

16/32-Bit

Architecture

XC2288H, XC2289H

16/32-Bit Single-Chip Microcontroller
with 32-Bit Performance

XC2000 Family / High Line

Data Sheet

V1.3 2011-07

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Page	Subjects (major changes since last revision)
11	Clarified available Flash and SRAM memory allocation.
77	USIC "QSPI" protocol shortcut removed due to ambiguity (interpreted as Queued SPI or Quad SPI).
108	Relaxed the conditions for short-term deviation of internal clock source frequency Δf_{INT} .
108	Added startup time from power-on t_{SPO}
112	Removed the 128MHz conditions for N_{WSFLE}
119	Added the minimum PLL free running frequency. Reduced the min/max bandwidth.
139	Added notes on required pad settings for FlexRay operation. Added note on FlexRay operation frequency.
139	Corrected parameter name f(oscdd) to f(osc) in FleyRay timing table.
147	Thermal resistance values updated.

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**16/32-Bit Single-Chip Microcontroller
with 32-Bit Performance
XC228xH (XC2000 Family)**

1 Summary of Features

For a quick overview and easy reference, the features of the XC228xH are summarized here.

- High-performance CPU with five-stage pipeline and MPU
 - 10 ns instruction cycle @ 100 MHz CPU clock (single-cycle execution)
 - One-cycle 32-bit addition and subtraction with 40-bit result
 - One-cycle multiplication (16 × 16 bit)
 - Background division (32 / 16 bit) in 21 cycles
 - One-cycle multiply-and-accumulate (MAC) instructions
 - Enhanced Boolean bit manipulation facilities
 - Zero-cycle jump execution
 - Additional instructions to support HLL and operating systems
 - Register-based design with multiple variable register banks
 - Fast context switching support with two additional local register banks
 - 16 Mbytes total linear address space for code and data
 - 1,024 Bytes on-chip special function register area (C166 Family compatible)
 - Integrated Memory Protection Unit (MPU)
- Interrupt system with 16 priority levels providing 112 interrupt nodes
 - Selectable external inputs for interrupt generation and wake-up
 - Fastest sample-rate 10 ns
- Eight-channel interrupt-driven single-cycle data transfer with Peripheral Event Controller (PEC), 24-bit pointers cover total address space
- Clock generation from internal or external clock sources, using on-chip PLL or prescaler
- Hardware CRC-Checker with Programmable Polynomial to Supervise On-Chip Memory Areas
- On-chip memory modules
 - 8 Kbytes on-chip stand-by RAM (SBRAM)
 - 2 Kbytes on-chip dual-port RAM (DPRAM)
 - 24 Kbytes on-chip data SRAM (DSRAM)
 - Up to 112 Kbytes on-chip program/data SRAM (PSRAM)
 - Up to 1,600 Kbytes on-chip program memory (Flash memory)
 - Memory content protection through Error Correction Code (ECC)

Summary of Features

- On-Chip Peripheral Modules
 - Two synchronizable A/D Converters with up to 24 channels, 10-bit resolution, conversion time below 1 μ s, optional data preprocessing (data reduction, range check), broken wire detection
 - Up to two 16-channel general purpose capture/compare units (CCx)
 - Up to four capture/compare units for flexible PWM signal generation (CCU6x)
 - Multi-functional general purpose timer unit with 5 timers
 - Up to 10 serial interface channels to be used as UART, LIN, high-speed synchronous channel (SPI/QSPI), IIC bus interface (10-bit addressing, 400 kbit/s), IIS interface
 - On-chip MultiCAN interface (Rev. 2.0B active) with up to 256 message objects (Full CAN/Basic CAN) on up to 6 CAN nodes and gateway functionality
 - FlexRay™ module (E-Ray) according to protocol specification V2.1, with 2 nodes
 - On-chip system timer and on-chip real time clock
- Up to 12 Mbytes external address space for code and data
 - Programmable external bus characteristics for different address ranges
 - Multiplexed or demultiplexed external address/data buses
 - Selectable address bus width
 - 16-bit or 8-bit data bus width
 - Five programmable chip-select signals
- Single power supply from 3.0 V to 5.5 V
- Power reduction and wake-up modes with flexible power management
- Programmable watchdog timer and oscillator watchdog
- Up to 118 general purpose I/O lines
- On-chip bootstrap loaders
- Supported by a full range of development tools including C compilers, macro-assembler packages, emulators, evaluation boards, HLL debuggers, simulators, logic analyzer disassemblers, programming boards
- On-chip debug support via Device Access Port (DAP) or JTAG interface
- 144-pin Green LQFP package, 0.5 mm (19.7 mil) pitch

Ordering Information

The ordering code for an Infineon microcontroller provides an exact reference to a specific product. This ordering code identifies:

- the derivative itself, i.e. its function set, the temperature range, and the supply voltage
- the temperature range:
 - SAF-...: -40°C to 85°C
 - SAH-...: -40°C to 110°C
 - SAK-...: -40°C to 125°C
- the package and the type of delivery.

For ordering codes for the XC228xH please contact your sales representative or local distributor.

This document describes several derivatives of the XC228xH group:

Basic Device Types are readily available and

Special Device Types are only available on request.

As this document refers to all of these derivatives, some descriptions may not apply to a specific product, in particular to the special device types.

For simplicity the term **XC228xH** is used for all derivatives throughout this document.

1.1 Basic Device Types

Basic device types are available and can be ordered through Infineon's direct and/or distribution channels.

Table 1 Synopsis of XC228xH Basic Device Types

Derivative ¹⁾	Flash Memory ²⁾	PSRAM ³⁾	Capt./Comp. Modules	ADC ⁴⁾ Chan.	Interfaces ⁴⁾
XC2288H-200F _x L	1,600 Kbytes	112 Kbytes	CC1/2 CCU60/1/2/3	16 + 8	4 CAN Nodes, 10 Serial Chan.

- 1) x is a placeholder for available speed grade in MHz. Can be 100 only.
- 2) Specific information about the on-chip Flash memory in [Table 3](#) and [Table 4](#).
- 3) All derivatives additionally provide 8 Kbytes SBRAM, 2 Kbytes DPRAM, and 24 Kbytes DSRAM.
- 4) Specific information about the available channels in [Table 5](#).
Analog input channels are listed for each Analog/Digital Converter module separately (ADC0 + ADC1).

1.2 Special Device Types

Special device types are only available for high-volume applications on request.

Table 2 Synopsis of XC228xH Special Device Types

Derivative ¹⁾	Flash Memory ²⁾	PSRAM ³⁾	Capt./Comp. Modules	ADC ⁴⁾ Chan.	Interfaces ⁴⁾
XC2288H-136F _x L	1,088 Kbytes	80 Kbytes	CC1/2 CCU60/1/2/3	16 + 8	4 CAN Nodes 10 Serial Chan.
XC2289H-136F _x L	1,088 Kbytes	80 Kbytes	CC1/2 CCU60/1/2/3	16 + 8	2 FlexRay Nodes 6 CAN Nodes 10 Serial Chan.
XC2289H-200F _x L	1,600 Kbytes	112 Kbytes	CC1/2 CCU60/1/2/3	16 + 8	2 FlexRay Nodes 6 CAN Nodes 10 Serial Chan.

- 1) x is a placeholder for available speed grade in MHz. Can be 80 or 100.
- 2) Specific information about the on-chip Flash memory in [Table 3](#) and [Table 4](#).
- 3) All derivatives additionally provide 8 Kbytes SBRAM, 2 Kbytes DPRAM, and 24 Kbytes DSRAM.
- 4) Specific information about the available channels in [Table 5](#).
Analog input channels are listed for each Analog/Digital Converter module separately (ADC0 + ADC1).

1.3 Definition of Feature Variants

The XC228xH types are offered with several Flash memory sizes. [Table 3](#) and [Table 4](#) describe the location of the available Flash memory.

Table 3 Continuous Flash Memory Ranges

Total Flash Size	1st Range ¹⁾	2nd Range	3rd Range
1,600 Kbytes	C0'0000 _H ... C0'FFFF _H	C1'0000 _H ... D8'FFFF _H	n.a.
1,088 Kbytes	C0'0000 _H ... C0'FFFF _H	C1'0000 _H ... CF'FFFF _H	D8'0000 _H ... D8'FFFF _H

1) The uppermost 4-Kbyte sector of the first Flash segment is reserved for internal use (C0'F000_H to C0'FFFF_H).

Table 4 Flash Memory Module Allocation (in Kbytes)

Total Flash Size	Flash 0 ¹⁾	Flash 1	Flash 2	Flash 3	Flash 4	Flash 5	Flash 6
1,600	256	255	256	256	256	256	64
1,088	256	255	256	256	-	-	64

1) The uppermost 4-Kbyte sector of the first Flash segment is reserved for internal use (C0'F000_H to C0'FFFF_H).

The XC228xH types are offered with different interface options. [Table 5](#) lists the available channels for each option.

Table 5 Interface Channel Association

Total Number	Available Channels / Message Objects
16 ADC0 channels	CH0 ... CH15
8 ADC1 channels	CH0 ... CH7
6 CAN nodes	CAN0, CAN1, CAN2, CAN3, CAN4, CAN5 256 message objects
4 CAN nodes	CAN0, CAN1, CAN2, CAN3 128 Message Objects
10 serial channels	U0C0, U0C1, U1C0, U1C1, U2C0, U2C1, U3C0, U3C1, U4C0, U4C1

Summary of Features

The XC228xH types are offered with several PSRAM memory sizes. **Figure 1** shows the allocation rules. For example 80 Kbytes of PSRAM will be allocated at E0'0000h-E1'3FFFh.

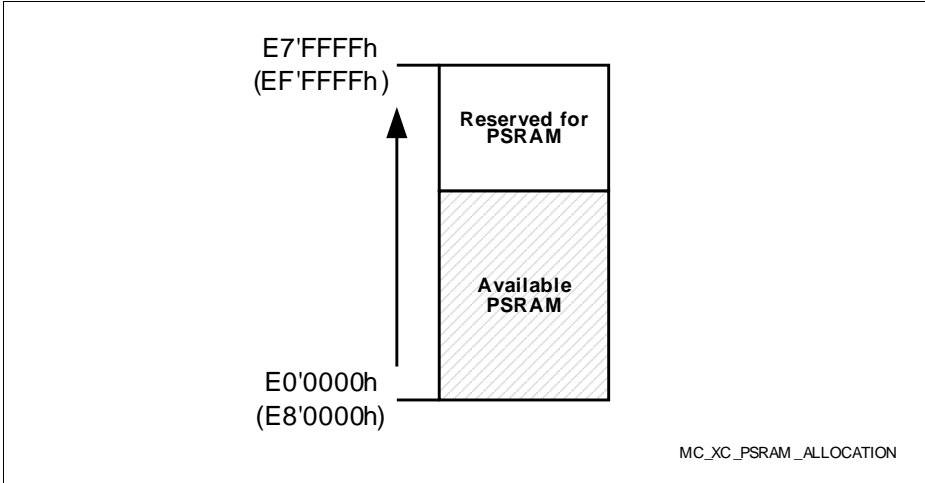


Figure 1 PSRAM Allocation

2 General Device Information

The XC228xH series (16/32-Bit Single-Chip Microcontroller with 32-Bit Performance) is a part of the Infineon XC2000 Family of full-feature single-chip CMOS microcontrollers. These devices extend the functionality and performance of the C166 Family in terms of instructions (MAC unit), peripherals, and speed. They combine high CPU performance (up to 100 million instructions per second) with extended peripheral functionality and enhanced IO capabilities. Optimized peripherals can be adapted flexibly to meet the application requirements. These derivatives utilize clock generation via PLL and internal or external clock sources. On-chip memory modules include program Flash, program RAM, and data RAM.

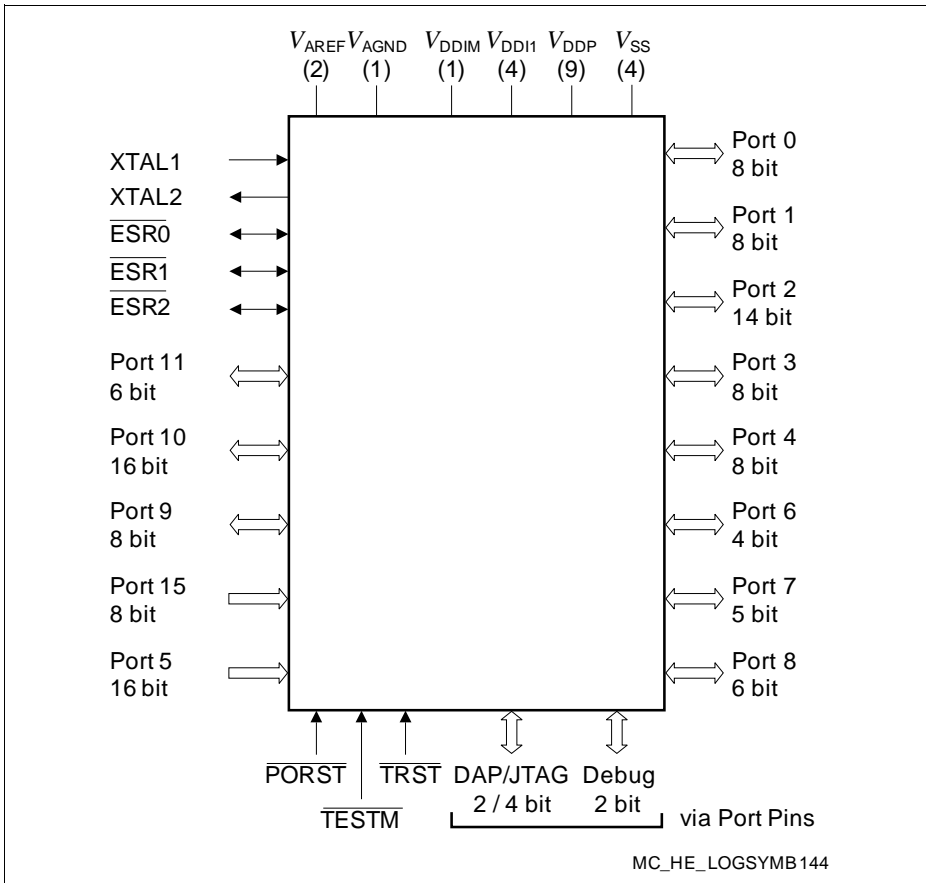


Figure 2 Logic Symbol

2.1 Pin Configuration and Definition

The pins of the XC228xH are described in detail in [Table 6](#), which includes all alternate functions. For further explanations please refer to the footnotes at the end of the table. The following figure summarizes all pins, showing their locations on the four sides of the package.

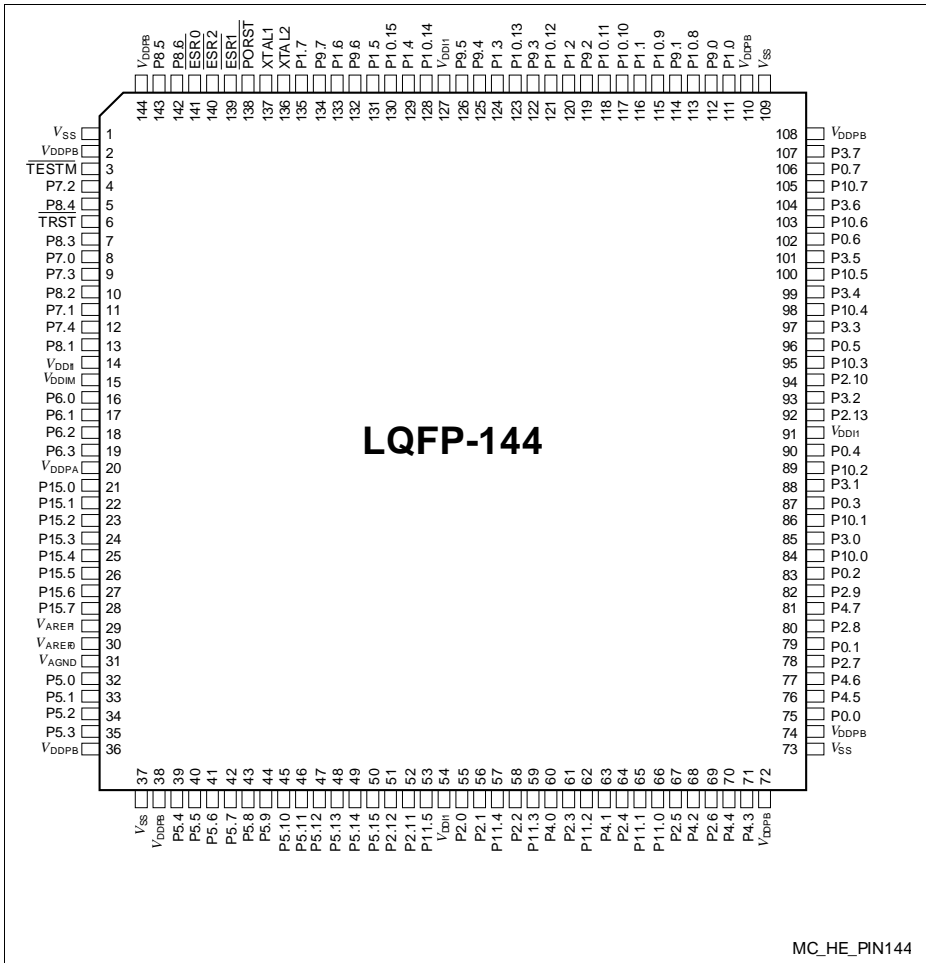


Figure 3 XC228xH Pin Configuration (top view)

Key to Pin Definitions

- **Ctrl.:** The output signal for a port pin is selected by bit field PC in the associated register Px_IOCry. Output O0 is selected by setting the respective bit field PC to 1x00_B, output O1 is selected by 1x01_B, etc. Output signal OH is controlled by hardware.
- **Type:** Indicates the pad type and its power supply domain (A, B, M, 1).
 - St: Standard pad
 - Sp: Special pad e.g. XTALx
 - DP: Double pad - can be used as standard or high speed pad
 - In: Input only pad
 - PS: Power supply pad

Table 6 Pin Definitions and Functions

Pin	Symbol	Ctrl.	Type	Function
3	TESTM	I	In/B	Testmode Enable Enables factory test modes, must be held HIGH for normal operation (connect to V_{DDPB}). An internal pull-up device will hold this pin high when nothing is driving it.
4	P7.2	O0 / I	St/B	Bit 2 of Port 7, General Purpose Input/Output
	EMUX0	O1	St/B	External Analog MUX Control Output 0 (ADC1)
	TxDC4	O2	St/B	CAN Node 4 Transmit Data Output
	TxDC5	O3	St/B	CAN Node 5 Transmit Data Output
	CCU62_CCP OS0A	I	St/B	CCU62 Position Input 0
	TDI_C	IH	St/B	JTAG Test Data Input If JTAG pos. C is selected during start-up, an internal pull-up device will hold this pin high when nothing is driving it.

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
5	P8.4	O0 / I	St/B	Bit 4 of Port 8, General Purpose Input/Output
	CCU60_COU T61	O1	St/B	CCU60 Channel 1 Output
	CCU62_CC6 1	O2	St/B	CCU62 Channel 1 Output
	CC1_CC2	O3	St/B	CC1 Channel 2 Output
	TMS_D	IH	St/B	JTAG Test Mode Selection Input If JTAG pos. D is selected during start-up, an internal pull-up device will hold this pin high when nothing is driving it.
	CCU62_CC6 1INB	I	St/B	CCU62 Channel 1 Input
	U4C1_DX1A	I	St/B	USIC4 Channel 1 Shift Clock Input
6	$\overline{\text{TRST}}$	I	In/B	Test-System Reset Input For normal system operation, pin $\overline{\text{TRST}}$ should be held low. A high level at this pin at the rising edge of $\overline{\text{PORST}}$ activates the XC228xH's debug system. In this case, pin $\overline{\text{TRST}}$ must be driven low once to reset the debug system. An internal pull-down device will hold this pin low when nothing is driving it.
7	P8.3	O0 / I	St/B	Bit 3 of Port 8, General Purpose Input/Output
	CCU60_COU T60	O1	St/B	CCU60 Channel 0 Output
	CCU62_CC6 0	O2	St/B	CCU62 Channel 0 Output
	U4C1_SELO 0	O3	St/B	USIC4 Channel 1 Select/Control 0 Output
	TDI_D	IH	St/B	JTAG Test Data Input If JTAG pos. D is selected during start-up, an internal pull-up device will hold this pin low when nothing is driving it.
	CCU62_CC6 0INB	I	St/B	CCU62 Channel 0 Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
8	P7.0	O0 / I	St/B	Bit 0 of Port 7, General Purpose Input/Output
	T3OUT	O1	St/B	GPT12E Timer T3 Toggle Latch Output
	T6OUT	O2	St/B	GPT12E Timer T6 Toggle Latch Output
	TDO_A	OH / IH	St/B	JTAG Test Data Output / DAP1 Input/Output If DAP pos. 0 or 2 is selected during start-up, an internal pull-down device will hold this pin low when nothing is driving it.
	ESR2_1	I	St/B	ESR2 Trigger Input 1
	RxDC4B	I	St/B	CAN Node 4 Receive Data Input
	U4C1_DX0C	I	St/B	USIC4 Channel 0 Receive Data Input
9	P7.3	O0 / I	St/B	Bit 3 of Port 7, General Purpose Input/Output
	EMUX1	O1	St/B	External Analog MUX Control Output 1 (ADC1)
	U0C1_DOUT	O2	St/B	USIC0 Channel 1 Shift Data Output
	U0C0_DOUT	O3	St/B	USIC0 Channel 0 Shift Data Output
	CCU62_CCP OS1A	I	St/B	CCU62 Position Input 1
	TMS_C	IH	St/B	JTAG Test Mode Selection Input If JTAG pos. C is selected during start-up, an internal pull-up device will hold this pin low when nothing is driving it.
	U0C1_DX0F	I	St/B	USIC0 Channel 1 Shift Data Input
10	P8.2	O0 / I	St/B	Bit 2 of Port 8, General Purpose Input/Output
	CCU60_CC6 2	O1	St/B	CCU60 Channel 2 Output
	TxDC1	O2	St/B	CAN Node 1 Transmit Data Output
	U1C1_DOUT	O3	St/B	USIC1 Channel 1 Shift Data output
	CCU60_CC6 2INB	I	St/B	CCU60 Channel 2 Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
11	P7.1	O0 / I	St/B	Bit 1 of Port 7, General Purpose Input/Output
	EXTCLK	O1	St/B	Programmable Clock Signal Output
	TXDC4	O2	St/B	CAN Node 4 Transmit Data Output
	U4C1_DOUT	O3	St/B	USIC4 Channel 1 Shift Data Output
	U4C1_DX0D	I	St/B	USIC4 Channel 1 Shift Data Input
	CCU62_CTR APA	I	St/B	CCU62 Emergency Trap Input
	BRKIN_C	I	St/B	OCDS Break Signal Input
12	P7.4	O0 / I	St/B	Bit 4 of Port 7, General Purpose Input/Output
	EMUX2	O1	St/B	External Analog MUX Control Output 2 (ADC1)
	U0C1_DOUT	O2	St/B	USIC0 Channel 1 Shift Data Output
	U0C1_SCLK OUT	O3	St/B	USIC0 Channel 1 Shift Clock Output
	CCU62_CCP OS2A	I	St/B	CCU62 Position Input 2
	TCK_C	IH	St/B	DAP0/JTAG Clock Input If JTAG pos. C is selected during start-up, an internal pull-up device will hold this pin high when nothing is driving it. If DAP pos. 2 is selected during start-up, an internal pull-down device will hold this pin low when nothing is driving it.
	U0C0_DX0D	I	St/B	USIC0 Channel 0 Shift Data Input
U0C1_DX1E	I	St/B	USIC0 Channel 1 Shift Clock Input	
13	P8.1	O0 / I	St/B	Bit 1 of Port 8, General Purpose Input/Output
	CCU60_CC6 1	O1	St/B	CCU60 Channel 1 Output
	CC1_CC1	O2	St/B	CC1 Channel 1 Output
	U4C1_MCLK OUT	O3	St/B	USIC4 Channel 1 Master Clock Output
	CCU60_CC6 1INB	I	St/B	CCU60 Channel 1 Input
	RxDC1F	I	St/B	CAN Node 1 Receive Data Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
16	P6.0	O0 / I	DA/A	Bit 0 of Port 6, General Purpose Input/Output
	EMUX0	O1	DA/A	External Analog MUX Control Output 0 (ADC0)
	TxDC2	O2	DA/A	CAN Node 2 Transmit Data Output
	BRKOUT	O3	DA/A	OCDS Break Signal Output
	ADCx_REQG TyG	I	DA/A	External Request Gate Input for ADC0/1
	U1C1_DX0E	I	DA/A	USIC1 Channel 1 Shift Data Input
17	P6.1	O0 / I	DA/A	Bit 1 of Port 6, General Purpose Input/Output
	EMUX1	O1	DA/A	External Analog MUX Control Output 1 (ADC0)
	T3OUT	O2	DA/A	GPT12E Timer T3 Toggle Latch Output
	U1C1_DOUT	O3	DA/A	USIC1 Channel 1 Shift Data Output
	ADCx_REQT RyE	I	DA/A	External Request Trigger Input for ADC0/1
	RxDC2E	I	DA/A	CAN Node 2 Receive Data Input
18	ESR1_6	I	DA/A	ESR1 Trigger Input 6
	P6.2	O0 / I	DA/A	Bit 2 of Port 6, General Purpose Input/Output
	EMUX2	O1	DA/A	External Analog MUX Control Output 2 (ADC0)
	T6OUT	O2	DA/A	GPT12E Timer T6 Toggle Latch Output
	U1C1_SCLK OUT	O3	DA/A	USIC1 Channel 1 Shift Clock Output
	U1C1_DX1C	I	DA/A	USIC1 Channel 1 Shift Clock Input
19	P6.3	O0 / I	DA/A	Bit 3 of Port 6, General Purpose Input/Output
	T3OUT	O2	DA/A	GPT12E Timer T3 Toggle Latch Output
	U1C1_SELO 0	O3	DA/A	USIC1 Channel 1 Select/Control 0 Output
	U1C1_DX2D	I	DA/A	USIC1 Channel 1 Shift Control Input
	ADCx_REQT RyF	I	DA/A	External Request Trigger Input for ADC0/1
21	P15.0	I	In/A	Bit 0 of Port 15, General Purpose Input
	ADC1_CH0	I	In/A	Analog Input Channel 0 for ADC1
22	P15.1	I	In/A	Bit 1 of Port 15, General Purpose Input
	ADC1_CH1	I	In/A	Analog Input Channel 1 for ADC1

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
23	P15.2	I	In/A	Bit 2 of Port 15, General Purpose Input
	ADC1_CH2	I	In/A	Analog Input Channel 2 for ADC1
	T5INA	I	In/A	GPT12E Timer T5 Count/Gate Input
24	P15.3	I	In/A	Bit 3 of Port 15, General Purpose Input
	ADC1_CH3	I	In/A	Analog Input Channel 3 for ADC1
	T5EUDA	I	In/A	GPT12E Timer T5 External Up/Down Control Input
25	P15.4	I	In/A	Bit 4 of Port 15, General Purpose Input
	ADC1_CH4	I	In/A	Analog Input Channel 4 for ADC1
	T6INA	I	In/A	GPT12E Timer T6 Count/Gate Input
26	P15.5	I	In/A	Bit 5 of Port 15, General Purpose Input
	ADC1_CH5	I	In/A	Analog Input Channel 5 for ADC1
	T6EUDA	I	In/A	GPT12E Timer T6 External Up/Down Control Input
27	P15.6	I	In/A	Bit 6 of Port 15, General Purpose Input
	ADC1_CH6	I	In/A	Analog Input Channel 6 for ADC1
28	P15.7	I	In/A	Bit 7 of Port 15, General Purpose Input
	ADC1_CH7	I	In/A	Analog Input Channel 7 for ADC1
29	V_{AREF1}	-	PS/A	Reference Voltage for A/D Converter ADC1
30	V_{AREF0}	-	PS/A	Reference Voltage for A/D Converter ADC0
31	V_{AGND}	-	PS/A	Reference Ground for A/D Converters ADC0/1
32	P5.0	I	In/A	Bit 0 of Port 5, General Purpose Input
	ADC0_CH0	I	In/A	Analog Input Channel 0 for ADC0
33	P5.1	I	In/A	Bit 1 of Port 5, General Purpose Input
	ADC0_CH1	I	In/A	Analog Input Channel 1 for ADC0
34	P5.2	I	In/A	Bit 2 of Port 5, General Purpose Input
	ADC0_CH2	I	In/A	Analog Input Channel 2 for ADC0
	TDI_A	I	In/A	JTAG Test Data Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
35	P5.3	I	In/A	Bit 3 of Port 5, General Purpose Input
	ADC0_CH3	I	In/A	Analog Input Channel 3 for ADC0
	T3INA	I	In/A	GPT12E Timer T3 Count/Gate Input
39	P5.4	I	In/A	Bit 4 of Port 5, General Purpose Input
	ADC0_CH4	I	In/A	Analog Input Channel 4 for ADC0
	CCU63_T12 HRB	I	In/A	External Run Control Input for T12 of CCU63
	T3EUDA	I	In/A	GPT12E Timer T3 External Up/Down Control Input
	TMS_A	I	In/A	JTAG Test Mode Selection Input
40	P5.5	I	In/A	Bit 5 of Port 5, General Purpose Input
	ADC0_CH5	I	In/A	Analog Input Channel 5 for ADC0
	CCU60_T12 HRB	I	In/A	External Run Control Input for T12 of CCU60
41	P5.6	I	In/A	Bit 6 of Port 5, General Purpose Input
	ADC0_CH6	I	In/A	Analog Input Channel 6 for ADC0
42	P5.7	I	In/A	Bit 7 of Port 5, General Purpose Input
	ADC0_CH7	I	In/A	Analog Input Channel 7 for ADC0
43	P5.8	I	In/A	Bit 8 of Port 5, General Purpose Input
	ADC0_CH8	I	In/A	Analog Input Channel 8 for ADC0
	ADC1_CH8	I	In/A	Analog Input Channel 8 for ADC1
	CCU6x_T12H RC	I	In/A	External Run Control Input for T12 of CCU60/1/2/3
	CCU6x_T13H RC	I	In/A	External Run Control Input for T13 of CCU60/1/2/3
	U2C0_DX0F	I	In/A	USIC2 Channel 0 Shift Data Input
44	P5.9	I	In/A	Bit 9 of Port 5, General Purpose Input
	ADC0_CH9	I	In/A	Analog Input Channel 9 for ADC0
	ADC1_CH9	I	In/A	Analog Input Channel 9 for ADC1
	CC2_T7IN	I	In/A	CAPCOM2 Timer T7 Count Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
45	P5.10	I	In/A	Bit 10 of Port 5, General Purpose Input
	ADC0_CH10	I	In/A	Analog Input Channel 10 for ADC0
	ADC1_CH10	I	In/A	Analog Input Channel 10 for ADC1
	BRKIN_A	I	In/A	OCDS Break Signal Input
	U2C1_DX0F	I	In/A	USIC2 Channel 1 Shift Data Input
	CCU61_T13 HRA	I	In/A	External Run Control Input for T13 of CCU61
46	P5.11	I	In/A	Bit 11 of Port 5, General Purpose Input
	ADC0_CH11	I	In/A	Analog Input Channel 11 for ADC0
	ADC1_CH11	I	In/A	Analog Input Channel 11 for ADC1
47	P5.12	I	In/A	Bit 12 of Port 5, General Purpose Input
	ADC0_CH12	I	In/A	Analog Input Channel 12 for ADC0
48	P5.13	I	In/A	Bit 13 of Port 5, General Purpose Input
	ADC0_CH13	I	In/A	Analog Input Channel 13 for ADC0
	CCU63_T13 HRF	I	In/A	External Run Control Input for T13 of CCU63
49	P5.14	I	In/A	Bit 14 of Port 5, General Purpose Input
	ADC0_CH14	I	In/A	Analog Input Channel 14 for ADC0
	CC1_T0IN	I	St/B	CAPCOM1 Timer T7 Count Input
50	P5.15	I	In/A	Bit 15 of Port 5, General Purpose Input
	ADC0_CH15	I	In/A	Analog Input Channel 15 for ADC0
	RxDC2F	I	In/A	CAN Node 2 Receive Data Input
51	P2.12	O0 / I	St/B	Bit 12 of Port 2, General Purpose Input/Output
	U0C0_SELO 4	O1	St/B	USIC0 Channel 0 Select/Control 4 Output
	U0C1_SELO 3	O2	St/B	USIC0 Channel 1 Select/Control 3 Output
	TXDC2	O3	St/B	CAN Node 2 Transmit Data Output
	READY	IH	St/B	External Bus Interface READY Input
	U4C0_DX0A	I	St/B	USIC4 Channel 0 Shift Data Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
52	P2.11	O0 / I	St/B	Bit 11 of Port 2, General Purpose Input/Output
	U0C0_SELO 2	O1	St/B	USIC0 Channel 0 Select/Control 2 Output
	U0C1_SELO 2	O2	St/B	USIC0 Channel 1 Select/Control 2 Output
	U3C1_DOUT	O3	St/B	USIC3 Channel 1 Shift Data Output
	BHE/WRH	OH	St/B	External Bus Interf. High-Byte Control Output Can operate either as Byte High Enable (BHE) or as Write strobe for High Byte (WRH).
53	P11.5	O0 / I	St/B	Bit 5 of Port 11, General Purpose Input/Output
	CCU61_CC6 0	O1	St/B	CCU61 Channel 0 Output
	CCU61_COU T63	O2	St/B	CCU61 Channel 3 Output
	U3C1_SELO 1	O3	St/B	USIC3 Channel 1 Select/Control 1 Output
	CCU61_CC6 0INB	I	St/B	CCU61 Channel 0 Input
	U3C1_DX2B	I	St/B	USIC3 Channel 1 Shift Control Input
	U4C0_DX2A	I	St/B	USIC4 Channel 0 Shift Control Input
55	P2.0	O0 / I	St/B	Bit 0 of Port 2, General Purpose Input/Output
	TxDC5	O1	St/B	CAN Node 5 Transmit Data Output
	CCU63_CC6 0	O2	St/B	CCU63 Channel 0 Output
	AD13	OH / IH	St/B	External Bus Interface Address/Data Line 13
	RxDC0C	I	St/B	CAN Node 0 Receive Data Input
	CCU63_CC6 0INB	I	St/B	CCU63 Channel 0 Input
	T5INB	I	St/B	GPT12E Timer T5 Count/Gate Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
56	P2.1	O0 / I	St/B	Bit 1 of Port 2, General Purpose Input/Output
	TxDC0	O1	St/B	CAN Node 0 Transmit Data Output
	CCU63_CC6 1	O2	St/B	CCU63 Channel 1 Output
	AD14	OH / IH	St/B	External Bus Interface Address/Data Line 14
	RxDC5C	I	St/B	CAN Node 5 Receive Data Input
	CCU63_CC6 1INB	I	St/B	CCU63 Channel 1 Input
	T5EUDB	I	St/B	GPT12E Timer T5 External Up/Down Control Input
	ESR1_5	I	St/B	ESR1 Trigger Input 5
57	P11.4	O0 / I	St/B	Bit 4 of Port 11, General Purpose Input/Output
	CCU61_CC6 2	O1	St/B	CCU61 Channel 2 Output
	U3C1_DOUT	O2	St/B	USIC3 Channel 1 Shift Data Output
	RxDC5B	I	St/B	CAN Node 5 Receive Data Input
	CCU61_CC6 2INB	I	St/B	CCU61 Channel 2 Input
	U3C1_DX0B	I	St/B	USIC3 Channel 1 Shift Data Input
58	P2.2	O0 / I	St/B	Bit 2 of Port 2, General Purpose Input/Output
	TxDC1	O1	St/B	CAN Node 1 Transmit Data Output
	CCU63_CC6 2	O2	St/B	CCU63 Channel 2 Output
	AD15	OH / IH	St/B	External Bus Interface Address/Data Line 15
	CCU63_CC6 2INB	I	St/B	CCU63 Channel 2 Input
	ESR2_5	I	St/B	ESR2 Trigger Input 5

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
59	P11.3	O0 / I	St/B	Bit 3 of Port 11, General Purpose Input/Output
	CCU61_COU T63	O1	St/B	CCU61 Channel 3 Output
	CCU61_COU T62	O2	St/B	CCU61 Channel 2 Output
	TxDC5	O3	St/B	CAN Node 5 Transmit Data Input
	CCU61_T13 HRF	I	St/B	External Run Control Input for T13 of CCU61
	U4C0_DX1A	I	St/B	USIC4 Channel 0 Shift Clock Input
60	P4.0	O0 / I	St/B	Bit 0 of Port 4, General Purpose Input/Output
	CC2_CC24	O3 / I	St/B	CAPCOM2 CC24IO Capture Inp./ Compare Out.
	CS0	OH	St/B	External Bus Interface Chip Select 0 Output
61	P2.3	O0 / I	St/B	Bit 3 of Port 2, General Purpose Input/Output
	U0C0_DOUT	O1	St/B	USIC0 Channel 0 Shift Data Output
	CCU63_COU T63	O2	St/B	CCU63 Channel 3 Output
	CC2_CC16	O3 / I	St/B	CAPCOM2 CC16IO Capture Inp./ Compare Out.
	A16	OH	St/B	External Bus Interface Address Line 16
	ESR2_0	I	St/B	ESR2 Trigger Input 0
	U0C0_DX0E	I	St/B	USIC0 Channel 0 Shift Data Input
	U0C1_DX0D	I	St/B	USIC0 Channel 1 Shift Data Input
RxDC0A	I	St/B	CAN Node 0 Receive Data Input	
62	P11.2	O0 / I	St/B	Bit 2 of Port 11, General Purpose Input/Output
	CCU61_CC6 1	O1	St/B	CCU61 Channel 1 Output
	U3C1_DOUT	O2	St/B	USIC3 Channel 1 Shift Data Output
	U4C0_SELO 1	O3	St/B	USIC4 Channel 0 Select/Control 1 Output
	CCU63_CCP OS2A	I	St/B	CCU63 Position Input 2
	CCU61_CC6 1INB	I	St/B	CCU61 Channel 1 Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
63	P4.1	O0 / I	St/B	Bit 1 of Port 4, General Purpose Input/Output
	U3C0_SELO3	O1	St/B	USIC3 Channel Select/Control 3 Output
	TxDC2	O2	St/B	CAN Node 2 Transmit Data Output
	CC2_CC25	O3 / I	St/B	CAPCOM2 CC25IO Capture Inp./ Compare Out.
	CS1	OH	St/B	External Bus Interface Chip Select 1 Output
	CCU62_CCP OS0B	I	St/B	CCU62 Position Input 0
	T4EADB	I	St/B	GPT12E Timer T4 External Up/Down Control Input
ESR1_8	I	St/B	ESR1 Trigger Input 8	
64	P2.4	O0 / I	St/B	Bit 4 of Port 2, General Purpose Input/Output
	U0C1_DOUT	O1	St/B	USIC0 Channel 1 Shift Data Output
	TxDC0	O2	St/B	CAN Node 0 Transmit Data Output
	CC2_CC17	O3 / I	St/B	CAPCOM2 CC17IO Capture Inp./ Compare Out.
	A17	OH	St/B	External Bus Interface Address Line 17
	ESR1_0	I	St/B	ESR1 Trigger Input 0
	U0C0_DX0F	I	St/B	USIC0 Channel 0 Shift Data Input
RxDC1A	I	St/B	CAN Node 1 Receive Data Input	
65	P11.1	O0 / I	St/B	Bit 1 of Port 11, General Purpose Input/Output
	CCU61_COU T61	O1	St/B	CCU61 Channel 1 Output
	TxDC0	O2	St/B	CAN Node 0 Transmit Data Output
	U3C1_SELO0	O3	St/B	USIC3 Channel 1 Select/Control 0 Output
	CCU63_CCP OS1A	I	St/B	CCU63 Position Input 1
	CCU61_CTR APD	I	St/B	CCU61 Emergency Trap Input
	U3C1_DX2A	I	St/B	USIC3 Channel 1 Shift Control Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
66	P11.0	O0 / I	St/B	Bit 0 of Port 11, General Purpose Input/Output
	CCU61_COU T60	O1	St/B	CCU61 Channel 0 Output
	U3C1_SCLK OUT	O2	St/B	USIC3 Channel 1 Shift Clock Output
	U4C0_SCLK OUT	O3	St/B	USIC4 Channel 0 Shift Clock Output
	U4C0_DX1D	I	St/B	USIC4 Channel 0 Shift Clock Input
	CCU63_CCP OS0A	I	St/B	CCU63 Position Input 0
	RxDC0F	I	St/B	CAN Node 0 Receive Data Input
	U3C1_DX1A	I	St/B	USIC3 Channel 1 Shift Clock Input
	ESR1_7	I	St/B	ESR1 Trigger Input 7
67	P2.5	O0 / I	St/B	Bit 5 of Port 2, General Purpose Input/Output
	U0C0_SCLK OUT	O1	St/B	USIC0 Channel 0 Shift Clock Output
	TxDC0	O2	St/B	CAN Node 0 Transmit Data Output
	CC2_CC18	O3 / I	St/B	CAPCOM2 CC18IO Capture Inp./ Compare Out.
	A18	OH	St/B	External Bus Interface Address Line 18
	U0C0_DX1D	I	St/B	USIC0 Channel 0 Shift Clock Input
	ESR1_10	I	St/B	ESR1 Trigger Input 10
	U3C1_DX0D	I	St/B	USIC3 Channel 1 Shift Data Input
68	P4.2	O0 / I	St/B	Bit 2 of Port 4, General Purpose Input/Output
	U3C0_SCLK OUT	O1	St/B	USIC3 Channel 0 Shift Clock Output
	TxDC2	O2	St/B	CAN Node 2 Transmit Data Output
	CC2_CC26	O3 / I	St/B	CAPCOM2 CC26IO Capture Inp./ Compare Out.
	CS2	OH	St/B	External Bus Interface Chip Select 2 Output
	T2INA	I	St/B	GPT12E Timer T2 Count/Gate Input
	CCU62_CCP OS1B	I	St/B	CCU62 Position Input 1
	U3C0_DX1B	I	St/B	USIC3 Channel 0 Shift Clock Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
69	P2.6	O0 / I	St/B	Bit 6 of Port 2, General Purpose Input/Output
	U0C0_SELO0	O1	St/B	USIC0 Channel 0 Select/Control 0 Output
	U0C1_SELO1	O2	St/B	USIC0 Channel 1 Select/Control 1 Output
	CC2_CC19	O3 / I	St/B	CAPCOM2 CC19IO Capture Inp./ Compare Out.
	A19	OH	St/B	External Bus Interface Address Line 19
	U0C0_DX2D	I	St/B	USIC0 Channel 0 Shift Control Input
	RxDC0D	I	St/B	CAN Node 0 Receive Data Input
	ESR2_6	I	St/B	ESR2 Trigger Input 6
70	P4.4	O0 / I	St/B	Bit 4 of Port 4, General Purpose Input/Output
	U3C0_SELO2	O1	St/B	USIC3 Channel 0 Select/Control 2 Output
	CC2_CC28	O3 / I	St/B	CAPCOM2 CC28IO Capture Inp./ Compare Out.
	CS4	OH	St/B	External Bus Interface Chip Select 4 Output
	CLKIN2	I	St/B	Clock Signal Input 2
	U3C0_DX2C	I	St/B	USIC3 Channel 0 Shift Control Input
	71	P4.3	O0 / I	St/B
U0C1_DOUT		O1	St/B	USIC0 Channel 1 Shift Data Output
CC2_CC27		O3 / I	St/B	CAPCOM2 CC27IO Capture Inp./ Compare Out.
CS3		OH	St/B	External Bus Interface Chip Select 3 Output
RxDC2A		I	St/B	CAN Node 2 Receive Data Input
T2EUDA		I	St/B	GPT12E Timer T2 External Up/Down Control Input
CCU62_CCP OS2B		I	St/B	CCU62 Position Input 2

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
75	P0.0	O0 / I	St/B	Bit 0 of Port 0, General Purpose Input/Output
	U1C0_DOUT	O1	St/B	USIC1 Channel 0 Shift Data Output
	CCU61_CC60	O3	St/B	CCU61 Channel 0 IOutput
	A0	OH	St/B	External Bus Interface Address Line 0
	U1C0_DX0A	I	St/B	USIC1 Channel 0 Shift Data Input
	CCU61_CC60INA	I	St/B	CCU61 Channel 0 Input
	ESR1_11	I	St/B	ESR1 Trigger Input 11
76	P4.5	O0 / I	St/B	Bit 5 of Port 4, General Purpose Input/Output
	U3C0_DOUT	O1	St/B	USIC3 Channel 0 Shift Data Output
	CC2_CC29	O3 / I	St/B	CAPCOM2 CC29IO Capture Inp./Compare Out.
	CCU61_CCP0S0A	I	St/B	CCU61 Position Input 0
	U3C0_DX0B	I	St/B	USIC3 Channel 0 Shift Data Input
	ESR2_10	I	St/B	ESR2 Trigger Input 10
77	P4.6	O0 / I	St/B	Bit 6 of Port 4, General Purpose Input/Output
	U3C0_DOUT	O1	St/B	USIC3 Channel 0 Shift Data Output
	CC2_CC30	O3 / I	St/B	CAPCOM2 CC30IO Capture Inp./ Compare Out.
	T4INA	I	St/B	GPT12E Timer T4 Count/Gate Input
	CCU61_CCP0S1A	I	St/B	CCU61 Position Input 1

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
78	P2.7	O0 / I	St/B	Bit 7 of Port 2, General Purpose Input/Output
	U0C1_SELO0	O1	St/B	USIC0 Channel 1 Select/Control 0 Output
	U0C0_SELO1	O2	St/B	USIC0 Channel 0 Select/Control 1 Output
	CC2_CC20	O3 / I	St/B	CAPCOM2 CC20IO Capture Inp./ Compare Out.
	A20	OH	St/B	External Bus Interface Address Line 20
	U0C1_DX2C	I	St/B	USIC0 Channel 1 Shift Control Input
	RxDC1C	I	St/B	CAN Node 1 Receive Data Input
	ESR2_7	I	St/B	ESR2 Trigger Input 7
	U4C0_DX0D	I	St/B	USIC4 Channel 0 Shift Data Input
79	P0.1	O0 / I	St/B	Bit 1 of Port 0, General Purpose Input/Output
	U1C0_DOUT	O1	St/B	USIC1 Channel 0 Shift Data Output
	TxDC0	O2	St/B	CAN Node 0 Transmit Data Output
	CCU61_CC61	O3	St/B	CCU61 Channel 1 Output
	A1	OH	St/B	External Bus Interface Address Line 1
	U1C0_DX0B	I	St/B	USIC1 Channel 0 Shift Data Input
	CCU61_CC61INA	I	St/B	CCU61 Channel 1 Input
	U1C0_DX1A	I	St/B	USIC1 Channel 0 Shift Clock Input
80	P2.8	O0 / I	DP/B	Bit 8 of Port 2, General Purpose Input/Output
	U0C1_SCLKOUT	O1	DP/B	USIC0 Channel 1 Shift Clock Output
	EXTCLK	O2	DP/B	Programmable Clock Signal Output 1)
	CC2_CC21	O3 / I	DP/B	CAPCOM2 CC21IO Capture Inp./ Compare Out.
	A21	OH	DP/B	External Bus Interface Address Line 21
	U0C1_DX1D	I	DP/B	USIC0 Channel 1 Shift Clock Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
81	P4.7	O0 / I	St/B	Bit 7 of Port 4, General Purpose Input/Output
	CC2_CC31	O3 / I	St/B	CAPCOM2 CC31IO Capture Inp./ Compare Out.
	T4EUDA	I	St/B	GPT12E Timer T4 External Up/Down Control Input
	CCU61_CCP OS2A	I	St/B	CCU61 Position Input 2
82	P2.9	O0 / I	St/B	Bit 9 of Port 2, General Purpose Input/Output
	U0C1_DOUT	O1	St/B	USIC0 Channel 1 Shift Data Output
	TxDC1	O2	St/B	CAN Node 1 Transmit Data Output
	CC2_CC22	O3 / I	St/B	CAPCOM2 CC22IO Capture Inp./ Compare Out.
	A22	OH	St/B	External Bus Interface Address Line 22
	CLKIN1	I	St/B	Clock Signal Input 1
	TCK_A	IH	St/B	DAP0/JTAG Clock Input If JTAG pos. A is selected during start-up, an internal pull-up device will hold this pin high when nothing is driving it. If DAP pos. 0 is selected during start-up, an internal pull-down device will hold this pin low when nothing is driving it.
83	P0.2	O0 / I	St/B	Bit 2 of Port 0, General Purpose Input/Output
	U1C0_SCLK OUT	O1	St/B	USIC1 Channel 0 Shift Clock Output
	TxDC0	O2	St/B	CAN Node 0 Transmit Data Output
	CCU61_CC6 2	O3	St/B	CCU61 Channel 2 Output
	A2	OH	St/B	External Bus Interface Address Line 2
	U1C0_DX1B	I	St/B	USIC1 Channel 0 Shift Clock Input
	CCU61_CC6 2INA	I	St/B	CCU61 Channel 2 Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
84	P10.0	O0 / I	St/B	Bit 0 of Port 10, General Purpose Input/Output
	U0C1_DOUT	O1	St/B	USIC0 Channel 1 Shift Data Output
	CCU60_CC60	O2	St/B	CCU60 Channel 0 Output
	AD0	OH / IH	St/B	External Bus Interface Address/Data Line 0
	CCU60_CC60INA	I	St/B	CCU60 Channel 0 Input
	ESR1_2	I	St/B	ESR1 Trigger Input 2
	U0C0_DX0A	I	St/B	USIC0 Channel 0 Shift Data Input
	U0C1_DX0A	I	St/B	USIC0 Channel 1 Shift Data Input
85	P3.0	O0 / I	St/B	Bit 0 of Port 3, General Purpose Input/Output
	U2C0_DOUT	O1	St/B	USIC2 Channel 0 Shift Data Output
	ESR1_1	I	St/B	ESR1 Trigger Input 1
	U2C0_DX0A	I	St/B	USIC2 Channel 0 Shift Data Input
	RxDC3B	I	St/B	CAN Node 3 Receive Data Input
	U2C0_DX1A	I	St/B	USIC2 Channel 0 Shift Clock Input
86	P10.1	O0 / I	St/B	Bit 1 of Port 10, General Purpose Input/Output
	U0C0_DOUT	O1	St/B	USIC0 Channel 0 Shift Data Output
	CCU60_CC61	O2	St/B	CCU60 Channel 1 Output
	AD1	OH / IH	St/B	External Bus Interface Address/Data Line 1
	CCU60_CC61INA	I	St/B	CCU60 Channel 1 Input
	U0C0_DX1A	I	St/B	USIC0 Channel 0 Shift Clock Input
	U0C0_DX0B	I	St/B	USIC0 Channel 0 Shift Data Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
87	P0.3	O0 / I	St/B	Bit 3 of Port 0, General Purpose Input/Output
	U1C0_SELO0	O1	St/B	USIC1 Channel 0 Select/Control 0 Output
	U1C1_SELO1	O2	St/B	USIC1 Channel 1 Select/Control 1 Output
	CCU61_COUT60	O3	St/B	CCU61 Channel 0 Output
	A3	OH	St/B	External Bus Interface Address Line 3
	U1C0_DX2A	I	St/B	USIC1 Channel 0 Shift Control Input
	RxDC0B	I	St/B	CAN Node 0 Receive Data Input
88	P3.1	O0 / I	St/B	Bit 1 of Port 3, General Purpose Input/Output
	U2C0_DOUT	O1	St/B	USIC2 Channel 0 Shift Data Output
	TxDC3	O2	St/B	CAN Node 3 Transmit Data Output
	U2C0_DX0B	I	St/B	USIC2 Channel 0 Shift Data Input
89	P10.2	O0 / I	St/B	Bit 2 of Port 10, General Purpose Input/Output
	U0C0_SCLKOUT	O1	St/B	USIC0 Channel 0 Shift Clock Output
	CCU60_CC62	O2	St/B	CCU60 Channel 2 Output
	U3C0_SELO1	O3	St/B	USIC3 Channel 0 Select/Control 1 Output
	AD2	OH / IH	St/B	External Bus Interface Address/Data Line 2
	CCU60_CC62INA	I	St/B	CCU60 Channel 2 Input
	U0C0_DX1B	I	St/B	USIC0 Channel 0 Shift Clock Input
	U3C0_DX2B	I	St/B	USIC3 Channel 0 Shift Control Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
90	P0.4	O0 / I	St/B	Bit 4 of Port 0, General Purpose Input/Output
	U1C1_SELO0	O1	St/B	USIC1 Channel 1 Select/Control 0 Output
	U1C0_SELO1	O2	St/B	USIC1 Channel 0 Select/Control 1 Output
	CCU61_COUT61	O3	St/B	CCU61 Channel 1 Output
	A4	OH	St/B	External Bus Interface Address Line 4
	U1C1_DX2A	I	St/B	USIC1 Channel 1 Shift Control Input
	RxDC1B	I	St/B	CAN Node 1 Receive Data Input
	ESR2_8	I	St/B	ESR2 Trigger Input 8
92	P2.13	O0 / I	St/B	Bit 13 of Port 2, General Purpose Input/Output
	U2C1_SELO2	O1	St/B	USIC2 Channel 1 Select/Control 2 Output
	U4C0_DOUT	O3	St/B	USIC4 Channel 0 Shift Data Output
	U4C0_DX0E	I	St/B	USIC4 Channel 0 Shift Data Input
	RxDC2D	I	St/B	CAN Node 2 Receive Data Input
93	P3.2	O0 / I	St/B	Bit 2 of Port 3, General Purpose Input/Output
	U2C0_SCLKOUT	O1	St/B	USIC2 Channel 0 Shift Clock Output
	TxDC3	O2	St/B	CAN Node 3 Transmit Data Output
	U2C0_DX1B	I	St/B	USIC2 Channel 0 Shift Clock Input
94	P2.10	O0 / I	St/B	Bit 10 of Port 2, General Purpose Input/Output
	U0C1_DOUT	O1	St/B	USIC0 Channel 1 Shift Data Output
	U0C0_SELO3	O2	St/B	USIC0 Channel 0 Select/Control 3 Output
	CC2_CC23	O3 / I	St/B	CAPCOM2 CC23IO Capture Inp./ Compare Out.
	A23	OH	St/B	External Bus Interface Address Line 23
	U0C1_DX0E	I	St/B	USIC0 Channel 1 Shift Data Input
	CAPINA	I	St/B	GPT12E Register CAPREL Capture Input
	U3C1_DX0A	I	St/B	USIC3 Channel 1 Shift Data Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
95	P10.3	O0 / I	St/B	Bit 3 of Port 10, General Purpose Input/Output
	CCU60_COU T60	O2	St/B	CCU60 Channel 0 Output
	AD3	OH / IH	St/B	External Bus Interface Address/Data Line 3
	U0C0_DX2A	I	St/B	USIC0 Channel 0 Shift Control Input
	U0C1_DX2A	I	St/B	USIC0 Channel 1 Shift Control Input
	U3C0_DX0A	I	St/B	USIC3 Channel 0 Shift Data Input
96	P0.5	O0 / I	St/B	Bit 5 of Port 0, General Purpose Input/Output
	U1C1_SCLK OUT	O1	St/B	USIC1 Channel 1 Shift Clock Output
	U1C0_SELO 2	O2	St/B	USIC1 Channel 0 Select/Control 2 Output
	CCU61_COU T62	O3	St/B	CCU61 Channel 2 Output
	A5	OH	St/B	External Bus Interface Address Line 5
	U1C1_DX1A	I	St/B	USIC1 Channel 1 Shift Clock Input
	U1C0_DX1C	I	St/B	USIC1 Channel 0 Shift Clock Input
	RXDC3E	I	St/B	CAN Node 3 Receive Data Input
97	P3.3	O0 / I	St/B	Bit 3 of Port 3, General Purpose Input/Output
	U2C0_SELO 0	O1	St/B	USIC2 Channel 0 Select/Control 0 Output
	U2C1_SELO 1	O2	St/B	USIC2 Channel 1 Select/Control 1 Output
	U4C0_SELO 0	O3	St/B	USIC4 Channel 0 Select/Control 0 Output
	U2C0_DX2A	I	St/B	USIC2 Channel 0 Shift Control Input
	RxDC3A	I	St/B	CAN Node 3 Receive Data Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
98	P10.4	O0 / I	St/B	Bit 4 of Port 10, General Purpose Input/Output
	U0C0_SELO3	O1	St/B	USIC0 Channel 0 Select/Control 3 Output
	CCU60_COUT61	O2	St/B	CCU60 Channel 1 Output
	U3C0_DOUT	O3	St/B	USIC3 Channel 0 Shift Data Output
	AD4	OH / IH	St/B	External Bus Interface Address/Data Line 4
	U0C0_DX2B	I	St/B	USIC0 Channel 0 Shift Control Input
	U0C1_DX2B	I	St/B	USIC0 Channel 1 Shift Control Input
	ESR1_9	I	St/B	ESR1 Trigger Input 9
99	P3.4	O0 / I	St/B	Bit 4 of Port 3, General Purpose Input/Output
	U2C1_SELO0	O1	St/B	USIC2 Channel 1 Select/Control 0 Output
	U2C0_SELO1	O2	St/B	USIC2 Channel 0 Select/Control 1 Output
	U0C0_SELO4	O3	St/B	USIC0 Channel 0 Select/Control 4 Output
	U2C1_DX2A	I	St/B	USIC2 Channel 1 Shift Control Input
	RxDC4A	I	St/B	CAN Node 4 Receive Data Input
100	P10.5	O0 / I	St/B	Bit 5 of Port 10, General Purpose Input/Output
	U0C1_SCLKOUT	O1	St/B	USIC0 Channel 1 Shift Clock Output
	CCU60_COUT62	O2	St/B	CCU60 Channel 2 Output
	U2C0_DOUT	O3	St/B	USIC2 Channel 0 Shift Data Output
	AD5	OH / IH	St/B	External Bus Interface Address/Data Line 5
	U0C1_DX1B	I	St/B	USIC0 Channel 1 Shift Clock Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
101	P3.5	O0 / I	St/B	Bit 5 of Port 3, General Purpose Input/Output
	U2C1_SCLK OUT	O1	St/B	USIC2 Channel 1 Shift Clock Output
	U2C0_SELO 2	O2	St/B	USIC2 Channel 0 Select/Control 2 Output
	U0C0_SELO 5	O3	St/B	USIC0 Channel 0 Select/Control 5 Output
	U2C1_DX1A	I	St/B	USIC2 Channel 1 Shift Clock Input
102	P0.6	O0 / I	St/B	Bit 6 of Port 0, General Purpose Input/Output
	U1C1_DOUT	O1	St/B	USIC1 Channel 1 Shift Data Output
	TxDC1	O2	St/B	CAN Node 1 Transmit Data Output
	CCU61_COU T63	O3	St/B	CCU61 Channel 3 Output
	A6	OH	St/B	External Bus Interface Address Line 6
	U1C1_DX0A	I	St/B	USIC1 Channel 1 Shift Data Input
	CCU61_CTR APA	I	St/B	CCU61 Emergency Trap Input
	U1C1_DX1B	I	St/B	USIC1 Channel 1 Shift Clock Input
103	P10.6	O0 / I	St/B	Bit 6 of Port 10, General Purpose Input/Output
	U0C0_DOUT	O1	St/B	USIC0 Channel 0 Shift Data Output
	TxDC4	O2	St/B	CAN Node 4 Transmit Data Output
	U1C0_SELO 0	O3	St/B	USIC1 Channel 0 Select/Control 0 Output
	AD6	OH / IH	St/B	External Bus Interface Address/Data Line 6
	U0C0_DX0C	I	St/B	USIC0 Channel 0 Shift Data Input
	U1C0_DX2D	I	St/B	USIC1 Channel 0 Shift Control Input
	CCU60_CTR APA	I	St/B	CCU60 Emergency Trap Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
104	P3.6	O0 / I	St/B	Bit 6 of Port 3, General Purpose Input/Output
	U2C1_DOUT	O1	St/B	USIC2 Channel 1 Shift Data Output
	TxDC4	O2	St/B	CAN Node 4 Transmit Data Output
	U0C0_SELO 6	O3	St/B	USIC0 Channel 0 Select/Control 6 Output
	U2C1_DX0A	I	St/B	USIC2 Channel 1 Shift Data Input
	U2C1_DX1B	I	St/B	USIC2 Channel 1 Shift Clock Input
105	P10.7	O0 / I	St/B	Bit 7 of Port 10, General Purpose Input/Output
	U0C1_DOUT	O1	St/B	USIC0 Channel 1 Shift Data Output
	CCU60_COU T63	O2	St/B	CCU60 Channel 3 Output
	AD7	OH / IH	St/B	External Bus Interface Address/Data Line 7
	U0C1_DX0B	I	St/B	USIC0 Channel 1 Shift Data Input
	CCU60_CCP OS0A	I	St/B	CCU60 Position Input 0
	RxDC4C	I	St/B	CAN Node 4 Receive Data Input
	T4INB	I	St/B	GPT12E Timer T4 Count/Gate Input
106	P0.7	O0 / I	St/B	Bit 7 of Port 0, General Purpose Input/Output
	U1C1_DOUT	O1	St/B	USIC1 Channel 1 Shift Data Output
	U1C0_SELO 3	O2	St/B	USIC1 Channel 0 Select/Control 3 Output
	TxDC3	O3	St/B	CAN Node 3 Transmit Data Output
	A7	OH	St/B	External Bus Interface Address Line 7
	U1C1_DX0B	I	St/B	USIC1 Channel 1 Shift Data Input
	CCU61_CTR APB	I	St/B	CCU61 Emergency Trap Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
107	P3.7	O0 / I	St/B	Bit 7 of Port 3, General Purpose Input/Output
	U2C1_DOUT	O1	St/B	USIC2 Channel 1 Shift Data Output
	U2C0_SELO 3	O2	St/B	USIC2 Channel 0 Select/Control 3 Output
	U0C0_SELO 7	O3	St/B	USIC0 Channel 0 Select/Control 7 Output
	U2C1_DX0B	I	St/B	USIC2 Channel 1 Shift Data Input
111	P1.0	O0 / I	St/B	Bit 0 of Port 1, General Purpose Input/Output
	U1C0_MCLK OUT	O1	St/B	USIC1 Channel 0 Master Clock Output
	U1C0_SELO 4	O2	St/B	USIC1 Channel 0 Select/Control 4 Output
	A8	OH	St/B	External Bus Interface Address Line 8
	ESR1_3	I	St/B	ESR1 Trigger Input 3
	CCU62_CTR APB	I	St/B	CCU62 Emergency Trap Input
	T6INB	I	St/B	GPT12E Timer T6 Count/Gate Input
112	P9.0	O0 / I	St/B	Bit 0 of Port 9, General Purpose Input/Output
	CCU63_CC6 0	O1	St/B	CCU63 Channel 0 Output
	CC1_CC6	O2	St/B	CAPCOM1 CC6 Compare Output
	CCU63_CC6 0INA	I	St/B	CCU63 Channel 0 Input
	T6EUDB	I	St/B	GPT12E Timer T6 External Up/Down Control Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
113	P10.8	O0 / I	St/B	Bit 8 of Port 10, General Purpose Input/Output
	U0C0_MCLK OUT	O1	St/B	USIC0 Channel 0 Master Clock Output
	U0C1_SELO 0	O2	St/B	USIC0 Channel 1 Select/Control 0 Output
	U2C1_DOUT	O3	St/B	USIC2 Channel 1 Shift Data Output
	AD8	OH / IH	St/B	External Bus Interface Address/Data Line 8
	CCU60_CCP OS1A	I	St/B	CCU60 Position Input 1
	U0C0_DX1C	I	St/B	USIC0 Channel 0 Shift Clock Input
	BRKIN_B	I	St/B	OCDS Break Signal Input
	T3EUDB	I	St/B	GPT12E Timer T3 External Up/Down Control Input
	ERAY_RxDA	I	St/B	ERAY Receive Data Input Channel A
	ESR2_11	I	St/B	ESR2 Trigger Input 11
114	P9.1	O0 / I	DP/B	Bit 1 of Port 9, General Purpose Input/Output
	CCU63_CC6 1	O1	DP/B	CCU63 Channel 1 Output
	CC1_CC5	O2	DP/B	CAPCOM1 CC5 Compare Output
	ERAY_TxDB	O3	DP/B	ERAY Transmit Data Output Channel B
	CCU63_CC6 1INA	I	DP/B	CCU63 Channel 1 Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
115	P10.9	O0 / I	DP/B	Bit 9 of Port 10, General Purpose Input/Output
	U0C0_SELO 4	O1	DP/B	USIC0 Channel 0 Select/Control 4 Output
	U0C1_MCLK OUT	O2	DP/B	USIC0 Channel 1 Master Clock Output
	ERAY_TxDA	O3	DP/B	ERAY Transmit Data Output Channel A
	AD9	OH / IH	DP/B	External Bus Interface Address/Data Line 9
	CCU60_CCP OS2A	I	DP/B	CCU60 Position Input 2
	TCK_B	IH	DP/B	DAP0/JTAG Clock Input If JTAG pos. B is selected during start-up, an internal pull-up device will hold this pin high when nothing is driving it. If DAP pos. 1 is selected during start-up, an internal pull-down device will hold this pin low when nothing is driving it.
T3INB	I	DP/B	GPT12E Timer T3 Count/Gate Input	
116	P1.1	O0 / I	St/B	Bit 1 of Port 1, General Purpose Input/Output
	CCU62_COU T62	O1	St/B	CCU62 Channel 2 Output
	U1C0_SELO 5	O2	St/B	USIC1 Channel 0 Select/Control 5 Output
	U2C1_DOUT	O3	St/B	USIC2 Channel 1 Shift Data Output
	A9	OH	St/B	External Bus Interface Address Line 9
	ESR2_3	I	St/B	ESR2 Trigger Input 3
	U2C1_DX0C	I	St/B	USIC2 Channel 1 Shift Data Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
117	P10.10	O0 / I	St/B	Bit 10 of Port 10, General Purpose Input/Output
	U0C0_SELO0	O1	St/B	USIC0 Channel 0 Select/Control 0 Output
	CCU60_COUT63	O2	St/B	CCU60 Channel 3 Output
	ERAY_TxENA	O3	St/B	ERAY Transmit Enable Output Channel A
	AD10	OH / IH	St/B	External Bus Interface Address/Data Line 10
	U0C0_DX2C	I	St/B	USIC0 Channel 0 Shift Control Input
	U0C1_DX1A	I	St/B	USIC0 Channel 1 Shift Clock Input
	TDI_B	IH	St/B	JTAG Test Data Input If JTAG pos. B is selected during start-up, an internal pull-up device will hold this pin high when nothing is driving it.
118	P10.11	O0 / I	St/B	Bit 11 of Port 10, General Purpose Input/Output
	U1C0_SCLKOUT	O1	St/B	USIC1 Channel 0 Shift Clock Output
	BRKOUT	O2	St/B	OCDS Break Signal Output
	U3C0_SELO0	O3	St/B	USIC3 Channel 0 Select/Control 0 Output
	AD11	OH / IH	St/B	External Bus Interface Address/Data Line 11
	U1C0_DX1D	I	St/B	USIC1 Channel 0 Shift Clock Input
	RxDC2B	I	St/B	CAN Node 2 Receive Data Input
	TMS_B	IH	St/B	JTAG Test Mode Selection Input If JTAG pos. B is selected during start-up, an internal pull-up device will hold this pin high when nothing is driving it.
	U3C0_DX2A	I	St/B	USIC3 Channel 0 Shift Control Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
119	P9.2	O0 / I	St/B	Bit 2 of Port 9, General Purpose Input/Output
	CCU63_CC6 2	O1	St/B	CCU63 Channel 2 Output
	CC1_CC4	O2	St/B	CAPCOM1 CC4 Compare Output
	ERAY_TxEN B	O3	St/B	ERAY Transmit Enable Output Channel B
	CCU63_CC6 2INA	I	St/B	CCU63 Channel 2 Input
	CAPINB	I	St/B	GPT12E Register CAPREL Capture Input
120	P1.2	O0 / I	St/B	Bit 2 of Port 1, General Purpose Input/Output
	CCU62_CC6 2	O1	St/B	CCU62 Channel 2 Output
	U1C0_SELO 6	O2	St/B	USIC1 Channel 0 Select/Control 6 Output
	U2C1_SCLK OUT	O3	St/B	USIC2 Channel 1 Shift Clock Output
	A10	OH	St/B	External Bus Interface Address Line 10
	ESR1_4	I	St/B	ESR1 Trigger Input 4
	CCU61_T12 HRB	I	St/B	External Run Control Input for T12 of CCU61
	CCU62_CC6 2INA	I	St/B	CCU62 Channel 2 Input
	U2C1_DX0D	I	St/B	USIC2 Channel 1 Shift Data Input
	U2C1_DX1C	I	St/B	USIC2 Channel 1 Shift Clock Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
121	P10.12	O0 / I	St/B	Bit 12 of Port 10, General Purpose Input/Output
	U1C0_DOUT	O1	St/B	USIC1 Channel 0 Shift Data Output
	TxDC2	O2	St/B	CAN Node 2 Transmit Data Output
	TDO_B	OH / IH	St/B	JTAG Test Data Output / DAP1 Input/Output If DAP pos. 1 is selected during start-up, an internal pull-down device will hold this pin low when nothing is driving it.
	AD12	OH / IH	St/B	External Bus Interface Address/Data Line 12
	U1C0_DX0C	I	St/B	USIC1 Channel 0 Shift Data Input
	U1C0_DX1E	I	St/B	USIC1 Channel 0 Shift Clock Input
122	P9.3	O0 / I	St/B	Bit 3 of Port 9, General Purpose Input/Output
	CCU63_COU T60	O1	St/B	CCU63 Channel 0 Output
	$\overline{\text{BRKOUT}}$	O2	St/B	OCDS Break Signal Output
123	P10.13	O0 / I	St/B	Bit 13 of Port 10, General Purpose Input/Output
	U1C0_DOUT	O1	St/B	USIC1 Channel 0 Shift Data Output
	TxDC3	O2	St/B	CAN Node 3 Transmit Data Output
	U1C0_SELO 3	O3	St/B	USIC1 Channel 0 Select/Control 3 Output
	$\overline{\text{WR/WRL}}$	OH	St/B	External Bus Interface Write Strobe Output Active for each external write access, when $\overline{\text{WR}}$, active for ext. writes to the low byte, when $\overline{\text{WRL}}$.
	U1C0_DX0D	I	St/B	USIC1 Channel 0 Shift Data Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
124	P1.3	O0 / I	St/B	Bit 3 of Port 1, General Purpose Input/Output
	CCU62_COU T63	O1	St/B	CCU62 Channel 3 Output
	U1C0_SELO 7	O2	St/B	USIC1 Channel 0 Select/Control 7 Output
	U2C0_SELO 4	O3	St/B	USIC2 Channel 0 Select/Control 4 Output
	A11	OH	St/B	External Bus Interface Address Line 11
	ESR2_4	I	St/B	ESR2 Trigger Input 4
	CCU62_T12 HRB	I	St/B	External Run Control Input for T12 of CCU62
125	P9.4	O0 / I	St/B	Bit 4 of Port 9, General Purpose Input/Output
	CCU63_COU T61	O1	St/B	CCU63 Channel 1 Output
	U2C0_DOUT	O2	St/B	USIC2 Channel 0 Shift Data Output
	CCU62_COU T63	O3	St/B	CCU62 Channel 3 Output
126	P9.5	O0 / I	St/B	Bit 5 of Port 9, General Purpose Input/Output
	CCU63_COU T62	O1	St/B	CCU63 Channel 2 Output
	U2C0_DOUT	O2	St/B	USIC2 Channel 0 Shift Data Output
	CCU62_COU T62	O3	St/B	CCU62 Channel 2 Output
	U2C0_DX0E	I	St/B	USIC2 Channel 0 Shift Data Input
	CCU60_CCP OS2B	I	St/B	CCU60 Position Input 2

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
128	P10.14	O0 / I	St/B	Bit 14 of Port 10, General Purpose Input/Output
	U1C0_SELO 1	O1	St/B	USIC1 Channel 0 Select/Control 1 Output
	U0C1_DOUT	O2	St/B	USIC0 Channel 1 Shift Data Output
	U3C0_SCLK OUT	O3	St/B	USIC3 Channel 0 Shift Clock Output
	$\overline{\text{RD}}$	OH	St/B	External Bus Interface Read Strobe Output
	ESR2_2	I	St/B	ESR2 Trigger Input 2
	U0C1_DX0C	I	St/B	USIC0 Channel 1 Shift Data Input
	RxDC3C	I	St/B	CAN Node 3 Receive Data Input
	U3C0_DX1A	I	St/B	USIC3 Channel 0 Shift Clock Input
129	P1.4	O0 / I	St/B	Bit 4 of Port 1, General Purpose Input/Output
	CCU62_COU T61	O1	St/B	CCU62 Channel 1 Output
	U1C1_SELO 4	O2	St/B	USIC1 Channel 1 Select/Control 4 Output
	U2C0_SELO 5	O3	St/B	USIC2 Channel 0 Select/Control 5 Output
	A12	OH	St/B	External Bus Interface Address Line 12
	U2C0_DX2B	I	St/B	USIC2 Channel 0 Shift Control Input
	RxDC5A	I	St/B	CAN Node 5 Receive Data Input
130	P10.15	O0 / I	St/B	Bit 15 of Port 10, General Purpose Input/Output
	U1C0_SELO 2	O1	St/B	USIC1 Channel 0 Select/Control 2 Output
	U0C1_DOUT	O2	St/B	USIC0 Channel 1 Shift Data Output
	U1C0_DOUT	O3	St/B	USIC1 Channel 0 Shift Data Output
	ALE	OH	St/B	External Bus Interf. Addr. Latch Enable Output
	U0C1_DX1C	I	St/B	USIC0 Channel 1 Shift Clock Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
131	P1.5	O0 / I	St/B	Bit 5 of Port 1, General Purpose Input/Output
	CCU62_COU T60	O1	St/B	CCU62 Channel 0 Output
	U1C1_SELO 3	O2	St/B	USIC1 Channel 1 Select/Control 3 Output
	BRKOUT	O3	St/B	OCDS Break Signal Output
	A13	OH	St/B	External Bus Interface Address Line 13
	U2C0_DX0C	I	St/B	USIC2 Channel 0 Shift Data Input
132	P9.6	O0 / I	St/B	Bit 6 of Port 9, General Purpose Input/Output
	CCU63_COU T63	O1	St/B	CCU63 Channel 3 Output
	CCU63_COU T62	O2	St/B	CCU63 Channel 2 Output
	CCU62_COU T61	O3	St/B	CCU62 Channel 1 Output
	CCU63 _CTRAPA	I	St/B	CCU63 Emergency Trap Input
	CCU60_CCP OS1B	I	St/B	CCU60 Position Input 1
	ERAY_RxDB	I	St/B	ERAY Receive Data Input Channel B
133	P1.6	O0 / I	St/B	Bit 6 of Port 1, General Purpose Input/Output
	CCU62_CC6 1	O1 / I	St/B	CCU62 Channel 1 Output
	U1C1_SELO 2	O2	St/B	USIC1 Channel 1 Select/Control 2 Output
	U2C0_DOUT	O3	St/B	USIC2 Channel 0 Shift Data Output
	A14	OH	St/B	External Bus Interface Address Line 14
	U2C0_DX0D	I	St/B	USIC2 Channel 0 Shift Data Input
	CCU62_CC6 1INA	I	St/B	CCU62 Channel 1 Input
	U4C1_DX0A	I	St/B	USIC4 Channel 1 Shift Data Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
134	P9.7	O0 / I	St/B	Bit 7 of Port 9, General Purpose Input/Output
	CCU62_COU T60	O1	St/B	CCU62 Channel 0 Output
	CCU62_COU T63	O2	St/B	CCU62 Channel 3 Output
	CCU63_CTR APB	I	St/B	CCU63 Emergency Trap Input
	U2C0_DX1D	I	St/B	USIC2 Channel 0 Shift Clock Input
	CCU60_CCP OS0B	I	St/B	CCU60 Position Input 0
135	P1.7	O0 / I	St/B	Bit 7 of Port 1, General Purpose Input/Output
	CCU62_CC6 0	O1	St/B	CCU62 Channel 0 Output
	U1C1_MCLK OUT	O2	St/B	USIC1 Channel 1 Master Clock Output
	U2C0_SCLK OUT	O3	St/B	USIC2 Channel 0 Shift Clock Output
	A15	OH	St/B	External Bus Interface Address Line 15
	U2C0_DX1C	I	St/B	USIC2 Channel 0 Shift Clock Input
	CCU62_CC6 0INA	I	St/B	CCU62 Channel 0 Input
	RxDC4E	I	St/B	CAN Node 4 Receive Data Input
136	XTAL2	O	Sp/M	Crystal Oscillator Amplifier Output
137	XTAL1	I	Sp/M	Crystal Oscillator Amplifier Input To clock the device from an external source, drive XTAL1, while leaving XTAL2 unconnected. Voltages on XTAL1 must comply to the core supply voltage V_{DDIM} .
	ESR2_9	I	St/B	ESR2 Trigger Input 9

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
138	PORST	I	In/B	Power On Reset Input A low level at this pin resets the XC228xH completely. A spike filter suppresses input pulses <10 ns. Input pulses >100 ns safely pass the filter. The minimum duration for a safe recognition should be 120 ns. An internal pull-up device will hold this pin high when nothing is driving it.
139	ESR1	O0 / I	St/B	External Service Request 1 After power-up, an internal weak pull-up device holds this pin high when nothing is driving it.
	RxDC0E	I	St/B	CAN Node 0 Receive Data Input
	U1C0_DX0F	I	St/B	USIC1 Channel 0 Shift Data Input
	U1C0_DX2C	I	St/B	USIC1 Channel 0 Shift Control Input
	U1C1_DX0C	I	St/B	USIC1 Channel 1 Shift Data Input
	U1C1_DX2B	I	St/B	USIC1 Channel 1 Shift Control Input
	U2C1_DX2C	I	St/B	USIC2 Channel 1 Shift Control Input
140	ESR2	O0 / I	St/B	External Service Request 2 After power-up, an internal weak pull-up device holds this pin high when nothing is driving it.
	RxDC1E	I	St/B	CAN Node 1 Receive Data Input
	CCU60_CTR APC	I	St/B	CCU60 Emergency Trap Input
	CCU61_CTR APC	I	St/B	CCU61 Emergency Trap Input
	CCU62_CTR APC	I	St/B	CCU62 Emergency Trap Input
	CCU63_CTR APC	I	St/B	CCU63 Emergency Trap Input
	U1C1_DX0D	I	St/B	USIC1 Channel 1 Shift Data Input
	U1C1_DX2C	I	St/B	USIC1 Channel 1 Shift Control Input
	U2C1_DX0E	I	St/B	USIC2 Channel 1 Shift Data Input
	U2C1_DX2B	I	St/B	USIC2 Channel 1 Shift Control Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
141	ESR0	O0 / I	St/B	External Service Request 0 After power-up, ESR0 operates as open-drain bidirectional reset with a weak pull-up.
	U1C0_DX0E	I	St/B	USIC1 Channel 0 Shift Data Input
	U1C0_DX2B	I	St/B	USIC1 Channel 0 Shift Control Input
142	P8.6	O0 / I	St/B	Bit 6 of Port 8, General Purpose Input/Output
	CCU60_COU T63	O1	St/B	CCU60 Channel 3 Output
	CC1_CC3	O2	St/B	CAPCOM1 CC3 Compare Output
	MCHK_MAT CH	O3	St/B	Memory Checker Match Output
	CCU60_CTR APB	I	St/B	CCU60 Emergency Trap Input
	BRKIN_D	I	St/B	OCDS Break Signal Input
	CCU62_CTR APD	I	St/B	CCU62 Emergency Trap Input
	U4C1_DX2A	I	St/B	USIC4 Channel 1 Shift Control Input
143	P8.5	O0 / I	St/B	Bit 5 of Port 8, General Purpose Input/Output
	CCU60_COU T62	O1	St/B	CCU60 Channel 2 Output
	CCU62_CC6 2	O2	St/B	CCU62 Channel 2 Output
	U4C1_SCLK OUT	O3	St/B	USIC4 Channel 1 Shift Clock Output
	U4C1_DX1C	I	St/B	USIC4 Channel 1 Shift Clock Input
	TCK_D	IH	St/B	DAP0/JTAG Clock Input If JTAG pos. D is selected during start-up, an internal pull-up device will hold this pin high when nothing is driving it.
	CCU62_CC6 2INB	I	St/B	CCU62 Channel 2 Input
15	V _{DDIM}	-	PS/M	Digital Core Supply Voltage for Domain M Decouple with a ceramic capacitor, see Data Sheet for details.

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
14, 54, 91, 127	V_{DD1}	-	PS/1	Digital Core Supply Voltage for Domain 1 Decouple with a ceramic capacitor, see Data Sheet for details. All V_{DD1} pins must be connected to each other.
20	V_{DDPA}	-	PS/A	Digital Pad Supply Voltage for Domain A Connect decoupling capacitors to adjacent V_{DDP}/V_{SS} pin pairs as close as possible to the pins. <i>Note: The A/D_Converters and ports P5, P6 and P15 are fed from supply voltage V_{DDPA}.</i>
2, 36, 38, 72, 74, 108, 110, 144	V_{DDPB}	-	PS/B	Digital Pad Supply Voltage for Domain B Connect decoupling capacitors to adjacent V_{DDP}/V_{SS} pin pairs as close as possible to the pins. <i>Note: The on-chip voltage regulators and all ports except P5, P6 and P15 are fed from supply voltage V_{DDPB}.</i>
1, 37, 73, 109	V_{SS}	-	PS/--	Digital Ground All V_{SS} pins must be connected to the ground-line or ground-plane. <i>Note: Also the exposed pad is connected internally to V_{SS}. To improve the EMC behavior, it is recommended to connect the exposed pad to the board ground. For thermal aspects, please refer to the Data Sheet. Board layout examples are given in an application note.</i>

- 1) To generate the reference clock output for bus timing measurement, f_{SYS} must be selected as source for EXTCLK and P2.8 must be selected as output pin. Also the high-speed clock pad must be enabled. This configuration is referred to as reference clock output signal CLKOUT.

2.2 Identification Registers

The identification registers describe the current version of the XC228xH and of its modules.

Table 7 XC228xH Identification Registers

Short Name	Value	Address	Notes
SCU_IDMANUF	1820 _H	00'F07E _H	
SCU_IDCHIP	4001 _H	00'F07C _H	marking EES-AA or ES-AA
	4002 _H	00'F07C _H	marking ES+AA, ES-AB or AB
SCU_IDMEM	318F _H	00'F07A _H	
SCU_IDPROG	1313 _H	00'F078 _H	
JTAG_ID	0018'A083 _H	---	marking EES-AA or ES-AA
	1018'A083 _H	---	marking ES+AA
	2018'A083 _H	---	marking ES-AB or AB

3 Functional Description

The architecture of the XC228xH combines advantages of RISC, CISC, and DSP processors with an advanced peripheral subsystem in a well-balanced design. On-chip memory blocks allow the design of compact systems-on-silicon with maximum performance suited for computing, control, and communication.

The on-chip memory blocks (program code memory and SRAM, dual-port RAM, data SRAM) and the generic peripherals are connected to the CPU by separate high-speed buses. Another bus, the LXBus, connects additional on-chip resources and external resources. This bus structure enhances overall system performance by enabling the concurrent operation of several subsystems of the XC228xH.

The block diagram gives an overview of the on-chip components and the advanced internal bus structure of the XC228xH.

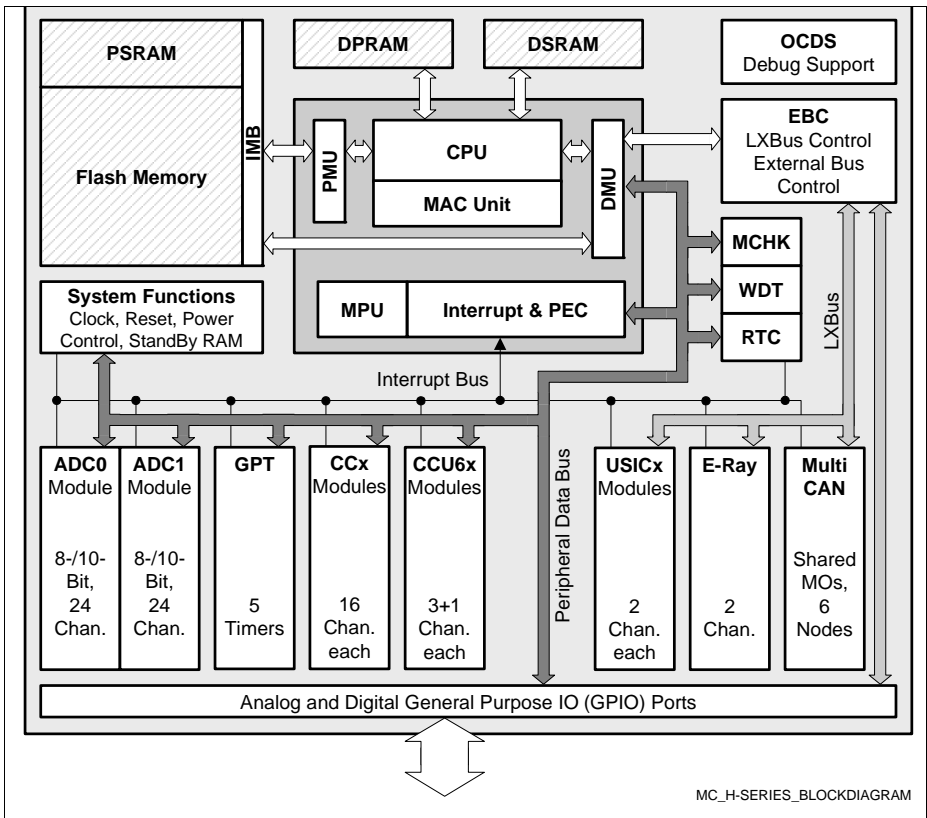


Figure 4 Block Diagram

3.1 Memory Subsystem and Organization

The memory space of the XC228xH is configured in the von Neumann architecture. In this architecture all internal and external resources, including code memory, data memory, registers and I/O ports, are organized in the same linear address space.

Table 8 XC228xH Memory Map ¹⁾

Address Area	Start Loc.	End Loc.	Area Size ²⁾	Notes
IMB register space	FF'FF00 _H	FF'FFFF _H	256 Bytes	
Reserved	F0'0000 _H	FF'FEFF _H	< 1 Mbyte	Minus IMB registers
Reserved for EPSRAM	E9'C000 _H	EF'FFFF _H	400 Kbytes	Mirrors EPSRAM
Emulated PSRAM	E8'0000 _H	E9'BFFF _H	up to 112 Kbytes	With Flash timing
Reserved for PSRAM	E1'C000 _H	E7'FFFF _H	400 Kbytes	Mirrors PSRAM
PSRAM	E0'0000 _H	E1'BFFF _H	up to 112 Kbytes	Program SRAM
Reserved for Flash	D9'0000 _H	DF'FFFF _H	448 Kbytes	
Flash 6	D8'0000 _H	D8'FFFF _H	64 Kbytes	
Flash 5	D4'0000 _H	D7'FFFF _H	256 Kbytes	
Flash 4	D0'0000 _H	D3'FFFF _H	256 Kbytes	
Flash 3	CC'0000 _H	CF'FFFF _H	256 Kbytes	
Flash 2	C8'0000 _H	CB'FFFF _H	256 Kbytes	
Flash 1	C4'0000 _H	C7'FFFF _H	256 Kbytes	
Flash 0	C0'0000 _H	C3'FFFF _H	256 Kbytes ³⁾	Minus res. seg.
External memory area	40'0000 _H	BF'FFFF _H	8 Mbytes	
External IO area ⁴⁾	21'0000 _H	3F'FFFF _H	1,984 Kbytes	
Reserved	20'C000 _H	20'FFFF _H	16 Kbytes	
USIC0–3 alternate regs.	20'B000 _H	20'BFFF _H	4 Kbytes	Accessed via EBC
MultiCAN alternate regs.	20'8000 _H	20'AFFF _H	12 Kbytes	Accessed via EBC
FlexRay registers	20'7000 _H	20'7FFF _H	4 Kbytes	Accessed via EBC
Reserved	20'6800 _H	20'6FFF _H	2 Kbytes	
USIC0–4 registers	20'4000 _H	20'67FF _H	10 Kbytes	Accessed via EBC
MultiCAN registers	20'0000 _H	20'3FFF _H	16 Kbytes	Accessed via EBC
External memory area	01'0000 _H	1F'FFFF _H	1984 Kbytes	
SFR area	00'FE00 _H	00'FFFF _H	0.5 Kbytes	

Table 8 XC228xH Memory Map (cont'd)¹⁾

Address Area	Start Loc.	End Loc.	Area Size²⁾	Notes
Dualport RAM (DPRAM)	00'F600 _H	00'FDFF _H	2 Kbytes	
Reserved for DPRAM	00'F200 _H	00'F5FF _H	1 Kbytes	
ESFR area	00'F000 _H	00'F1FF _H	0.5 Kbytes	
XSFR area	00'E000 _H	00'EFFF _H	4 Kbytes	
Data SRAM (DSRAM)	00'8000 _H	00'DFFF _H	24 Kbytes	
External memory area	00'0000 _H	00'7FFF _H	32 Kbytes	

- 1) Accesses to the shaded areas are reserved. In devices with external bus interface these accesses generate external bus accesses.
- 2) The areas marked with “<” are slightly smaller than indicated, see column “Notes”.
- 3) The uppermost 4-Kbyte sector of the first Flash segment is reserved for internal use (C0'F000_H to C0'FFFF_H).
- 4) Several pipeline optimizations are not active within the external IO area. This is necessary to control external peripherals properly.

This common memory space consists of 16 Mbytes organized as 256 segments of 64 Kbytes; each segment contains four data pages of 16 Kbytes. The entire memory space can be accessed bitwise or wordwise. Portions of the on-chip DPRAM and the register spaces (ESFR/SFR) additionally are directly bit addressable.

The internal data memory areas and the Special Function Register areas (SFR and ESFR) are mapped into segment 0, the system segment.

The Program Management Unit (PMU) handles all code fetches and, therefore, controls access to the program memories such as Flash memory and PSRAM.

The Data Management Unit (DMU) handles all data transfers and, therefore, controls access to the DSRAM and the on-chip peripherals.

Both units (PMU and DMU) are connected to the high-speed system bus so that they can exchange data. This is required if operands are read from program memory, code or data is written to the PSRAM, code is fetched from external memory, or data is read from or written to external resources. These include peripherals on the LXBus such as USIC or MultiCAN. The system bus allows concurrent two-way communication for maximum transfer performance.

Up to 112 Kbytes of on-chip Program SRAM (PSRAM) are provided to store user code or data. The PSRAM is accessed via the PMU and is optimized for code fetches. A section of the PSRAM with programmable size can be write-protected.

Note: The actual size of the PSRAM depends on the quoted device type.

Functional Description

Up to 24 Kbytes of on-chip Data SRAM (DSRAM) are used for storage of general user data. The DSRAM is accessed via a separate interface and is optimized for data access.

2 Kbytes of on-chip Dual-Port RAM (DPRAM) provide storage for user-defined variables, for the system stack, and for general purpose register banks. A register bank can consist of up to 16 word-wide (R0 to R15) and/or byte-wide (RL0, RH0, ..., RL7, RH7) General Purpose Registers (GPRs).

The upper 256 bytes of the DPRAM are directly bit addressable. When used by a GPR, any location in the DPRAM is bit addressable.

8 Kbytes of on-chip Stand-By SRAM (SBRAM) provide storage for system-relevant user data that must be preserved while the major part of the device is powered down. The SBRAM is accessed via a specific interface and is powered in domain M.

1024 bytes (2 × 512 bytes) of the address space are reserved for the Special Function Register areas (SFR space and ESFR space). SFRs are word-wide registers which are used to control and monitor functions of the different on-chip units. Unused SFR addresses are reserved for future members of the XC2000 Family. In order to ensure upward compatibility they should either not be accessed or written with zeros.

In order to meet the requirements of designs where more memory is required than is available on chip, up to 12 Mbytes (approximately, see [Table 8](#)) of external RAM and/or ROM can be connected to the microcontroller. The External Bus Interface also provides access to external peripherals.

The on-chip Flash memory stores code, constant data, and control data. The 1,600 Kbytes of on-chip Flash memory consist of 1 module of 64 Kbytes (preferably for data storage) and 6 modules of 256 Kbytes. Each module is organized in 4-Kbyte sectors.

The uppermost 4-Kbyte sector of segment 0 (located in Flash module 0) is used internally to store operation control parameters and protection information.

Note: The actual size of the Flash memory depends on the chosen device type.

Each sector can be separately write protected¹⁾, erased and programmed (in blocks of 128 Bytes). The complete Flash area can be read-protected. A user-defined password sequence temporarily unlocks protected areas. The Flash modules combine 128-bit read access with protected and efficient writing algorithms for programming and erasing. Dynamic error correction provides extremely high read data security for all read access operations. Access to different Flash modules can be executed in parallel.

For Flash parameters, please see [Section 4.6](#).

1) To save control bits, sectors are clustered for protection purposes, they remain separate for programming/erasing.

Memory Content Protection

The contents of on-chip memories can be protected against soft errors (induced e.g. by radiation) by activating the parity mechanism or the Error Correction Code (ECC).

The parity mechanism can detect a single-bit error and prevent the software from using incorrect data or executing incorrect instructions.

The ECC mechanism can detect and automatically correct single-bit errors. This supports the stable operation of the system.

It is strongly recommended to activate the ECC mechanism wherever possible because this dramatically increases the robustness of an application against such soft errors.

3.2 External Bus Controller

All external memory access operations are performed by a special on-chip External Bus Controller (EBC). The EBC also controls access to resources connected to the on-chip LXBus (MultiCAN and the USIC modules). The LXBus is an internal representation of the external bus that allows access to integrated peripherals and modules in the same way as to external components.

The EBC can be programmed either to Single Chip Mode, when no external memory is required, or to an external bus mode with the following selections¹⁾:

- Address Bus Width with a range of 0 ... 24-bit
- Data Bus Width 8-bit or 16-bit
- Bus Operation Multiplexed or Demultiplexed

The bus interface uses Port 10 and Port 2 for addresses and data. In the demultiplexed bus modes, the lower addresses are output separately on Port 0 and Port 1. The number of active segment address lines is selectable, restricting the external address space to 8 Mbytes ... 64 Kbytes. This is required when interface lines shall be assigned to Port 2.

External $\overline{\text{CS}}$ signals (address windows plus default) can be generated and output on Port 4 in order to save external glue logic. External modules can be directly connected to the common address/data bus and their individual select lines.

Important timing characteristics of the external bus interface are programmable (with registers TCONCSx/FCONCSx) to allow the user to adapt it to a wide range of different types of memories and external peripherals.

Access to very slow memories or modules with varying access times is supported by a special 'Ready' function. The active level of the control input signal is selectable.

In addition, up to four independent address windows may be defined (using registers ADDRSELx) to control access to resources with different bus characteristics. These address windows are arranged hierarchically where window 4 overrides window 3, and window 2 overrides window 1. All accesses to locations not covered by these four address windows are controlled by TCONCS0/FCONCS0. The currently active window can generate a chip select signal.

The external bus timing is based on the rising edge of the reference clock output CLKOUT. The external bus protocol is compatible with that of the standard C166 Family.

1) Bus modes are switched dynamically if several address windows with different mode settings are used.

3.3 Central Processing Unit (CPU)

The core of the CPU consists of a 5-stage execution pipeline with a 2-stage instruction-fetch pipeline, a 16-bit arithmetic and logic unit (ALU), a 32-bit/40-bit multiply and accumulate unit (MAC), a register-file providing three register banks, and dedicated SFRs. The ALU features a multiply-and-divide unit, a bit-mask generator, and a barrel shifter.

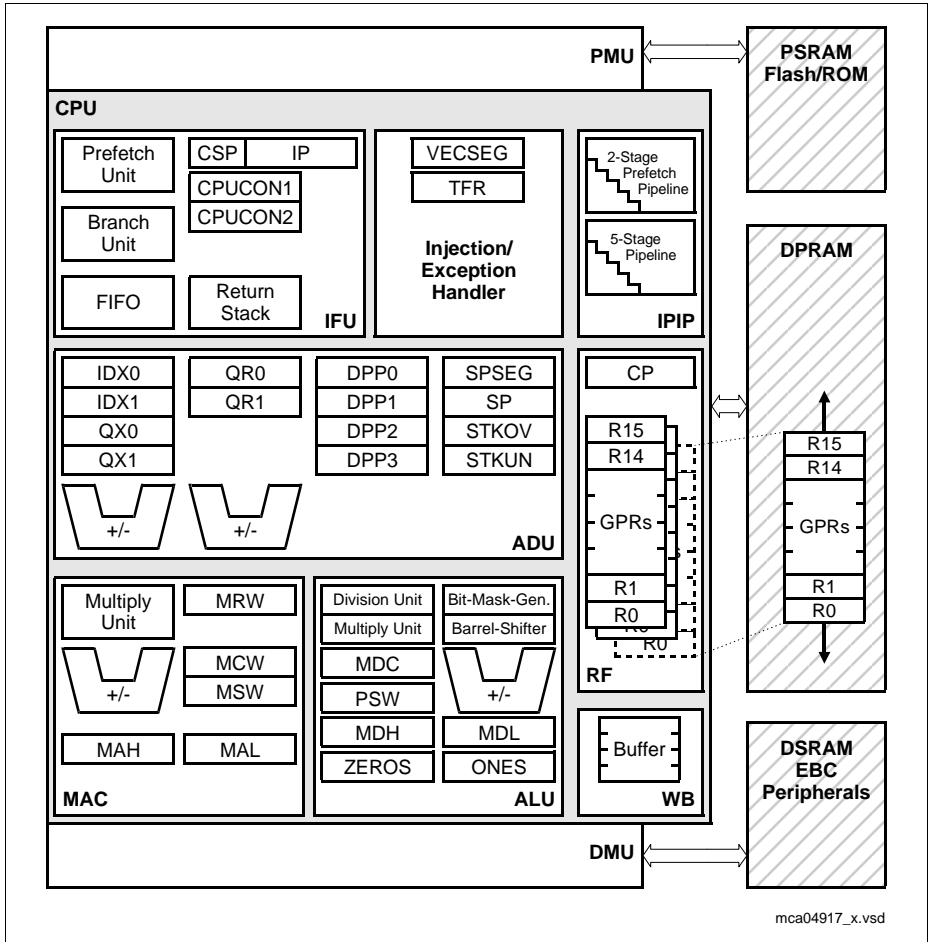


Figure 5 CPU Block Diagram

Functional Description

With this hardware most XC228xH instructions are executed in a single machine cycle of 10 ns @ 100-MHz CPU clock. For example, shift and rotate instructions are always processed during one machine cycle, no matter how many bits are shifted. Also, multiplication and most MAC instructions execute in one cycle. All multiple-cycle instructions have been optimized so that they can be executed very fast; for example, a 32-/16-bit division is started within 4 cycles while the remaining cycles are executed in the background. Another pipeline optimization, the branch target prediction, eliminates the execution time of branch instructions if the prediction was correct.

The CPU has a register context consisting of up to three register banks with 16 word-wide GPRs each at its disposal. One of these register banks is physically allocated within the on-chip DPRAM area. A Context Pointer (CP) register determines the base address of the active register bank accessed by the CPU at any time. The number of these register bank copies is only restricted by the available internal RAM space. For easy parameter passing, a register bank may overlap others.

A system stack of up to 32 Kwords is provided for storage of temporary data. The system stack can be allocated to any location within the address space (preferably in the on-chip RAM area); it is accessed by the CPU with the stack pointer (SP) register. Two separate SFRs, STKOV and STKUN, are implicitly compared with the stack pointer value during each stack access to detect stack overflow or underflow.

The high performance of the CPU hardware implementation can be best utilized by the programmer with the highly efficient XC228xH instruction set. This includes the following instruction classes:

- Standard Arithmetic Instructions
- DSP-Oriented Arithmetic Instructions
- Logical Instructions
- Boolean Bit Manipulation Instructions
- Compare and Loop Control Instructions
- Shift and Rotate Instructions
- Prioritize Instruction
- Data Movement Instructions
- System Stack Instructions
- Jump and Call Instructions
- Return Instructions
- System Control Instructions
- Miscellaneous Instructions

The basic instruction length is either 2 or 4 bytes. Possible operand types are bits, bytes and words. A variety of direct, indirect or immediate addressing modes are provided to specify the required operands.

3.4 Memory Protection Unit (MPU)

The XC228xH's Memory Protection Unit (MPU) protects user-specified memory areas from unauthorized read, write, or instruction fetch accesses. The MPU can protect the whole address space including the peripheral area. This completes established mechanisms such as the register security mechanism or stack overrun/underrun detection.

Four Protection Levels support flexible system programming where operating system, low level drivers, and applications run on separate levels. Each protection level permits different access restrictions for instructions and/or data.

Every access is checked (if the MPU is enabled) and an access violating the permission rules will be marked as invalid and leads to a protection trap.

A set of protection registers for each protection level specifies the address ranges and the access permissions. Applications requiring more than 4 protection levels can dynamically re-program the protection registers.

3.5 Memory Checker Module (MCHK)

The XC228xH's Memory Checker Module calculates a checksum (fractional polynomial division) on a block of data, often called Cyclic Redundancy Code (CRC). It is based on a 32-bit linear feedback shift register and may, therefore, also be used to generate pseudo-random numbers.

The Memory Checker Module is a 16-bit parallel input signature compression circuitry which enables error detection within a block of data stored in memory, registers, or communicated e.g. via serial communication lines. It reduces the probability of error masking due to repeated error patterns by calculating the signature of blocks of data.

The polynomial used for operation is configurable, so most of the commonly used polynomials may be used. Also, the block size for generating a CRC result is configurable via a local counter. An interrupt may be generated if testing the current data block reveals an error.

An autonomous CRC compare circuitry is included to enable redundant error detection, e.g. to enable higher safety integrity levels.

The Memory Checker Module provides enhanced fault detection (beyond parity or ECC) for data and instructions in volatile and non volatile memories. This is especially important for the safety and reliability of embedded systems.

3.6 Interrupt System

The architecture of the XC228xH supports several mechanisms for fast and flexible response to service requests; these can be generated from various sources internal or external to the microcontroller. Any of these interrupt requests can be programmed to be serviced by the Interrupt Controller or by the Peripheral Event Controller (PEC).

Using a standard interrupt service the current program execution is suspended and a branch to the interrupt vector table is performed. With the PEC just one cycle is 'stolen' from the current CPU activity to perform the PEC service. A PEC service implies a single byte or word data transfer between any two memory locations with an additional increment of either the PEC source pointer, the destination pointer, or both. An individual PEC transfer counter is implicitly decremented for each PEC service except when performing in the continuous transfer mode. When this counter reaches zero, a standard interrupt is performed to the corresponding source-related vector location. PEC services are particularly well suited to supporting the transmission or reception of blocks of data. The XC228xH has eight PEC channels, each with fast interrupt-driven data transfer capabilities.

With a minimum interrupt response time of $7/11^{1)}$ CPU clocks, the XC228xH can react quickly to the occurrence of non-deterministic events.

Interrupt Nodes and Source Selection

The interrupt system provides 112 physical nodes with separate control register containing an interrupt request flag, an interrupt enable flag and an interrupt priority bit field. Most interrupt sources are assigned to a dedicated node. A particular subset of interrupt sources shares a set of nodes. The source selection can be programmed using the interrupt source selection (ISSR) registers.

External Request Unit (ERU)

A dedicated External Request Unit (ERU) is provided to route and preprocess selected on-chip peripheral and external interrupt requests. The ERU features 4 programmable input channels with event trigger logic (ETL) a routing matrix and 4 output gating units (OGU). The ETL features rising edge, falling edge, or both edges event detection. The OGU combines the detected interrupt events and provides filtering capabilities depending on a programmable pattern match or miss.

Trap Processing

The XC228xH provides efficient mechanisms to identify and process exceptions or error conditions that arise during run-time, the so-called 'Hardware Traps'. A hardware trap causes an immediate system reaction similar to a standard interrupt service (branching

1) Depending if the jump cache is used or not.

Functional Description

to a dedicated vector table location). The occurrence of a hardware trap is also indicated by a single bit in the trap flag register (TFR). Unless another higher-priority trap service is in progress, a hardware trap will interrupt any ongoing program execution. In turn, hardware trap services can normally not be interrupted by standard or PEC interrupts.

Depending on the package option up to 3 External Service Request (ESR) pins are provided. The ESR unit processes their input values and allows to implement user controlled trap functions (System Requests SR0 and SR1). In this way reset, wakeup and power control can be efficiently realized.

Software interrupts are supported by the 'TRAP' instruction in combination with an individual trap (interrupt) number. Alternatively to emulate an interrupt by software a program can trigger interrupt requests by writing the Interrupt Request (IR) bit of an interrupt control register.

3.7 On-Chip Debug Support (OCDS)

The On-Chip Debug Support system built into the XC228xH provides a broad range of debug and emulation features. User software running on the XC228xH can be debugged within the target system environment.

The OCDS is controlled by an external debugging device via the debug interface. This either consists of the 2-pin Device Access Port (DAP) or of the JTAG port conforming to IEEE-1149. The debug interface can be completed with an optional break interface.

The debugger controls the OCDS with a set of dedicated registers accessible via the debug interface (DAP or JTAG). In addition the OCDS system can be controlled by the CPU, e.g. by a monitor program. An injection interface allows the execution of OCDS-generated instructions by the CPU.

Multiple breakpoints can be triggered by on-chip hardware, by software, or by an external trigger input. Single stepping is supported, as is the injection of arbitrary instructions and read/write access to the complete internal address space. A breakpoint trigger can be answered with a CPU halt, a monitor call, a data transfer, or/and the activation of an external signal.

Tracing of data can be obtained via the debug interface, or via the external bus interface for increased performance.

Tracing of program execution is supported by the XC2000 Family emulation device. With this device the DAP can operate on clock rates of up to 20 MHz.

The DAP interface uses two interface signals, the JTAG interface uses four interface signals, to communicate with external circuitry. The debug interface can be amended with two optional break lines.

3.8 Capture/Compare Units (CC1 and CC2)

Each CAPCOM unit supports generation and control of timing sequences on up to 16 channels with a maximum resolution of one system clock cycle (eight cycles in staggered mode). The CAPCOM unit is typically used to handle high-speed I/O tasks such as pulse and waveform generation, pulse width modulation (PWM), digital to analog (D/A) conversion, software timing, or time recording with respect to external events.

Two 16-bit timers with reload registers provide two independent time bases for the capture/compare register array.

The input clock for the timers is programmable to several prescaled values of the internal system clock, or may be derived from an overflow/underflow of timer T6 in module GPT2. This provides a wide range of variation for the timer period and resolution and allows precise adjustments to the application specific requirements. In addition, external count inputs allow event scheduling for the capture/compare registers relative to external events.

The capture/compare register array contains 16 dual purpose capture/compare registers, each of which may be individually allocated to either CAPCOM timer and programmed for capture or compare function.

All registers have each one port pin associated with it which serves as an input pin for triggering the capture function, or as an output pin to indicate the occurrence of a compare event.

When a capture/compare register has been selected for capture mode, the current contents of the allocated timer will be latched ('captured') into the capture/compare register in response to an external event at the port pin which is associated with this register. In addition, a specific interrupt request for this capture/compare register is generated. Either a positive, a negative, or both a positive and a negative external signal transition at the pin can be selected as the triggering event.

The contents of all registers which have been selected for one of the five compare modes are continuously compared with the contents of the allocated timers.

When a match occurs between the timer value and the value in a capture/compare register, specific actions will be taken based on the selected compare mode.

Table 9 Compare Modes

Compare Modes	Function
Mode 0	Interrupt-only compare mode; Several compare interrupts per timer period are possible
Mode 1	Pin toggles on each compare match; Several compare events per timer period are possible

Table 9 Compare Modes (cont'd)

Compare Modes	Function
Mode 2	Interrupt-only compare mode; Only one compare interrupt per timer period is generated
Mode 3	Pin set '1' on match; pin reset '0' on compare timer overflow; Only one compare event per timer period is generated
Double Register Mode	Two registers operate on one pin; Pin toggles on each compare match; Several compare events per timer period are possible
Single Event Mode	Generates single edges or pulses; Can be used with any compare mode

When a capture/compare register has been selected for capture mode, the current contents of the allocated timer will be latched ('captured') into the capture/compare register in response to an external event at the port pin associated with this register. In addition, a specific interrupt request for this capture/compare register is generated. Either a positive, a negative, or both a positive and a negative external signal transition at the pin can be selected as the triggering event.

The contents of all registers selected for one of the five compare modes are continuously compared with the contents of the allocated timers.

When a match occurs between the timer value and the value in a capture/compare register, specific actions will be taken based on the compare mode selected.

Functional Description

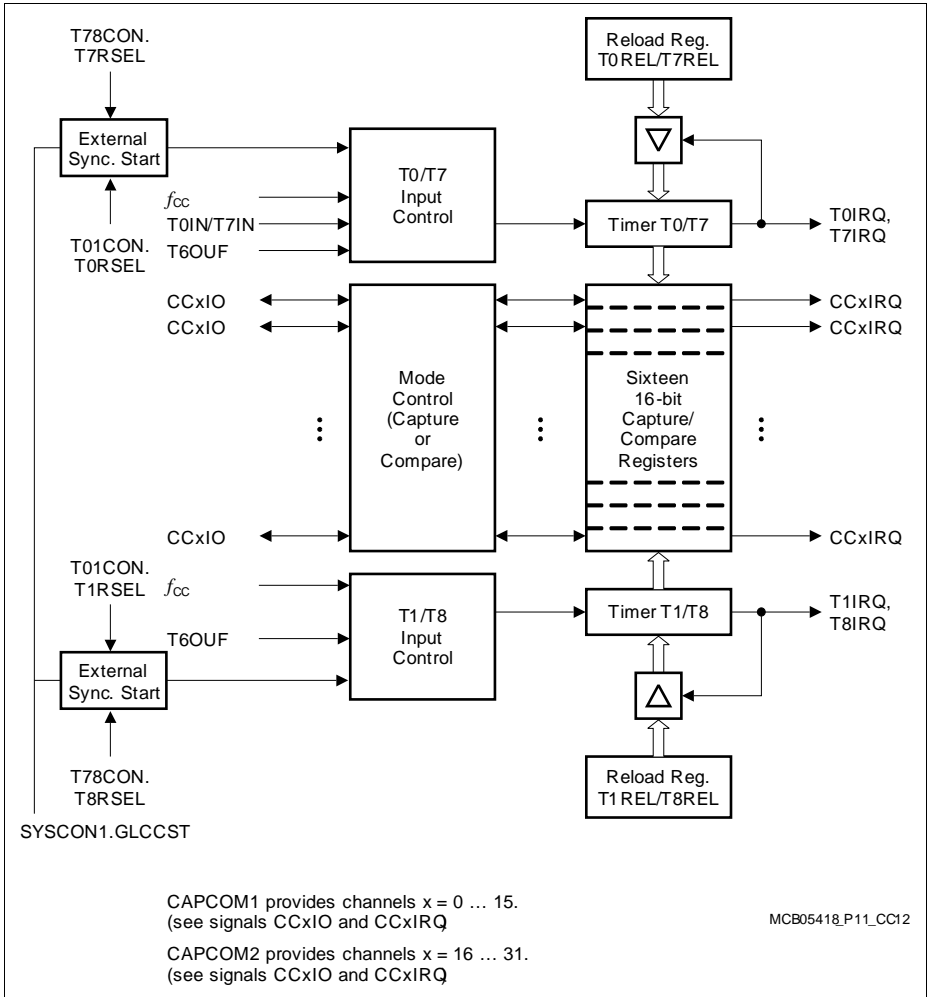


Figure 6 CAPCOM Unit Block Diagram

3.9 Capture/Compare Units CCU6x

The XC228xH types feature the CCU60, CCU61, CCU62 and CCU63 unit(s).

CCU6 is a high-resolution capture and compare unit with application-specific modes. It provides inputs to start the timers synchronously, an important feature in devices with several CCU6 modules.

The module provides two independent timers (T12, T13), that can be used for PWM generation, especially for AC motor control. Additionally, special control modes for block commutation and multi-phase machines are supported.

Timer 12 Features

- Three capture/compare channels, where each channel can be used either as a capture or as a compare channel.
- Supports generation of a three-phase PWM (six outputs, individual signals for high-side and low-side switches)
- 16-bit resolution, maximum count frequency = peripheral clock
- Dead-time control for each channel to avoid short circuits in the power stage
- Concurrent update of the required T12/13 registers
- Center-aligned and edge-aligned PWM can be generated
- Single-shot mode supported
- Many interrupt request sources
- Hysteresis-like control mode
- Automatic start on a HW event (T12HR, for synchronization purposes)

Timer 13 Features

- One independent compare channel with one output
- 16-bit resolution, maximum count frequency = peripheral clock
- Can be synchronized to T12
- Interrupt generation at period match and compare match
- Single-shot mode supported
- Automatic start on a HW event (T13HR, for synchronization purposes)

Additional Features

- Block commutation for brushless DC drives implemented
- Position detection via Hall sensor pattern
- Automatic rotational speed measurement for block commutation
- Integrated error handling
- Fast emergency stop without CPU load via external signal ($\overline{\text{CTRAP}}$)
- Control modes for multi-channel AC drives
- Output levels can be selected and adapted to the power stage

Functional Description

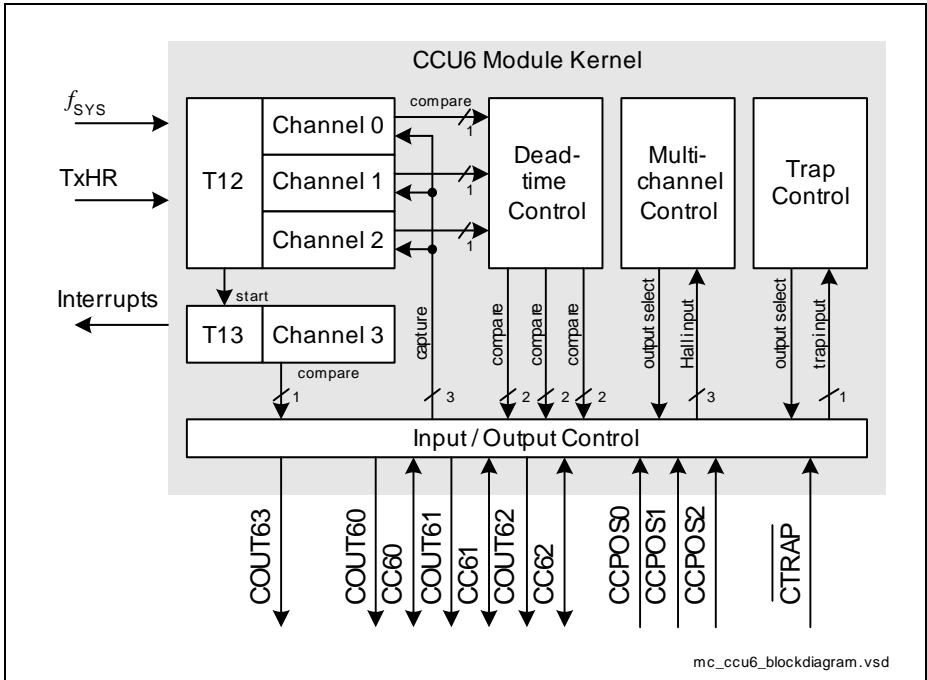


Figure 7 CCU6 Block Diagram

Timer T12 can work in capture and/or compare mode for its three channels. The modes can also be combined. Timer T13 can work in compare mode only. The multi-channel control unit generates output patterns that can be modulated by timer T12 and/or timer T13. The modulation sources can be selected and combined for signal modulation.

3.10 General Purpose Timer (GPT12E) Unit

The GPT12E unit is a very flexible multifunctional timer/counter structure which can be used for many different timing tasks such as event timing and counting, pulse width and duty cycle measurements, pulse generation, or pulse multiplication.

The GPT12E unit incorporates five 16-bit timers organized in two separate modules, GPT1 and GPT2. Each timer in each module may either operate independently in a number of different modes or be concatenated with another timer of the same module.

Each of the three timers T2, T3, T4 of **module GPT1** can be configured individually for one of four basic modes of operation: Timer, Gated Timer, Counter, and Incremental Interface Mode. In Timer Mode, the input clock for a timer is derived from the system clock and divided by a programmable prescaler. Counter Mode allows timer clocking in reference to external events.

Pulse width or duty cycle measurement is supported in Gated Timer Mode, where the operation of a timer is controlled by the 'gate' level on an external input pin. For these purposes each timer has one associated port pin (TxIN) which serves as a gate or clock input. The maximum resolution of the timers in module GPT1 is 4 system clock cycles.

The counting direction (up/down) for each timer can be programmed by software or altered dynamically by an external signal on a port pin (TxEUD), e.g. to facilitate position tracking.

In Incremental Interface Mode the GPT1 timers can be directly connected to the incremental position sensor signals A and B through their respective inputs TxIN and TxEUD. Direction and counting signals are internally derived from these two input signals, so that the contents of the respective timer Tx corresponds to the sensor position. The third position sensor signal TOP0 can be connected to an interrupt input.

Timer T3 has an output toggle latch (T3OTL) which changes its state on each timer overflow/underflow. The state of this latch may be output on pin T3OUT e.g. for time out monitoring of external hardware components. It may also be used internally to clock timers T2 and T4 for measuring long time periods with high resolution.

In addition to the basic operating modes, T2 and T4 may be configured as reload or capture register for timer T3. A timer used as capture or reload register is stopped. The contents of timer T3 is captured into T2 or T4 in response to a signal at the associated input pin (TxIN). Timer T3 is reloaded with the contents of T2 or T4, triggered either by an external signal or a selectable state transition of its toggle latch T3OTL. When both T2 and T4 are configured to alternately reload T3 on opposite state transitions of T3OTL with the low and high times of a PWM signal, this signal can be continuously generated without software intervention.

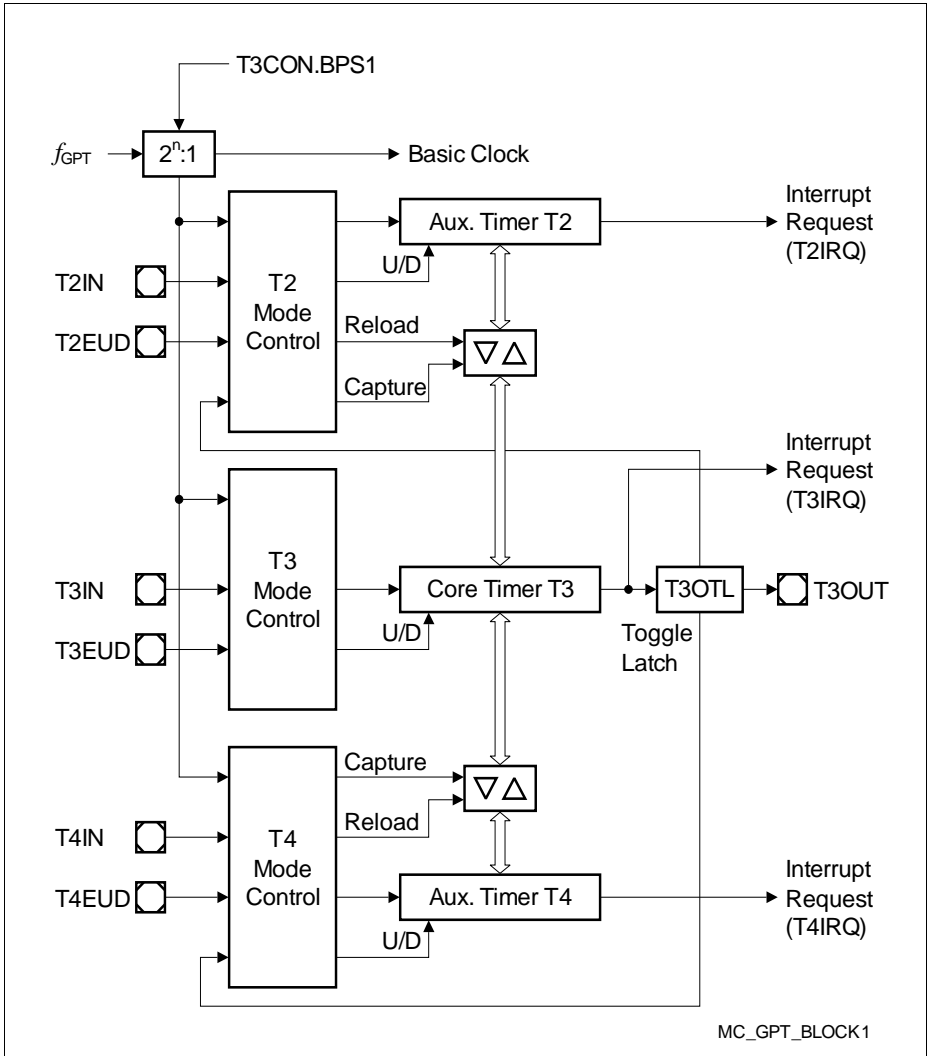


Figure 8 Block Diagram of GPT1

Functional Description

With its maximum resolution of 2 system clock cycles, the **GPT2 module** provides precise event control and time measurement. It includes two timers (T5, T6) and a capture/reload register (CAPREL). Both timers can be clocked with an input clock which is derived from the CPU clock via a programmable prescaler or with external signals. The counting direction (up/down) for each timer can be programmed by software or altered dynamically with an external signal on a port pin (TxEUD). Concatenation of the timers is supported with the output toggle latch (T6OTL) of timer T6, which changes its state on each timer overflow/underflow.

The state of this latch may be used to clock timer T5, and/or it may be output on pin T6OUT. The overflows/underflows of timer T6 can also be used to clock the CAPCOM2 timers and to initiate a reload from the CAPREL register.

The CAPREL register can capture the contents of timer T5 based on an external signal transition on the corresponding port pin (CAPIN); timer T5 may optionally be cleared after the capture procedure. This allows the XC228xH to measure absolute time differences or to perform pulse multiplication without software overhead.

The capture trigger (timer T5 to CAPREL) can also be generated upon transitions of GPT1 timer T3 inputs T3IN and/or T3EUD. This is especially advantageous when T3 operates in Incremental Interface Mode.

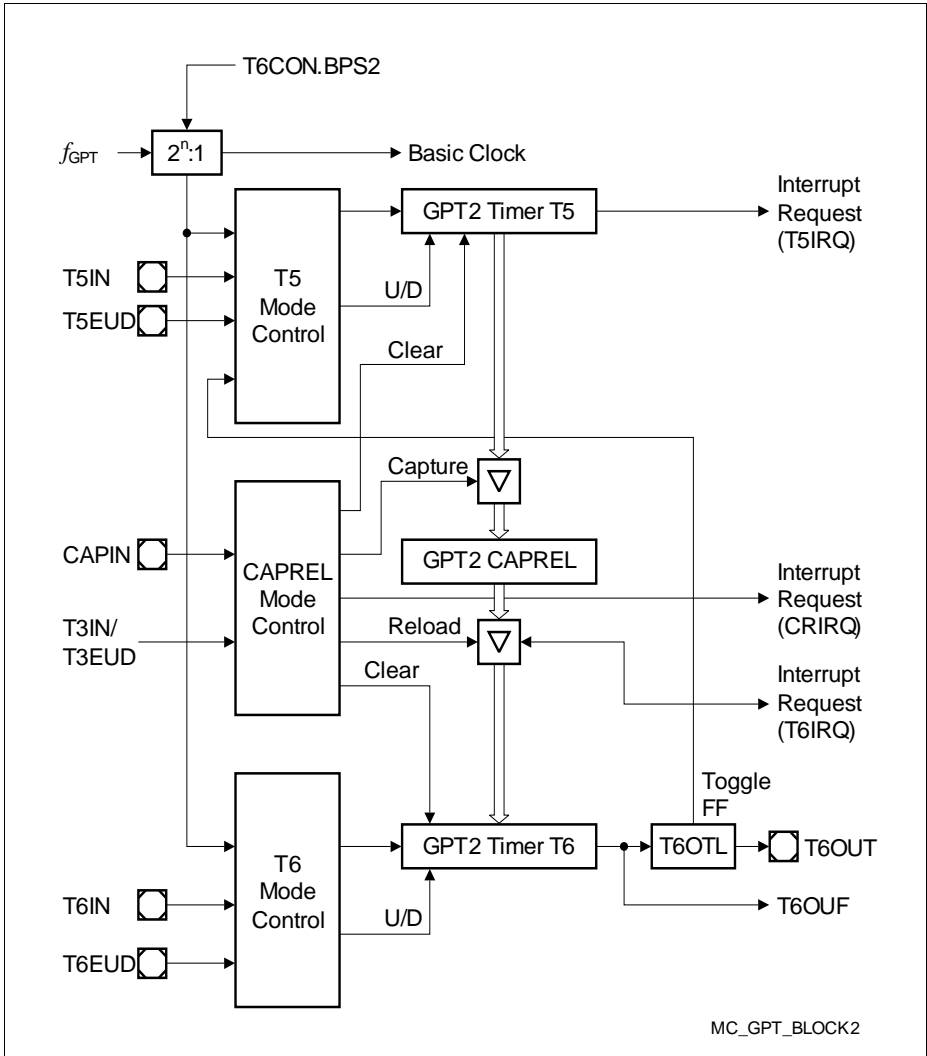


Figure 9 Block Diagram of GPT2

3.11 Real Time Clock

The Real Time Clock (RTC) module of the XC228xH can be clocked with a clock signal selected from internal sources or external sources (pins).

The RTC basically consists of a chain of divider blocks:

- Selectable 32:1 and 8:1 dividers (on - off)
- The reloadable 16-bit timer T14
- The 32-bit RTC timer block (accessible via registers RTCH and RTCL) consisting of:
 - a reloadable 10-bit timer
 - a reloadable 6-bit timer
 - a reloadable 6-bit timer
 - a reloadable 10-bit timer

All timers count up. Each timer can generate an interrupt request. All requests are combined to a common node request.

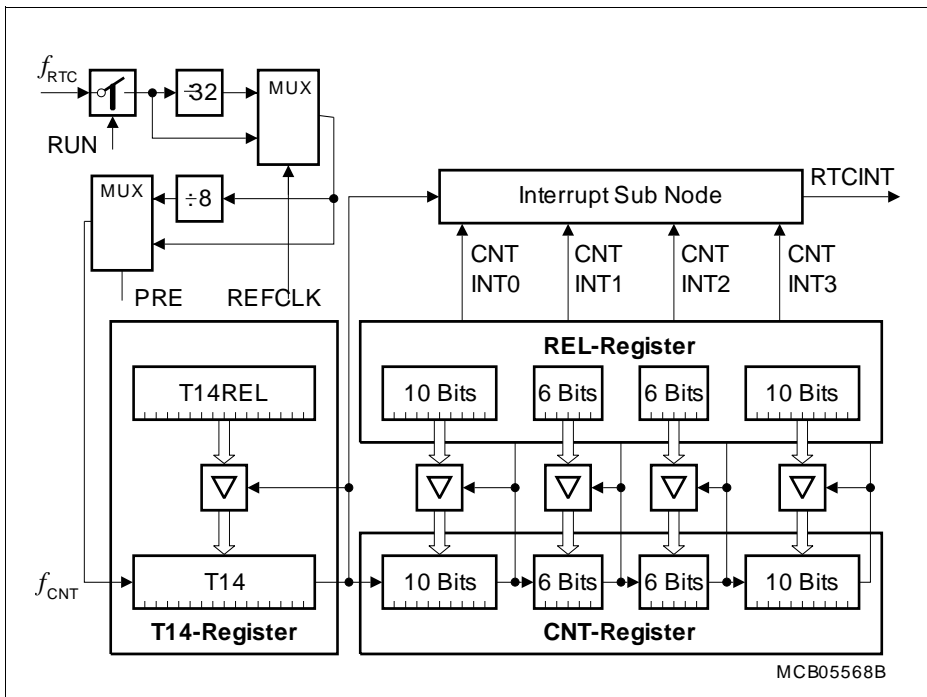


Figure 10 RTC Block Diagram

Note: The registers associated with the RTC are only affected by a power reset.

Functional Description

The RTC module can be used for different purposes:

- System clock to determine the current time and date
- Cyclic time-based interrupt, to provide a system time tick independent of CPU frequency and other resources
- 48-bit timer for long-term measurements
- Alarm interrupt at a defined time

3.12 A/D Converters

For analog signal measurement, up to two 10-bit A/D converters (ADC0, ADC1) with 16 + 8 multiplexed input channels and a sample and hold circuit have been integrated on-chip. 4 inputs can be converted by both A/D converters. Conversions use the successive approximation method. The sample time (to charge the capacitors) and the conversion time are programmable so that they can be adjusted to the external circuit. The A/D converters can also operate in 8-bit conversion mode, further reducing the conversion time.

Several independent conversion result registers, selectable interrupt requests, and highly flexible conversion sequences provide a high degree of programmability to meet the application requirements. Both modules can be synchronized to allow parallel sampling of two input channels.

For applications that require more analog input channels, external analog multiplexers can be controlled automatically. For applications that require fewer analog input channels, the remaining channel inputs can be used as digital input port pins.

The A/D converters of the XC228xH support two types of request sources which can be triggered by several internal and external events.

- Parallel requests are activated at the same time and then executed in a predefined sequence.
- Queued requests are executed in a user-defined sequence.

In addition, the conversion of a specific channel can be inserted into a running sequence without disturbing that sequence. All requests are arbitrated according to the priority level assigned to them.

Data reduction features reduce the number of required CPU access operations allowing the precise evaluation of analog inputs (high conversion rate) even at a low CPU speed. Result data can be reduced by limit checking or accumulation of results.

The Peripheral Event Controller (PEC) can be used to control the A/D converters or to automatically store conversion results to a table in memory for later evaluation, without requiring the overhead of entering and exiting interrupt routines for each data transfer. Each A/D converter contains eight result registers which can be concatenated to build a result FIFO. Wait-for-read mode can be enabled for each result register to prevent the loss of conversion data.

In order to decouple analog inputs from digital noise and to avoid input trigger noise, those pins used for analog input can be disconnected from the digital input stages. This can be selected for each pin separately with the Port x Digital Input Disable registers.

The Auto-Power-Down feature of the A/D converters minimizes the power consumption when no conversion is in progress.

Broken wire detection for each channel and a multiplexer test mode provide information to verify the proper operation of the analog signal sources (e.g. a sensor system).

3.13 Universal Serial Interface Channel Modules (USIC)

The XC228xH features the USIC modules USIC0, USIC1, USIC2, USIC3, USIC4. Each module provides two serial communication channels.

The Universal Serial Interface Channel (USIC) module is based on a generic data shift and data storage structure which is identical for all supported serial communication protocols. Each channel supports complete full-duplex operation with a basic data buffer structure (one transmit buffer and two receive buffer stages). In addition, the data handling software can use FIFOs.

The protocol part (generation of shift clock/data/control signals) is independent of the general part and is handled by protocol-specific preprocessors (PPPs).

The USIC's input/output lines are connected to pins by a pin routing unit. The inputs and outputs of each USIC channel can be assigned to different interface pins, providing great flexibility to the application software. All assignments can be made during runtime.

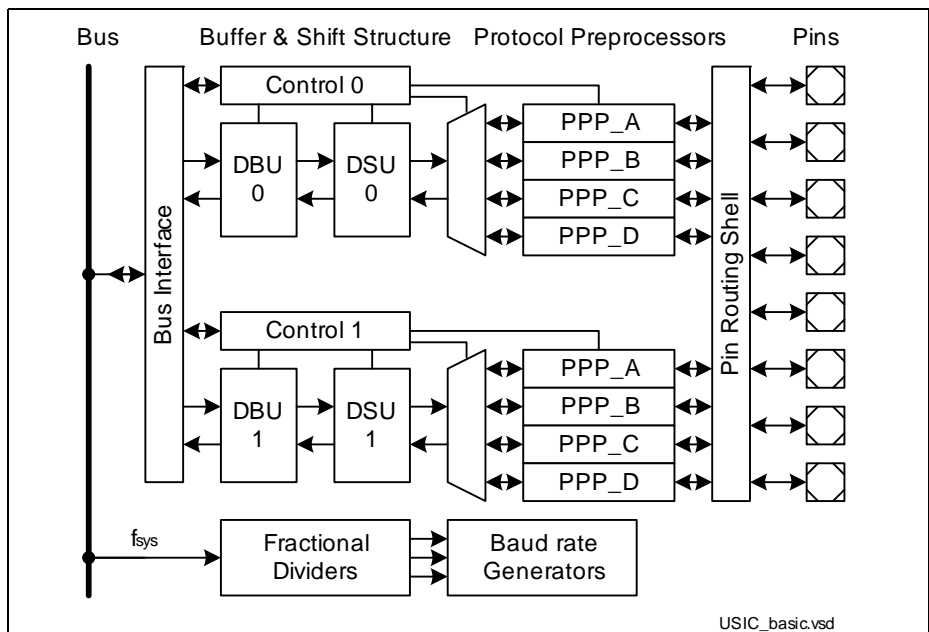


Figure 11 General Structure of a USIC Module

The regular structure of the USIC module brings the following advantages:

- Higher flexibility through configuration with same look-and-feel for data management
- Reduced complexity for low-level drivers serving different protocols
- Wide range of protocols with improved performances (baud rate, buffer handling)

Target Protocols

Each USIC channel can receive and transmit data frames with a selectable data word width from 1 to 16 bits in each of the following protocols:

- **UART** (asynchronous serial channel)
 - module capability: maximum baud rate = $f_{\text{SYS}} / 4$
 - data frame length programmable from 1 to 63 bits
 - MSB or LSB first
- **LIN** Support (Local Interconnect Network)
 - module capability: maximum baud rate = $f_{\text{SYS}} / 16$
 - checksum generation under software control
 - baud rate detection possible by built-in capture event of baud rate generator
- **SSC/SPI** (synchronous serial channel with or without data buffer)
 - module capability: maximum baud rate = $f_{\text{SYS}} / 2$, limited by loop delay
 - number of data bits programmable from 1 to 63, more with explicit stop condition
 - MSB or LSB first
 - optional control of slave select signals
- **IIC** (Inter-IC Bus)
 - supports baud rates of 100 kbit/s and 400 kbit/s
- **IIS** (Inter-IC Sound Bus)
 - module capability: maximum baud rate = $f_{\text{SYS}} / 2$

Note: Depending on the selected functions (such as digital filters, input synchronization stages, sample point adjustment, etc.), the maximum achievable baud rate can be limited. Please note that there may be additional delays, such as internal or external propagation delays and driver delays (e.g. for collision detection in UART mode, for IIC, etc.).

3.14 MultiCAN Module

The MultiCAN module contains independently operating CAN nodes with Full-CAN functionality which are able to exchange Data and Remote Frames using a gateway function. Transmission and reception of CAN frames is handled in accordance with CAN specification V2.0 B (active). Each CAN node can receive and transmit standard frames with 11-bit identifiers as well as extended frames with 29-bit identifiers.

All CAN nodes share a common set of message objects. Each message object can be individually allocated to one of the CAN nodes. Besides serving as a storage container for incoming and outgoing frames, message objects can be combined to build gateways between the CAN nodes or to set up a FIFO buffer.

Note: The number of CAN nodes and message objects depends on the selected device type.

The message objects are organized in double-chained linked lists, where each CAN node has its own list of message objects. A CAN node stores frames only into message objects that are allocated to its own message object list and it transmits only messages belonging to this message object list. A powerful, command-driven list controller performs all message object list operations.

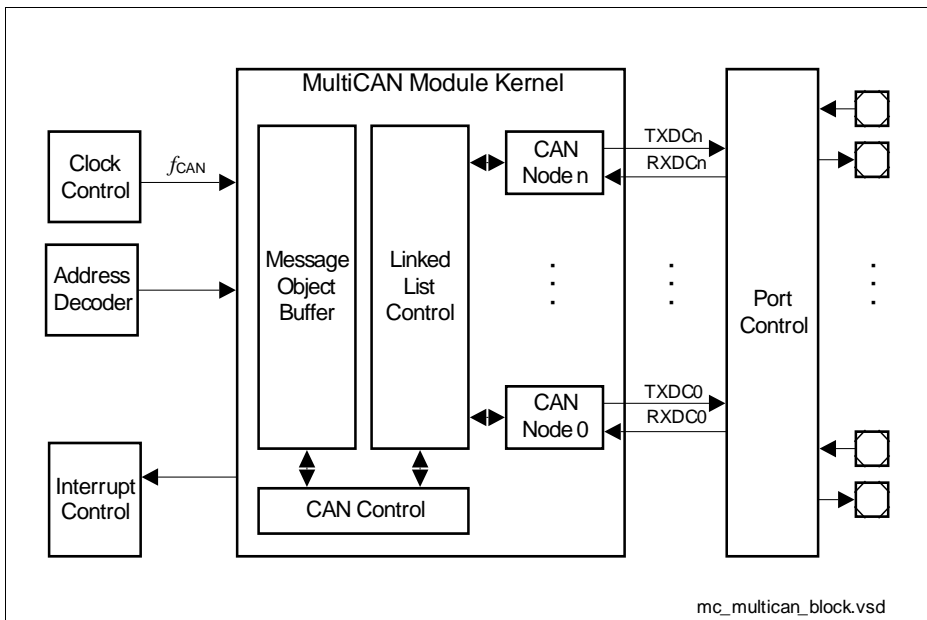


Figure 12 Block Diagram of MultiCAN Module

MultiCAN Features

- CAN functionality conforming to CAN specification V2.0 B active for each CAN node (compliant to ISO 11898)
- Independent CAN nodes
- Set of independent message objects (shared by the CAN nodes)
- Dedicated control registers for each CAN node
- Data transfer rate up to 1 Mbit/s, individually programmable for each node
- Flexible and powerful message transfer control and error handling capabilities
- Full-CAN functionality for message objects:
 - Can be assigned to one of the CAN nodes
 - Configurable as transmit or receive objects, or as message buffer FIFO
 - Handle 11-bit or 29-bit identifiers with programmable acceptance mask for filtering
 - Remote Monitoring Mode, and frame counter for monitoring
- Automatic Gateway Mode support
- 16 individually programmable interrupt nodes
- Analyzer mode for CAN bus monitoring

3.15 FlexRay™ Protocol Controller (E-Ray)

The E-Ray module performs communication according to the FlexRay™ ¹⁾ protocol specification V2.1. With maximum specified clock the bitrate can be programmed to values up to 10 Mbit/s. Additional bus driver (BD) hardware is required for connection to the physical layer.

Figure 13 shows a global view of the E-Ray interface.

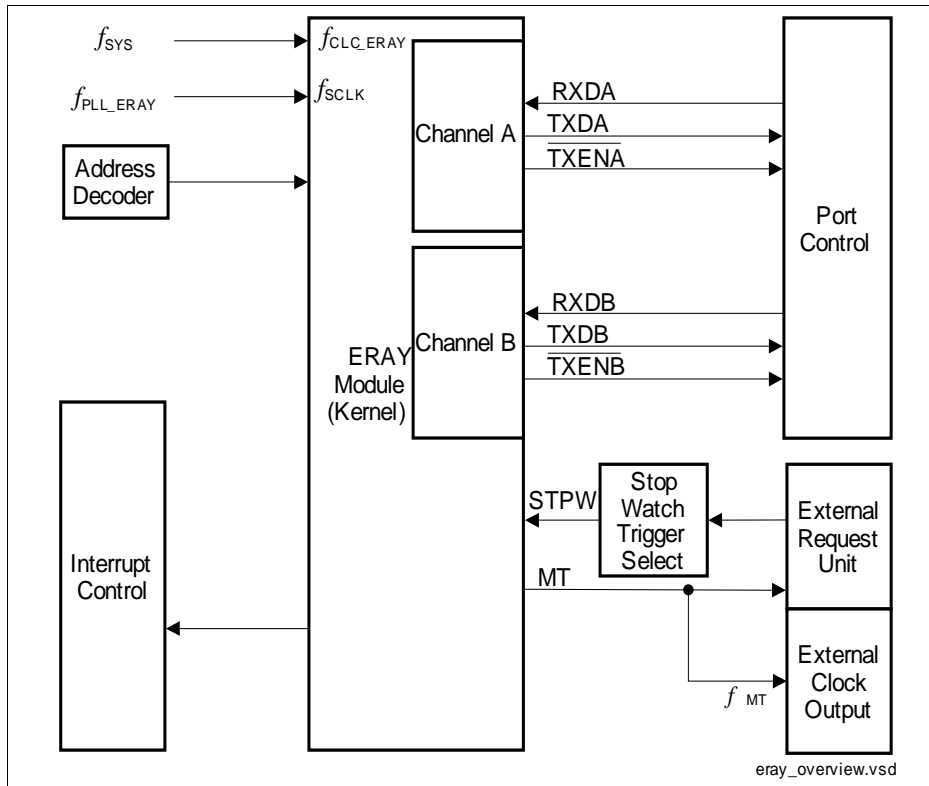


Figure 13 General Block Diagram of the E-Ray Interface

The E-Ray module communicates with the external world via three I/O lines each channel. The RXDA_x and RXDB_x lines are the receive data input signals, TXDA and TXDB lines are the transmit output signals, and TXENA and TXENB the transmit enable signals.

1) FlexRay™ is a trademark of the FlexRay Consortium.

Functional Description

Clock control, address decoding, and service request control are managed outside the E-Ray module kernel.

FlexRay™ Module Features

For communication on a FlexRay™ network, individual Message Buffers with up to 254 data bytes are configurable. The message storage consists of a single-ported Message RAM that holds up to 128 Message Buffers. All functions concerning the handling of messages are implemented in the Message Handler. Those functions are the acceptance filtering, the transfer of messages between the two FlexRay™ Channel Protocol Controllers and the Message RAM, maintaining the transmission schedule as well as providing message status information.

The register set of the E-Ray IP-module can be accessed directly by an external Host via the module's Host interface. These registers are used to control/configure/monitor the FlexRay™ Channel Protocol Controllers, Message Handler, Global Time Unit, System Universal Control, Frame and Symbol Processing, Network Management, Service Request Control, and to access the Message RAM via Input / Output Buffer.

The E-Ray IP-module supports the following features:

- Conformance with FlexRay™ protocol specification V2.1
- Data rates of up to 10 Mbit/s on each channel
- Up to 128 Message Buffers configurable
- 8 Kbytes of Message RAM for storage of e.g. 128 Message Buffers with max. 48 bytes data field or up to 30 Message Buffers with 254 bytes Data Sections
- Configuration of Message Buffers with different payload lengths possible
- One configurable receive FIFO
- Each Message Buffer can be configured as receive buffer, as transmit buffer or as part of the receive FIFO
- Host access to Message Buffers via Input and Output Buffer.
Input Buffer: Holds message to be transferred to the Message RAM
Output Buffer: Holds message read from the Message RAM
- Filtering for slot counter, cycle counter, and channel
- Maskable module service requests
- Network Management supported
- Four service request lines
- Automatic delayed read access to Output Command Request Register (OBCR) if a data transfer from Message RAM to Output Shadow Buffer (initiated by a previous write access to the OBCR) is ongoing.
- Automatic delayed read access to Input Command Request Register (IBCR) if a data transfer from Input Shadow Buffer to Message RAM to (initiated by a previous write access to the IBCR) is ongoing.
- Four Input Buffers for building up transmission Frames in parallel.
- Flag indicating which Input Buffer is currently accessible by the host.

3.16 System Timer

The System Timer consists of a programmable prescaler and two concatenated timers (10 bits and 6 bits). Both timers can generate interrupt requests. The clock source can be selected and the timers can also run during power reduction modes.

Therefore, the System Timer enables the software to maintain the current time for scheduling functions or for the implementation of a clock.

3.17 Watchdog Timer

The Watchdog Timer is one of the fail-safe mechanisms which have been implemented to prevent the controller from malfunctioning for longer periods of time.

The Watchdog Timer is always enabled after an application reset of the chip. It can be disabled and enabled at any time by executing the instructions DISWDT and ENWDT respectively. The software has to service the Watchdog Timer before it overflows. If this is not the case because of a hardware or software failure, the Watchdog Timer overflows, generating a prewarning interrupt and then a reset request.

The Watchdog Timer is a 16-bit timer clocked with the system clock divided by 16,384 or 256. The Watchdog Timer register is set to a prespecified reload value (stored in WDTREL) in order to allow further variation of the monitored time interval. Each time it is serviced by the application software, the Watchdog Timer is reloaded and the prescaler is cleared.

Time intervals between 3.2 μ s and 13.4 s can be monitored (@ 80 MHz).

Time intervals between 2.56 μ s and 10.71 s can be monitored (@ 100 MHz).

The default Watchdog Timer interval after power-up is 6.5 ms (@ 10 MHz).

3.18 Clock Generation

The Clock Generation Unit can generate the system clock signal f_{SYS} for the XC228xH from a number of external or internal clock sources:

- External clock signals with pad voltage or core voltage levels
- External crystal or resonator using the on-chip oscillator
- On-chip clock source for operation without crystal/resonator
- Wake-up clock (ultra-low-power) to further reduce power consumption

The programmable on-chip PLL with multiple prescalers generates a clock signal for maximum system performance from standard crystals, a clock input signal, or from the on-chip clock source. See also [Section 4.7.2](#).

The Oscillator Watchdog (OWD) generates an interrupt if the crystal oscillator frequency falls below a certain limit or stops completely. In this case, the system can be supplied with an emergency clock to enable operation even after an external clock failure.

All available clock signals can be output on one of two selectable pins.

3.19 Parallel Ports

The XC228xH provides up to 119 I/O lines which are organized into 11 input/output ports and 2 input ports. All port lines are bit-addressable, and all input/output lines can be individually (bit-wise) configured via port control registers. This configuration selects the direction (input/output), push/pull or open-drain operation, activation of pull devices, and edge characteristics (shape) and driver characteristics (output current) of the port drivers. The I/O ports are true bidirectional ports which are switched to high impedance state when configured as inputs. During the internal reset, all port pins are configured as inputs without pull devices active.

All port lines have alternate input or output functions associated with them. These alternate functions can be programmed to be assigned to various port pins to support the best utilization for a given application. For this reason, certain functions appear several times in **Table 10**.

All port lines that are not used for alternate functions may be used as general purpose I/O lines.

Table 10 Summary of the XC228xH's Ports

Port	Width	I/O	Connected Modules
P0	8	I/O	EBC (A7...A0), CCU6, USIC, CAN
P1	8	I/O	EBC (A15...A8), CCU6, USIC
P2	14	I/O	EBC (READY, $\overline{\text{BHE}}$, A23...A16, AD15...AD13, D15...D13), CAN, CC2, GPT12E, USIC, DAP/JTAG
P3	8	I/O	CAN, USIC
P4	8	I/O	EBC ($\overline{\text{CS4}}$... $\overline{\text{CS0}}$), CC2, CAN, GPT12E, USIC
P5	16	I	Analog Inputs, CCU6, DAP/JTAG, GPT12E, CAN, CC1
P6	4	I/O	ADC, CAN, GPT12E
P7	5	I/O	CAN, GPT12E, SCU, DAP/JTAG, CCU6, ADC, USIC
P8		I/O	CCU6, DAP/JTAG, CC1, USIC
P9	8	I/O	CCU6, DAP/JTAG, CAN, ERAY, CC1
P10	16	I/O	EBC (ALE, $\overline{\text{RD}}$, $\overline{\text{WR}}$, AD12...AD0, D12...D0), CCU6, USIC, DAP/JTAG, CAN, ERAY
P11	6	I/O	CCU6, USIC, CAN
P15	8	I	Analog Inputs, GPT12E

3.20 Power Management

The XC228xH provides the means to control the power it consumes either at a given time or averaged over a certain duration.

Three mechanisms can be used (and partly in parallel):

- **Supply Voltage Management** permits the temporary reduction of the supply voltage of major parts of the logic or even its complete disconnection. This drastically reduces the power consumed because it eliminates leakage current, particularly at high temperature.
Several power reduction modes provide the best balance of power reduction and wake-up time.
- **Clock Generation Management** controls the frequency of internal and external clock signals. Clock signals for currently inactive parts of logic are disabled automatically. The user can drastically reduce the consumed power by reducing the XC228xH system clock frequency.
External circuits can be controlled using the programmable frequency output EXTCLK.
- **Peripheral Management** permits temporary disabling of peripheral modules. Each peripheral can be disabled and enabled separately. The CPU can be switched off while the peripherals can continue to operate.

Wake-up from power reduction modes can be triggered either externally with signals generated by the external system, or internally by the on-chip wake-up timer. This supports intermittent operation of the XC228xH by generating cyclic wake-up signals. Full performance is available to quickly react to action requests while the intermittent sleep phases greatly reduce the average system power consumption.

Note: When selecting the supply voltage and the clock source and generation method, the required parameters must be carefully written to the respective bit fields, to avoid unintended intermediate states. Recommended sequences are provided which ensure the intended operation of power supply system and clock system. Please refer to the Programmer's Guide.

3.21 Instruction Set Summary

Table 11 lists the instructions of the XC228xH.

The addressing modes that can be used with a specific instruction, the function of the instructions, parameters for conditional execution of instructions, and the opcodes for each instruction can be found in the “**Instruction Set Manual**”.

This document also provides a detailed description of each instruction.

Table 11 Instruction Set Summary

Mnemonic	Description	Bytes
ADD(B)	Add word (byte) operands	2 / 4
ADDC(B)	Add word (byte) operands with Carry	2 / 4
SUB(B)	Subtract word (byte) operands	2 / 4
SUBC(B)	Subtract word (byte) operands with Carry	2 / 4
MUL(U)	(Un)Signed multiply direct GPR by direct GPR (16- × 16-bit)	2
DIV(U)	(Un)Signed divide register MDL by direct GPR (16-/16-bit)	2
DIVL(U)	(Un)Signed long divide reg. MD by direct GPR (32-/16-bit)	2
CPL(B)	Complement direct word (byte) GPR	2
NEG(B)	Negate direct word (byte) GPR	2
AND(B)	Bitwise AND, (word/byte operands)	2 / 4
OR(B)	Bitwise OR, (word/byte operands)	2 / 4
XOR(B)	Bitwise exclusive OR, (word/byte operands)	2 / 4
BCLR/BSET	Clear/Set direct bit	2
BMOV(N)	Move (negated) direct bit to direct bit	4
BAND/BOR/BXOR	AND/OR/XOR direct bit with direct bit	4
BCMP	Compare direct bit to direct bit	4
BFLDH/BFLDL	Bitwise modify masked high/low byte of bit-addressable direct word memory with immediate data	4
CMP(B)	Compare word (byte) operands	2 / 4
CMPD1/2	Compare word data to GPR and decrement GPR by 1/2	2 / 4
CMPI1/2	Compare word data to GPR and increment GPR by 1/2	2 / 4
PRIOR	Determine number of shift cycles to normalize direct word GPR and store result in direct word GPR	2
SHL/SHR	Shift left/right direct word GPR	2

Table 11 Instruction Set Summary (cont'd)

Mnemonic	Description	Bytes
ROL/ROR	Rotate left/right direct word GPR	2
ASHR	Arithmetic (sign bit) shift right direct word GPR	2
MOV(B)	Move word (byte) data	2 / 4
MOVBS/Z	Move byte operand to word op. with sign/zero extension	2 / 4
JMPA/I/R	Jump absolute/indirect/relative if condition is met	4
JMPS	Jump absolute to a code segment	4
JB(C)	Jump relative if direct bit is set (and clear bit)	4
JNB(S)	Jump relative if direct bit is not set (and set bit)	4
CALLA/I/R	Call absolute/indirect/relative subroutine if condition is met	4
CALLS	Call absolute subroutine in any code segment	4
PCALL	Push direct word register onto system stack and call absolute subroutine	4
TRAP	Call interrupt service routine via immediate trap number	2
PUSH/POP	Push/pop direct word register onto/from system stack	2
SCXT	Push direct word register onto system stack and update register with word operand	4
RET(P)	Return from intra-segment subroutine (and pop direct word register from system stack)	2
RETS	Return from inter-segment subroutine	2
RETI	Return from interrupt service subroutine	2
SBRK	Software Break	2
SRST	Software Reset	4
IDLE	Enter Idle Mode	4
PWRDN	Unused instruction ¹⁾	4
SRVWDT	Service Watchdog Timer	4
DISWDT/ENWDT	Disable/Enable Watchdog Timer	4
EINIT	End-of-Initialization Register Lock	4
ATOMIC	Begin ATOMIC sequence	2
EXTR	Begin EXTENDED Register sequence	2
EXTP(R)	Begin EXTENDED Page (and Register) sequence	2 / 4
EXTS(R)	Begin EXTENDED Segment (and Register) sequence	2 / 4

Table 11 Instruction Set Summary (cont'd)

Mnemonic	Description	Bytes
NOP	Null operation	2
CoMUL/CoMAC	Multiply (and accumulate)	4
CoADD/CoSUB	Add/Subtract	4
Co(A)SHR	(Arithmetic) Shift right	4
CoSHL	Shift left	4
CoLOAD/STORE	Load accumulator/Store MAC register	4
CoCMP	Compare	4
CoMAX/MIN	Maximum/Minimum	4
CoABS/CoRND	Absolute value/Round accumulator	4
CoMOV	Data move	4
CoNEG/NOP	Negate accumulator/Null operation	4

1) The Enter Power Down Mode instruction is not used in the XC228xH, due to the enhanced power control scheme. PWRDN will be correctly decoded, but will trigger no action.

4 Electrical Parameters

The operating range for the XC228xH is defined by its electrical parameters. For proper operation the specified limits must be respected when integrating the device in its target environment.

4.1 General Parameters

These parameters are valid for all subsequent descriptions, unless otherwise noted.

Table 12 Absolute Maximum Rating Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Output current on a pin when high value is driven	I_{OH} SR	-30	–	–	mA	
Output current on a pin when low value is driven	I_{OL} SR	–	–	30	mA	
Overload current	I_{OV} SR	-10	–	10	mA	¹⁾
Absolute sum of overload currents	$\Sigma I_{OV} $ SR	–	–	100	mA	¹⁾
Junction Temperature	T_J SR	-40	–	150	°C	
Storage Temperature	T_{ST} SR	-65	–	150	°C	
Digital supply voltage for IO pads and voltage regulators	V_{DDP} SR	-0.5	–	6.0	V	
Voltage on any pin with respect to ground (V_{SS})	V_{IN} SR	-0.5	–	$V_{DDP} + 0.5$	V	$V_{IN} \leq V_{DDP(max)}$

¹⁾ Overload condition occurs if the input voltage V_{IN} is out of the absolute maximum rating range. In this case the current must be limited to the listed values by design measures.

Note: Stresses above the values listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only. Functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for an extended time may affect device reliability. During absolute maximum rating overload conditions ($V_{IN} > V_{DDP}$ or $V_{IN} < V_{SS}$) the voltage on V_{DDP} pins with respect to ground (V_{SS}) must not exceed the values defined by the absolute maximum ratings.

4.1.1 Operating Conditions

The following operating conditions must not be exceeded to ensure correct operation of the XC228xH. All parameters specified in the following sections refer to these operating conditions, unless otherwise noticed.

Note: Typical parameter values refer to room temperature and nominal supply voltage, minimum/maximum parameter values also include conditions of minimum/maximum temperature and minimum/maximum supply voltage. Additional details are described where applicable.

Table 13 Operating Conditions

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Voltage Regulator Buffer Capacitance for DMP_M	C_{EVRM} SR	1.0	–	4.7	μF	¹⁾
Voltage Regulator Buffer Capacitance for DMP_1	C_{EVR1} SR	0.68	–	2.2	μF	²⁾¹⁾
External Load Capacitance	C_L SR	–	20 ³⁾	–	pF	pin out driver= default ⁴⁾
System frequency	f_{SYS} SR	–	–	100	MHz	⁵⁾
Overload current for analog inputs ⁶⁾	I_{OVA} SR	-2	–	5	mA	not subject to production test
Overload current for digital inputs ⁶⁾	I_{OVD} SR	-5	–	5	mA	not subject to production test
Overload current coupling factor for analog inputs ⁷⁾	K_{OVA} CC	–	2.5×10^{-4}	1.5×10^{-3}	–	$I_{OV} < 0$ mA; not subject to production test
		–	1.0×10^{-6}	1.0×10^{-4}	–	$I_{OV} > 0$ mA; not subject to production test

Electrical Parameters
Table 13 Operating Conditions (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Overload current coupling factor for digital I/O pins	K_{OVD} CC	–	1.0×10^{-2}	3.0×10^{-2}		$I_{OV} < 0$ mA; not subject to production test
		–	1.0×10^{-4}	5.0×10^{-3}		$I_{OV} > 0$ mA; not subject to production test
Absolute sum of overload currents	$\Sigma I_{OV} $ SR	–	–	50	mA	not subject to production test
Digital core supply voltage for domain M ⁸⁾	V_{DDIM} CC	–	1.5	–		
Digital core supply voltage for domain 1 ⁸⁾	V_{DDI1} CC	–	1.5	–		
Digital supply voltage for IO pads and voltage regulators	V_{DDP} SR	3.0	–	5.5	V	
Digital ground voltage	V_{SS} SR	–	0	–	V	

- 1) To ensure the stability of the voltage regulators the EVRs must be buffered with ceramic capacitors. Separate buffer capacitors with the recommended values shall be connected as close as possible to each V_{DDIM} and V_{DDI1} pin to keep the resistance of the board tracks below 2 Ohm. Connect all V_{DDI1} pins together. The minimum capacitance value is required for proper operation under all conditions (e.g. temperature). Higher values slightly increase the startup time.
- 2) Use one Capacitor for each pin.
- 3) This is the reference load. For bigger capacitive loads, use the derating factors listed in the PAD properties section.
- 4) The timing is valid for pin drivers operating in default current mode (selected after reset). Reducing the output current may lead to increased delays or reduced driving capability (C_L).
- 5) The operating frequency range may be reduced for specific device types. This is indicated in the device designation (...FxxL). 80 MHz devices are marked ...F80L.
- 6) Overload conditions occur if the standard operating conditions are exceeded, i.e. the voltage on any pin exceeds the specified range: $V_{OV} > V_{Ithmax}$ ($I_{OV} > 0$) or $V_{OV} < V_{Lmin}$ ($I_{OV} < 0$). The absolute sum of input overload currents on all pins may not exceed 50 mA. The supply voltages must remain within the specified limits. Proper operation under overload conditions depends on the application. Overload conditions must not occur on pin XTAL1 (powered by V_{DDIM}).

Electrical Parameters

- 7) An overload current (I_{OV}) through a pin injects a certain error current (I_{INJ}) into the adjacent pins. This error current adds to the respective pins leakage current (I_{OZ}). The amount of error current depends on the overload current and is defined by the overload coupling factor K_{OV} . The polarity of the injected error current is inverse compared to the polarity of the overload current that produces it. The total current through a pin is $|I_{TOT}| = |I_{OZ}| + (|I_{OV}| K_{OV})$. The additional error current may distort the input voltage on analog inputs.
- 8) Value is controlled by on-chip regulator

4.2 Voltage Range definitions

The XC228xH timing depends on the supply voltage. If such a dependency exists the timing values are given for 2 voltage areas commonly used. The voltage areas are defined in the following tables.

Table 14 Upper Voltage Range Definition

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Digital supply voltage for IO pads and voltage regulators	V_{DDP} SR	4.5	5	5.5	V	

Table 15 Lower Voltage Range Definition

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Digital supply voltage for IO pads and voltage regulators	V_{DDP} SR	3.0	3.3	4.5	V	

4.2.1 Parameter Interpretation

The parameters listed in the following include both the characteristics of the XC228xH and its demands on the system. To aid in correctly interpreting the parameters when evaluating them for a design, they are marked accordingly in the column "Symbol":

CC (Controller Characteristics):

The logic of the XC228xH provides signals with the specified characteristics.

SR (System Requirement):

The external system must provide signals with the specified characteristics to the XC228xH.

4.3 DC Parameters

These parameters are static or average values that may be exceeded during switching transitions (e.g. output current).

The XC228xH can operate within a wide supply voltage range from 3.0 V to 5.5 V. However, during operation this supply voltage must remain within 10 percent of the selected nominal supply voltage. It cannot vary across the full operating voltage range.

Because of the supply voltage restriction and because electrical behavior depends on the supply voltage, the parameters are specified separately for the upper and the lower voltage range.

During operation, the supply voltages may only change with a maximum speed of $dV/dt < 1 \text{ V/ms}$.

Leakage current is strongly dependent on the operating temperature and the voltage level at the respective pin. The maximum values in the following tables apply under worst case conditions, i.e. maximum temperature and an input level equal to the supply voltage.

The value for the leakage current in an application can be determined by using the respective leakage derating formula (see tables) with values from that application.

The pads of the XC228xH are designed to operate in various driver modes. The DC parameter specifications refer to the pad current limits specified in [Section 4.7.4](#).

Pullup/Pulldown Device Behavior

Most pins of the XC228xH feature pullup or pulldown devices. For some special pins these are fixed; for the port pins they can be selected by the application.

The specified current values indicate how to load the respective pin depending on the intended signal level. **Figure 14** shows the current paths.

The shaded resistors shown in the figure may be required to compensate system pull currents that do not match the given limit values.

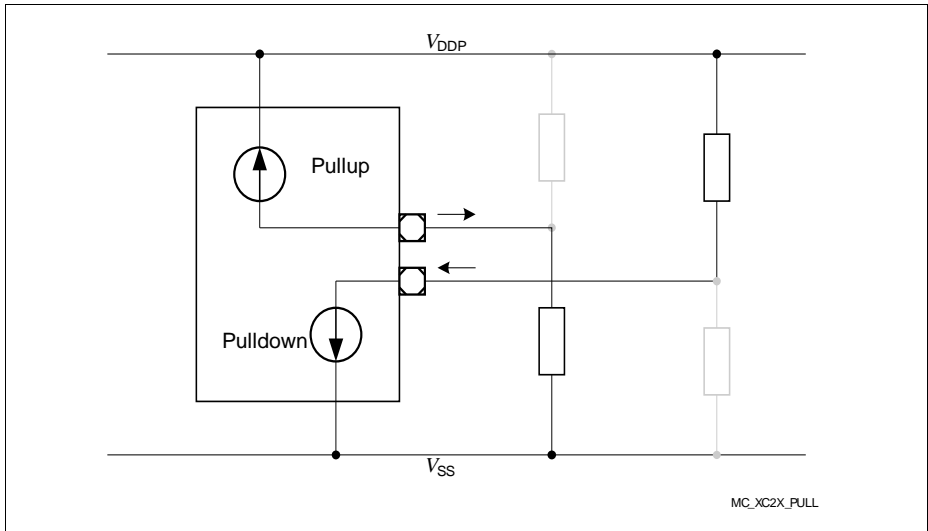


Figure 14 Pullup/Pulldown Current Definition

4.3.1 DC Parameters for Upper Voltage Area

Keeping signal levels within the limits specified in this table ensures operation without overload conditions. For signal levels outside these specifications, also refer to the specification of the overload current I_{OV} .

Note: Operating Conditions apply.

Table 16 is valid under the following conditions: $V_{DDP,typ.} 5\text{ V}$; $V_{DDP} \geq 4.5\text{ V}$; $V_{DDP} \leq 5.5\text{ V}$

Table 16 DC Characteristics for Upper Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Pin capacitance (digital inputs/outputs). To be doubled for double bond pins. ¹⁾	$C_{IO\ CC}$	–	–	10	pF	not subject to production test
Input Hysteresis ²⁾	HYS CC	$0.11 \times V_{DDP}$	–	–	V	$R_S = 0\text{ Ohm}$
Absolute input leakage current on pins of analog ports ³⁾	$ I_{OZ1} $ CC	–	10	200	nA	$V_{IN} > V_{SS}\text{ V}$; $V_{IN} < V_{DDP}$
Absolute input leakage current for all other pins. To be doubled for double bond pins. ³⁾¹⁾⁴⁾	$ I_{OZ2} $ CC	–	0.2	5	μA	$T_J \leq 110\text{ }^\circ\text{C}$; $V_{IN} > V_{SS}\text{ V}$; $V_{IN} < V_{DDP}$
		–	0.2	15	μA	$T_J \leq 150\text{ }^\circ\text{C}$; $V_{IN} > V_{SS}\text{ V}$; $V_{IN} < V_{DDP}$
Pull Level Force Current ⁵⁾	$ I_{PLF} $ SR	250	–	–	μA	$V_{IN} \geq V_{IHmin}(pull_down_enabled)$; $V_{IN} \leq V_{ILmax}(pull_up_enabled)$
Pull Level Keep Current ⁶⁾	$ I_{PLK} $ SR	–	–	30	μA	$V_{IN} \geq V_{IHmin}(pull_up_enabled)$; $V_{IN} \leq V_{ILmax}(pull_down_enabled)$
Input high voltage (all except XTAL1)	V_{IH} SR	$0.7 \times V_{DDP}$	–	$V_{DDP} + 0.3$	V	
Input low voltage (all except XTAL1)	V_{IL} SR	-0.3	–	$0.3 \times V_{DDP}$	V	

Electrical Parameters

Table 16 DC Characteristics for Upper Voltage Range (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Output High voltage ⁷⁾	V_{OH} CC	$V_{DDP} - 1.0$	—	—	V	$I_{OH} \geq I_{OHmax}$
		$V_{DDP} - 0.4$	—	—	V	$I_{OH} \geq I_{OHnom}$ ⁸⁾
Output Low Voltage ⁷⁾	V_{OL} CC	—	—	0.4	V	$I_{OL} \leq I_{OLnom}$ ⁹⁾
		—	—	1.0	V	$I_{OL} \leq I_{OLmax}$

- 1) Because each double bond pin is connected to two pads (standard pad and high-speed pad), it has twice the normal value. For a list of affected pins refer to the pin definitions table in chapter 2.
- 2) Not subject to production test - verified by design/characterization. Hysteresis is implemented to avoid metastable states and switching due to internal ground bounce. It cannot suppress switching due to external system noise under all conditions.
- 3) If the input voltage exceeds the respective supply voltage due to ground bouncing ($V_{IN} < V_{SS}$) or supply ripple ($V_{IN} > V_{DDP}$), a certain amount of current may flow through the protection diodes. This current adds to the leakage current. An additional error current (I_{INJ}) will flow if an overload current flows through an adjacent pin. Please refer to the definition of the overload coupling factor K_{OV} .
- 4) The given values are worst-case values. In production test, this leakage current is only tested at 125 °C; other values are ensured by correlation. For derating, please refer to the following descriptions: Leakage derating depending on temperature (T_J = junction temperature [°C]): $I_{OZ} = 0.05 \times e^{(1.5 + 0.028 \times T_J)}$ [μA]. For example, at a temperature of 95 °C the resulting leakage current is 3.2 μA. Leakage derating depending on voltage level ($DV = V_{DDP} - V_{PIN}$ [V]): $I_{OZ} = I_{OZtempmax} - (1.6 \times DV)$ (μA). This voltage derating formula is an approximation which applies for maximum temperature.
- 5) Drive the indicated minimum current through this pin to change the default pin level driven by the enabled pull device.
- 6) Limit the current through this pin to the indicated value so that the enabled pull device can keep the default pin level.
- 7) The maximum deliverable output current of a port driver depends on the selected output driver mode. This specification is not valid for outputs which are switched to open drain mode. In this case the respective output will float and the voltage is determined by the external circuit.
- 8) As a rule, with decreasing output current the output levels approach the respective supply level ($V_{OL} \rightarrow V_{SS}$, $V_{OH} \rightarrow V_{DDP}$). However, only the levels for nominal output currents are verified.
- 9) As a rule, with decreasing output current the output levels approach the respective supply level ($V_{OL} \rightarrow V_{SS}$, $V_{OH} \rightarrow V_{DDP}$). However, only the levels for nominal output currents are verified.

4.3.2 DC Parameters for Lower Voltage Area

Keeping signal levels within the limits specified in this table ensures operation without overload conditions. For signal levels outside these specifications, also refer to the specification of the overload current I_{OV} .

Note: Operating Conditions apply.

Table 17 is valid under the following conditions: V_{DDP} typ. 3.3 V; $V_{DDP} \geq 3.0$ V; $V_{DDP} \leq 4.5$ V

Table 17 DC Characteristics for Lower Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Pin capacitance (digital inputs/outputs). To be doubled for double bond pins. ¹⁾	C_{IO} CC	–	–	10	pF	not subject to production test
Input Hysteresis ²⁾	HYS CC	0.07 x V_{DDP}	–	–	V	$R_S = 0$ Ohm
Absolute input leakage current on pins of analog ports ³⁾	$ I_{OZ1} $ CC	–	10	200	nA	$V_{IN} > V_{SS}$ V; $V_{IN} < V_{DDP}$
Absolute input leakage current for all other pins. To be doubled for double bond pins. ³⁾¹⁾⁴⁾	$ I_{OZ2} $ CC	–	0.2	2.5	μ A	$T_J \leq 110$ °C; $V_{IN} > V_{SS}$ V; $V_{IN} < V_{DDP}$
		–	0.2	8	μ A	$T_J \leq 150$ °C; $V_{IN} > V_{SS}$ V; $V_{IN} < V_{DDP}$
Pull Level Force Current ⁵⁾	$ I_{PLF} $ SR	150	–	–	μ A	$V_{IN} \geq V_{IHmin}$ (pull down_enabled) ; $V_{IN} \leq V_{ILmax}$ (pull up_enabled)
Pull Level Keep Current ⁶⁾	$ I_{PLK} $ SR	–	–	10	μ A	$V_{IN} \geq V_{IHmin}$ (pull up_enabled) ; $V_{IN} \leq V_{ILmax}$ (pull down_enabled)
Input high voltage (all except XTAL1)	V_{IH} SR	0.7 x V_{DDP}	–	$V_{DDP} + 0.3$	V	

Electrical Parameters

Table 17 DC Characteristics for Lower Voltage Range (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Input low voltage (all except XTAL1)	V_{IL} SR	-0.3	–	$0.3 \times V_{DDP}$	V	
Output High voltage ⁷⁾	V_{OH} CC	$V_{DDP} - 1.0$	–	–	V	$I_{OH} \geq I_{OHmax}$
		$V_{DDP} - 0.4$	–	–	V	$I_{OH} \geq I_{OHnom}$ ⁸⁾
Output Low Voltage ⁷⁾	V_{OL} CC	–	–	0.4	V	$I_{OL} \leq I_{OLnom}$ ⁹⁾
		–	–	1.0	V	$I_{OL} \leq I_{OLmax}$

- 1) Because each double bond pin is connected to two pads (standard pad and high-speed pad), it has twice the normal value. For a list of affected pins refer to the pin definitions table in chapter 2.
- 2) Not subject to production test - verified by design/characterization. Hysteresis is implemented to avoid metastable states and switching due to internal ground bounce. It cannot suppress switching due to external system noise under all conditions.
- 3) If the input voltage exceeds the respective supply voltage due to ground bouncing ($V_{IN} < V_{SS}$) or supply ripple ($V_{IN} > V_{DDP}$), a certain amount of current may flow through the protection diodes. This current adds to the leakage current. An additional error current (I_{INJ}) will flow if an overload current flows through an adjacent pin. Please refer to the definition of the overload coupling factor K_{OV} .
- 4) The given values are worst-case values. In production test, this leakage current is only tested at 125 °C; other values are ensured by correlation. For derating, please refer to the following descriptions: Leakage derating depending on temperature (T_J = junction temperature [°C]): $I_{OZ} = 0.05 \times e^{(1.5 + 0.028 \times T_J)}$ [μA]. For example, at a temperature of 95 °C the resulting leakage current is 3.2 μA. Leakage derating depending on voltage level ($DV = V_{DDP} - V_{PIN}$ [V]): $I_{OZ} = I_{OZtempmax} - (1.6 \times DV)$ (μA). This voltage derating formula is an approximation which applies for maximum temperature.
- 5) Drive the indicated minimum current through this pin to change the default pin level driven by the enabled pull device.
- 6) Limit the current through this pin to the indicated value so that the enabled pull device can keep the default pin level.
- 7) The maximum deliverable output current of a port driver depends on the selected output driver mode. This specification is not valid for outputs which are switched to open drain mode. In this case the respective output will float and the voltage is determined by the external circuit.
- 8) As a rule, with decreasing output current the output levels approach the respective supply level (VOL->VSS, VOH->VDDP). However, only the levels for nominal output currents are verified.
- 9) As a rule, with decreasing output current the output levels approach the respective supply level (V_{OL} -> V_{SS} , V_{OH} -> V_{DDP}). However, only the levels for nominal output currents are verified.

4.3.3 Power Consumption

The power consumed by the XC228xH depends on several factors such as supply voltage, operating frequency, active circuits, and operating temperature. The power consumption specified here consists of two components:

- The switching current I_S depends on the device activity
- The leakage current I_{LK} depends on the device temperature

To determine the actual power consumption, always both components, switching current I_S and leakage current I_{LK} must be added:

$$I_{DDP} = I_S + I_{LK}$$

Note: The power consumption values are not subject to production test. They are verified by design/characterization.

To determine the total power consumption for dimensioning the external power supply, also the pad driver currents must be considered.

The given power consumption parameters and their values refer to specific operating conditions:

- **Active mode:**
Regular operation, i.e. peripherals are active, code execution out of Flash.
- **Stopover mode:**
Crystal oscillator and PLL stopped, Flash switched off, clock in domain DMP_1 stopped.
- **Standby mode:**
Voltage domain DMP_1 switched off completely, power supply control switched off. DMP_M domain is supplied by ultra low power electronic voltage regulator (ULPEVR). The alternate regulator EVR_M is switched off.

Note: The maximum values cover the complete specified operating range of all manufactured devices.

The typical values refer to average devices under typical conditions, such as nominal supply voltage, room temperature, application-oriented activity.

After a power reset, the decoupling capacitors for V_{DDIM} and V_{DDI1} are charged with the maximum possible current.

For additional information, please refer to [Section 5.2, Thermal Considerations](#).

Note: Operating Conditions apply.

Table 18 Switching Power Consumption

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Power supply current (active) with all peripherals active and EVVRs on ¹⁾	I_{SACT_FR} CC	–	40 + 0.9 x f_{SYS} ²⁾	40 + 1.4 x f_{SYS} ²⁾	mA	power_mode= active ; voltage_range= both ³⁾⁴⁾⁵⁾
Power supply current in standby mode	I_{SSB} CC	–	95	200	μA	power_mode= standby ; voltage_range= lower ⁶⁾
		–	120	330	μA	power_mode= standby ; voltage_range= upper ⁶⁾
Power supply current in stopover mode, EVVRs on	I_{SSO} CC	–	1.4	4.0	mA	power_mode= stopover ; voltage_range= both ⁴⁾

1) If the FlexRay module clock (running @ 80 MHz) is turned off (KSCFG.MODEN=0) the value is reduced by 15 mA.

2) f_{SYS} in MHz

3) The pad supply voltage pins (V_{DDPB}) provide the input current for the on-chip EVVRs and the current consumed by the pin output drivers. A small current is consumed because the drivers input stages are switched. In Fast Startup Mode (with the Flash modules deactivated), the typical current is reduced to 3 + 0.6 x f_{SYS} .

4) Please consider the additional conditions described in section "Active Mode Power Supply Current".

5) The pad supply voltage has only a minor influence on this parameter.

6) These values are valid if the voltage validation circuits for V_{DDPB} (SWD) and V_{DDIM} (PVC_M) are off. Leaving SWD and PVC_M active adds another 90 μA.

Active Mode Power Supply Current

The actual power supply current in active mode not only depends on the system frequency but also on the configuration of the XC228xH's subsystem.

Besides the power consumed by the device logic the power supply pins also provide the current that flows through the pin output drivers.

A small current is consumed because the drivers' input stages are switched.

The IO power domains can be supplied separately. Power domain A (V_{DDPA}) supplies the A/D converters and Port 6. Power domain B (V_{DDPB}) supplies the on-chip EVVRs and all other ports.

Electrical Parameters

During operation domain A draws a maximum current of 1.5 mA for each active A/D converter module from V_{DDPA} .

In Fast Startup Mode (with the Flash modules deactivated), the typical current is reduced to $3 + 0.6 \times f_{SYS}$ mA.

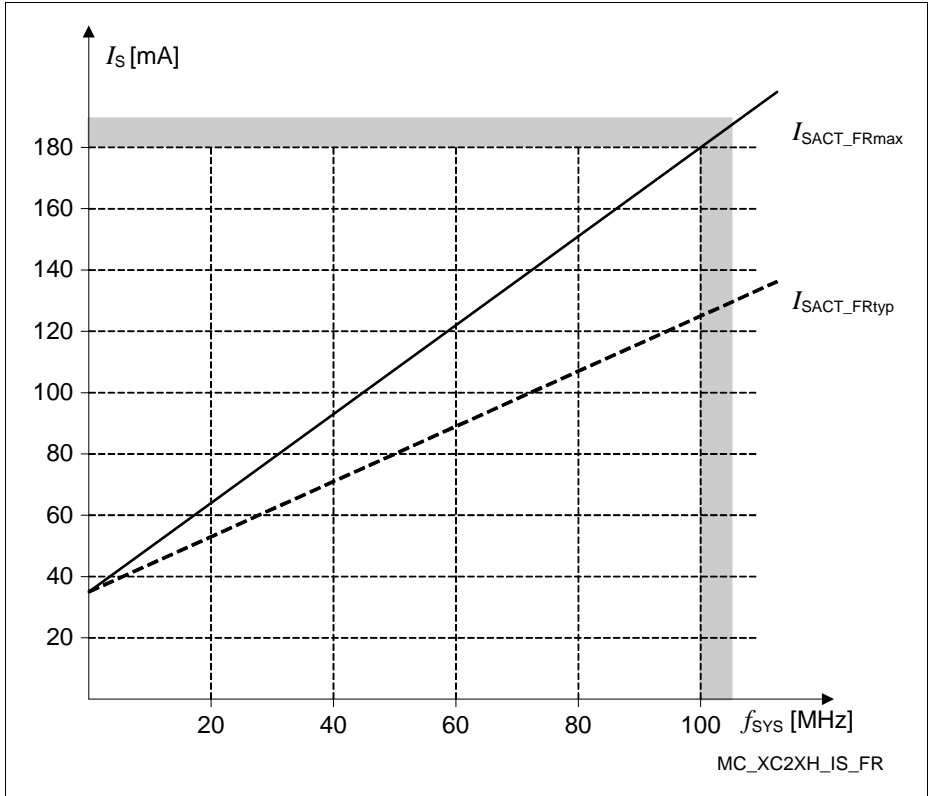


Figure 15 Supply Current in Active Mode as a Function of Frequency

Note: Operating Conditions apply.

Table 19 Leakage Power Consumption

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Leakage supply current (DMP_1 off) ¹⁾²⁾	I_{LK0} CC	–	20	35	μA	$T_J = 25\text{ }^\circ\text{C}$
		–	115	330	μA	$T_J = 85\text{ }^\circ\text{C}$
		–	270	880	μA	$T_J = 125\text{ }^\circ\text{C}$
		–	420	1.445	μA	$T_J = 150\text{ }^\circ\text{C}$
Leakage supply current (DMP_1 powered) ¹⁾²⁾	I_{LK1} CC	–	0.04	0.06	mA	$T_J = 25\text{ }^\circ\text{C}$
		–	0.7	1.8	mA	$T_J = 85\text{ }^\circ\text{C}$
		–	3.1	8.6	mA	$T_J = 125\text{ }^\circ\text{C}$
		–	6.6	19.2	mA	$T_J = 150\text{ }^\circ\text{C}$

- 1) The supply current caused by leakage depends mainly on the junction temperature and the supply voltage. The temperature difference between the junction temperature T_J and the ambient temperature T_A must be taken into account. As this fraction of the supply current does not depend on device activity, it must be added to other power consumption values.
- 2) All inputs (including pins configured as inputs) are set at 0 V to 0.1 V or at $V_{DDP} - 0.1$ V to V_{DDP} and all outputs (including pins configured as outputs) are disconnected.

Note: A fraction of the leakage current flows through domain DMP_A (pin V_{DDPA}). This current can be calculated as $7,000 \times e^{-\alpha}$, with $\alpha = 5000 / (273 + 1.3 \times T_J)$. For $T_J = 150^\circ\text{C}$, this results in a current of 160 μA .

The leakage power consumption can be calculated according to the following formulas:

$$I_{LK0} = 500,000 \times e^{-\alpha} \text{ with } \alpha = 3000 / (273 + B \times T_J)$$

Parameter B must be replaced by

- 1.0 for typical values
- 1.6 for maximum values

$$I_{LK1} = 600,000 \times e^{-\alpha} \text{ with } \alpha = 5000 / (273 + B \times T_J)$$

Parameter B must be replaced by

- 1.1 for typical values
- 1.4 for maximum values

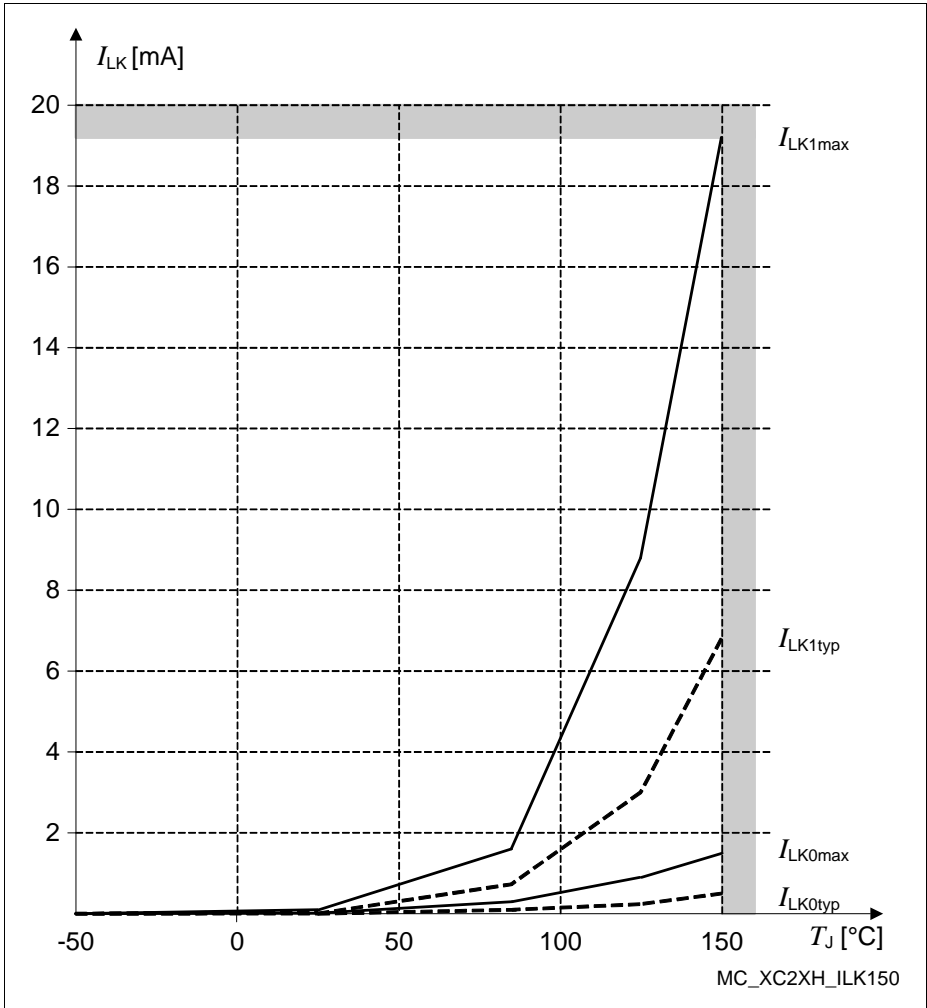


Figure 16 Leakage Supply Current as a Function of Temperature

4.4 Analog/Digital Converter Parameters

These parameters describe the conditions for optimum ADC performance.

Note: Operating Conditions apply.

Table 20 ADC Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Switched capacitance at an analog input	C_{AINSW} CC	–	–	6	pF	not subject to production test
Total capacitance at an analog input	C_{AINT} CC	–	–	14	pF	not subject to production test
Switched capacitance at the reference input	C_{AREFSW} CC	–	–	10	pF	not subject to production test
Total capacitance at the reference input	C_{AREFT} CC	–	–	21	pF	not subject to production test
Differential Non-Linearity Error	$ EA_{DNL} $ CC	–	0.8	1	LSB	
Gain Error	$ EA_{GAIN} $ CC	–	0.4	0.8	LSB	
Integral Non-Linearity	$ EA_{INL} $ CC	–	0.8	1.2	LSB	
Offset Error	$ EA_{OFF} $ CC	–	0.5	0.8	LSB	
Analog clock frequency	f_{ADCI} SR	0.5	–	16.5	MHz	voltage_range=lower
		0.5	–	20	MHz	voltage_range=upper
Input resistance of the selected analog channel	R_{AIN} CC	–	–	2	kOh m	not subject to production test
Input resistance of the reference input	R_{AREF} CC	–	–	2	kOh m	not subject to production test

Table 20 ADC Parameters (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Broken wire detection delay against VAGND ¹⁾	t_{BWG} CC	–	–	50 ²⁾		
Broken wire detection delay against VAREF ¹⁾	t_{BWR} CC	–	–	50 ³⁾		
Conversion time for 8-bit result ¹⁾	t_{c8} CC	(11 + STC) x t_{ADCl} + 2 x t_{SYS}	–	–		
Conversion time for 10-bit result ¹⁾	t_{c10} CC	(13 + STC) x t_{ADCl} + 2 x t_{SYS}	–	–		
Total Unadjusted Error	TUE CC	–	1	2	LSB	⁴⁾
Wakeup time from analog powerdown, fast mode	t_{WAF} CC	–	–	4	μs	
Wakeup time from analog powerdown, slow mode	t_{WAS} CC	–	–	15	μs	
Analog reference ground	V_{AGND} SR	$V_{\text{SS}} - 0.05$	–	1.5	V	
Analog input voltage range	V_{AIN} SR	V_{AGND}	–	V_{AREF}	V	⁵⁾
Analog reference voltage	V_{AREF} SR	$V_{\text{AGND}} + 1.0$	–	$V_{\text{DDPA}} + 0.05$	V	

1) This parameter includes the sample time (also the additional sample time specified by STC), the time to determine the digital result and the time to load the result register with the conversion result. Values for the basic clock t_{ADCl} depend on programming.

2) The broken wire detection delay against V_{AGND} is measured in numbers of consecutive precharge cycles at a conversion rate of not more than 500 μs . Result below 10% (66_H)

3) The broken wire detection delay against V_{AREF} is measured in numbers of consecutive precharge cycles at a conversion rate of not more than 10 μs . This function is influenced by leakage current, in particular at high temperature. Result above 80% (332_H)

Electrical Parameters

- 4) TUE is tested at $V_{AREF} = V_{DDPA} = 5.0\text{ V}$, $V_{AGND} = 0\text{ V}$. It is verified by design for all other voltages within the defined voltage range. The specified TUE is valid only if the absolute sum of input overload currents on analog port pins (see I_{OV} specification) does not exceed 10 mA, and if V_{AREF} and V_{AGND} remain stable during the measurement time.
- 5) V_{AIN} may exceed V_{AGND} or V_{AREF} up to the absolute maximum ratings. However, the conversion result in these cases will be $X000_H$ or $X3FF_H$, respectively.

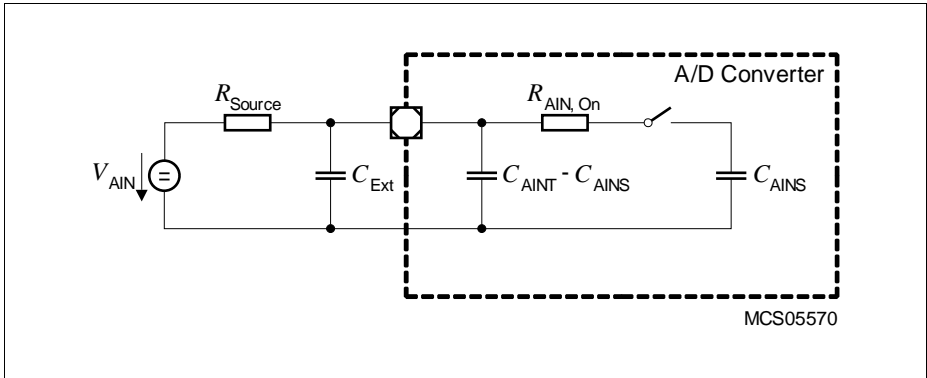


Figure 17 Equivalent Circuitry for Analog Inputs

Electrical Parameters

Sample time and conversion time of the XC228xH's A/D converters are programmable. The timing above can be calculated using **Table 21**.

The limit values for f_{ADCl} must not be exceeded when selecting the prescaler value.

Table 21 A/D Converter Computation Table

GLOBCTR.5-0 (DIVA)	A/D Converter Analog Clock f_{ADCl}	INPCRx.7-0 (STC)	Sample Time¹⁾ t_s
000000 _B	f_{SYS}	00 _H	$t_{\text{ADCl}} \times 2$
000001 _B	$f_{\text{SYS}} / 2$	01 _H	$t_{\text{ADCl}} \times 3$
000010 _B	$f_{\text{SYS}} / 3$	02 _H	$t_{\text{ADCl}} \times 4$
:	$f_{\text{SYS}} / (\text{DIVA}+1)$:	$t_{\text{ADCl}} \times (\text{STC}+2)$
111110 _B	$f_{\text{SYS}} / 63$	FE _H	$t_{\text{ADCl}} \times 256$
111111 _B	$f_{\text{SYS}} / 64$	FF _H	$t_{\text{ADCl}} \times 257$

1) The selected sample time is doubled if broken wire detection is active (due to the presampling phase).

Converter Timing Example A:

Assumptions: $f_{\text{SYS}} = 100 \text{ MHz}$ (i.e. $t_{\text{SYS}} = 10 \text{ ns}$), DIVA = 03_H, STC = 00_H

Analog clock $f_{\text{ADCl}} = f_{\text{SYS}} / 4 = 25 \text{ MHz}$, i.e. $t_{\text{ADCl}} = 40 \text{ ns}$

Sample time $t_s = t_{\text{ADCl}} \times 2 = 80 \text{ ns}$

Conversion 10-bit:

$$t_{\text{C10}} = 13 \times t_{\text{ADCl}} + 2 \times t_{\text{SYS}} = 13 \times 40 \text{ ns} + 2 \times 10 \text{ ns} = 0.540 \mu\text{s}$$

Conversion 8-bit:

$$t_{\text{C8}} = 11 \times t_{\text{ADCl}} + 2 \times t_{\text{SYS}} = 11 \times 40 \text{ ns} + 2 \times 10 \text{ ns} = 0.460 \mu\text{s}$$

Converter Timing Example B:

Assumptions: $f_{\text{SYS}} = 40 \text{ MHz}$ (i.e. $t_{\text{SYS}} = 25 \text{ ns}$), DIVA = 02_H, STC = 03_H

Analog clock $f_{\text{ADCl}} = f_{\text{SYS}} / 3 = 13.3 \text{ MHz}$, i.e. $t_{\text{ADCl}} = 75 \text{ ns}$

Sample time $t_s = t_{\text{ADCl}} \times 5 = 375 \text{ ns}$

Conversion 10-bit:

$$t_{\text{C10}} = 16 \times t_{\text{ADCl}} + 2 \times t_{\text{SYS}} = 16 \times 75 \text{ ns} + 2 \times 25 \text{ ns} = 1.25 \mu\text{s}$$

Conversion 8-bit:

$$t_{\text{C8}} = 14 \times t_{\text{ADCl}} + 2 \times t_{\text{SYS}} = 14 \times 75 \text{ ns} + 2 \times 25 \text{ ns} = 1.10 \mu\text{s}$$

4.5 System Parameters

The following parameters specify several aspects which are important when integrating the XC228xH into an application system.

Note: These parameters are not subject to production test but verified by design and/or characterization.

Note: Operating Conditions apply.

Table 22 Various System Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Short-term deviation of internal clock source frequency ¹⁾	Δf_{INT} CC	-1	–	1	%	$\Delta T_J \leq 10^\circ\text{C}$
Internal clock source frequency	f_{INT} CC	4.8	5.0	5.2	MHz	
Wakeup clock source frequency ²⁾	f_{WU} CC	400	–	700	kHz	FREQSEL= 00
		210	–	390	kHz	FREQSEL= 01
		140	–	260	kHz	FREQSEL= 10
		110	–	200	kHz	FREQSEL= 11
Startup time from power-on with code execution from Flash	t_{SPO} CC	1.9	2.6	3.2	ms	$f_{\text{WU}}= 500 \text{ kHz}$
Startup time from standby mode with code execution from Flash	t_{SSB} CC	3.2	4.2	4.9	ms	$f_{\text{WU}}= 140 \text{ kHz}$
		2.0	2.7	3.3	ms	$f_{\text{WU}}= 500 \text{ kHz}$
Startup time from stopover mode with code execution from PSRAM	t_{SSO} CC	11 / f_{WU} ³⁾	–	12 / f_{WU} ³⁾	μs	
Core voltage (PVC) supervision level	V_{PVC} CC	$V_{\text{LV}} - 0.03$	V_{LV}	$V_{\text{LV}} + 0.07$ ⁴⁾	V	⁵⁾
Supply watchdog (SWD) supervision level	V_{SWD} CC	$V_{\text{LV}} - 0.10$ ⁶⁾	V_{LV}	$V_{\text{LV}} + 0.15$	V	voltage_range= lower ⁵⁾
		$V_{\text{LV}} - 0.15$	V_{LV}	$V_{\text{LV}} + 0.15$	V	voltage_range= upper ⁵⁾

1) The short-term frequency deviation refers to a timeframe of a few hours and is measured relative to the current frequency at the beginning of the respective timeframe. This parameter is useful to determine a time span for re-triggering a LIN synchronization.

Electrical Parameters

- 2) This parameter is tested for the fastest and the slowest selection. The medium selections are not subject to production test - verified by design/characterization
- 3) f_{WU} in MHz
- 4) This value includes a hysteresis of approximately 50 mV for rising voltage.
- 5) V_{LV} = selected SWD voltage level
- 6) The limit $V_{LV} - 0.10$ V is valid for the OK1 level. The limit for the OK2 level is $V_{LV} - 0.15$ V.

Conditions for t_{SPO} Timing Measurement

The time required for the transition from **Power-on** to **Base** mode is called t_{SPO} . It is measured under the following conditions:

Precondition: The pad supply is valid, i.e. V_{DDPB} is above 3.0V and remains above 3.0V even though the XC228xH is starting up. No debugger is attached.

Start condition: Power-on reset is removed ($\overline{PORST} = 1$).

End condition: External pin toggle caused by first user instruction executed from FLASH after startup.

Conditions for t_{SSB} Timing Measurement

The time required for the transition from **Standby** to **Base** mode is called t_{SSB} . It is measured under the following conditions:

Precondition: The **Standby** mode has been entered using the procedure defined in the Programmer's Guide.

Start condition: Pin toggle on \overline{ESR} pin triggering the startup sequence.

End condition: External pin toggle caused by first user instruction executed from FLASH after startup.

Conditions for t_{SSO} Timing Measurement

The time required for the transition from **Stopover** to **Stopover Waked-Up** mode is called t_{SSO} . It is measured under the following conditions:

Precondition: The **Stopover** mode has been entered using the procedure defined in the Programmer's Guide.

Start condition: Pin toggle on \overline{ESR} pin triggering the startup sequence.

End condition: External pin toggle caused by first user instruction executed from PSRAM after startup.

Coding of bit fields LEVxV in SWD and PVC Configuration Registers

Table 23 Coding of bit fields LEVxV in Register SWDCON0

Code	Default Voltage Level	Notes ¹⁾
0000 _B	2.9 V	
0001 _B	3.0 V	LEV1V: reset request
0010 _B	3.1 V	
0011 _B	3.2 V	
0100 _B	3.3 V	
0101 _B	3.4 V	
0110 _B	3.6 V	
0111 _B	4.0 V	
1000 _B	4.2 V	
1001 _B	4.5 V	LEV2V: no request
1010 _B	4.6 V	
1011 _B	4.7 V	
1100 _B	4.8 V	
1101 _B	4.9 V	
1110 _B	5.0 V	
1111 _B	5.5 V	

1) The indicated default levels are selected automatically after a power reset.

Table 24 Coding of bit fields LEVxV in Registers PVCyCONz

Code	Default Voltage Level	Notes ¹⁾
000 _B	0.95 V	
001 _B	1.05 V	
010 _B	1.15 V	
011 _B	1.25 V	
100 _B	1.35 V	LEV1V: reset request
101 _B	1.45 V	LEV2V: interrupt request ²⁾
110 _B	1.55 V	
111 _B	1.65 V	

1) The indicated default levels are selected automatically after a power reset.

Electrical Parameters

- 2) Due to variations of the tolerance of both the Embedded Voltage Regulators (EVR) and the PVC levels, this interrupt can be triggered inadvertently, even though the core voltage is within the normal range. It is, therefore, recommended not to use this warning level.

4.6 Flash Memory Parameters

The XC228xH is delivered with all Flash sectors erased and with no protection installed. The data retention time of the XC228xH's Flash memory (i.e. the time after which stored data can still be retrieved) depends on the number of times the Flash memory has been erased and programmed.

Note: These parameters are not subject to production test but verified by design and/or characterization.

Note: Operating Conditions apply.

Table 25 Flash Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Parallel Flash module program/erase limit depending on Flash read activity	N_{PP} SR	–	–	7 ¹⁾		$N_{FL_RD} \leq 1$
		–	–	1 ²⁾		$N_{FL_RD} > 1$
Flash erase endurance for security pages	N_{SEC} SR	10	–	–	cycles	$t_{RET} \geq 20$ years
Flash wait states ³⁾	N_{WSFLAS} H SR	1	–	–		$f_{SYS} \leq 8$ MHz
		2	–	–		$f_{SYS} \leq 13$ MHz
		3	–	–		$f_{SYS} \leq 17$ MHz
		4	–	–		$f_{SYS} > 17$ MHz
Flash wait state extension ⁴⁾	N_{WSFLE} SR	0	–	–		$f_{SYS} \leq 80$ MHz
		1	–	–		$f_{SYS} > 80$ MHz
Erase time per sector/page	t_{ER} CC	–	7 ⁵⁾	8.0	ms	
Programming time per page	t_{PR} CC	–	3 ⁵⁾	3.5	ms	
Data retention time	t_{RET} CC	20	–	–	years	$N_{ER} \leq 1.000$ cycles
Drain disturb limit	N_{DD} SR	32	–	–	cycles	

Electrical Parameters

Table 25 Flash Parameters (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Number of erase cycles	N_{ER} SR	–	–	15.000	cycles	$t_{RET} \geq 5$ years; Valid for up to 64 user selected sectors (data storage)
		–	–	1.000	cycles	$t_{RET} \geq 20$ years

- 1) All Flash module(s) can be erased/programmed while code is executed and/or data is read from only one Flash module or from PSRAM. The Flash module that delivers code/data can, of course, not be erased/programmed.
- 2) Flash module 6 can be erased/programmed while code is executed and/or data is read from any other Flash module.
- 3) Value of IMB_IMBCTRL.WSFLASH.
- 4) Value of IMB_IMBCTRL.WSFLE.
- 5) Programming and erase times depend on the internal Flash clock source. The control state machine needs a few system clock cycles. This increases the stated durations noticeably only at extremely low system clock frequencies.

Access to the XC228xH Flash modules is controlled by the IMB. Built-in prefetch mechanisms optimize the performance for sequential access.

Flash access waitstates only affect non-sequential access. Due to prefetch mechanisms, the performance for sequential access (depending on the software structure) is only partially influenced by waitstates.

4.7 AC Parameters

These parameters describe the dynamic behavior of the XC228xH.

4.7.1 Testing Waveforms

These values are used for characterization and production testing (except pin XTAL1).

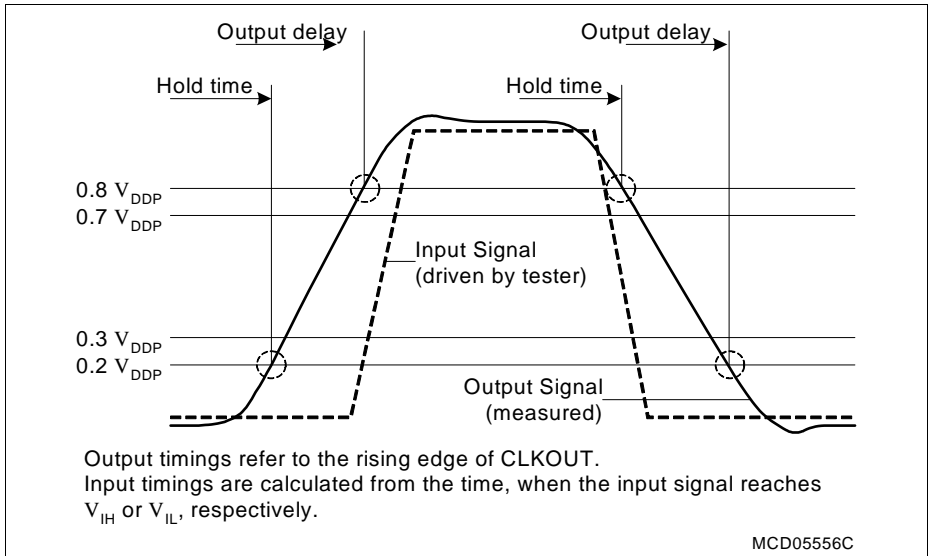


Figure 18 Input Output Waveforms

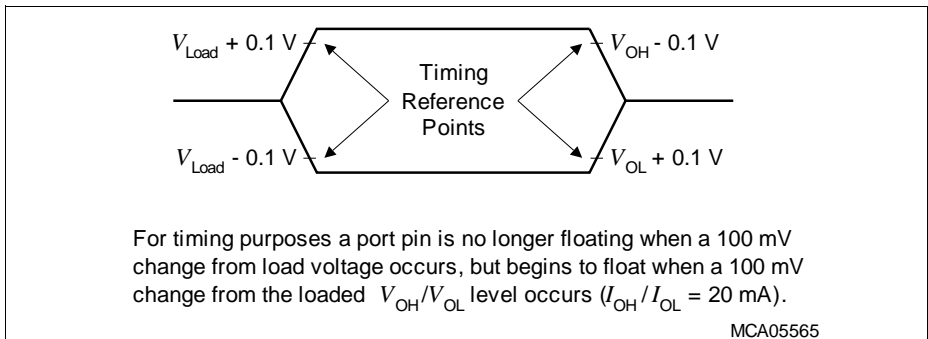


Figure 19 Floating Waveforms

4.7.2 Definition of Internal Timing

The internal operation of the XC228xH is controlled by the internal system clock f_{SYS} . Because the system clock signal f_{SYS} can be generated from a number of internal and external sources using different mechanisms, the duration of the system clock periods (TCS) and their variation (as well as the derived external timing) depend on the mechanism used to generate f_{SYS} . This must be considered when calculating the timing for the XC228xH.

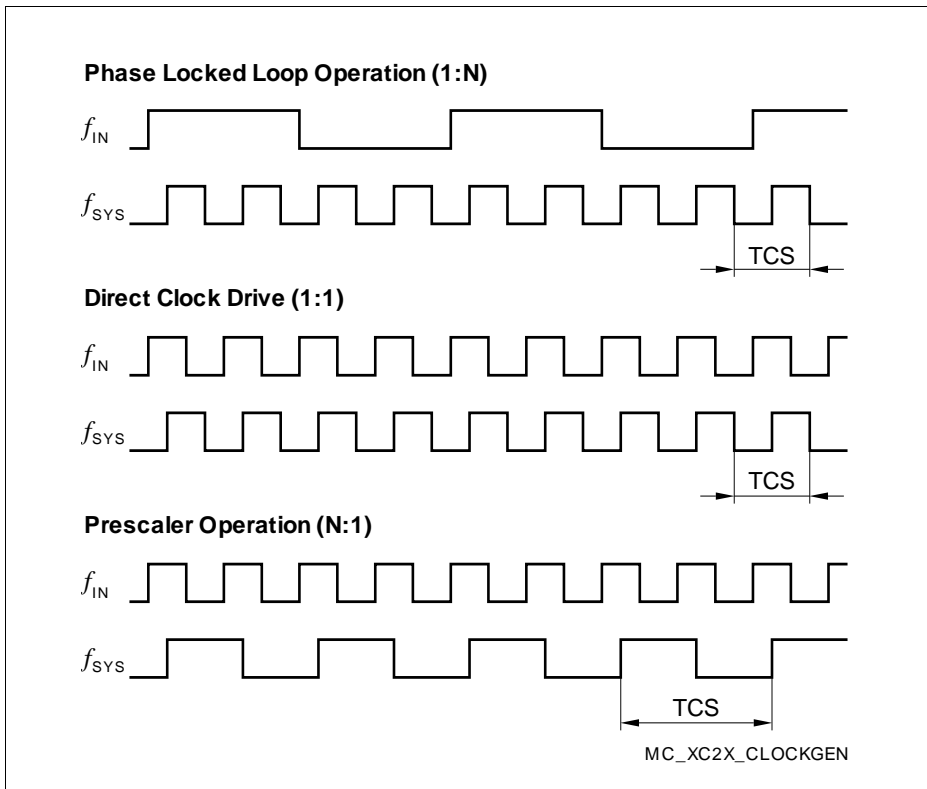


Figure 20 Generation Mechanisms for the System Clock

Note: The example of PLL operation shown in [Figure 20](#) uses a PLL factor of 1:4; the example of prescaler operation uses a divider factor of 2:1.

The specification of the external timing (AC Characteristics) depends on the period of the system clock (TCS).

Direct Drive

When direct drive operation is selected (SYSCON0.CLKSEL = 11_B), the system clock is derived directly from the input clock signal CLKIN1:

$$f_{\text{SYS}} = f_{\text{IN}}$$

The frequency of f_{SYS} is the same as the frequency of f_{IN} . In this case the high and low times of f_{SYS} are determined by the duty cycle of the input clock f_{IN} .

Selecting Bypass Operation from the XTAL1¹⁾ input and using a divider factor of 1 results in a similar configuration.

Prescaler Operation

When prescaler operation is selected (SYSCON0.CLKSEL = 10_B, PLLCON0.VCOBY = 1_B), the system clock is derived either from the crystal oscillator (input clock signal XTAL1) or from the internal clock source through the output prescaler K1 (= K1DIV+1):

$$f_{\text{SYS}} = f_{\text{OSC}} / K1.$$

If a divider factor of 1 is selected, the frequency of f_{SYS} equals the frequency of f_{OSC} . In this case the high and low times of f_{SYS} are determined by the duty cycle of the input clock f_{OSC} (external or internal).

The lowest system clock frequency results from selecting the maximum value for the divider factor K1:

$$f_{\text{SYS}} = f_{\text{OSC}} / 1024.$$

4.7.2.1 Phase Locked Loop (PLL)

When PLL operation is selected (SYSCON0.CLKSEL = 10_B, PLLCON0.VCOBY = 0_B), the on-chip phase locked loop is enabled and provides the system clock. The PLL multiplies the input frequency by the factor **F** ($f_{\text{SYS}} = f_{\text{IN}} \times \mathbf{F}$).

F is calculated from the input divider P (= PDIV+1), the multiplication factor N (= NDIV+1), and the output divider K2 (= K2DIV+1):

$$(\mathbf{F} = \mathbf{N} / (\mathbf{P} \times \mathbf{K2})).$$

The input clock can be derived either from an external source at XTAL1 or from the on-chip clock source.

The PLL circuit synchronizes the system clock to the input clock. This synchronization is performed smoothly so that the system clock frequency does not change abruptly.

Adjustment to the input clock continuously changes the frequency of f_{SYS} so that it is locked to f_{IN} . The slight variation causes a jitter of f_{SYS} which in turn affects the duration of individual TCSs.

1) Voltages on XTAL1 must comply to the core supply voltage V_{DDIM} .

Electrical Parameters

The timing in the AC Characteristics refers to TCSs. Timing must be calculated using the minimum TCS possible under the given circumstances.

The actual minimum value for TCS depends on the jitter of the PLL. Because the PLL is constantly adjusting its output frequency to correspond to the input frequency (from crystal or oscillator), the accumulated jitter is limited. This means that the relative deviation for periods of more than one TCS is lower than for a single TCS (see formulas and [Figure 21](#)).

This is especially important for bus cycles using waitstates and for the operation of timers, serial interfaces, etc. For all slower operations and longer periods (e.g. pulse train generation or measurement, lower baudrates, etc.) the deviation caused by the PLL jitter is negligible.

The value of the accumulated PLL jitter depends on the number of consecutive VCO output cycles within the respective timeframe. The VCO output clock is divided by the output prescaler K2 to generate the system clock signal f_{SYS} . The number of VCO cycles is $K2 \times T$, where **T** is the number of consecutive f_{SYS} cycles (TCS).

The maximum accumulated jitter (long-term jitter) D_{Tmax} is defined by:

$$D_{Tmax} [\text{ns}] = \pm(220 / (K2 \times f_{SYS}) + 4.3)$$

This maximum value is applicable, if either the number of clock cycles $T > (f_{SYS} / 1.2)$ or the prescaler value $K2 > 17$.

In all other cases for a timeframe of **T** × TCS the accumulated jitter D_T is determined by:

$$D_T [\text{ns}] = D_{Tmax} \times [(1 - 0.058 \times K2) \times (T - 1) / (0.83 \times f_{SYS} - 1) + 0.058 \times K2]$$

f_{SYS} in [MHz] in all formulas.

Example, for a period of 3 TCSs @ 33 MHz and $K2 = 4$:

$$D_{max} = \pm(220 / (4 \times 33) + 4.3) = 5.97 \text{ ns (Not applicable directly in this case!)}$$

$$\begin{aligned} D_3 &= 5.97 \times [(1 - 0.058 \times 4) \times (3 - 1) / (0.83 \times 33 - 1) + 0.058 \times 4] \\ &= 5.97 \times [0.768 \times 2 / 26.39 + 0.232] \\ &= 1.7 \text{ ns} \end{aligned}$$

Example, for a period of 3 TCSs @ 33 MHz and $K2 = 2$:

$$D_{max} = \pm(220 / (2 \times 33) + 4.3) = 7.63 \text{ ns (Not applicable directly in this case!)}$$

$$\begin{aligned} D_3 &= 7.63 \times [(1 - 0.058 \times 2) \times (3 - 1) / (0.83 \times 33 - 1) + 0.058 \times 2] \\ &= 7.63 \times [0.884 \times 2 / 26.39 + 0.116] \\ &= 1.4 \text{ ns} \end{aligned}$$

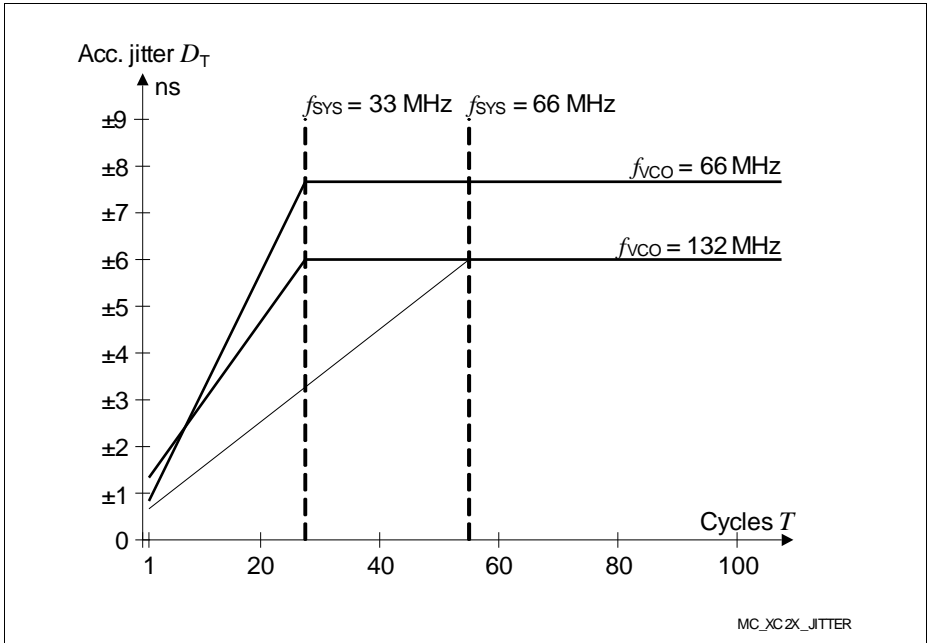


Figure 21 **Approximated Accumulated PLL Jitter**

Note: The specified PLL jitter values are valid if the capacitive load per pin does not exceed $C_L = 20$ pF.

The maximum peak-to-peak noise on the pad supply voltage (measured between V_{DDPB} pin 144 and V_{SS} pin 1) is limited to a peak-to-peak voltage of $V_{PP} = 50$ mV. This can be achieved by appropriate blocking of the supply voltage as close as possible to the supply pins and using PCB supply and ground planes.

PLL frequency band selection

Different frequency bands can be selected for the VCO so that the operation of the PLL can be adjusted to a wide range of input and output frequencies:

Table 26 System PLL Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
VCO output frequency	f_{VCO} CC	50	–	110	MHz	$VCOSSEL=00b$; $VCOmode=$ controlled
		10	–	40	MHz	$VCOSSEL=00b$; $VCOmode=$ free running
		100	–	200	MHz	$VCOSSEL=01b$; $VCOmode=$ controlled
		20	–	80	MHz	$VCOSSEL=01b$; $VCOmode=$ free running

FlexRay PLL Parameters

This PLL unit is dedicated for usage with the FlexRay module. FlexRay operation requires a high precision clock signal to fulfil the protocol timing requirements. If the system clock is generated by use of a quartz it is mandatory to adjust the quartz parameters in your target system. Please refer also to [Chapter 4.7.3](#) for details.

Table 27 FlexRay PLL Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
VCO output frequency	f_{VCA} CC	480	–	480	MHz	$VCOmode=$ con trolled
		80	–	340	MHz	$VCOmode=$ fre e running

4.7.2.2 Wakeup Clock

When wakeup operation is selected ($SYSCON0.CLKSEL = 00_B$), the system clock is derived from the low-frequency wakeup clock source:

$$f_{SYS} = f_{WU}$$

In this mode, a basic functionality can be maintained without requiring an external clock source and while minimizing the power consumption.

4.7.2.3 Selecting and Changing the Operating Frequency

When selecting a clock source and the clock generation method, the required parameters must be carefully written to the respective bit fields, to avoid unintended intermediate states.

Many applications change the frequency of the system clock (f_{SYS}) during operation in order to optimize system performance and power consumption. Changing the operating frequency also changes the switching currents, which influences the power supply.

To ensure proper operation of the on-chip EVRs while they generate the core voltage, the operating frequency shall only be changed in certain steps. This prevents overshoots and undershoots of the supply voltage.

To avoid the indicated problems, recommended sequences are provided which ensure the intended operation of the clock system interacting with the power system.

Please refer to the Programmer's Guide.

4.7.3 External Clock Input Parameters

These parameters specify the external clock generation for the XC228xH. The clock can be generated in two ways:

- By connecting a **crystal or ceramic resonator** to pins XTAL1/XTAL2
- By supplying an **external clock signal**
 - This clock signal can be supplied either to pin XTAL1 (core voltage domain) or to pin CLKIN1 or CLKIN2 (IO voltage domain)

If connected to CLKIN1 or CLKIN2, the input signal must reach the defined input levels V_{IL} and V_{IH} . If connected to XTAL1, a minimum amplitude V_{AX1} (peak-to-peak voltage) is sufficient for the operation of the on-chip oscillator.

Note: The given clock timing parameters ($t_1 \dots t_4$) are only valid for an external clock input signal.

Note: Operating Conditions apply.

Table 28 External Clock Input Characteristics

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Oscillator frequency	f_{OSC} SR	4	–	40	MHz	Input= Clock Signal
		4	–	20	MHz	Input= Crystal or Ceramic Resonator
XTAL1 input current absolute value	$ I_{IL} $ CC	–	–	20	μ A	
Input clock high time	t_1 SR	6	–	–	ns	
Input clock low time	t_2 SR	6	–	–	ns	
Input clock rise time	t_3 SR	–	8	8	ns	
Input clock fall time	t_4 SR	–	8	8	ns	
Input voltage amplitude on XTAL1 ¹⁾	V_{AX1} SR	0.3 x V_{DDIM}	–	–	V	$f_{OSC} \geq 4$ MHz; $f_{OSC} < 16$ MHz
		0.4 x V_{DDIM}	–	–	V	$f_{OSC} \geq 16$ MHz; $f_{OSC} < 25$ MHz
		0.5 x V_{DDIM}	–	–	V	$f_{OSC} \geq 25$ MHz; $f_{OSC} \leq 40$ MHz
Input voltage range limits for signal on XTAL1	V_{IX1} SR	-1.7 + V_{DDIM}	–	1.7	V	²⁾

Electrical Parameters

- 1) The amplitude voltage V_{AX1} refers to the offset voltage V_{OFF} . This offset voltage must be stable during the operation and the resulting voltage peaks must remain within the limits defined by V_{IX1} .
- 2) Overload conditions must not occur on pin XTAL1.

Note: For crystal or ceramic resonator operation, it is strongly recommended to measure the oscillation allowance (negative resistance) in the final target system (layout) to determine the optimum parameters for oscillator operation.

The manufacturers of crystals and ceramic resonators offer an oscillator evaluation service. This evaluation checks the crystal/resonator specification limits to ensure a reliable oscillator operation.



Figure 22 External Clock Drive XTAL1

4.7.4 Pad Properties

The output pad drivers of the XC228xH can operate in several user-selectable modes. Strong driver mode allows controlling external components requiring higher currents such as power bridges or LEDs. Reducing the driving power of an output pad reduces electromagnetic emissions (EME). In strong driver mode, selecting a slower edge reduces EME.

The dynamic behavior, i.e. the rise time and fall time, depends on the applied external capacitance that must be charged and discharged. Timing values are given for a capacitance of 20 pF, unless otherwise noted.

In general, the performance of a pad driver depends on the available supply voltage V_{DDP} . Therefore the following tables list the pad parameters for the upper voltage range and the lower voltage range, respectively.

Note: These parameters are not subject to production test but verified by design and/or characterization.

Note: Operating Conditions apply.

Table 29 is valid under the following conditions: V_{DDPtyp} : 5 V; $V_{DDP} \geq 4.5$ V; $V_{DDP} \leq 5.5$ V

Table 29 Standard Pad Parameters for Upper Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Maximum output driver current (absolute value) ¹⁾	I_{Omax} CC	–	–	4.0	mA	Driver_Strength = Medium
		–	–	10	mA	Driver_Strength = Strong
		–	–	0.5	mA	Driver_Strength = Weak
Nominal output driver current (absolute value)	I_{Onom} CC	–	–	1.0	mA	Driver_Strength = Medium
		–	–	2.5	mA	Driver_Strength = Strong
		–	–	0.1	mA	Driver_Strength = Weak

Electrical Parameters

Table 29 Standard Pad Parameters for Upper Voltage Range (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Rise and Fall times (10% - 90%)	$t_{RF\ CC}$	–	–	23 + 0.6 x C_L	ns	$C_L \geq 20\text{ pF}$; $C_L \leq 100\text{ pF}$; Driver_Strength = Medium
		–	–	11.6 + 0.22 x C_L	ns	$C_L \geq 20\text{ pF}$; $C_L \leq 100\text{ pF}$; Driver_Strength = Strong ; Driver_Edge= Medium
		–	–	4.2 + 0.14 x C_L	ns	$C_L \geq 20\text{ pF}$; $C_L \leq 100\text{ pF}$; Driver_Strength = Strong ; Driver_Edge= Sharp
		–	–	20.6 + 0.22 x C_L	ns	$C_L \geq 20\text{ pF}$; $C_L \leq 100\text{ pF}$; Driver_Strength = Strong ; Driver_Edge= Slow
		–	–	212 + 1.9 x C_L	ns	$C_L \geq 20\text{ pF}$; $C_L \leq 100\text{ pF}$; Driver_Strength = Weak

1) An output current above $|I_{Oxnom}|$ may be drawn from up to three pins at the same time. For any group of 16 neighboring output pins, the total output current in each direction (ΣI_{OL} and $\Sigma -I_{OH}$) must remain below 50 mA.

Table 30 Standard Pad Parameters for Lower Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Maximum output driver current (absolute value) ¹⁾	I_{Omax} CC	–	–	2.5	mA	Driver_Strength = Medium
		–	–	10	mA	Driver_Strength = Strong
		–	–	0.5	mA	Driver_Strength = Weak
Nominal output driver current (absolute value)	I_{Onom} CC	–	–	1.0	mA	Driver_Strength = Medium
		–	–	2.5	mA	Driver_Strength = Strong
		–	–	0.1	mA	Driver_Strength = Weak

Electrical Parameters

Table 30 Standard Pad Parameters for Lower Voltage Range (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Rise and Fall times (10% - 90%)	$t_{RF\ CC}$	–	–	37 + 0.65 x C_L	ns	$C_L \geq 20\text{ pF}$; $C_L \leq 100\text{ pF}$; Driver_Strength = Medium
		–	–	24 + 0.3 x C_L	ns	$C_L \geq 20\text{ pF}$; $C_L \leq 100\text{ pF}$; Driver_Strength = Strong ; Driver_Edge= Medium
		–	–	6.2 + 0.24 x C_L	ns	$C_L \geq 20\text{ pF}$; $C_L \leq 100\text{ pF}$; Driver_Strength = Strong ; Driver_Edge= Sharp
		–	–	34 + 0.3 x C_L	ns	$C_L \geq 20\text{ pF}$; $C_L \leq 100\text{ pF}$; Driver_Strength = Strong ; Driver_Edge= Slow
		–	–	500 + 2.5 x C_L	ns	$C_L \geq 20\text{ pF}$; $C_L \leq 100\text{ pF}$; Driver_Strength = Weak

1) An output current above $|I_{Oxnom}|$ may be drawn from up to three pins at the same time. For any group of 16 neighboring output pins, the total output current in each direction (ΣI_{OL} and $\Sigma -I_{OH}$) must remain below 50 mA.

4.7.5 External Bus Timing

The following parameters specify the behavior of the XC228xH bus interface.

Note: These parameters are not subject to production test but verified by design and/or characterization.

Note: Operating Conditions apply.

Table 31 Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
CLKOUT Cycle Time ¹⁾	t_5 CC	–	$1 / f_{SYS}$	–	ns	
CLKOUT high time	t_6 CC	2	–	–		
CLKOUT low time	t_7 CC	2	–	–		
CLKOUT rise time	t_8 CC	–	–	3	ns	
CLKOUT fall time	t_9 CC	–	–	3		

1) The CLKOUT cycle time is influenced by PLL jitter. For longer periods the relative deviation decreases (see PLL deviation formula).

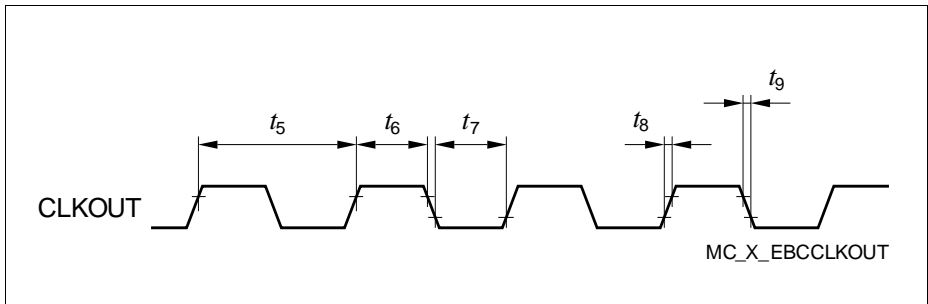


Figure 23 CLKOUT Signal Timing

Note: The term CLKOUT refers to the reference clock output signal which is generated by selecting f_{SYS} as the source signal for the clock output signal EXTCLK on pin P2.8 and by enabling the high-speed clock driver on this pin.

Variable Memory Cycles

External bus cycles of the XC228xH are executed in five consecutive cycle phases (AB, C, D, E, F). The duration of each cycle phase is programmable (via the TCONCSx registers) to adapt the external bus cycles to the respective external module (memory, peripheral, etc.).

The duration of the access phase can optionally be controlled by the external module using the READY handshake input.

This table provides a summary of the phases and the ranges for their length.

Table 32 Programmable Bus Cycle Phases (see timing diagrams)

Bus Cycle Phase	Parameter	Valid Values	Unit
Address setup phase, the standard duration of this phase (1 ... 2 TCS) can be extended by 0 ... 3 TCS if the address window is changed	tpAB	1 ... 2 (5)	TCS
Command delay phase	tpC	0 ... 3	TCS
Write Data setup/MUX Tristate phase	tpD	0 ... 1	TCS
Access phase	tpE	1 ... 32	TCS
Address/Write Data hold phase	tpF	0 ... 3	TCS

Note: The bandwidth of a parameter (from minimum to maximum value) covers the whole operating range (temperature, voltage) as well as process variations. Within a given device, however, this bandwidth is smaller than the specified range. This is also due to interdependencies between certain parameters. Some of these interdependencies are described in additional notes (see standard timing).

Note: Operating Conditions apply.

Table 33 is valid under the following conditions: $C_L = 20$ pF; voltage_range= upper ; voltage_range= upper

Table 33 External Bus Timing for Upper Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Output valid delay for \overline{RD} , \overline{WR} (L/H)	t_{10} CC	–	7	13	ns	
Output valid delay for BHE, ALE	t_{11} CC	–	7	14	ns	
Address output valid delay for A23 ... A0	t_{12} CC	–	8	14	ns	

Electrical Parameters

Table 33 External Bus Timing for Upper Voltage Range (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Address output valid delay for AD15 ... AD0 (MUX mode)	t_{13} CC	–	8	15	ns	
Output valid delay for \overline{CS}	t_{14} CC	–	7	13	ns	
Data output valid delay for AD15 ... AD0 (write data, MUX mode)	t_{15} CC	–	8	15	ns	
Data output valid delay for D15 ... D0 (write data, DEMUX mode)	t_{16} CC	–	8	15	ns	
Output hold time for \overline{RD} , \overline{WR} (L/H)	t_{20} CC	-2	6	8	ns	
Output hold time for \overline{BHE} , ALE	t_{21} CC	-2	6	10	ns	
Address output hold time for AD15 ... AD0	t_{23} CC	-3	6	8	ns	
Output hold time for \overline{CS}	t_{24} CC	-3	6	11	ns	
Data output hold time for D15 ... D0 and AD15 ... AD0	t_{25} CC	-3	6	8	ns	
Input setup time for READY, D15 ... D0, AD15 ... AD0	t_{30} SR	25	15	–	ns	
Input hold time READY, D15 ... D0, AD15 ... AD0 ¹⁾	t_{31} SR	0	-7	–	ns	

1) Read data are latched with the same internal clock edge that triggers the address change and the rising edge of RD. Address changes before the end of RD have no impact on (demultiplexed) read cycles. Read data can change after the rising edge of RD.

Table 34 is valid under the following conditions: $C_L = 20$ pF; voltage_range= lower ; voltage_range= lower

Table 34 External Bus Timing for Lower Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Output valid delay for \overline{RD} , \overline{WR} (L/H)	t_{10} CC	–	11	20	ns	
Output valid delay for BHE, ALE	t_{11} CC	–	10	21	ns	
Address output valid delay for A23 ... A0	t_{12} CC	–	11	22	ns	
Address output valid delay for AD15 ... AD0 (MUX mode)	t_{13} CC	–	10	22	ns	
Output valid delay for \overline{CS}	t_{14} CC	–	10	13	ns	
Data output valid delay for AD15 ... AD0 (write data, MUX mode)	t_{15} CC	–	10	22	ns	
Data output valid delay for D15 ... D0 (write data, DEMUX mode)	t_{16} CC	–	10	22	ns	
Output hold time for \overline{RD} , \overline{WR} (L/H)	t_{20} CC	-2	8	10	ns	
Output hold time for \overline{BHE} , ALE	t_{21} CC	-2	8	10	ns	
Address output hold time for AD15 ... AD0	t_{23} CC	-3	8	10	ns	
Output hold time for \overline{CS}	t_{24} CC	-3	8	11	ns	
Data output hold time for D15 ... D0 and AD15 ... AD0	t_{25} CC	-3	8	10	ns	
Input setup time for READY, D15 ... D0, AD15 ... AD0	t_{30} SR	29	17	–	ns	
Input hold time READY, D15 ... D0, AD15 ... AD0 ¹⁾	t_{31} SR	0	-9	–	ns	

1) Read data are latched with the same internal clock edge that triggers the address change and the rising edge of RD. Address changes before the end of RD have no impact on (demultiplexed) read cycles. Read data can change after the rising edge of RD.

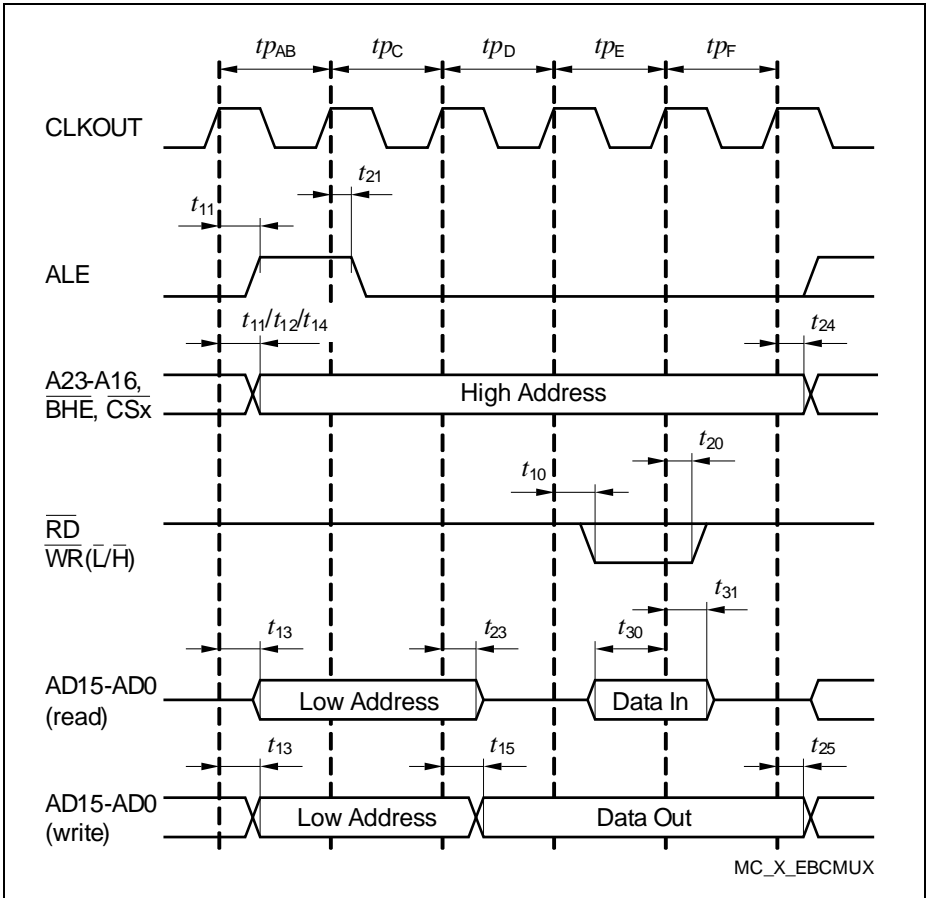


Figure 24 Multiplexed Bus Cycle

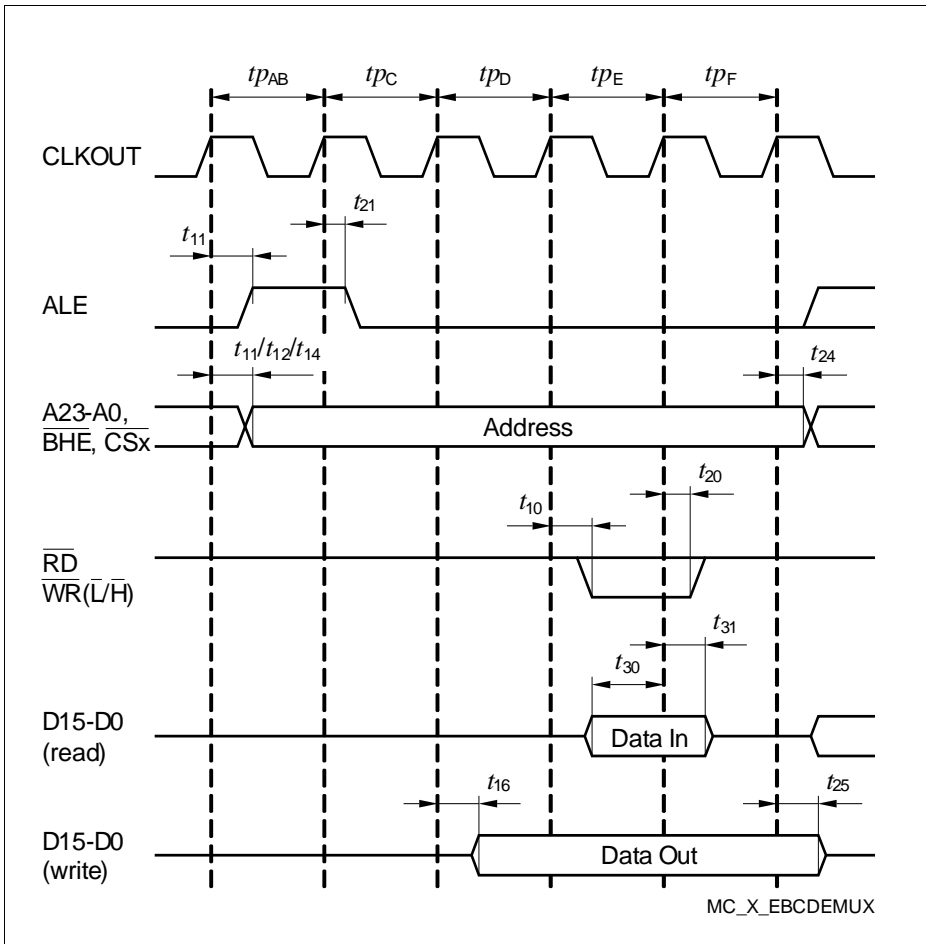


Figure 25 Demultiplexed Bus Cycle

4.7.5.1 Bus Cycle Control with the READY Input

The duration of an external bus cycle can be controlled by the external circuit using the READY input signal. The polarity of this input signal can be selected.

Synchronous READY permits the shortest possible bus cycle but requires the input signal to be synchronous to the reference signal CLKOUT.

An asynchronous READY signal puts no timing constraints on the input signal but incurs a minimum of one waitstate due to the additional synchronization stage. The minimum

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duration of an asynchronous READY signal for safe synchronization is one CLKOUT period plus the input setup time.

An active READY signal can be deactivated in response to the trailing (rising) edge of the corresponding command (RD or WR).

If the next bus cycle is controlled by READY, an active READY signal must be disabled before the first valid sample point in the next bus cycle. This sample point depends on the programmed phases of the next cycle.

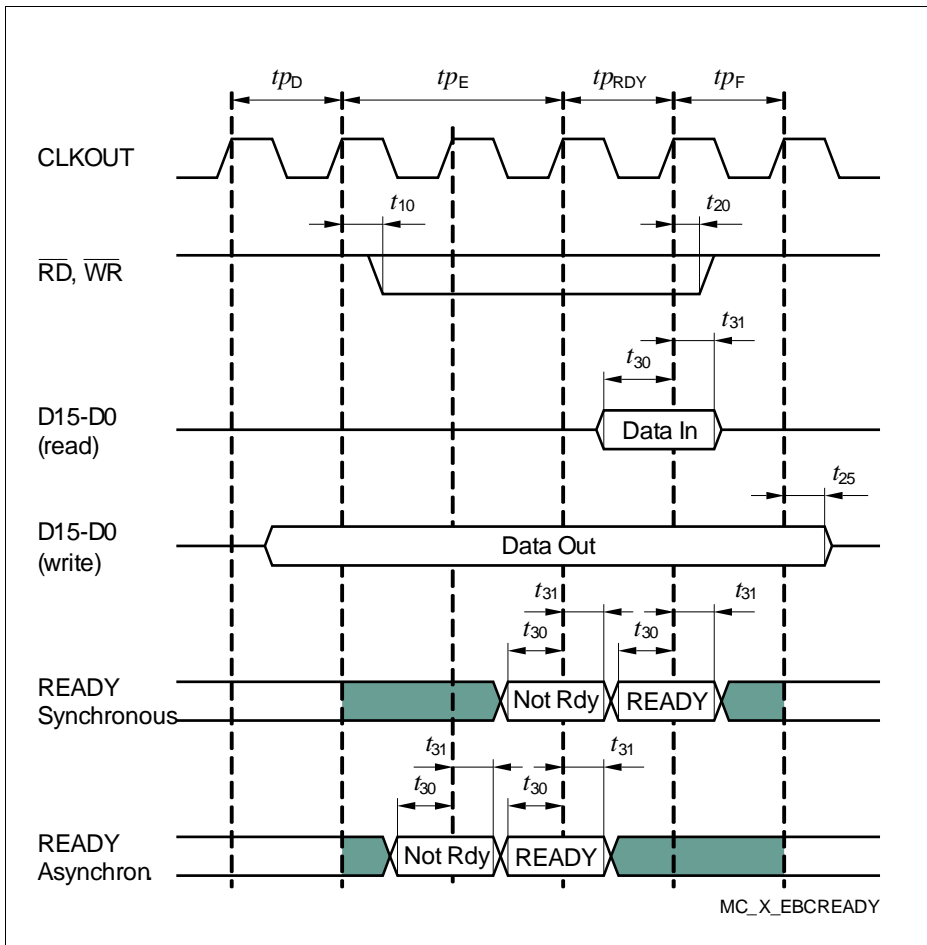


Figure 26 READY Timing

Electrical Parameters

Note: If the READY input is sampled inactive at the indicated sampling point ("Not Rdy") a READY-controlled waitstate is inserted (t_{pRDY}), sampling the READY input active at the indicated sampling point ("Ready") terminates the currently running bus cycle.

Note the different sampling points for synchronous and asynchronous READY. This example uses one mandatory waitstate (see t_{pE}) before the READY input value is used.

4.7.6 Synchronous Serial Interface Timing

The following parameters are applicable for a USIC channel operated in SSC mode.

Note: These parameters are not subject to production test but verified by design and/or characterization.

Note: Operating Conditions apply.

Table 35 is valid under the following conditions: $C_L = 20$ pF; SSC= master ; voltage_range= upper

Table 35 USIC SSC Master Mode Timing for Upper Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Slave select output SELO active to first SCLKOUT transmit edge	t_1 CC	$t_{SYS} - 8^{1)}$	–	–	ns	
Slave select output SELO inactive after last SCLKOUT receive edge	t_2 CC	$t_{SYS} - 6^{1)}$	–	–	ns	
Data output DOUT valid time	t_3 CC	-6	–	9	ns	
Receive data input setup time to SCLKOUT receive edge	t_4 SR	31	–	–	ns	
Data input DX0 hold time from SCLKOUT receive edge	t_5 SR	-4	–	–	ns	

1) $t_{SYS} = 1 / f_{SYS}$

Table 36 is valid under the following conditions: $C_L = 20$ pF; SSC= master ; voltage_range= lower

Table 36 USIC SSC Master Mode Timing for Lower Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Slave select output SELO active to first SCLKOUT transmit edge	t_1 CC	$t_{SYS} - 10^{1)}$	–	–	ns	
Slave select output SELO inactive after last SCLKOUT receive edge	t_2 CC	$t_{SYS} - 9^{1)}$	–	–	ns	
Data output DOUT valid time	t_3 CC	-7	–	11	ns	
Receive data input setup time to SCLKOUT receive edge	t_4 SR	40	–	–	ns	
Data input DX0 hold time from SCLKOUT receive edge	t_5 SR	-5	–	–	ns	

1) $t_{SYS} = 1 / f_{SYS}$

Table 37 is valid under the following conditions: $C_L = 20$ pF; SSC= slave ; voltage_range= upper

Table 37 USIC SSC Slave Mode Timing for Upper Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Select input DX2 setup to first clock input DX1 transmit edge ¹⁾	t_{10} SR	7	–	–	ns	
Select input DX2 hold after last clock input DX1 receive edge ¹⁾	t_{11} SR	7	–	–	ns	
Receive data input setup time to shift clock receive edge ¹⁾	t_{12} SR	7	–	–	ns	

Electrical Parameters

Table 37 USIC SSC Slave Mode Timing for Upper Voltage Range (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Data input DX0 hold time from clock input DX1 receive edge ¹⁾	t_{13} SR	5	–	–	ns	
Data output DOUT valid time	t_{14} CC	7	–	33	ns	

1) These input timings are valid for asynchronous input signal handling of slave select input, shift clock input, and receive data input (bits DXnCR.DSEN = 0).

Table 38 is valid under the following conditions: $C_L = 20$ pF; SSC= slave ; voltage_range= lower

Table 38 USIC SSC Slave Mode Timing for Lower Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Select input DX2 setup to first clock input DX1 transmit edge ¹⁾	t_{10} SR	7	–	–	ns	
Select input DX2 hold after last clock input DX1 receive edge ¹⁾	t_{11} SR	7	–	–	ns	
Receive data input setup time to shift clock receive edge ¹⁾	t_{12} SR	7	–	–	ns	
Data input DX0 hold time from clock input DX1 receive edge ¹⁾	t_{13} SR	5	–	–	ns	
Data output DOUT valid time	t_{14} CC	8	–	41	ns	

1) These input timings are valid for asynchronous input signal handling of slave select input, shift clock input, and receive data input (bits DXnCR.DSEN = 0).



Figure 27 USIC - SSC Master/Slave Mode Timing

Note: This timing diagram shows a standard configuration where the slave select signal is low-active and the serial clock signal is not shifted and not inverted.

4.7.7 FlexRay Interface Timing

The following parameters are applicable to the FlexRay channels.

Notes

1. These parameters are not subject to production test but verified by design and/or characterization.
2. Valid for FlexRay operation frequency 80 MHz.
3. Valid for pad Driver_Strength=Strong; Driver_Edge=Sharp.
4. Must use double pad (see [Chapter 2.1](#)) when operated at 3.3 V.
5. Operating Conditions apply.

Table 39 ERAY FlexRay Interface Timing

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Time span from last BSS to FES without the influence of quartz tolerance (d10Bit_TX) ¹⁾	t_{60} CC	997.75	–	1002.2 5	ns	$f_{OSC} = 20$ MHz; $C_L \leq 25$ pF
TxD data valid from fsample flip flop txd_reg => TxDA, TxDB (dTxAsym) ²⁾³⁾	$t_{61_minus_t62}$ CC	–	–	1.5	ns	$C_L \leq 25$ pF
Time span between last BSS and FES without influence of quartz tolerance (d10Bit_RX) ¹⁾⁴⁾	t_{63} SR	966.5	–	1046.0	ns	$f_{OSC} = 20$ MHz
RxD capture by fsample (RxDA/RxDB => sampling flip-flop) (dRxAsym) ⁴⁾	$t_{64_minus_t65}$ CC	–	–	3.5	ns	

1) PLL jitter included.

2) Refers to delays caused by the asymmetries of the output drivers of the digital logic and the GPIO pad drivers. Quartz tolerance and PLL jitter are not included.

3) TxD output drivers have an asymmetry of rising and falling edges of $|t_F - t_R| \leq 1$ ns.

4) Valid for output slopes of the Bus Driver of $dRxSlope \leq 5$ ns at $C_L = 15$ pF, $20\% \times V_{DDP}$ to $80\% \times V_{DDP}$, according to the FlexRay Electrical Physical Layer Specification V2.1 B.

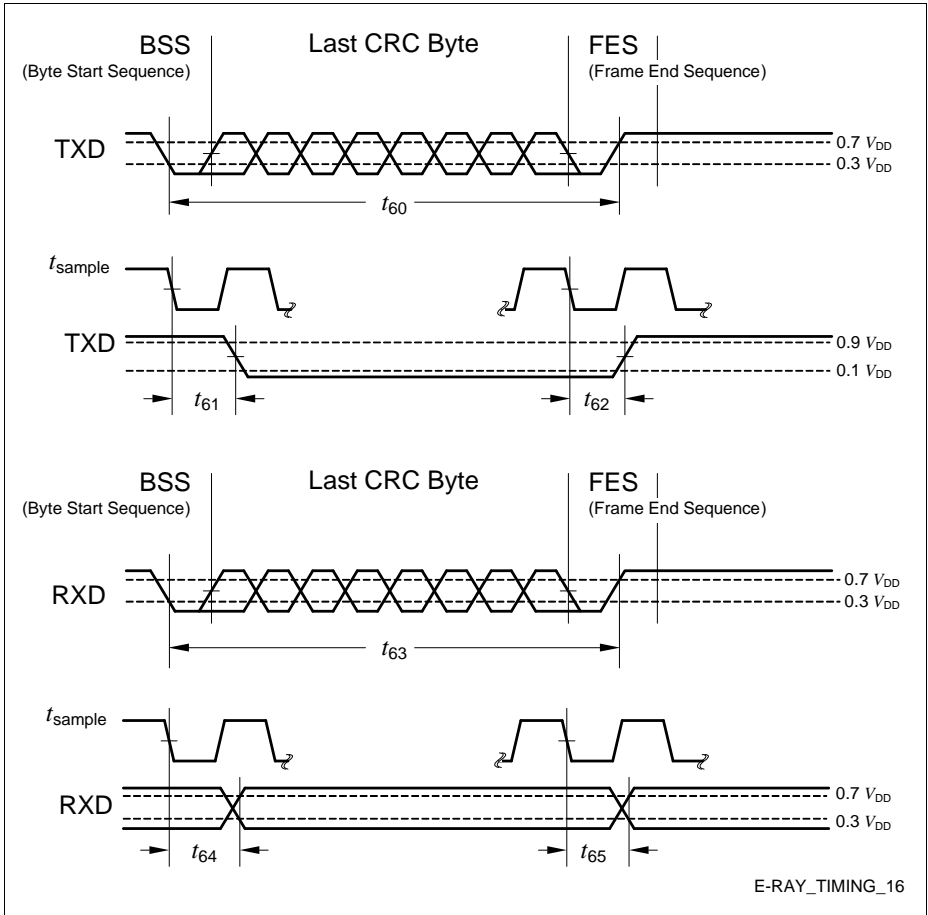


Figure 28 FlexRay Timing

4.7.8 Debug Interface Timing

The debugger can communicate with the XC228xH either via the 2-pin DAP interface or via the standard JTAG interface.

Debug via DAP

The following parameters are applicable for communication through the DAP debug interface.

Note: These parameters are not subject to production test but verified by design and/or characterization.

Note: Operating Conditions apply.

Table 40 is valid under the following conditions: $C_L = 20$ pF; voltage_range= upper

Table 40 DAP Interface Timing for Upper Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
DAP0 clock period ¹⁾	t_{11} SR	25	–	–	ns	
DAP0 high time	t_{12} SR	8	–	–	ns	
DAP0 low time ¹⁾	t_{13} SR	8	–	–	ns	
DAP0 clock rise time	t_{14} SR	–	–	4	ns	
DAP0 clock fall time	t_{15} SR	–	–	4	ns	
DAP1 setup to DAP0 rising edge	t_{16} SR	3	–	–	ns	pad_type= high speed ²⁾
		6	–	–	ns	pad_type= standard
DAP1 hold after DAP0 rising edge	t_{17} SR	4	–	–	ns	pad_type= high speed ²⁾
		6	–	–	ns	pad_type= standard
DAP1 valid per DAP0 clock period ³⁾	t_{19} CC	19	21	–	ns	pad_type= high speed ²⁾
		17	20	–	ns	pad_type= standard

1) See the DAP chapter for clock rate restrictions in the Active::IDLE protocol state.

2) Available high speed pins can be found in the pin definitions table in chapter 2.

3) The Host has to find a suitable sampling point by analyzing the sync telegram response.

Table 41 is valid under the following conditions: $C_L = 20$ pF; voltage_range= lower

Table 41 DAP Interface Timing for Lower Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
DAP0 clock period ¹⁾	t_{11} SR	25	–	–	ns	
DAP0 high time	t_{12} SR	8	–	–	ns	
DAP0 low time ¹⁾	t_{13} SR	8	–	–	ns	
DAP0 clock rise time	t_{14} SR	–	–	4	ns	
DAP0 clock fall time	t_{15} SR	–	–	4	ns	
DAP1 setup to DAP0 rising edge	t_{16} SR	3	–	–	ns	pad_type= high speed ²⁾
		6	–	–	ns	pad_type= standard
DAP1 hold after DAP0 rising edge	t_{17} SR	4	–	–	ns	pad_type= high speed ²⁾
		6	–	–	ns	pad_type= standard
DAP1 valid per DAP0 clock period ³⁾	t_{19} CC	19	21	–	ns	pad_type= high speed ²⁾
		12	17	–	ns	pad_type= standard

1) See the DAP chapter for clock rate restrictions in the Active::IDLE protocol state.

2) Available high speed pins can be found in the pin definitions table in chapter 2.

3) The Host has to find a suitable sampling point by analyzing the sync telegram response.

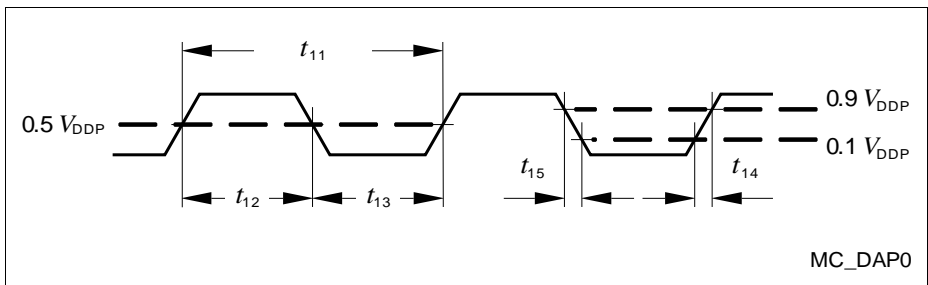


Figure 29 Test Clock Timing (DAP0)

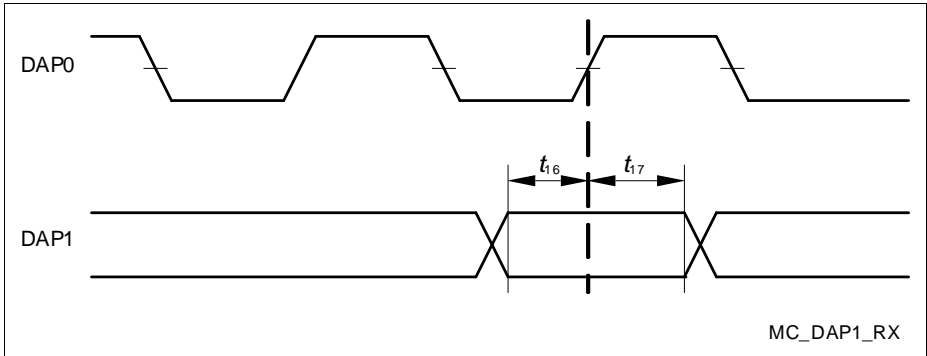


Figure 30 DAP Timing Host to Device

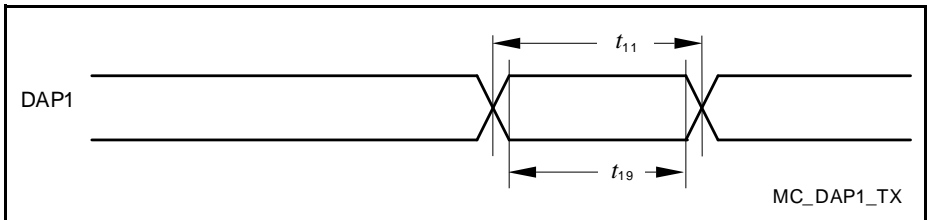


Figure 31 DAP Timing Device to Host

Note: The transmission timing is determined by the receiving debugger by evaluating the sync-request synchronization pattern telegram.

Debug via JTAG

The following parameters are applicable for communication through the JTAG debug interface. The JTAG module is fully compliant with IEEE1149.1-2000.

Note: These parameters are not subject to production test but verified by design and/or characterization.

Note: Operating Conditions apply.

Table 42 is valid under the following conditions: $C_L = 20$ pF; voltage_range= upper

Table 42 JTAG Interface Timing for Upper Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
TCK clock period	t_1 SR	50	–	–	ns	1)
TCK high time	t_2 SR	16	–	–	ns	

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Table 42 JTAG Interface Timing for Upper Voltage Range (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
TCK low time	t_3 SR	16	–	–	ns	
TCK clock rise time	t_4 SR	–	–	8	ns	
TCK clock fall time	t_5 SR	–	–	8	ns	
TDI/TMS setup to TCK rising edge	t_6 SR	6	–	–	ns	
TDI/TMS hold after TCK rising edge	t_7 SR	6	–	–	ns	
TDO valid from TCK falling edge (propagation delay) ²⁾	t_8 CC	–	25	29	ns	
TDO high impedance to valid output from TCK falling edge ³⁾²⁾	t_9 CC	–	25	29	ns	
TDO valid output to high impedance from TCK falling edge ²⁾	t_{10} CC	–	25	29	ns	
TDO hold after TCK falling edge ²⁾	t_{18} CC	5	–	–	ns	

1) Under typical conditions, the JTAG interface can operate at transfer rates up to 20 MHz.

2) The falling edge on TCK is used to generate the TDO timing.

3) The setup time for TDO is given implicitly by the TCK cycle time.

Table 43 is valid under the following conditions: $C_L = 20$ pF; voltage_range= lower

Table 43 JTAG Interface Timing for Lower Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
TCK clock period	t_1 SR	50	–	–	ns	
TCK high time	t_2 SR	16	–	–	ns	
TCK low time	t_3 SR	16	–	–	ns	
TCK clock rise time	t_4 SR	–	–	8	ns	
TCK clock fall time	t_5 SR	–	–	8	ns	
TDI/TMS setup to TCK rising edge	t_6 SR	6	–	–	ns	

Table 43 JTAG Interface Timing for Lower Voltage Range (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
TDI/TMS hold after TCK rising edge	t_7 SR	6	—	—	ns	
TDO valid from TCK falling edge (propagation delay) ¹⁾	t_8 CC	—	32	36	ns	
TDO high impedance to valid output from TCK falling edge ²⁾¹⁾	t_9 CC	—	32	36	ns	
TDO valid output to high impedance from TCK falling edge ¹⁾	t_{10} CC	—	32	36	ns	
TDO hold after TCK falling edge ¹⁾	t_{18} CC	5	—	—	ns	

- 1) The falling edge on TCK is used to generate the TDO timing.
- 2) The setup time for TDO is given implicitly by the TCK cycle time.

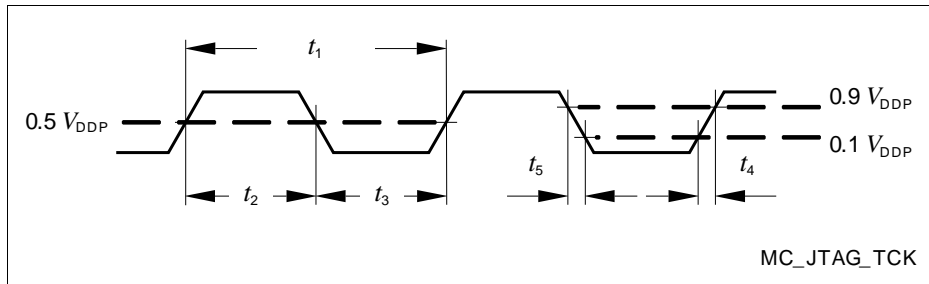


Figure 32 Test Clock Timing (TCK)

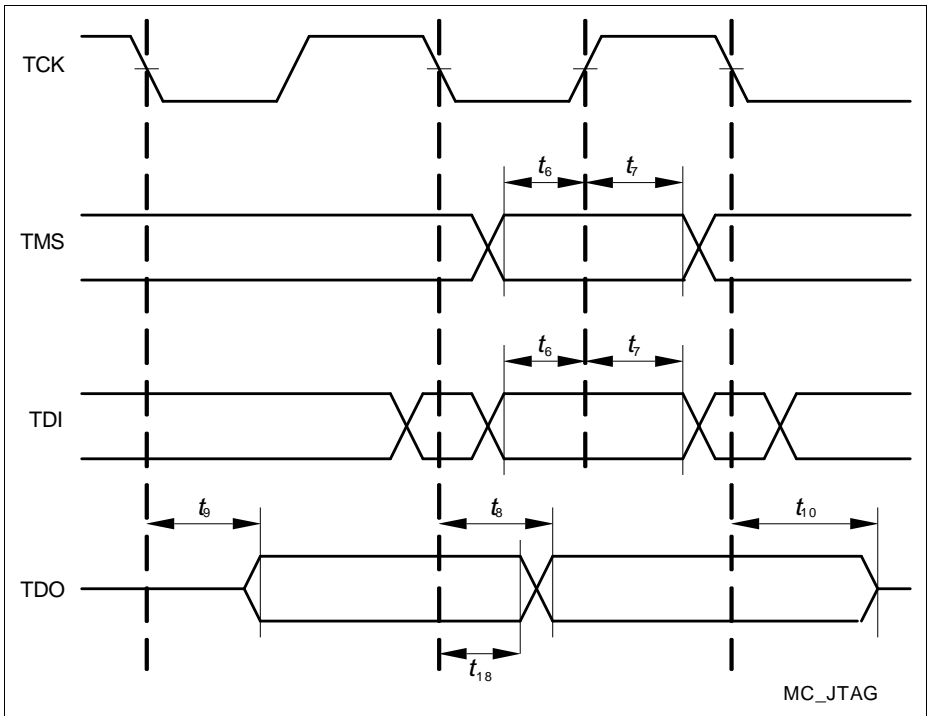


Figure 33 JTAG Timing

5 Package and Reliability

The XC2000 Family devices use the package type PG-LQFP (Plastic Green - Low Profile Quad Flat Package). The following specifications must be regarded to ensure proper integration of the XC228xH in its target environment.

5.1 Packaging

These parameters specify the packaging rather than the silicon.

Table 44 Package Parameters (PG-LQFP-144-13)

Parameter	Symbol	Limit Values		Unit	Notes
		Min.	Max.		
Exposed Pad Dimension	$E_x \times E_y$	–	8.0×8.0	mm	–
Power Dissipation	P_{DISS}	–	< 1	W	–
Thermal resistance Junction-Ambient	$R_{\Theta JA}$	–	39	K/W	2-layer, no vias ¹⁾
			33	K/W	4-layer, unsoldered ²⁾
			24	K/W	4-layer, no vias, soldered ³⁾
			19	K/W	4-layer, vias, soldered ⁴⁾

1) Device mounted on a 2-layer JEDEC board (JESD 51-3) without thermal vias; exposed pad soldered or not soldered.

2) Device mounted on a 4-layer JEDEC board (JESD 51-7) with or without thermal vias; exposed pad not soldered.

3) Device mounted on a 4-layer JEDEC board (JESD 51-7) without thermal vias; exposed pad soldered.

4) Device mounted on a 4-layer JEDEC board (JESD 51-7) with thermal vias; exposed pad soldered.

Note: To improve the EMC behavior, it is recommended to connect the exposed pad to the board ground, independent of the thermal requirements.

Board layout examples are given in an application note.

Package Compatibility Considerations

The XC228xH is a member of the XC2000 Family of microcontrollers. It is also compatible to a certain extent with members of similar families or subfamilies.

Each package is optimized for the device it houses. Therefore, there may be slight differences between packages of the same pin-count but for different device types. In particular, the size of the Exposed Pad (if present) may vary.

If different device types are considered or planned for an application, it must be ensured that the board layout fits all packages under consideration.

Package Outlines

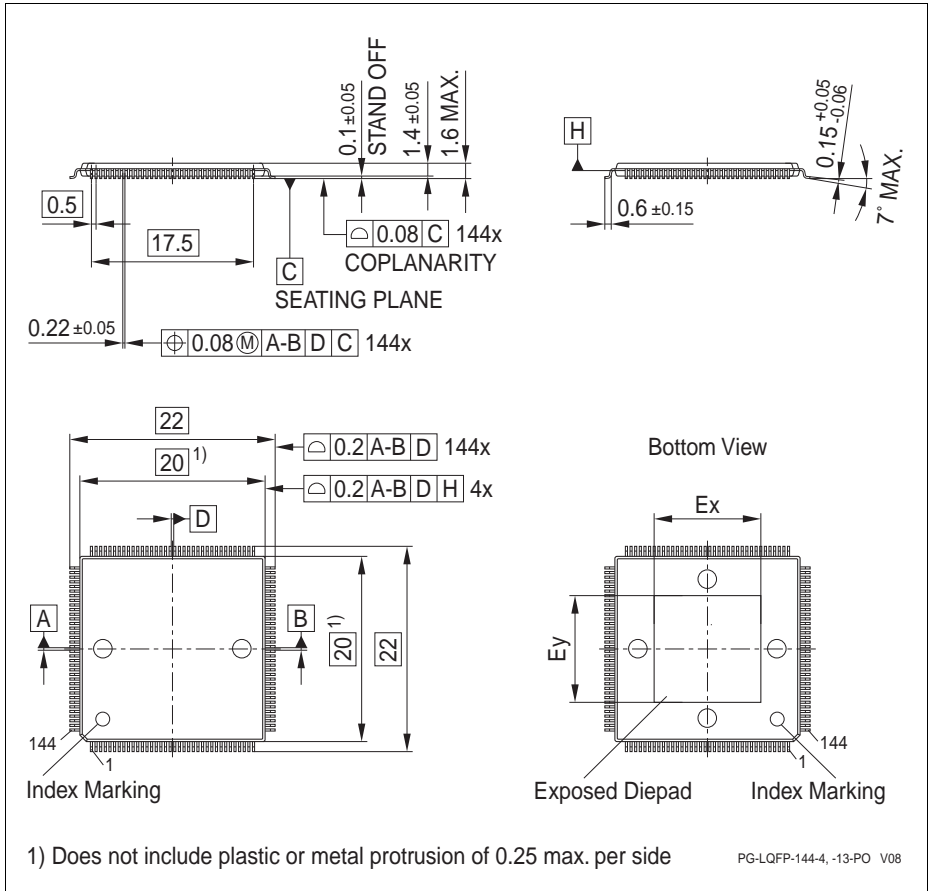


Figure 34 PG-LQFP-144-13 (Plastic Green Thin Quad Flat Package)

All dimensions in mm.

You can find complete information about Infineon packages, packing and marking in our Infineon Internet Page "Packages": <http://www.infineon.com/packages>

5.2 Thermal Considerations

When operating the XC228xH in a system, the total heat generated in the chip must be dissipated to the ambient environment to prevent overheating and the resulting thermal damage.

The maximum heat that can be dissipated depends on the package and its integration into the target board. The “Thermal resistance $R_{\Theta JA}$ ” quantifies these parameters. The power dissipation must be limited so that the average junction temperature does not exceed 150 °C.

The difference between junction temperature and ambient temperature is determined by $\Delta T = (P_{INT} + P_{IOSTAT} + P_{IODYN}) \times R_{\Theta JA}$

The internal power consumption is defined as

$$P_{INT} = V_{DDP} \times I_{DDP} \text{ (switching current and leakage current).}$$

The static external power consumption caused by the output drivers is defined as

$$P_{IOSTAT} = \Sigma((V_{DDP} - V_{OH}) \times I_{OH}) + \Sigma(V_{OL} \times I_{OL})$$

The dynamic external power consumption caused by the output drivers (P_{IODYN}) depends on the capacitive load connected to the respective pins and their switching frequencies.

If the total power dissipation for a given system configuration exceeds the defined limit, countermeasures must be taken to ensure proper system operation:

- Reduce V_{DDP} , if possible in the system
- Reduce the system frequency
- Reduce the number of output pins
- Reduce the load on active output drivers

5.3 Quality Declarations

The operation lifetime of the XC228xH depends on the applied temperature profile in the application. For a typical example, please refer to [Table 46](#); for other profiles, please contact your Infineon counterpart to calculate the specific lifetime within your application.

Table 45 Quality Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Operation lifetime	t_{OP} CC	–	–	20	a	See Table 46 and Table 47
ESD susceptibility according to Human Body Model (HBM)	V_{HBM} SR	–	–	2 000	V	EIA/JESD22-A114-B
Moisture sensitivity level	MSL CC	–	–	3	–	JEDEC J-STD-020C

Table 46 Typical Usage Temperature Profile

Operating Time (Sum = 20 years)	Operating Temperat.	Notes
1 200 h	$T_J = 150^\circ\text{C}$	Normal operation
3 600 h	$T_J = 125^\circ\text{C}$	Normal operation
7 200 h	$T_J = 110^\circ\text{C}$	Normal operation
12 000 h	$T_J = 100^\circ\text{C}$	Normal operation
$7 \times 21\ 600$ h	$T_J = 0\dots 10^\circ\text{C}, \dots, 60\dots 70^\circ\text{C}$	Power reduction

Table 47 Long Time Storage Temperature Profile

Operating Time (Sum = 20 years)	Operating Temperat.	Notes
2 000 h	$T_J = 150^\circ\text{C}$	Normal operation
16 000 h	$T_J = 125^\circ\text{C}$	Normal operation
6 000 h	$T_J = 110^\circ\text{C}$	Normal operation
151 200 h	$T_J \leq 150^\circ\text{C}$	No operation

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- Поставка специализированных компонентов военного и аэрокосмического уровня качества (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Actel, Aeroflex, Peregrine, VPT, Syfer, Eurofarad, Texas Instruments, MS Kennedy, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

Компания «Океан Электроники» является официальным дистрибьютором и эксклюзивным представителем в России одного из крупнейших производителей разъемов военного и аэрокосмического назначения «JONHON», а так же официальным дистрибьютором и эксклюзивным представителем в России производителя высокотехнологичных и надежных решений для передачи СВЧ сигналов «FORSTAR».



JONHON

«JONHON» (основан в 1970 г.)

Разъемы специального, военного и аэрокосмического назначения:

(Применяются в военной, авиационной, аэрокосмической, морской, железнодорожной, горно- и нефтедобывающей отраслях промышленности)

«FORSTAR» (основан в 1998 г.)

ВЧ соединители, коаксиальные кабели, кабельные сборки и микроволновые компоненты:

(Применяются в телекоммуникациях гражданского и специального назначения, в средствах связи, РЛС, а так же военной, авиационной и аэрокосмической отраслях промышленности).



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