

### FEATURES

4 A peak output current

Working voltage

High-side or low-side relative to input: 537 V peak

High frequency operation: 1 MHz maximum

3.3 V to 5 V CMOS input logic

4.5 V to 18 V output drive

Secondary UVLO

ADuM3224A/ADuM4224A UVLO at 4.1 V  $V_{DDA}/V_{DDB}$

ADuM3224B/ADuM4224B UVLO at 6.9 V  $V_{DDA}/V_{DDB}$

ADuM3224C/ADuM4224C UVLO at 10.5 V  $V_{DDA}/V_{DDB}$

Precise timing characteristics

59 ns maximum isolator and driver propagation delay

5 ns maximum channel-to-channel matching

CMOS input logic levels

High common-mode transient immunity: >25 kV/ $\mu$ s

Enhanced system-level ESD performance per IEC 61000-4-x

High junction temperature operation: 125°C

Default low output

Safety and regulatory approvals

ADuM3224 narrow-body, 16-lead SOIC

UL recognition per UL 1577

3000 V rms for 1 minute SOIC long package

CSA Component Acceptance Notice 5A

VDE certificate of conformity (pending)

DIN V VDE V 0884-10 (VDE V 0884-10):2006-12

$V_{IORM}$  = 560 V peak

ADuM4224 wide-body, 16-lead SOIC

UL recognition per UL 1577

5000 V rms for 1 minute SOIC long package

CSA Component Acceptance Notice 5A

VDE certificate of conformity (pending)

DIN V VDE V 0884-10 (VDE V 0884-10):2006-12

$V_{IORM}$  = 849 V peak

Qualified for automotive applications

### APPLICATIONS

Switching power supplies

Isolated IGBT/MOSFET gate drives

Industrial inverters

### FUNCTIONAL BLOCK DIAGRAM



Figure 1.

### GENERAL DESCRIPTION

The ADuM3224/ADuM4224<sup>1</sup> are 4 A isolated, half-bridge gate drivers that employ the Analog Devices, Inc., iCoupler® technology to provide independent and isolated high-side and low-side outputs. The ADuM3224 provides 3000 V rms isolation in the narrow-body, 16-lead SOIC package, and the ADuM4224 provides 5000 V rms isolation in the wide-body, 16-lead SOIC package. Combining high speed CMOS and monolithic transformer technology, these isolation components provide outstanding performance characteristics superior to the alternatives, such as the combination of pulse transformers and gate drivers.

The ADuM3224/ADuM4224 isolators each provide two independent isolated channels. They operate with an input supply ranging from 3.0 V to 5.5 V, providing compatibility with lower voltage systems. In comparison to gate drivers employing high voltage level translation methodologies, the ADuM3224/ADuM4224 offer the benefit of true, galvanic isolation between the input and each output. Each output can be continuously operated up to 537 V peak relative to the input, thereby supporting low-side switching to negative voltages. The differential voltage between the high-side and low-side can be as high as 800 V peak.

As a result, the ADuM3224/ADuM4224 provide reliable control over the switching characteristics of IGBT/MOSFET configurations over a wide range of positive or negative switching voltages.

<sup>1</sup> Protected by U.S. Patents 5,952,849; 6,873,065; and 7,075,239. Other patents pending.

## TABLE OF CONTENTS

Features .....	1	Pin Configuration and Function Descriptions.....	10
Applications.....	1	Typical Performance Characteristics .....	11
Functional Block Diagram .....	1	Applications Information .....	14
General Description .....	1	Printed Circuit Board Layout .....	14
Revision History .....	2	Undervoltage Lockout .....	14
Specifications.....	3	Propagation Delay-Related Parameters.....	14
Electrical Characteristics—5 V Operation.....	3	Thermal Limitations and Switch Load Characteristics.....	14
Electrical Characteristics—3.3 V Operation .....	4	Output Load Characteristics.....	14
Package Characteristics .....	5	Bootstrapped Half-Bridge Operation .....	15
Insulation and Safety Related Specifications .....	5	DC Correctness and Magnetic Field Immunity .....	15
Regulatory Information.....	6	Power Consumption .....	17
DIN V VDE V 0884-10 (VDE V 0884-10) Insulation		Insulation Lifetime .....	17
Characteristics .....	7	Outline Dimensions .....	18
Recommended Operating Conditions .....	8	Ordering Guide .....	19
Absolute Maximum Ratings.....	9	Automotive Products.....	19
ESD Caution.....	9		

## REVISION HISTORY

### 11/15—Rev. A to Rev. B

Changes to Power Consumption Section.....	17
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### 11/14—Rev. 0 to Rev. A

Changes to Features Section and General Description	
Section.....	1
Changes to Table 5.....	5
Changes to Regulatory Information Section, Table 6, and	
Table 7 .....	6
Changes to Table 8 and Table 9.....	7

### 12/13—Revision 0: Initial Version

## SPECIFICATIONS

### ELECTRICAL CHARACTERISTICS—5 V OPERATION

All voltages are relative to their respective ground.  $4.5\text{ V} \leq V_{DD1} \leq 5.5\text{ V}$ ,  $4.5\text{ V} \leq V_{DDA} \leq 18\text{ V}$ ,  $4.5\text{ V} \leq V_{DDB} \leq 18\text{ V}$ , unless stated otherwise. All minimum/maximum specifications apply over  $T_j = -40^\circ\text{C}$  to  $+125^\circ\text{C}$ . All typical specifications are at  $T_j = 25^\circ\text{C}$ ,  $V_{DD1} = 5\text{ V}$ ,  $V_{DDA} = V_{DDB} = 12\text{ V}$ . Switching specifications are tested with CMOS signal levels.

Table 1.

Parameter	Symbol	Min	Typ	Max	Unit	Test Conditions/Comments
DC SPECIFICATIONS						
Input Supply Current, Quiescent	$I_{DDI(Q)}$		1.4	2.4	mA	
Output Supply Current, Per Channel, Quiescent	$I_{DDO(Q)}$		2.3	3.2	mA	
Supply Current at 1 MHz						
$V_{DD1}$ Supply Current	$I_{DD1(Q)}$		1.6	2.5	mA	Up to 1 MHz, no load
$V_{DDA}/V_{DDB}$ Supply Current	$I_{DDA(Q)}/I_{DDB(Q)}$		5.6	8.0	mA	Up to 1 MHz, no load
Input Currents	$I_{IA}, I_{IB}$	-1	+0.01	+1	$\mu\text{A}$	$0\text{ V} \leq V_{IA}, V_{IB} \leq V_{DD1}$
Logic High Input Threshold	$V_{IH}$	$0.7 \times V_{DD1}$			V	
Logic Low Input Threshold	$V_{IL}$			$0.3 \times V_{DD1}$	V	
Logic High Output Voltages	$V_{OAH}, V_{OBH}$	$V_{DDA}/V_{DDB} - 0.1$	$V_{DDA}/V_{DDB}$		V	$I_{OX} = -20\text{ mA}, V_{IX} = V_{IXH}$
Logic Low Output Voltages	$V_{OAL}, V_{OBL}$		0.0	0.15	V	$I_{OX} = +20\text{ mA}, V_{IX} = V_{IXL}$
Undervoltage Lockout, $V_{DDA}/V_{DDB}$ Supply						
A Grade						
Positive Going Threshold	$V_{DDAUV+}, V_{DDBUV+}$		4.1	4.4	V	
Negative Going Threshold	$V_{DDAUV-}, V_{DDBUV-}$	3.2	3.6		V	
Hysteresis	$V_{DDAUVH}, V_{DDBUVH}$		0.5		V	
B Grade						
Positive Going Threshold	$V_{DDAUV+}, V_{DDBUV+}$		6.9	7.4	V	
Negative Going Threshold	$V_{DDAUV-}, V_{DDBUV-}$	5.7	6.2		V	
Hysteresis	$V_{DDAUVH}, V_{DDBUVH}$		0.7		V	
C Grade						
Positive Going Threshold	$V_{DDAUV+}, V_{DDBUV+}$		10.5	11.1	V	
Negative Going Threshold	$V_{DDAUV-}, V_{DDBUV-}$	8.9	9.6		V	
Hysteresis	$V_{DDAUVH}, V_{DDBUVH}$		0.9		V	
Output Short-Circuit Pulsed Current <sup>1</sup>	$I_{OA(SC)}, I_{OB(SC)}$	2.0	4.0		A	$V_{DDA}/V_{DDB} = 12\text{ V}$
Output Pulsed Source Resistance	$R_{OA}, R_{OB}$	0.3	1.1	3.0	$\Omega$	$V_{DDA}/V_{DDB} = 12\text{ V}$
Output Pulsed Sink Resistance	$R_{OA}, R_{OB}$	0.3	0.6	3.0	$\Omega$	$V_{DDA}/V_{DDB} = 12\text{ V}$
SWITCHING SPECIFICATIONS						
Pulse Width <sup>2</sup>	PW	50			ns	$C_L = 2\text{ nF}, V_{DDA}/V_{DDB} = 12\text{ V}$
Maximum Data Rate <sup>3</sup>		1			MHz	$C_L = 2\text{ nF}, V_{DDA}/V_{DDB} = 12\text{ V}$
Propagation Delay <sup>4</sup>	$t_{DHL}, t_{DLH}$	31	43	54	ns	$C_L = 2\text{ nF}, V_{DDA}/V_{DDB} = 12\text{ V}$ ; see Figure 20
ADuM3224A/ADuM4224A	$t_{DHL}, t_{DLH}$	35	47	59	ns	$C_L = 2\text{ nF}, V_{DDA}/V_{DDB} = 4.5\text{ V}$ ; see Figure 20
Propagation Delay Skew <sup>5</sup>	$t_{PSK}$			12	ns	$C_L = 2\text{ nF}, V_{DDA}/V_{DDB} = 12\text{ V}$ ; see Figure 20
Channel-to-Channel Matching <sup>6</sup>	$t_{PSKCD}$		1	5	ns	$C_L = 2\text{ nF}, V_{DDA}/V_{DDB} = 12\text{ V}$ ; see Figure 20
	$t_{PSKCD}$		1	7	ns	$C_L = 2\text{ nF}, V_{DDA}/V_{DDB} = 4.5\text{ V}$ ; see Figure 20
Output Rise/Fall Time (10% to 90%)	$t_R/t_F$	6	12	18	ns	$C_L = 2\text{ nF}, V_{DDA}/V_{DDB} = 12\text{ V}$ ; see Figure 20
Dynamic Input Supply Current Per Channel	$I_{DDI(D)}$		0.05		mA/Mbps	$V_{DDA}/V_{DDB} = 12\text{ V}$
Dynamic Output Supply Current Per Channel	$I_{DDO(D)}$		1.65		mA/Mbps	$V_{DDA}/V_{DDB} = 12\text{ V}$
Refresh Rate	$f_r$		1.2		Mbps	

<sup>1</sup> Short-circuit duration less than 1  $\mu\text{s}$ . Average power must conform to the limit shown in the Absolute Maximum Ratings section.

<sup>2</sup> The minimum pulse width is the shortest pulse width at which the specified timing parameter is guaranteed.

<sup>3</sup> The maximum data rate is the fastest data rate at which the specified timing parameter is guaranteed.

<sup>4</sup> The  $t_{DLH}$  propagation delay is measured from the time of the input rising logic high threshold,  $V_{IH}$ , to the output rising 10% level of the  $V_{OX}$  signal. The  $t_{DHL}$  propagation delay is measured from the input falling logic low threshold,  $V_{IL}$ , to the output falling 90% threshold of the  $V_{OX}$  signal. See Figure 20 for waveforms of propagation delay parameters.

<sup>5</sup>  $t_{PSK}$  is the magnitude of the worst-case difference in  $t_{DLH}$  and/or  $t_{DHL}$  that is measured between units at the same operating temperature, supply voltages, and output load within the recommended operating conditions. See Figure 20 for waveforms of propagation delay parameters.

<sup>6</sup> Channel-to-channel matching is the absolute value of the difference in propagation delays between the two channels.

**ELECTRICAL CHARACTERISTICS—3.3 V OPERATION**

All voltages are relative to their respective ground.  $3.0\text{ V} \leq V_{DD1} \leq 3.6\text{ V}$ ,  $4.5\text{ V} \leq V_{DDA} \leq 18\text{ V}$ ,  $4.5\text{ V} \leq V_{DDB} \leq 18\text{ V}$ , unless stated otherwise. All minimum/maximum specifications apply over  $T_j = -40^\circ\text{C}$  to  $+125^\circ\text{C}$ . All typical specifications are at  $T_j = 25^\circ\text{C}$ ,  $V_{DD1} = 3.3\text{ V}$ ,  $V_{DDA} = V_{DDB} = 12\text{ V}$ . Switching specifications are tested with CMOS signal levels.

**Table 2.**

Parameter	Symbol	Min	Typ	Max	Unit	Test Conditions/Comments
<b>DC SPECIFICATIONS</b>						
Input Supply Current, Quiescent	$I_{DDI(Q)}$		0.87	1.4	mA	
Output Supply Current, Per Channel, Quiescent	$I_{DDO(Q)}$		2.3	3.2	mA	
Supply Current at 1 MHz						
$V_{DD1}$ Supply Current	$I_{DD1(Q)}$		1.1	1.5	mA	Up to 1 MHz, no load
$V_{DDA}/V_{DDB}$ Supply Current	$I_{DDA(Q)}/I_{DDB(Q)}$		5.6	8.0	mA	Up to 1 MHz, no load
Input Currents	$I_{IA}, I_{IB}$	-10	+0.01	+10	$\mu\text{A}$	$0\text{ V} \leq V_{IA}, V_{IB} \leq V_{DD1}$
Logic High Input Threshold	$V_{IH}$	$0.7 \times V_{DD1}$			V	
Logic Low Input Threshold	$V_{IL}$			$0.3 \times V_{DD1}$	V	
Logic High Output Voltages	$V_{OAH}, V_{OBH}$	$V_{DDA}/V_{DDB} - 0.1$	$V_{DDA}/V_{DDB}$		V	$I_{OX} = -20\text{ mA}$ , $V_{IX} = V_{IXH}$
Logic Low Output Voltages	$V_{OAL}, V_{OBL}$		0.0	0.15	V	$I_{OX} = +20\text{ mA}$ , $V_{IX} = V_{IXL}$
Undervoltage Lockout, $V_{DDA}/V_{DDB}$ Supply						
A Grade						
Positive Going Threshold	$V_{DDAUV+}, V_{DDBUV+}$		4.1	4.4	V	
Negative Going Threshold	$V_{DDAUV-}, V_{DDBUV-}$	3.2	3.6		V	
Hysteresis	$V_{DDAUVH}, V_{DDBUVH}$		0.5		V	
B Grade						
Positive Going Threshold	$V_{DDAUV+}, V_{DDBUV+}$		6.9	7.4	V	
Negative Going Threshold	$V_{DDAUV-}, V_{DDBUV-}$	5.7	6.2		V	
Hysteresis	$V_{DDAUVH}, V_{DDBUVH}$		0.7		V	
C Grade						
Positive Going Threshold	$V_{DDAUV+}, V_{DDBUV+}$		10.5	11.1	V	
Negative Going Threshold	$V_{DDAUV-}, V_{DDBUV-}$	8.9	9.6		V	
Hysteresis	$V_{DDAUVH}, V_{DDBUVH}$		0.9		V	
Output Short-Circuit Pulsed Current <sup>1</sup>	$I_{OA(SC)}, I_{OB(SC)}$	2.0	4.0		A	$V_{DDA}/V_{DDB} = 12\text{ V}$
Output Pulsed Source Resistance	$R_{OA}, R_{OB}$	0.3	1.1	3.0	$\Omega$	$V_{DDA}/V_{DDB} = 12\text{ V}$
Output Pulsed Sink Resistance	$R_{OA}, R_{OB}$	0.3	0.6	3.0	$\Omega$	$V_{DDA}/V_{DDB} = 12\text{ V}$
<b>SWITCHING SPECIFICATIONS</b>						
Pulse Width <sup>2</sup>	PW	50			ns	$C_L = 2\text{ nF}$ , $V_{DDA}/V_{DDB} = 12\text{ V}$
Maximum Data Rate <sup>3</sup>		1			MHz	$C_L = 2\text{ nF}$ , $V_{DDA}/V_{DDB} = 12\text{ V}$
Propagation Delay <sup>4</sup>	$t_{DHL}, t_{DLH}$	35	47	59	ns	$C_L = 2\text{ nF}$ , $V_{DDA}/V_{DDB} = 12\text{ V}$ , see Figure 20
ADuM3224A/ADuM4224A	$t_{DHL}, t_{DLH}$	37	51	65	ns	$C_L = 2\text{ nF}$ , $V_{DDA}/V_{DDB} = 4.5\text{ V}$ , see Figure 20
Propagation Delay Skew <sup>5</sup>	$t_{PSK}$			12	ns	$C_L = 2\text{ nF}$ , $V_{DDA}/V_{DDB} = 12\text{ V}$ , see Figure 20
Channel-to-Channel Matching <sup>6</sup>	$t_{PSKCD}$		1	5	ns	$C_L = 2\text{ nF}$ , $V_{DDA}/V_{DDB} = 12\text{ V}$ , see Figure 20
	$t_{PSKCD}$		1	7	ns	$C_L = 2\text{ nF}$ , $V_{DDA}/V_{DDB} = 4.5\text{ V}$ , see Figure 20
Output Rise/Fall Time (10% to 90%)	$t_R/t_F$	6	12	22	ns	$C_L = 2\text{ nF}$ , $V_{DDA}/V_{DDB} = 12\text{ V}$ , see Figure 20
Dynamic Input Supply Current Per Channel	$I_{DDI(D)}$		0.05		mA/Mbps	$V_{DDA}/V_{DDB} = 12\text{ V}$
Dynamic Output Supply Current Per Channel	$I_{DDO(D)}$		1.65		mA/Mbps	$V_{DDA}/V_{DDB} = 12\text{ V}$
Refresh Rate	$f_r$		1.1		Mbps	

<sup>1</sup> Short-circuit duration less than 1  $\mu\text{s}$ . Average power must conform to the limit shown in the Absolute Maximum Ratings section.

<sup>2</sup> The minimum pulse width is the shortest pulse width at which the specified timing parameter is guaranteed.

<sup>3</sup> The maximum data rate is the fastest data rate at which the specified timing parameter is guaranteed.

<sup>4</sup> The  $t_{DLH}$  propagation delay is measured from the time of the input rising logic high threshold,  $V_{IH}$ , to the output rising 10% level of the  $V_{OX}$  signal. The  $t_{DHL}$  propagation delay is measured from the input falling logic low threshold,  $V_{IL}$ , to the output falling 90% threshold of the  $V_{OX}$  signal. See Figure 20 for waveforms of propagation delay parameters.

<sup>5</sup>  $t_{PSK}$  is the magnitude of the worst-case difference in  $t_{DLH}$  and/or  $t_{DHL}$  that is measured between units at the same operating temperature, supply voltages, and output load within the recommended operating conditions. See Figure 20 for waveforms of propagation delay parameters.

<sup>6</sup> Channel-to-channel matching is the absolute value of the difference in propagation delays between the two channels.

**PACKAGE CHARACTERISTICS**

Table 3.

Parameter	Symbol	Min	Typ	Max	Unit	Test Conditions/Comments
Resistance (Input-to-Output)	$R_{I-O}$		10 <sup>12</sup>		$\Omega$	f = 1 MHz
Capacitance (Input-to-Output)	$C_{I-O}$		2.0		pF	
Input Capacitance	$C_i$		4.0		pF	
IC Junction-to-Ambient Thermal Resistance						
ADuM3224	$\theta_{JA}$		76		°C/W	
ADuM4224	$\theta_{JA}$		45		°C/W	
IC Junction-to-Case Thermal Resistance						
ADuM3224	$\theta_{JC}$		42		°C/W	
ADuM4224	$\theta_{JC}$		29		°C/W	

**INSULATION AND SAFETY RELATED SPECIFICATIONS****ADuM3224 Specifications**

Table 4.

Parameter	Symbol	Value	Unit	Test Conditions/Comments
Rated Dielectric Insulation Voltage		3000	V rms	1 minute duration
Minimum External Air Gap (Clearance)	L(I01)	4.0 min	mm	Measured from input terminals to output terminals, shortest distance through air
Minimum External Tracking (Creepage)	L(I02)	4.0 min	mm	Measured from input terminals to output terminals, shortest distance path along body
Minimum Internal Gap (Internal Clearance)		0.017 min	mm	Insulation distance through insulation
Tracking Resistance (Comparative Tracking Index)	CTI	>400	V	DIN IEC 112/VDE 0303 Part 1
Isolation Group		II		Material Group (DIN VDE 0110, 1/89, Table 1)

**ADuM4224 Specifications**

Table 5.

Parameter	Symbol	Value	Unit	Test Conditions/Comments
Rated Dielectric Insulation Voltage		5000	V rms	1 minute duration
Minimum External Air Gap (Clearance)	L(I01)	7.6 min	mm	Measured from input terminals to output terminals, shortest distance through air
Minimum External Tracking (Creepage)	L(I02)	7.6 min	mm	Measured from input terminals to output terminals, shortest distance path along body
Minimum Internal Gap (Internal Clearance)		0.017 min	mm	Insulation distance through insulation
Tracking Resistance (Comparative Tracking Index)	CTI	>400	V	DIN IEC 112/VDE 0303 Part 1
Isolation Group		II		Material Group (DIN VDE 0110, 1/89, Table 1)

**REGULATORY INFORMATION**

The [ADuM3224](#) is approved or pending approval by the organizations listed in Table 6.

**Table 6.**

UL	CSA	VDE (Pending)
Recognized under UL 1577 Component Recognition Program <sup>1</sup>	Approved under CSA Component Acceptance Notice 5A	Certified according to DIN V VDE V 0884-10 (VDE V 0884-10): 2006-12 <sup>2</sup>
Single/Protection 3000 V rms Isolation Voltage	Basic insulation per CSA 60950-1-07 and IEC 60950-1, 380 V rms (537 V peak) maximum working voltage	Reinforced insulation, 560 V peak
File E214100	File 205078	File 2471900-4880-0001

<sup>1</sup> In accordance with UL 1577, each [ADuM3224](#) is proof tested by applying an insulation test voltage  $\geq 3600$  V rms for 1 second (current leakage detection limit = 6  $\mu$ A).

<sup>2</sup> In accordance with DIN V VDE V 0884-10, each [ADuM3224](#) is proof tested by applying an insulation test voltage  $\geq 1050$  V peak for 1 second (partial discharge detection limit = 5 pC). An asterisk (\*) marking branded on the component designates DIN V VDE V 0884-10 approval.

The [ADuM4224](#) is approved or pending approval by the organizations listed in Table 7.

**Table 7.**

UL	CSA	VDE (Pending)
Recognized under UL 1577 Component Recognition Program <sup>1</sup>	Approved under CSA Component Acceptance Notice 5A	Certified according to DIN V VDE V 0884-10 (VDE V 0884-10): 2006-12 <sup>2</sup>
Single/Protection 5000 V rms Isolation Voltage	Reinforced insulation per CSA 60950-1-07 and IEC 60950-1, 380 V rms (537 V peak) maximum working voltage Basic insulation per CSA 60950-1-07 and IEC 60950-1, 760 V rms (1074 V peak) maximum working voltage	Reinforced insulation, 849 V peak
File E214100	File 205078	File 2471900-4880-0001

<sup>1</sup> In accordance with UL 1577, each [ADuM4224](#) is proof tested by applying an insulation test voltage  $\geq 6000$  V rms for 1 second (current leakage detection limit = 10  $\mu$ A).

<sup>2</sup> In accordance with DIN V VDE V 0884-10, each [ADuM4224](#) is proof tested by applying an insulation test voltage  $\geq 1590$  V peak for 1 second (partial discharge detection limit = 5 pC). An asterisk (\*) marking branded on the component designates DIN V VDE V 0884-10 approval.

**DIN V VDE V 0884-10 (VDE V 0884-10) INSULATION CHARACTERISTICS**

These isolators are suitable for reinforced isolation only within the safety limit data. Maintenance of the safety data is ensured by protective circuits. The asterisk (\*) marking on the package denotes DIN V VDE V 0884-10 approval for a 560 V peak working voltage.

**Table 8. ADuM3224 VDE Characteristics (Pending)**

Description	Test Conditions/Comments	Symbol	Characteristic	Unit
Installation Classification per DIN VDE 0110 For Rated Mains Voltage $\leq 150$ V rms For Rated Mains Voltage $\leq 300$ V rms For Rated Mains Voltage $\leq 400$ V rms			I to IV I to III I to II	
Climatic Classification			40/105/21	
Pollution Degree per DIN VDE 0110, Table 1			2	
Maximum Working Insulation Voltage		$V_{IORM}$	560	V peak
Input-to-Output Test Voltage, Method B1	$V_{IORM} \times 1.875 = V_{pd(m)}$ , 100% production test, $t_{ini} = t_m = 1$ sec, partial discharge $< 5$ pC	$V_{pd(m)}$	1050	V peak
Input-to-Output Test Voltage, Method A	$V_{IORM} \times 1.5 = V_{pd(m)}$ , $t_{ini} = 60$ sec, $t_m = 10$ sec, partial discharge $< 5$ pC	$V_{pd(m)}$		
After Environmental Tests Subgroup 1			896	V peak
After Input and/or Safety Test Subgroup 2 and Subgroup 3	$V_{IORM} \times 1.2 = V_{pd(m)}$ , $t_{ini} = 60$ sec, $t_m = 10$ sec, partial discharge $< 5$ pC	$V_{pd(m)}$	672	V peak
Highest Allowable Overvoltage		$V_{IOTM}$	4242	V peak
Surge Isolation Voltage	$V_{PEAK} = 10$ kV, 1.2 $\mu$ s rise time, 50 $\mu$ s, 50% fall time	$V_{IOSM}$	6000	V peak
Safety-Limiting Values	Maximum value allowed in the event of a failure (see Figure 2)			
Maximum Junction Temperature		$T_S$	150	$^{\circ}$ C
Safety Total Dissipated Power		$P_S$	1.64	W
Insulation Resistance at $T_S$	$V_{IO} = 500$ V	$R_S$	$>10^9$	$\Omega$

**Table 9. ADuM4224 VDE Characteristics (Pending)**

Description	Test Conditions/Comments	Symbol	Characteristic	Unit
Installation Classification per DIN VDE 0110 For Rated Mains Voltage $\leq 150$ V rms For Rated Mains Voltage $\leq 300$ V rms For Rated Mains Voltage $\leq 400$ V rms			I to IV I to III I to II	
Climatic Classification			40/105/21	
Pollution Degree per DIN VDE 0110, Table 1			2	
Maximum Working Insulation Voltage		$V_{IORM}$	849	V peak
Input-to-Output Test Voltage, Method B1	$V_{IORM} \times 1.875 = V_{pd(m)}$ , 100% production test, $t_{ini} = t_m = 1$ sec, partial discharge $< 5$ pC	$V_{pd(m)}$	1592	V peak
Input-to-Output Test Voltage, Method A	$V_{IORM} \times 1.5 = V_{pd(m)}$ , $t_{ini} = 60$ sec, $t_m = 10$ sec, partial discharge $< 5$ pC	$V_{pd(m)}$		
After Environmental Tests Subgroup 1			1273	V peak
After Input and/or Safety Test Subgroup 2 and Subgroup 3	$V_{IORM} \times 1.2 = V_{pd(m)}$ , $t_{ini} = 60$ sec, $t_m = 10$ sec, partial discharge $< 5$ pC	$V_{pd(m)}$	1018	V peak
Highest Allowable Overvoltage		$V_{IOTM}$	7071	V peak
Surge Isolation Voltage	$V_{PEAK} = 10$ kV, 1.2 $\mu$ s rise time, 50 $\mu$ s, 50% fall time	$V_{IOSM}$	6000	V peak
Safety Limiting Values	Maximum value allowed in the event of a failure (see Figure 3)			
Maximum Junction Temperature		$T_S$	150	$^{\circ}$ C
Safety Total Dissipated Power		$P_S$	2.77	W
Insulation Resistance at $T_S$	$V_{IO} = 500$ V	$R_S$	$>10^9$	$\Omega$



Figure 2. ADuM3224 Thermal Derating Curve, Dependence of Safety-Limiting Values on Case Temperature, per DIN V VDE V 0884-10



Figure 3. ADuM4224 Thermal Derating Curve, Dependence of Safety-Limiting Values on Case Temperature, per DIN V VDE V 0884-10

## RECOMMENDED OPERATING CONDITIONS

Table 10.

Parameter	Symbol	Rating
Operating Junction Temperature	$T_J$	-40°C to +125°C
Supply Voltages <sup>1</sup>		
$V_{DD1}$	$V_{DD1}$	3.0 V to 5.5 V
$V_{DDA}, V_{ddb}$	$V_{DDA}, V_{ddb}$	4.5 V to 18 V
$V_{DD1}$ Rise Time	$t_{VDD1}$	1 V/ $\mu$ s
$V_{DDA}, V_{ddb}$ Rise Time	$t_{VDDA}, t_{Vddb}$	10 V/ $\mu$ s
Maximum Input Signal Rise and Fall Times	$t_{VIA}, t_{VIB}$	1 ms
Common-Mode Transient, Static <sup>2</sup>		-50 kV/ $\mu$ s to +50 kV/ $\mu$ s
Common-Mode Transient Immunity, Dynamic <sup>3</sup>		-25 kV/ $\mu$ s to +25 kV/ $\mu$ s

- <sup>1</sup> All voltages are relative to their respective ground. See the Applications Information section for information on immunity to external magnetic fields.
- <sup>2</sup> Static common-mode transient immunity is defined as the largest  $dv/dt$  between  $GND_1$  and  $GND_A/GND_B$  with inputs held either high or low such that the output voltage remains either above  $0.8 \times V_{DDA}/V_{ddb}$  for  $V_{IA}/V_{IB} = \text{high}$ , or 0.8 V for  $V_{IA}/V_{IB} = \text{low}$ . Operation with transients above the recommended levels can cause momentary data upsets.
- <sup>3</sup> Dynamic common-mode transient immunity is defined as the largest  $dv/dt$  between  $GND_1$  and  $GND_A/GND_B$  with the switching edge coincident with the transient test pulse. Operation with transients above the recommended levels can cause momentary data upsets.



## ABSOLUTE MAXIMUM RATINGS

Ambient temperature = 25°C, unless otherwise noted.

Table 11.

Parameter	Rating
Storage Temperature (T <sub>ST</sub> )	–55°C to +150°C
Operating Junction Temperature (T <sub>J</sub> )	–40°C to +150°C
Supply Voltages <sup>1</sup>	
V <sub>DD1</sub>	–0.5 V to +7.0 V
V <sub>DDA</sub> , V <sub>DDB</sub>	–0.5 V to +20 V
Input Voltage (V <sub>IA</sub> , V <sub>IB</sub> , DISABLE) <sup>1</sup>	–0.5 V to V <sub>DD1</sub> + 0.5 V
Output Voltage <sup>1</sup>	
V <sub>OA</sub>	–0.5 V to V <sub>DDA</sub> + 0.5 V
V <sub>OB</sub>	–0.5 V to V <sub>DDB</sub> + 0.5 V
Average Output Current, per Pin (I <sub>O</sub> ) <sup>2</sup>	–35 mA to +35 mA
Common-Mode Transients (CM <sub>H</sub> , CM <sub>L</sub> ) <sup>3</sup>	–100 kV/μs to +100 kV/μs

<sup>1</sup> All voltages are relative to their respective ground.

<sup>2</sup> See Figure 2 and Figure 3 for information on maximum allowable current for various temperatures.

<sup>3</sup> Refers to common-mode transients across the insulation barrier. Common-mode transients exceeding the absolute maximum rating can cause latch-up or permanent damage.

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

### ESD CAUTION



**ESD (electrostatic discharge) sensitive device.** Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

Table 12. Maximum Continuous Working Voltage<sup>1</sup>

Parameter	Max	Unit	Constraint
AC Voltage, Bipolar Waveform	565	V peak	50-year minimum lifetime
AC Voltage, Unipolar Waveform	1131	V peak	50-year minimum lifetime
DC Voltage	1131	V peak	50-year minimum lifetime

<sup>1</sup> Refers to the continuous voltage magnitude imposed across the isolation barrier. See the Insulation Lifetime section for more information.

Table 13. ADuM3224/ADuM4224 (Positive Logic) Truth Table<sup>1</sup>

DISABLE	V <sub>IA</sub> Input	V <sub>IB</sub> Input	V <sub>DD1</sub> State	V <sub>DDA</sub> /V <sub>DDB</sub> State	V <sub>OA</sub> Output	V <sub>OB</sub> Output	Notes
L	L	L	Powered	Powered	L	L	Outputs return to the input state within 1 μs of DISABLE = L assertion.
L	L	H	Powered	Powered	L	H	Outputs return to the input state within 1 μs of DISABLE = L assertion.
L	H	L	Powered	Powered	H	L	Outputs return to the input state within 1 μs of DISABLE = L assertion.
L	H	H	Powered	Powered	H	H	Outputs return to the input state within 1 μs of DISABLE = L assertion.
H	X	X	Powered	Powered	L	L	Outputs take on default low state within 3 μs of DISABLE = H assertion.
L	L	L	Unpowered	Powered	L	L	Outputs return to the input state within 1 μs of V <sub>DD1</sub> power restoration.
X	X	X	Powered	Unpowered	L	L	Outputs return to the input state within 50 μs of V <sub>DDA</sub> /V <sub>DDB</sub> power restoration.

<sup>1</sup> X = don't care, L = low, and H = high.

## PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



## NOTES

1. NC = NO CONNECT. NOT INTERNALLY CONNECTED.

11791-003

Figure 4. Pin Configuration

Table 14. ADuM3224/ADuM4224 Pin Function Descriptions

Pin No. <sup>1</sup>	Mnemonic	Description
1	V <sub>IA</sub>	Logic Input A.
2	V <sub>IB</sub>	Logic Input B.
3, 8	V <sub>DD1</sub>	Input Supply Voltage.
4	GND <sub>1</sub>	Ground Reference for Input Logic Signals.
5	DISABLE	Input Disable. Disables the isolator inputs and refresh circuits. Outputs take on the default low state within 3 μs of a DISABLE = high assertion. Outputs return to the input state within 1 μs of a DISABLE = low assertion.
6, 7, 12, 13	NC	No Connect. These pins are not internally connected.
9	GND <sub>B</sub>	Ground Reference for Output B.
10	V <sub>OB</sub>	Output B.
11	V <sub>DDB</sub>	Output B Supply Voltage.
14	GND <sub>A</sub>	Ground Reference for Output A.
15	V <sub>OA</sub>	Output A.
16	V <sub>DDA</sub>	Output A Supply Voltage.

<sup>1</sup> Pin 3 and Pin 8 are internally connected; connecting both pins to the V<sub>DD1</sub> supply is recommended.

### TYPICAL PERFORMANCE CHARACTERISTICS



Figure 5. Output Waveform for 2 nF Load with 12 V Output Supply



Figure 6. Output Matching and Rise Time Waveforms for 2 nF Load with 12 V Output Supply



Figure 7. Typical ADuM3224 Maximum Load vs. Switching Frequency ( $R_G = 1 \Omega$ )



Figure 8. Typical ADuM4224 Maximum Load vs. Switching Frequency ( $R_G = 1 \Omega$ )



Figure 9. Typical  $I_{DD1}$  Supply Current vs. Frequency



Figure 10. Typical  $I_{DDA}$ ,  $I_{DDB}$  Supply Current vs. Frequency with 2 nF Load



Figure 11. Typical Propagation Delay vs. Junction Temperature



Figure 14. Typical Rise/Fall Time Variation vs. Output Supply Voltage



Figure 12. Typical Propagation Delay vs. Input Supply Voltage;  $V_{DDA}, V_{DDB} = 12\text{ V}$



Figure 15. Typical Propagation Delay (PD) Channel-to-Channel Matching vs. Output Supply Voltage



Figure 13. Typical Propagation Delay vs. Output Supply Voltage,  $V_{DD1} = 5\text{ V}$



Figure 16. Typical Propagation Delay (PD) Channel-to-Channel Matching vs. Temperature;  $V_{DDA}, V_{DDB} = 12\text{ V}$



Figure 17. Typical Output Resistance ( $R_{out}$ ) vs. Output Supply Voltage



Figure 18. Typical Source/Sink Output Current vs. Output Supply Voltage

## APPLICATIONS INFORMATION

### PRINTED CIRCUIT BOARD LAYOUT

The ADuM3224/ADuM4224 digital isolators require no external interface circuitry for the logic interfaces. Power supply bypassing is required at the input and output supply pins, as shown in Figure 19. Use a small ceramic capacitor with a value between 0.01  $\mu\text{F}$  and 0.1  $\mu\text{F}$  to provide a good high frequency bypass. On the output power supply pin,  $V_{\text{DDA}}$  or  $V_{\text{DDB}}$ , it is recommended to also add a 10  $\mu\text{F}$  capacitor to provide the charge required to drive the gate capacitance at the ADuM3224/ADuM4224 outputs. On the output supply pin, avoid the use of vias on the bypass capacitor, or use multiple vias to reduce the inductance in the bypassing. The total lead length between both ends of the smaller capacitor and the input or output power supply pin must not exceed 5 mm. For specific layout guidelines, refer to the [AN-1109 Application Note, Recommendations for Control of Radiated Emissions with iCoupler Devices](#).



Figure 19. Recommended PCB Layout

### UNDERVOLTAGE LOCKOUT

For the output of a channel of the ADuM3224/ADuM4224 to be valid, both the  $V_{\text{DD1}}$  and the  $V_{\text{DDA}}$  or  $V_{\text{DDB}}$  power supplies must be above the positive going undervoltage lockout (UVLO) threshold. If, during operation, the supply voltage drops below the negative going UVLO threshold, the output is brought low to protect the switch from being underdriven. The  $V_{\text{DD1}}$  threshold is typically around 2.5 V. There are three options for the secondary supply thresholds, which can be selected by the different grades (see the Ordering Guide). The UVLO of each output channel acts independently of the other, but in the case of a  $V_{\text{DD1}}$  UVLO, both channels are brought low.

### PROPAGATION DELAY-RELATED PARAMETERS

Propagation delay is a parameter that describes the time it takes a logic signal to propagate through a component. The propagation delay to a logic low output can differ from the propagation delay to a logic high output. The ADuM3224/ADuM4224 specify  $t_{\text{DLH}}$  (see Figure 20) as the time between the rising input high logic threshold,  $V_{\text{IH}}$ , to the output rising 10% threshold. Likewise, the falling propagation delay,  $t_{\text{DHL}}$ , is defined as the time between the input falling logic low threshold,  $V_{\text{IL}}$ , and the output falling 90% threshold. The rise and fall times are dependent on the loading conditions and are not included in the propagation delay, which is the industry standard for gate drivers.



Figure 20. Propagation Delay Parameters

Channel-to-channel matching refers to the maximum amount that the propagation delay differs between channels within a single ADuM3224/ADuM4224 component.

Propagation delay skew refers to the maximum amount that the propagation delay differs between multiple ADuM3224/ADuM4224 components operating under the same conditions.

### THERMAL LIMITATIONS AND SWITCH LOAD CHARACTERISTICS

For isolated gate drivers, the necessary separation between the input and output circuits prevents the use of a single thermal pad beneath the device, and heat is, therefore, dissipated mainly through the package pins.

Package thermal dissipation limits the performance of output load vs. switching frequency, as illustrated in Figure 7 and Figure 8 for the maximum load capacitance that can be driven with a 1  $\Omega$  series gate resistance for different values of output voltage. For example, Figure 7 shows that a typical ADuM3224 can drive a large MOSFET with 140 nC gate charge at 8 V output (which is equivalent to a 17 nF load) up to a frequency of about 300 kHz.

Do not allow the internal junction temperature of the ADuM3224/ADuM4224 to exceed the maximum junction temperature of 150°C. Operation above this value causes damage to the device. There is no internal thermal shutdown to protect the ADuM3224/ADuM4224. If a thermal shutdown is desired, see the ADuM3223/ADuM4223 data sheet.

### OUTPUT LOAD CHARACTERISTICS

The ADuM3224/ADuM4224 output signals depend on the characteristics of the output load, which is typically an N-channel MOSFET. The driver output response to an N-channel MOSFET load can be modeled with a switch output resistance ( $R_{\text{SW}}$ ), an inductance due to the printed circuit board trace ( $L_{\text{TRACE}}$ ), a series gate resistor ( $R_{\text{GATE}}$ ), and a gate-to-source capacitance ( $C_{\text{GS}}$ ), as shown in Figure 21.



Figure 21. RLC Model of the Gate of an N-Channel MOSFET

$R_{SW}$  is the switch resistance of the internal ADuM3224/ADuM4224 driver output, which is about 1.1  $\Omega$ .  $R_{GATE}$  is the intrinsic gate resistance of the MOSFET and any external series resistance. A MOSFET that requires a 4 A gate driver has a typical intrinsic gate resistance of about 1  $\Omega$  and a gate-to-source capacitance,  $C_{GS}$ , of between 2 nF and 10 nF.  $L_{TRACE}$  is the inductance of the printed circuit board trace, typically a value of 5 nH or less for a well designed layout with a very short and wide connection from the ADuM3224/ADuM4224 output to the gate of the MOSFET.

The following equation defines the Q factor of the resistor/inductor/capacitor (RLC) circuit, which indicates how the ADuM3224/ADuM4224 output responds to a step change. For a well damped output, Q is less than 1. Adding a series gate resistance dampens the output response.

$$Q = \frac{1}{(R_{SW} + R_{GATE})} \times \sqrt{\frac{L_{TRACE}}{C_{GS}}}$$

In Figure 5, the ADuM3224/ADuM4224 output waveforms for a 12 V output are shown for a  $C_{GS}$  of 2 nF. Note the small amount of ringing of the output in Figure 5 with a  $C_{GS}$  of 2 nF,  $R_{SW}$  of 1.1  $\Omega$ ,  $R_{GATE}$  of 0  $\Omega$ , and a calculated Q factor of 0.75, where less than 1 is desired for good damping.

Output ringing can be reduced by adding a series gate resistance to dampen the response. For applications of less than 1 nF load, it is recommended to add a series gate resistor of about 2  $\Omega$  to 5  $\Omega$ .

### BOOTSTRAPPED HALF-BRIDGE OPERATION

The ADuM3224/ADuM4224 are well suited to the operation of two output gate signals that are referenced to separate grounds, as in the case of a half-bridge configuration. Because isolated auxiliary supplies are often expensive, it is beneficial to reduce the amount of supplies. One method to perform this is to use a bootstrap configuration for the high-side supply of the ADuM3224/ADuM4224. In this topology, the decoupling capacitor,  $C_A$ , acts as the energy storage for the high-side supply, and is filled whenever the low-side switch is closed, bringing  $GND_A$  to  $GND_B$ . During the charging time of  $C_A$ , the  $dv/dt$  of

the  $V_{DDA}$  voltage must be controlled to reduce the possibility of glitches on the output. Keeping the  $dv/dt$  below 10 V/ $\mu$ s is recommended for the ADuM3224/ADuM4224. This can be controlled by introducing a series resistance,  $R_{BOOT}$ , into the charging path of  $C_A$ . As an example, if  $V_{AUX}$  is 12 V,  $C_A$  has a total capacitance of 10  $\mu$ F, and the forward voltage drop of the bootstrap diode is 1 V.

$$V_{BOOT} = \frac{V_{AUX} - V_{D_{BOOT}}}{C_A \times \frac{dv}{dt_{MAX}}} = \frac{12 \text{ V} - 1 \text{ V}}{10 \mu\text{F} \times 10 \text{ V}/\mu\text{s}} = 0.11 \Omega$$

### DC CORRECTNESS AND MAGNETIC FIELD IMMUNITY

Positive and negative logic transitions at the isolator input cause narrow ( $\sim$ 1 ns) pulses to be sent to the decoder via the transformer. The decoder is bistable and is, therefore, either set or reset by the pulses, indicating input logic transitions. In the absence of logic transitions of more than 1  $\mu$ s at the input, a periodic set of refresh pulses indicative of the correct input state are sent to ensure dc correctness at the output.

If the decoder receives no internal pulses for more than about 3  $\mu$ s, the input side is assumed to be unpowered or nonfunctional, in which case, the isolator output is forced to a default low state by the watchdog timer circuit. In addition, the outputs are in a low default state while the power is coming up before the UVLO threshold is crossed.

The ADuM3224/ADuM4224 are immune to external magnetic fields. The limitation on the ADuM3224/ADuM4224 magnetic field immunity is set by the condition in which induced voltage in the transformer receiving coil is sufficiently large to either falsely set or reset the decoder. The following analysis defines the conditions under which this can occur. The 3 V operating condition of the ADuM3224/ADuM4224 is examined because it represents the most susceptible mode of operation. The pulses at the transformer output have an amplitude greater than 1.0 V. The decoder has a sensing threshold at about 0.5 V, therefore establishing a 0.5 V margin in which induced voltages can be tolerated. The voltage induced across the receiving coil is given by

$$V = (-d\beta/dt) \sum \pi r_n^2, n = 1, 2, \dots, N$$

where:

$\beta$  is the magnetic flux density (gauss).

$r_n$  is the radius of the nth turn in the receiving coil (cm).

$N$  is the number of turns in the receiving coil.



Figure 22. Circuit of Bootstrapped Half-Bridge Operation

Given the geometry of the receiving coil in the ADuM3224/ADuM4224 and an imposed requirement that the induced voltage be, at most, 50% of the 0.5 V margin at the decoder, a maximum allowable magnetic field is calculated, as shown in Figure 23.



Figure 23. Maximum Allowable External Magnetic Flux Density

For example, at a magnetic field frequency of 1 MHz, the maximum allowable magnetic field of 0.08 kgauss induces a voltage of 0.25 V at the receiving coil. This is about 50% of the sensing threshold and does not cause a faulty output transition. Similarly, if such an event were to occur during a transmitted pulse (and had the worst-case polarity), the received pulse is reduced from >1.0 V to 0.75 V, still well above the 0.5 V sensing threshold of the decoder.

The preceding magnetic flux density values correspond to specific current magnitudes at given distances away from the ADuM3224/ADuM4224 transformers. Figure 24 expresses these allowable current magnitudes as a function of frequency for selected distances. As shown in Figure 24, the ADuM3224/ADuM4224 are immune and only can be affected by extremely large currents operated at a high frequency and very close to the component. For the 1 MHz example, a 0.2 kA current must be placed 5 mm from the ADuM3224/ADuM4224 to affect the operation of the component.

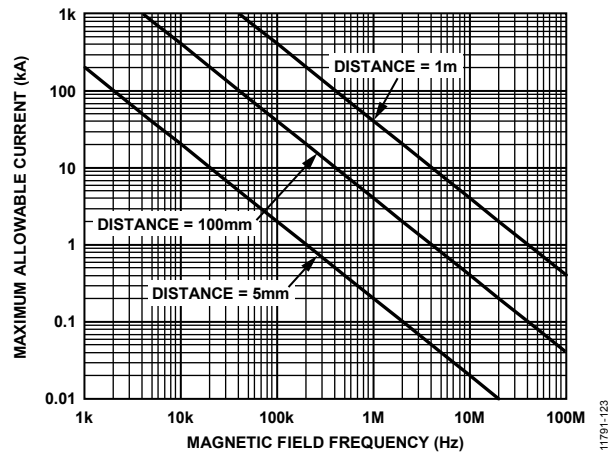


Figure 24. Maximum Allowable Current for Various Current-to-ADuM3224/ADuM4224 Spacings



## POWER CONSUMPTION

The supply current at a given channel of the [ADuM3224/ADuM4224](#) isolator is a function of the supply voltage, channel data rate, and channel output load.

During the driving of a MOSFET gate, the driver must dissipate power. This power is not insignificant and can lead to thermal shutdown (TSD) if considerations are not made. The gate of a MOSFET can be simulated approximately as a capacitive load. Due to Miller capacitance and other nonlinearities, it is common practice to take the stated input capacitance,  $C_{ISS}$ , of a given MOSFET and multiply it by a factor of 5 to arrive at a conservative estimate to approximate the load being driven. With this value, the estimated total power dissipation per channel due to switching action is given by

$$P_{DISS} = C_{EST} \times (V_{DDx})^2 \times f_s$$

where:

$$C_{EST} = C_{ISS} \times 5.$$

$f_s$  is the switching frequency.

Alternately, use the gate charge to obtain a more precise value for  $P_{DISS}$ .

$$P_{DISS} = Q_{GATE} \times V_{DDx} \times f_s$$

where:

$Q_{GATE}$  is the gate charge for the MOSFET.

$f_s$  is the switching frequency.

This power dissipation is shared between the internal on resistances of the internal gate driver switches and the external gate resistances,  $R_{GON}$  and  $R_{GOFF}$ . The ratio of the internal gate resistances to the total series resistance allows the calculation of losses seen within the [ADuM3224/ADuM4224](#) chips per channel.

$$P_{DISS\_IC} = P_{DISS} \times \frac{1}{2} \times \left( \frac{R_{DSON\_P}}{R_{EXT\_X} + R_{DSON\_P}} + \frac{R_{DSON\_N}}{R_{EXT\_X} + R_{DSON\_N}} \right)$$

Taking the power dissipation found inside the chip and multiplying it by  $\theta_{JA}$  gives the rise above ambient temperature that the [ADuM3224/ADuM4224](#) experiences, multiplied by two to reflect that there are two channels.

$$T_j = \theta_{JA} \times 2 \times P_{DISS\_IC} + T_{AMB}$$

For the device to remain within specification,  $T_j$  must not exceed 125°C. If  $T_j$  exceeds 150°C (typical), the device enters TSD.

Quiescent power dissipation may also be added to give a more accurate number for temperature rise, but the switching power losses are often the largest source of power dissipation, and quiescent losses can often be ignored. To calculate the total supply current, the quiescent supply currents for each input and output channel corresponding to  $I_{DD1(Q)}$ ,  $I_{DDA(Q)}$ , and  $I_{DDB(Q)}$  are added. The full equation for the  $T_j$  becomes

$$T_j = \theta_{JA} \times (2 \times P_{DISS\_IC} + V_{DD1} \times I_{DD1(Q)} + V_{DDA} \times I_{DDA(Q)} + V_{DDB} \times I_{DDB(Q)}) + T_{AMB}$$

Figure 9 provides total input  $I_{DD1}$  supply current as a function of data rate for both input channels. Figure 10 provides total  $I_{DDA}$  or  $I_{DDB}$  supply current as a function of data rate for both outputs loaded with 2 nF capacitance.

## INSULATION LIFETIME

All insulation structures eventually break down when subjected to voltage stress over a sufficiently long period. The rate of insulation degradation is dependent on the characteristics of the voltage waveform applied across the insulation. In addition to the testing performed by the regulatory agencies, Analog Devices carries out an extensive set of evaluations to determine the lifetime of the insulation structure within the [ADuM3224/ADuM4224](#).

Analog Devices performs accelerated life testing using voltage levels higher than the rated continuous working voltage. Acceleration factors for several operating conditions are determined. These factors allow calculation of the time to failure at the actual working voltage.

The values shown in Table 12 summarize the peak voltage for 50 years of service life for a bipolar ac operating condition and the maximum CSA/VDE approved working voltages. In many cases, the approved working voltage is higher than the 50-year service life voltage. Operation at these high working voltages can lead to shortened insulation life in some cases.

The insulation lifetime of the [ADuM3224/ADuM4224](#) depends on the voltage waveform type imposed across the isolation barrier. The *iCoupler* insulation structure degrades at different rates depending on whether the waveform is bipolar ac, unipolar ac, or dc. Figure 25, Figure 26, and Figure 27 illustrate these different isolation voltage waveforms.

A bipolar ac voltage environment is the worst case for the *iCoupler* products and is the 50-year operating lifetime that Analog Devices recommends for maximum working voltage. In the case of unipolar ac or dc voltage, the stress on the insulation is significantly lower. This allows operation at higher working voltages while still achieving a 50-year service life. Treat any cross insulation voltage waveform that does not conform to Figure 26 or Figure 27 as a bipolar ac waveform, and limit its peak voltage to the 50-year lifetime voltage value listed in Table 12.

Note that the voltage presented in Figure 26 is shown as sinusoidal for illustration purposes only. It is meant to represent any voltage waveform varying between 0 V and some limiting value. The limiting value can be positive or negative, but the voltage cannot cross 0 V.



Figure 25. Bipolar AC Waveform



Figure 26. Unipolar AC Waveform



Figure 27. DC Waveform

OUTLINE DIMENSIONS



COMPLIANT TO JEDEC STANDARDS MS-012-AC  
 CONTROLLING DIMENSIONS ARE IN MILLIMETERS; INCH DIMENSIONS  
 (IN PARENTHESES) ARE ROUNDED-OFF MILLIMETER EQUIVALENTS FOR  
 REFERENCE ONLY AND ARE NOT APPROPRIATE FOR USE IN DESIGN.

Figure 28. 16-Lead Standard Small Outline Package [SOIC\_N]  
 Narrow Body  
 (R-16)

Dimensions shown in millimeters and (inches)

060606-A



COMPLIANT TO JEDEC STANDARDS MS-013-AA  
 CONTROLLING DIMENSIONS ARE IN MILLIMETERS; INCH DIMENSIONS  
 (IN PARENTHESES) ARE ROUNDED-OFF MILLIMETER EQUIVALENTS FOR  
 REFERENCE ONLY AND ARE NOT APPROPRIATE FOR USE IN DESIGN.

Figure 29. 16-Lead Standard Small Outline Package [SOIC\_W]  
 Wide Body  
 (RW-16)

Dimensions shown in millimeters and (inches)

03-27-2007-B

## ORDERING GUIDE

Model <sup>1, 2</sup>	No. of Channels	Output Peak Current (A)	Minimum Output Voltage (V)	Temperature Range	Package Description	Package Option	Ordering Quantity
ADuM3224WARZ	2	4	4.5	-40°C to +125°C	16-Lead SOIC_N	R-16	
ADuM3224WARZ-RL7	2	4	4.5	-40°C to +125°C	16-Lead SOIC_N, 7" Tape and Reel	R-16	1,000
ADuM3224WBRZ	2	4	7.5	-40°C to +125°C	16-Lead SOIC_N	R-16	
ADuM3224WBRZ-RL7	2	4	7.5	-40°C to +125°C	16-Lead SOIC_N, 7" Tape and Reel	R-16	1,000
ADuM3224WCRZ	2	4	11.5	-40°C to +125°C	16-Lead SOIC_N	R-16	
ADuM3224WCRZ-RL7	2	4	11.5	-40°C to +125°C	16-Lead SOIC_N, 7" Tape and Reel	R-16	1,000
ADuM4224WARWZ	2	4	4.5	-40°C to +125°C	16-Lead SOIC_W	RW-16	
ADuM4224WARWZ-RL	2	4	4.5	-40°C to +125°C	16-Lead SOIC_W, 13" Tape and Reel	RW-16	1,000
ADuM4224WBRWZ	2	4	7.5	-40°C to +125°C	16-Lead SOIC_W	RW-16	
ADuM4224WBRWZ-RL	2	4	7.5	-40°C to +125°C	16-Lead SOIC_W, 13" Tape and Reel	RW-16	1,000
ADuM4224WCRWZ	2	4	11.5	-40°C to +125°C	16-Lead SOIC_W	RW-16	
ADuM4224WCRWZ-RL	2	4	11.5	-40°C to +125°C	16-Lead SOIC_W, 13" Tape and Reel	RW-16	1,000

<sup>1</sup> Z = RoHS Compliant Part.

<sup>2</sup> W = Qualified for Automotive Applications.

## AUTOMOTIVE PRODUCTS

The [ADuM3224W](#) and [ADuM4224W](#) models are available with controlled manufacturing to support the quality and reliability requirements of automotive applications. Note that these automotive models may have specifications that differ from the commercial models; therefore, designers should review the Specifications section of this data sheet carefully. Only the automotive grade products shown are available for use in automotive applications. Contact your local Analog Devices account representative for specific product ordering information and to obtain the specific Automotive Reliability reports for these models.

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Наши преимущества:

- Поставка оригинальных импортных электронных компонентов напрямую с производств Америки, Европы и Азии, а так же с крупнейших складов мира;
- Широкая линейка поставок активных и пассивных импортных электронных компонентов (более 30 млн. наименований);
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
- Экспресс доставка в любую точку России;
- Помощь Конструкторского Отдела и консультации квалифицированных инженеров;
- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
- Поставка электронных компонентов под контролем ВП;
- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001;
- При необходимости вся продукция военного и аэрокосмического назначения проходит испытания и сертификацию в лаборатории (по согласованию с заказчиком);
- Поставка специализированных компонентов военного и аэрокосмического уровня качества (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Actel, Aeroflex, Peregrine, VPT, Syfer, Eurofarad, Texas Instruments, MS Kennedy, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

Компания «Океан Электроники» является официальным дистрибьютором и эксклюзивным представителем в России одного из крупнейших производителей разъемов военного и аэрокосмического назначения «JONHON», а так же официальным дистрибьютором и эксклюзивным представителем в России производителя высокотехнологичных и надежных решений для передачи СВЧ сигналов «FORSTAR».



## JONHON

«JONHON» (основан в 1970 г.)

Разъемы специального, военного и аэрокосмического назначения:

(Применяются в военной, авиационной, аэрокосмической, морской, железнодорожной, горно- и нефтедобывающей отраслях промышленности)

«FORSTAR» (основан в 1998 г.)

ВЧ соединители, коаксиальные кабели,  
кабельные сборки и микроволновые компоненты:

(Применяются в телекоммуникациях гражданского и специального назначения, в средствах связи, РЛС, а так же военной, авиационной и аэрокосмической отраслях промышленности).



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