequence. **Case A** describes a startup sequence that does not use the current-limiting feature and slowly turns on the MOSFET by limiting the gate dv/dt. **Case B** uses the current-limiting feature and turns on the MOSFET as fast as possible while still preventing a high inrush current.

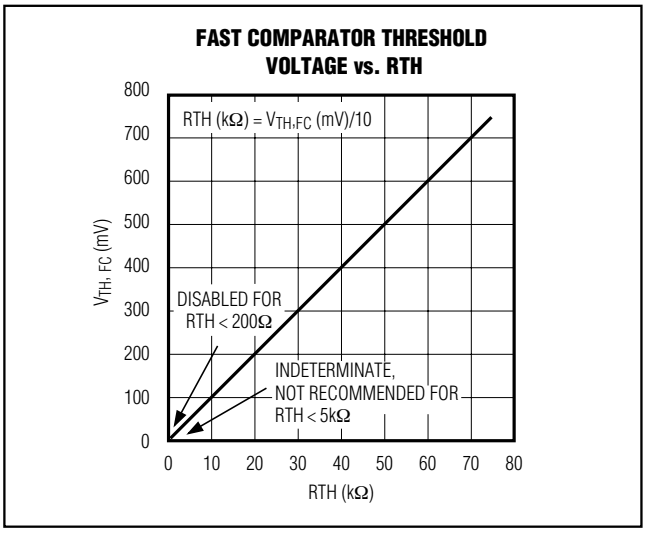


Figure 7. Fast Comparator Threshold vs. RTH

Case A. Startup Without Current Regulation

There are three ways to turn on the MOSFET without reaching the fast comparator current limit:

1) If the board capacitance is small, the inrush current is low.
2) If the gate capacitance is high, the MOSFET turns on slowly.
3) The fast comparator can be disabled (MAX4273 only).

In all three cases, t_{ON} is determined only by the charge required to enhance the MOSFET. Effectively, the small gate-charging current limits the output voltage dv/dt. This time can be extended by connecting an external capacitor between GATE and GND (MAX4271/MAX4272) (Figure 9) or between GATE and CEXT (MAX4273). The turn-on time is dominated by the external gate capacitance if this value is considerably higher than the MOSFET gate capacitance. Table 6 shows the timing required to enhance the recommended MOSFET with or without the external capacitor; Figure 3 shows the related waveforms and timing diagrams. (See Time to Charge Gate vs. C_{GATE} and Startup Time with C_{BOARD} = 0 in the *Typical Operating Characteristics*.) Remember that a high gate capacitance also increases the turn-off time (t_{OFF}), except in the case of a MAX4273 fast fault.

If an external gate capacitor is not used, R_S is not necessary. R_S prevents MOSFET self-oscillations that can occur when C_{GATE} is high while C_{BOARD} is low.

Electrical characteristics as specified by the manufacturer's data sheet are:

FDS6670A: C_{ISS} = 3200pF, Q_T(MAX) = 50nC, $R_{DS(ON)}$ = 8.2mΩ

IRF7401: C_{ISS} = 1600pF, Q_T(MAX) = 48nC, $R_{DS(ON)}$ = 22mΩ

MMSF5N03HD: C_{ISS} = 1200pF, Q_T(MAX) = 21nC, $R_{DS(ON)}$ = 40mΩ

Table 5. Startup and Retry Timing Parameters

PART	t_{ON}		t_{RETRY}	
	DEFAULT (µs)	EXTERNALLY SET	DEFAULT (µs)	EXTERNALLY SET
MAX4271	10	t_{ON} (ms) = 0.31 x **CTIM** (nF)	No retry available	
MAX4272	10	t_{ON} (ms) = 0.31 x **CTIM** (nF)	320	t_{RETRY} (ms) = 32 x t_{ON} = 10 x **CTIM** (nF)
MAX4273*	10	t_{ON} (ms) = 0.31 x **CTON** (nF)	320	t_{RETRY} (ms) =10 x **CTIM** (nF)

*MAX4273 retry feature disabled by connecting CTIM to V_{IN}.

3V to 12V Current-Limiting Hot-Swap Controllers with Autoretry, DualSpeed/BiLevel Fault Protection

Table 6. "No-Overcurrent" Turn-On Timing (Startup Without Current Limit)

DEVICE	C_{GATE} (nF)	MOSFET t_{ON} (ms)			MOSFET t_{OFF} (ms)		
		V_{IN} = 3V	V_{IN} = 5V	V_{IN} = 12V	V_{IN} = 3V	V_{IN} = 5V	V_{IN} = 12V
Fairchild FDS6670A	0	0.22	0.16	0.19	0.07	0.13	0.145
	22	2.3	2	3.2	0.54	1.1	1.95
International Rectifier IRF7401	0	0.175	0.130	0.16	0.075	0.13	0.16
	22	1.9	1.8	3.5	0.54	1.1	2.0
Motorola MMSF5N03HD	0	0.101	0.074	0.073	0.033	0.067	0.085
	22	2.0	1.8	3.2	0.470	1.0	1.95

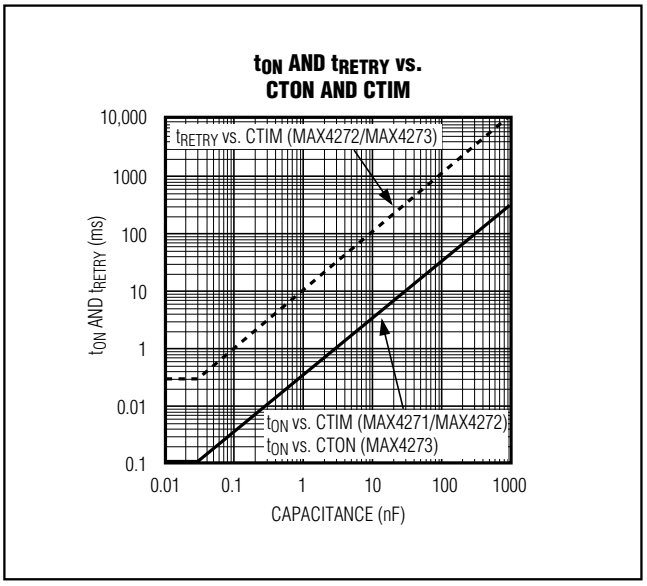


Figure 8. Startup and Retry Timeout

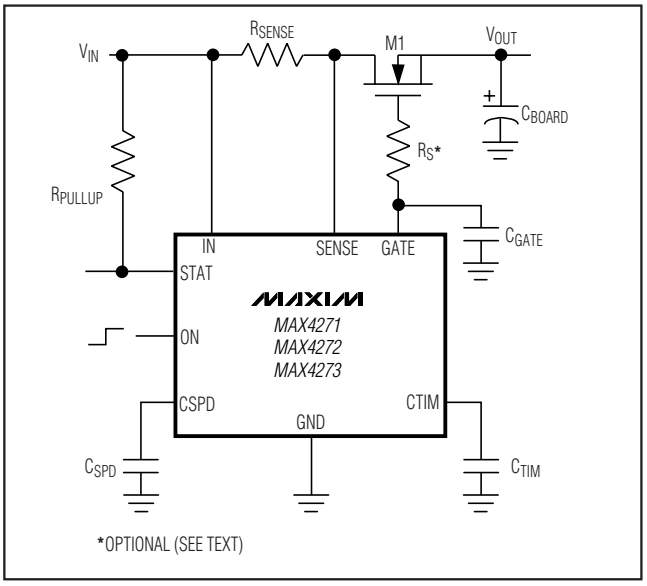


Figure 9. Operation with an External Capacitor

Case B. Startup With Current Regulation

In applications where the board capacitance (C_{BOARD}) at V_{OUT} is high, the inrush current causes a voltage drop across R_{SENSE} that exceeds the fast comparator threshold ($V_{FC,TH}$). In this case, the current charging C_{BOARD} can be considered constant and the turn-on time is determined by:

$$t_{ON} = C_{BOARD} \times V_{IN} / I_{FAST,SET}$$

where the maximum load current $I_{FAST,SET} = V_{FC,TH} / R_{SENSE}$. Figure 3 shows the waveforms and timing diagrams for a startup transient with current regulation. (See Startup Time (C_{BOARD} = 470µF) in the *Typical Operating Characteristics*.) When operating under this condition, an external gate capacitor is not required. Adding an external gate capacitor at GATE to GND reduces the regulated current ripple but increases t_{OFF} by increasing the gate delay (t_d) (Figure 4).

The actual startup time is determined by the longer of the two timings of Case A and Case B. Set the startup timer t_{START} at 2 × t_{ON} to guarantee enough time for the output voltage to settle; also take into consideration device parameter variation.

Retry

The retry timer defines the dead time before the IC tries to restart a startup sequence following a fault detection. This feature is available on the MAX4272/MAX4273.

Before selecting the retry timer value, determine how long a temporary high-current fault condition may be present. In the event of a permanent fault, the automatic retry will effectively force current pulses through the

MOSFET with a duty cycle equal to t_{ON}/t_{RETRY} and with a current equal to $I_{FAST,SET}$. Therefore, particular care has to be taken when choosing between immediate retry and board space needed to manage the power dissipation capabilities of the MOSFET (see *Thermal Considerations*). The duty cycle is fixed to 1/32 for the MAX4272, but can be varied in the MAX4273 by choosing CTIM and C_{TON} independently.

Additional External Gate Capacitance (CEXT)

An external gate capacitance can be connected at GATE. This increases the time required to enhance the MOSFET and further limits the output rise time. In the MAX4271/MAX4272, connect the external capacitor between GATE and GND. In the MAX4273, the external capacitor can be connected between GATE and CEXT or GND. If the capacitor is connected to CEXT, it is discharged to ground during a slow comparator fault but it is left floating during a fast comparator fault; this allows the device to turn off the external MOSFET faster during critical faults. (CEXT is biased at V_{IN}; therefore, use a nonpolarized capacitor). Capacitance connected from GATE to CEXT does little to decrease the regulated current ripple. Add a small capacitor (5nF) from GATE to GND. See the charging and discharging time vs. C_{GATE} graphs in the *Typical Operating Characteristics*.

Slow Comparator Response Time (CSPD)

The slow comparator threshold is set at 50mV, and its response time is determined by the external capacitor connected to CSPD (Figure 10).

A minimum response time of 20µs (typ) is achieved by leaving this pin floating. This time is determined internally and is not affected by stray capacitance at CSPD (up to 100pF).

Set the slow comparator response time to be longer than the normal operation load transients (see *Slow Comparator*).

ON and Reset Comparators

The ON comparator controls the ON/OFF function of these devices. The ON comparator is a precision voltage comparator that can be used for temperature monitoring or as an additional UVLO (Figure 11). The MAX4273 also features an uncommitted delayed comparator. This comparator can be used for voltage monitoring, power sequencing, or for generating a power-on reset signal for on-card microprocessors (Figure 12). Both comparator threshold voltages are set at $V_{REF}/2$ = 0.6V with a 3mV (typ) hysteresis.

The uncommitted comparator OUTC output is an open-drain output, and it is asserted low when its input voltage (INC) is below the threshold voltage. It goes into a high-impedance state 150ms after the voltage has risen above the threshold. The delay for negative-going edges is 10µs.

Figure 13 shows the MAX4273 used to monitor precisely the temperature of an external device such as the MOSFET. This configuration uses the uncommitted comparator to set the UVLO at a higher level by running its output into the ON comparator's input.

The ON comparator initiates startup when its input voltage (V_{ON}) rises above the threshold voltage and turns off the MOSFET when the voltage falls below the threshold. The propagation delay is 10µs going high or low. The ON comparator is also used to reset the MAX4271/MAX4273 (when CTIM = V_{IN}) after a fault condition (see *Latched/Autoretry*).

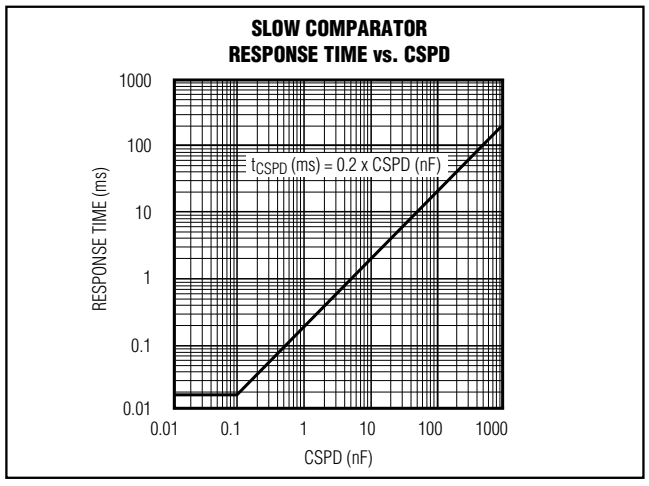


Figure 10. Slow Comparator Response Time vs. CSPD

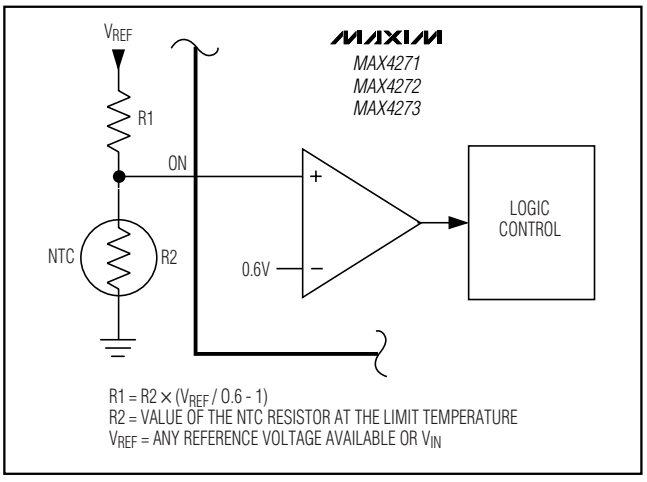


Figure 11. Temperature Monitoring and Protection

3V to 12V Current-Limiting Hot-Swap Controllers with Autoretry, DualSpeed/BiLevel Fault Protection

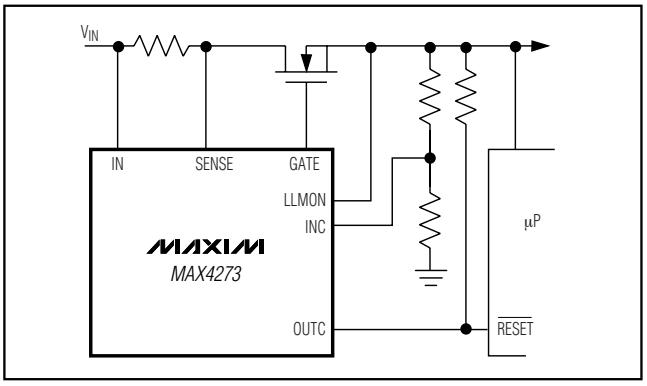


Figure 12. Power-On Reset

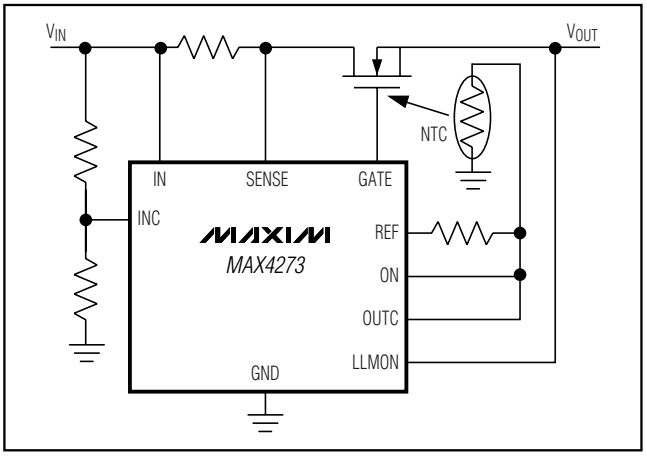


Figure 13. Power-On Reset and Temperature

The ON and INC comparator inputs and the STAT and OUTC can be pulled to voltages up to 14V independently of V_{IN}, thus allowing parts to be daisy-chained and not be turned on through the internal protection diodes. In some applications, it is useful to use connectors with staggered leads. In Figure 14, the ON pin forces the removable board to be powered up only when all connections are made.

Auxiliary Vcc

The auxiliary V_{CC} is available on the MAX4273 and is used to sustain the input voltage required for the device to operate during a short-circuit condition on the board. When a short occurs, the main system power supply could collapse and the MAX4273 will not have enough voltage to keep the gate drive operational and turn off the external MOSFET. If the fault is not removed, the system could remain in a sustained short-circuit state.

Connect a 1µF capacitor from the AUXVCC pin to GND. This capacitor will deliver the necessary energy to the gate drive until the MOSFET is turned off and the main supply recovers. The 1µF capacitor is charged from V_{IN} through an internal switch during normal operation.

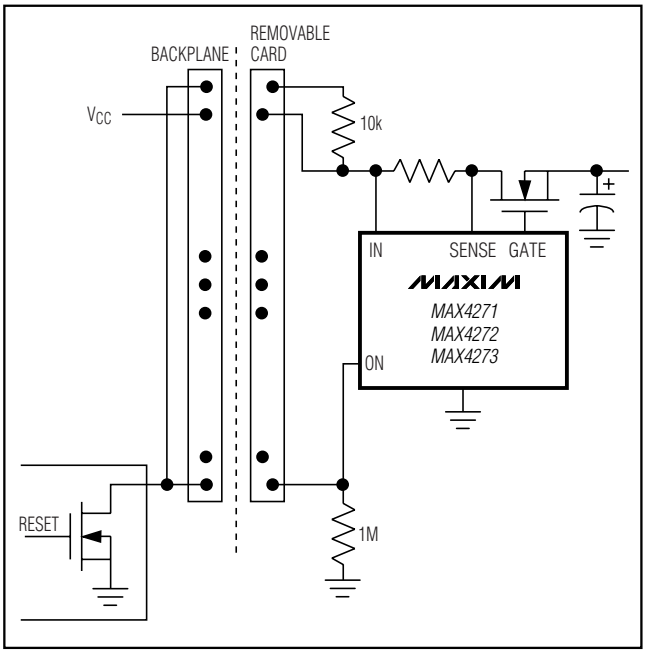


Figure 14. Fail-Safe Connector

Maximum Load Capacitance

The MAX4271/MAX4272/MAX4273 can be used on the backplane to regulate current upon insertion of a removable card (Figure 16). This allows multiple cards with different input capacitance to be inserted into the same slot even if the card doesn't have on-board hot-swap protection.

The MAX4271/MAX4272/MAX4273 current-limiting feature is active during the startup period set by CTIM. The startup period can be triggered if V_{IN} is connected to ON through a trace on the card. Once t_{START} has expired (timed out), the load capacitance has to be charged or a fault condition is detected. To ensure startup with a fixed CTIM, t_{START} has to be longer than the time required to charge the board capacitance. The maximum load capacitance is calculated as follows:

$$C_{BOARD} < t_{START} \times I_{FAST,SET} / V_{IN}$$

Input Transients

The voltage at V_{IN} must be above the UVLO during inrush and fault conditions. When a short condition occurs on the board, the fault current can be higher than the fast comparator current limit. The gate voltage

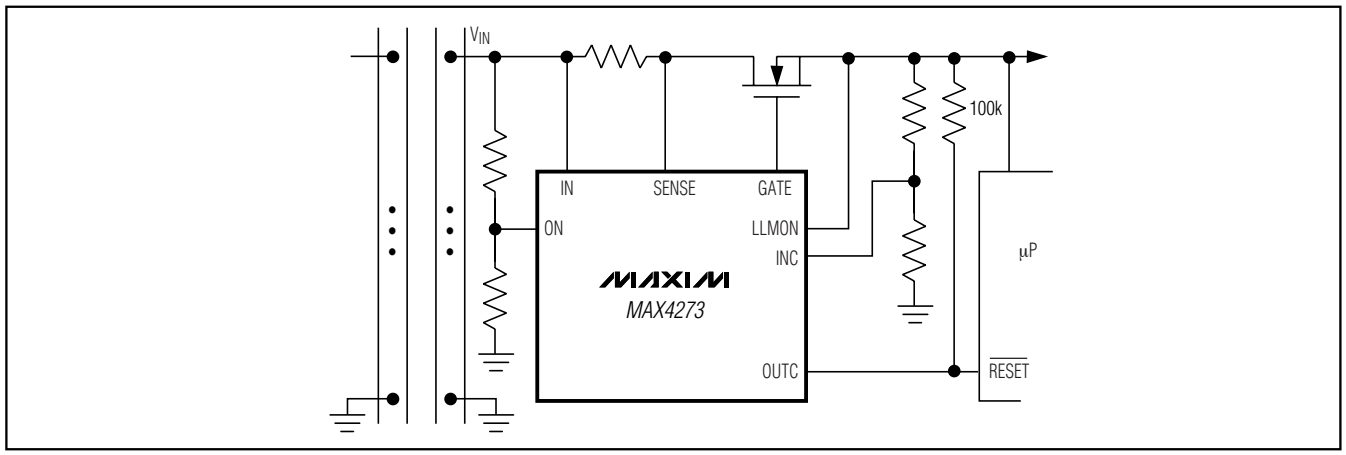


Figure 15. Adjustable Undervoltage Lockout and Output Voltage Reset Generator

is discharged immediately, but note that the MOSFET is not completely off until $V_{GS} < V_{TH}$. If the main system power supply collapses below UVLO, the MAX4271/MAX4272/MAX4273 will force the device to restart in startup mode with a 150ms delay once the supply has recovered. The main system power supply must be able to deliver this fault current without excessive voltage drop.

The MOSFET is turned off in a very short time; therefore, the resulting dv/dt can be considerable. The backplane delivering the power to the external card must have a fairly low inductance to limit the voltage transients caused by the removal of a fault. Bypassing the input with a small capacitor alleviates false UVLO trips due to these transients.

MOSFET Thermal Considerations

During normal operation, the MOSFET dissipates little power, it is fully turned on, and its $R_{DS(ON)}$ is minimal. The power dissipated in normal operation is $P_D = I_{LOAD}^2 \times R_{DS(ON)}$. A considerable amount of power is dissipated during the startup and turn-off transients. The design must take into consideration the worst-case scenario of a continuous short-circuit fault present on the board. Two cases need to be considered:

1) The single turn-on with the device latched after a fault (when using MAX4271 or MAX4273 in latched mode)
2) The continuous automatic retry (when using the MAX4272 or MAX4273 in retry mode)

Use the following equation to calculate the maximum transient thermal resistance (in °C/W) required for an output short to ground:

$$Z_{\theta JA}\ (max) = (T_{JMAX} - T_A) / (V_{IN} \times I_{FAST,\ SET})$$

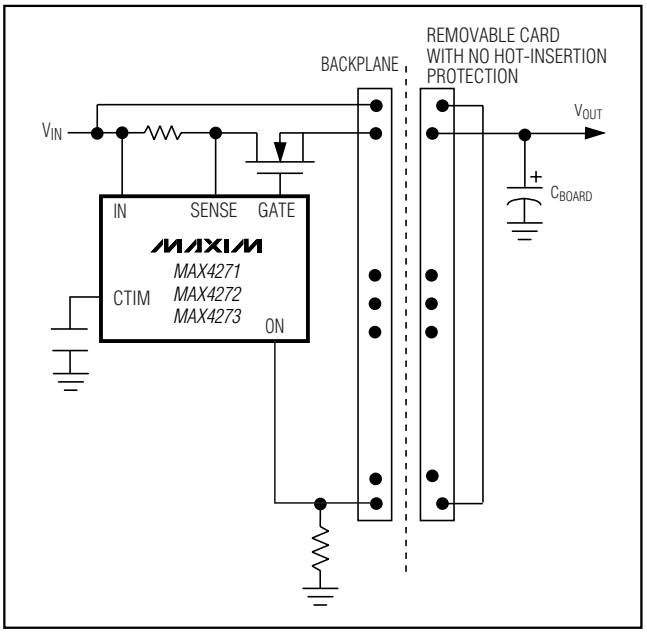


Figure 16. Using the MAX4271/MAX4272/MAX4273 on a Backplane

MOSFET manufacturers typically include curves for the transient thermal resistance, $Z_{\theta JA}$, of the package (Figure 17). Find the thermal impedance of the MOSFET by using t_{START} as the pulse duration and by choosing the single pulse curve for latched mode parts or by choosing the duty cycle = 0.03 curve for the MAX4272 (the duty cycle is fixed at 32:1). If the $Z_{\theta JA}$ required is less than that of the package, reduce t_{START}, reduce $I_{FAST,SET}$, use a heatsink on the MOSFET, or choose one with better thermal characteristics.

3V to 12V Current-Limiting Hot-Swap Controllers with Autoretry, DualSpeed/BiLevel Fault Protection

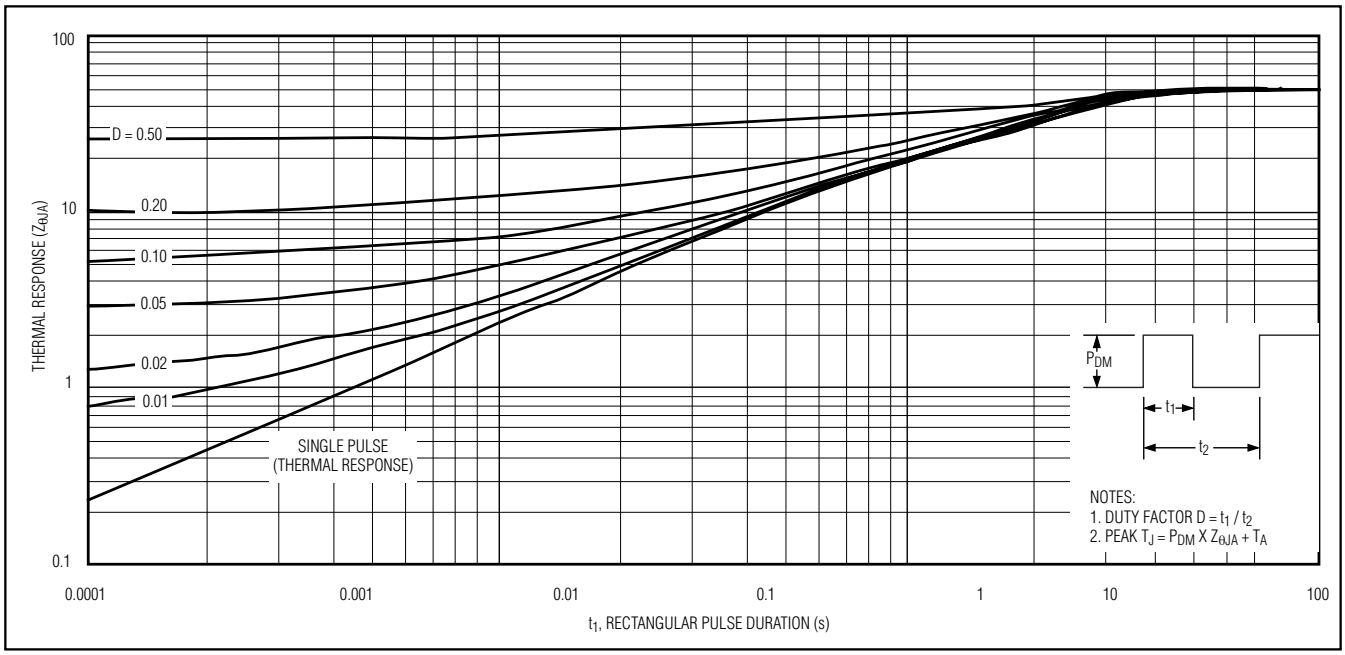


Figure 17. Example Curves (from IRF7413A) for Maximum Effective Transient Thermal Impedance, Junction to Ambient

For the MAX4273 in retry mode, the duty cycle can be adjusted. Use the MOSFET $Z_{\theta JA}$ curve and the t_{START} pulse duration to choose a maximum duty cycle, D. Calculate the retry time:

$$t_{RETRY} = t_{START} / D$$

Use Figure 8 to determine CTIM given t_{RETRY}.

Design Procedure (MAX4273)

GIVEN:

- Hot-swap 5V supply to a 1000µF card
- MOSFET IRF7413A: $R_{DS(on)}$ = 0.0135Ω
- Operating current = 4A
- Overload current = 5A
- System current limit = 10A
- Retry enabled

PROCEDURE:

1) Select the current sense resistor:

Slow comparator threshold = 50mV

Overload current = 5A

Current-sense resistor value = 50mV/5A = 10mΩ

The device will initiate a slow fault if the load current is greater than 5A for longer than 20µs after startup.

2) Set fast comparator threshold (RTH).

The MOSFET pulsed drain current limit is 58A. Another consideration for setting the current limit is the system requirement. Systems may glitch if 58A load transients are present. For this example, the load transient will be limited to 10A:

System current limit = 10A

Current-sense resistor = 10mΩ

Fast comparator threshold = 10mΩ × 10A = 100mV

Select RTH from Figure 7, RTH = 10kΩ

3) Set Startup timer.

Startup current = System current limit = 10A

V_{IN} = 5V

C_{BOARD} = 1000µF

t_{ON} = 1000µF × 5V / 10A = 500µs

Give a factor of 2 guardband on the startup timer.

t_{START} = 2 × t_{ON} = 1.0ms

From Figure 8, C_{TON} = 3000pF.

4) Select Retry Timeout.

V_{IN} = 5V.

I_{MAX} = 10A.

Peak junction temperature, T_J = 150°C$_J$

Peak ambient temperature, T_A = 85°C$_A$

Calculate the MOSFET thermal resistance required for a short to ground.

$Z_{\theta JA}$ = (150°C - 85°C) / (5V × 10A) = 1.30°C/W

Using t_{START} = 1ms as the pulse duration, use Figure 17 to select a duty cycle. The duty cycle should be about 0.01 or less. This implies t_{RETRY} = t_{START} / 0.01 = 100ms. From Figure 8, C_{TIM} = 10nF.

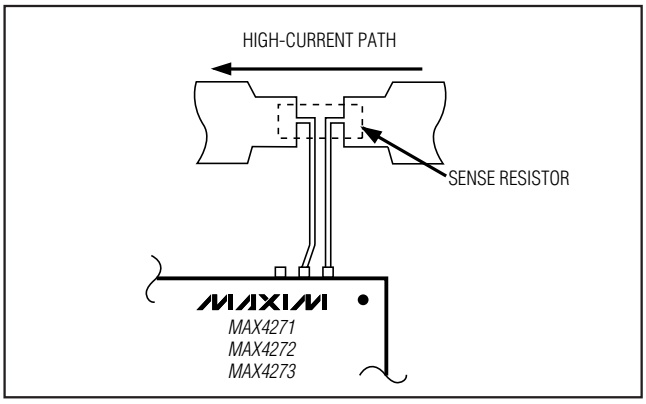


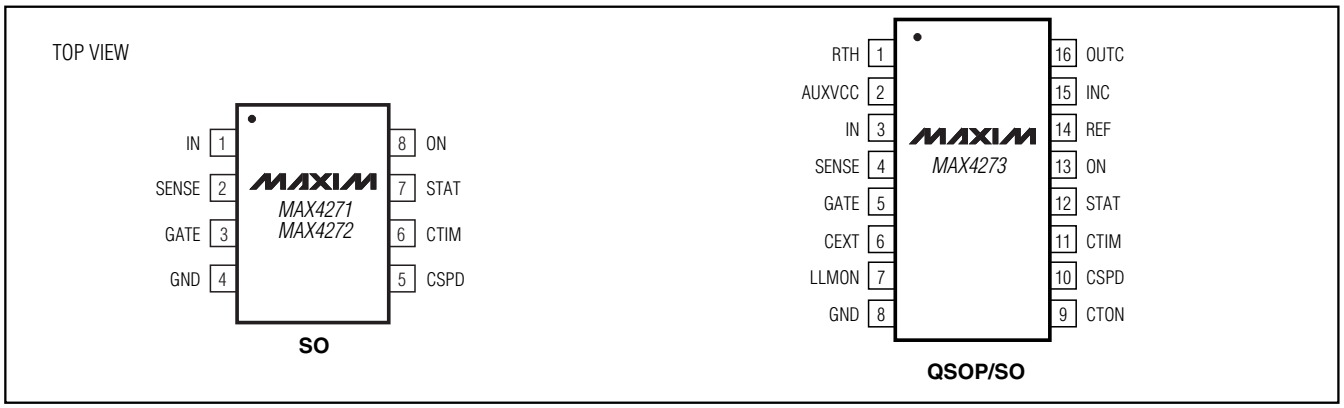
Figure 18. Kelvin Connections for the Current-Sense Resistors

Layout Considerations

To take full tracking advantage of the switch response time to an output fault condition, it is important to keep all traces as short as possible and to maximize the high-current trace dimensions to reduce the effect of undesirable parasitic inductance. Place the MAX4271/MAX4272/MAX4273 close to the card's connector. Use a ground plane to minimize impedance and inductance. Minimize the current-sense resistor trace length (<10mm), and ensure accurate current sensing with Kelvin connections (Figure 18).

When the output is short circuited, the voltage drop across the external MOSFET becomes large. Hence, the power dissipation across the switch increases, as does the die temperature. An efficient way to achieve good power dissipation on a surface-mount package is to lay out two copper pads directly under the MOSFET package on both sides of the board. Connect the two pads to the ground plane through vias, and use enlarged copper mounting pads on the top side of the board.

Pin Configurations



3V to 12V Current-Limiting Hot-Swap Controllers with Autoretry, DualSpeed/BiLevel Fault Protection

Package Information

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to **www.maxim-ic.com/packages**.)

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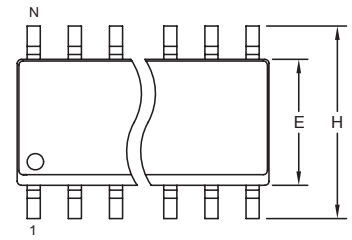
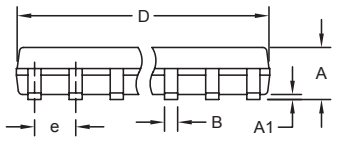


TOP VIEW

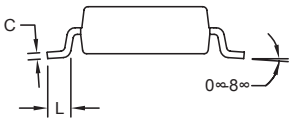
	INCHES		MILLIMETERS	
DIM	MIN	MAX	MIN	MAX
A	0.053	0.069	1.35	1.75
A1	0.004	0.010	0.10	0.25
B	0.014	0.019	0.35	0.49
C	0.007	0.010	0.19	0.25
e	0.050 BSC		1.27 BSC	
E	0.150	0.157	3.80	4.00
H	0.228	0.244	5.80	6.20
L	0.016	0.050	0.40	1.27

VARIATIONS:

	INCHES		MILLIMETERS			
DIM	MIN	MAX	MIN	MAX	N	MS012
D	0.189	0.197	4.80	5.00	8	AA
D	0.337	0.344	8.55	8.75	14	AB
D	0.386	0.394	9.80	10.00	16	AC



FRONT VIEW



SIDE VIEW

NOTES:
1. D&E DO NOT INCLUDE MOLD FLASH.
2. MOLD FLASH OR PROTRUSIONS NOT TO EXCEED 0.15mm (.006").
3. LEADS TO BE COPLANAR WITHIN 0.10mm (.004").
4. CONTROLLING DIMENSION: MILLIMETERS.
5. MEETS JEDEC MS012.
6. N = NUMBER OF PINS.

DALLAS SEMICONDUCTOR MAXIM

PROPRIETARY INFORMATION

TITLE: PACKAGE OUTLINE, .150" SOIC

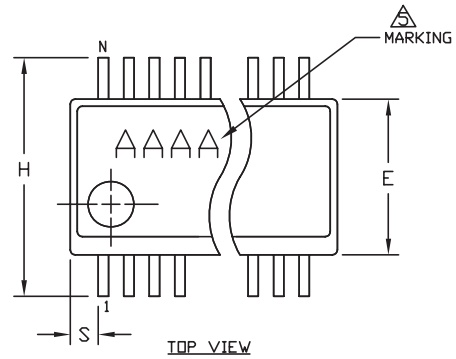
APPROVAL	DOCUMENT CONTROL NO.	REV.	
	21-0041	B	1/1

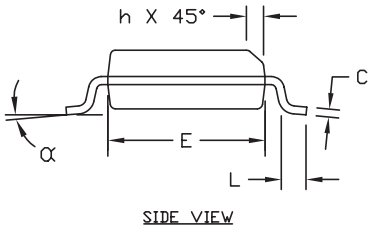
3V to 12V Current-Limiting Hot-Swap Controllers with Autoretry, DualSpeed/BiLevel Fault Protection

Package Information (continued)

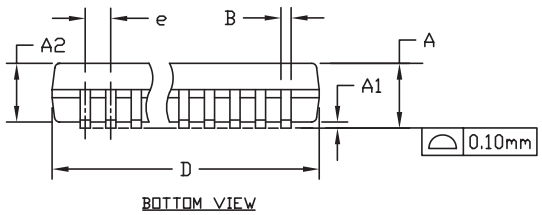
(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to **www.maxim-ic.com/packages**.)

QSOP.EPS





DIM	INCHES MIN	INCHES MAX	MILLIMETERS MIN	MILLIMETERS MAX
A	.053	.069	1.35	1.75
A1	.004	.010	.102	.254
A2	.049	.065	1.245	1.651
B	.008	.012	0.20	0.30
C	.0075	.0098	0.191	0.249
D	SEE VARIATIONS			
E	.150	.157	3.81	3.99
e	.025 BSC		0.635 BSC	
H	.230	.244	5.84	6.20
h	.010	.016	0.25	0.41
L	.016	.035	0.41	0.89
N	SEE VARIATIONS			
α	0°	8°	0°	8°



VARIATIONS:

	INCHES MIN.	INCHES MAX.	MILLIMETERS MIN.	MILLIMETERS MAX.	N	PKG CODES
D	.189	.196	4.80	4.98	16	E16-1, E16M-1, E16-3F, E16-4, E16-5, E16-6, E16-8F
S	.0020	.0070	0.05	0.18		
D	.337	.344	8.56	8.74	20	E20-1, E20-2
S	.0500	.0550	1.270	1.397		
D	.337	.344	8.56	8.74	24	E24-1, E24-2, E24-3F
S	.0250	.0300	0.635	0.762		
D	.386	.393	9.80	9.98	28	E28-1, E28M-1, E28-2
S	.0250	.0300	0.635	0.762		

NOTES:
1. D & E DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS.
2. MOLD FLASH OR PROTRUSIONS NOT TO EXCEED .006" PER SIDE.
3. CONTROLLING DIMENSIONS: INCHES.
4. MEETS JEDEC MO137.
5. MARKING SHOWN IS FOR PKG. ORIENTATION ONLY.
6. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND PbFREE (+) PKG. CODES.

—DRAWING NOT TO SCALE—

MAXIM

TITLE:
PACKAGE OUTLINE
QSOP .150", .025" LEAD PITCH

APPROVAL	DOCUMENT CONTROL NO.	REV.	
	21-0055	G	1/1

3V to 12V Current-Limiting Hot-Swap Controllers with Autoretry, DualSpeed/BiLevel Fault Protection

Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	4/00	Initial release	—
1	7/00	Removed future product designation for MAX4273	1
1a	4/01	Updated footer	1
2	12/07	Included Package codes in Ordering Information table, Updated Note 5 and Package Outlines	1, 4, 24

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Поставка электронных компонентов

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Наши преимущества:

- Поставка оригинальных импортных электронных компонентов напрямую с производств Америки, Европы и Азии, а так же с крупнейших складов мира;
- Широкая линейка поставок активных и пассивных импортных электронных компонентов (более 30 млн. наименований);
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
- Экспресс доставка в любую точку России;
- Помощь Конструкторского Отдела и консультации квалифицированных инженеров;
- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
- Поставка электронных компонентов под контролем ВП;
- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001;
- При необходимости вся продукция военного и аэрокосмического назначения проходит испытания и сертификацию в лаборатории (по согласованию с заказчиком);
- Поставка специализированных компонентов военного и аэрокосмического уровня качества (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Actel, Aeroflex, Peregrine, VPT, Syfer, Eurofarad, Texas Instruments, MS Kennedy, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

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