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FAN54053 High Efficiency, 1.55 A, Li-Ion Switching Charger with Power Path, USB-OTG, in a Small Solution Footprint

Features

- Fully Integrated, High-Efficiency Switch-Mode Charger for Single-Cell Li-Ion and Li-Polymer Batteries
- Power Path Circuit Ensures Fast System Startup with a Dead Battery when VBUS is Connected
- 1.55 A Maximum Charge Current
- Programmable High Accuracy Float Voltage:
 - ±0.5% at 25°C
 - ±1% from 0 to 125°C
- ±5% Input and Charge Current Regulation Accuracy
- Temperature-Sense Input for JEITA Compliance
- Thermal Regulation and Shutdown
- 4.2 V at 2.3 A Production Test Support
- 5 V, 500 mA Boost Mode for USB OTG
- 28 V Absolute Maximum Input Voltage
- 6 V Maximum Input Operating Voltage
- Programmable through High-Speed I²C Interface (3.4 Mb/s) with Fast Mode Plus Compatibility
 - Input Current
 - Fast-Charge / Termination Current
 - Float Voltage
 - Termination Enable
- 3 MHz Synchronous Buck PWM Controller with Wide Duty Cycle Range
- Small Footprint 1 µH External Inductor
- Safety Timer with Reset Control
- Dynamic Input Voltage Control
- Very Low Battery Current when Charger Inactive

Applications

- Cell Phones, Smart Phones, PDAs
- Tablet, Portable Media Players
- Gaming Device, Digital Cameras

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Description

The FAN54053 is a 1.55 A USB-compliant switch-mode charger featuring power path operation, USB OTG boost support, JEITA temperature control, and production test mode support, in a small 25 bump, 0.4 mm pitch WLCSP package.

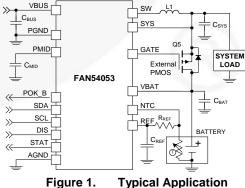
To facilitate fast system startup, the IC includes a power path circuit, which disconnects the battery from the system rail, ensuring that the system can power up quickly following a VBUS connection. The power path circuit ensures that the system rail stays up when the charger is plugged in, even if the battery is dead or shorted.

The charging parameters; float voltage, input voltage regulation, input current, charging current, and other operating modes are programmable through an I^2C Interface that operates up to 3.4 Mbps. The charger and boost regulator circuits switch at 3 MHz to minimize the size of external passive components.

The FAN54053 provides battery charging in three phases: conditioning, constant current and constant voltage. The IC automatically restarts the charge cycle when the battery falls below a voltage threshold. If the input source is removed, the IC enters a high-impedance mode blocking battery current from leaking to the input. Charge status is reported back to the host through the I^2 C port.

Dynamic input voltage control prevents a weak adapter's voltage from collapsing, ensuring charging capability from such adapters.

The FAN54053 is available in a space saving 2.4 mm x 2.0 mm WLCSP package.



April 2015

Part Number	Temperature Range	Packa	ige	PN Bits: IC_INFO[5:3]	Pack	king Methoo
FAN54053UCX	-40 to 85°C	25-Bump, Wafer-Lu Package (WLCSP	evel Chip-Scale), 0.4 mm Pitch	010	Тар	be and Reel
Table 1. Feature S	Summary					
Part Number	Slave Address	Automatic Charg	ge Battery Ab Behavi			Watchdog mer Defaul
FAN54053	1101011	No	On	POK_	В	Disabled
	CHARGE PUMP BUS OVP DWER OK IEUS & VBUS CONTROL BUS OVP DWER OK IIEUS & VBUS CONTROL IIEUS CON IIEUS CO		Q1 Q1 Q1 Q2 Q2 PGND PGND CC and CV Battery Charger TEL SEN			C _{MID}
PMID Greater than V _{BAT}	Q1/		S Greater than V _B ,	YS AT	Q4A ON	• Q4B OFF

Component	Description	Vendor	Parameter	Тур.	Unit
1.4	4 11 000/ 0.7.4 0040	Toko DFE201610E-1R0M	L	1.0	μH
L1	1 μH, 20%, 2.7 A, 2016	or Equivalent	DCR (Series R)	48	mΩ
$C_{\text{BAT,}}C_{\text{SYS}}$	10 μF, 20%, 6.3 V, X5R, 0603	Murata: GRM188R60J106M TDK: C1608X5R0J106M	С	10	μF
C _{MID}	4.7 μF, 10%, 10 V, X5R, 0603	Murata: GRM188R61A475K TDK: C1608X5R1A475K	C ⁽¹⁾	4.7	μF
C _{BUS,}	1.0 μF, 10%, 25 V, X5R, 0603	Murata GRM188R61E105K TDK:C1608X5R1E105M	С	1.0	μF
Q5	PMOS,12 V, 16 mΩ, MLP2x2	Fairchild FDMA905P	R _{DS(ON)}	16	mΩ
C _{REF}	1 μF, 10%, 6.3 V, X5R, 0402		С	1.0	μF

Note:

1. 10 V rating is sufficient for C_{MID} since PMID is protected from over-voltage surges on VBUS by Q3.

in a Small Solution Footprint

FAN54053— High Efficiency, 1.55 A, Li-Ion Switching Charger with Power Path, USB-OTG,

Pin Configuration SDA PGND PMID VBUS SW (A5) (A4) (A3) (A2) (A1) (A1) (A2) (A3) (A4) (A5) SCL **B**2 (B5) **B**4 **B**3 (B1) (B1) (B2) (B3) (B4) (B5) DIS GATE (C5) (C4) **C**3 C2 (C1) (C4) (C1) (C2) (C3) (C5) STAT SYS VBAT NTC (D5) (D4) (D3) (D2) (D1)(D1) (D2) (D3) (D4) (D5) POK_B AGND REF (E5) (E2) (E4) (E3) (E1) (E2) (E3) (E4) (E5)

Figure 3. **Top View** (E1)

Figure 4. **Bottom View**

Pin Definitions

Pin #	Name	Description
A1	SDA	I ² C Interface Serial Data. This pin should not be left floating.
B1	SCL	I ² C Interface Serial Clock. This pin should not be left floating.
C1	DIS	Disable . If this pin is held HIGH, Q1 and Q3 are turned off, creating a HIGH Z condition at VBUS and the PWM converter is disabled.
D1	STAT	Status . Open-drain output indicating charge status. The IC pulls this pin LOW when charge is in progress; can be used to signal the host processor when a fault condition occurs.
E1	POK_B	Power OK . Open-drain output that pulls LOW when VBUS is plugged in and the battery has risen above V_{LOWV} . This signal is used to signal the host processor that it can begin to draw significant current.
A2 – D2	PGND	Power Ground . Power return for gate drive and power transistors. The connection from this pin to the bottom of C_{MID} should be as short as possible.
E2	AGND	Analog Ground. All IC signals are referenced to this node.
A3 – C3	SW	Switching Node. Connect to output inductor.
D3 – E3	SYS	System Supply . Output voltage of the switching charger and input to the power path controller. Bypass SYS to PGND with a 10 μ F capacitor.
A4 – C4	PMID	Power Input Voltage . Power input to the charger regulator, bypass point for the input current sense. Bypass with a minimum of a 4.7 μ F, 6.3 V capacitor to PGND.
D4 – E4	VBAT	Battery Voltage . Connect to the positive (+) terminal of the battery pack. Bypass with a 10 μ F capacitor to PGND. VBAT is a power path connection.
A5 – B5	VBUS	Charger Input Voltage and USB-OTG output voltage. Bypass with a 1 μ F capacitor to PGND.
C5	GATE	External MOSFET Gate . This pin controls the gate of an external P-channel MOSFET transistor used to augment the internal ideal diode. The source of the P-channel MOSFET should be connected to SYS and the drain should be connected to VBAT.
D5	NTC	Thermistor input . The IC compares this node with taps on a resistor divider from REF to inhibit auto- charging when the battery temperature is outside of permitted fast-charge limits.
E5	REF	Reference Voltage. REF is a 1.8 V regulated output.

Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

Symbol		Parameter		Min.	Max.	Unit	
		Continuous		-0.3	00.0		
V _{BUS}	Voltage on VBUS Pin	Pulsed, 100 ms Max	-1.0	28.0	V		
	Voltage on PMID Voltage Pin		-0.3	7.0	V		
Vı	Voltage on SW, SYS, VBAT, ST	e on SW, SYS, VBAT, STAT, DIS Pins					
Vo	Voltage on Other Pins		-0.3	6.5 ⁽²⁾	V		
dV _{BUS} dt	Maximum V_{BUS} Slope Above 5.5	mum V_{BUS} Slope Above 5.5 V when Boost or Charger Active			4	V/µs	
	Electrostatic Discharge	Human Body Model	per JESD22-A114	2000		V	
500	Protection Level	Charged Device Mo	del per JESD22-C101	500		- V	
ESD	IEC 61000-4-2 System ESD ⁽³⁾	USB Connector	Air Gap	15			
	TEC 61000-4-2 System ESD	Pins (V_{BUS} to GND)	Contact	8		– kV	
TJ	Junction Temperature			-40	+150	°C	
T _{STG}	Storage Temperature			-65	+150	°C	
TL	Lead Soldering Temperature, 10	Seconds			+260	°C	

Notes:

2. Lesser of 6.5 V or V_1 + 0.3 V.

3. Guaranteed if $C_{BUS} \ge 1 \ \mu F$ and $C_{MID} \ge 4.7 \ \mu F$.

Recommended Operating Conditions

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to absolute maximum ratings.

Symbol	Parameter		Min.	Max.	Unit
V _{BUS}	Supply Voltage		4	6	V
V _{BAT(MAX)}	Maximum Battery Voltage when Boost enabled			4.5	V
$-\frac{dV_{BUS}}{dt}$	Negative VBUS Slew Rate during VBUS Short Circuit,	T _A <u><</u> 60°C		4	V/ve
dt	$C_{MID} \leq 4.7 \ \mu$ F, see VBUS Short While Charging	T _A <u>></u> 60°C		2	V/μs
T _A	Ambient Temperature		-30	+85	°C
TJ	Junction Temperature (see Register Bit section)		-30	+120	°C

Thermal Properties

Junction-to-ambient thermal resistance is a function of application and board layout. This data is measured with four-layer 2s2p boards in accordance to JEDEC standard JESD51. Special attention must be paid not to exceed junction temperature $T_{J(max)}$ at a given ambient temperature T_A .

Symbol	Parameter	Typical	Unit
θ_{JA}	Junction-to-Ambient Thermal Resistance	50	°C/W
θ_{JB}	Junction-to-PCB Thermal Resistance	20	°C/W

Electrical Specifications

Unless otherwise specified: according to the circuit of Figure 1; recommended operating temperature range for T_J and T_A ; $V_{BUS}=5.0 \text{ V}$; HZ_MODE="0"; OPA_MODE="0" (Charge Mode); SCL, SDA=0 or 1.8 V; and typical values are for $T_J=25^{\circ}$ C. Min. and Max. values are not tested in production, but are determined by characterization.

Symbol	Parameter		Conditions	Min.	Тур.	Max.	Unit
Power Sup	plies	•				1	
		PWM Switching			25		mA
I _{VBUS}	VBUS Current	V _{BAT} > V _{OREG} I _{BUSLIM} = 500 mA			6		mA
		$0^{\circ}C < T_J < 85^{\circ}C,$ $V_{BAT} > V_{LOWV}$	HZ_MODE = "1",		190	280	μA
I _{BAT_HZ}	Battery Discharge Current in High-Impedance Mode	DIS pin HIGH, or V _{BAT} =4.35 V	HZ_MODE = "1",		<1.25	10	μΑ
I _{BUS_HZ}	Battery Leakage Current to V _{BUS} in High-Impedance Mode	DIS pin HIGH, or V_{BUS} Shorted to G	HZ_MODE = "1", Ground, V _{BAT} =4.35 V	-5.0	-0.2		μΑ
Charger Vo	oltage Regulation						
	Charge Voltage Range			3.51		4.45	V
V _{OREG}		$T_A = 25^{\circ}C, V_{OREG}$	= 4.35 V	-0.5		+0.5	%
	Charge Voltage Accuracy	T _J =0 to 125°C		-1		+1	%
Charging C	Current Regulation (Fast Charge	ge)		•			
	Output Charge Current	V _{LOWV} < V _{BAT} <	IO_LEVEL = "0"	550		1550	mA
I _{OCHRG}	Range	V _{OREG}	IO_LEVEL = "1" (default)	165	200	230	mA
	Charge Current Accuracy	IO_LEVEL = "0"		-5		+5	%
Weak Batte	ery Detection	•		•		•	
	Weak Battery Threshold Range			3.35		3.75	V
V_{LOWV}	Weak Battery Threshold Accuracy			-5		+5	%
	Weak Battery Deglitch Time	Rising Voltage, 2	mV Overdrive		32		ms
PWM Char	ging Threshold			•		•	•
VBATMIN	Rising PWM Charging Threshold			3.1	3.2	3.3	V
VBATFALL	Falling PWM Charging Threshold				3.0		V
Logic Leve	els: DIS, SDA, SCL	1					
VIH	High-Level Input Voltage			1.05			V
VIL	Low-Level Input Voltage					0.4	V
I _{IN}	Input Bias Current	Input Tied to GNE) or V _{BUS}		0.01	1.00	μA
R _{PD}	DIS Pull-Down Resistance				1		MΩ
Charge Ter	rmination Detection	L					
	Termination Current Range	$V_{BAT} > V_{OREG} - V_F$	RCH, VBUS > VSLP	50		400	mA
	Termination Current	I _{TERM} Setting < 10		-15		+15	
I _(TERM)	Accuracy	I _{TERM} Setting ≥ 20		-5		+5	- %
	Termination Current Deglitch Time ⁽⁴⁾				32		ms

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Electrical Specifications (Continued)

Unless otherwise specified: according to the circuit of Figure 1; recommended operating temperature range for T_J and T_A ; $V_{BUS}=5.0$ V; HZ_MODE="0"; OPA_MODE="0" (Charge Mode); SCL, SDA=0 or 1.8 V; and typical values are for $T_J=25^{\circ}$ C. Min. and Max. values are not tested in production, but are determined by characterization.

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit
Power Path ((Q4) Control (Precharge)					
		IO_LEVEL = "1" (default)	165	200	235	mA
	Power Path Maximum	IO_LEVEL = "0", I _{BUSLIM} ≤ "01"	165	200	235	mA
IPP	Charge Current	IO_LEVEL = "0", I _{BUSLIM} > "01", I _{OCHARGE} ≤ "02"	375	450	520	mA
		IO_LEVEL = "0", I _{BUSLIM} > "01", I _{OCHARGE} > "02"	610	200 200	840	mA
	VBAT to SYS Threshold	(SYS-VBAT) Falling	-6	-5	-3	mV
V _{THSYS}	for Q4 and Gate Transition While Charging	(SYS-VBAT) Rising	-1	+1	2	mV
Production 1	Fest Mode					
V _{BAT(PTM)} ⁽⁴⁾	Production Test Output Voltage	1 mA < I _{BAT} < 2 A, V _{BUS} =5.5 V	4.116	4.200	4.284	V
I _{BAT(PTM)} ⁽⁴⁾	Production Test Output Current	20% Duty with Max. Period 10 ms	2.3			А
Battery Tem	perature Monitor (NTC)	·		•	•	
T1	T1 (0°C) Temperature Threshold		71.9	73.9	75.9	
T2	T2 (10°C) Temperature Threshold		62.6	64.6	66.6	% 0
Т3	T3 (45°C) Temperature Threshold		31.9	32.9	34.9	Vref
T4	T4 (60°C) Temperature Threshold		21.3	23.3	25.3	
Input Power	Source Detection	-				
V _{IN(MIN)1}	VBUS Input Voltage Rising	To Initiate and Pass VBUS Validation		4.35	4.45	V
V _{IN(MIN)2}	Minimum VBUS during Charge	During Charging		3.71	3.94	V
$t_{VBUS_VALID}^{(4)}$	VBUS Validation Time			30		ms
V _{BUS} Control	Loop					
VBUSLIM	VBUS Loop Setpoint Accuracy		-3		+3	%
Input Curren	t Limit					
		I _{BUSLIM} = "00"	450	475	500	
I _{BUSLIM}	Charger Input Current Limit Threshold	I _{BUSLIM} = "01"		760		mA
		I _{BUSLIM} = "10"	972	1080	1188	
V _{REF} Bias Ge	nerator		I			
	Bias Regulator Voltage	V _{BUS} > V _{IN(MIN)1}		1.8		V
V_{REF}	Short-Circuit Current Limit			2.5		mA
Battery Rech	arge Threshold	•			•	
\ <i>\</i>	Recharge Threshold	Below V _{OREG}	100	120	150	m٧
V _{RCH}	Deglitch Time	V _{BAT} Falling Below V _{RCH} Threshold		130		ms
STAT, POK_	B Output					
V _{STAT(OL)}	STAT Output Low	I _{STAT} = 10 mA			0.4	V
I _{STAT(OH)}	STAT High Leakage Current	V _{STAT} = 5 V			1	μA

FAN54053-High Efficiency, 1.55 A, Li-Ion Switching Charger with Power Path, USB-OTG, in a Small Solution Footprint

Electrical Specifications (Continued)

Unless otherwise specified: according to the circuit of Figure 1; recommended operating temperature range for T_J and T_A ; $V_{BUS}=5.0 \text{ V}$; HZ_MODE="0"; OPA_MODE="0" (Charge Mode); SCL, SDA=0 or 1.8 V; and typical values are for $T_J=25^{\circ}$ C. Min. and Max. values are not tested in production, but are determined by characterization.

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit
Battery Dete	ction		1			
I _{DETECT}	Battery Detection Current before Charge Done (Sink Current) ⁽⁵⁾	Begins after Termination Detected and		-1.9		mA
t DETECT	Battery Detection Time	Vbat ≤ Voreg − Vrch		262		ms
Sleep Comp	arator					
V_{SLP}	Sleep-Mode Entry Threshold, V _{BUS} – V _{BAT}	2.3 V \leq V _{BAT} \leq V _{OREG} , V _{BUS} Falling	0	0.04	0.10	V
Power Switc	hes (see Figure 2)	·			•	
	Q3 On Resistance (VBUS to PMID)	$I_{IN(LIMIT)} = 500 \text{ mA}$		180	400	
-	Q1 On Resistance (PMID to SW)			130	225	mΩ
R _{DS(ON)}	Q2 On Resistance (SW to GND)			150	225	
	Q4 On Resistance (SYS to VBAT)	V _{BAT} =4.35 V		70	100	mΩ
I _{SYNC}	Synchronous to Non-Synchronous Current Cut-Off Threshold ⁽⁶⁾	Low-Side MOSFET (Q2) Cycle-by-Cycle Current Limit		180		mA
Charger PW	M Modulator					
f _{SW}	Oscillator Frequency		2.7	3.0	3.3	MHz
D _{MAX}	Maximum Duty Cycle				100	%
D _{MIN}	Minimum Duty Cycle			0		%
Boost Mode	Operation (OPA_MODE=1)	1				
		2.5 V < V _{BAT} < 4.5 V, I _{LOAD} from 0 to 200 mA	4.80	5.07	5.20	V
V _{BOOST}	Boost Output Voltage at VBUS	$3.0~V < V_{BAT} < 4.5~V,~I_{LOAD}$ from 0 to 500 mA	4.77	5.07	5.20	V
IBAT(BOOST)	Boost Mode Quiescent Current	PFM Mode, $V_{BAT} = 3.6 \text{ V}$, $I_{LOAD} = 0 \text{ A}$		250	350	μA
I _{LIMPK(BST)}	Q2 Peak Current Limit		1350	1550	1950	mA
	Minimum Battery Voltage for Boost	While Boost Active		2.32		
UVLOBST	Operation	To Start Boost Regulator		2.48	2.70	V
VBUS Load	Resistance					
_		Normal Operation		500		kΩ
R _{VBUS}	VBUS to PGND Resistance	VBUS Validation		100		Ω
Protection a	nd Timers					
	VBUS Over-Voltage Shutdown	V _{BUS} Rising	6.09	6.29	6.49	V
VBUSOVP	Hysteresis	V _{BUS} Falling		100		mV
ILIMPK(CHG)	Q1 Cycle-by-Cycle Peak Current Limit	Charge Mode		3		A
	Battery Short-Circuit Threshold	V _{BAT} Rising	1.95	2.00	2.07	V
V _{SHORT}	Hysteresis			100		mV
I _{SHORT}	Linear Charging Current	V _{BAT} < V _{SHORT}		30		mA
T	Thermal Shutdown Threshold ⁽⁴⁾	T _J Rising		145		
TSHUTDWN	Hysteresis ⁽⁴⁾	TJ Falling		25		°C
T _{CF}	Thermal Regulation Threshold ⁽⁴⁾	Charge Current Reduction Begins		120	1	°C
t _{INT}	Detection Interval	~		2.1	1	s

Electrical Specifications (Continued)

Unless otherwise specified: according to the circuit of Figure 1; recommended operating temperature range for T_J and T_A ; V_{BUS} =5.0 V; HZ_MODE; OPA_MODE=0; (Charge Mode); SCL, SDA=0 or 1.8 V; and typical values are for T_J =25°C. Min. and Max. values are not tested in production, but are determined by characterization.

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit
	32-Second Timer ⁽⁷⁾	Charger Enabled	20.5	25.2	28.0	
t _{32S}		Charger Disabled	18.0	25.2	34.0	S
t _{15MIN}	15-Minute Timer	15-Minute Mode	12.0	13.5	15.0	min
Δt_{LF}	Low-Frequency Timer Accuracy	Charger Inactive	-23		27	%

Notes:

- 4. Guaranteed by design; not tested in production.
- 5. Negative current is current flowing from the battery to VBUS (discharging the battery).
- 6. Q2 always turns on for 60 ns, then turns off if current is below I_{SYNC}.
- 7. This tolerance (%) applies to all timers on the IC, including soft-start and deglitching timers.

I²C Timing Specifications

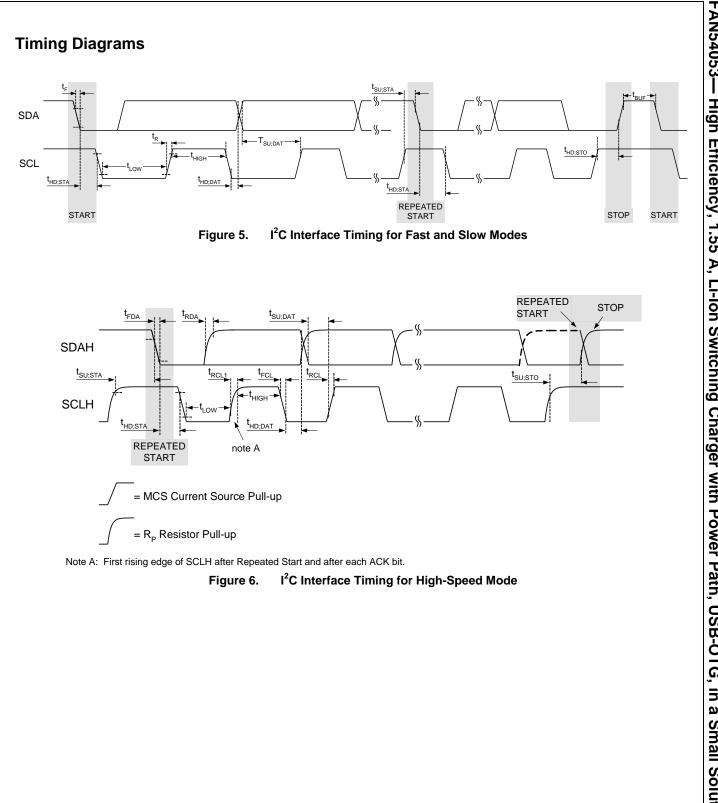
Guaranteed by design.

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit
		Standard Mode			100	
		Fast Mode			400	
f _{SCL}	SCL Clock Frequency	Fast Mode Plus			1000	kHz
		High-Speed Mode, $C_B \leq 100 \text{ pF}$			3400	
		High-Speed Mode, $C_B \leq 400 \text{ pF}$			1700	
		Standard Mode		4.7		
t _{BUF}	BUS-free Time between STOP and START Conditions	Fast Mode		1.3		μS
		Fast Mode Plus		0.5		
		Standard Mode		4		μs
	START or Repeated START Hold	Fast Mode		600		ns
t _{hd;sta}	Time	Fast Mode Plus		260		ns
		High-Speed Mode		160		ns
		Standard Mode		4.7		μS
		Fast Mode		1.3		μS
t _{LOW}	SCL LOW Period	Fast Mode Plus		0.5		μs
		High-Speed Mode, $C_B \leq 100 \text{ pF}$		160		ns
		High-Speed Mode, $C_B \leq 400 \text{ pF}$		320		ns
		Standard Mode		4		μS
t _{HIGH} SCL HI		Fast Mode		600		ns
	SCL HIGH Period	Fast Mode Plus		260		ns
		High-Speed Mode, $C_B \leq 100 \text{ pF}$		60		ns
		High-Speed Mode, $C_B \leq 400 \text{ pF}$		120		ns
		Standard Mode		4.7		μS
		Fast Mode		600		ns
t _{SU;STA}	Repeated START Setup Time	Fast Mode Plus		260		ns
		High-Speed Mode		160		ns
		Standard Mode		250		
		Fast Mode		100		
t _{SU;DAT}	Data Setup Time	Fast Mode Plus		50		ns
		High-Speed Mode		10		
		Standard Mode	0	1	3.45	μs
		Fast Mode	0		900	ns
t _{HD;DAT}	Data Hold Time	Fast Mode Plus	0	1	450	ns
		High-Speed Mode, C _B <u>≤</u> 100 pF	0	1	70	ns
		High-Speed Mode, $C_B \le 400 \text{ pF}$	0	1	150	ns
		Standard Mode	20+0	.1C _B	1000	
		Fast Mode	20+0		300	
t _{RCL}	SCL Rise Time	Fast Mode Plus	20+0	.1C _B	120	ns
		High-Speed Mode, $C_B \leq 100 \text{ pF}$		10	80	
		High-Speed Mode, $C_B \leq 400 \text{ pF}$		20	160	

I²C Timing Specifications (Continued)

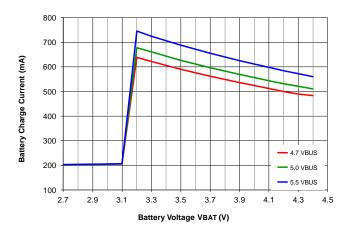
Guaranteed by design.

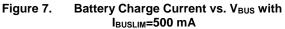
Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit	
		Standard Mode	20+0	1C _B	300		
		Fast Mode	20+0	.1C _B	300		
t _{FCL}	SCL Fall Time	Fast Mode Plus	20+0	.1C _B	120	ns	
		High-Speed Mode, C _B <u>≤</u> 100 pF		10	40		
		High-Speed Mode, $C_B \leq 400 \text{ pF}$		20	80		
4	Rise Time of SCL after a Repeated	High-Speed Mode, $C_B \leq 100 \text{ pF}$		10	80		
t _{RCL1}	START Condition and after ACK Bit	High-Speed Mode, C _B <u>≤</u> 400 pF		20	160	ns	
		Standard Mode	20+0	1C _B	1000		
		Fast Mode	20+0	.1C _B	300		
t _{RDA}	SDA Rise Time	Fast Mode Plus	20+0	.1C _B	120	ns	
		High-Speed Mode, C _B <u><</u> 100 pF		10	80		
		High-Speed Mode, $C_B \leq 400 \text{ pF}$		20	160		
		Standard Mode	20+0	1C _B	300		
		Fast Mode	20+0	.1C _в	300		
t _{FDA}	SDA Fall Time	Fast Mode Plus	20+0	.1C _B	120	ns	
		High-Speed Mode, C _B <u><</u> 100 pF		10	80		
		High-Speed Mode, C _B <u><</u> 400 pF		20	160		
		Standard Mode		4		μS	
	Stan Canditian Satur Time	Fast Mode		600		ns	
t _{su;sто}	Stop Condition Setup Time	Fast Mode Plus		120		ns	
		High-Speed Mode		160		ns	
C _B	Capacitive Load for SDA and SCL				400	pF	

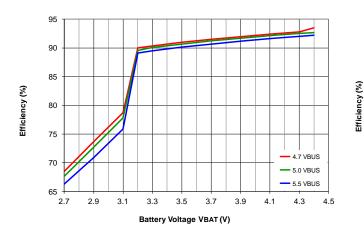


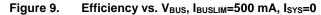
Charge Mode Typical Characteristics

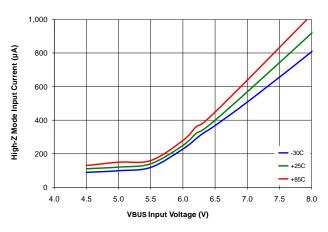
Unless otherwise specified, circuit of Figure 1, V_{OREG}=4.35 V, I_{OCHARGE}=950 mA, V_{BUS}=5.0 V, and T_A=25°C.

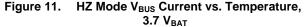


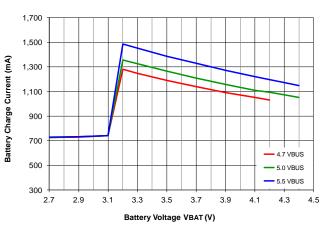


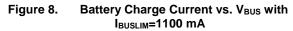












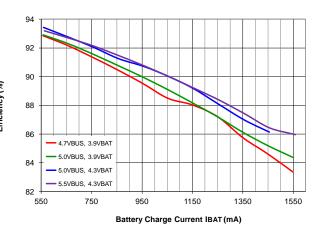


Figure 10. Efficiency vs. Charging Current, IBUSLIM=No Limit

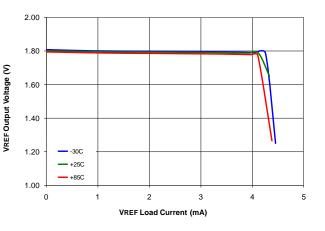
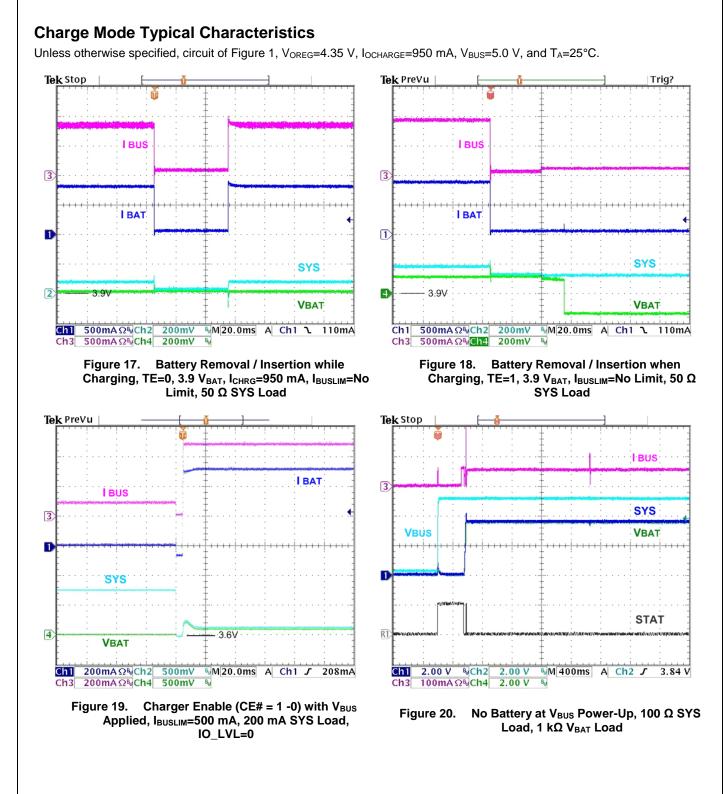


Figure 12. V_{REF} vs. Load Current, Over-Temperature



Charge Mode Typical Characteristics Unless otherwise specified, circuit of Figure 1, VOREG=4.35 V, IOCHARGE=950 mA, VBUS=5.0 V, and TA=25°C. Tek PreVu **Tek** PreVu I BUS I BUS 3 BAT BAT 1 SYS VBUS VBUS SYS VBAT VBAT 2 3:6V 2 3.1V R1 RT 200mV 200mA Ω%Ch2 200mV M 100ms A Ch1 J 104mA M 100ms A Ch1 J 100mA 500mA Ω%Ch2 Ch1 Ch1 Ch3 200mA Ω% Ch4 200mV % Ch3 500mA Ω% Ch4 200mV B_b Figure 13. Charger Startup at V_{BUS} Plug-In, 500 mA Figure 14. Charger Startup at V_{BUS} Plug-In, 1100 mA I_{BUSLIM} , 3.1 V_{BAT} , 50 Ω SYS Load, CE# = 0, I_{BUSLIM} , 3.6 V_{BAT} , 700 mA SYS Load, CE# = 0, IO_LVL=1 IO_LVL=0 **Tek** Pre∨u Tek Stop VBUS I BUS IBUS 3 I BAT 3 BAT 1 1 SYS SYS VBAT VBAT 4 3:7V 3:1V -2 Ch1 200mA Ω[®]Ch2 Ch1 500mA Ω% Ch2 100mV Ch3 500mA Ω% Ch4 100mV 200mV M 100ms A Ch1 J 72.0mA 100mV M40.0ms A Ch1 J 70.0mA Ch3 200mA Ω% Ch4 200mV Figure 15. Charger Startup at VBUS Plug-In Using

300 mA Current Limited Source, 500 mA I_{BUSLIM} , 3.1 V_{BAT}, 200 mA SYS Load, CE# = 0, IO_LVL=0 Figure 16. Charger Startup with HZ Bit Reset, 500 mA I_{BUSLIM}, 950 mA I_{CHARGE}, 50 Ω SYS Load, CE# = 0



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GSM Typical Characteristics

A 2.0 A GSM pulse applied at V_{BAT} with 5 μs rise / fall time. Simultaneous to GSM pulse, 50 Ω additional load applied at SYS.

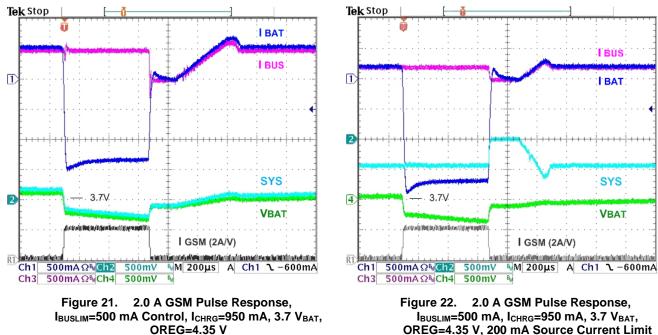


Figure 22. 2.0 A GSM Pulse Response, IBUSLIM=500 mA, ICHRG=950 mA, 3.7 VBAT, OREG=4.35 V, 200 mA Source Current Limit

I BUS

BAT

SYS

VBAT.

Boost Mode Typical Characteristics

Unless otherwise specified, using circuit of Figure 1, V_{BAT} =3.6 V, T_A =25°C.

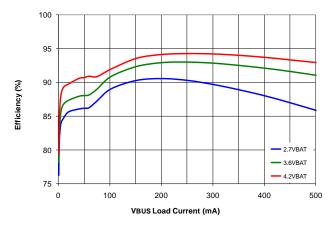


Figure 23. Efficiency vs. I_{BUS} Over V_{BAT}

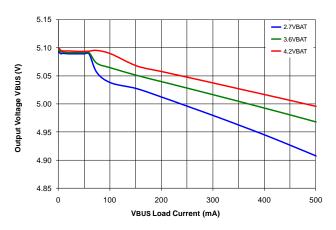


Figure 25. Regulation vs. I_{BUS} Over V_{BAT}

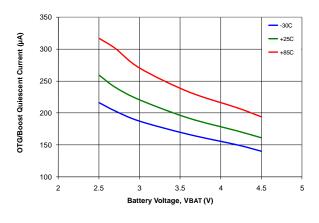


Figure 27. Quiescent Current (I_Q) vs. V_{BAT} Over-Temperature

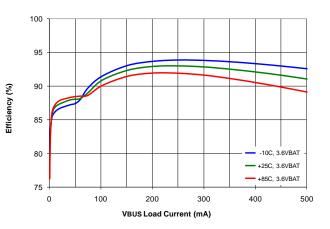


Figure 24. Efficiency vs. I_{BUS} Over-Temperature, 3.6 V_{BAT}

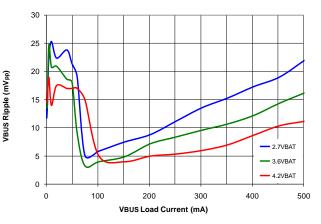
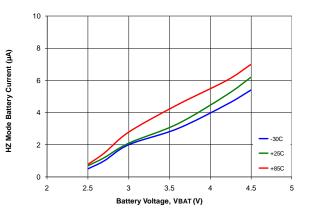
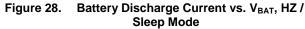
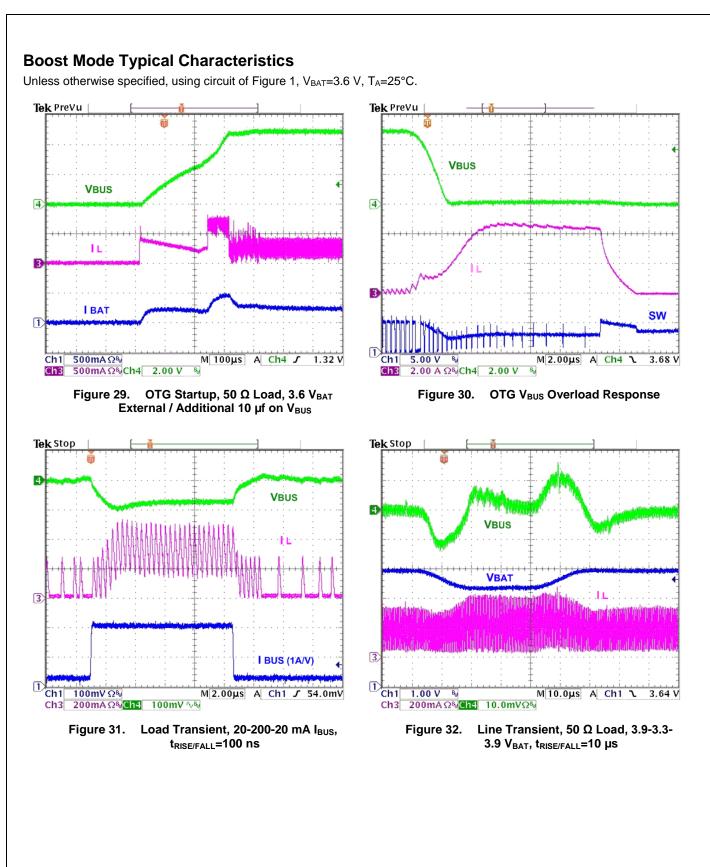


Figure 26. Output Ripple vs. I_{BUS} Over V_{BAT}





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Circuit Description / Overview

When charging batteries with a current-limited input source, such as USB, a switching charger's high efficiency over a wide range of output voltages minimizes charging time.

FAN54053 combines a highly integrated synchronous buck regulator for charging with a synchronous boost regulator, which can supply 5 V to USB On-The-Go (OTG) peripherals. The FAN54053 employs synchronous rectification for both the charger and boost regulators to maintain high efficiency over a wide range of battery voltages and charge states.

The FAN54053 has four operating modes:

- 1. Charge Mode: Charges a single-cell Li-ion or Li-polymer battery.
- Boost Mode: Provides 5 V power to USB-OTG with an integrated synchronous rectification boost regulator, using the battery as input.
- High-Impedance Mode: Both the boost and charging circuits are OFF in this mode. Current flow from VBUS to the battery or from the battery to VBUS is blocked in this mode. This mode consumes very little current from VBUS or the battery.
- Production Test Mode: This mode provides 4.35 V output on VBAT and supplies a load current of up to 2.3 A.

Charge Mode and Registers

Charge Mode

In Charge Mode, FAN54053 employs six regulation loops:

- Input Current: Limits the amount of current drawn from VBUS. This current is sensed internally and can be programmed through the I²C interface.
- Charging Current: Limits the maximum charging current. This current is sensed using an internal sense MOSFET.
- VBUS Voltage: This loop is designed to prevent the input supply from being dragged below V_{BUSLIM} (typically 4.5 V) when the input power source is current limited. An example of this would be a travel charger. This loop cuts back the current when V_{BUS} approaches V_{BUSLIM}, allowing the input source to run in current limit.
- 4. Charge Voltage: The regulator is restricted from exceeding this voltage. As the internal battery voltage rises, the battery's internal impedance works in conjunction with the charge voltage regulation to decrease the amount of current flowing to the battery. Battery charging is completed when the current through Q4 drops below the I_{TERM} threshold.
- 5. Pre-charge: When V_{BAT} is below V_{BATMIN} , Q4 operates as a linear current source and modulates its current to ensure that the voltage on SYS stays above 3.4 V.
- Temperature: If the IC's junction temperature reaches 120°C, charge current is reduced until the IC's temperature is below 120°C.

PWM Controller in Charge Mode

The IC uses a current-mode PWM controller to regulate the output voltage and battery charge currents. The synchronous rectifier (Q2) has a negative current limit that turns off Q2 at 140 mA to prevent current flow from the battery.

Battery Charging Curve

If the battery voltage is below V_{SHORT}, a linear current source pre-charges the battery until V_{BAT} reaches V_{SHORT}. The PWM charging circuit is then started and the battery is charged with a constant current if sufficient input power is available. The current slew rate is limited to prevent overshoot.

During the current regulation phase of charging, I_{BUSLIM} or the programmed charging current limits the amount of current available to charge the battery and power the system.

During the voltage regulation phase of charging, assuming that V_{OREG} is programmed to the cell's fully charged "float" voltage, the current that the battery accepts with the PWM regulator limiting its output (sensed at VBAT) to V_{OREG} declines.

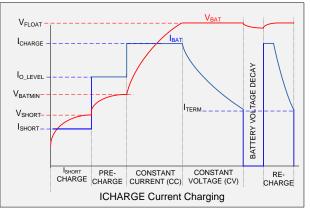


Figure 33. Charge Curve, I_{CHARGE} Not Limited by I_{BUSLIM}

The FAN54053 is designed to work with a current-limited input source at VBUS as shown below:

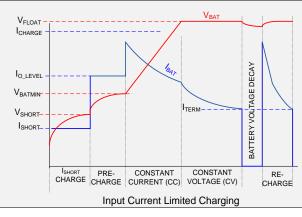


Figure 34. Charge Curve, IBUSLIM Limits ICHARGE

The following charging parameters can be programmed by the host through ${\rm I}^2{\rm C}$:

Parameter	Name	Register
Output Voltage Regulation	V _{OREG}	REG2[7:2]
Battery Charging Current Limit	IOCHARGE	REG4[6:3]
Input Current Limit	IBUSLIM	REG1[7:6]
Charge Termination Limit	I _{TERM}	REG4[2:0]
Weak Battery Voltage	VLOWV	REG1[5:4]
VBUS Control	V _{BUSLIM}	REG5[2:0]

Table 3. Programmable Charging Parameters

Output Voltage Regulation (VOREG)

The charger output or "float" voltage can be programmed by the OREG (REG02[7:2]) bits from 3.51 V to 4.45 V in 20 mV increments. The defauilt setting is 3.55 V.

See OREG Register Bit Definitions

Battery Charging Current Limit (IOCHARGE)

When the IO_LEVEL bit is set (default), the $I_{OCHARGE}$ bits are ignored and charge current is set to 200 mA.

See IOCHARGE Register Bit Definitions

Input Current Limiting (IBUSLIM)

To minimize charging time without overloading VBUS current limitations, the IC's input current limit can be programmed by the I_{BUSLIM} (REG1[7:6]) bits.

For the FAN54053, no charging occurs automatically at V_{BUS} POR, so the input current limit is established by the I_{BUSLIM} bits.

See IBUSLIM Register Bit Definitions

Termination Limit (I_{TERM})

Charge current termination can be enabled or disabled using the TE (REG01h[3]) bit. By default TE = "0", therefore, termination is disabled and charging does not terminate at the programmed I_{TERM} level.

When TE = "1", and V_{BAT} reaches V_{OREG}, the charging current is reduced, limited by the battery's ESR and its internal cell voltage. When the charge current falls below I_{TERM}; PWM charging stops; but the STAT pin remains LOW. The STAT pin then goes HIGH and the STAT bits change to CHARGE DONE (10), provided the battery and charger are still connected. If V_{BAT} falls to V_{RCH} below V_{OREG}, the Fast Charge cycle starts again.

Post-charging can be enabled to "top-off" the battery to a lower termination current threshold than I_{TERM} . The PC_EN bit (REG07h[3]) must be set to "1" before the battery charging current reaches I_{TERM} . The lower termination current is set by the PC_IT (REG07h[2:0] bits. Post-charging begins after normal charging is ended (as described above) with the PC_ON (REG11h[2]) monitor bit set to "1".

During post-charging, the STAT pin is HIGH, indicating that the charge current is below the I_{TERM} level. Once the current reaches the threshold for post-charging completion (PC_IT),

PWM charging stops and the PC_ON bit changes back to "0". If the charging current goes above I_{TERM} without first falling to PC_IT, the PC_ON bit can be reset by using any of these methods: V_{BAT} moving below and above V_{BATMIN} , a V_{BUS} POR, or the CE# or HZ_MODE bit cycled. If V_{BAT} falls to V_{RCH} below V_{OREG} , the Fast Charge cycle starts again.

See ITERM Register Bit Definitions

Weak Battery Voltage (V_{LOWV})

The FAN54053 monitors the level of the battery with respect to a programmable V_{LOWV} (REG01h<5:4>) threshold (default 3.7 V). The V_{LOWV} threshold defines the voltage level of the battery at which the system is guaranteed to be fully operational when only powered by the battery.

The POK_B pin pulls LOW once V_{BAT} reaches V_{LOWV} , and remains LOW as long as the IC is in Fast Charge. The IC will remain in Fast Charge as long as VBAT > 3.0 V.

See VLOWV Register Bit Definitions

VBUS Control loop (V_{BUSLIM})

The IC includes a control loop that limits input current in case a current-limited source is supplying $V_{\text{BUS}}.$

The control increases the charging current until either:

- I_{BUSLIM} or I_{OCHARGE} is reached OR
- V_{BUS} = V_{BUSLIM}.

If V_{BUS} collapses to V_{BUSLIM}, the VBUS loop reduces its current to keep V_{BUS} = V_{BUSLIM}. When the VBUS control loop is limiting the charge current, the VLIM bit (REG05h[3]) is set.

See VBUSLIM Register Bit Definitions

Charger Operation

VBUS Plug In and Safety Timer

At VBUS plug in, the TMR_RST (Reg00h[7]) bit must be set within 2 seconds of V_{BUS} rising above V_{(INMIN)1} or all registers, except for SAFETY (REG06h), are set to their default values. This functionality occurs regardless of the state of the CE# and WD_DIS bit. If plug in occurs with the device in a HZ or Charge Done state and the TMR_RST bit is not set within 2 seconds of V_{BUS} rising above V_{(INMIN)1}, all register, except for SAFETY, will reset when the device enters PWM Charging or Recharge.

By default, the safety timers do not run in the FAN54053. A Watchdog (t_{32s}) timer can be enabled by setting the WD_DIS register bit, (REG13h[1]) to "0". When WD_DIS = "0", charging is controlled by the host with the t_{32S} timer running to ensure that the host is alive. Setting the TMR_RST bit resets the t_{32S} timer. If the t_{32S} timer times out; all registers, except SAFETY, are set to their default values (including WD_DIS and CE#), the FAULT bits are set to "110", and STAT is pulsed.

V_{BUS} POR / Non-Compliant Charger Rejection

256 ms after VBUS is connected, the IC pulses the STAT pin and sets the VBUS_CON bit. Before starting to supply current, the IC applies a 110 Ω load from VBUS to GND. V_{BUS} must remain above $V_{\text{IN}(\text{MIN})1}$ and below VBUS_OVP for $t_{\text{VBUS}_\text{VALID}}$ (32 ms) before the IC initiates charging or supplies power to SYS.

The VBUS validation sequence always occurs before significant current is available to be drawn from VBUS (for example, after a VBUS OVP fault or a V_{RCH} (recharge initiation). t_{VBUS_VALID} ensures that unfiltered 50/60 Hz chargers and other non-compliant chargers are rejected.

USB-Friendly Boot Sequence

The FAN54053 does not automatically initiate charging at V_{BUS} POR. Instead, prior to receiving host commands, the buck is enabled to provide power to SYS while Q4 and Q5 remain off until register bit CE# (REG01[2]) is set to "0" through the I²C interface, allowing charging through Q4.

Startup with No Battery

The FAN54053 has Battery Absent Behavior enabled. At V_{BUS} POR with the battery absent, the PWM will run, providing 3.55 V to the system from the input source with current limited by the default I_{BUSLIM} setting.

Startup with a Dead Battery

At V_{BUS} POR, if V_{BAT} < V_{SHORT}, all registers, including the SAFETY register, are reset to their default values and the DBAT_B (REG02h[1]) bit is reset. CE# = "1", so charging is disabled.

If the battery's protection switch is open, the PWM will run, providing 3.55 V to the system from the input source with current limited by the default I_{BUSLIM} setting. This allows the host processor to awaken and establish host control. Once this occurs, the host's low level software can program the CE# bit to "0" and a linear current source closes the battery protection switch. When V_{BAT} voltage rises above V_{SHORT} and sufficient power is available, PWM charging begins and the battery is charged through the BATFET, Q4. The IO_LEVEL (REG05h[5]) bit is set to "1" by default which limits charge current to 200 mA.

With CE# = "1" once V_{BAT} rises above V_{SHORT}, DBAT_B is set. With CE# = "0" once V_{BAT} rises above V_{BATMIN}, DBAT_B is set.

Power Path Operation

As long as V_{BAT} < V_{BATMIN}, Q4 operates as a linear current source, (Precharge) with its current (I_{PP}) limited to 200 mA when IO_LEVEL (REG05h[5]) is set to its default value of "1". If IO_LEVEL is set to "0" and I_{INLIM} > "01", charge current is limited to 450 mA when I_{OCHARGE} \leq 750 mA, and 730 mA when I_{OCHARGE} > 750 mA. Providing the input current is not limited by the I_{BUSLIM} setting or the current available from the source, during precharge, the IC regulates SYS to 3.55 V and provides the IO_LEVEL limited current to the battery.

System power always has the highest priority when power from the buck is limited ensuring SYS does not fall below 3.4 V. This is managed by folding back the current to charge the battery until charge current is reduced to 0 A.

After V_{BAT} reaches $V_{BATMIN},\ Q4$ closes and is used as a current-sense element to limit current (I_{OCHRG}) per the I^2C register settings. This is accomplished by limiting the PWM modulator's current (Fast Charge). If SYS drops more than 5 mV (V_{THSYS}) below V_{BAT} and CE# = "0", Q4 and Q5 are turned on (GATE is pulled LOW). If CE# = "1", only Q5 is turned on. Once SYS voltage becomes higher than $V_{BAT},\ Q5$ is turned off and Q4 again serves as the current-sense element to limit I_{OCHRG} .

If CE# = "1" and DIS pin is high or CE# = "1" and HZ_MODE = "1" while $V_{BAT} > V_{LOWV}$, so as to prevent the system from crashing, Q4 and Q5 are enabled. Q4 and Q5 are also both turned on when the IC enters SLEEP Mode ($V_{BUS} < V_{BAT}$).

POK_B (see Table 4)

The POK_B pin and POK_B (REG11h[5]) bit are intended to provide feedback to the baseband processor that the battery is strong enough to allow the device to fully function. Whenever the IC is operating in precharge, POK_B is HIGH. On exiting Precharge, POK_B remains HIGH until V_{BAT} > V_{LOWV}. REG01h[5:4] sets the V_{LOWV} threshold.POK_B pulls LOW once V_{BAT} reaches V_{LOWV}, and remains LOW as long as the IC is in Fast Charge and the IC will remain in Fast Charge as long as V_{BAT} > 3.0 V. If the battery voltage falls below 3.0 V the IC enters Precharge. If WD_DIS = "0" and the T_{32S} expires during charging, the POK_B pin will go high.

If the battery was above V_{LOWV} and has fallen below the level, the POK_B bit can be set to change the state of the pin to be high. This setting of the bit and pin can be used to signal the system into a low-power state, preventing excessive loading from the system while attempting to recharge a depleted battery.

The STAT pin pulses any time the POK_B pin changes.

Operating Mode	V _{BUS}	V _{BAT}	CE#	PWM	V _{SYS}	Q4	Q5	GATE	POK_B
BUS Disconnected									
OFF	< V _{BAT OR} < V _{IN(MIN)2}	> V _{SHORT}	х	OFF	<u>≺</u> V _{BAT}	ON	ON	LOW	HIGH
	VBUS	8 Plug in with B	attery	Protecti	ion Swi	tch Ope	n		
DWM			1			055	055		HIGH
PWM	Valid	OPEN	0	ON	Voreg	OFF	OFF	HIGH	Indeterminate ⁽⁸⁾
30 mA Linear Charging ⁽⁹⁾	Valid	< V _{SHORT}	0	ON	3.55	OFF	OFF	HIGH	HIGH
Charge Mode									
Precharge	Valid	> V _{SHORT} and < V _{BATMIN}	0	ON	3.55	Linear	OFF	HIGH	HIGH
Precharge: I _{SYS} + I _{pp} > I _{PWM} , I _{BAT} < I _{PP}	Valid	< V _{BATMIN}	0	ON	< 3.55	Linear	OFF	HIGH	HIGH
Fast Charge	Valid	> V _{BATMIN} and < V _{LOWV}	0	ON	> V _{BAT}	ON	OFF	HIGH	HIGH
		> V _{LOWV}							LOW
		Battery Voltage	Falling	from Fa	ast Char	ge			1
Precharge	Valid	VBATFALL	0	ON	3.55	ON	OFF	HIGH	HIGH
		Battery S	Supplen	nenting	SYS				
Supplemental Mode :		> V _{SHORT}	1	ON	$< V_{BAT}$	Х	ON	LOW	Х
I _{SYS} > I _{PWM}	Valid	> V_{BATMIN} and > V_{SYS} + V_{THSYS}	0	ON	< V _{BAT}	х	ON	LOW	х

Note:

8. When V_{BAT} is open it can float to V_{SYS} , and POK_B = HIGH when $V_{BAT} < V_{LOWV}$ and POK_B = LOW when $V_{BAT} > V_{LOWV}$.

9. 30 mA Linear Charging operating mode assumes the host has programmed CE# = "0" during PWM Operating Mode.

Charger Status / Fault Status

The STAT pin indicates the operating condition of the IC and provides a fault indicator for interrupt driven systems.

Table 5. STAT Pin Function

EN_STAT	Charge State	STAT Pin
0	Х	OPEN
Х	Normal Conditions	OPEN
1	Charging	LOW
Х	Fault (Charging or Boost)	128 μs Pulse, then OPEN

The FAULT bits (REG00h[2:0]) indicate the type of fault in Charge Mode.

Monitor Registers (REG10h, REG11h)

Additional status monitoring bits enable the host processor to have more visibility into the status of the IC. The monitor bits are real-time status indicators and are not internally debounced or otherwise time qualified. The state of the MONITOR register bits listed in High-Impedance Mode is valid only when V_{BUS} is valid.

Charge Mode Control Bits

The CE# (REG01h[2]) bit is set to "1' by default, therefore, charging is disabled.

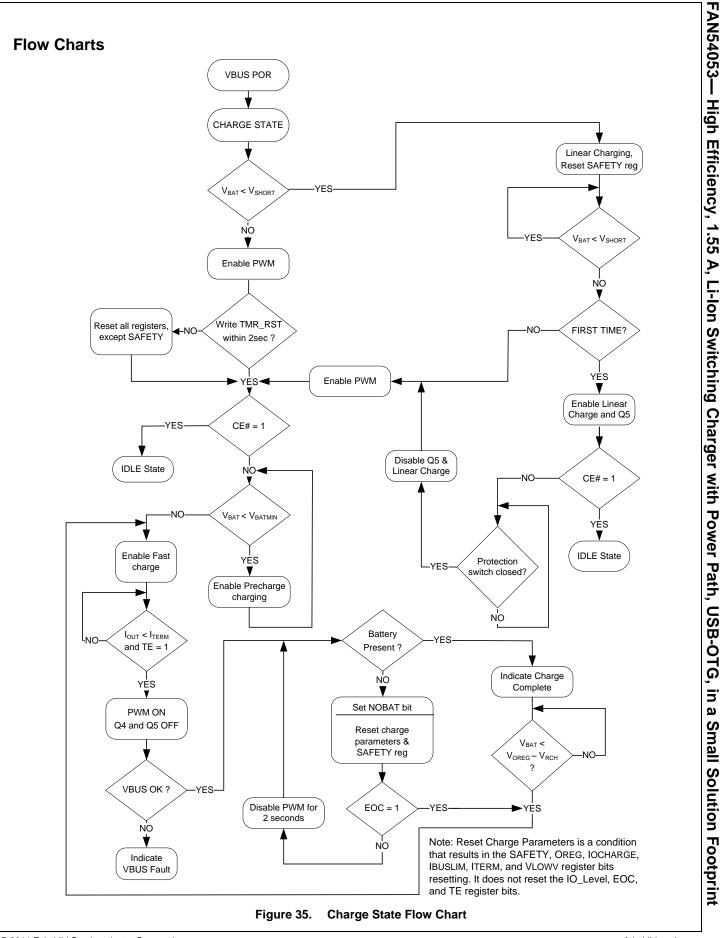
Setting the RESET (REG04h[7]) bit clears all registers (except SAFETY). The CE# bit will only be cleared if RESET occurs with a valid VBUS and $V_{BAT} < V_{LOWV}$. If HZ_MODE or the WD_DIS bit was set when the RESET bit is set, this bit is also cleared. Refer to the Register Bit Definitions section for more details.

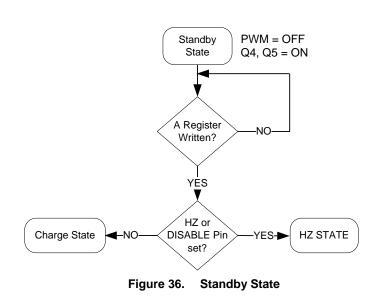
The HZ_MODE (REG01h[1]) and DIS pin will put the device in High-Impedance Mode. If HZ_MODE = "1" or DIS pin is HIGH, so as to prevent the system from crashing, Q4 and Q5 are enabled.

The functionality of the HZ_MODE (REG01h[1]) bit and DIS pin has a dependence upon V_{BAT} voltage level and the WD_DIS (REG13h[1]) bit state. Refer to Table 5 for details.

Table 6.	DIS Pin, HZ	_MODE and WD_	DIS bits Operation
----------	-------------	---------------	--------------------

Conditions	Functionality
	Setting either the HZ_MODE bit through I ² C or the DIS pin to HIGH will disable the charger and put the IC into High-Impedance Mode.
WD DIS = 1 and $V_{BAT} > V_{LOWV}$	Resetting the HZ_MODE bit or the DIS pin to LOW will allow charging to resume.
	While in High-Impedance mode, if V_{BAT} drops below V_{LOWV} , all registers (except SAFETY), including HZ_MODE and CE# are reset. Note that charge parameters will need to be reprogrammed in order to completely charge the battery.
WD_DIS = 1 and $V_{BAT} < V_{LOWV}$	Setting either the HZ_MODE bit through I^2C or the DIS pin to HIGH will reset all registers (except SAFETY), including HZ_MODE and CE#.
WD_DIS = 0 and $V_{BAT} > V_{LOWV}$	Setting either the HZ_MODE bit through I^2C or the DIS pin to HIGH will stop the t_{32S} timer from advancing (does not reset it), disable the charger, and put the IC into High-Impedance Mode.
	Resetting the HZ_MODE bit or the DIS pin to LOW will allow charging to resume.
WD_DIS = 0 and $V_{BAT} < V_{LOWV}$	Setting either the HZ_MODE bit through I^2C or the DIS pin to HIGH will disable the charger and put the IC into High-Impedance Mode. The T_{32S} timer will continue to run. If the T_{32S} timer is allowed to overflow, all registers (except SAFETY) are reset, including WD_DIS, HZ_MODE and CE#.





Non-Charging States

Sleep Mode

When V_{BUS} falls below V_{BAT} + V_{SLP} and V_{BUS} is above $V_{\text{IN(MIN)2}}$, the IC enters Sleep Mode to prevent the battery from draining into VBUS. During Sleep Mode, reverse current is disabled by body switching Q1.

Idle State

The Idle State is related to the condition of the battery. During Idle mode the Switch Mode Power Supply (SMPS) is operating, but the battery is not being charged for one or more of the following conditions exists: the Safety Timer expires, charging is complete, or the BATFET is disabled by the Charge Enable bit, CE# = "1".

The PWM Buck continues to supply power to the system, but the Battery is no longer being charged and the BATFET is disabled.

Standby State

The Standby State is an intermediate state where the switch mode supply is off due to either bad input power, the device has been put in High-Impedance Mode, or the die temperature is too hot.

Charger Protection

Battery Temperature (NTC) Monitor

The FAN54053 reduces the maximum charge current and termination voltage if an NTC measuring battery temperature (T_{BAT}) indicates that it is outside the fast-charging limits (T2 to T3), as described in the JEITA specification¹. There are four temperature thresholds that change battery charger operation: T1, T2, T3, and T4, shown in Table 7.

Table 7. Battery Temperature Thresholds

For use with 10 k Ω NTC, β = 3380, and R_{REF} = 10 k Ω .

Threshold	Temperature	% of V _{REF}
T1	0°C	73.9
T2	10°C	64.6
Т3	45°C	32.9
T4	60°C	23.3

Table 8. Charge Parameters vs. T_{BAT}

Т _{ват} (°С)	I _{CHARGE}	V _{FLOAT}
Below T1	Charging to VB	AT Disabled
Between T1 and T2	I _{OCHARGE} / 2 ⁽¹⁰⁾	4.0 V
Between T2 and T3	I _{OCHARGE}	V _{OREG}
Between T3 and T4	I _{OCHARGE} / 2 ⁽¹⁰⁾	4.0 V
Above T4	Charging to VBAT Disabled	

Note:

10. If I_{OCHARGE} is programmed to less than 650 mA, the charge current is limited to 340 mA.

Thermistors with other β values can be used, with some shift in the corresponding temperature threshold, as shown in Table 9.

Table 9.	Thermistor Temperature Thresholds
$R_{REF} = R_{TH}$	_{RM} at 25°C.

Parameter	Various Thermistors			
R _{THRM(25°C)}	10 kΩ	10 kΩ	47 kΩ	100 kΩ
β	3380	3940	4050	4250
T1	0°C	3°C	6	8
T2	10°C	12°C	13	14
Т3	45°C	42°C	41	40
T4	60°C	55°C	53	51

The host processor can disable temperature-driven control of charging parameters by writing "1" to the TEMP_DIS bit. Since TEMP_DIS is reset whenever the IC resets its registers, the temperature controls are enforced whenever the IC is auto-charging, since auto-charge is always preceeded by a reset of registers.

To disable the thermistor circuit, tie the NTC pin to GND. Before enabling the charger, the IC tests to see if NTC is shorted to GND. If NTC is shorted to GND, no thermistor readings occur and the NTC_OK and NTC1-NTC4 is reset.

The IC first measures the NTC immediately prior to entering any PWM charging state, then measures the NTC once per second, updating the result in NTC1-NTC4 bits (REG 12h[3:0]).

Table 10. NTC1-NTC4 Decoding

		-		
Т _{ват} (°С)	NTC4	NTC3	NTC2	NTC1
Above T4	1	1	1	1
Between T3 and T4	0	1	1	1
Between T2 and T3	0	0	1	1
Between T1 and T2	0	0	0	1
Below T1	0	0	0	0

Safety Register Settings

The IC contains a SAFETY register (REG06h) that prevents the values in OREG (REG02h[7:2]) and IOCHARGE (REG04h[7:4]) from exceeding the values of VSAFE (REG06h[3:0]) and ISAFE (REG06h[7:4]) in the SAFETY register.

After V_{BAT} rises above V_{SHORT} , the SAFETY register is loaded with its default value and may be written to only before writing to any other register. The same 8-bit value should be written to the SAFETY register twice to set the register value. After writing to any other register, the SAFETY register is locked until V_{BAT} falls below V_{SHORT} .

If the host attempts to write a value higher than VSAFE or ISAFE to OREG or I_{OCHARGE}, respectively; the VSAFE, ISAFE value appears as the OREG, I_{OCHARGE} register value, respectively.

¹ Japan Electronics and Information Technology Industries Association (JEITA) and Battery Association of Japan. "A Guide to the Safe Use of Secondary Lithium Ion Batteries in Notebook-type Personal Computers," April 28, 2007.

The Safety register is reset when the battery is below 3.0 V and power is removed from the VBUS if CE# = "1". The Safety register can be re-written to if it is the first I2C write anytime after the VBUS removal.

See VSAFE and ISAFE Register Bit Definitions

Thermal Regulation and Shutdown

When the IC's junction temperature reaches TCF (about 120°C), the charger reduces its output current to 550 mA to prevent overheating. If the temperature increases beyond $T_{SHUTDOWN}$; charging is suspended, the FAULT bits are set to 101, and STAT is pulsed high. In Suspend Mode, all timers stop and the state of the IC's logic is preserved. Charging resumes at programmed current after the die cools to about 120°C.

Note that as power dissipation increases, the effective θ_{JA} decreases due to the larger difference between the die temperature and ambient.

Charge Mode Input Supply Protection

Input Supply Low-Voltage Detection

The IC continuously monitors V_{BUS} during charging. If V_{BUS} falls below $V_{\text{IN(MIN)2}},$ the IC:

- 1. Terminates charging
- 2. Pulses the STAT pin, sets the STAT bits to "11", and sets the FAULT bits to "011".

If V_{BUS} recovers above the $V_{\text{IN}(\text{MIN})1}$ rising threshold after time t_{INT} (about two seconds), the charging process is repeated. This function prevents the USB power bus from collapsing or oscillating when the IC is connected to a suspended USB port or a low-current-capable OTG device.

Input Over-Voltage Detection

When the V_{BUS} exceeds VBUS_{OVP}, the IC:

- 1. Turns off Q3
- 2. Suspends charging
- 3. Sets the FAULT bits to "001", sets the STAT bits to "11", and pulses the STAT pin.

When VBUS falls about 100 mV below VBUSOVP, the fault is cleared and charging resumes after VBUS is revalidated.

SYS Short During Discharge / Supplemental Mode

Caution should be taken to ensure the SYS pin is not shorted when connected to a battery. This condition can induce high current flow through the BATFET (Q4) and the external FET (Q5) until the battery's own safety circuit trips. The resulting high current can damage the IC.

VBUS Short While Charging

If VBUS is shorted with a very low impedance while the IC is charging with $I_{\text{BUSLIM}} = 100$ mA, the IC may not meet datasheet specifications until power is removed. To trigger this condition, V_{BUS} must be driven from 5 V to GND with a high slew rate. Achieving this slew rate requires a 0 Ω short to the USB cable less than 10 cm from the connector.

Charge Mode Battery Detection & Protection

V_{BAT} Over-Voltage Protection

The OREG voltage regulation loop prevents V_{BAT} from overshooting V_{OREG} by more than 50 mV when the battery is removed. When the PWM charger runs with no battery, the TE bit is not set and a battery is inserted that is charged to a voltage higher than V_{OREG} ; PWM pulses stop. If no further pulses occur for 30 ms, the IC sets the FAULT bits to "100", sets the STAT bits to "11", and pulses the STAT pin

Battery Detection during Charging

The IC can detect the presence, absence, or removal of a battery if the termination bit (TE) is set to "1" and CE# = "0". During normal charging, once V_{BAT} is close to V_{OREG} and the charge current falls below I_{TERM}; the PWM charger continues to provide power to SYS and Q4 is turned off. It then turns on a discharge current, I_{DETECT}, for t_{DETECT}. If V_{BAT} is still above $V_{OREG} - V_{RCH}$, the battery is present and the IC sets the STAT bits to "10" (Charge Done). If V_{BAT} is below V_{OREG} - V_{RCH}, the battery is absent and the IC:

- 1. Sets the charging parameters to their default values.
- 2. Sets the FAULT bits to "111" (Battery Absent) and sets the NOBAT bit.
- 3. If EOC = "0", the IC turns off the PWM for t_{INT} , then resumes charging and retries Battery Detection. If the battery is still absent, the process repeats with the "No Battery" fault re-enunciated.
- If EOC = "1", the PWM remains on to provide power to SYS, but charge termination and the battery absent test are performed every t_{INT}.

Linear Charging

If the battery voltage is below the short-circuit threshold (V_{SHORT}); a linear current source, I_{SHORT} , charges V_{BAT} until $V_{BAT} > V_{SHORT}$.

Production Test Mode (PTM)

PTM provides 4.20 V at up to 2.3 A to VBAT when V_{BUS} = 5.5 V ±5%.

The IC enters PTM when the PROD (REG05h[6]) bit is set after the NOBAT (REG11h[3]) bit has been set. The NOBAT bit indicates that the IC has detected battery absence. A battery absence detection test is performed automatically at current termination. The steps for entering PTM should include: set the TE (REG01h[3]) bit high, set the CE# (REG01h[2]) bit low, wait for the NOBAT bit to set HIGH, then set the PROD bit to "1" to enter PTM. Battery absence detection is completed within 500 ms from the time that CE# is set.

In PTM, the GATE bit (REG11h[7]) is LOW, Q5 is on, and all auxiliary control loops are disabled. Only the OREG loop is active, which controls V_{BAT} to 4.20 V, regardless of the OREG register setting. Thermal shutdown remains active.

During PTM, high current pulses (load currents greater than 1.5 A) must be limited to 20% duty cycle with a minimum period of 10 ms.

Boost Mode

Boost Mode can be enabled if the IC is in 32-Second Mode by setting the OPA_MODE REG01h[0]) bit HIGH and clearing the HZ MODE bit.

Table 11. Enabling Boost	Table	11.	Enabling	Boost
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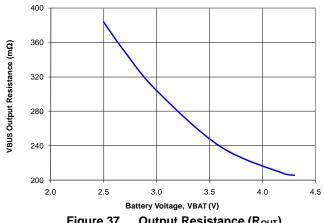
HZ_MODE	OPA_MODE	BOOST
0	1	Enabled
1	Х	Disabled
Х	0	Disabled

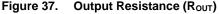
To remain in Boost Mode, the TMR RST must be set by the host before the t325 timer times out. If t325 times out in Boost Mode; the IC resets all registers, pulses the STAT pin, sets the FAULT bits to 110, and resets the BOOST bit. VBUS POR or reading R0 clears the fault condition.

Boost PWM Control

The IC uses a minimum on-time and computed minimum offtime to regulate V_{BUS}. The regulator achieves excellent transient response by employing current-mode modulation. This technique causes the regulator to exhibit a load line. During fast charging, the output voltage drops slightly as the input current rises. With a constant V_{BAT}, this appears as a constant output resistance.

The "droop" caused by the output resistance when a load is applied allows the regulator to respond smoothly to load transients with no undershoot from the load line. This can be seen in Figure 31 and Figure 37.





 V_{BUS} as a function of I_{LOAD} can be computed when the regulator is in PWM Mode (continuous conduction) as:

$$V_{OUT} = 5.07 - R_{OUT} \bullet I_{LOAD}$$
 EQ. 1

At V_{BAT} =3.0 V and I_{LOAD} =300 mA, V_{BUS} drops to:

$$V_{OUT} = 5.07 - 0.30 \bullet 0.3 = 4.98V$$
 FO 2

At V_{BAT} =3.6 V and I_{LOAD} =500 mA, V_{BUS} drops to:

$$V_{OUT} = 5.07 - 0.24 \bullet 0.5 = 4.95V$$
 FQ. 3

PFM Mode

If $V_{BUS} > VREF_{BOOST}$ (nominally 5.07 V) when the minimum off-time ends, the regulator enters PFM Mode. Boost pulses are inhibited until V_{BUS} < VREF_{BOOST}. The minimum on-time is increased to enable the output to pump up sufficiently with each PFM boost pulse. Therefore, the regulator behaves like a constant on-time regulator, with the bottom of its output voltage ripple at 5.07 V in PFM Mode.

Table 12.	Boost PWM	Operating	States
			U UU

Mode	Description	Invoked When
LIN	Linear Startup	$V_{BAT} > V_{BUS}$
SS	Boost Soft-Start	$V_{BUS} < V_{BST}$
BST	Boost Operating Mode	V _{BAT} > UVLO _{BST} and SS Completed

Startup

When the boost regulator is shut down, current flow is prevented from V_{BAT} to V_{BUS}, as well as reverse flow from VBUS to VBAT.

LIN State

When EN rises, if V_{BAT} > UVLO_{BST}; the regulator first attempts to bring PMID within 400 mV of V_{BAT} using an internal 450 mA current limited source from V_{BAT} (LIN State). If PMID has not achieved V_{BAT} – 400 mV after 560 µs, a fault state is initiated.

SS State

When PMID > V_{BAT} – 400 mV, the boost regulator begins switching with a reduced peak current limit of about 50% of its normal current limit. The output slews up until V_{BUS} is within 5% of its setpoint; at which time, the regulation loop is closed and the current limit is set to 100%.

If the output fails to achieve 95% of its setpoint (V_{BST}) within 128 µs, the current limit is increased to 100%. If the output fails to achieve 95% of its setpoint after this second 384 μ s period, a fault state is initiated.

BST State

This is the normal operating mode of the regulator. The regulator uses a minimum toFF-minimum toN modulation <u>VIN</u>, which scheme. The minimum t_{OFF} is proportional to VOUT

keeps the regulator's switching frequency reasonably constant in CCM. $t_{ON(MIN)}$ is proportional to V_{BAT} and is a higher value if the inductor current reached 0 before tOFE(MIN) in the prior cycle.

To ensure V_{BUS} does not overshoot the regulation point, the boost switch remains off as long as $V_{FB} > V_{REF(BST)}$.

Boost Faults

If a BOOST fault occurs:

- 1. The STAT pin pulses.
- OPA_MODE bit is reset. 2.
- 3. The power stage is in High-Impedance Mode.
- 4. The FAULT bits (REG0[2:0]) are set per Table 13.

Restart After Boost Faults

OPA_MODE is reset on boost faults. Boost Mode can only be re-enabled by setting the OPA_MODE bit.

Table 13. Fault Bits During Boost Mode

Fa	ult l	Bit	Foult (REC00b[2:0]) Description
B2	B1	B0	Fault (REG00h[2:0]) Description
0	0	0	Normal (no fault)
0	0	1	$V_{BUS} > VBUS_{OVP}$
0	1	0	V_{BUS} fails to achieve the voltage required to advance to the next state during soft-start or sustained (>50 µs) current limit during the BST state.
0	1	1	V _{BAT} < UVLO _{BST}
1	0	0	NA: This code does not appear.
1	0	1	Thermal shutdown
1	1	0	Timer fault; all registers reset.
1	1	1	NA: This code does not appear.

I²C Interface

The FAN54053's serial interface is compatible with Standard, Fast, Fast Plus, and High-Speed Mode I^2C bus specifications. The FAN54053 SCL line is an input and the SDA line is a bi-directional open-drain output; it can only pull down the bus when active. The SDA line only pulls LOW during data reads and when signaling ACK. All data is shifted in MSB (bit 7) first.

Slave Address

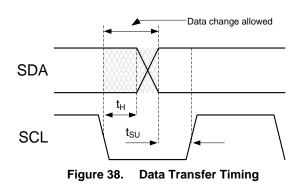
Table 14. I²C Slave Address Byte

7	6	5	4	3	2	1	0
1	1	0	1	0	1	1	R/W

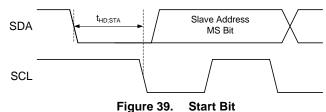
In hex notation, the slave address assumes a 0 LSB. The hex slave address is D6h for all parts in the family. Other slave addresses can be accommodated upon request. *Contact a Fairchild Semiconductor representative.*

Bus Timing

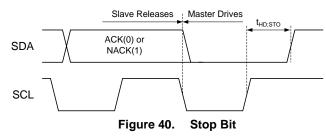
Shown in Figure 38, data is normally transferred when SCL is LOW. Data is clocked in on the rising edge of SCL. Typically, data transitions shortly at or after the falling edge of SCL to allow ample time for the data to set up before the next SCL rising edge.



Each bus transaction begins and ends with SDA and SCL HIGH. A transaction begins with a START condition, which is defined as SDA transitioning from 1 to 0 with SCL HIGH, as shown in Figure 39.



Transactions end with a STOP condition, which is SDA transitioning from 0 to 1 with SCL HIGH, as shown in Figure 40.



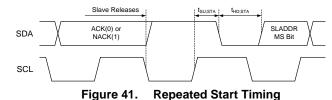
During a read from the FAN54053 Figure 43 the master issues a Repeated Start after sending the register address and before resending the slave address. The Repeated Start is a 1-to-0 transition on SDA while SCL is HIGH, as shown in Figure 41.

High-Speed (HS) Mode

The protocols for High-Speed (HS), Low-Speed (LS), and Fast-Speed (FS) Modes are identical except the bus speed for HS Mode is 3.4 MHz. HS Mode is entered when the bus master sends the HS master code 00001XXX after a start condition. The master code is sent in Fast or Fast Plus Mode (less than 1 MHz clock); slaves do not ACK the transmission.

The master then generates a repeated start condition that causes all slaves on the bus to switch to HS Mode. The master then sends I²C packets, as described above, using the HS Mode clock rate and timing.

The bus remains in HS Mode until a stop bit is sent by the master. While in HS Mode, packets are separated by repeated start conditions Figure 41.



Read and Write Transactions

The figures below outline the sequences for data read and write. Bus control is signified by the shading of the packet,

and

Slave Drives Bus

defined as All addresses and data are MSB first.

Master Drives Bus

Table 15. Bit Definitions for Figure 42- Figure 45

Symbol	Definition
S	START, see Figure 39
А	ACK. The slave drives SDA to 0 to acknowledge the preceding packet.
Ā	NACK. The slave sends a 1 to NACK the preceding packet.
R	Repeated START, see Figure 41
Р	STOP, see Figure 40

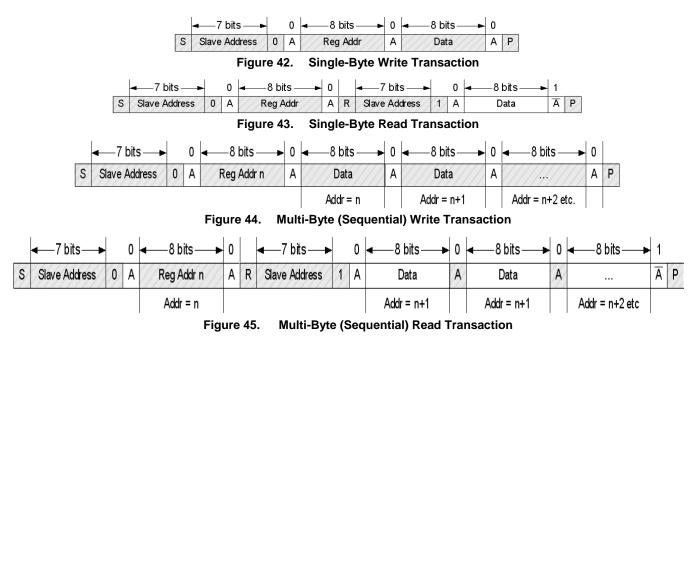
Multi-Byte (Sequential) Read and Write Transactions

Sequential Write

The Slave Address, Reg Addr address, and the first data byte are transmitted to the FAN54053 in the same way as in a byte write Figure 42. However, instead of generating a Stop condition, the master transmits additional bytes that are written to consecutive sequential registers after the falling edge of the eighth bit. After the last byte written and its ACK bit received, the master issues a STOP bit. The IC contains an 8-bit counter that increments the address pointer after each byte is written.

Sequential Read

Sequential reads are initiated in the same way as a singlebyte read , except that once the slave transmits the first data byte, the master issues an acknowledge instead of a STOP condition. This directs the slave's I^2C logic to transmit the next sequentially addressed 8-bit word. The FAN54053 contains an 8-bit counter that increments the address pointer after each byte is read, which allows the entire memory contents to be read during one I^2C transaction.



Register Descriptions

The Twelve user-accessible IC registers are defined in Table 17.

Table 16. I²C Register Map

Registe	r				BIT NA	ME						
Name	REG#	7	6	5	4	3	2	1	0			
CONTROL0	0H	TMR_RST	EN_STAT	ST	TAT	BOOST	FAULT					
CONTROL1	1H	IBUS	LIM	VLC	VWV	TE	CE#	HZ_MODE	OPA_MODE			
OREG	2H			OREG	3			DBAT_B	EOC			
IC_INFO	ЗH	Vendor	Code		PN			REVISION				
IBAT	4H	RESET		IOCHA	RGE		ITERM					
VBUS_ CONTROL	5H	Reserved	PROD	IO_LEVEL	VBUS_CON	SP	VBUSLIM					
SAFETY	6H		ISAI	FE				VSAFE				
POST_ CHARGING	7H	Reserved	Reserved	VBUS	_LOAD	PC_EN	PC_IT					
MONITOR0	10H	ITERM_CMP	VBAT_CMP	LINCHG	T_120	ICHG	IBUS	VBUS_VALID	CV			
MONITOR1	11H	GATE	VBAT	POK_B	DIS_LEVEL	NOBAT	PC_ON	Reserved	Reserved			
NTC	12H	Rese	rved	TEMP_DIS	NTC_OK	NTC4	NTC3 NTC2		NTC1			
WD_CONTROL	13H	Reserved	Reserved	Reserved	Reserved	Reserved	EN_REG	WD_DIS	Reserved			
RESTART	FA		RESTART									

Table 17. Register Bit Definitions

This table defines the operation of each register bit. Default values are in **bold** text.

Bit	Name	Value	Туре						Description	
	CONTR	OL0			Regi	ster Ad	Idress	s: 00h((0) Default Value=0100 0000 (4	10h)
7	TMR_RST	0	W			sets the bit alway			iting a 0 has no effect.	
6	EN_STAT	0	R/W		nts STA ciate fau		om goi	ng LOV	V during charging; STAT pin still pulses to	
		1		Enabl	es STA	T pin to	be LO\	W wher	n IC is charging	
				Bit 5 0	t 4 0				Description Standby	
5:4	STAT	00	R	0 1 1						
<u>_</u>	BOOST	0	R	IC is r	IC is not in Boost Mode					
3	BOOST	1		IC is in Boost Mode						
						Fault Bi		Туре	FAULT Description	
					2	1	0			
					0	0	0	R	Normal (No Fault)	
					0	0	1	R	VBUS OVP	
			See		0	1	0	RC	Sleep Mode	
2:0	FAULT	000	table to the		0	1	1	R	Poor Input Source	
			right.		1	0	0	R	Battery OVP	
					1	0	1	R	Thermal Shutdown	
					1	1	0	RC	Timer Fault	
					1	1	1	RC	No Battery	
				For B	oost Mo	ode fault	s, see	Table 1	3.	

Bit	Name	Value	Туре				Descript	ion				
CONT	ROL1			Register Address: 01h(1)			Default Value=0011 0100 (34h)				
				Input current limit	Bi	t		(4)				
				7		6	IBUSL	_{IM} (mA)				
7:6	IBUSLIM	00	R/W	0	_	0		75				
				0		1	-	' 60				
				1		0	-	080				
			-	1		1	No	Limit				
				Weak battery voltage thr	es	hold						
				5	Bi	t 4	V_{LO}	wv (V)				
5:4	VLOWV	11	R/W	0	_	0		3.4				
				0	ľ	1	:	3.5				
				1		0	:	3.6				
				1		1	:	3.7				
3	TE	0	R/W	Setting the TE bit to a 1	wil	l ena	ble Charge Te	ermination.				
2	CE#	1	R/W	This is an active low bit a the bit is reset, it will retu								
1	HZ_MODE	0	R/W	Setting this bit to a "1" pu High Impendance mode.		the o	device in					
0	OPA_MODE	0	R/W	The device is in Charge Mode when the OPA_MODE bit = 0 and in Boost Operation when the bit = 1.				See Table 11				

		r	1	r											
Bit	Name	Value	Туре					De	escript	ion					
OREG				Register	Addres	s: 02h(2)			Default	Va	lue=000	0 1000	(08h)	
				Charger output "float" voltage; programmable from 3.51 to 4.45 V in 20 mV increments.											
				Dec	Hex	VOREG		Dec	Hex	VOREG		Dec	Hex	VOREG	
				0	00	3.51		16	10	3.83		32	20	4.15	
				1	01	3.53		17	11	3.85		33	21	4.17	
				2	02	3.55		18	12	3.87		34	22	4.19	
				3	03	3.57		19	13	3.89		35	23	4.21	
				4	04	3.59		20	14	3.91		36	24	4.23	
				5	05	3.61		21	15	3.93		37	25	4.25	
7:2	OREG	000010	R/W	6	06	3.63		22	16	3.95		38	26	4.27	
				7	07	3.65		23	17	3.97		39	27	4.29	
				8	08	3.67		24	18	3.99		40	28	4.31	
					9	09	3.69		25	19	4.01		41	29	4.33
				10	0A	3.71		26	1A	4.03		42	2A	4.35	
				11	0B	3.73		27	1B	4.05		43	2B	4.37	
				12	0C	3.75		28	1C	4.07		44	2C	4.39	
				13	0D	3.77		29	1D	4.09		45	2D	4.41	
				14	0E	3.79		30	1E	4.11		46	2E	4.43	
				15	0F	3.81		31	1F	4.13		47-63	2F-3F	4.45	
		0	R/W	Indicates bit is igno		IC detec	ted	a dead	battery	after VBL	IS_	POR. W	riting a	0 to this	
1	DBAT_B	1		VBUS_P If the hos	OR. st sets th		le th	ne IC is (charging	$r < V_{SHOR}$	-				
0	EOC	0	R/W	then result the proce	umes ch ess repe	arging an ats with t	d re he "	etries Ba No Batte	ttery De ery" fau	, the IC to tection. If t re-enun State Flo	f the ciat	e battery ed, and	/ is still a		
		1		If no battery is detected when a full battery (end of charge) is reached, the PWM charger stays on, allowing the host processor to continue to run with no battery.											
IC_INF	0			Register	Addres	s: 03h(3)			Default	Va	lue=100	1 0XXX	(9Xh)	
7:6	Vendor Code	10	R	Identifies	s Fairchil	d Semico	ndu	uctor as	the IC s	upplier					
5:3	PN	010	R	Part num	ber bits	see the	Orc	lering In	formatic	n					
2:0	REV		R	IC Revis				-			se	three bi	ts		

Bit	Name	Value	Туре					Des	cription			
IBAT				Register Address: 04	h(4)				Defa	ault Value='	1000 (0001 (81h
				Conditions		Fu	ncti	onal	ity			
				Valid V _{BUS} , V _{BAT} > V _{LOV}	$\label{eq:Valid} \begin{array}{l} \mbox{Valid} \ \mbox{V}_{BUS}, \ \mbox{V}_{BAT} > \ \mbox{V}_{LOWV} \end{array} \begin{array}{l} \mbox{Setting the RESIS} \\ \mbox{SAFETY and CE} \\ \mbox{HZ}_MODE. \end{array}$					cluding WD_	ĎIS a	nd
7	RESET	1	W	Valid V _{BUS} , V _{BAT} < V _{LOV}	NV					clears all re D_DIS, HZ_		
				Absent V _{BUS} Setting the RESET bit clears all registers (exc SAFETY and CE#) including WD_DIS and HZ_MODE.								
				Writing a 0 has no effe	ct; re	ead	retu	rns 1				
				Programs the maximum charge current (550 mA default)								
				F	6	В 5	it 4	3	I _{OCHAR}	_{GE} (mA)		
					0	0	0	0	5	50		
				0	0	0	1		50			
				0	0	1	0		'50			
6:3	IOCHARGE	0000 R/W		0	0	1	1		350			
					0	1	0	0		950		
					0	1	0	1		050		
					0	1	1	0		150		
					0	1	1	1		250		
					1	0	0	0		350 450		
					-	010-				450 550		
									•	550		
				Sets the current used for		-	ing	ermi	nation			
					Bit	1	_			(mA)		
				2	1 0	0			50	0		
				0	0	1			10			
2:0	ITERM	001	R/W	0	1	0			15			
				0	1	1			20			
				1	0	0			25			
				1	0	1	\uparrow		30			
				1	1	0			35	50		
				1	1	1			40)0		

Bit	Name	Value	Туре	Description
	CONTROL		- 71	Register Address: 05h(5) Default Value=001X X100
7	 Reserved	0	R	This bit always returns 0
,	Reserved	0	R/W	Charger operates in Normal Mode.
6	PROD	1	11/10	Charger operates in Production Test Mode.
		1		
5	IO_LEVEL	0	R/W	Battery current is controlled by I _{OCHARGE} and I _{BUSLIM} bits while Fast Charging. During Precharge Mode, battery current is limited to 450 mA when I _{OCHARGE} \leq 750 mA and 730 mA when I _{OCHARGE} > 750 mA. I _{BUSLIM} bits must be set to "10" or "11" or IO_LEVEL current will remain at 200 mA.
		1		Battery current control is set to 200 mA for Fast Charge and Precharge Mode.
4	VBUS_CON		R	1 Indicates that V_{BUS} is above 4.4 V (rising) or 3.8 V (falling). When VBUS_CON changes from 0 to 1, a STAT pulse occurs.
0		0	R	VBUS control loop is not active (V _{BUS} is able to stay above V _{BUSLIM})
3	VLIM	1		VBUS control loop is active and V _{BUS} is being regulated to V _{BUSLIM}
				VBUS control voltage reference
				Bit VBUSLIM (V) 2 1 0 0 0 4.213 0 0 1
2:0	VBUSLIM	100	R/W	0 1 0 4.373
				0 1 1 4.453
				1 0 4.533 1 0 1 4.613
				1 1 0 4.693
				1 1 1 4.773
SAFE	ТҮ			Register Address: 06h(6) Default Value=0100 0000 (4Ah)
				Sets the maximum I _{OCHARGE} value used by the control circuit
				Bit Iocharge(Max) (mA) 7 6 5 4
				0 0 0 1 650 0 0 1 0 750
7:4	ISAFE	0110	R/W	
7.4	ISAFE	0110	R/VV	0 1 0 0 950
				0 1 0 1 1,050
				0 1 1 0 <u>1,050</u> 0 1 1 1 1 1,250
				0 1 1 1 1,250 1 0 0 0 1,350
				1 0 0 1 1,450
				1010-1111 1,550
				Sets the maximum V _{OREG} used by the control circuit
				Bit VOREG(MAX) (V) 3 2 1 0
				0 0 0 4.21 0 0 0 1 4.23
				0 0 0 1 4.23 0 0 1 0 4.25
3:0	VSAFE	1010	R/W	0 1 0 0 4.29
				0 1 1 0 4.33 0 1 1 1 4.35
1	1		1	1100-1111 4.45

FAN54053— High Efficiency, 1.55 A, Li-Ion Switching Charger with Power Path, USB-OTG, in a Small Solution Footprint

Bit	Name	Value	Туре	Description					
POST	_CHARGING			Register Address: 07h(7) Default Value=0000 0001 (01h)					
7:6	Reserved	00	R	These bits always return 0					
				After charger termination, in the charge done state, these bits control VBUS loadin to improve detection of AC power removal from the AC adapter.					
				[5:4] VBUS Loading in Charge Done State:					
5:4	VBUS_LOAD	00	R/W	00 None					
				01 Load VBUS for 4 ms every two seconds					
				10 Load VBUS for 131 ms every two seconds					
				11 Load VBUS for 135 ms every two seconds					
3	PC_EN	0	R/W	Post charging or background charging feature is disabled					
		1		Post charging or background charging feature is enabled					
				Sets the termination current for post charging					
				Bit PC_IT (mA)					
				$\begin{array}{c ccccccccccccccccccccccccccccccccccc$					
2:0	PC_IT	001	R/W	0 1 0 150					
				1 0 0 250 1 0 1 300					
				1 1 1 400					
MONI	TOR0			Register Address: 10h (16) Default Value=XXX0 XXXX (XXh)					
7	ITERM_CMP		R	ITERM comparator output, 1 when I _{CHARGE} > I _{TER} M reference					
6	VBAT_CMP		R	Output of VBAT comparator, 1 when $V_{BAT} < V_{BUS}$					
5	LINCHG		R	1 when 30 mA linear charger ON (V _{BAT} < V _{SHORT})					
4	T_120		R	Thermal regulation comparator, 1 when the die temperature is greater than 120°C. If battery is being charged in Precharge mode, the charge current is limited to 200 mA and in Fast Charge, 550 mA.					
3	ICHG		R	0 indicates the ICHARGE loop is controlling the battery charge current.					
2	IBUS		R	0 indicates the IBUS (input current) loop is controlling the battery charge current.					
1	VBUS_VALID		R	1 indicates V_{BUS} has passed validation and is capable of charging.					
0	CV		R	1 indicates the constant-voltage loop (OREG) is controlling the charger and all current limiting loops have released.					
MONI	TOR1			Register Address: 11h (17) Default Value=XX1X XXX0					
7	GATE		R	The GATE bit indicates the state of the GATE pin. If the bit is "0", the pin is low, driving the PFET, Q5 on. A "1" will disable Q5, but current can still flow from batter to the system through Q5's body diode.					
6	VBAT		R	A "1" indicates $V_{BAT} > V_{BATMIN}$ in PP charging or $V_{BAT} > V_{LOWV}$ in PWM charging. A "0" indicates $V_{BAT} < V_{BATMIN}$ in PP charging or $V_{BAT} < V_{LOWV}$ in PWM charging.					
5	POK_B	1	R/W	POK_B indicates the state of the POK_B pin (see section on POK_B). This bit can be set to a 1 if VBAT has fallen below V_{LOWV} , in turn the open drain POK_B pin will be Hi-Z.					
4	DIS_LEVEL		R	This pin indicates the state of the DIS pin. A "1" indicates the DIS pin is high and the device is in a Hi-Z state on the input and the PWM controller is not running.					
3	NOBAT		R	A "1" on this bit indicates that the device has determined there is no battery connected.					
2	PC_ON		R	A "1" on this bit indicates that Post charging (background charging) is in progress.					
1:0	Reserved	0	R	These bits always return 0.					

Bit	Name	Value	Туре	Description		
NTC		I	1	Register Address: 12H (18) De	efault Value=000X XXXX	
7:6	Reserved	00	R	These bits always return 0.		
	TEMP_DIS	0	R/W	NTC Temperature measurement results affect charge parameters.		
5		1		NTC Temperature measurement results do not a measurements continue to be updated every sec		
4	NTC_OK		R	0 if NTC is either shorted to GND, open, or shorted to REF.		
3	NTC4		R	1 indicates that NTC is above the T4 threshold.		
2	NTC3		R	1 indicates that NTC is above the T3 threshold.	See Battery Temperature (NTC)	
1	NTC2		R	1 indicates that NTC is above the T2 threshold.	Monitor	
0	NTC1		R	1 indicates that NTC is above the T1 threshold.		
WD_CONTROL				Register Address: 13h (19) De	fault Value = 0110 1110 (6Eh)	
7	Reserved	0	R	This bit always returns 0		
6	Reserved	1	R	This bit always returns 1		
5	Reserved	1	R	This bit always returns 1		
4	Reserved	0	R	This bit always returns 0		
3	Reserved	1	R	This bit always returns 1		
2	EN_VREG	1	R/W	The EN_VREG defaults to a "1" enabling the regulator. To disable the regulator, set the bit to a "0".		
1	WD_DIS	1	R/W	A "1" disables the Watchdog (t_{32s}) and t_{15MIN} timers. Setting the bit to a "0" will enable the timers (See Safety Timer Section for further information).		
0	Reserved	0	R	This bit always returns 0		
RESTART				Register Address: FAh (250) De	efault Value = 1111 1111 (FFh)	
7:0	RESTART		W	Writing B5h restarts charging when the IC is in the charge done state. This register reads back FF.		

PCB Layout Recommendation

Bypass capacitors should be placed as close to the IC as possible. In particular, the total loop length for CMID should be minimized to reduce overshoot and ringing on the SW, PMID, and VBUS pins. Power and ground pins should be

routed directly to their bypass capacitors using the top copper layer. The copper area connecting to the IC should be maximized to improve thermal performance.

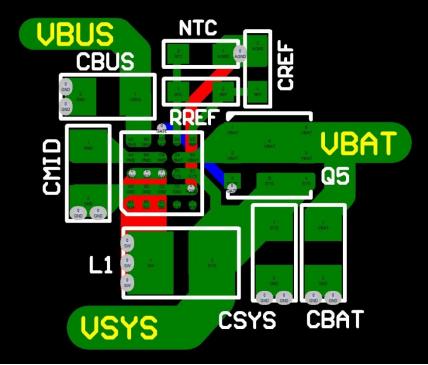
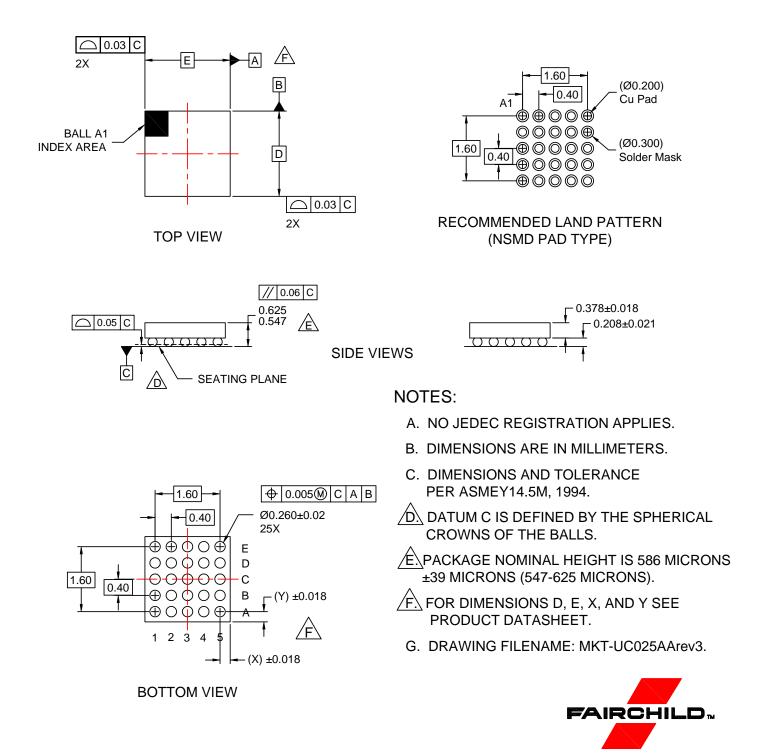


Figure 46. PCB Layout Recommendation

Product-Specific Dimensions

Product	D	E	Х	Y
FAN54053UCX	2.40 ±0.030	2.00 ±0.030	0.180	0.380



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