Half-Bridge Driver

The NCV7726A is a twelve channel half-bridge driver with protection features designed specifically for automotive and industrial motion control applications. The product has independent controls and diagnostics, and the drivers can be operated in forward, reverse, brake, and high impedance states. The device is controlled via a 16 bit SPI interface and is daisy chain compatible.

Features

- Low Quiescent Current Sleep Mode
- High−Side and Low−Side Drivers Connected in Half−Bridge Configurations
- Integrated Freewheeling Protection (LS and HS)
- 500 mA Typical, 1.1 A Peak Current
- $R_{DS(on)} = 0.85 \Omega$ (typ)
- 5 MHz SPI Communication
- 16 Bit Frame Error Detection
- Daisy Chain Compatible with Multiple of 8 bit Devices
- Compliance with 3.3 V and 5 V Systems
- Undervoltage and Overvoltage Lockout
- Discriminated Fault Reporting
- Overcurrent Protection
- Overtemperature Protection
- Low−Side Underload Detection
- Exposed Pad Package
- NCV Prefix for Automotive and Other Applications Requiring Unique Site and Control Change Requirements; AEC−Q100 Qualified and PPAP Capable
- These Devices are Pb−Free, Halogen Free/BFR Free and are RoHS Compliant

Typical Applications

- Automotive
- Industrial
- DC Motor Management for HVAC Application

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ORDERING INFORMATION

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

Figure 1. Typical Application

PIN FUNCTION DESCRIPTION The pin-out for the Half-Bridge Driver in SSOP24 NB EP package is shown in the table below.

MAXIMUM RATINGS (Voltages are with respect to GND)

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. See or download ON Semiconductor's Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

ATTRIBUTES

2. Based on JESD51-7, 1.6 mm thick FR4, 2S2P PCB with 600 mm² 2 oz. copper and 18 thermal vias to 80x80 mm 1 oz. internal spreader planes. Simulated with each channel dissipating 0.2 W.

RECOMMENDED OPERATING CONDITIONS

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(-40^{\circ}C \leq T_J \leq 150^{\circ}C, 5.5 V \leq VSx \leq 40 V, 3.15 V \leq V_{CC} \leq 5.25 V, EN = V_{CC},$ unless otherwise specified.)

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

[3.](#page-7-0) Not production tested.

[4.](#page-7-0) This is the minimum time the user must wait between SPI commands.

[5.](#page-7-0) This is the minimum time the user must wait between consecutive SRR requests.

ELECTRICAL CHARACTERISTICS (continued)

 $(-40^{\circ}C \leq T_J \leq 150^{\circ}C, 5.5 V \leq VSx \leq 40 V, 3.15 V \leq V_{CC} \leq 5.25 V, EN = V_{CC}$, unless otherwise specified.)

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ELECTRICAL CHARACTERISTICS (continued)

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Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

3. Not production tested.

4. This is the minimum time the user must wait between SPI commands.

5. This is the minimum time the user must wait between consecutive SRR requests.

CHARACTERISTIC TIMING DIAGRAMS

Figure 5. Detailed SPI Timing

4.5 2.30 ACTIVE MODE V_{CC} CURRENT (mA) $VSx = 13.2 V$ **ACTIVE MODE VCC CURRENT (mA)** 4.0 $VSx = 13.2 V$ SLEEP MODE CURRENT (µA) 2.25 **SLEEP MODE CURRENT (-**3.5 3.0 2.20 150°C $V_{\text{CC}} = 5.25$ V 2.5 125°C 2.15 2.0 $V_{\text{CC}} = 5.00 \text{ V}$ 25°C 1.5 2.10 −40°C 1.0 $V_{\text{CC}} = 3.15 \text{ V}$ 2.05 0.5 $0 \underline{\mathsf{I}}$
-50 2.00 −50 0 50 100 150 3.0 3.5 4.0 4.5 5.0 5.5 **TEMPERATURE (°C) V_{CC}, VOLTAGE (V)** Figure 6. IqTot vs. Temperature **Figure 7. I(V_{CC})** Active Mode vs. V(V_{CC}) 2.0 BODY DIODE FORWARD VOLTAGE (V) 1.05 **BODY DIODE FORWARD VOLTAGE (V)** 1.8 $VSx = 13.2 V$ 1.00 1.6 0.95 **)** 1.4 **RDS(on) (** 0.90 LSx 1.2 0.85 1.0 LSx HSx 0.80 $I_f = 0.5 A$ 0.8 HSx 0.6 -50 0.75 -50 −50 0 50 100 150 −50 0 50 100 150 C) TEMPERATURE (°C) **TEMPERATURE (°C)** Figure 8. R_{DS(on)} vs. Temperature **Figure 9. Body Diode Voltage vs. Temperature** 2.0 0.02 LSx sdSrc, IsdSnk, OVERCURRENT (A) **IsdSrc, IsdSnk, OVERCURRENT (A)** 1.5 0 1.0 −0.02 **ৰু_**০.০4 0.5 **LEAKAGE (-** $VSx = 13.2 V$ 0 V_{CC} = 5.0 V −0.06 LSx −0.5 −0.08 −1.0 −0.10 **HS_x** −0.12 −1.5 HSx –2.0 L
−50 —0.14 —
−50 −50 0 50 100 150 −50 0 50 100 150 200 **TEMPERATURE (°C)** C) TEMPERATURE (°C)

TYPICAL CHARACTERISTICS

Figure 10. Over Current vs. Temperature Figure 11. Leakage vs. Temperature

DETAILED OPERATING DESCRIPTION

General Overview

The NCV7726A is comprised of twenty four NMOS power drivers. The drivers are arranged as twelve half−bridge output channels, allowing for six independent full−bridge configured loads. Output control and status reporting is handled via the SPI (Serial Peripheral Interface) communications port.

Each output is characterized for a typical 0.5 A DC load and has a maximum 2.0 A surge capability (at $VSx =$ 13.2 V). Maximum allowable junction temperature is 150°C and may constrain the maximum load current and/or limit the number of drivers active at once.

An active−high enable function (EN) allows global control of the outputs and provides a low quiescent current sleep mode when the device is not being utilized. An internal pull−down resistor is provided on the input to ensure the device enters sleep mode if the input signal is lost.

After EN transitions from low to high, the V_{CC} POR cycle will proceed and bring the device into normal operation. The device configuration registers can then be programmed via SPI. Bringing EN low clears all registers (no configuration or status data is stored), disables the drivers, and enters sleep mode.

SPI Communication

16−bit full duplex SPI communication has been implemented for device configuration, driver control, and reading the status data. In addition to the 16−bit status data, a pseudo−bit (PRE_15) can also be retrieved from the SO output.

The device must be enabled $(EN = H)$ for SPI communication. The SPI inputs are TTL compatible and the SO output high level is defined by the applied V_{CC} . The active−low CSB input has a pull−up resistor and the remaining inputs have pull−down resistors to bias them to known states when SPI communication is inactive.

The latched thermal shutdown (TSD) status bit PRE_15 is available on SO until the first rising SCLK edge after CSB goes low. The following conditions must be met for a valid TSD read to be captured:

- 1. SCLK and SI are low before the CSB cycle;
- 2. CSB transitions from high to low;
- 3. CSB setup time (TcsbSup: Figure [5](#page-9-0), #5) is satisfied.

Figure 12 shows the SPI communication frame format, and Tables [1](#page-12-0) and [2](#page-13-0) define the command input and diagnostic status output bits.

Figure 12. SPI Communication Frame Format

Communication is implemented as follows and is also illustrated in Figures 12 and [14](#page-15-0):

- 1. SI and SCLK are set to low before the CSB cycle.
- 2. CSB goes low to begin a serial data frame; pseudo−bit PRE_15 is immediately available at SO.
- 3. SI data is shifted in on every rising edge of SCLK, starting with the most significant bit (MSB), SRR.
- 4. SI data is recognized on every falling edge of the SCLK.
- 5. Current SO data is simultaneously shifted out on every rising edge of SCLK, starting with the MSB (OCS).
- 6. CSB goes high to end the frame and SO becomes tri−state.
- 7. The last 16 bits clocked into SI are transferred to the device's data register if no frame error is detected, otherwise the entire frame is ignored and the previous input data is preserved.

Table 1. SPI COMMAND INPUT DEFINITIONS

*All command input bits are set to 0 at V_{CC} power−on reset.
**Latched faults are cleared and outputs can be re−programmed if no fault exists after SRR asserted.

*All status output bits are set to 0 at Vcc power−on reset (POR).

**HBCRx is forced to 0 when HBSTx = 0 via POR, SPI, or fault.

Table [2](#page-13-0). SPI STATUS OUTPUT DEFINITIONS (continued)

*All status output bits are set to 0 at Vcc power−on reset (POR). **HBCRx is forced to 0 when HBSTx = 0 via POR, SPI, or fault.

Frame Error Detection

The NCV7726A employs frame error detection to help ensure input data integrity. SCLK is compared to an *n x 8* bit counter and a valid frame (CSB H−L−H cycle) has integer multiples of 8 SCLK cycles. For the first 16 bits shifted into SI, SCLK is compared to a modulo16 counter $(n = 2)$, and SCLK is compared to a modulo 8 counter $(n = 1, 2, ...m)$ thereafter. This variable modulus facilitates daisy chain operation with devices using different word lengths.

The last 16 bits clocked into SI are transferred to the NCV7726A's data register if no frame error is detected, otherwise the entire frame is ignored and the previous input data is preserved.

Daisy Chain Operation

Daisy chain operation is possible with multiple 16−bit and 8−bit devices that have a compatible SPI protocol. The clock phase and clock polarity with respect to the data for all the devices in the chain must be the same as the NCV7726A.

CSB and SCLK are parallel connected to every device in the chain while SO and SI are series connected between each device.

The master's MOSI is connected to the SI of the first device and the first device's SO is connected to the next device's SI. The SO of the final device in the chain is connected to the master's MISO.

The hardware configuration for the NCV7726A daisy chained with an 8− bit SPI device is shown in Figure [13](#page-15-0). A 24−bit frame made of 16−bit word 'A' and 8−bit word 'B' is sent from the master. Command word B is sent first followed by word A. The master simultaneously receives status word B first followed by word A. The progression of data from the MCU through the sequential devices is illustrated in Figure 14.

Compliance with the illustrated frame format is required for proper daisy chain operation. Situations should be avoided where an incorrect multiple of 8 bits is sent to the devices, but the frame length does not cause a frame error in the devices. For example, the word order could be inadvertently interleaved or reversed. Invalid data is accepted by the NCV7726A in such scenarios and possibly by other devices in the chain, depending on their frame error implementation. Data is received as a command by the device at the beginning of the chain, but the device at the end of the chain may receive status data from the preceding device as a command.

CMD [x, n] = Command Word to Device 'x', Length 'n' STA [x, n] = Status Word from Device 'x', Length 'n'

Figure 13. Daisy Chain Configuration

Figure 14. Daisy Chain – 24 bit Frame Format

TSD Bit in Daisy Chain Operation

The SO path is designed to allow TSD status retrieval in a daisy chain configuration using NCV7726A or other devices with identical SPI functionality. The TSD status bit is OR'd with SI and then multiplexed with the device's usual status data (Figure [15\)](#page-16-0).

CSB is held high and SI and SCLK are held low by the master before the start of the SPI frame. TSD status is immediately available as bit PRE 15 at SO (SO = TSD) when CSB goes low to begin the frame. The usual status data (SO = STA) becomes available after the first rising SCLK edge.

The TSD status automatically propagates through the chain from the SO output of the previous device to the SI input of the next. This is shown in Figures [16](#page-16-0) and [17,](#page-16-0) first without a TSD fault in either device (Figure [16\)](#page-16-0), and then subsequently with a latched TSD fault (TSD = 1) in device "A" propagating through to device "B" (Figure [17\)](#page-16-0).

Since the TSD status of any device propagates automatically through the entire chain, it is not possible to determine which device (or devices) has a fault (TSD = 1). The usual status data from each device will need to be examined to determine where a fault (or faults) may exist.

Figure 15. TSD SPI Link

Figure 16. Daisy Chain Without TSD Fault

Figure 17. Daisy Chain With TSD Fault

Power Up/Down Control

The V_{CC} supply input powers the device's logic core. A V_{CC} power−on reset (POR) function provides controlled power−up/down. V_{CC} POR initializes the command input and status output registers to their default states (0x00), and ensures that the bridge output and SO drivers maintain Hi−Z as power is applied. SPI communication and normal device operation can proceed once V_{CC} rises above the POR threshold and EN remains high.

The VS1 and VS2 supply inputs power their respective output drivers (refer to Figure [2](#page-2-0) and the PIN FUNCTION DESCRIPTION). The VSx inputs are monitored to ensure that the supply stays within the recommended operating range. If the VSx supply moves into either of the VS undervoltage or overvoltage regions, the output drivers are switched to Hi−Z but command and status data is preserved. Output drivers will remain on if $O_UO = 0$ during an overvoltage condition.

Driver Control

The NCV7726A has the flexibility to control each half−bridge driver channel via SPI. Actual driver output state is determined by the command input and the current fault status bits as shown in Figure 18 and Table 3.

The channels are divided into two groups and each group is selected by the HBSEL input bit (see Table [1](#page-12-0)). High−side (HSx) and low−side (LSx) drivers of the same channel cannot be active at the same time, and non−overlap delays are imposed when switching between HSx and LSx drivers in the same channel. This control design thus prevents current shoot−through.

After the device has powered up and the drivers are allowed to turn on, the drivers remain on until commanded off via SPI or until a fault condition occurs.

DIAGNOSTICS, PROTECTIONS, STATUS REPORTING AND RESET

Overview

The NCV7726A employs diagnostics designed to prevent destructive overstress during a fault condition. Diagnostics are classified as either supervisory or protection functions (Table 4). Supervisory functions provide status information about device conditions. Protection functions provide status information and activate fault management behaviors.

Diagnostics resulting in output shutdown and latched status may depend on a qualifier and may require user intervention for output recovery and status memory clear. Diagnostics resulting in output lockout and non−latched status (VSOV or VSUV) may recover and clear automatically. Output configurations can be changed during output lockout. Outputs assume the new configurations or resume the previous configurations when an auto−recover fault is resolved. Table 5 shows output states during faults and output recovery modes, and Table 6 shows the status memory and memory clear modes.

Table 4. DIAGNOSTIC CLASSES AND FUNCTIONS

Name	Class	Function
TSD	Protection	Thermal Shutdown
OCS	Protection	Overcurrent Shutdown
PSF	Protection	Under/overvoltage Lockout (OVLO = 1)
ULD	Protection	Underload Shutdown
HBSTX	Supervisory	Half-Bridge X Output Status
HBCRX	Supervisory	Half-Bridge X Config Status
TW	Supervisory	Thermal Warning

Table 5. OUTPUT STATE VS. FAULT AND OUTPUT RECOVERY

*OUTx returns to its previous state (Y_n) or new state $(Yn+1)$ if fault is removed.

Table 6. STATUS MEMORY VS. FAULT AND MEMORY CLEAR

*Status memory returns to its no−fault state if fault is removed.

Status Information Retrieval

Current status information as selected by HBSEL is retrieved during each SPI frame. To preserve device configuration and output states, the previous SI data pattern must be sent during the status retrieval frame.

Status information is prevented from being updated during a SPI frame but new status becomes available after CSB goes high at the end of the frame provided the frame did not contain an SRR request. For certain device faults, it may not be possible to determine which channel (or channels) has a particular fault (or faults) since notification may be via a single global status bit. The complete status data from all channels may need to be examined to determine where a fault may exist.

Status Register Reset − SRR

Sending SRR = 1 clears status memory and re-activates faulted outputs for channels as selected by HBSEL. The previous SI data pattern must be sent with SRR to preserve device configuration and output states. SRR takes effect at the rising edge of CSB and a timer (Tsrr) is started. Tsrr is the minimum time the user must wait between consecutive SRR requests. If a fault is still present when SRR is sent, protection can be re-engaged and shutdown will recur. The status registers can also be reset by toggling the EN pin or status registers can also be
by VCC power – on reset.

Diagnostics Details

The following sections describe individual diagnostics and behaviors. In each description and illustration, a SPI frame is assumed to always be valid and the SI data pattern sent for HBCNFx and HBENx is the same as the previous frame. Actual results can depend on asynchronous fault events and SPI clock frequency and frame rate.

Undervoltage Lockout

Global Notification, Global Operation

Undervoltage detection and lockout control is provided by monitoring the VS1, VS2 and V_{CC} supply inputs. Undervoltage hysteresis is provided to ensure clean detection transitions. Undervoltage timing is shown in Figure 19.

Undervoltage at either VSx input turns off all outputs and sets the power supply fail (PSF) status bit. The outputs return to their previously programmed state and the PSF status bit is cleared when VSx rises above the hysteresis voltage level. SPI is available and programmed output enable and configuration states are maintained if proper VCC is present during VSx undervoltage. V_{CC} undervoltage turns all outputs off and clears the command input and status output registers.

Overvoltage Lockout

Global Notification, Global Operation

Overvoltage detection and lockout control is provided by monitoring the VS1 and VS2 supply inputs. Overvoltage hysteresis is provided to ensure clean detection transitions. Overvoltage timing is shown in Figure 20.

Overvoltage at either VSx input turns off all outputs if the overvoltage lockout input bit is set (OVLO = 1,

 $HBSEL = X$), and sets the power supply fail (PSF) status bit (see Tables 5 and 6). The outputs return to their previously programmed state and the PSF status bit is cleared when VSx falls below the hysteresis voltage level.

To reduce stress, it is recommended to operate the device with OVLO bit asserted to ensure that the drivers turn off during a load dump scenario.

Figure 20. Overvoltage Timing

Overcurrent Shutdown

Global Notification per HBSEL, Per Half−Bridge Operation

Overcurrent detection and shutdown control is provided by monitoring each HS and LS driver. Overcurrent timing is shown in Figure 21. Overcurrent in either driver starts a channel's overcurrent delay timer. If overcurrent exists after the delay, both drivers are latched off and the overcurrent (OCS) status bit is set. The OCS bit is cleared and channels α are re – activated by sending SRR = 1. The channel group select (HBSEL) input bit determines which channels are affected by SRR.

A persistent overcurrent cause should be resolved prior to re-activation to avoid repetitive stress on the drivers. Extended exposure to stress may affect device reliability.

Figure 21. Overcurrent Timing

Underload Shutdown

Global Notification per HBSEL, Shutdown Control per HBSEL, Per Half−Bridge Operation

Underload detection and shutdown control is provided by monitoring each LS driver. Underload timing is shown in Figure 22. Underload at a LS driver starts the global underload delay timer. If underload occurs in another channel after the global timer has been started, the delay for any subsequent underload will be the remainder of the timer. The timer runs continuously with a persistent underload condition.

If underload exists after the delay and if the underload shutdown (ULDSC) command bit is set, both HS and LS drivers are latched off and the underload (ULD) status bit is set; otherwise the drivers remain on and the ULD bit is set

(see Tables 5 and 6). The ULD bit is cleared and channels are re $-$ activated by sending $SRR = 1$. The channel group select (HBSEL) input bit determines which channels are affected by SRR and also determines which half-bridges are latched off via the ULDSC command bit (see Table 1).

Underload may result from a fault (e.g. open-load) condition or normal circuit behavior (e.g. L/R tau). In motor applications it is often desirable to actively brake the motor by turning on both HS or LS drivers in two half-bridge channels. If the configuration is two LS drivers (LS brake), an underload will result as the motor current decays normally. Utilizing HS brake instead will avoid underload notification.

Figure 22. Underload Timing

Thermal Warning and Thermal Shutdown

Global Notification, Per Half−Bridge Operation

Thermal warning (TW) and thermal shutdown (TSD) detection and control are provided for each half-bridge by monitoring the driver pair's thermal sensor. Thermal hysteresis is provided for each of the warning and shutdown functions to ensure clean detection transitions. Since TW notification precedes TSD, software polling of the TW bit enables avoidance of thermal shutdown. Thermal warning and shutdown timing is shown in Figure 23.

The TW status bit is set when a half-bridge's sensor temperature exceeds the warning level (TJ > Twr), and the bit is automatically cleared when sensor temperature falls below the warning hysteresis level (TJ < TwHy). A channel's output state is unaffected by TW.

When sensor temperature exceeds the shutdown level (TJ > Tsd), the channel's HS and LS drivers are latched off, the TW bit is/remains set, and the TSD (PRE_15) bit is set. The TSD bit is cleared and all affected channels in a group are re-activated (TJ < TsdHy) by sending SRR = 1. The channel group select (HBSEL) input bit determines which channels are affected by SSR.

Figure 23. Thermal Warning and Shutdown Timing

The latched thermal shutdown (TSD) information is available on SO after CSB goes low until the first rising SCLK edge. The following procedures must be met for a true TSD reading:

- 1. SCLK and SI are low before the CSB cycle. Violating these conditions will results in an undetermined SPI behavior or/and an incorrect TSD reading.
- 2. CSB transitioning from high to low.
- 3. CSB setup time (TcsbSup) is satisfied and the data is captured before the first SCLK rising edge.

THERMAL PERFORMANCE ESTIMATES

Figure 24. Transient R(t) vs. Pulse Time for 2 oz Spreader

PACKAGE DIMENSIONS

NOTES: 1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.

- 2. CONTROLLING DIMENSION: MILLIMETERS. 3. DIMENSION b DOES NOT INCLUDE DAMBAR PROTRUSION. DAMBAR PROTRUSION SHALL BE 0.10 MAX. AT MMC. DAMBAR CANNOT BE LOCATED ON THE LOWER RADIUS OF THE FOOT. DIMENSION b APPLIES TO THE FLAT SECTION OF THE LEAD BETWEEN 0.10 TO 0.25 FROM THE LEAD TIP.
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- 6. DATUMS A AND B ARE DETERMINED AT DATUM PLANE H.
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