

## MAX86140/ MAX86141

## Best-in-Class Optical Pulse Oximeter and Heart-Rate Sensor for Wearable Health

### General Description

The MAX86140/MAX86141 are ultra-low-power, completely integrated, optical data acquisition systems. On the transmitter side, the MAX86140/MAX86141 have three programmable high-current LED drivers that can be configured to drive up to six LEDs using an external 3x2:1 mux. With two MAX86140/MAX86141 devices working in master-slave mode, the LED drivers can drive up to twelve LEDs using an external 3x2:1 mux. On the receiver side, MAX86140 consists of a single optical readout channel, while the MAX86141 has two optical readout channels that can operate simultaneously. The devices have low-noise signal conditioning analog front-end (AFE) including 19-bit ADC, an industry-lead ambient light cancellation (ALC) circuit, and a picket fence detect and replace function. Due to the low power consumption, compact size, ease of use, and industry-lead ambient light rejection capability of MAX86140/MAX86141, the devices are ideal for a wide variety of optical-sensing applications, such as pulse oximetry and heart rate detection.

The MAX86140/MAX86141 operate on a 1.8V main supply voltage and a 3.1V to 5.5V LED driver supply voltage. Both devices support a standard SPI compatible interface and fully autonomous operation. Each device has a large 128-word built-in FIFO. The MAX86140/MAX86141 is available in compact wafer-level package (WLP) (2.048 x 1.848mm) with 0.4mm ball pitch.

### Applications

- Wearable Devices for Fitness, Wellness and Medical Applications
- Optimized for Wrist, Finger, Ear, and Other Locations
- Optimized Performance to Detect
  - Optical Heart Rate
  - Oxygen Saturation (SpO<sub>2</sub>)
  - Muscle Oxygen Saturation (SmO<sub>2</sub> and StO<sub>2</sub>)

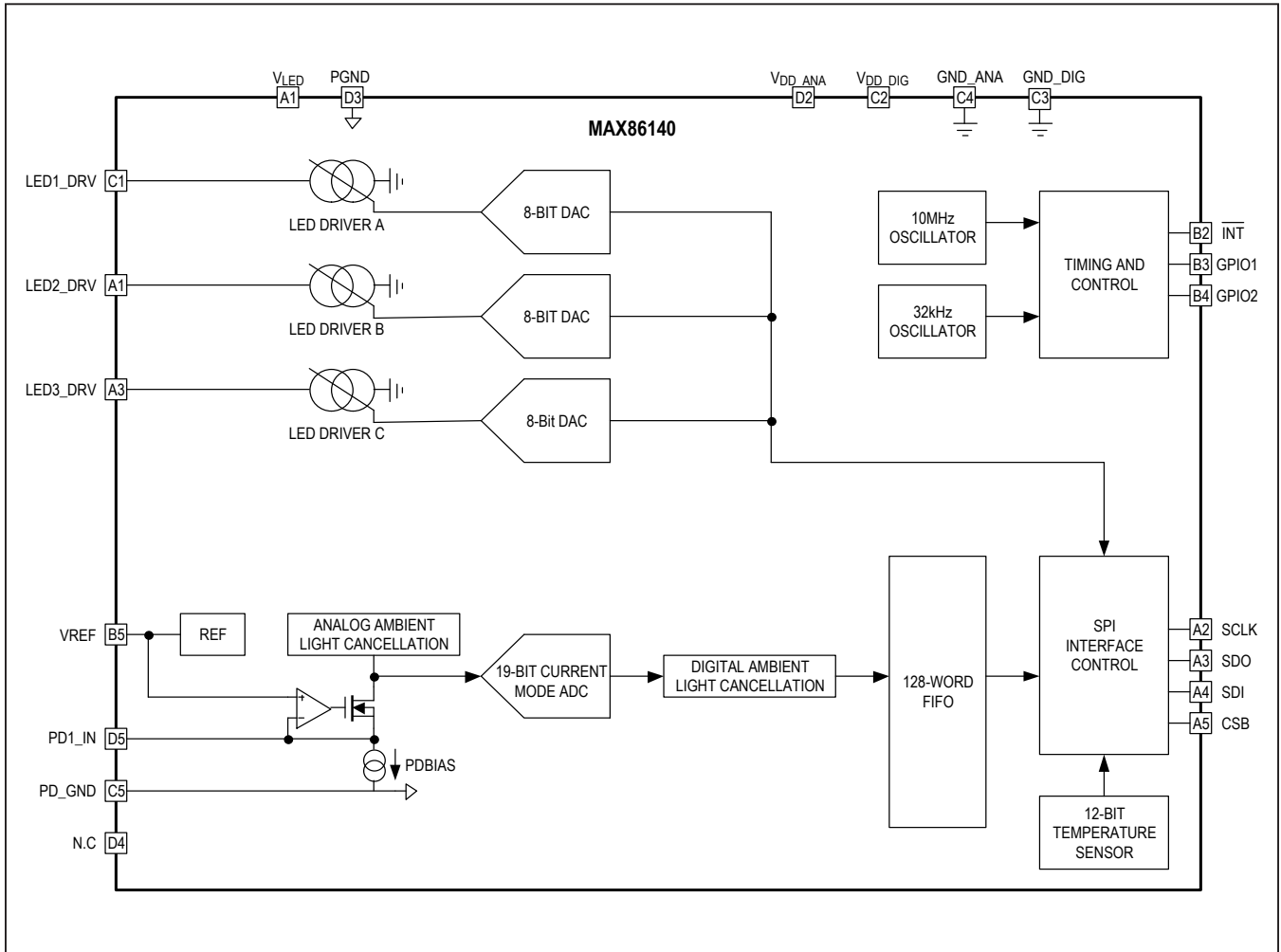
### Benefits and Features

- Complete Single and Dual-Channel Optical Data Acquisition System
- Optimized Architecture for Transmissive and Reflective Heart Rate or SpO<sub>2</sub> Monitoring
- Low Dark Current Noise of < 50pA RMS (Sample to Sample Variance)
- Lower Effective Dark Current Noise Achievable Through Multiple Sample Modes and On-Chip Averaging
- High-Resolution, 19-Bit Charge Integrating ADC
- Three Low-Noise, 8-Bit LED Current DACs
- Excellent Dynamic Range > 89dB in White Card Loop-Back Test (Sample-to-Sample Variance)
- Dynamic Range Extendable to > 104dB for SpO<sub>2</sub> and > 110dB for HRM with Multiple Sample Modes and On-Chip Averaging
- Excellent Ambient Range and Rejection Capability
  - > 100μA Ambient Photodetector Current
  - > 70dB Ambient Rejection at 120Hz
- Ultra-Low-Power Operation for Wearable Devices
  - Low-Power Operation, Optical Readout Channel < 10μA (typ) at 25sps
  - Short Exposure Integration Period of 14.8μs, 29.4μs, 58.7μs, 117.3μs
  - Low Shutdown Current = 0.6μA (typ)
- Rejection of Fast Ambient Transients
- Miniature 2.048 x 1.848mm, 5 x 4 0.4mm Ball Pitch WLP
- -40°C to +85°C Operating Temperature Range

*[Ordering Information](#) appears at end of data sheet.*

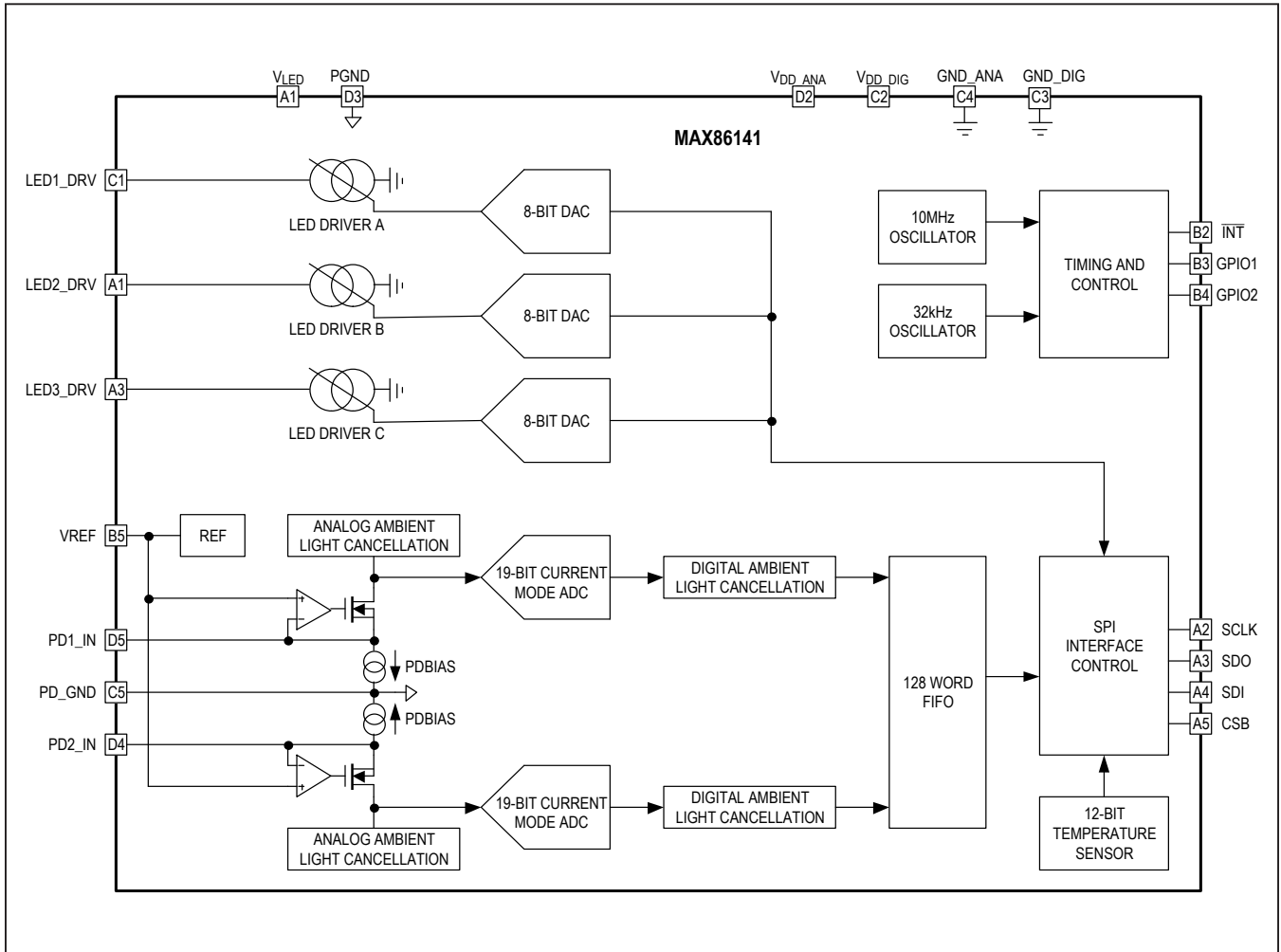
Detailed Block Diagrams

MAX86140



Detailed Block Diagrams (continued)

MAX86141



### Absolute Maximum Ratings

V <sub>DD_ANA</sub> to GND_ANA.....	-0.3V to +2.2V	PD1_IN to GND_ANA.....	-0.3V to +2.2V
V <sub>DD_DIG</sub> to GND_ANA.....	-0.3V to +2.2V	PD2_IN to GND_ANA.....	-0.3V to +2.2V
V <sub>DD_ANA</sub> to V <sub>DD_DIG</sub> .....	-0.3V to +0.3V	PD_GND to GND_ANA.....	-0.3V to +0.3V
PGND to GND_ANA.....	-0.3V to +0.3V	All other pins to GND_ANA.....	-0.3V to +2.2V
SCLK, SDO, SDI, CSB, INT, GPIO1, GPIO2 to GND_ANA.....	-0.3V to +6.0V	Output Short-Circuit Duration.....	Continuous
GND_DIG to GND_ANA.....	-0.3V to +0.3V	Continuous Input Current Into Any Pin (except LEDx_DRV Pins).....	±20mA
V <sub>LED</sub> to PGND.....	-0.3V to +6.0V	Continuous Power Dissipation (WLP (derate 5.5mW/°C above +70°C)).....	440mW
LED1_DRV to PGND.....	-0.3V to V <sub>LED</sub> + 0.3V	Operating Temperature Range.....	-40°C to +85°C
LED2_DRV to PGND.....	-0.3V to V <sub>LED</sub> + 0.3V	Storage Temperature Range.....	-40°C to +105°C
LED3_DRV to PGND.....	-0.3V to V <sub>LED</sub> + 0.3V		

Stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### Package Information

#### 5 x 4 WLP

PACKAGE CODE	N201A2+1
Outline Number	<a href="#">21-100134</a>
Land Pattern Number	Refer to <a href="#">Application Note 1891</a>
<b>Thermal Resistance, Four-Layer Board:</b>	
Junction to Ambient ( $\theta_{JA}$ )	55.49°C/W
Junction to Case ( $\theta_{JC}$ )	N/A

For the latest package outline information and land patterns (footprints), go to [www.maximintegrated.com/packages](http://www.maximintegrated.com/packages). Note that a “+”, “#”, or “-” in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to [www.maximintegrated.com/thermal-tutorial](http://www.maximintegrated.com/thermal-tutorial).

### Electrical Characteristics

(V<sub>DD\_ANA</sub> = 1.8V, V<sub>DD\_DIG</sub> = 1.8V, V<sub>LED</sub> = 5.0V, PPG<sub>x</sub>\_ADC\_RGE = 16μA, PPG\_SR = 1024sps, PPG\_TINT = 14.8μs, LED\_SETLNG = 6μs, LED<sub>x</sub>\_RGE = 31mA, C<sub>PD</sub> = 65pF, PDBIAS<sub>x</sub> = 0x1, I<sub>exposure</sub> = 1μA, T<sub>A</sub> = +25°C, min/max are from T<sub>A</sub> = -40°C to +85°C, unless otherwise noted. (Note 1, 2))

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
<b>Readout Channel</b>						
ADC Resolution				19		bits
ADC Full Scale Input Current		PPG <sub>x</sub> _ADC_RGE = 0x0		4.0		μA
		PPG <sub>x</sub> _ADC_RGE = 0x1		8.0		
		PPG <sub>x</sub> _ADC_RGE = 0x2		16.0		
		PPG <sub>x</sub> _ADC_RGE = 0x3		32.0		
ADC Integration Time	t <sub>INT</sub>	PPG_TINT = 0x0		14.8		μs
		PPG_TINT = 0x1		29.4		
		PPG_TINT = 0x2		58.7		
		PPG_TINT = 0x3		117.3		
Minimum PPG Sample Rate		PPG_SR = 0x0A		8		sps
Maximum PPG Sample Rate		PPG_SR = 0x13		4096		sps
Sample Rate Error		From nominal as indicated in the PPG_SR table	-2		+2	%
Maximum DC Ambient Light Rejection	ALR	ALC = on, ALC_OVF = 1		200		μA
AC Ambient Light Rejection	AC_ALRR	ALC = on, I <sub>ambient</sub> = 1μA DC with ±0.4μA pk-pk 120Hz Sinwave		70		dB
DC Ambient Light Rejection		ALC = on, I <sub>ambient</sub> modulated between 0μA and 30μA, LED_SETLNG = 12μs, PPG_TINT = 117.3μs		0.5		nA
Dark Current Offset	DC_O	ALC = ON, PDBIAS <sub>x</sub> = 0x0, ADD_OFFSET = 1		±1		Counts
Dark Current Input Referred Noise		PPG_TINT = 14.8μs		262		pArms
		PPG_TINT = 29.4μs		128		
		PPG_TINT = 58.7μs		83		pArms
		PPG_TINT = 117.3μs		56		
Maximum Photodiode Input Capacitance	C <sub>pd</sub>	I <sub>ambient</sub> = 0μA, less than 1nA of code shift	PDBIAS <sub>x</sub> = 0x1		65	pF
			PDBIAS <sub>x</sub> = 0x5		130	
			PDBIAS <sub>x</sub> = 0x6		260	
			PDBIAS <sub>x</sub> = 0x7		520	
VDD DC PSR		I <sub>ambient</sub> = 0μA, V <sub>DD_ANA</sub> = V <sub>DD_DIG</sub> = 1.7V to 2.0V	-560	-330	+560	LSB/V
<b>LED Driver</b>						
LED Current Resolution				8		Bits
Driver DNL		LED <sub>x</sub> _RGE = 124mA	-1		1	LSB
Driver INL		LED <sub>x</sub> _RGE = 124mA		0.6		LSB

**Electrical Characteristics (continued)**

( $V_{DD\_ANA} = 1.8V$ ,  $V_{DD\_DIG} = 1.8V$ ,  $V_{LED} = 5.0V$ ,  $PPGX\_ADC\_RGE = 16\mu A$ ,  $PPG\_SR = 1024sps$ ,  $PPG\_TINT = 14.8\mu s$ ,  $LED\_SETLNG = 6\mu s$ ,  $LEDx\_RGE = 31mA$ ,  $C_{PD} = 65pF$ ,  $PDBIASx = 0x1$ ,  $I_{exposure} = 1\mu A$ ,  $T_A = +25^\circ C$ , min/max are from  $T_A = -40^\circ C$  to  $+85^\circ C$ , unless otherwise noted. (Note 1, 2))

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
Full Scale LED Current	$I_{LED}$	$LEDx\_PA = 0xFF$	$LEDx\_RGE = 0x0$		31	mA	
			$LEDx\_RGE = 0x1$		62		
			$LEDx\_RGE = 0x2$		93		
			$LEDx\_RGE = 0x3$	117	124		129
Minimum output voltage	$V_{OL}$	$LEDx\_PA = 0xFF$ , 95% of the desired LED current	$LEDx\_RGE = 0x0$		160	253	mV
			$LEDx\_RGE = 0x1$		317		
			$LEDx\_RGE = 0x2$		495		
			$LEDx\_RGE = 0x3$		700		
LED Driver DC PSR		$V_{LEDx\_DRV} = 0.9V$ , $LEDx\_PA = 0xFF$ , $V_{LED} = 3.1V$ to $5.5V$ , $LEDx\_RGE = 124mA$		-1	+400	$\mu A/V$	
		$V_{DD\_ANA} = V_{DD\_DIG} = 1.7V$ to $2.0V$ , $T_A = +25^\circ C$ , $LEDx\_PA = 0xFF$		110	1410		
LED1 Driver Compliance Interrupt	$LED1_{COMP}$			180		mV	
<b>Internal Die Temperature Sensor</b>							
Temperature Sensor Accuracy		$T_A = +25^\circ C$		1		$^\circ C$	
Temperature Sensor Minimum Range		Temperature error $< 5^\circ C$		-40		$^\circ C$	
Temperature Sensor Maximum Range		Temperature error $< 5^\circ C$		85		$^\circ C$	
Temperature ADC Acquisition Time				29		ms	
<b>Power Supply</b>							
Power Supply Voltage	$V_{DD\_ANA}$ , $V_{DD\_DIG}$	Verified during PSRR Test	1.7	1.8	2.0	V	
LED Supply Voltage	$V_{LED}$	Verified during PSRR Test	3.1		5.5	V	

**Electrical Characteristics (continued)**

(V<sub>DD\_ANA</sub> = 1.8V, V<sub>DD\_DIG</sub> = 1.8V, V<sub>LED</sub> = 5.0V, PPG<sub>x\_ADC\_RGE</sub> = 16μA, PPG<sub>\_SR</sub> = 1024sps, PPG<sub>\_TINT</sub> = 14.8μs, LED<sub>\_SETLNG</sub> = 6μs, LED<sub>x\_RGE</sub> = 31mA, C<sub>PD</sub> = 65pF, PDBIAS<sub>x</sub> = 0x1, I<sub>exposure</sub> = 1μA, T<sub>A</sub> = +25°C, min/max are from T<sub>A</sub> = -40°C to +85°C, unless otherwise noted. (Note 1, 2))

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS		
VDD Supply Current	I <sub>DD</sub>	MAX86140, Single LED Exposure/Sample. PPG <sub>_SR</sub> = 4096sps, LP <sub>_MODE</sub> = 0x0, LED <sub>x</sub> PA = 0mA		660	780	μA		
		MAX86140, Single LED Exposure/Sample, PW = 14.8μs, LP <sub>_MODE</sub> = 0x1, LED <sub>x</sub> _PA = 0mA	PPG <sub>_SR</sub> = 256sps	80				
			PPG <sub>_SR</sub> = 100sps	32				
			PPG <sub>_SR</sub> = 50sps	16				
			PPG <sub>_SR</sub> = 25sps	8.5				
		MAX86140, Double LED Exposure/Sample, PPG <sub>_SR</sub> = 84sps, PW = 14.8μs, LP <sub>_MODE</sub> = 0x1, LED <sub>x</sub> _PA = 0mA	Single pulse	42		μA		
			Double pulse	89		μA		
		MAX86141, Single LED Exposure/Sample. PPG <sub>_SR</sub> = 4096sps, LP <sub>_MODE</sub> = 0x0, LED <sub>x</sub> PA = 0mA	I <sub>DD</sub>	MAX86141, Single LED Exposure/Sample, LP <sub>_BOOST</sub> = 1, LED <sub>x</sub> _PA = 0mA		978	1170	μA
				MAX86141, Single LED Exposure/Sample, LP <sub>_BOOST</sub> = 1, LED <sub>x</sub> _PA = 0mA	PPG <sub>_SR</sub> = 256sps	115.5		
					PPG <sub>_SR</sub> = 100sps	46		
PPG <sub>_SR</sub> = 50sps	23							
PPG <sub>_SR</sub> = 25sps	11							
MAX86141, Two LED Exposure/Sample, PPG <sub>_SR</sub> = 84sps, LP <sub>_BOOST</sub> = 1, LED <sub>x</sub> _PA = 0mA	Single pulse			60		μA		
	Double pulse			130		μA		
Die Temperature mode, SPS = 1, Optical channel(s) disabled			8		μA			
VLED Supply Current	I <sub>LED</sub>	Single LED exposure per Sample, PPG <sub>_TINT</sub> = 117.3μs, Single-Pulse, PPG <sub>_SR</sub> = 256sps, LED <sub>x</sub> _PA = 0mA		0.22		μA		
		Single LED exposure per Sample, PPG <sub>_TINT</sub> = 117.3μs, Single-Pulse, LED <sub>x</sub> _PA = 62mA	PPG <sub>_SR</sub> = 256sps	1880				
			PPG <sub>_SR</sub> = 100sps	735				
			PPG <sub>_SR</sub> = 50sps	370				
			PPG <sub>_SR</sub> =25sps	185				
		Two LED exposure per sample, PPG <sub>_TINT</sub> = 117.3μs, LED <sub>x</sub> _PA = 62mA, PPG <sub>_SR</sub> = 84sps	Single-Pulse	1240		μA		
Dual-Pulse	2480							
VDD Current in Shutdown		T <sub>A</sub> = +25°C	0.6	2.5	μA			
VLED Current in Shutdown		T <sub>A</sub> = +25°C		1	μA			

### Electrical Characteristics (continued)

( $V_{DD\_ANA} = 1.8V$ ,  $V_{DD\_DIG} = 1.8V$ ,  $V_{LED} = 5.0V$ ,  $PPGx\_ADC\_RGE = 16\mu A$ ,  $PPG\_SR = 1024sps$ ,  $PPG\_TINT = 14.8\mu s$ ,  $LED\_SETLNG = 6\mu s$ ,  $LEDx\_RGE = 31mA$ ,  $C_{PD} = 65pF$ ,  $PDBIASx = 0x1$ ,  $I_{exposure} = 1\mu A$ ,  $T_A = +25^\circ C$ , min/max are from  $T_A = -40^\circ C$  to  $+85^\circ C$ , unless otherwise noted. (Note 1, 2))

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
<b>Digital I/O Characteristics</b>						
SDO Output Low Voltage	$V_{OL\_SDO}$	$I_{SINK} = 2mA$			0.4	V
SDO Output High Voltage	$V_{OH\_SDO}$	$I_{SOURCE} = 2mA$	$V_{DD}-0.4$			V
Open-Drain Output Low Voltage	$V_{OL\_OD}$	$I_{SINK} = 6mA$ , INTB, GPIO1, GPIO2			0.4	V
Input Voltage Low	$V_{IL}$	SDI, SCLK, CSB, GPIO1, GPIO2			0.4	V
Input Voltage High	$V_{IH}$	SDI, SCLK, CSB, GPIO1, GPIO2	1.4			V
Input Hysteresis	$V_{HYS}$	SDI, SCLK, CSB		330		mV
		GPIO1, GPIO2		240		
Input Leakage Current	$I_{IN}$	$V_{IN} = 0V$ , $T_A = +25^\circ C$ (SDI, SCLK, CSB, GPIO1, GPIO2)		0.01	1	$\mu A$
Input Capacitance	$C_{IN}$	SDI, SCLK, CSB, GPIO1, GPIO2		10		pF
<b>SPI Timing Characteristics (Note 3)</b>						
SCLK Frequency	$f_{SCLK}$				8	MHz
SCLK Period	$t_{CP}$		125			ns
SCLK Pulse Width High	$t_{CH}$		40			ns
SCLK Pulse Width Low	$t_{CL}$		40			ns
CSB Fall to SCLK Rise Setup Time	$t_{CSS0}$	To 1 <sup>st</sup> SCLK rising edge	20			ns
CSB Fall to SCLK Rise Hold Time	$t_{CSH0}$	Applies to inactive rising edge preceding 1 <sup>st</sup> rising edge	5			ns
CSB Rise to SCLK Rise Hold Time	$t_{CSH1}$	Applies to 24th rising edge	500			ns
SCLK Rise to CSB Fall	$t_{CSF}$	Applies to 24th rising edge	500			ns
CSB Pulse Width High	$t_{CSPW}$		250			ns
SDI to SCLK Rise Setup Time	$t_{DS}$		10			ns
SDI to SCLK Rise Hold Time	$t_{DH}$		10			ns
SCLK Fall to SDO Transition	$t_{DOT}$	$C_{LOAD} = 50pF$			35	ns
CSB Fall to SDO Enabled	$t_{DOE}$	$C_{LOAD} = 0pF$	12			ns
CSB Rise to SDO Hi-Z	$t_{DOZ}$	Disable Time			25	ns
GPIO1 External Sync Pulse Width	$t_{PLGPIO1}$		5			$\mu s$
GPIO2 External Clock Input (Note 4)	$f_{GPIO2}$	External Sample Reference Clock on GPIO2	31900		32868	Hz
GPIO2 External Clock Pulse Width	$t_{PWGPIO2}$		1			$\mu s$

**Note 1:** All devices are 100% production tested at  $T_A = +25^\circ C$ . Specifications over temperature limits are guaranteed by Maxim Integrated's bench or proprietary automated test equipment (ATE) characterization.

**Note 2:** All other register settings assumed to be POR status unless otherwise noted.

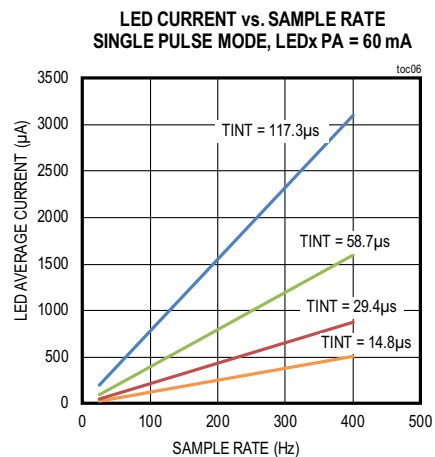
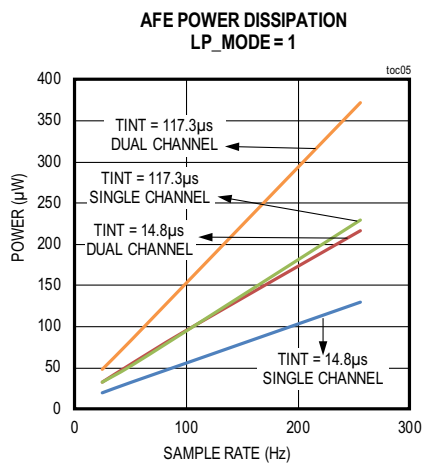
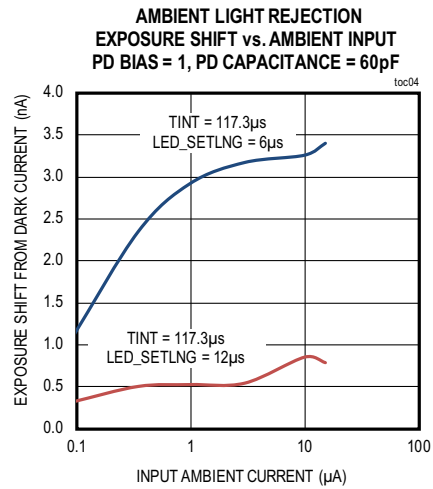
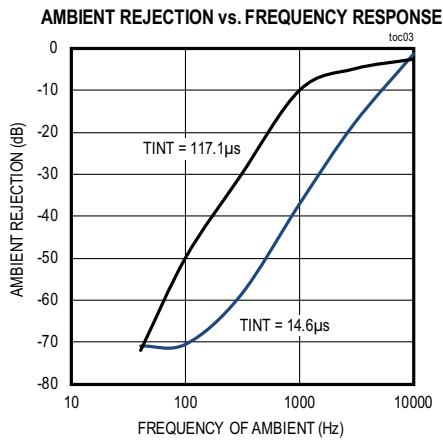
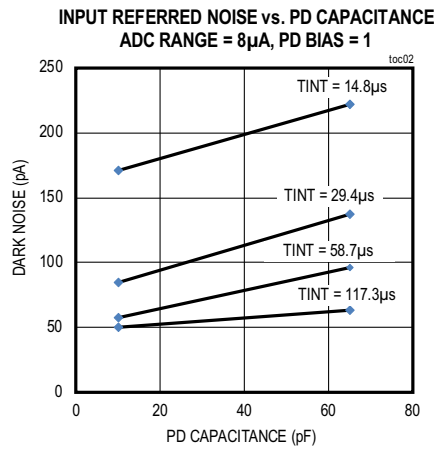
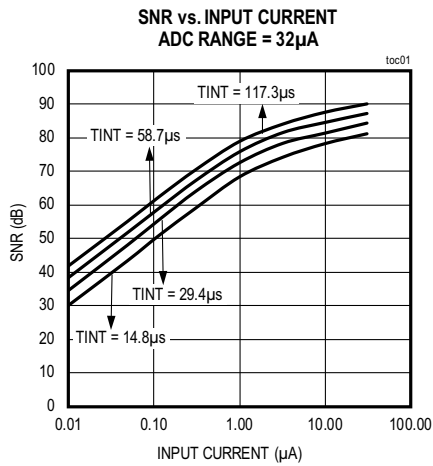
**Note 3:** Guaranteed by design. Not production tested.

**Note 4:** See *Register Map/PPG Configuration 2 (0x12)* section for the sample rate by the external clock frequency. The sample rate shifts if the external clock frequency drifts.



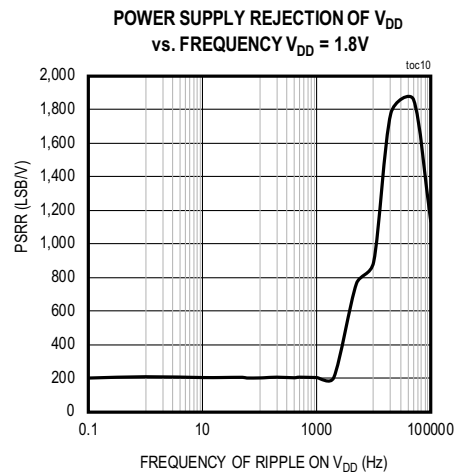
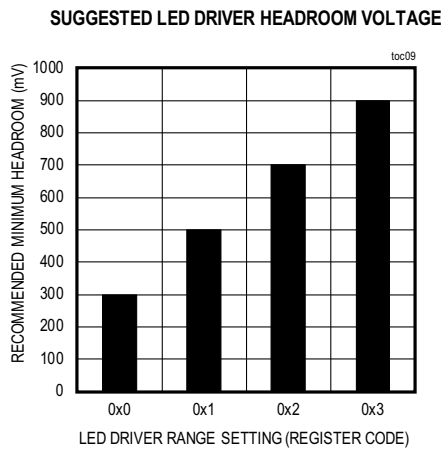
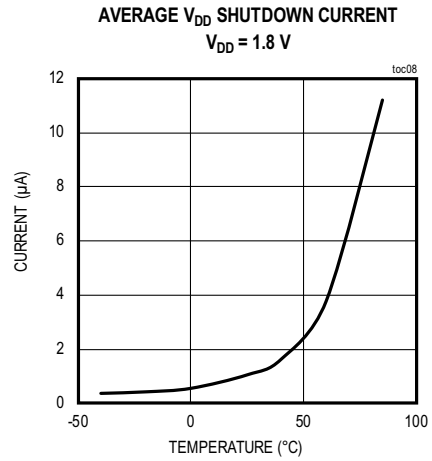
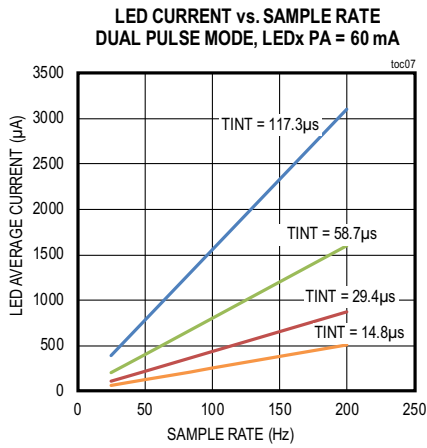
Typical Operating Characteristics

( $V_{DD\_ANA} = V_{DD\_DIG} = 1.8V$ ,  $V_{LED} = 5.0V$ ,  $GND\_ANA = GND\_DIG = PGND = 0V$ ,  $T_A = +25^\circ C$ , unless otherwise noted.)



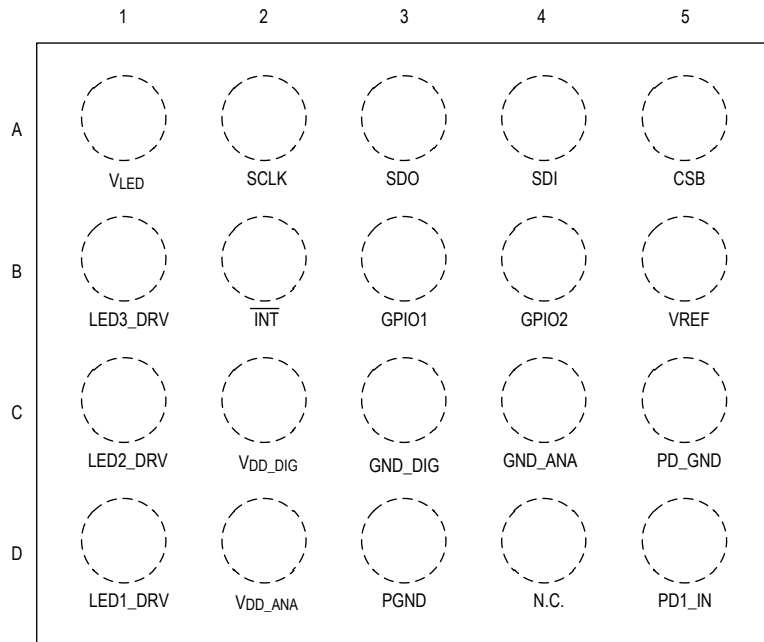
Typical Operating Characteristics (continued)

( $V_{DD\_ANA} = V_{DD\_DIG} = 1.8V$ ,  $V_{LED} = 5.0V$ ,  $GND\_ANA = GND\_DIG = PGND = 0V$ ,  $T_A = +25^\circ C$ , unless otherwise noted.)

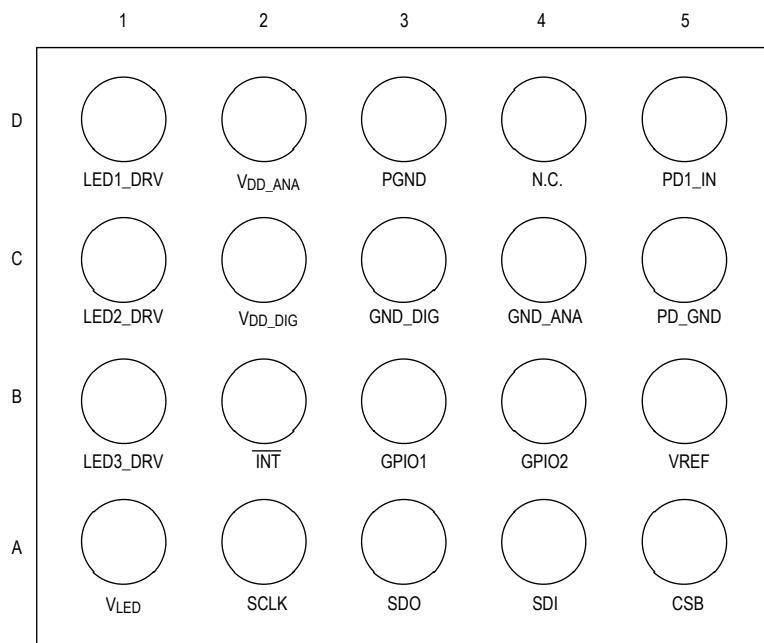


Pin Configurations

MAX86140



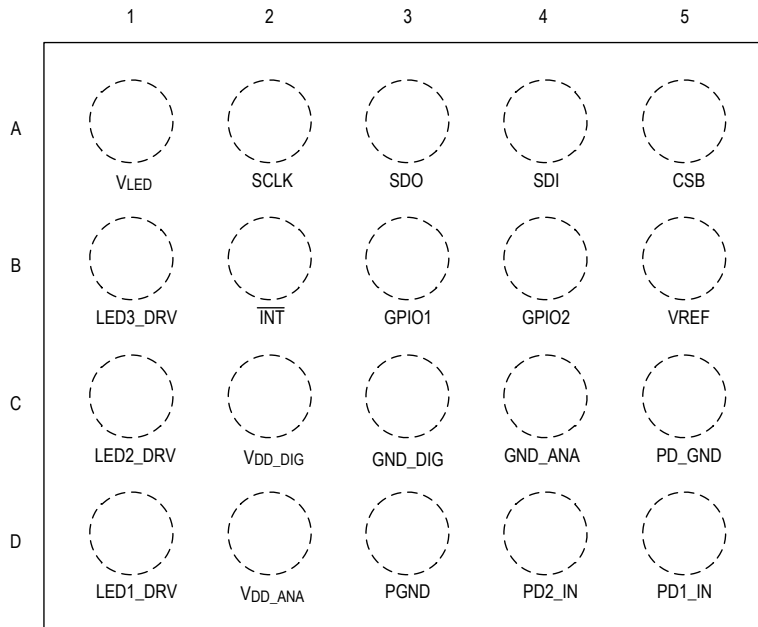
BUMPS FACING DOWN, 0.4MM PITCH



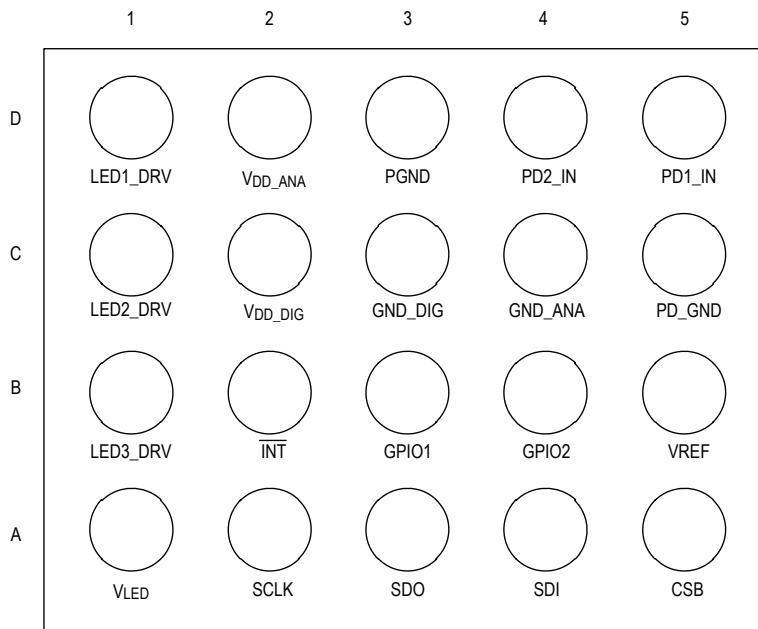
BUMPS FACING UP, 0.4MM PITCH

Pin Configurations (continued)

MAX86141



BUMPS FACING DOWN, 0.4MM PITCH

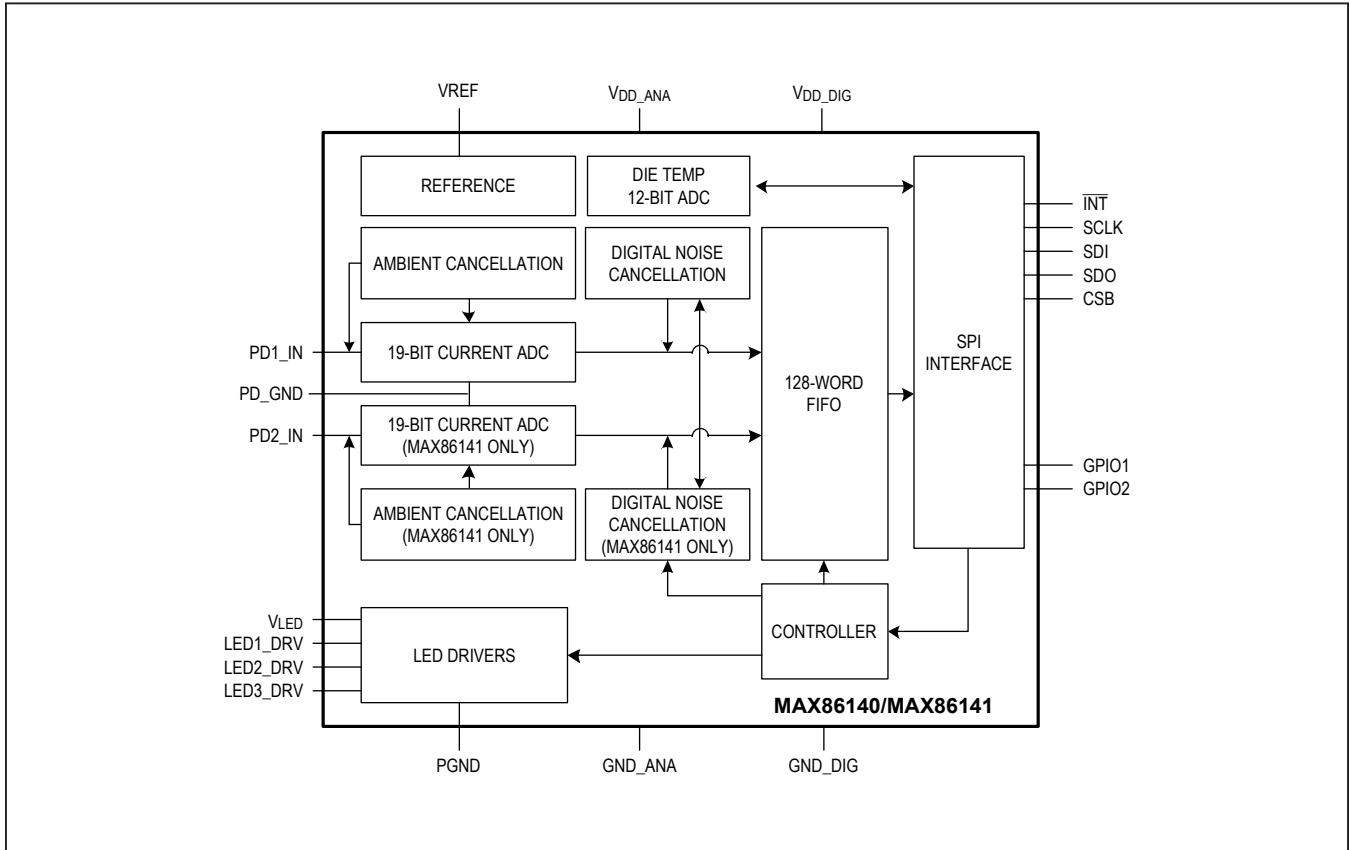


BUMPS FACING UP, 0.4MM PITCH

## Pin Description

PIN		NAME	FUNCTION
MAX86140	MAX86141		
<b>Power</b>			
C2	C2	V <sub>DD_DIG</sub>	Digital Logic Supply. Connect to externally-regulated supply. Bypass to GND_DIG
C3	C3	GND_DIG	Digital Logic and Digital Pad Return. Connect to PCB Ground.
D2	D2	V <sub>DD_ANA</sub>	Analog Supply. Connect to externally-regulated supply. Bypass with a 0.1µF capacitor as close as possible to bump and a 10µF capacitor to GND_ANA.
C4	C4	GND_ANA	Analog Power Return. Connect to PCB Ground.
A1	A1	V <sub>LED</sub>	LED Power Supply Input. Connect to external voltage supply. Bypass with a 10µF capacitor to PGND.
D3	D3	PGND	LED Power Return. Connect to PCB Ground.
<b>Control Interface</b>			
A2	A2	SCLK	SPI Clock
A3	A3	SDO	SPI Data Output
A4	A4	SDI	SPI Data Input
A5	A5	CSB	SPI Chip select
B2	B2	INT̄	Interrupt. Programmable open-drain Interrupt output signal pin (active-low).
B3	B3	GPIO1	General Purpose I/O. Open-drain when programmed as output (active-low).
B4	B4	GPIO2	General Purpose I/O. Open-drain when programmed as output (active-low).
<b>Optical Pins</b>			
—	D4	PD2_IN	Photodiode Cathode Input
D5	D5	PD1_IN	Photodiode Cathode Input
C5	C5	PD_GND	Photodiode Anode
D1	D1	LED1_DRV	LED Output Driver 1. Connect the LED cathode to LED1_DRV and its anode to the V <sub>LED</sub> supply.
C1	C1	LED2_DRV	LED Output Driver 2. Connect the LED cathode to LED2_DRV and its anode to the V <sub>LED</sub> supply.
B1	B1	LED3_DRV	LED Output Driver 3. Connect the LED cathode to LED3_DRV and its anode to the V <sub>LED</sub> supply.
<b>Reference</b>			
B5	B5	VREF	Internal Reference Decoupling Point. Bypass with a 1µF capacitor to GND_ANA.
<b>N.C.</b>			
D4		N.C.	No Connection. Connect to unconnected PCB pad for mechanical stability. N.C. pins should not be connected to any signal, power, or ground pins.

Simplified Block Diagram



## Detailed Description

The MAX86140/MAX86141 are complete integrated optical data acquisition systems, ideal for optical pulse oximetry and heart rate detection applications. Both parts have been designed for the demanding requirements of mobile and wearable devices and require minimal external hardware components are necessary for integration into a wearable device. They include high-resolution, optical readout signal processing channels with robust ambient light cancellation and high-current LED driver DACs to form a complete optical readout signal chain.

The MAX86140/MAX86141 are fully adjustable through software registers and the digital output data is stored in a 128-word FIFO within the IC. The FIFO allows the MAX86140/MAX86141 to be connected to a microcontroller or processor on a shared bus, where the data is not being read continuously from the MAX86140/MAX86141's registers. Both operate in fully autonomous modes for low power battery applications.

The MAX86140 consists of a single optical readout channel, while the MAX86141 incorporates dual optical readout channels that operate simultaneously. Both parts have three LED drivers and are well suited for a wide variety of optical sensing applications.

The MAX86140/MAX86141 operate on a 1.8V main supply voltage, with a separate 3.1V to 5.5V LED driver power supply. Both devices have flexible timing and shut-down configurations as well as control of individual blocks so an optimized measurement can be made at minimum power levels.

## Optical Subsystem

The optical subsystem in the MAX86140/MAX86141 is composed of ambient light cancellation (ALC), a continuous-time sigma-delta ADC, and proprietary discrete time filter. ALC incorporates a proprietary scheme to cancel ambient light generated photodiode current, allowing the sensor to work in high ambient light conditions. The optical ADC has programmable full-scale ranges of 4 $\mu$ A to 32 $\mu$ A. The internal ADC is a continuous time oversampling sigma delta converter with 19-bit resolution.

The ADC output data rate can be programmed from 8sps (samples per second) to 4096sps per channel. The MAX86140/MAX86141 includes a proprietary discrete time filter to reject 50Hz/60Hz interference and changing residual ambient light from the sensor measurements.

The MAX86140/MAX86141 supports Dynamic Power Down mode (Low-Power mode) in which the power consumption is decreased between samples. This mode is only supported for sample rates 128sps and below. For more details on the power consumption at each sample rates, refer to the [Electrical Characteristics](#) table.

## LED Driver

The MAX86140/MAX86141 integrates three precision LED driver-current DACs that modulate LED pulses for a variety of optical measurements. The LED current DACs have 8-bits of dynamic range with four programmable full-scale ranges of 31mA, 62mA, 94mA, and 124mA. The LED drivers are low dropout current sources, allowing for low-noise, power-supply independent LED currents to be sourced at the lowest supply voltage possible; therefore minimizing LED power consumption. The LED pulse width can be programmed from 14.8 $\mu$ s to 117.3 $\mu$ s to allow the algorithms to optimize SpO<sub>2</sub> and HR accuracy at the lowest dynamic power consumption dictated by the application.

## FIFO Configuration

The FIFO is 128 sample depth and is designed to support various data types, as shown in [Table 2](#). Each sample width is 3 bytes, which includes a 5-bit tag width. The tag embedded in the FIFO\_DATA is used to identify the source of each sample data. The description of each Tag is as shown in [Table 3](#).

## LED Sequence Control (address 0x20 ~ 0x22)

The data format in the FIFO, as well as the sequencing of exposures, are controlled by the LED Sequence Registers through LEDC1 through LEDC6. There are six LED Sequence Data Items available, as shown in [Table 1](#). The exposure sequence cycles through the LED Sequence bit fields, starting from LEDC1 to LEDC6. The first LED Sequence field set to NONE (0000) ends the sequence.

**Table 1. LED Sequence Control Registers**

ADDRESS	REGISTER NAME	DE-FAULT VALUE	B7	B6	B5	B4	B3	B2	B1	B0
0x20	LED Sequence Register 1	00	LEDC2[3:0]				LEDC1[3:0]			
0x21	LED Sequence Register 2	00	LEDC4[3:0]				LEDC3[3:0]			
0x22	LED Sequence Register 3	00	LEDC6[3:0]				LEDC5[3:0]			

Table 2 lists the codes for exposures selected in the LED sequence control registers.

**Table 2. LED Sequence Register Data Type**

LEDCN[3:0]	DATA TYPE
0000	NONE
0001	LED1
0010	LED2
0011	LED3
0100	LED1 and LED2 pulsed simultaneously
0101	LED1 and LED3 pulsed simultaneously
0110	LED2 and LED3 pulsed simultaneously
0111	LED1, LED2, and LED3 pulsed simultaneously
1000	Pilot on LED1
1001	DIRECT AMBIENT
1010	LED4 (external mux control)
1011	LED5 (external mux control)
1100	LED6 (external mux control)
1101	Reserved
1110	Reserved
1111	Reserved



Table 3 shows the format of the FIFO data along with the associated Tag. In a sample if a picket fence event is detected, the predicted value is pushed to the FIFO along with its tag (PPFx\_LEDCx\_DATA).

**Table 3. FIFO Data and Tag**

TAG[4:0]	DATA TYPE	COMMENTS
00001	PPG1 LEDC1 DATA	If LEDC1 is non-zero
00010	PPG1 LEDC2 DATA	If LEDC1 and LEDC2 are non-zero
00011	PPG1 LEDC3 DATA	If LEDC1, LEDC2 and LEDC3 are non-zero
00100	PPG1 LEDC4 DATA	If LEDC1, LEDC2, LEDC3, and LEDC4 are non-zero
00101	PPG1 LEDC5 DATA	If LEDC1, LEDC2, LEDC3, LEDC4, and LEDC5 are non-zero
00110	PPG1 LEDC6 DATA	If LEDC1, LEDC2, LEDC3, LEDC4, LEDC5, and LEDC6 are non-zero
00111	PPG2 LEDC1 DATA	If LEDC1 is non-zero
01000	PPG2 LEDC2 DATA	If LEDC1 and LEDC2 are non-zero
01001	PPG2 LEDC3 DATA	If LEDC1, LEDC2, and LEDC3 are non-zero
01010	PPG2 LEDC4 DATA	If LEDC1, LEDC2, LEDC3, and LEDC4 are non-zero
01011	PPG2 LEDC5 DATA	If LEDC1, LEDC2, LEDC3, LEDC4, and LEDC5 are non-zero
01100	PPG2 LEDC6 DATA	If LEDC1, LEDC2, LEDC3, LEDC4, LEDC5, and LEDC6 are non-zero
01101	PPF1 LEDC1 DATA	If LEDC1 is non-zero (Picket Fence Event)
01110	PPF1 LEDC2 DATA	If LEDC1 and LEDC2 are non-zero (Picket Fence Event)
01111	PPF1 LEDC3 DATA	If LEDC1, LEDC2, and LEDC3 are non-zero (Picket Fence Event)
10000	Reserved	
10001	Reserved	
10010	Reserved	
10011	PPF2 LEDC1 DATA	If LEDC1 is non-zero (Picket Fence Event)
10100	PPF2 LEDC2 DATA	If LEDC1 and LEDC2 are non-zero (Picket Fence Event)
10101	PPF2 LEDC3 DATA	If LEDC1, LEDC2, and LEDC3 are non-zero (Picket Fence Event)
10110	Reserved	
10111	Reserved	
11000	Reserved	
11001	PROX1 DATA	Only PILOT LED1 for LEDC1 is used
11010	PROX2 DATA	Only PILOT LED1 for LEDC1 is used
11011	Reserved	
11100	Reserved	
11101	Reserved	
11110	INVALID DATA	This tag indicates that there was an attempt to read an empty FIFO
11111	TIME STAMP	If TIME_STAMP_EN = 1, this is TIME_STAMP

There are seven registers that control how the FIFO is configured and read out. These registers are illustrated below.

**Table 4. PPG Configuration**

AD-DRESS	REGISTER NAME	B7	B6	B5	B4	B3	B2	B1	B0
0x04	FIFO Write Pointer	-			FIFO_WR_PTR[6:0]				
0x05	FIFO Read Pointer	-			FIFO_RD_PTR[6:0]				
0x06	Overflow Counter	-			OVF_COUNTER[6:0]				
0x07	FIFO Data Counter	FIFO_DATA_COUNT[7:0]							
0x08	FIFO Data Register	FIFO_DATA[7:0]							
0x09	FIFO Configuration 1	-	FIFO_A_FULL[6:0]						
0x0A	FIFO Configuration 2	-	-	TIME_STAMP_EN	FLUSH_FIFO	FIFO_STAT_CLR	A_FULL_TYPE	FIFO_RO	-

**Write Pointer (Register 0x04)**

FIFO\_WR\_PTR[6:0] points to the FIFO location where the next item will be written. This pointer advances for each item pushed on to the FIFO by the internal conversion process. The write pointer is a 7-bit counter and will wrap around to count 0x00 on the next item after count 0x7F.

**Read Pointer (Register 0x05)**

FIFO\_RD\_PTR[6:0] points to the location from where the next item from the FIFO will be read via the serial interface. This advances each time an item is read from the FIFO. The read pointer can be both read and written to. This allows an item to be reread from the FIFO if it has not already been overwritten. The read pointer is updated from a 7-bit counter and will wrap around to count 0x00 from count 0x7F.

**Overflow Counter (Register 0x06)**

OVF\_COUNTER[6:0] logs the number of items lost if the FIFO is not read in a timely fashion. This counter holds/saturates at count value 0x7F. When a complete item is popped from the FIFO (when the read pointer advances), the OVF\_COUNTER is reset to zero. This counter is essentially a debug tool. It should be read immediately before reading the FIFO in order to check if an overflow condition has occurred.

**FIFO Data Counter (Register 0x07)**

FIFO\_DATA\_COUNT[7:0] is a read-only register which holds the number of items available in the FIFO for the host to read. This increments when a new item is pushed to the FIFO, and decrements when the host reads an item from the FIFO.

**FIFO Data (Register 0x08)**

FIFO\_DATA[7:0] is a read-only register used to retrieve data from the FIFO. It is important to burst read the item from the FIFO. Each item is three bytes. So burst reading three bytes at FIFO\_DATA register via the serial interface advances the FIFO\_RD\_PTR. The format and data type of the data stored in the FIFO is determined by the Tag associated with data. Readout from the FIFO follows a progression defined by LED Sequence Control registers as well. This configuration is best illustrated by a few examples.

Assume it is desired to perform a SpO<sub>2</sub> measurement and also monitor the ambient level on the photodiode to adjust the IR and red LED intensity. To perform this measurement, configure the following registers:

<b>LED Sequence Control</b>	
LEDC1 = 0x1	(LED1 exposure)
LEDC2 = 0x2	(LED2 exposure)
LEDC3 = 0x9	(DIRECT AMBIENT exposure)
LEDC4 = 0x0	(NONE)
LEDC5 = 0x0	(NONE)
LEDC6 = 0x0	(NONE)
<b>PPG Configuration</b>	
PPG1_ADC_RGE[1:0]	(PPG1 Gain Range Control)
PPG2_ADC_RGE[1:0]	(PPG2 Gain Range Control)
PPG_TINT[1:0]	(LED Pulse-Width Control)
PPG_SR[3:0]	(Sample Rate)
<b>LED Pulse Amplitude</b>	
LED1_PA[7:0]	(LED1 Drive Current)
LED2_PA[7:0]	(LED2 Drive Current)

When done so the sample sequence and the data format in the FIFO follows the following time/location sequence.

tag 1, PPG1 LED1 data  
tag 7, PPG2 LED1 data  
tag 2, PPG1 LED2 data  
tag 8, PPG2 LED2 data  
tag 3, PPG1 Ambient data  
tag 9, PPG2 Ambient data  
tag 1, PPG1 LED1 data  
tag 7, PPG2 LED1 data  
tag 2, PPG1 LED2 data  
tag 8, PPG2 LED2 data  
tag 3, PPG1 Ambient data  
tag 9, PPG2 Ambient data

.

.

.

tag 1, PPG1 LED1 data  
tag 7, PPG2 LED1 data  
tag 2, PPG1 LED2 data  
tag 8, PPG2 LED2 data  
tag 3, PPG1 Ambient data  
tag 9, PPG2 Ambient data

where:

PPGm LED1 data = the ambient corrected exposure data from LED1 in PPGm channel  
PPGm LED2 data = the ambient corrected exposure data from LED2 in PPGm channel  
PPGm Ambient data = the direct ambient sample in PPGm channel  
m = 1 of PPG1 channel, and 2 for PPG2 channel

For a second example, assume it is desired to pulse LED1 and LED2 simultaneously while also monitoring the ambient level.

LED Sequence Control	
LEDC1 = 0x4	(LED1 and LED2 exposure)
LEDC2 = 0x9	(DIRECT AMBIENT exposure)
LEDC3 = 0x0	(NONE)
LEDC4 = 0x0	(NONE)
LEDC5 = 0x0	(NONE)
LEDC6 = 0x0	(NONE)

In this case, the sequencing in the FIFO will then be:

```

tag 1, PPG1 LED1+LED2 data
tag 7, PPG2 LED1+LED2 data
tag 2, PPG 1 Ambient data
tag 8, PPG 2 Ambient data
tag 1, PPG1 LED1+LED2 data
tag 7, PPG2 LED1+LED2 data
tag 2, PPG1 Ambient data
tag 8, PPG2 Ambient data

```

.  
.  
.

```

tag 1, PPG1 LED1+LED2 data
tag 7, PPG2 LED1+LED2 data
tag 2, PPG1 Ambient data
tag 8, PPG2 Ambient data

```

where:

PPGm LED1+LED2 data = the ambient corrected exposure data from LED1 and LED2 for PPGm channel  
PPGm Ambient data = the direct ambient corrected sample for PPGm channel

The number of bytes of data for the PPG channel is given by:  $2 \times 3 \times K \times N$

where:

K = the number of active exposures as defined in the LED Sequence Control registers 0x20, 0x21, and 0x22.  
N = the number of samples in the FIFO

To calculate the number of available items one can perform the following pseudo-code:

```

read the OVF_COUNTER register
read the FIFO_DATA_COUNT register
if OVF_COUNTER == 0 //no overflow occurred
    NUM_AVAILABLE_SAMPLES = FIFO_DATA_COUNT
else
    NUM_AVAILABLE_SAMPLES = 128 // overflow occurred and data has been lost
endif

```

[Table 6](#) shows the FIFO data format depends on the data type being stored. Optical data, whether full ambient corrected LED exposure, ambient corrected proximity or direct ambient sampled data is left-justified, as shown in [Table 6](#). Bits F23:F19 of the FIFO word contains the tag that identifies the data.

**Table 6. Optical FIFO Data Format**

FIFO DATA FORMAT (FIFO_DATA[23:14])									
Tag (TAG[4:0])					ADC Value (FIFO_DATA[18:14])				
F23	F22	F21	F20	F19	F18	F17	F16	F15	F14
T4	T3	T2	T1	T0	O18	O17	O16	O15	O14

FIFO DATA FORMAT (FIFO_DATA[13:4])									
ADC Value (FIFO_DATA[13:4])									
F13	F12	F11	F10	F9	F8	F7	F6	F5	F4
O13	O12	O11	O10	O9	O8	O7	O6	O5	O4

FIFO DATA FORMAT (FIFO_DATA[3:0])			
ADC Value (FIFO_DATA[3:0])			
F3		F2	
O3		O2	
F1		F0	
O1		O0	

**FIFO\_A\_FULL (address 0x09)**

The FIFO\_A\_FULL[6:0] field in the FIFO Configuration 1 register (0x09) sets the watermark for the FIFO and determines when the A\_FULL bit in the Interrupt\_Status register (0x00) gets asserted. The A\_FULL bit will be set when the FIFO contains 128 minus FIFO\_A\_FULL[6:0] items. When the FIFO is almost full, if the A\_FULL\_EN mask bit in the Interrupt\_Enable register (0x03) is set, then A\_FULL bit gets asserted in the Interrupt Status 1 register and this bit is routed to the INT pin on the serial interface. This condition should prompt the applications processor to read samples off of the FIFO before it fills. The A\_FULL bit is cleared when the status register is read.

The application processor can read both the FIFO\_WR\_PTR and FIFO\_RD\_PTR to calculate the number of items available in the FIFO, or just read the OVF\_COUNTER and FIFO\_DATA\_COUNT registers, and read as many items as it needs to empty the FIFO. Alternatively, if the applications always responds much faster than the selected sample rate, it could just read 128 minus FIFO\_A\_FULL[6:0] items when it gets A\_FULL interrupt and be assured that all data from the FIFO is read.

**FIFO\_RO (Address 0x0A)**

The FIFO\_RO bit in the FIFO Configuration 2 register (0x0A) determines whether samples get pushed on to the FIFO when it is full. If push is enabled when FIFO is full, old samples are lost. If FIFO\_RO is not set, the new sample is dropped and the FIFO is not updated.

**A\_FULL\_TYPE (Address 0x0A)**

The A\_FULL\_TYPE bit defines the behavior of the A\_FULL interrupt. If the A\_FULL\_TYPE bit is set low, the A\_FULL

interrupt gets asserted when the A\_FULL condition is detected and cleared by status register read, but reasserts for every sample if the A\_FULL condition persists. If A\_FULL\_TYPE bit is set high, the A\_FULL interrupt gets asserted only when a new A\_FULL condition is detected. The interrupt gets cleared on Interrupt Status 1 register read, and does not reassert for every sample until a new A\_FULL condition is detected.

**FIFO\_STAT\_CLR (Address 0x0A)**

The FIFO\_STAT\_CLR bit defines whether the A-FULL interrupt should get cleared by FIFO\_DATA register read. If FIFO\_STAT\_CLR is set low, A\_FULL and DATA\_RDY interrupts do not get cleared by FIFO\_DATA register read but get cleared by status register read. If FIFO\_STAT\_CLR is set high, A\_FULL and DATA\_RDY interrupts get cleared by a FIFO\_DATA register read or a status register read.

**FLUSH\_FIFO (Address 0x0A)**

The FIFO Flush bit is used for flushing the FIFO. The FIFO becomes empty and the FIFO\_WR\_PTR[6:0], FIFO\_RD\_PTR[6:0], FIFO\_DATA\_COUNT[7:0] and OVF\_COUNTER[6:0] get reset to zero. FLUSH\_FIFO is a self-clearing bit.

**TIME\_STAMP\_EN (Address 0x0A)**

When TIME\_STAMP\_EN bit is set to 1, the 19 bits time stamp gets pushed to the FIFO along with its Tag for every 8 samples. This timestamp is useful for aligning data from two devices after the host reads the FIFOs of those devices. When TIME\_STAMP\_EN bit is set to 0, the sample counter is not pushed to FIFO.

### Pseudo-Code Example of Initializing the Optical AFE

The following pseudo-code shows an example of configuring MAX86140/MAX86141 for a SpO2 applications, where LED1 and LED2 are IR and red LED, respectively.

```

DEVICE OPEN
START;
// AFE Initialization
WRITE 0x1 to RESET[0];           // Soft Reset (Register 0x0D[0])
DELAY 1ms;
READ Interrupt Status 1;         // Clear Interrupt (Register 0x00)
READ Interrupt Status 2;         // Clear Interrupt (Register 0x01)
WRITE 0x1 to SHDN[0];           // Shutdown (Register 0x0D[1])
WRITE 0x3 to PPG_TINT[1:0];      // Pulse Width = 123.8ms (Register 0x11[1:0])
WRITE 0x2 to PPG1_ADC_RGE1:0];   // ADC Range = 16µA (Register 0x11[3:2])
WRITE 0x2 to PPG2_ADC_RGE1:0];   // ADC Range = 16µA (Register 0x11[3:2])
// For MAX86141 when used in Dual Channel only
WRITE 0x0 to SMP_AVE[2:0];       // Sample Averaging = 1 (Register 0x12[2:0])
WRITE 0x00 to PPG_SR[4:0];       // Sample Rate = 25sps (Register 0x12[7:3])
WRITE 0x3 to LED_SETLNG[1:0];    // LED Settling Time = 12ms (Register 0x13[7:6])
WRITE 0x01 to PDBIAS1[2:0];      // PD 1 Biasing for Cpd = 0~65pF (Register 0x15[2:0])
WRITE 0x01 to PDBIAS2[2:0];      // PD 1 Biasing for Cpd = 0~65pF (Register 0x15[2:0])
// For MAX86141 when used in Dual Channel only
WRITE 0x3 to LED1_RGE[1:0];      // LED Driver 1 Range = 124mA (Register 0x15[2:0])
WRITE 0x3 to LED2_RGE[1:0];      // LED Driver 2 Range = 124mA (Register 0x15[2:0])
WRITE 0x20 to LED1_DRV[1:0];     // LED 1 Drive Current = 15.36mA (Register 0x23[7:0])
WRITE 0x20 to LED2_DRV[1:0];     // LED 2 Drive Current = 15.36mA (Register 0x24[7:0])
WRITE 0x1 to LP_Mode[0];         // Low Power mode enabled
// FIFO Configuration
WRITE 0x10 to FIFO_A_FULL[6:0];  // FIFO INT triggered condition (Register 0x09[6:0])
WRITE 0x1 to FIFO_RO;           // FIFO Roll Over enabled (Register 0x0A[1])
WRITE 0x1 to A_FULL_EN;         // FIFO_A_FULL interrupt enabled (Register 0x02[7])
WRITE 0x1 to LEDC1[3:0];        // LED1 exposure configured in time slot 1
WRITE 0x2 to LEDC2[3:0];        // LED2 exposure configured in time slot 1
WRITE 0x0 to LEDC3[3:0];
WRITE 0x0 to LEDC4[3:0];
WRITE 0x0 to LEDC5[3:0];
WRITE 0x0 to LEDC6[3:0];
WRITE 0x0 to SHDN[0];           // Start Sampling STOP;

```

### Pseudo-Code for Interrupt Handling with FIFO\_A\_FULL

The following pseudo-code shows an example on handling the Interrupt when using A\_FULL Interrupt.

```

void irqHandler(void)
{
    uint8_t intStatus;
    //Read Status
    ReadReg(0x00, &intStatus);
    if ( intStatus & 0x80 ) { //A FULL RDY
        device_data_read(); //Data Read Routine
    }
}

```

**Pseudo-Code Example of Reading Data from FIFO**

Example pseudo-code for reading data from FIFO when using single photodiode channel and two LED channels.

```
#define FIFO_SAMPLES (128-0x10) // FIFO_A_FULL[6:0] = 0x10
void device_data_read(void) {
    int i;
    uint8_t sampleCnt;
    uint8_t dataBuf[FIFO_SAMPLES*3]; //(128 - FIFO_A_FULL[6:0]) samples, 3 byte/channel
    uint8_t tag1[FIFO_SAMPLES/2]; //(128 - FIFO_A_FULL[6:0])/2channels samples
    uint8_t tag2[FIFO_SAMPLES/2]; //(128 - FIFO_A_FULL[6:0])/2channels samples
    int led1[FIFO_SAMPLES/2]; //(128 - FIFO_A_FULL[6:0])/2channels samples
    int led2[FIFO_SAMPLES/2]; //(128 - FIFO_A_FULL[6:0])/2channels samples

    ReadReg(0x07, &sampleCnt); // sampleCnt should be the same value as FIFO_SAMPLES
    //Read FIFO
    ReadFifo(dataBuf, sampleCnt * 3);

    for ( i = 0; i < sampleCnt/2/*channels*/; i++ ) {
        tag1[i] = (dataBuf[i*6+0] >> 3) & 0x1f;
        led1[i] = ((dataBuf[i*6+0] << 16) | (dataBuf[i*6+1] << 8) | (dataBuf[i*6+2]))
            & 0x7fff;
        tag2[i] = (dataBuf[i*6+3] >> 3) & 0x1f;
        led2[i] = ((dataBuf[i*6+3] << 16) | (dataBuf[i*6+4] << 8) | (dataBuf[i*6+5]))
            & 0x7fff;
    }
}
```

Example pseudo-code for reading data from FIFO when using dual photodiode channels and two LED channels.

```
#define FIFO_SAMPLES (128-0x10) // FIFO_A_FULL[6:0] = 0x10
void device_data_read(void) {
    int i;
    uint8_t sampleCnt;
    uint8_t dataBuf[FIFO_SAMPLES*3]; //(128 - FIFO_A_FULL[6:0]) samples, 3 byte/channel
    uint8_t tag1A[FIFO_SAMPLES/4]; //(128 - FIFO_A_FULL[6:0])/4channels samples
    uint8_t tag1B[FIFO_SAMPLES/4]; //(128 - FIFO_A_FULL[6:0])/4channels samples
    uint8_t tag2A [FIFO_SAMPLES/4]; //(128 - FIFO_A_FULL[6:0])/4channels samples
    uint8_t tag2B [FIFO_SAMPLES/4]; //(128 - FIFO_A_FULL[6:0])/4channels samples
    int led1A[FIFO_SAMPLES/4];
    int led1B[FIFO_SAMPLES/4];
    int led2A[FIFO_SAMPLES/4];
    int led2B[FIFO_SAMPLES/4];

    ReadReg(0x07, &sampleCnt); // sampleCnt should be the same value as FIFO_SAMPLES
    //Read FIFO
    ReadFifo(dataBuf, sampleCnt * 3);

    for ( i = 0; i < sampleCnt/4/*channels*/; i++ ) {
        tag1A[i] = (dataBuf[i*12+0] >> 3) & 0x1f;
        led1A[i] = ((dataBuf[i*12+0] << 16) | (dataBuf[i*12+1] << 8) |
            dataBuf[i*12+2])) & 0x7fff; // LED1, PD1
        tag1B[i] = (dataBuf[i*12+3] >> 3) & 0x1f;
        led1B[i] = ((dataBuf[i*12+3] << 16) | (dataBuf[i*12+4] << 8) |
            (dataBuf[i*12+5])) & 0x7fff; // LED1, PD2
        tag2A[i] = (dataBuf[i*12+6] >> 3) & 0x1f;
        led2A[i] = ((dataBuf[i*12+6] << 16) | (dataBuf[i*12+7] << 8) |
            (dataBuf[i*12+8])) & 0x7fff; // LED2, PD1
        tag2B[i] = (dataBuf[i*12+9] >> 3) & 0x1f;
        led2B[i] = ((dataBuf[i*12+9] << 16) | (dataBuf[i*12+10] << 8) |
            (dataBuf[i*12+11])) & 0x7fff; // LED2, PD2
    }
}
```

**Optical Timing**

The MAX86140/MAX86141 optical controller is capable of being configured to make a variety of measurements. Each LED exposure is ambient light compensated before the ADC conversion.

The controller can be configured to pulse one, two or three LED drivers sequentially so as to make measurements at multiple wavelengths as is done in a pulse oximetry measurements or simultaneously to drive multiple LEDs as is done with heart rate measurements on the wrist.

The controller is also configurable to measure direct ambient level for every exposure sample. The direct ambient measurement can be used to adjust the LED drive

level to compensate for increased noise levels when high interfering ambient signals are present.

The following optical timing diagrams illustrate several possible measurement configurations.

**One LED Pulsing with No Direct Ambient Sampling**

The optical timing diagram below represents just LED1 pulsing during the exposure time with no direct ambient sampling enabled. This timing mode would be used when heart rate is being measured with a single green LED. In this mode a single optical sampled value will appear successively in the FIFO.

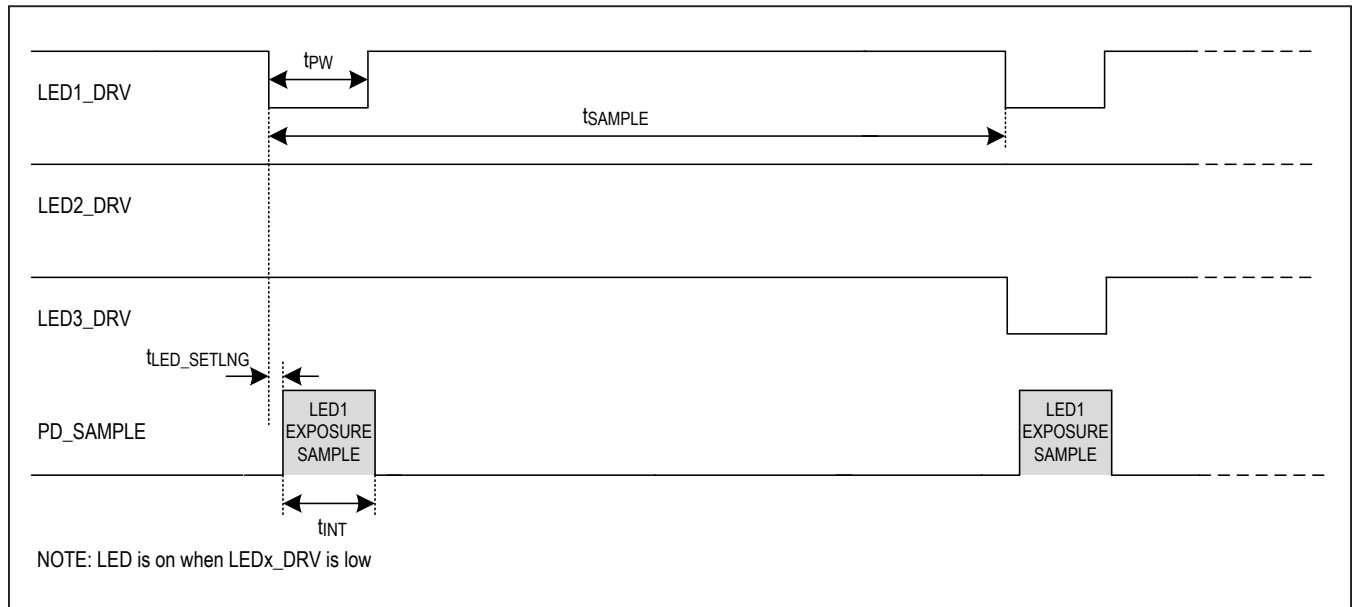


Figure 1. Timing for LED1 Pulsing with No Direct Ambient Sampling



**One LED Pulsing with Direct Ambient Sampling**

The optical timing diagram below represents just LED1 pulsing during the exposure time with direct ambient sampling enabled. This timing mode would be used when heart rate is being measured with a single, green LED. In this mode a single optical sampled value followed by the ambient sampled value will appear successively in the FIFO.

**Two LEDs Pulse Simultaneously with Direct Ambient Sampling**

The optical timing diagram below represents both LED1 and LED2 pulsing simultaneously with direct ambient sampling enabled. This timing mode would be used when heart rate is being measured with two green LEDs. In this mode a single optical sampled value followed by the ambient sampled value will appear in successive the FIFO locations. The direct ambient sampling is typically used to compensate the LED drive levels as the optical noise level can be elevated from ambient shot noise.

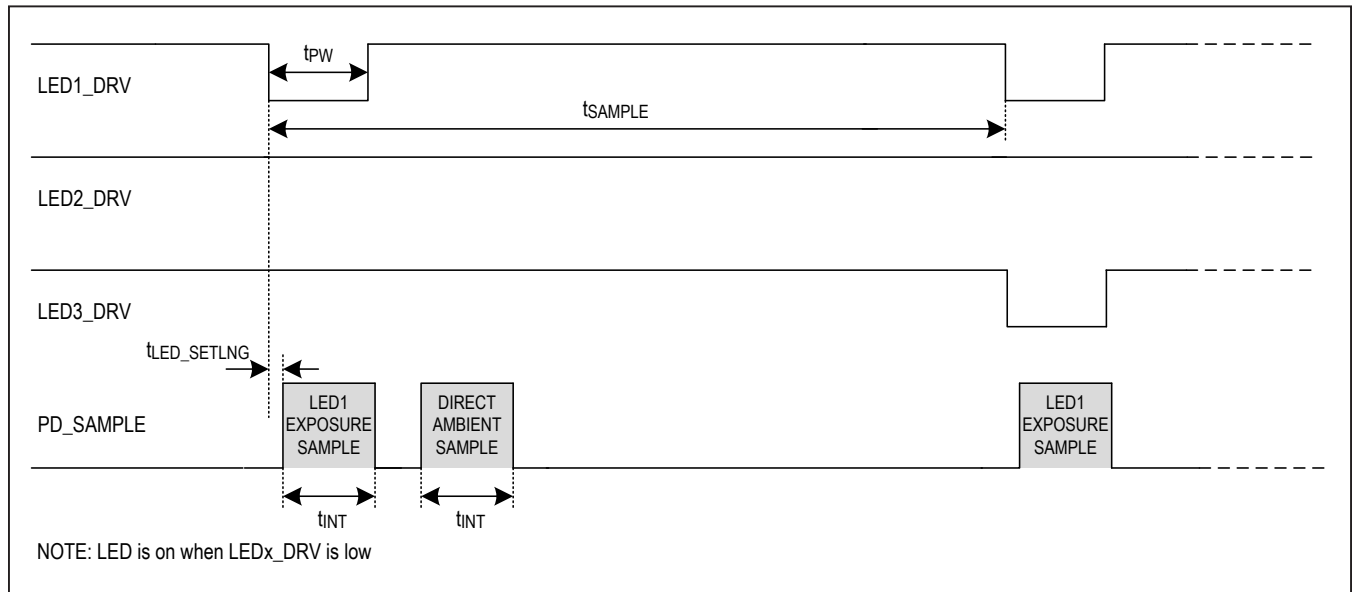


Figure 2. Timing for LED1 Pulsing with Direct Ambient Sampling

**All LED Pulsing Simultaneously with Direct Ambient Sampling**

The optical timing diagram below represents all three LEDs pulsing simultaneously with direct ambient sampling enabled. This timing mode would be used when heart rate is being measured with three green LEDs. In this

mode, a single optical sampled value, followed by the ambient sampled value, will appear in successive the FIFO locations. The direct ambient sampling is typically used to compensate the LED drive levels as the optical noise level can be elevated from ambient shot noise.

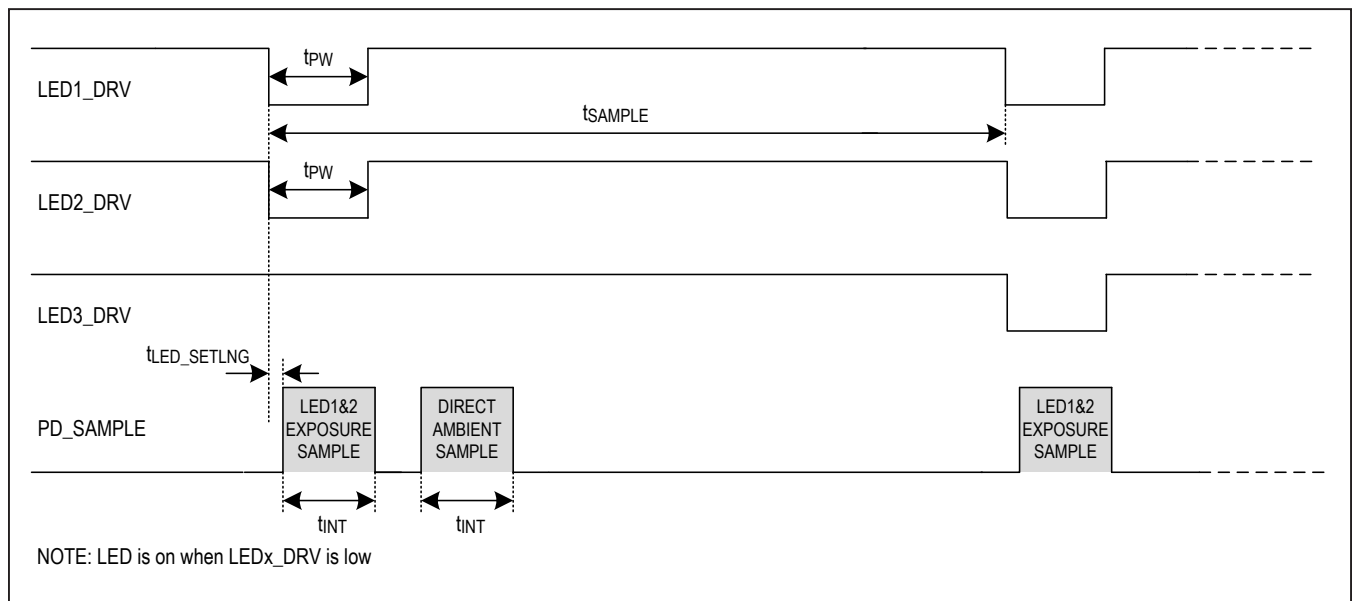


Figure 3. Timing for LED1 and LED2 Pulsing Simultaneously with Direct Ambient Sampling

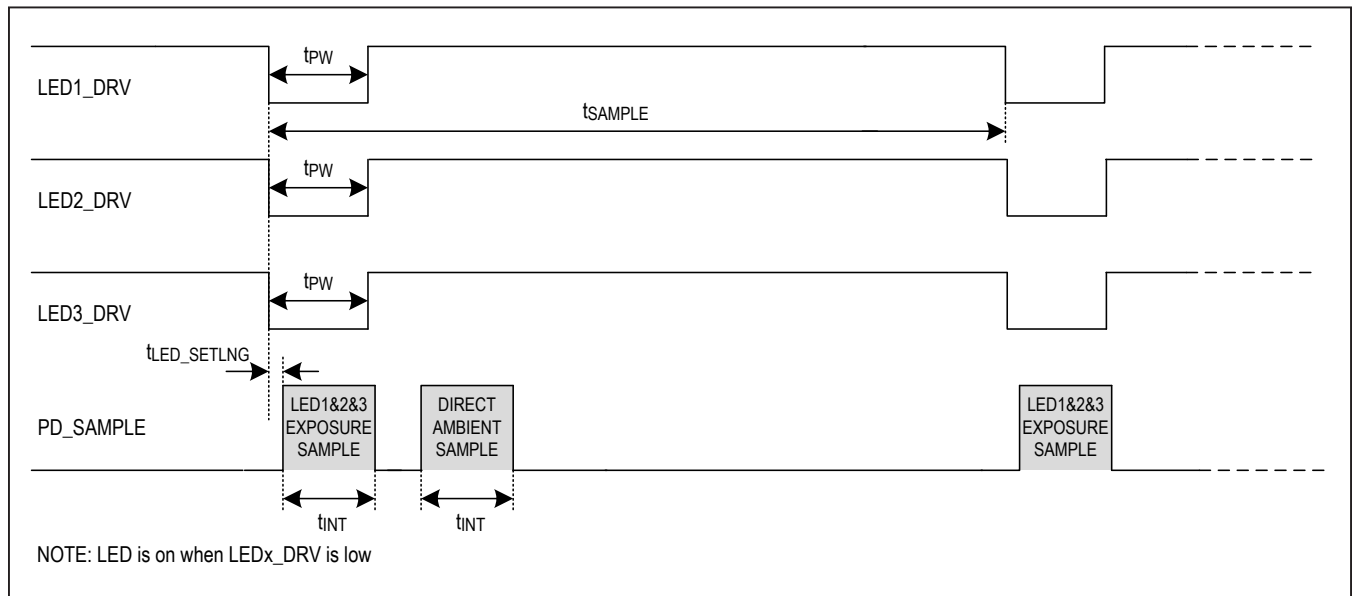


Figure 4. Timing for LED1, LED2, and LED3 Pulsing Simultaneously with Direct Ambient Sampling

**Two LEDs Pulse Sequentially with Direct Ambient Sampling**

The timing diagram below illustrates the optical timing when both LED1 and LED2 are enabled to pulse sequentially and direct ambient sampling is also enabled. This timing mode would be used when SpO<sub>2</sub> is being measured with IR and red LEDs. The optical sampled value for each LED will appear successively, followed by the direct ambient sampled value in the FIFO. when SpO<sub>2</sub> is being measured with IR and red LEDs. The optical sampled value

for each LED will appear successively, followed by the direct ambient sampled value in the FIFO.

**All LEDs Pulse Sequentially with Direct Ambient Sampling**

The optical timing diagram below illustrates the three LEDs pulsing sequentially, followed by a direct ambient sample. This timing mode would be used when heart rate on a green LED is combined with and SpO<sub>2</sub> measurement using IR and red LEDs.

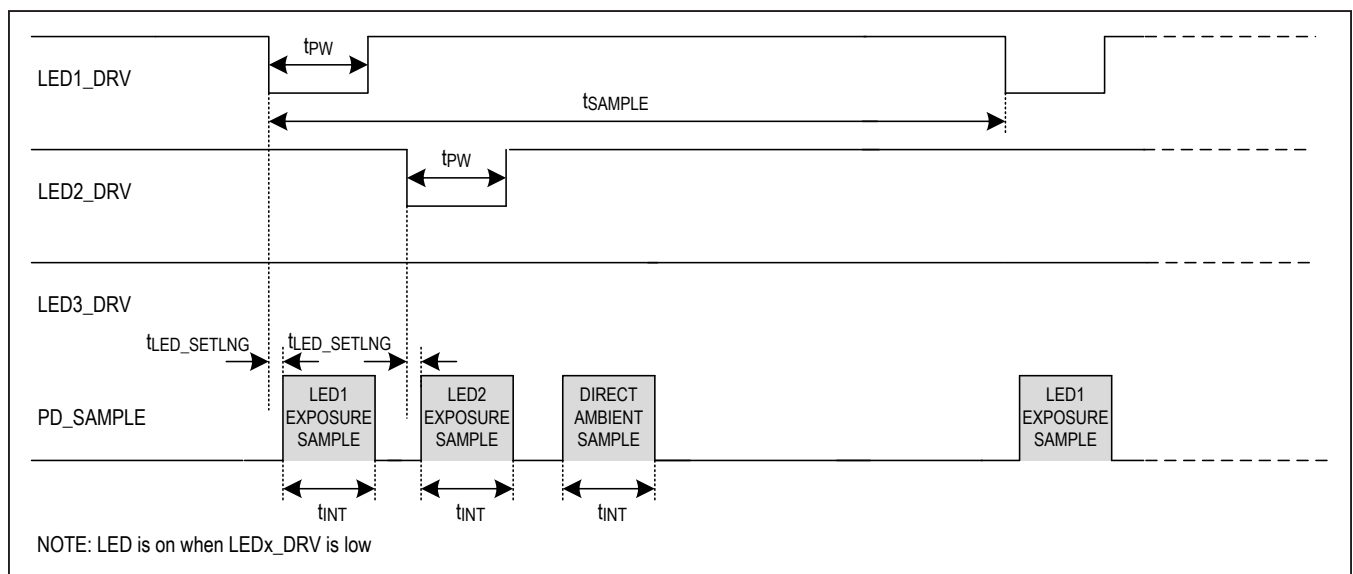


Figure 5. Timing for LED1 and LED2 Pulsing Sequentially with Direct Ambient Sampling

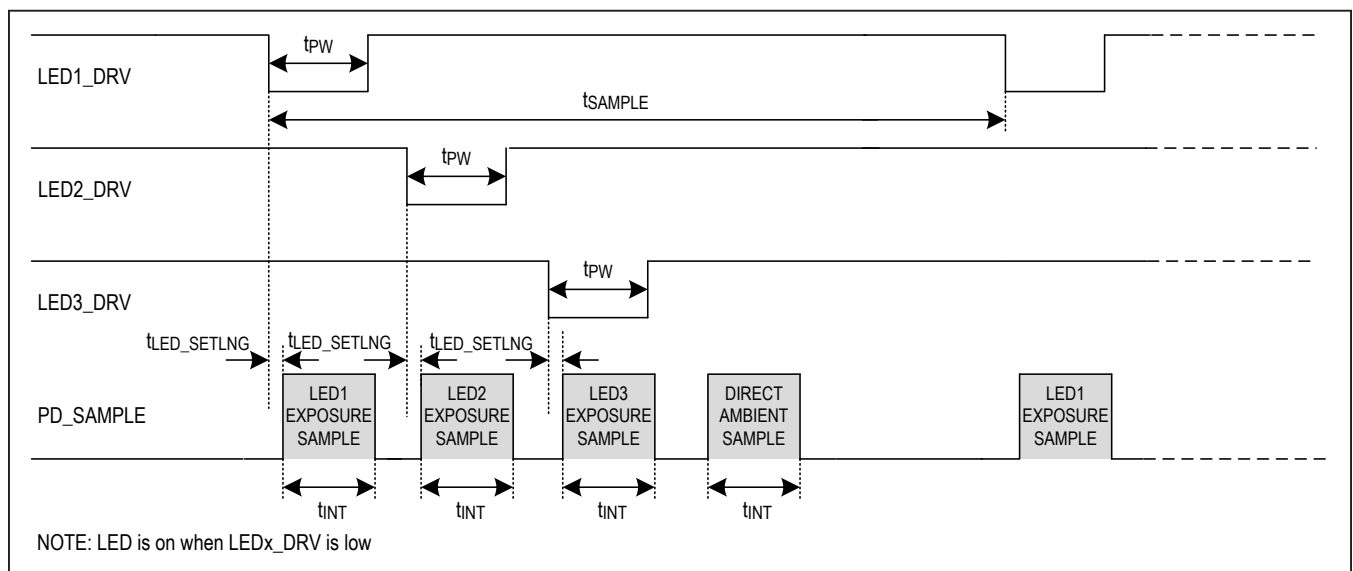


Figure 6. Timing for LED1, LED2, and LED3 Pulsing Sequentially with Direct Ambient Sampling

**GPIO Configuration**

The MAX86140/MAX86141 support several means by which they can synchronize to external sensors, muxes, and be extended to allow for more flexibility in the measurement configuration. This functionality is extended through the GPIO1 and GPIO2 pins and is selected by the GPIO CTRL bit field in the PPG SYNC Control register (0x10). The following describes option and the functional state of GPIO1 and GPIO2 as well as the part behavior.

**GPIO CTRL[3:0] 0000 and 0001: Stand Alone With and Without External Mux**

**Table 7. GPIO Mode 0000 and 0001**

GPIO CTRL	GPIO1 FUNCTION	GPIO2 FUNCTION	COMMENT
0000	Tristate or Mux Control	Disabled	GPIO1 will be active if any of the LEDCn[3:0] states A, B, or C are enabled in the exposure sequence. In this case, GPIO1 will be low during exposures on LED4, LED5 or LED6, otherwise it will be high. If LEDCn[3:0] state A, B, or C is not enabled in the exposure sequence, GPIO1 will be tristate unless externally pulled up. GPIO2 is disabled. Sample and exposure timing is controlled by the internal 32768Hz oscillator.
0001	Tristate or Mux Control	Input 32768Hz or 32000Hz Clock Input	GPIO1 will be active if any of the LEDCn[3:0] states A, B, or C are enabled in the exposure sequence. In this case, GPIO1 will be low during exposures on LED4, LED5 or LED6; otherwise, it will be high. If LEDCn[3:0] state A, B, or C is not enabled in the exposure sequence, GPIO1 will be tristate unless externally pulled up. GPIO2 is an input 32768/32000Hz. Sample and exposure timing is controlled by GPIO2 clock input.

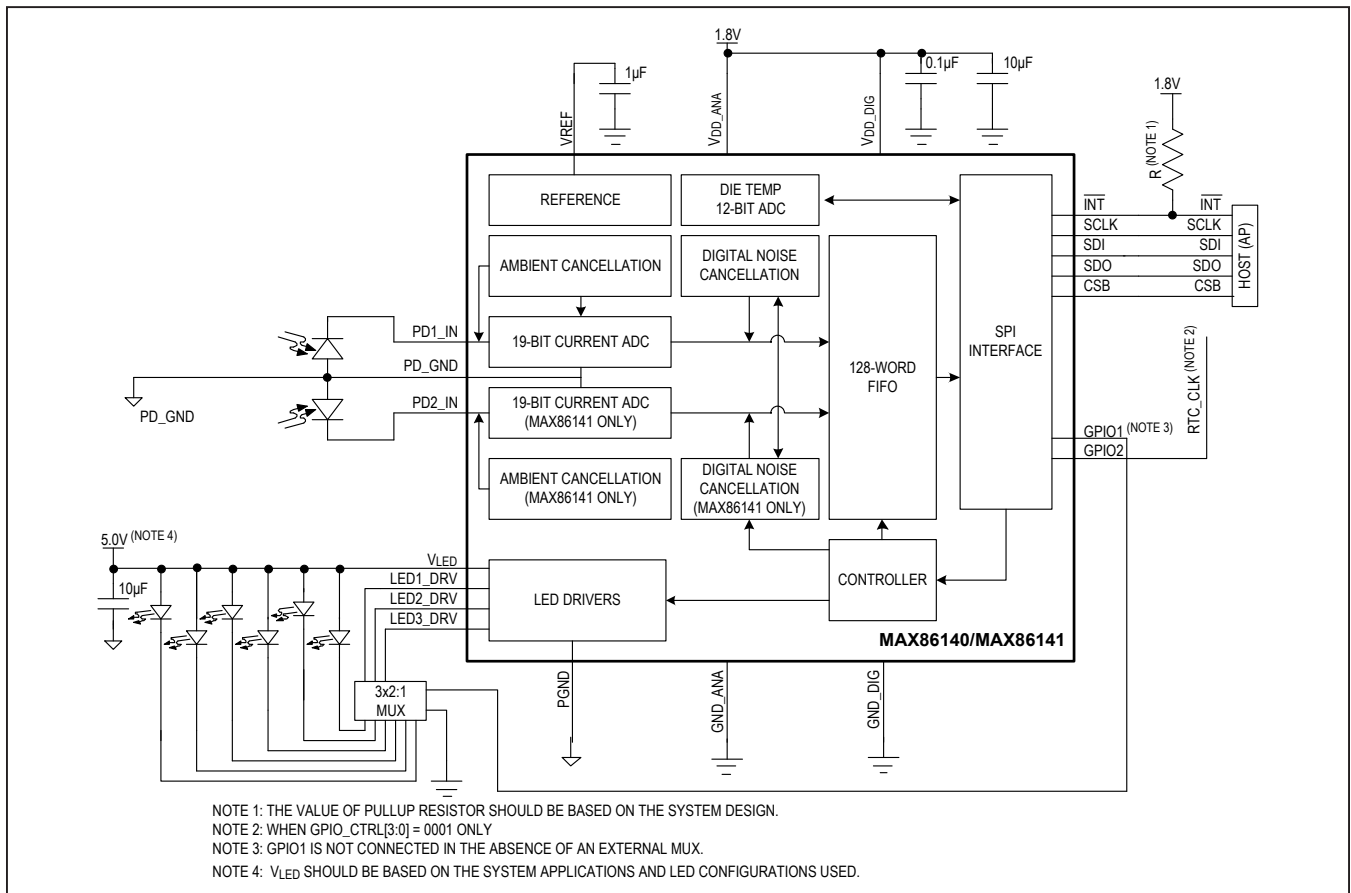


Figure 7. Block Diagram for GPIO CTRL[3:0] 0000 and 0001 Without External Mux

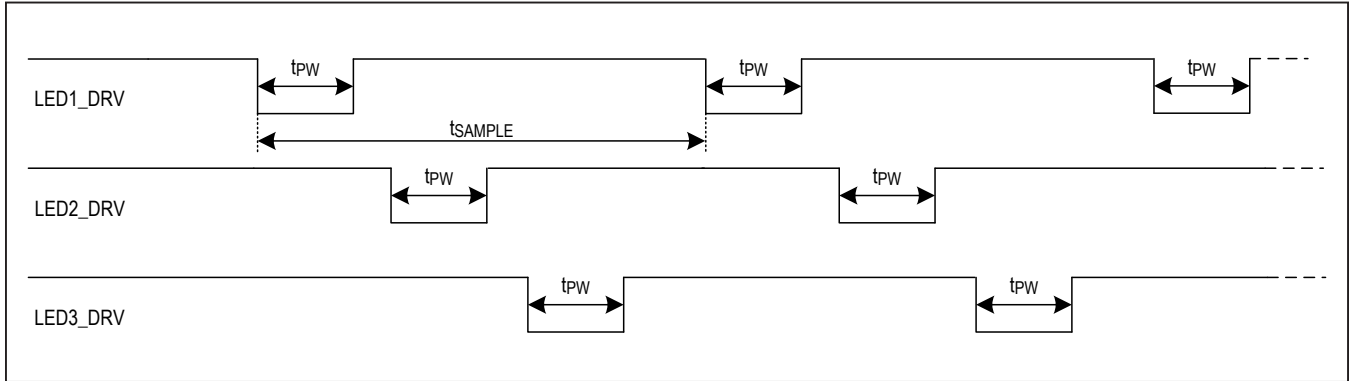


Figure 8. Timing Diagram for GPIO CTRL[3:0] 0000 and 0001 Without External Mux

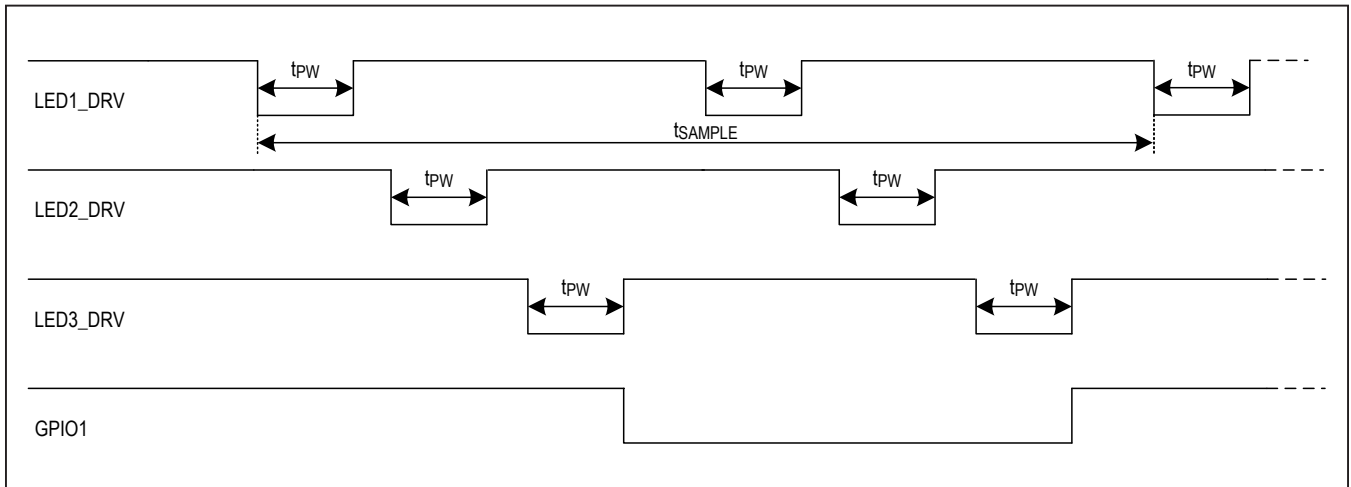


Figure 9. Timing Diagram for GPIO CTRL[3:0] 0000 and 0001 with External Mux

GPIO CTRL[3:0] 0010: Start of Sample Input with and without External Mux

Table 8. GPIO Mode 0010

GPIO CTRL	GPIO1 FUNCTION	GPIO2 FUNCTION	COMMENT
0010	Input Sample Trigger	Tristate or Mux Control	GPIO1 is defined as a sample trigger input (Slave). This input can come from an external source or from another MAX86140/ MAX86141 in master sample mode. GPIO2 will be active if any of the LEDCn[3:0] states A, B, or C are enabled in the exposure sequence. In this case, GPIO2 will be low during exposures on LED4, LED5 or LED6, otherwise it will be high. If LEDCn[3:0] state A, B, or C is not enabled in the exposure sequence, GPIO2 will be tristate unless externally pulled up. Exposure timing is controlled by internal oscillator.

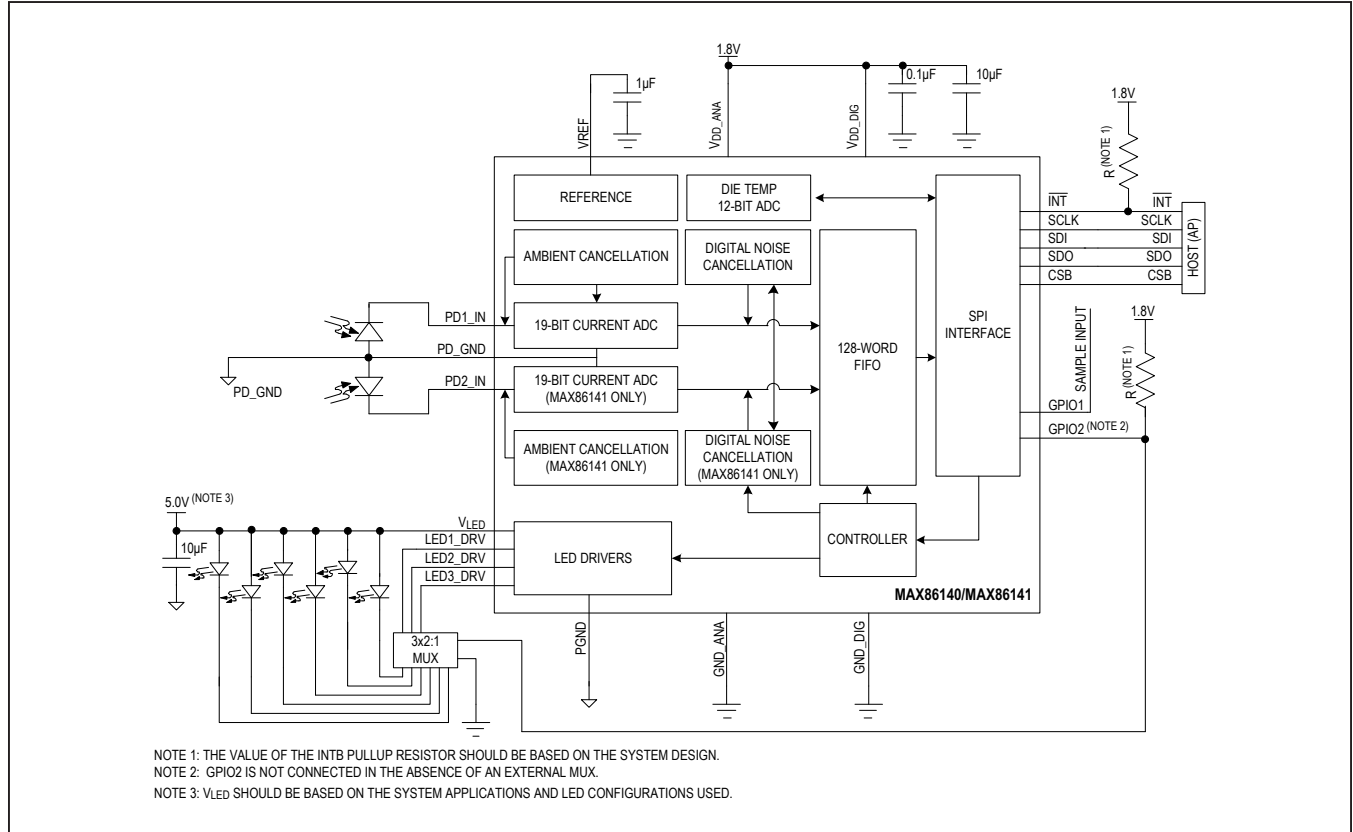


Figure 10. Block Diagram for GPIO CTRL[3:0] 0010 Without External Mux

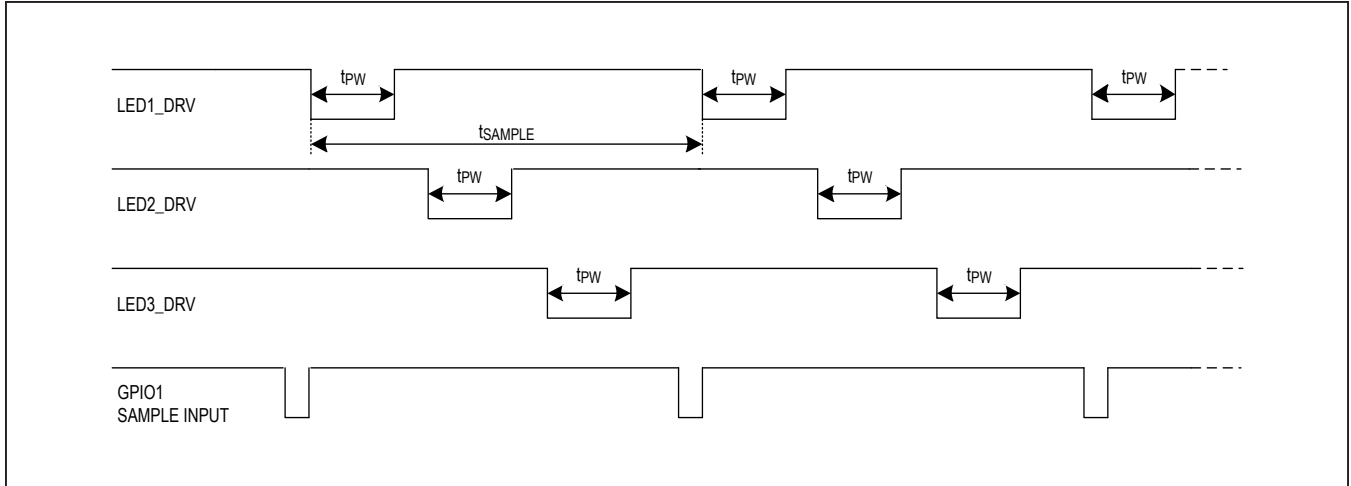


Figure 11. Timing Diagram for GPIO CTRL[3:0] 0010 Without External Mux

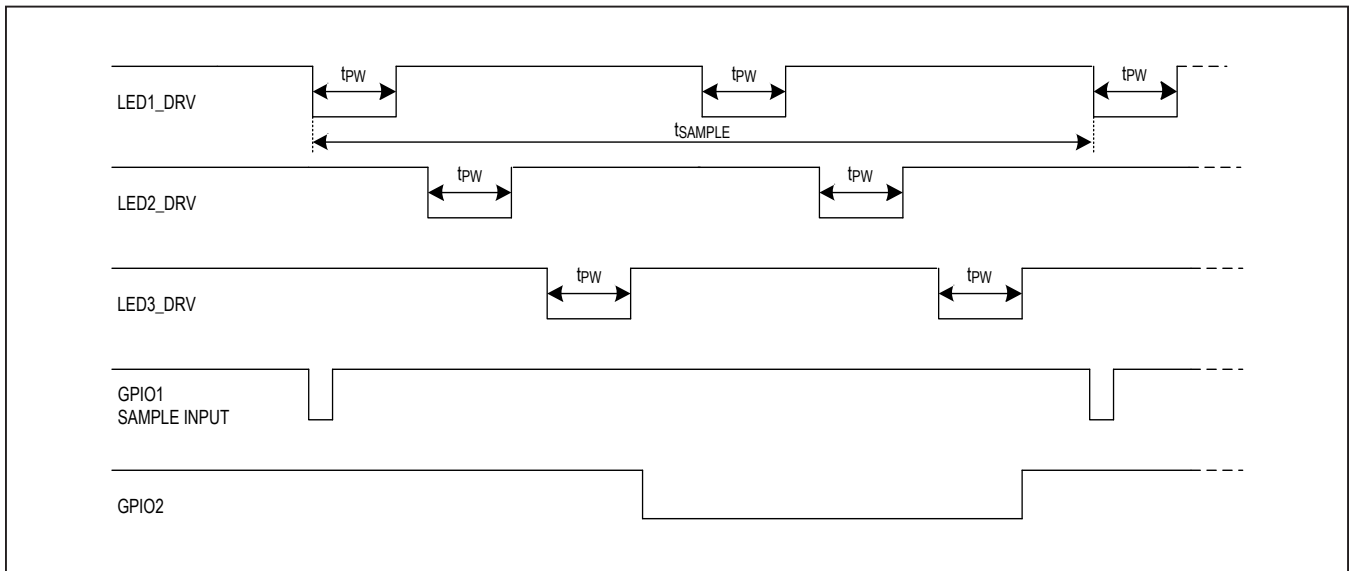


Figure 12. Timing Diagram for GPIO CTRL[3:0] 0010 with External Mux

**GPIO CTRL[3:0] 0011: Start of Sample Input with External Clock**

**Table 9. GPIO Mode 0011**

GPIO CTRL	GPIO1 FUNCTION	GPIO2 FUNCTION	COMMENT
0011	Input Sample Trigger	Input 32768Hz or 32000Hz Clock Input	GPIO1 is defined as a sample trigger input (Slave). This input can come from an external source or from another MAX86140/ MAX86141 in master sample mode. GPIO2 is an input 32768/32000Hz clock input. Exposure timing is controlled by GPIO2 clock input.

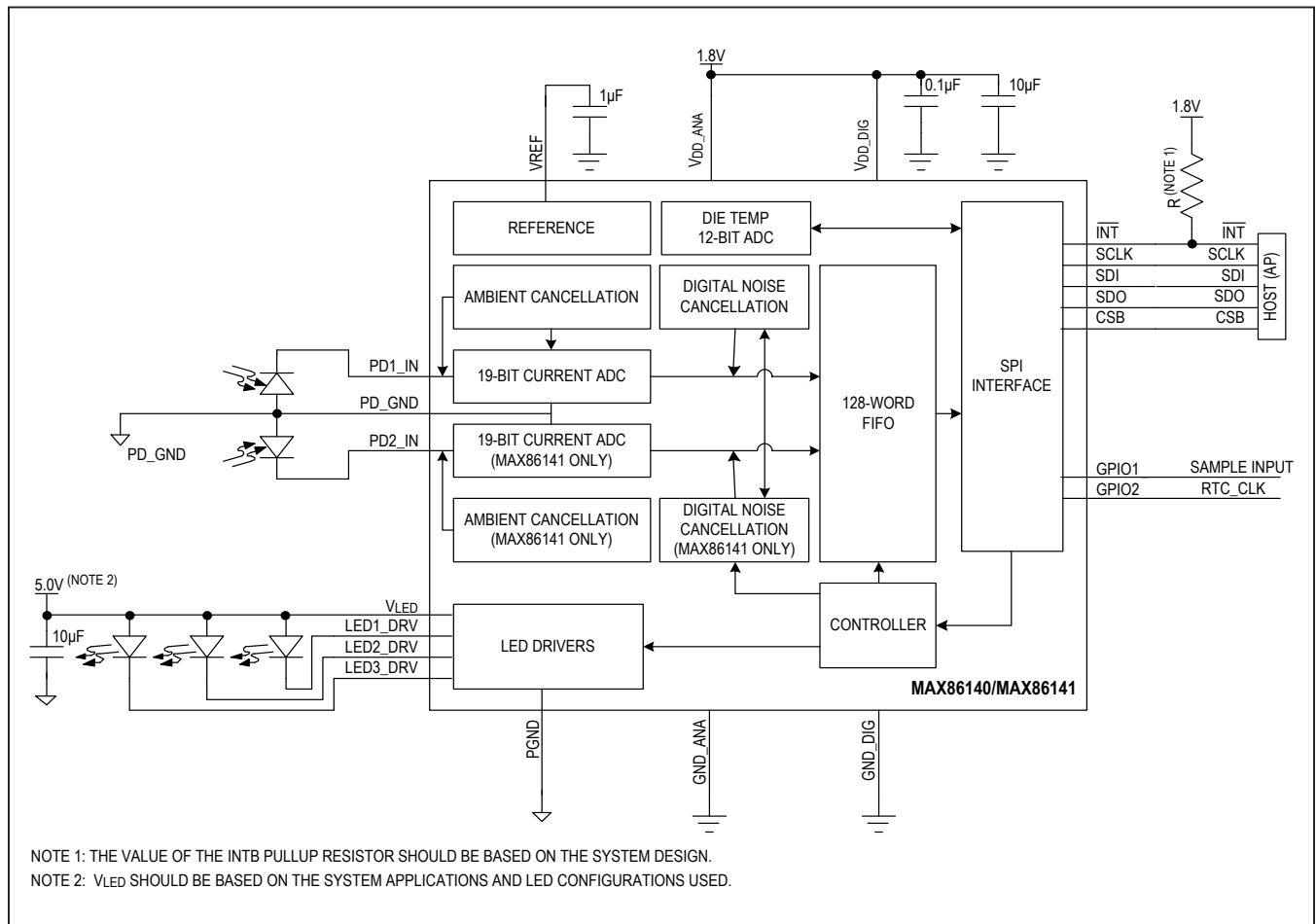


Figure 13. Block Diagram for GPIO CTRL[3:0] 0011



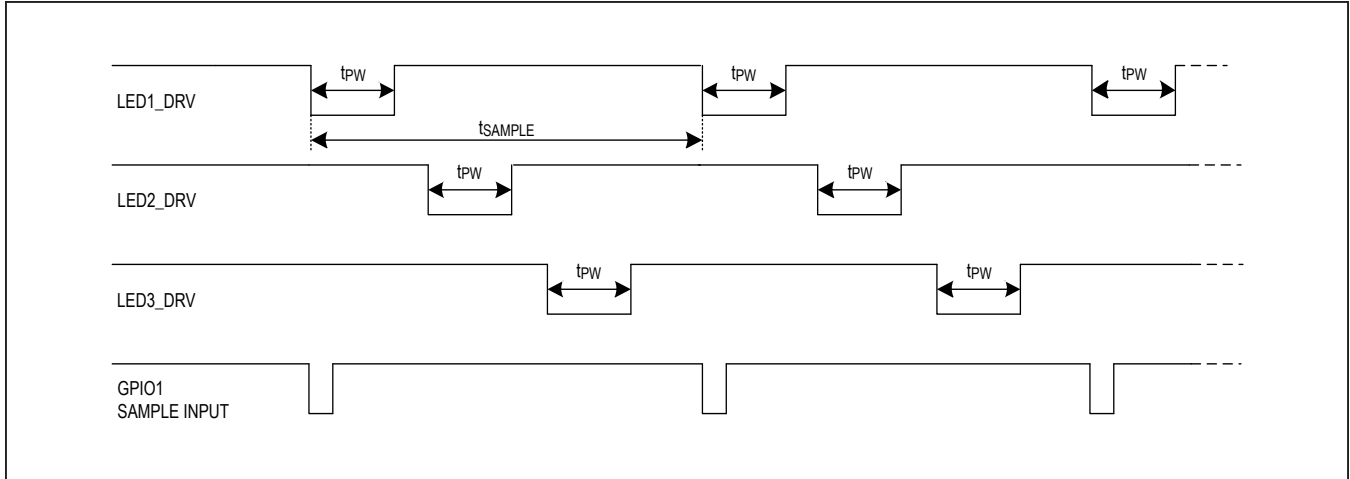


Figure 14. Timing Diagram for GPIO CTRL[3:0] 0011

**GPIO CTRL[3:0] 0100: Start of Sample Output With and Without External Mux**

**Table 10. GPIO Mode 0100**

GPIO CTRL	GPIO1 FUNCTION	GPIO2 FUNCTION	COMMENT
0100	Active Output Master Sample Output	Tristate or Mux Control	GPIO1 is defined as a master sample output. The GPIO1 output can be used to trigger a second sensor. When used with a second MAX86140/MAX86141 set to slave sample mode, the master sample timing will drive slave sample time. GPIO2 will be active if any of the LEDCn[3:0] states A, B, or C are enabled in the exposure sequence. In this case, GPIO2 will be low during exposures on LED4, LED5 or LED6, otherwise it will be high. If LEDCn[3:0] state A, B, or C is not enabled in the exposure sequence, GPIO2 will be tristate unless externally pulled up. Sample and exposure timing is controlled by internal oscillator.

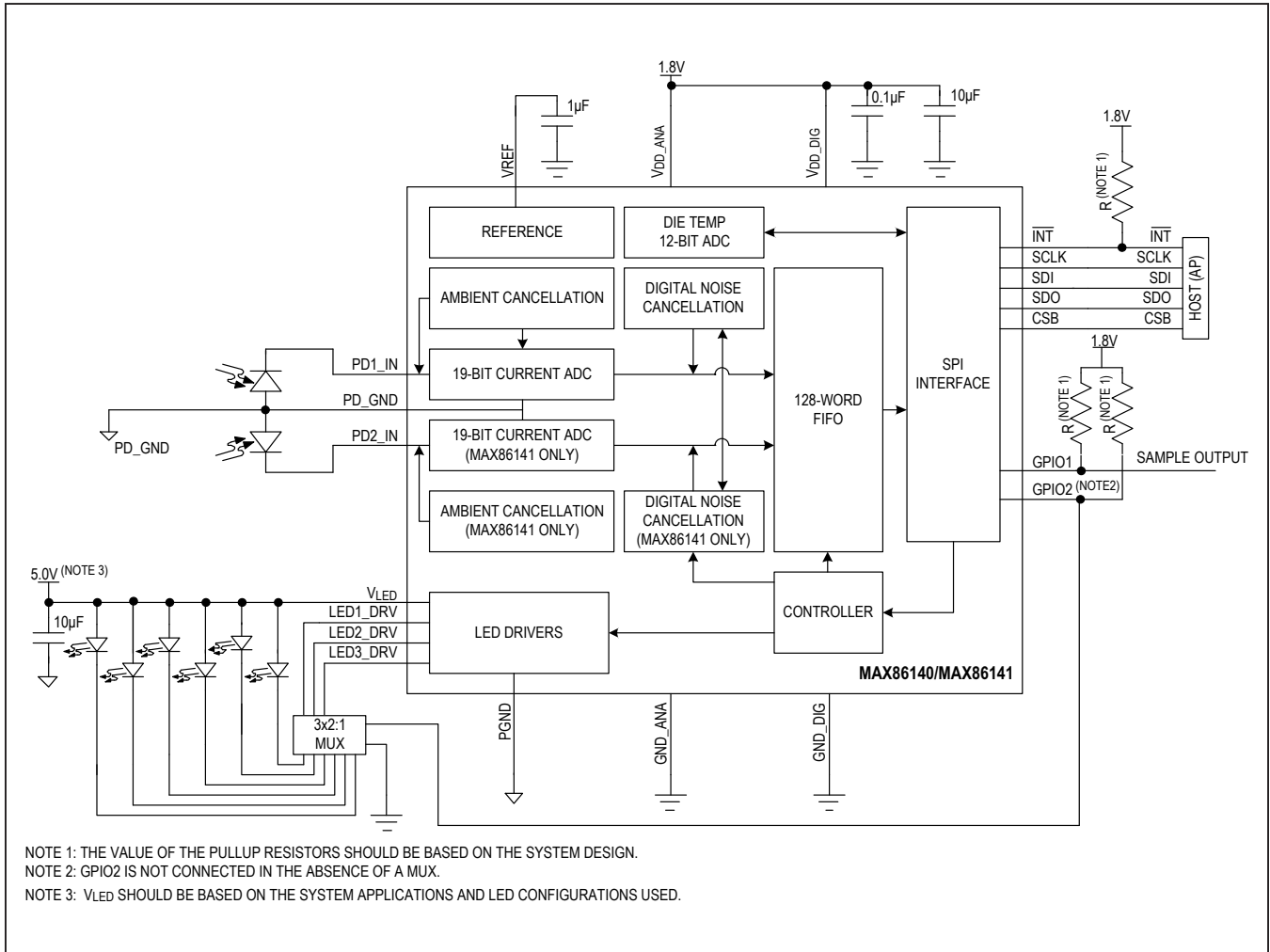


Figure 15. Block Diagram for GPIO CTRL[3:0] 0100 Without External Mux

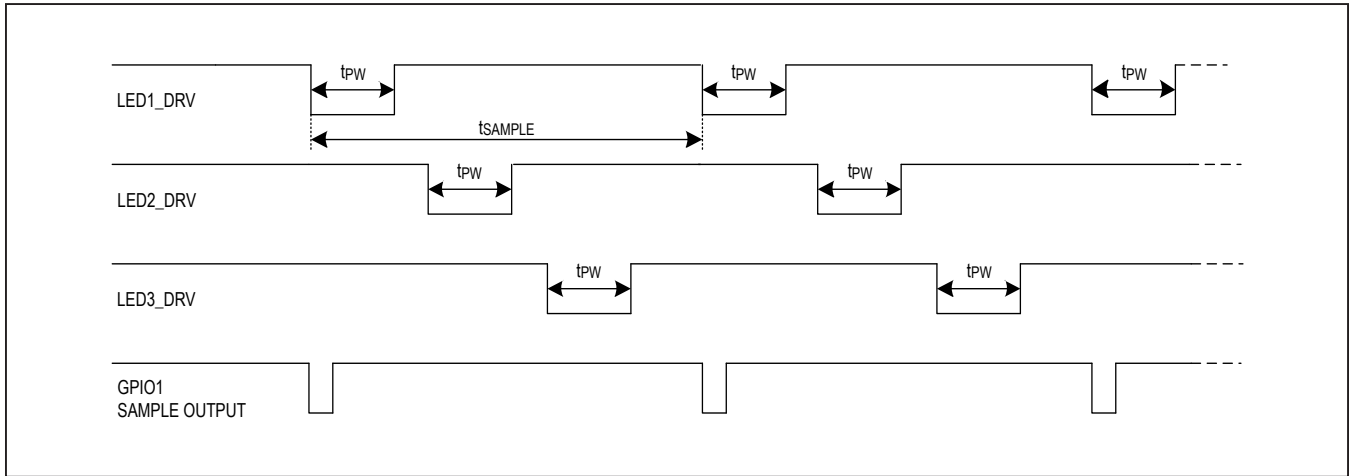


Figure 16. Timing Diagram for GPIO CTRL[3:0] 0100 Without External Mux

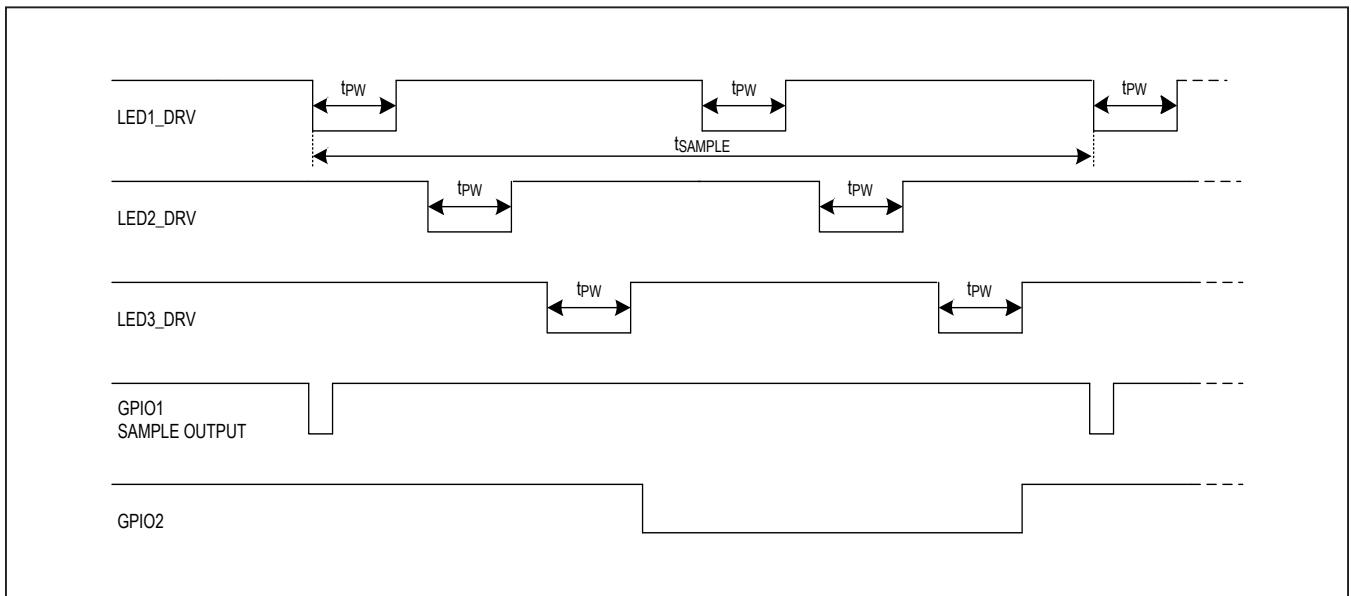


Figure 17. Timing Diagram for GPIO CTRL[3:0] 0100 with External Mux

**GPIO CTRL[3:0] 0101: Start of Sample Output with RTC Input Clock**

**Table 11. GPIO Mode 0101**

GPIO CTRL	GPIO1 FUNCTION	GPIO2 FUNCTION	COMMENT
0101	Active Output Master Sample Output	Input 32768/32000Hz Clock Input	GPIO1 is defined as a master sample output. The GPIO1 output can be used to trigger a second sensor. When used with a second MAX86140/MAX86141 set to slave sample mode, the master sample timing will drive slave sample time. GPIO2 is an input 32768/32000Hz. Exposure timing is controlled by GPIO2 clock input.

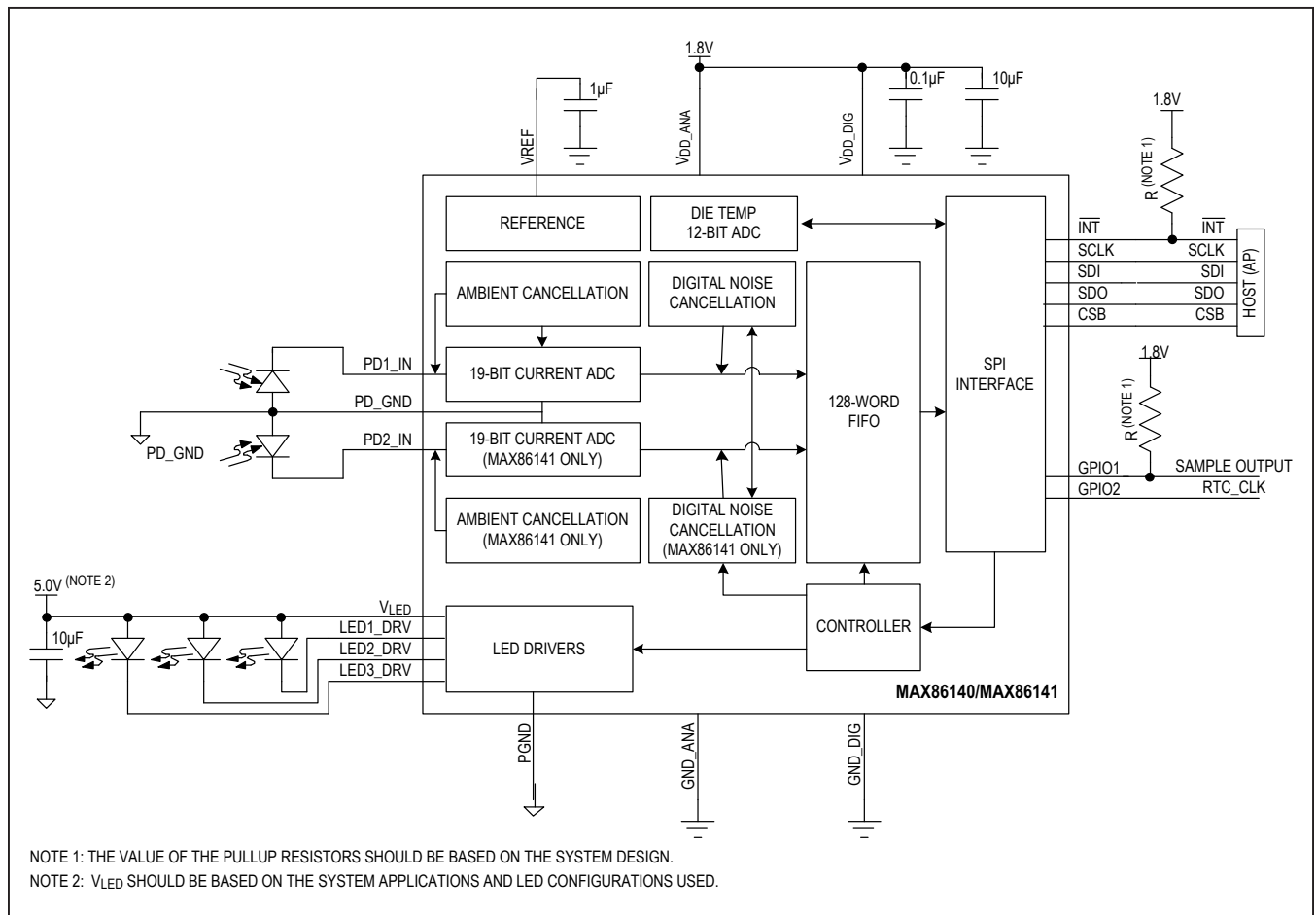


Figure 18. Block Diagram for GPIO CTRL[3:0] 0101

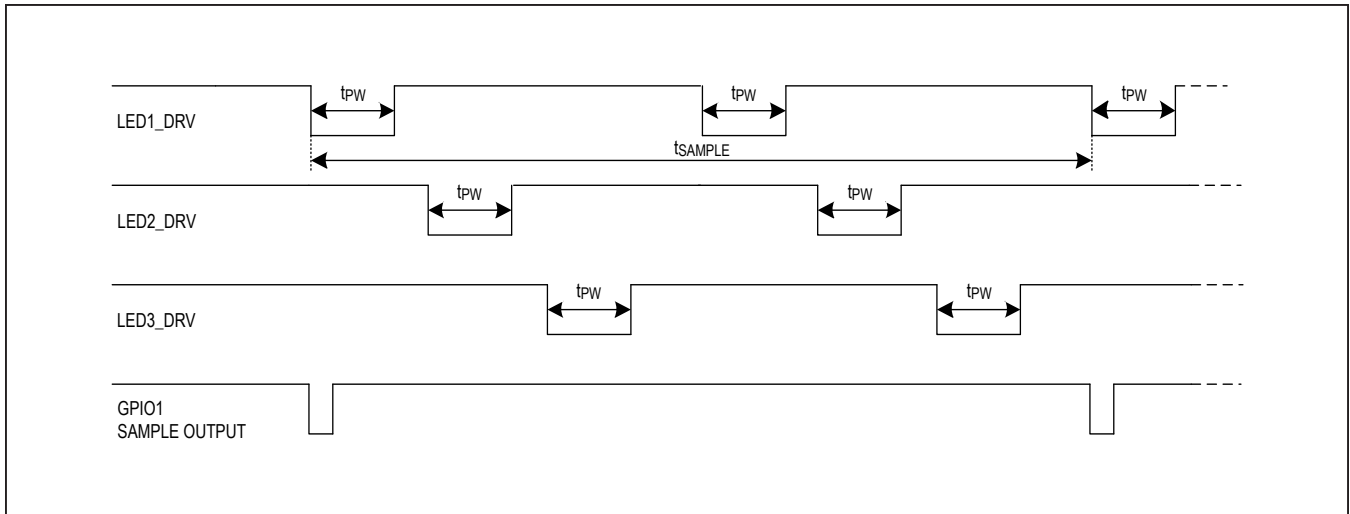


Figure 19. Timing Diagram for GPIO CTRL[3:0] 0101

**GPIO CTRL[3:0] 0110 and 0111: Master/Slave with External Mux**

**Table 12. GPIO Mode 0110 and 0111**

GPIO CTRL	GPIO1 FUNCTION	GPIO2 FUNCTION	COMMENT
0110	Input Exposure Trigger	Tristate or Mux Control	GPIO1 is defined as an exposure trigger input (Slave). This input can come from an external source or from another MAX86140 in master sample mode. Both sample and exposure timing is controlled by the GPIO1 input. GPIO2 will be active if any of the LEDCn[3:0] states A, B, or C are enabled in the exposure sequence. In this case, GPIO2 will be high during exposures on LED4, LED5 or LED6, otherwise it will be low. If LEDCn[3:0] state A, B, or C is not enabled in the exposure sequence, GPIO2 will be tristate.
0111	Active Output Master Exposure Output	Tristate or Mux Control	GPIO1 is defined as a master sample output. The GPIO1 output can be used to trigger and second sensor. When used with a second MAX86140 set to slave exposure mode, the master exposure timing will drive slave exposure time. GPIO2 will be active if any of the LEDCn[3:0] states A, B, or C are enabled in the exposure sequence. In this case, GPIO2 will be high during exposures on LED4, LED5 or LED6, otherwise it will be low. If LEDCn[3:0] state A, B, or C is not enabled in the exposure sequence, GPIO2 will be tristate. Sample and exposure time is controlled internally

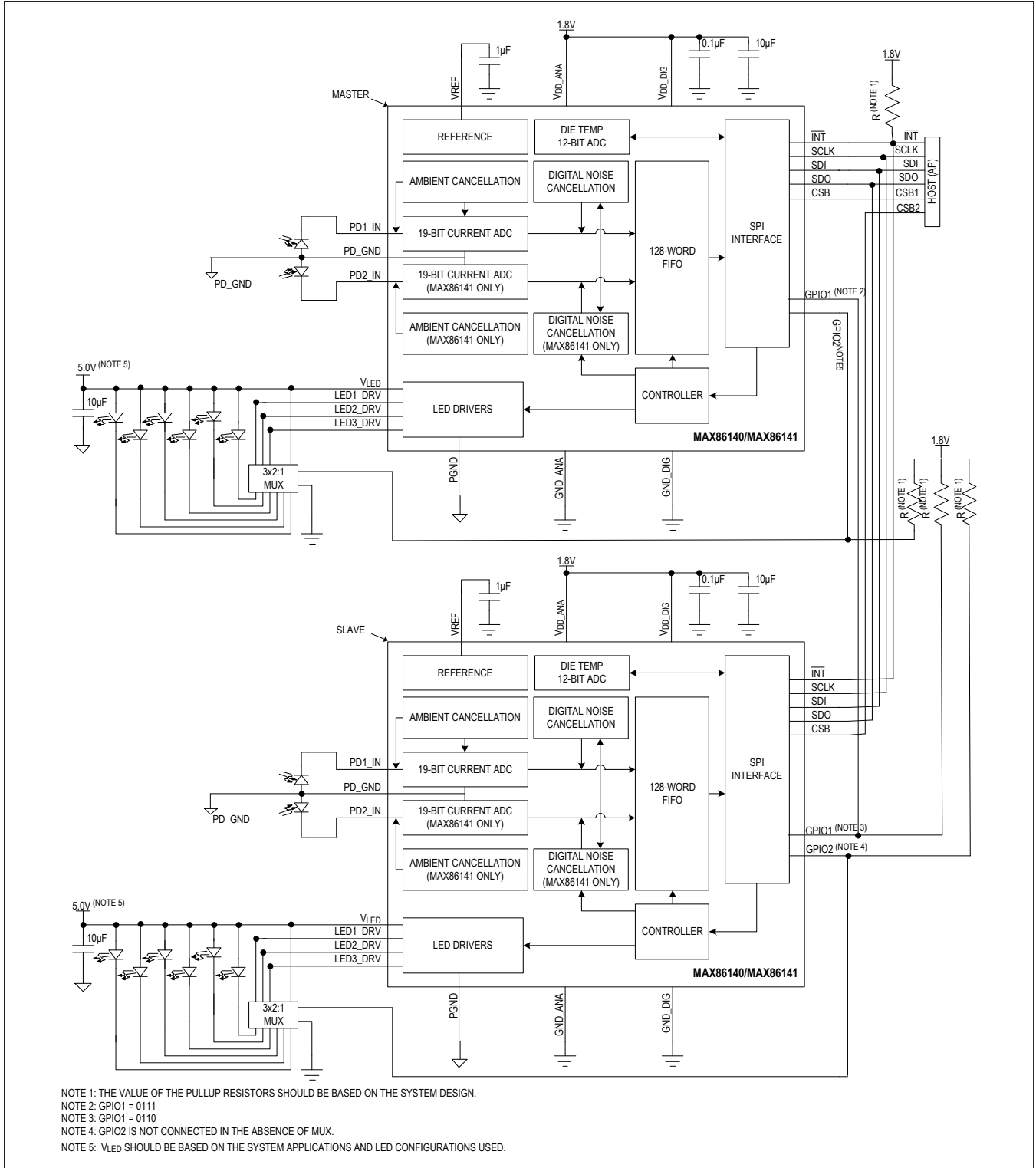


Figure 20. Block Diagram for GPIO CTRL[3:0] 0110 and 0111 With Two External Muxes

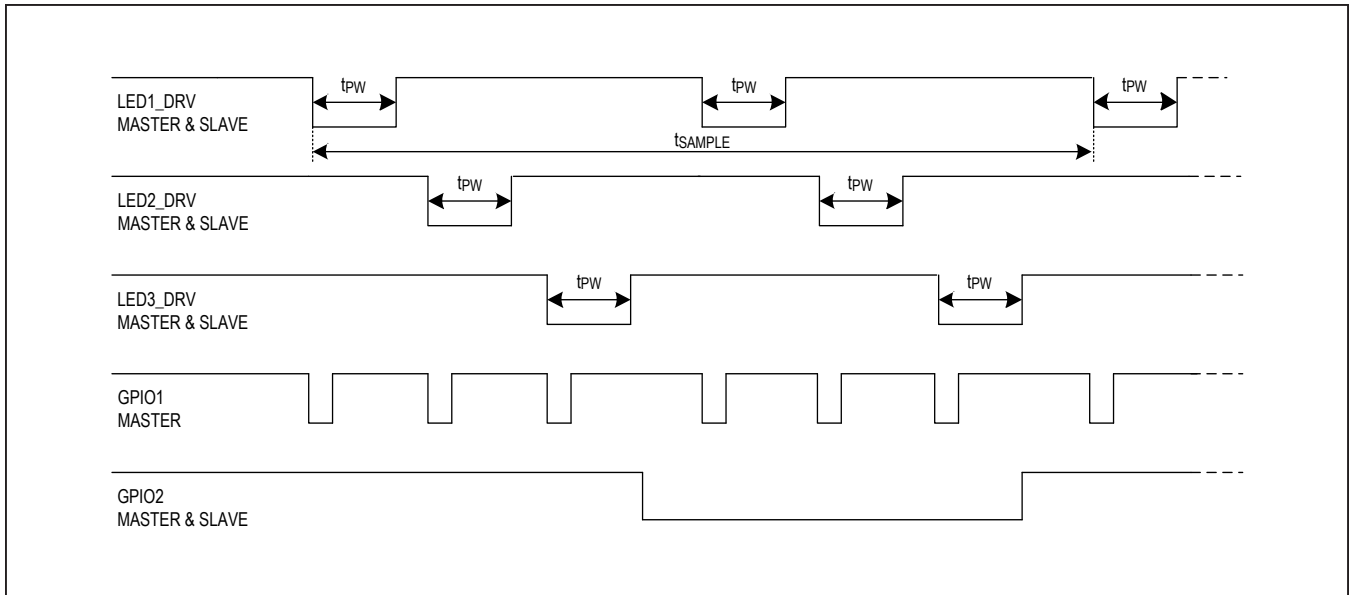


Figure 21. Timing Diagram for GPIO CTRL[3:0] 0110 and 0111 With External Mux

**GPIO CTRL[3:0] 0110 and 1000: Master/Slave with and without External Mux**

**Table 13. GPIO Mode 0110 and 1000**

GPIO CTRL	GPIO1 FUNCTION	GPIO2 FUNCTION	COMMENT
0110	Input Exposure Trigger	Tristate or Mux Control	GPIO1 is defined as an exposure trigger input (Slave). This input can come from an external source or from another MAX86140/MAX86141 in master sample mode. Both sample and exposure timing is controlled by the GPIO1 input. GPIO2 will be active if any of the LEDCn[3:0] states A, B, or C are enabled in the exposure sequence. In this case, GPIO2 will be low during exposures on LED4, LED5, or LED6, otherwise it will be high. If LEDCn[3:0] state A, B, or C is not enabled in the exposure sequence, GPIO2 will be tristate unless externally pulled up.
1000	Active Output Master Exposure Output	Input 32768/32000Hz Clock Input	GPIO1 is defined as a master sample output. The GPIO1 output can be used to trigger and second sensor. When used with a second MAX86140 /MAX86141 set to slave exposure mode, the master exposure timing will drive slave exposure time. GPIO2 is an input 32768/32000Hz. Sample and exposure timing is controlled by GPIO2 clock input.

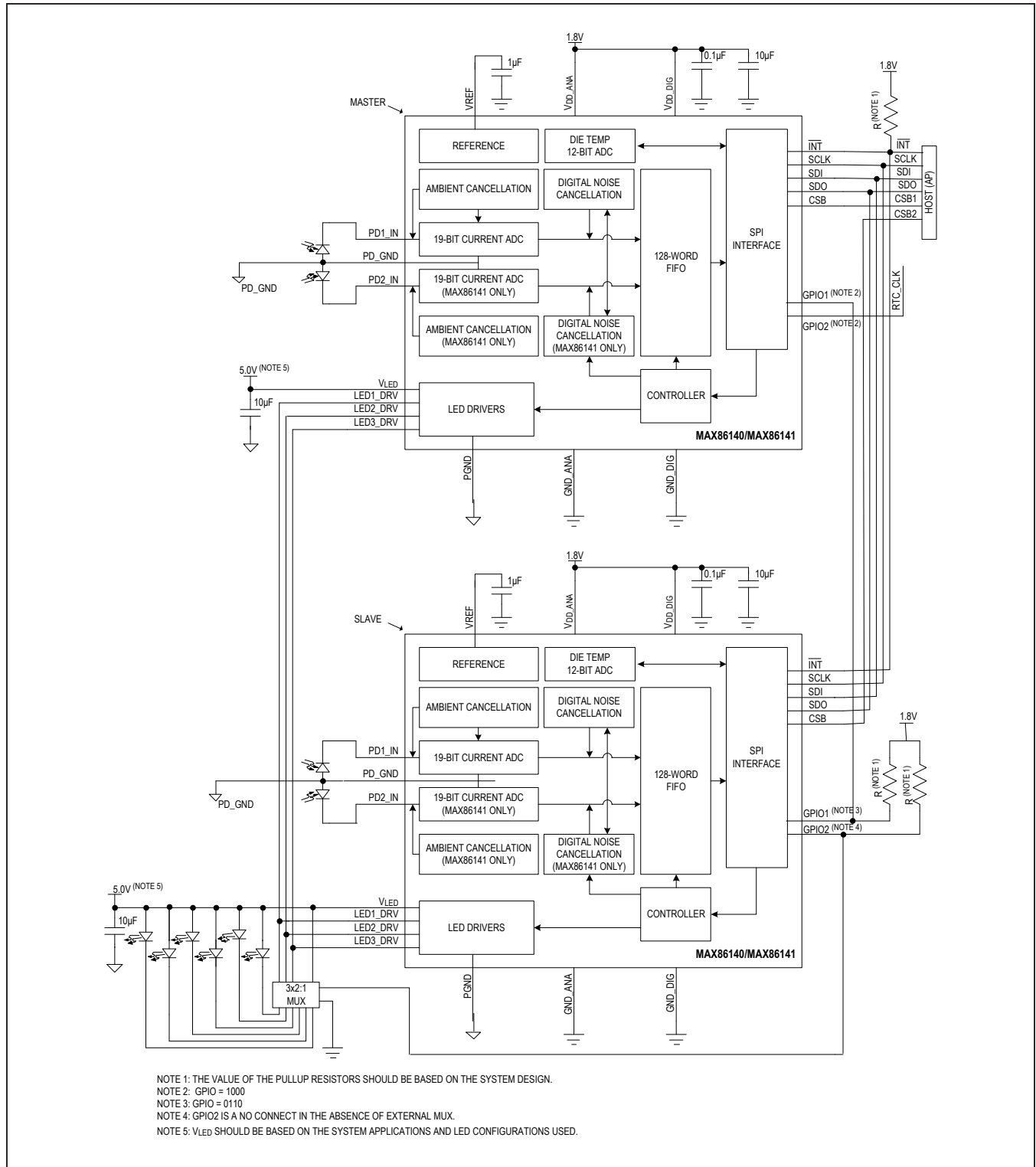


Figure 22. Block Diagram for GPIO CTRL[3:0] 0110 and 1000 Without External Mux



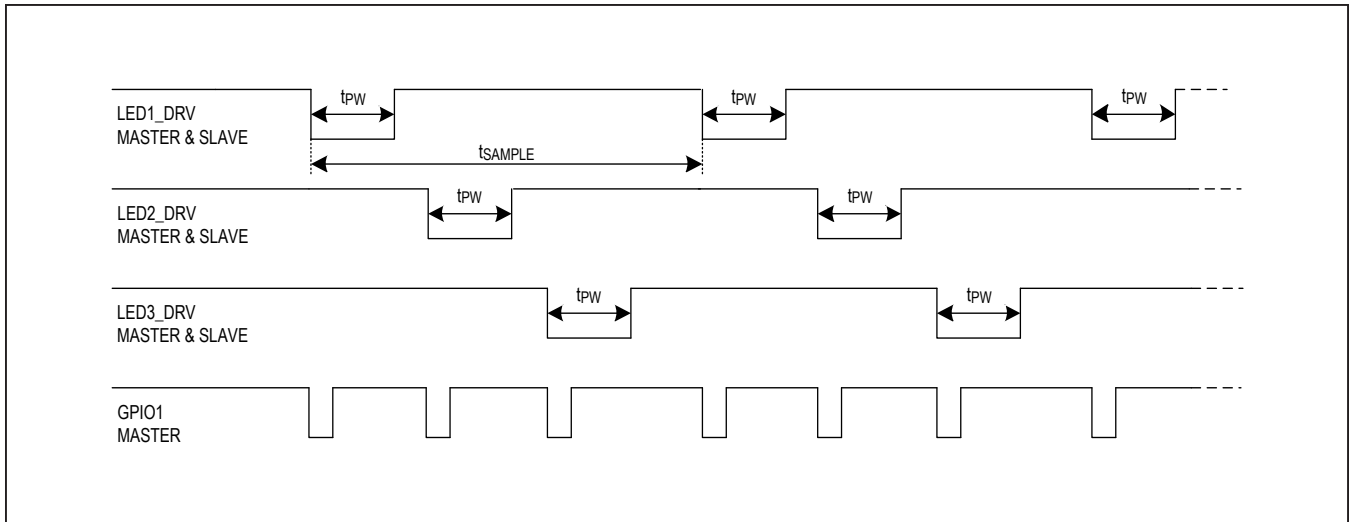


Figure 23. Timing Diagram for GPIO CTRL[3:0] 0110 and 1000 Without External Mux

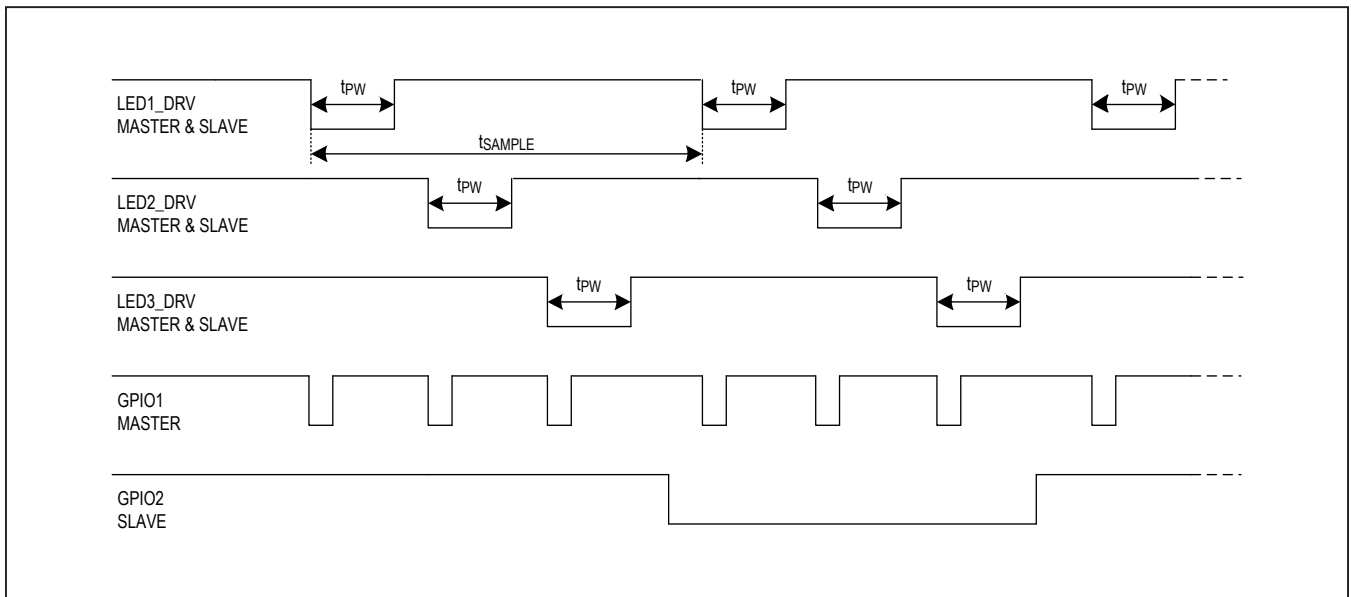


Figure 24. Timing Diagram for GPIO CTRL[3:0] 0110 and 1000 with External Mux

**GPIO CTRL[3:0] 1001 Hardware Sync**

**Table 14. GPIO Mode 1001**

GPIO CTRL	GPIO1 FUNCTION	GPIO2 FUNCTION	COMMENT
1001	Input HW_FORCE_SYNC	Input 32768Hz or 32000Hz Clock Input	GPIO1 is defined as a start of sample sync input. The rising edge of GPIO1 causes the present sample sequence to be terminated and reinitiated on the next rising edge of GPIO2 input. GPIO2 is an input 32768/32000Hz. Sample and exposure timing is controlled by GPIO2 clock input.

**GPIO CTRL[3:0] 1010 Hardware Sync**

**Table 15. GPIO Mode 1010**

GPIO CTRL	GPIO1 FUNCTION	GPIO2 FUNCTION	COMMENT
1010	Input Sample Sync ONE_SHOT	Tristate or Mux Control	GPIO1 is defined as a start of the power-up sequence for one sample. The falling edge of GPIO1 starts the power-up sequence followed by the exposure sequence as programmed in the LEDCn[3:0] registers. After the sample data is pushed to the FIFO, the device waits for the next Sample Sync pulse on GPIO1. GPIO2 is active if any of the exposure sequence registers LEDCn[3:0] are set to value 0xA, 0xB, or 0xC. In this case, GPIO2 is low during exposures on LED4, LED5, or LED6, respectively. Otherwise, GPIO2 is high. If the exposure sequence registers LEDCn[3:0] are not state 0xA, 0xB, or 0xC, GPIO2 is tristate. Sample period and exposure time are controlled internally.

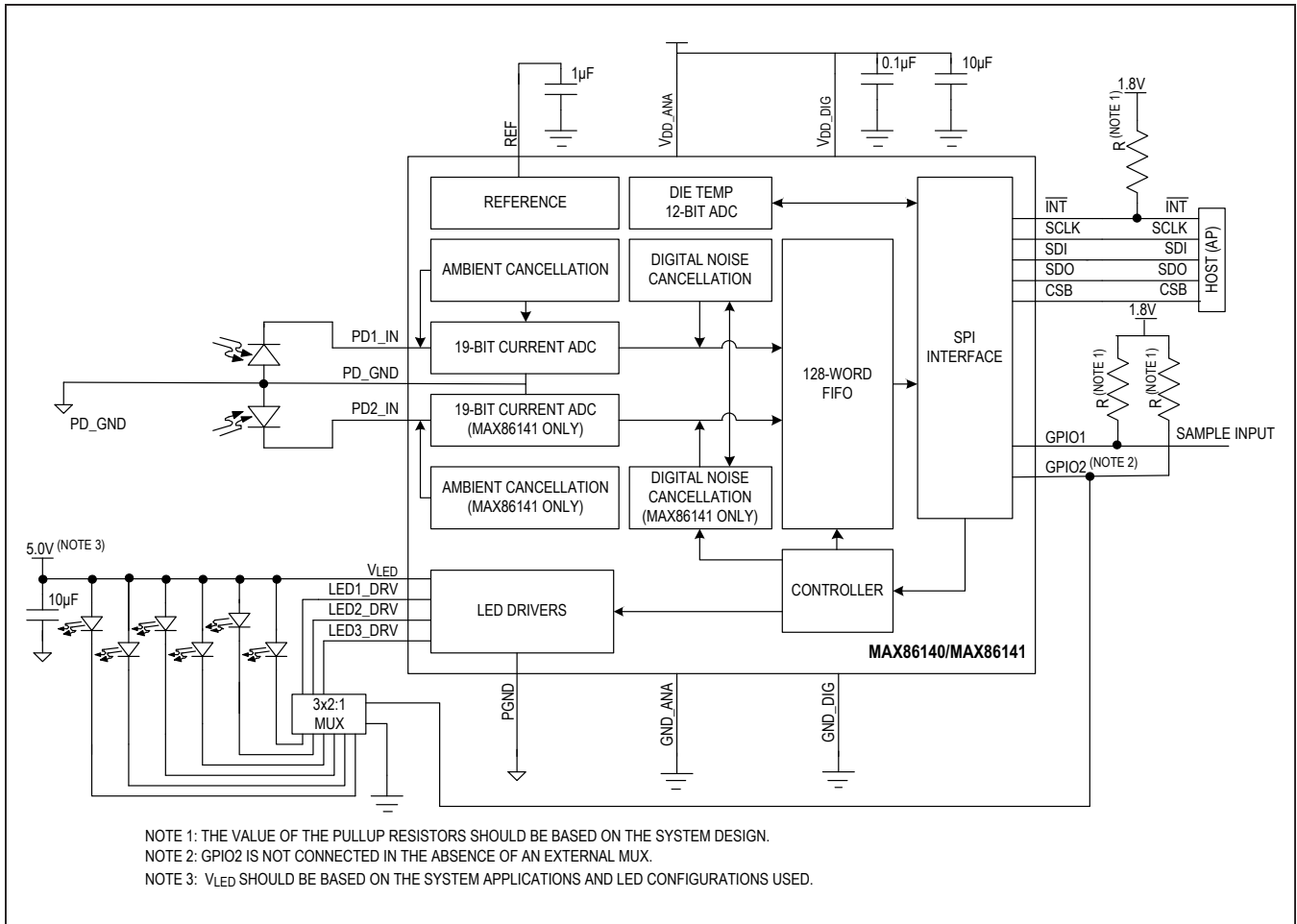


Figure 25. Block Diagram for GPIO CTRL[3:0] 1010

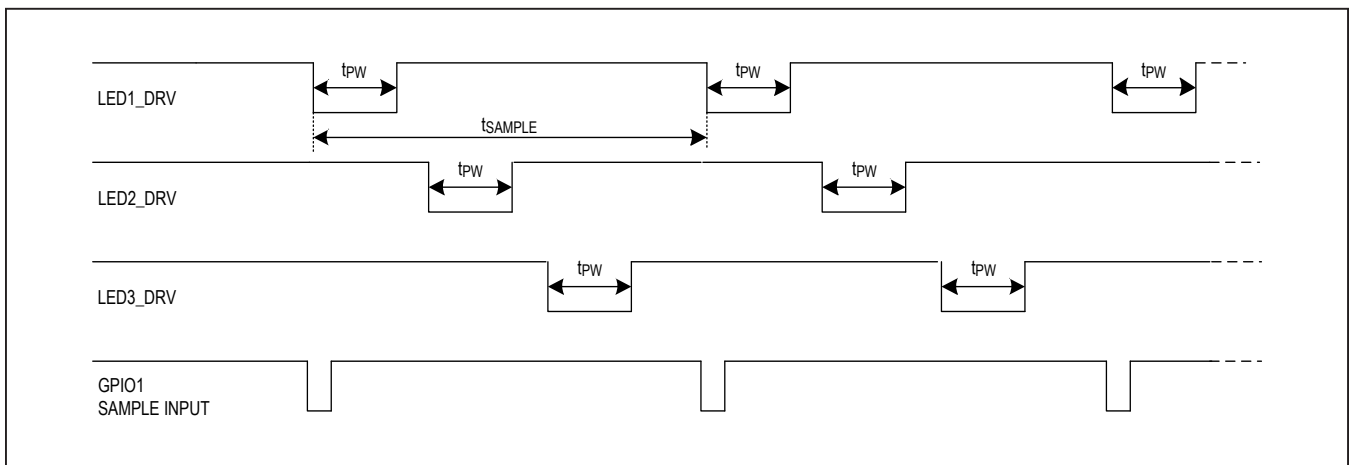


Figure 26. Timing Diagram for GPIO CTRL[3:0] 1010 Without External Mux

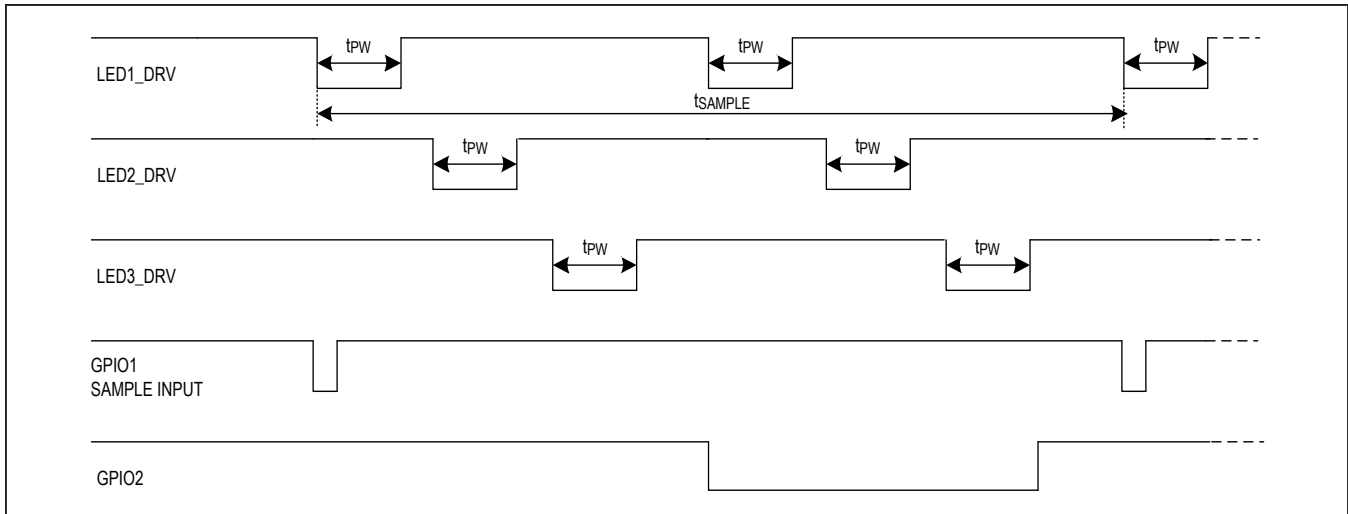


Figure 27. Timing Diagram for GPIO CTRL[3:0] 1010 With External Mux

### Proximity Mode Function

The MAX86140/MAX86141 includes an optical proximity function that could significantly reduce energy consumption and extend battery life when the sensor is not in contact with the skin. Proximity mode is enabled by setting PROX\_INT\_EN bit field to 1 in the Interrupt Enable 2 register (address 0x02[4]), setting a threshold in the PROX\_INT\_THRESH register (address 0x14) and assigning an LED current in the PILOT\_PA (address 0x29). Proximity mode also requires that LED Sequence Register 1, field LEDC1 (address [3:0]) to be assigned to a specific measurement and that measurement is correctly connected to a light source. The LEDC1 measurement is used to detect the optical presents of a reflecting object in proximity mode and thus must be valid for proximity mode to work.

When enabled, the Proximity Detect Interrupt (register 0x01[4]) will be asserted and proximity mode will be entered when the value of the measurement assigned to LEDC1 drops below the PROX\_INT\_THRESH. When entering proximity mode, the MAX86140/MAX86141 will drop the current to the LED(s) assigned to LEDC1 to PILOT\_PA value, reduce the sample rate to 8sps and operates in Low Power mode. The intent here is to both reduce the consumed LED current and MAX86140/MAX86141 power to a minimum during situations where there is no reflective returned signal. It is also intended to reduce the emitted light to a minimum or even below that perceivable by the human eye.

When the proximity mode is enabled and the measurement assigned to LEDC1 with the LED current in PILOT\_PA

exceeds the PROX\_INT\_THRESH, the MAX86140/MAX86141 will also generate a Proximity Detect Interrupt (register 0x01[4]). In such an event MAX86140/MAX86141 will switch to normal mode, changing the sample rate to that assigned in PPG Configuration 2 register (address 0x12) bit field PPG\_SR and the LED current assigned to the measurement of LEDC1. Therefore, the MAX86140 is able to switch to proximity mode and back to normal mode without microprocessor interaction.

The threshold applied to PROX\_INT\_THRESH should be well below that of a usable signal at the maximum LED current applied to LEDC1 but high enough to not be triggered by noise from distant objects. Further the current assigned to PILOT\_PA should be much lower than that assigned to LEDx\_PA in normal mode. This will ensure that the signal obtained from LEDC1 will drop significantly when entering proximity mode, thus providing enough hysteresis to eliminate multiple interrupts being generated at the proximity/normal mode transition.

To guarantee that MAX86140/MAX86141 will successfully transition from proximity mode to normal mode, the PROX\_INT\_THRESH should be low enough and the PILOT\_PA high enough to ensure that the device mounted on the darkest of skins will return a signal above the PROX\_INT\_THRESH at the PILOT\_PA current.

Note that proximity mode is only available to LEDC1 measurements that are made with PD1\_IN optical channel without an external mux. When proximity mode is active, LEDC2~LEDC6 will be ignored. The threshold applied to PROX\_INT\_THRESH register are in units of 2048LSBs.

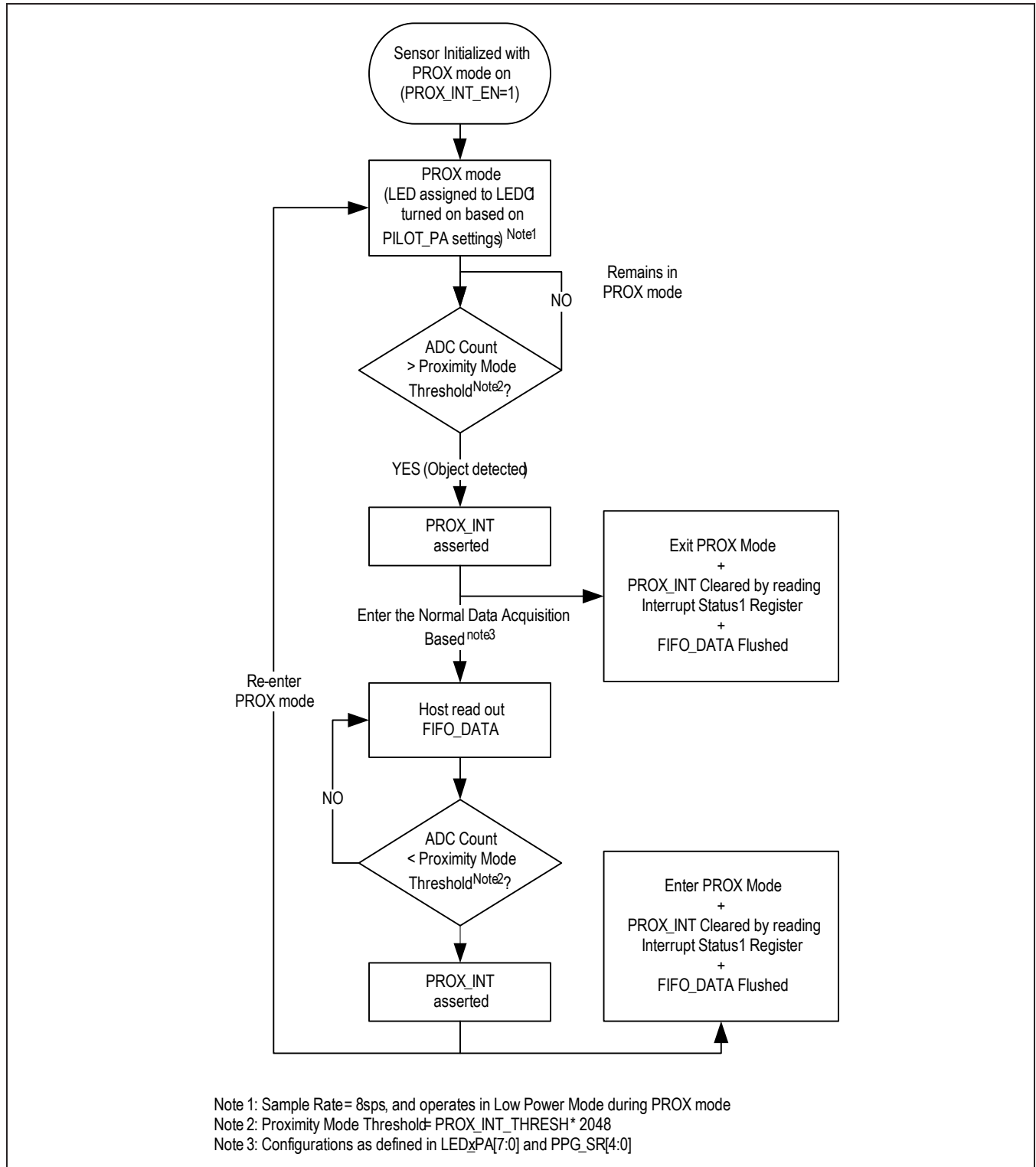


Figure 28. Proximity Function Flow Diagram

### Picket Fence Detect-and-Replace Function

Under typical situations, the rate of change of ambient light is such that the ambient signal level during exposure can be accurately predicted and high levels of ambient rejection are obtained. However, it is possible to have situations where the ambient light level changes extremely rapidly, for example when in a car with direct sunlight exposure passes under a bridge and into a dark shadow. In these situations, it is possible for the MAX86140/MAX86141 ambient light correction (ALC) circuit to fail and produce an erroneous estimation of the ambient light during the exposure interval. The built-in picket fence detect-and-replace function corrects the final PPG results in case of an ALC circuit failure due to these extreme conditions.

The picket fence detect-and-replace function works on the basis that the extreme conditions causing a failure of the ALC are rare events. These events resulting in a large deviation from the past sample history of a normal PPG riding on a motion effect signal, which normally would change relatively slowly with respect to the sampling interval. Under these conditions, it is possible to detect sample values that are well outside the normal sample to sample deviation and replace those samples with an extrapolated value based on the relatively recent history of samples.

The picket fence detect-and-replace function is disabled by default and is enabled by setting PF\_ENABLE (register 0x16[7]) bit to 1. The function begins with detecting a picket fence event. Detection is done by taking the absolute value of the difference between the present ADC converted value and a predicted point, called an estimation error, and comparing this estimation error to a threshold. If the estimation error exceeds the threshold, then the present ADC converted point is considered a picket fence event.

The predicted point referred to above is computed in one of two ways, set by the value in the PF\_ORDER (address 0x16[6]) bit. If PF\_ORDER = 0 the predicted point is simply the previous ADC converted point. If PF\_ORDER = 1, the predicted point is a least square fit extrapolation based on the previous four picket fence outputs, which, under normal circumstances, is identical to the ADC converted inputs.

The threshold used in detecting a picket fence event is a low pass version of the running estimation error computed above times a multiplier. The multiplier used is set by the THRESHOLD\_SIGMA\_MULT (address 0x16[1:0]) bits and can be 4, 8, 16, or 32 times the running low-pass filter output of the estimation error.

The low-pass filter function is controlled by two parameters, the IIR\_TC (address 0x16[5:4]) bits and IIR\_INIT\_VALUE (address 0x16[3:2]) bits. The IIR\_TC bits control the filter's time constant and are adjustable from 8 to 64 samples. The IIR\_INIT\_VALUE bits control the initial values for the IIR low pass filter when the algorithm is initialized.

Figure 29 illustrates the function in block diagram form. If the picket fence detect-and-replace function is enabled the input from the ADC,  $s(n)$  generates  $p(n)$  in a way that is dependent on the value of the PF\_ORDER bit. Value  $s(n)$  is subtracted from  $p(n)$  and turned into a positive number  $d(n)$  and fed into the IIR low-pass filter producing value  $lpf(n)$ . The output of the low pass filter  $lpf(n)$  is then multiplied by a user constant, THRESHOLD\_SIGMA\_MULT to produce the picket fence threshold, PFT(n). The value  $d(n)$  is then compared to this threshold and if greater than the PFT(n), the point  $s(n)$  is replaced with the point  $p(n)$ .

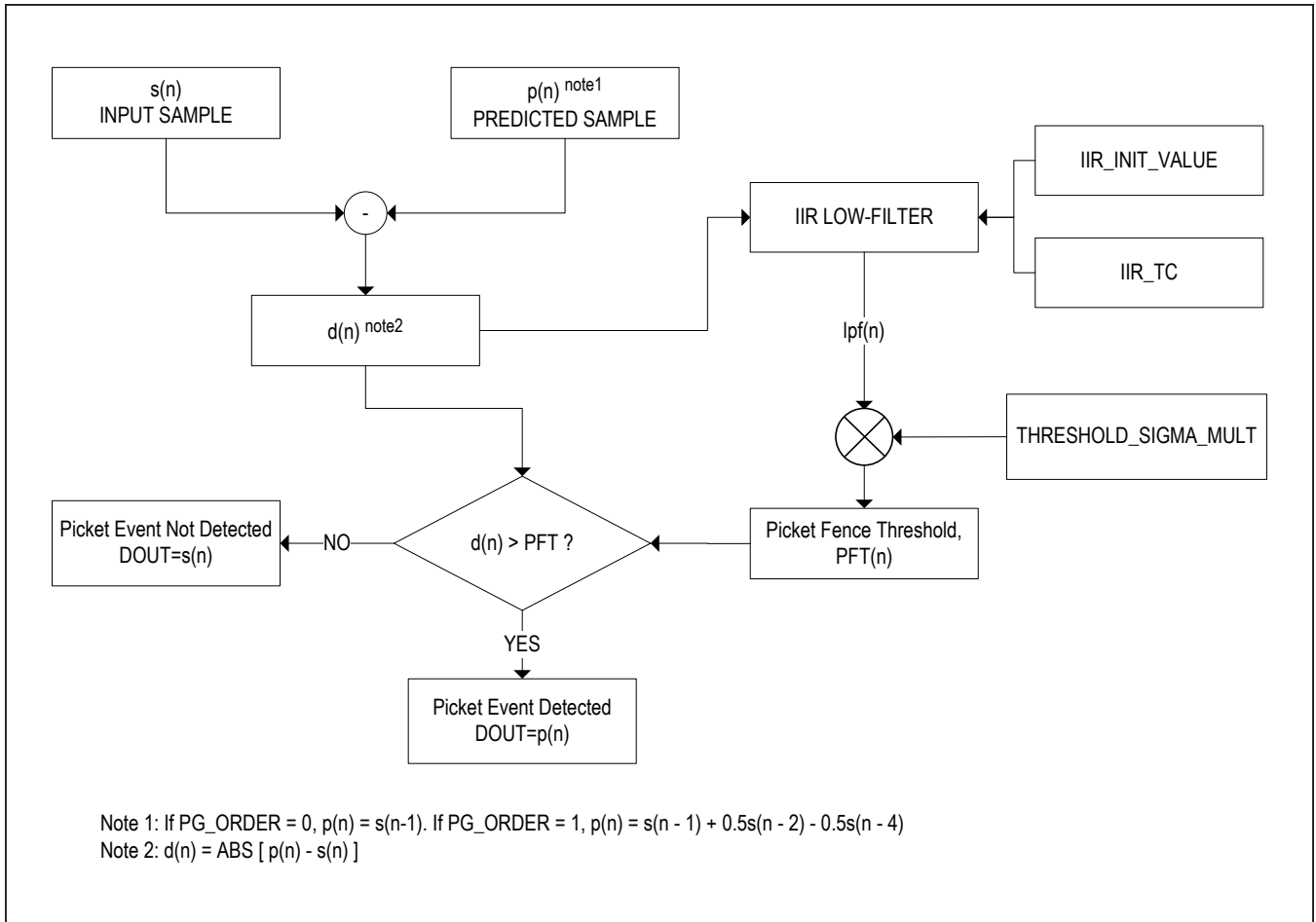


Figure 29. Picket Fence Detect-and-Replace Function Flow

This scheme essentially produces a threshold that tracks the past returned optical signal with a band width based on the past historical change sample to sample. [Figure 30](#) illustrates graphically how the threshold detection scheme works on a real PPG signal. Note that the black trace is

the real ADC sample points, the red trace is the output of the low-pass filter of the error estimation mirrored around the ADC points and the blue traces are the threshold values.

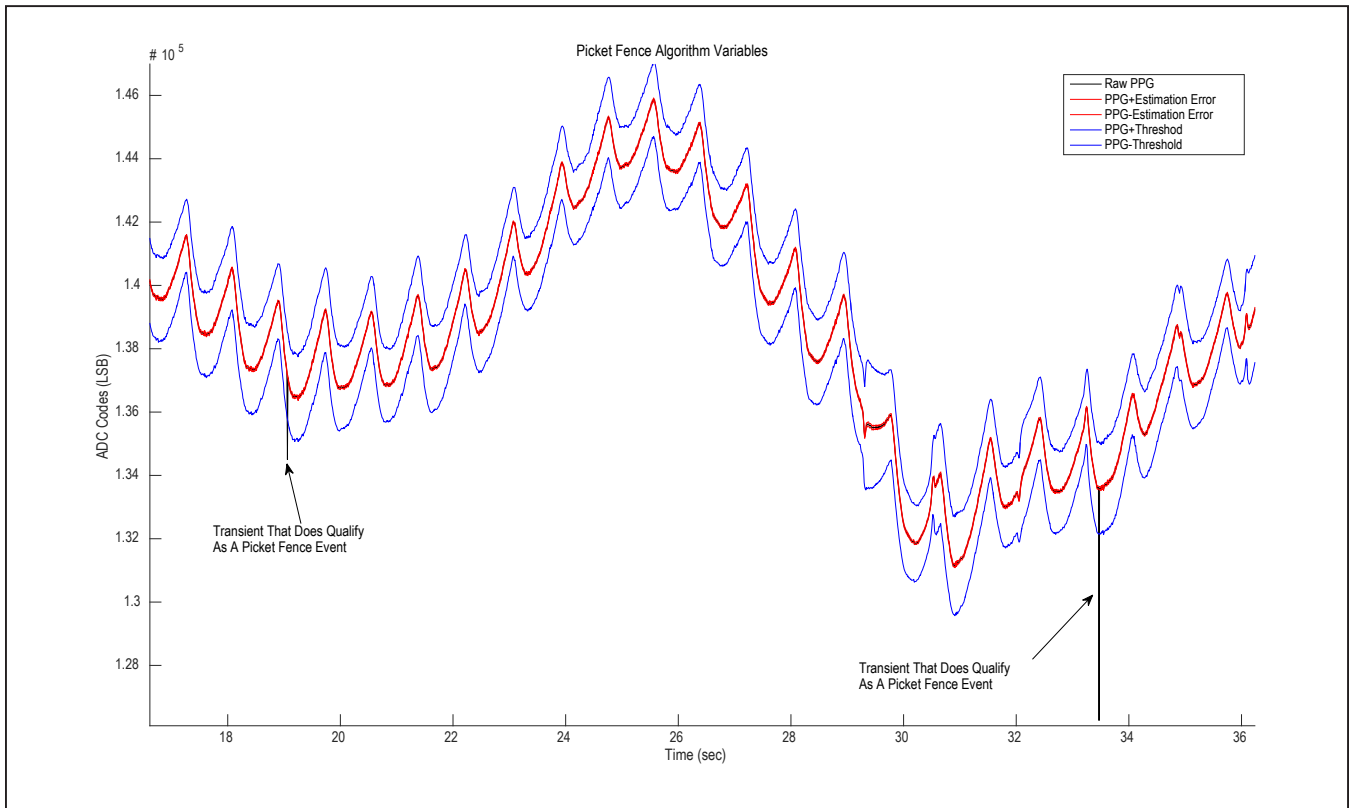


Figure 30. Picket Fence Variables In A PPG Waveform



The recommended settings for the picket fence detect-and-replace function are the default power on reset values for all registers except THRESHOLD\_SIGMA\_MULT bits. Here it is recommended that the 32x value 0x3 be used so only large excursions are classified as picket fence events. Lower values of THRESHOLD\_SIGMA\_MULT can cause the algorithm to go off track with extremely noisy waveform.

### Photodiode Biasing

The MAX86140/MAX86141 provides multiple photodiode biasing options (see [Table 16](#)). These options allow the MAX86140/MAX86141 to operate with a large range of photodiode capacitance. The PDBIASx settings adjust the PD\_IN bias point impedance to ensure that the photodiode settles rapidly enough to support the sample timing.

With a higher PD bias setting, the input-referred noise of the MAX86140/MAX86141 is increased. The relationship between Photodiode bias and noise with increasing photodiode capacitance is illustrated in the "Input Referred Noise vs. PD Capacitance" graph of the [Typical Operating Characteristics](#) section. Because of the increased noise with a higher PDBIASx setting, the lowest recommended PDBIASx settings should be used for a given photodiode capacitance.

### Layout Guidelines

The MAX86140/MAX86141 is a high dynamic range analog front-end (AFE) and its performance can be adversely impacted by the physical printed circuit board (PCB) layout. It is important that all bypass recommendations in the pin table are followed. Specifically, it is recommended that the VDD\_ANA and VDD\_DIG pins be shorted at the PCB. It is also recommended that GND\_ANA, GND\_DIG, and PGND be shorted to a single PCB ground plane. These three pins should be shorted on the edge of the WLP grid array. The pins should be connected using a single via to the PCB ground plane as close to the AFE as possible. Use multiple vias if grounds are not coplanar.

The combined VDD\_ANA and VDD\_DIG pins should then be decoupled with a 22µF 0402 and a 0.1µF ceramic chip capacitor to the PCB ground plane. Note that effective capacitance is approximately 10µF with voltage derating. In addition, the VREF pin should be decoupled to the PCB GND plane with a 1.0µF ceramic capacitor. The voltage on the VREF pin is nominally 1.21V, so a 6.3V rated ceramic capacitor should be adequate for this purpose. It is recommended that all decoupling caps use individual vias to the PCB GND plane to avoid mutual impedance coupling between decoupled supplies when sharing vias.

The most critical aspect of the PCB layout of MAX86140/MAX86141 is the handling of the PD\_IN and PD\_GND nodes. Parasitic capacitive coupling to the PD\_IN can result in additional noise being injected into the MAX86140/MAX86141 front-end. To minimize external interference coupling to PD\_IN, it is recommended that the PD\_IN node be fully shielded by the PD\_GND node. An example of this recommendation is shown below. In the three layers shown, the PD\_IN node is shielded with a coplanar PD\_GND trace on the top layer, the layer on which the MAX86140/MAX86141 is mounted. Ensure that no traces route adjacent to the PD\_IN vias. PD\_GND fill should be isolating PD\_IN vias from other traces. There should be no traces or vias other than PDx\_IN going through PD\_GND. Do not route traces over PD\_GND (i.e., the PCB GND plane should be adjacent to PD\_GND). On the bottom layer, the photodiode cathode is entirely shielded with the PD\_GND shield, which is also the photodiode anode. Note, also, that the PD\_GND shield also is extended below the photodiode. This is done because, in most photodiodes, the cathode is the bulk of the silicon. Therefore, shielding beneath the photodiode will terminate the capacitance to the bulk or cathode side to the reference node (PD\_GND). Finally, the PD\_GND pin should be attached to the PCB GND in only one point. This is shown on the top layer.

**Table 16. Recommended PDBIASx Values Based on the Photodiode Capacitance**

PDBIASx[2:0]	PHOTODIODE CAPACITANCE
0x001	0pF to 65pF
0x101	65pF to 130pF
0x110	130pF to 260pF
0x111	260pF to 520pF
All other values	Not recommended

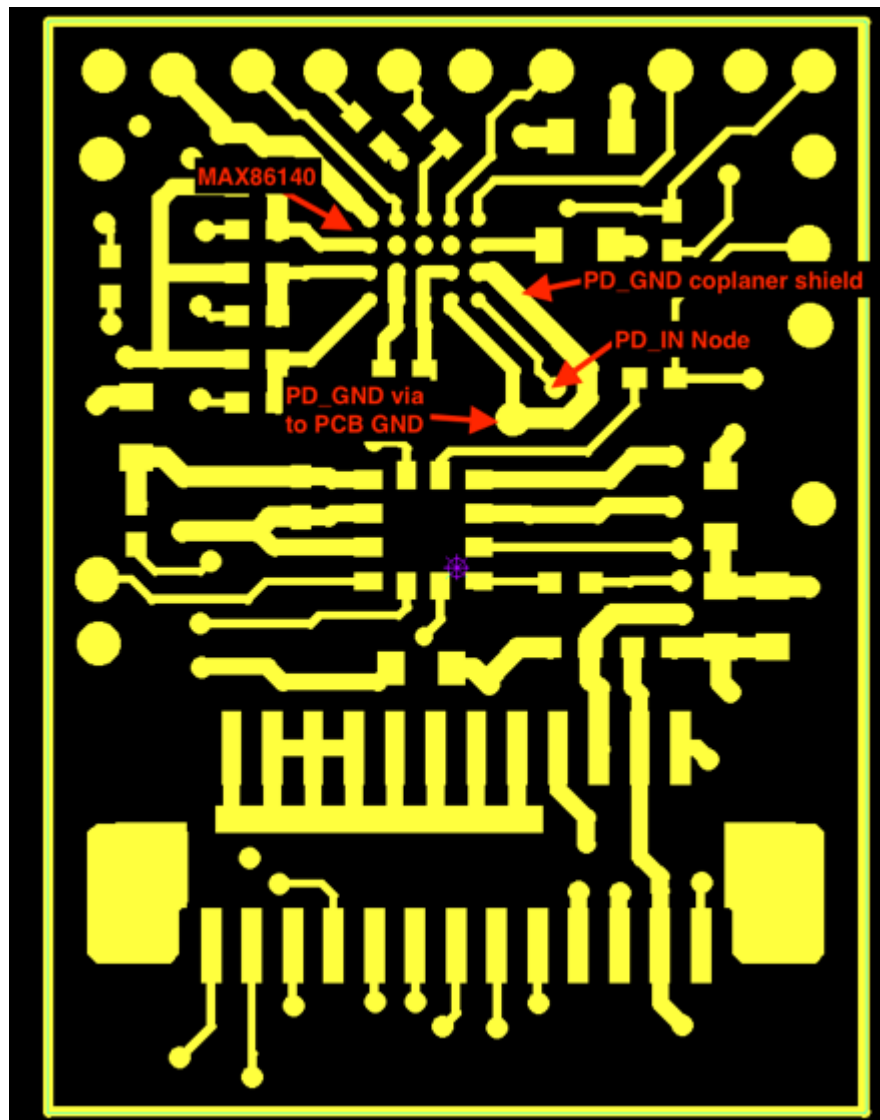


Figure 31. Example PCB Layout, Layer 1 (Top, MAX86140)

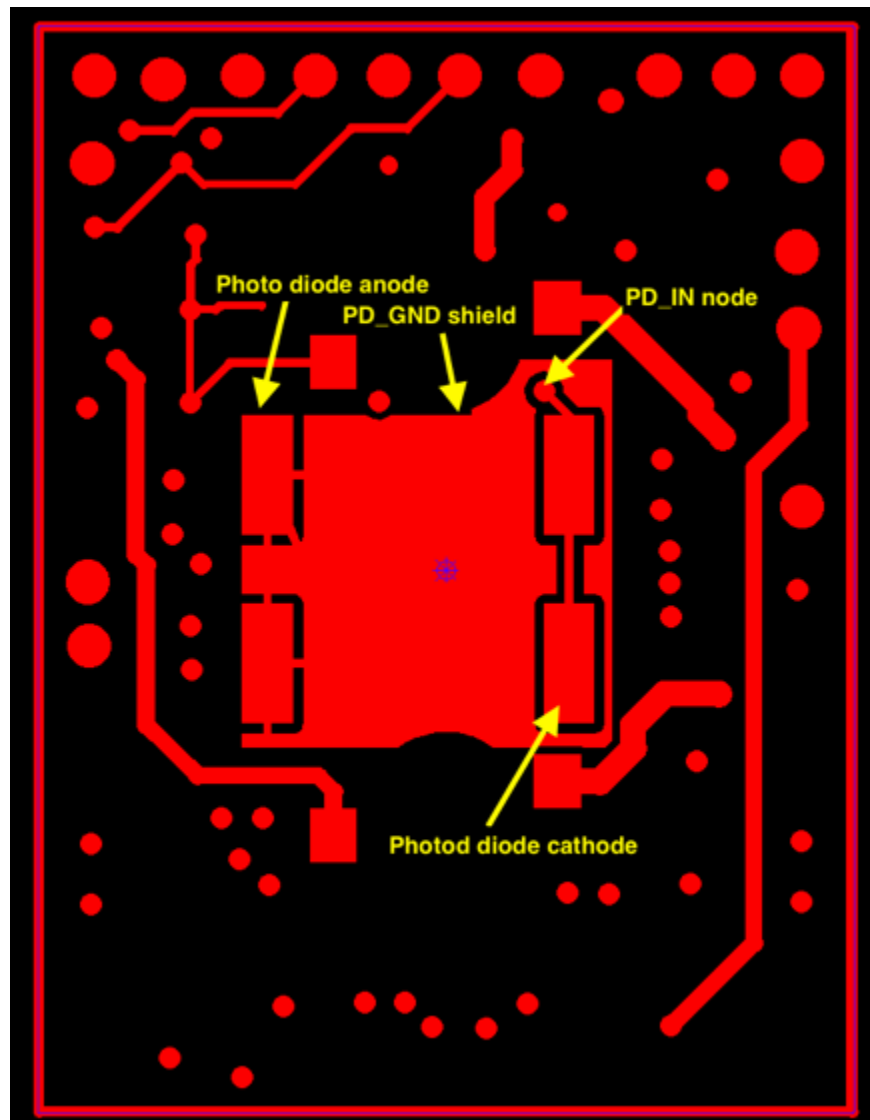


Figure 32. Example PCB Layout, Layer 6 (Bottom, Optical Layer)

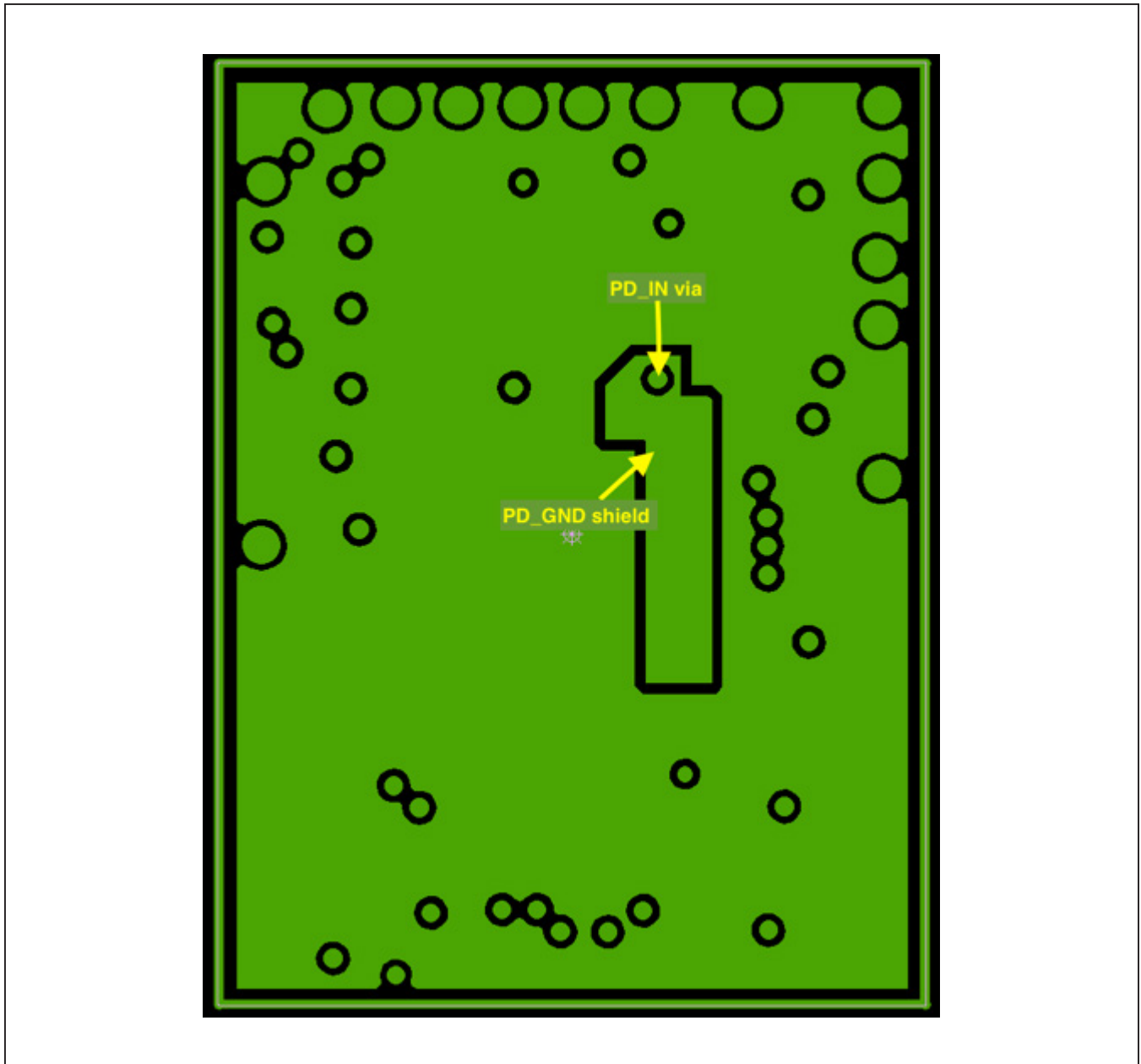


Figure 33. Example PCB Layout, Layer5 (Ground Layer)

## SPI Timing

### Detailed SPI Timing

The detailed SPI timing is illustrated below. The timings indicated are all specified in the [Electrical Characteristics](#) table.

### Single-Word SPI Register Read/Write Transaction

The MAX86140/MAX86141 is SPI/QSPI/Micro-wire/DSP compatible. The operation of the SPI interface is shown below. Data is strobed into the MAX86140/MAX86141 on the SCLK rising edge while clocked out on the SCLK falling edge. All single-word SPI read and write operations are done in a 3-byte, 24 clock cycle SPI instruction framed by a CSB low interval. The content of the SPI operation consists of a one-byte register address (A[7:0]), followed by

a one-byte command word that defines the transaction as write or read, followed by a single-byte data word either written to, or read from, the register location provided in the first byte.

Write mode operations will be executed on the 24<sup>th</sup> SCLK rising edge using the first three bytes of data available. In write mode, any data supplied after the 24<sup>th</sup> SCLK rising edge will be ignored. Subsequent writes require CSB to deassert high and then assert low for the next write command. A rising CSB edge preceding the 24<sup>th</sup> rising edge of SCLK by  $t_{CSA}$  (detailed SPI timing diagram), will result in the transaction being aborted.

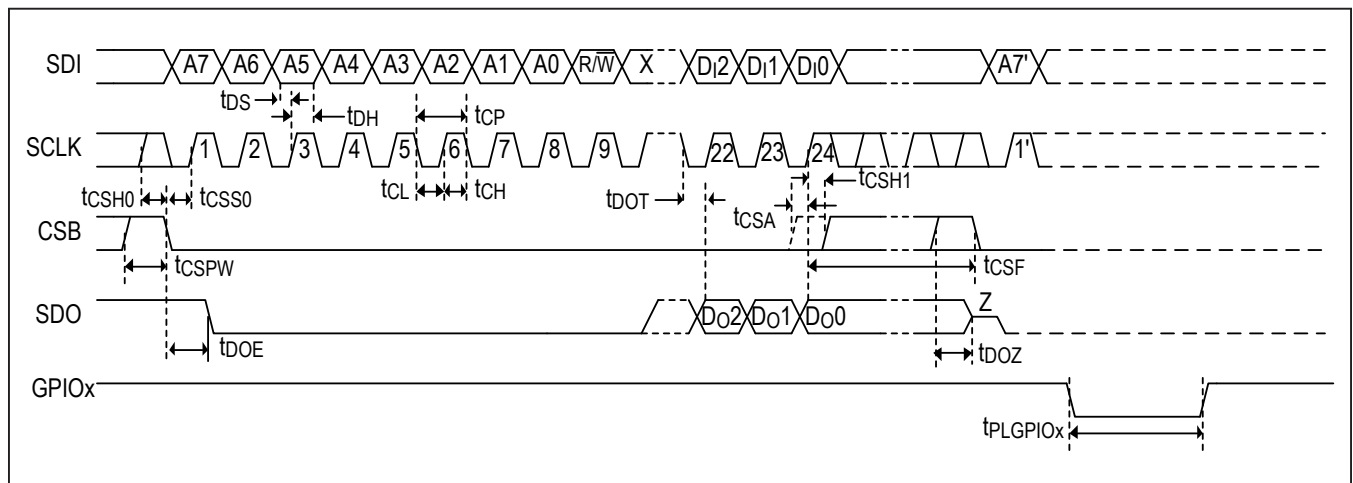


Figure 34. Detailed SPI Timing Diagram

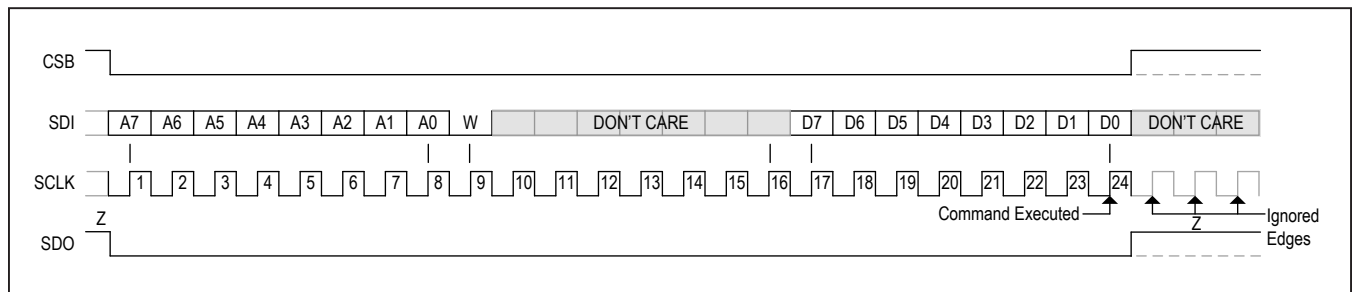


Figure 35. SPI Write Transaction

Read mode operations will access the requested data on the 16th SCLK rising edge, and present the MSB of the requested data on the following SCLK falling edge, allowing the  $\mu$ C to latch the data MSB on the 17th SCLK rising edge. Configuration and status registers are available through normal mode readback sequences. FIFO reads must be performed with a burst mode FIFO read (see [SPI FIFO Burst Mode Read Transaction](#)). If more than 24 SCLK rising edges are provided in a normal read sequence, the excess edges will be ignored and the device will read back zeros.

**SPI FIFO Burst Mode Read Transaction**

The MAX86140/MAX86141 provides a FIFO burst read mode to increase data transfer efficiency. The first 16 SCLK cycles operate exactly as described for the normal

read mode, the first byte being the register address, the second being a read command. The subsequent SCLKs consist of FIFO data, 24 SCLKs per word. All words in the FIFO should be read with a single FIFO burst read command.

Each FIFO sample consists of 3 bytes per sample and thus requires 24 SCLKs per sample to readout. The first byte (SCLK 17 to 24) consists of a tag indicating the data type of the subsequent bits. Following the tag is the MSBs of the subsequent data (MSB, MSB-1, and MSB-2). The next byte (SCLK 25 to 32) consists of data bits MSB-3 to MSB-11. The final byte of each sample (SCLK 33 to 40) consists of the data LSB bits. The number of words in the FIFO depends on the FIFO configuration. See [FIFO Configuration](#) for more details the FIFO configuration and readout.

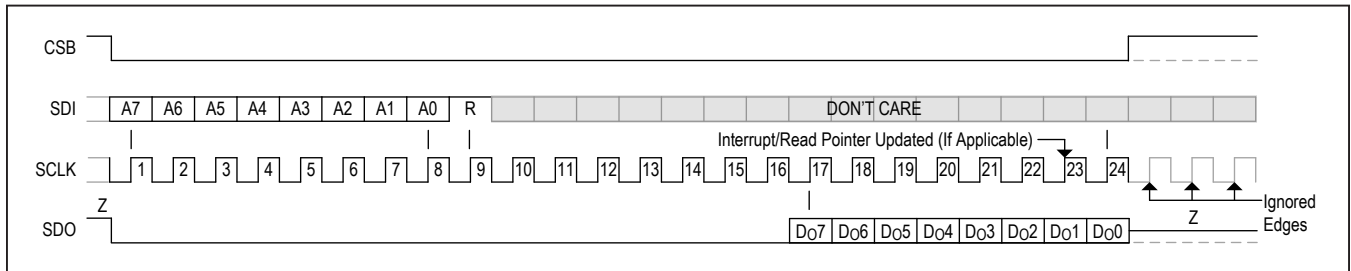


Figure 36. SPI Read Transaction

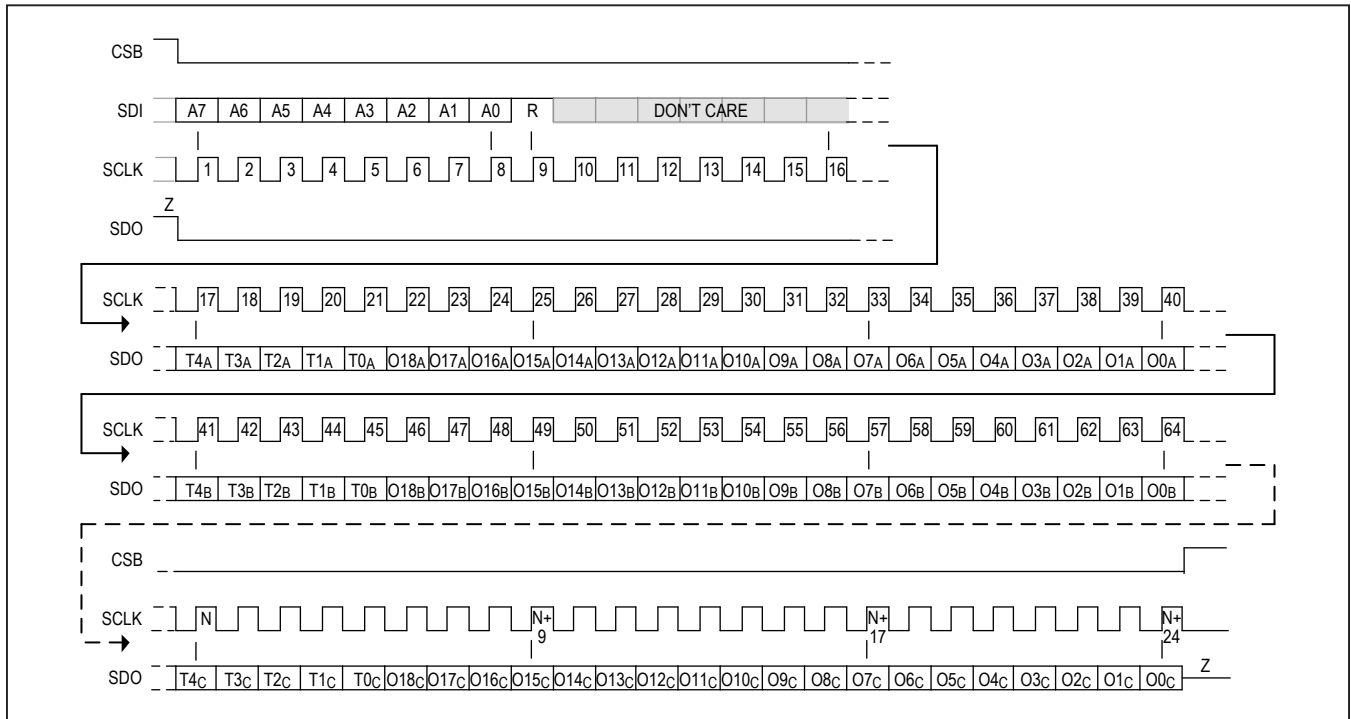


Figure 37. SPI FIFO Burst Mode Read Transaction

## Register Map

ADDRESS	NAME	MSB							LSB	
<b>Status</b>										
0x00	Interrupt Status 1[7:0]	A_FULL	DATA_RDY	ALC_OVF	PROX_INT	LED_COMPB	DIE_TEMP_RDY	VDD_OOR	PWR_RDY	
0x01	Interrupt Status 2[7:0]	-	-	-	-	-	-	-	SHA_DONE	
0x02	Interrupt Enable 1[7:0]	A_FULL_EN	DATA_RDY_EN	ALC_OVF_EN	PROX_INT_EN	LED_COMPB_EN	DIE_TEMP_RDY_EN	VDD_OOR_EN	-	
0x03	Interrupt Enable 2[7:0]	-	-	-	-	-	-	-	SHA_DONE_EN	
<b>FIFO</b>										
0x04	FIFO Write Pointer[7:0]	-	FIFO_WR_PTR[6:0]							
0x05	FIFO Read Pointer[7:0]	-	FIFO_RD_PTR[6:0]							
0x06	Over Flow Counter[7:0]	-	OVF_COUNTER[6:0]							
0x07	FIFO Data Counter[7:0]	FIFO_DATA_COUNT[7:0]								
0x08	FIFO Data Register[7:0]	FIFO_DATA[7:0]								
0x09	FIFO Configuration 1[7:0]	-	FIFO_A_FULL[6:0]							
0x0A	FIFO Configuration 2[7:0]	-	-	-	FLUSH_FIFO	FIFO_STAT_CLR	A_FULL_TYPE	FIFO_RO	-	
<b>System Control</b>										
0x0D	System Control[7:0]	-	-	-	-	SINGLE_PPG	LP_MODE	SHDN	RESET	
<b>PPG Configuration</b>										
0x10	PPG Sync Control[7:0]	TIME_STAMP_EN	-	-	SW_FORCE_SYNC	GPIO_CTRL[3:0]				
0x11	PPG Configuration 1[7:0]	ALC_DISABLE	ADD_OFFSET	PPG2_ADC_RGE[1:0]		PPG1_ADC_RGE[1:0]		PPG_TINT[1:0]		
0x12	PPG Configuration 2[7:0]	PPG_SR[4:0]					SMP_AVE[2:0]			
0x13	PPG Configuration 3[7:0]	LED_SETLNG[1:0]		DIG_FILT_SEL	-	-	BURST_RATE[1:0]		BURST_EN	
0x14	Prox Interrupt Threshold[7:0]	PROX_INT_THRESH[7:0]								
0x15	Photo Diode Bias[7:0]	-	PDBIAS2[2:0]			-	PDBIAS1[2:0]			

Register Map (continued)

ADDRESS	NAME	MSB					LSB
<b>PPG Picket Fence Detect and Replace</b>							
0x16	Picket Fence[7:0]	PF_ENABLE	PF_ORDER	IIR_TC[1:0]	IIR_INIT_VALUE[1:0]	THRESHOLD_SIGMA_MULT[1:0]	
<b>LED Sequence Control</b>							
0x20	LED Sequence Register 1[7:0]	LEDC2[3:0]			LEDC1[3:0]		
0x21	LED Sequence Register 2[7:0]	LEDC4[3:0]			LEDC3[3:0]		
0x22	LED Sequence Register 3[7:0]	LEDC6[3:0]			LEDC5[3:0]		
<b>LED Pulse Amplitude</b>							
0x23	LED1 PA[7:0]	LED1_DRV[7:0]					
0x24	LED2 PA[7:0]	LED2_DRV[7:0]					
0x25	LED3_PA[7:0]	LED3_DRV[7:0]					
0x26	LED4 PA[7:0]	LED4_DRV[7:0]					
0x27	LED5 PA[7:0]	LED5_DRV[7:0]					
0x28	LED6 PA[7:0]	LED6_DRV[7:0]					
0x29	LED PILOT PA[7:0]	PILOT_PA[7:0]					
0x2A	LED Range 1[7:0]	-	-	LED3_RGE[1:0]	LED2_RGE[1:0]	LED1_RGE[1:0]	
0x2B	LED Range 2[7:0]	-	-	LED6_RGE[1:0]	LED5_RGE[1:0]	LED4_RGE[1:0]	
<b>PPG1_HI_RES_DAC</b>							
0x2C	S1 HI RES DAC1[7:0]	S1_HRES_DAC1_OVR	-	S1_HRES_DAC1[5:0]			
0x2D	S2 HI RES DAC1[7:0]	S2_HRES_DAC1_OVR	-	S2_HRES_DAC1[5:0]			
0x2E	S3 HI RES DAC1[7:0]	S3_HRES_DAC1_OVR	-	S3_HRES_DAC1[5:0]			
0x2F	S4 HI RES DAC1[7:0]	S4_HRES_DAC1_OVR	-	S4_HRES_DAC1[5:0]			
0x30	S5 HI RES DAC1[7:0]	S5_HRES_DAC1_OVR	-	S5_HRES_DAC1[5:0]			
0x31	S6 HI RES DAC1[7:0]	S6_HRES_DAC1_OVR	-	S6_HRES_DAC1[5:0]			



Register Map (continued)

ADDRESS	NAME	MSB							LSB
<b>PPG2_HI_RES_DAC</b>									
0x32	S1 HI RES DAC2[7:0]	S1_HRES_DAC2_OVR	-					S1_HRES_DAC2[5:0]	
0x33	S2 HI RES DAC2[7:0]	S2_HRES_DAC2_OVR	-					S2_HRES_DAC2[5:0]	
0x34	S3 HI RES DAC2[7:0]	S3_HRES_DAC2_OVR	-					S3_HRES_DAC2[5:0]	
0x35	S4 HI RES DAC2[7:0]	S1_HRES_DAC2_OVR	-					S4_HRES_DAC2[5:0]	
0x36	S5 HI RES DAC2[7:0]	S2_HRES_DAC2_OVR	-					S5_HRES_DAC2[5:0]	
0x37	S6 HI RES DAC2[7:0]	S3_HRES_DAC2_OVR	-					S6_HRES_DAC2[5:0]	
<b>Die Temperature</b>									
0x40	Die Temperature Configuration[7:0]	-	-	-	-	-	-	-	TEMP_EN
0x41	Die Temperature Integer[7:0]	TEMP_INT[7:0]							
0x42	Die Temperature Fraction[7:0]	-	-	-	-	TEMP_FRAC[3:0]			
<b>SHA256</b>									
0xF0	SHA Command[7:0]	SHA_CMD[7:0]							
0xF1	SHA Configuration[7:0]	-	-	-	-	-	-	SHA_EN	SHA_START
<b>Memory</b>									
0xF2	Memory Control[7:0]	-	-	-	-	-	-	MEM_WR_EN	BANK_SEL
0xF3	Memory Index[7:0]	MEM_IDX[7:0]							
0xF4	Memory Data[7:0]	MEM_DATA[7:0]							
<b>Part ID</b>									
0xFF	Part ID[7:0]	PART_ID[7:0]							

**Interrupt Status 1 (0x00)**

BIT	7	6	5	4	3	2	1	0
Field	A_FULL	DATA_RDY	ALC_OVF	PROX_INT	LED_COMPB	DIE_TEMP_RDY	VDD_OOR	PWR_RDY
Reset	0x0	0x0	0x0	0x0	0x0	0x0	0x0	0x0
Access Type	Read Only	Read Only	Read Only	Read Only	Read Only	Read Only	Read Only	Read Only

**A\_FULL**

This is a read-only bit. This bit is cleared when the Interrupt Status 1 Register is read. It is also cleared when FIFO\_DATA register is read, if FIFO\_STAT\_CLR = 1.

VALUE	ENUMERATION	DECODE
0	OFF	Normal Operation
1	ON	Indicates that the FIFO buffer will overflow the threshold set by FIFO_A_FULL[6:0] on the next sample.

**DATA\_RDY**

This is a read-only bit and it is cleared by reading the Interrupt Status 1 register (0x00). It is also cleared by reading the FIFO\_DATA register if FIFO\_STAT\_CLR = 1.

VALUE	ENUMERATION	DECODE
0	OFF	Normal Operation
1	ON	This interrupt triggers when there is a new data in the FIFO.

**ALC\_OVF**

This is a read-only bit. The interrupt is cleared by reading the Interrupt Status 1 register (0x00).

VALUE	ENUMERATION	DECODE
0	OFF	Normal Operation
1	ON	This interrupt triggers when the ambient light cancellation function of the photodiode has reached its maximum limit due to overflow, and therefore, ambient light is affecting the output of the ADC.

**PROX\_INT**

VALUE	ENUMERATION	DECODE
0	OFF	Normal Operation
1	ON	Indicates that the ADC reading of the LED configured in LEDC1 has crossed the proximity threshold. If PROX_INT_EN is 0, then the prox mode is disabled and the exposure sequence configured in LED Sequence Control Registers begins immediately. This bit is cleared when the Interrupt Status 1 Register is read.

### LED\_COMPB

LED is not compliant. At the end of each sample, if the LED driver is not compliant, LED\_COMPB interrupt is asserted if LED\_COMPB\_EN is set to 1. It is a read-only bit and is cleared when the status register is read.

VALUE	ENUMERATION	DECODE
0	COMPLIANT	LED driver is compliant
1	NOT_COMPLIANT	LED driver is not compliant

### DIE\_TEMP\_RDY

This is a read-only bit and is automatically cleared when the temperature data is read, or when the Interrupt Status 1 Register is read.

VALUE	ENUMERATION	DECODE
0	OFF	Normal Operation
1	ON	Indicates that the TEMP ADC has finished it's current conversion.

### VDD\_OOR

This is a read-only bit. It is automatically cleared when the Interrupt Status 1 register is read.

The detection circuitry has a 10ms delay time, and will continue to trigger as long as the VDD\_ANA is out of range.

VALUE	ENUMERATION	DECODE
0	OFF	Normal operation
1	ON	Indicates that VDD_ANA is greater than 2.05V or less than 1.65V.

### PWR\_RDY

This is a read-only bit and indicates that VDD had gone below the UVLO threshold. This bit is not triggered by a soft reset. This bit is cleared when either Interrupt Status 1 Register is read, or by setting SHDN bit to 1.

VALUE	ENUMERATION	DECODE
0	OFF	Normal Operation
1	ON	Indicates that VBATT went below the UVLO threshold.

### Interrupt Status 2 (0x01)

BIT	7	6	5	4	3	2	1	0
Field	–	–	–	–	–	–	–	SHA_DONE
Reset	–	–	–	–	–	–	–	0x0
Access Type	–	–	–	–	–	–	–	Read Only

### SHA\_DONE

SHA256 Authentication Done status bit is set to 1 when the authentication algorithm completes. This is a read-only bit and gets cleared when the Status Register is read.

VALUE	ENUMERATION	DECODE
0x0		SHA Authentication not done
0x1		SHA Authentication done

**Interrupt Enable 1 (0x02)**

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	A_FULL_EN	DATA_RDY_EN	ALC_OVF_EN	PROX_INT_EN	LED_COMPB_EN	DIE_TEMP_RDY_EN	VDD_OOR_EN	–
<b>Reset</b>	0x0	0x0	0x0	0x0	0x0	0x0	0x0	–
<b>Access Type</b>	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read	–

**A\_FULL\_EN**

VALUE	ENUMERATION	DECODE
0	OFF	A_FULL interrupt is disabled
1	ON	A_FULL interrupt in enabled

**DATA\_RDY\_EN**

VALUE	ENUMERATION	DECODE
0	OFF	DATA_RDY interrupt is disabled
1	ON	DATA_RDY interrupt is enabled.

**ALC\_OVF\_EN**

VALUE	ENUMERATION	DECODE
0	OFF	ALC_OVF interrupt is disabled
1	ON	ALC_OVF interrupt in enabled

**PROX\_INT\_EN**

When PROX\_INT\_EN is enabled, the exposure programmed in the LEDC1 Sequence Register is used for proximity detection. If the ADC reading for this exposure is below 2048 times the threshold programmed in PROX\_INT\_THRESH register, the device is in proximity mode. Otherwise, it is in normal mode.

When the device is in proximity mode, the sample rate used is 8Hz, and the device starts data acquisition in pilot mode, using only one exposure of the LED programmed in LEDC1 register, and the LED current programmed in PILOT\_PA register.

When the device is in normal mode, the sample rate used is as defined under PPG\_SR register, and the device starts data acquisition in normal mode, using all the exposures programmed in the LED Sequence registers and appropriate LED currents.

PROX\_INT interrupt is asserted when the devices enters proximity mode or normal mode if PROX\_INT\_EN is programmed to 1.

VALUE	ENUMERATION	DECODE
0	OFF	Proximity mode and PROX_INT interrupt are disabled
1	ON	Proximity mode and PROX_INT interrupt are enabled

**LED\_COMPB\_EN**

VALUE	ENUMERATION	DECODE
0	DISABLE	LED_COMPB interrupt is disabled
1	ENABLE	LED_COMPB interrupt is enabled

**DIE\_TEMP\_RDY\_EN**

VALUE	ENUMERATION	DECODE
0	OFF	DIE_TEMP_RDY interrupt is disabled
1	ON	DIE_TEMP_RDY interrupt is enabled

**VDD\_OOR\_EN**

VALUE	ENUMERATION	DECODE
0	OFF	Disables the VDD_OVR interrupt
1	ON	Enables the VDD_OVR interrupt

**Interrupt Enable 2 (0x03)**

BIT	7	6	5	4	3	2	1	0
Field	–	–	–	–	–	–	–	SHA_DONE_EN
Reset	–	–	–	–	–	–	–	0x0
Access Type	–	–	–	–	–	–	–	Write, Read

**SHA\_DONE\_EN**

Enable SHA\_DONE Interrupt

VALUE	ENUMERATION	DECODE
0x0		SHA_DONE interrupt disabled
0x1		SHA_DONE interrupt enabled

**FIFO Write Pointer (0x04)**

BIT	7	6	5	4	3	2	1	0
Field	–	FIFO_WR_PTR[6:0]						
Reset	–	0x0						
Access Type	–	Read Only						

### FIFO\_WR\_PTR

This points to the location where the next sample will be written. This pointer advances for each sample pushed on to the circular FIFO.

See [FIFO Configuration](#) for details.

### FIFO Read Pointer (0x05)

BIT	7	6	5	4	3	2	1	0
Field	–	FIFO_RD_PTR[6:0]						
Reset	–	0x0						
Access Type	–	Write, Read						

### FIFO\_RD\_PTR

The FIFO Read Pointer points to the location from where the processor gets the next sample from the FIFO via the serial interface. This advances each time a sample is popped from the circular FIFO.

The processor may also write to this pointer after reading the samples. This allows rereading (or retrying) samples from the FIFO. However, writing to FIFO\_RD\_PTR may have adverse effects if it results in the FIFO being almost full.

Refer to [FIFO Configuration](#) for details.

### Overflow Counter (0x06)

BIT	7	6	5	4	3	2	1	0
Field	–	OVF_COUNTER[6:0]						
Reset	–	0x0						
Access Type	–	Read Only						

### OVF\_COUNTER

When FIFO is full, any new samples will result in new or old samples getting lost, depending on FIFO\_RO. OVF\_COUNTER counts the number of samples lost. It saturates at 0x7F.

Refer to [FIFO Configuration](#) for details.

### FIFO Data Counter (0x07)

BIT	7	6	5	4	3	2	1	0
Field	FIFO_DATA_COUNT[7:0]							
Reset	0x0							
Access Type	Read Only							

### FIFO\_DATA\_COUNT

This is a read-only register that holds the number of items available in the FIFO for the host to read. This increments when a new item is pushed to the FIFO and decrements when the host reads an item from the FIFO.

Refer to [FIFO Configuration](#) for details.

**FIFO Data Register (0x08)**

BIT	7	6	5	4	3	2	1	0
Field	FIFO_DATA[7:0]							
Reset	0x0							
Access Type	Read Only							

**FIFO\_DATA**

This is a read-only register and is used to get data from the FIFO. Refer to [FIFO Configuration](#) for details.

**FIFO Configuration 1 (0x09)**

BIT	7	6	5	4	3	2	1	0
Field	–	FIFO_A_FULL[6:0]						
Reset	–	0x3F						
Access Type	–	Write, Read						

**FIFO\_A\_FULL**

These bits indicate how many new samples can be written to the FIFO before the interrupt is asserted. For example, if set to 0xF, the interrupt triggers when there are 15 empty spaces left (113 entries), and so on.

Refer to [FIFO Configuration](#) for details.

FIFO_A_FULL<6:0>	FREE SPACE BEFORE INTERRUPT	# OF SAMPLES IN FIFO
0	0	128
1	1	127
2	2	126
3	3	125
----	----	----
126	126	2
127	127	1

**FIFO Configuration 2 (0x0A)**

BIT	7	6	5	4	3	2	1	0
Field	–	–	–	FLUSH_ FIFO	FIFO_ STAT_CLR	A_FULL_ TYPE	FIFO_RO	–
Reset	–	–	–	0x0	0x0	0x0	0x0	–
Access Type	–	–	–	Write, Read	Write, Read	Write, Read	Write, Read	–

### FLUSH\_FIFO

When this bit is set to '1', the FIFO gets flushed, FIFO\_WR\_PTR and FIFO\_RD\_PTR are reset to zero and FIFO\_DATA\_COUNT becomes 0. The contents of the FIFO are lost.

FIFO\_FLUSH is a self-clearing bit.

Refer to [FIFO Configuration](#) for details.

### FIFO\_STAT\_CLR

This defines whether the A-FULL interrupt should get cleared by FIFO\_DATA register read.

Refer to [FIFO Configuration](#) for details.

VALUE	ENUMERATION	DECODE
0	RD_DATA_NOCLR	A_FULL and DATA_RDY interrupts do not get cleared by FIFO_DATA register read. They get cleared by status register read.
1	RD_DATA_CLR	A_FULL and DATA_RDY interrupts get cleared by FIFO_DATA register read or status register read.

### A\_FULL\_TYPE

This defines the behavior of the A\_FULL interrupt.

VALUE	ENUMERATION	DECODE
0	AFULL_RPT	A_FULL interrupt gets asserted when the a_full condition is detected. It is cleared by status register read, but re-asserts for every sample if the a_full condition persists.
1	AFULL_ONCE	A_FULL interrupt gets asserted only when the a_full condition is detected. The interrupt gets cleared on status register read, and does not re-assert for every sample until a new a-full condition is detected.

### FIFO\_RO

Push enable when FIFO is full:

This bit controls the behavior of the FIFO when the FIFO becomes completely filled with data.

Push to FIFO is enabled when FIFO is full if FIFO\_RO = 1 and old samples are lost. Both FIFO\_WR\_PTR increments for each sample after the FIFO is full. FIFO\_RD\_PTR also increments for each sample pushed to the FIFO.

Push to FIFO is disabled when FIFO is full if FIFO\_RO = 0 and new samples are lost. FIFO\_WR\_PTR does not increment for each sample after the FIFO is full.

When the device is in PROX mode, push to FIFO is enabled independent of FIFO\_RO setting.

Refer to [FIFO Configuration](#) for details.

VALUE	ENUMERATION	DECODE
0	OFF	The FIFO stops on full.
1	ON	The FIFO automatically rolls over on full.



### System Control (0x0D)

BIT	7	6	5	4	3	2	1	0
Field	–	–	–	–	SINGLE_PPG	LP_MODE	SHDN	RESET
Reset	–	–	–	–	0x0	0x0	0x0	0x0
Access Type	–	–	–	–	Write, Read	Write, Read	Write, Read	Write, Read

### SINGLE\_PPG

In signal PP devices, this bit is ignored. In dual PPG devices, if this bit is 0, use two PPG channels; otherwise, use only PPG1 channel.

VALUE	ENUMERATION	DECODE
0x0	DUAL_PPG	Both PPG channels are enabled
0x1	SINGLE_PPG	Only PPG1 channel is enabled

### LP\_MODE

In low power mode, the sensor can be dynamically powered down between samples to conserve power. This dynamic power down mode option only supports samples rates of 256sps and below.

VALUE	ENUMERATION	DECODE
0	OFF	Dynamic power down is disabled.
1	ON	Dynamic power down is enabled. The device automatically enters low power mode between samples for samples rates 256sps and below. This mode is not available for higher sample rates.

### SHDN

The part can be put into a power-save mode by setting this bit to one. While in power-save mode, all configuration registers retain their values, and write/read operations function as normal. All interrupts are cleared to zero in this mode.

VALUE	ENUMERATION	DECODE
0	OFF	The part is in normal operation. No action taken.
1	ON	The part can be put into a power-save mode by writing a '1' to this bit. While in this mode all configuration registers remain accessible and retain their data. ADC conversion data contained in the registers are previous values. Writeable registers also remain accessible in shutdown. All interrupts are cleared. In this mode the oscillator is shutdown and the part draws minimum current. If this bit is asserted during an active conversion then the conversion is aborted.

### RESET

When this bit is set, the part undergoes a forced power-on-reset sequence. All configuration, threshold, and data registers including distributed registers are reset to their power-on-state. This bit then automatically becomes '0' after the reset sequence is completed.

VALUE	ENUMERATION	DECODE
0	OFF	The part is in normal operation. No action taken.
1	ON	The part undergoes a forced power-on-reset sequence. All configuration, threshold and data registers including distributed registers are reset to their power-on-state. This bit then automatically becomes '0' after the reset sequence is completed.

**PPG Sync Control (0x10)**

BIT	7	6	5	4	3	2	1	0
Field	TIME_STAMP_EN	–	–	SW_FORCE_SYNC	GPIO_CTRL[3:0]			
Reset	0x0	–	–	0x0	0x0			
Access Type	Write, Read	–	–	Write, Read	Write, Read			

**TIME\_STAMP\_EN**

Enable pushing TIME\_STAMP to FIFO. Refer to FIFO Configuration for details.

VALUE	ENUMERATION	DECODE
0x0	DISABLE	TIME_STAMP is not pushed to FIFO
0x1	ENABLE	TIME_STAMP is pushed to FIFO for a block of eight samples.

**SW\_FORCE\_SYNC**

Writing a 1 to this bit, aborts current sample and starts a new sample. This is a self clearing bit.

**GPIO\_CTRL**

The table below shows how the two GPIO ports are control for different modes of operation.

When two devices are configured to work as master-slave device pairs, they have to be configured identical for the following configuration register fields:

- PPG\_SR
- PPG\_TINT
- SMP\_AVE
- TIME\_STAMP\_EN
- FIFO\_A\_FULL
- FIFO\_ROLLS\_ON\_FULL

Number of LED Sequence Registers (LEDC1 to LEDC6) programmed should be same in both the devices. In Exposure Trigger mode, if Ambient is programmed in one of the registers, it needs to be in the same LEDCx register in both the devices.

GPIO\_CTRL register for both the devices should be programmed to be either Sample Trigger or Exposure Trigger.

It is also important to configure the Slave first and then the Master.

DATA\_RDY or A\_FULL interrupt should be enabled only on the Master. When interrupt is asserted read the Master first and then the Slave. Read same number of items from both devices.

Refer to GPIO Configuration for details.

GPIO_CTRL [3:0]	GPIO1 FUNCTION	GPIO2 FUNCTION	COMMENT
0000	Tristate or Mux Control	Disabled	GPIO1 will be active if any of the LEDCn[3:0] states A, B, or C are enabled in the exposure sequence. In this case, GPIO1 will be low during exposures on LED4, LED5, or LED6, otherwise it will be high. If LEDCn[3:0] state A, B, or C is not enabled in the exposure sequence, GPIO1 will be tristate unless externally pulled up. GPIO2 is disabled. Sample and exposure timing is controlled by the internal 32768Hz oscillator.
0001	Tristate or Mux Control	Input 32768Hz or 32000Hz Clock Input	GPIO1 will be active if any of the LEDCn[3:0] states A, B, or C are enabled in the exposure sequence. In this case, GPIO1 will be low during exposures on LED4, LED5 or LED6, otherwise it will be high. If LEDCn[3:0] state A, B, or C is not enabled in the exposure sequence, GPIO1 will be tristate unless externally pulled up. GPIO2 is an input 32768/32000Hz. Sample and exposure timing is controlled by GPIO2 clock input.

GPIO_CTRL [3:0]	GPIO1 FUNCTION	GPIO2 FUNCTION	COMMENT
0010	Input Sample Trigger	Tristate or Mux Control	GPIO1 is defined as a sample trigger input (Slave). This input can come from an external source or from another MAX86140/MAX86141 in master sample mode. GPIO2 will be active if any of the LEDCn[3:0] states A, B, or C are enabled in the exposure sequence. In this case, GPIO2 will be low during exposures on LED4, LED5, or LED6, otherwise it will be high. If LEDCn[3:0] state A, B, or C is not enabled in the exposure sequence, GPIO2 will be tristate unless externally pulled up. Exposure timing is controlled by internal oscillator.
0011	Input Sample Trigger	Input 32768Hz or 32000Hz Clock Input	GPIO1 is defined as a sample trigger input (Slave). This input can come from an external source or from another MAX86140/MAX86141 in master sample mode. GPIO2 is an input 32768/32000Hz clock input. Exposure timing is controlled by GPIO2 clock input.
0100	Active Output Master Sample Output	Tristate or Mux Control	GPIO1 is defined as a master sample output. The GPIO1 output can be used to trigger a second sensor. When used with a second MAX86140/MAX86141 set to slave sample mode, the master sample timing will drive slave sample time. GPIO2 will be active if any of the LEDCn[3:0] states A, B, or C are enabled in the exposure sequence. In this case, GPIO2 will be low during exposures on LED4, LED5 or LED6, otherwise it will be high. If LEDCn[3:0] state A, B, or C is not enabled in the exposure sequence, GPIO2 will be tristate unless externally pulled up. Sample and exposure timing is controlled by internal oscillator.
0101	Active Output Master Sample Output	Input 32768/32000Hz Clock Input	GPIO1 is defined as a master sample output. The GPIO1 output can be used to trigger and second sensor. When used with a second MAX86140/MAX86141 set to slave sample mode, the master sample timing will drive slave sample time. GPIO2 is an input 32768/32000Hz. Exposure timing is controlled by GPIO2 clock input.
0110	Input Exposure Trigger	Tristate or Mux Control	GPIO1 is defined as an exposure trigger input (Slave). This input can come from an external source or from another MAX86140/MAX86141 in master sample mode. Both sample and exposure timing is controlled by the GPIO1 input. GPIO2 will be active if any of the LEDCn[3:0] states A, B, or C are enabled in the exposure sequence. In this case, GPIO2 will be low during exposures on LED4, LED5, or LED6, otherwise it will be high. If LEDCn[3:0] state A, B, or C is not enabled in the exposure sequence, GPIO2 will be tristate unless externally pulled up.
0111	Active Output Master Exposure Output	Tristate or Mux Control	GPIO1 is defined as a master sample output. The GPIO1 output can be used to trigger and second sensor. When used with a second MAX86140/MAX86141 set to slave exposure mode, the master exposure timing will drive slave exposure time. GPIO2 will be active if any of the LEDCn[3:0] states A, B, or C are enabled in the exposure sequence. In this case, GPIO2 will be low during exposures on LED4, LED5 or LED6, otherwise it will be high. If LEDCn[3:0] state A, B, or C is not enabled in the exposure sequence, GPIO2 will be tristate unless externally pulled up. Sample and exposure time is controlled internally
1000	Active Output Master Exposure Output	Input 32768/32000Hz Clock Input	GPIO1 is defined as a master sample output. The GPIO1 output can be used to trigger second sensor. When used with a second MAX86140/MAX86141 set to slave exposure mode, the master exposure timing will drive slave exposure time. GPIO2 is an input 32768/32000Hz. Sample and exposure timing is controlled by GPIO2 clock input.
1001	Input HW_FORCE_SYNC	Input 32768Hz or 32000Hz Clock Input	GPIO1 is defined as a start of sample sync input. The falling edge of GPIO1 causes the present sample sequence to be terminated and reinitiated on the next rising edge of GPIO2 input. GPIO2 is an input 32768/32000Hz. Sample and exposure timing is controlled by GPIO2 clock input.

**PPG Configuration 1 (0x11)**

BIT	7	6	5	4	3	2	1	0
Field	ALC_DISABLE	ADD_OFFSET	PPG2_ADC_RGE[1:0]		PPG1_ADC_RGE[1:0]		PPG_TINT[1:0]	
Reset	0x0	0	0x0		0x0		0x3	
Access Type	Write, Read	Write, Read	Write, Read		Write, Read		Write, Read	

**ALC\_DISABLE**

VALUE	ENUMERATION	DECODE
0	OFF	ALC is enabled
1	ON	ALC is disabled

**ADD\_OFFSET**

ADD\_OFFSET is an option designed for dark current measurement. By adding offset to the PPG Data would allow dark current measurement without clipping the signal below 0.

When ADD\_OFFSET is set to 1, an offset is added to the PPG Data to be able to measure the dark current. The offset is 8192 counts if PPG\_SR is programmed for single pulse mode. The offset is 4096 counts if PPG\_SR is programmed for dual pulse mode.

**PPG2\_ADC\_RGE**

These bits set the ADC range of the SPO<sub>2</sub> sensor, as shown in the table below.

PPG_ADC_RGE[1:0]	LSB (pA)	FULL SCALE (nA)
00	78125	4096
01	15.625	8192
10	31.25	16384
11	62.5	32768

**PPG1\_ADC\_RGE**

These bits set the ADC range of the SPO<sub>2</sub> sensor, as shown in the table below.

PPG_ADC_RGE[1:0]	LSB (pA)	FULL SCALE (nA)
00	7,8125	4096
01	15.625	8192
10	31.25	16384
11	62.5	32768

**PPG\_TINT**

These bits set the pulse width of the LED drivers and the integration time of PPG ADC as shown in the table below.

$$t_{PW} = t_{TINT} + t_{LED\_SETLNG} + 0.5\mu s$$

PPG_TINT[1:0]	TPW, PULSE WIDTH (μS)	TTINT, INTEGRATION TIME (μS)	RESOLUTION BITS
00	21.3	14.8	19
01	35.9	29.4	19
10	65.2	58.7	19
11	123.8	117.3	19

**PPG Configuration 2 (0x12)**

BIT	7	6	5	4	3	2	1	0
Field	PPG_SR[4:0]					SMP_AVE[2:0]		
Reset	0x11					0x0		
Access Type	Write, Read					Write, Read		

**PPG\_SR**

These bits set the effective sampling rate of the PPG sensor as shown in the table below. The default on-chip sampling clock frequency is 32768Hz.

**Note:** If a sample rate is set that can not be supported by the selected pulse width and number of exposures per sample, then the highest available sample rate will be automatically set. The user can read back this register to confirm the sample rate.

SAMPLING CLOCK FREQUENCY	32768Hz	32000Hz	
PPG_SR[4:0]	Samples per Second	Samples per Second	Pulses Per Sample, N
0x00	24.995	24.409	1
0x01	50.027	48.855	1
0x02	84.021	82.051	1
0x03	99.902	97.561	1
0x04	199.805	195.122	1
0x05	399.610	390.244	1
0x06	24.995	24.409	2
0x07	50.027	48.855	2
0x08	84.021	82.051	2
0x09	99.902	97.561	2
0x0A	8.000	7.8125	1
0x0B	16.000	15.625	1
0x0C	32.000	31.250	1
0x0D	64.000	62.500	1
0x0E	128.000	125.000	1
0x0F	256.000	250.000	1
0x10	512.000	500.000	1
0x11	1024.000	1000.000	1
0x12	2048.000	2000.000	1
0x13	4096.000	4000.000	1
0x14-1F	Reserved	Reserved	Reserved

Maximum Sample rates (sps) supported for all the Integration Time (PPG\_TINT) and Number of Exposures:

NUMBER OF EXPOSURE PER SAMPLE	PPG_TINT = 0 (14.8µs)	PPG_TINT = 1 (29.4µs)	PPG_TINT = 2 (58.7µs)	PPG_TINT = 3 (117.3µs)
1 Exposure, N = 1	4096	2048	2048	1024
2 Exposures, N = 1	2048	1024	1024	512
3 Exposures, N = 1	1024	1024	512	512
4 Exposures, N = 1	1024	512	512	400
5 Exposures, N = 1	512	512	512	256
6 Exposures, N = 1	512	512	400	256
1 Exposure, N = 2	100	100	100	100
2 Exposures, N = 2	100	84	84	84
3 Exposures, N = 2	50	50	50	50
4 Exposures, N = 2	25	25	25	25
5 Exposures, N = 2	25	25	25	25
6 Exposures, N = 2	25	25	25	25

### SMP\_AVE

To reduce the amount of data throughput, adjacent samples (in each individual channel) can be averaged and decimated on the chip by setting this register.

These bits set the number of samples that are averaged on chip before being written to the FIFO.

SMP_AVE[2:0]	SAMPLE AVERAGE
000	1 (no averaging)
001	2
010	4
011	8
100	16
101	32
110	64
111	128

When BURST\_EN is 1, SMP\_AVE defines the number of conversions per burst. Depending on the BURST\_RATE programmed and the PPG\_SR used, it may not be possible to accommodate some of SMP\_AVE values. In that case, SMP\_AVE will take the highest value that can be accommodated. If SMP\_AVE = 0 cannot be accommodated, burst mode is disabled.

**Note:** PPG\_SR itself depends on Number of conversions per sample (LEDC1 to LEDC6) and the LED Integration time (PPG\_TINT).

The following table shows the maximum SMP\_AVE allowed for various configurations of BURST\_RATE and PPG\_SR:

PPG_SR USED	BURST_RATE = 0 (8Hz)	BURST_RATE = 1 (32Hz)	BURST_RATE = 2 (84Hz)	BURST_RATE = 3 (256Hz)
0 (25Hz, N = 1)	1	DIS	DIS	DIS
1 (50Hz, N = 1)	2	0	DIS	DIS
2 (84Hz, N = 1)	3	1	DIS	DIS
3 (100Hz, N = 1)	3	1	DIS	DIS
4 (200Hz, N = 1)	4	2	0	DIS
5 (400Hz, N = 1)	5	3	1	DIS
6 (25Hz, N = 2)	1	DIS	DIS	DIS
7 (50Hz, N = 2)	2	0	DIS	DIS
8 (84Hz, N = 2)	3	1	DIS	DIS
9 (100Hz, N = 2)	3	1	DIS	DIS
A (8Hz, N = 1)	DIS	DIS	DIS	DIS
B (16Hz, N = 1)	0	DIS	DIS	DIS
C (32Hz, N = 1)	1	DIS	DIS	DIS
D (64Hz, N = 1)	2	0	DIS	DIS
E (128Hz, N = 1)	3	1	0	DIS
F (256Hz, N = 1)	4	2	1	DIS
10 (512Hz, N = 1)	5	3	2	DIS
11 (1024Hz, N = 1)	6	4	3	0
12 (2048Hz, N = 1)	7	5	4	1
13 (4096Hz, N = 1)	7	6	5	2

**PPG Configuration 3 (0x13)**

BIT	7	6	5	4	3	2	1	0
Field	LED_SETLNG[1:0]		DIG_FILT_SEL	–	–	BURST_RATE[1:0]		BURST_EN
Reset	0x1		0x0	–	–	0x0		0x0
Access Type	Write, Read		Write, Read	–	–	Write, Read		Write, Read

**LED\_SETLNG**

Delay from rising-edge of LED to start of ADC integration. This allows for the LED current to settle before the start of ADC integration.

LED_SETLNG[1:0]	DELAY (μs)
00	4.0
01	6.0 (default)
10	8.0
11	12.0

### DIG\_FILT\_SEL

Select digital filter type.

VALUE	DECODE
0x0	Use CDM
0x1	Use FDM

### BURST\_RATE

BURST_RATE[1:0]	FREQUENCY OF BURST (Hz)
00	8
01	32
10	84
11	256

### BURST\_EN

When Burst Mode is disabled, PPG data conversions are continuous at the sample rate defined by PPG\_SR register. When Burst mode is enabled, a burst of PPG data conversions occur at the sample rate defined by PPG\_SR register. Number of conversion in the burst is defined by the SMP\_AVE register. Average data from the burst of data conversions is pushed to the FIFO. The burst repeats at the rate defined in BURST\_RATE[2:0] register. If the number of conversions cannot be accommodated, the device will use the next highest number of conversions.

If the effective PPG\_SR is too slow to accommodate the burst rate programmed, BURST\_EN is automatically set to 0, and the device runs in continuous mode.

**Note:** Each data conversion cycle is a sequence of conversions defined in the LEDC1 to LEDC6 registers.

VALUE	ENUMERATION	DECODE
0x0		Disable Burst Conversion mode
0x1		Enable Burst Conversion Mode

### Prox Interrupt Threshold (0x14)

BIT	7	6	5	4	3	2	1	0
Field	PROX_INT_THRESH[7:0]							
Reset	0x00							
Access Type	Write, Read							

### PROX\_INT\_THRESH

This register sets the LED1 ADC count that will trigger the transition between proximity mode and normal mode. The threshold is defined as the 8 MSB bits of the ADC count. For example, if PROX\_INT\_THRESH[7:0] = 0x01, then an ADC value of 2048 (decimal) or higher triggers the PROX interrupt. If PROX\_INT\_THRESH[7:0] = 0xFF, then only a saturated ADC triggers the interrupt.

See the [Proximity Mode Function](#) section in the detailed description for more details on the operation of proximity mode.



**Photo Diode Bias (0x15)**

BIT	7	6	5	4	3	2	1	0
Field	–	PDBIAS2[2:0]			–	PDBIAS1[2:0]		
Reset	–	0x0			–	0x0		
Access Type	–	Write, Read			–	Write, Read		

**PDBIAS2**

See [Photodiode Biasing](#) for more information.

PDBIAS2[2:0]	PHOTODIODE CAPACITANCE
001	0pF to 65pF
101	65pF to 130pF
110	130pF to 260pF
111	260pF to 520pF
All other values	Not recommended

**PDBIAS1**

See [Photodiode Biasing](#) for more information.

PDBIAS1[2:0]	PHOTODIODE CAPACITANCE
001	0pF to 65pF
101	65pF to 130pF
110	130pF to 260pF
111	260pF to 520pF
All other values	Not recommended

**Picket Fence (0x16)**

BIT	7	6	5	4	3	2	1	0
Field	PF_ENABLE	PF_ORDER	IIR_TC[1:0]		IIR_INIT_VALUE[1:0]		THRESHOLD_SIGMA_MULT[1:0]	
Reset	0x0	0x1	0x00		0x00		0x00	
Access Type	Write, Read	Write, Read	Write, Read		Write, Read		Write, Read	

**PF\_ENABLE**

Refer to [Picket Fence Detect-and-Replace Function](#) for details.

PF\_ENABLE set to 1 enabled the picket fence detect-and-replace function.

VALUE	ENUMERATION	DECODE
0	OFF	Disable (default)
1	ON	Enable Detect and Replace

### PF\_ORDER

PF\_ORDER determines which prediction method is used: the last sample or a linear fit to the previous four samples. Refer to [Picket Fence Detect-and-Replace Function](#) for details.

VALUE	ENUMERATION	DECODE
0	OFF	Last Sample (1 point)
1	ON	Fit 4 points to a line for prediction (default)

### IIR\_TC

IIR\_TC<1:0> determines the IIR filter bandwidth where the lowest setting has the narrowest bandwidth of a first-order filter.

Refer to [Picket Fence Detect-and-Replace Function](#) for details.

IIR_TC[1:0]	COEFFICIENT	SAMPLES TO 90%
00	1/64	146
01	1/32	72
10	1/16	35
11	1/8	17

### IIR\_INIT\_VALUE

This IIR filter estimates the true standard deviation between the actual and predicted sample and tracks the ADC Range setting.

Refer to [Picket Fence Detect-and-Replace Function](#) for details.

IIR_INIT_VALUE[1:0]	CODE
00	64
01	48
10	32
11	24

### THRESHOLD\_SIGMA\_MULT

GAIN resulting from the SIGMA\_MULT<1:0> setting determines the number of standard deviations of the delta between the actual and predicted sample beyond which a picket-fence event is triggered.

Refer to [Picket Fence Detect-and-Replace Function](#) for details.

THRESHOLD_SIGMA_MULT[1:0]	GAIN
00	4
01	8
10	16
11	32

### LED Sequence Register 1 (0x20)

BIT	7	6	5	4	3	2	1	0
Field	LEDC2[3:0]				LEDC1[3:0]			
Reset	0x0				0x0			
Access Type	Write, Read				Write, Read			

**LEDC2**

These bits set the data type for LED Sequence 2 of the FIFO.  
See [FIFO Configuration](#) for more information.

**LEDC1**

These bits set the data type for LED Sequence 1 of the FIFO.  
See [FIFO Configuration](#) for more information.

**LED Sequence Register 2 (0x21)**

BIT	7	6	5	4	3	2	1	0
Field	LEDC4[3:0]				LEDC3[3:0]			
Reset	0x0				0x0			
Access Type	Write, Read				Write, Read			

**LEDC4**

These bits set the data type for LED Sequence 4 of the FIFO.  
See [FIFO Configuration](#) for more information.

**LEDC3**

These bits set the data type for LED Sequence 3 of the FIFO.  
See [FIFO Configuration](#) for more information.

**LED Sequence Register 3 (0x22)**

BIT	7	6	5	4	3	2	1	0
Field	LEDC6[3:0]				LEDC5[3:0]			
Reset	0x0				0x0			
Access Type	Write, Read				Write, Read			

**LEDC6**

These bits set the data type for LED Sequence 6 of the FIFO.  
See [FIFO Configuration](#) for more information.

**LEDC5**

These bits set the data type for LED Sequence 5 of the FIFO.  
See [FIFO Configuration](#) for more information.

**LED1 PA (0x23)**

BIT	7	6	5	4	3	2	1	0
Field	LED1_DRV[7:0]							
Reset	0x00							
Access Type	Write, Read							

**LED1\_DRV**

These bits set the nominal drive current of LED 1, as shown in the table below.

LEDX_RGE[1:0]	00	01	10	11
LEDx_PA[7:0]	LED Current (mA)	LED Current (mA)	LED Current (mA)	LED Current (mA)
00000000	0.00	0.00	0.00	0.00
00000001	0.12	0.24	0.36	0.48
00000010	0.24	0.48	0.73	0.97
00000011	0.36	0.73	1.09	1.45
.....				
11111100	30.6	61.3	91.9	122.5
11111101	30.8	61.5	92.3	123.0
11111110	30.9	61.8	92.6	123.5
11111111	31.0	62.0	93.0	124.0
LSB	0.12	0.24	0.36	0.48

**LED2\_PA (0x24)**

BIT	7	6	5	4	3	2	1	0
Field	LED2_DRV[7:0]							
Reset	0x00							
Access Type	Write, Read							

**LED2\_DRV**

These bits set the nominal drive current of LED 2. See *LED1\_DRV* for description.

**LED3\_PA (0x25)**

BIT	7	6	5	4	3	2	1	0
Field	LED3_DRV[7:0]							
Reset	0x00							
Access Type	Write, Read							

**LED3\_DRV**

These bits set the nominal drive current of LED 2. See *LED1\_DRV* for description.

**LED4\_PA (0x26)**

BIT	7	6	5	4	3	2	1	0
Field	LED4_DRV[7:0]							
Reset	0x00							
Access Type	Write, Read							

**LED4\_DRV**

These bits set the nominal drive current of LED 4. See *LED1\_DRV* for description.

**LED5 PA (0x27)**

<b>BIT</b>	<b>7</b>	<b>6</b>	<b>5</b>	<b>4</b>	<b>3</b>	<b>2</b>	<b>1</b>	<b>0</b>
<b>Field</b>	LED5_DRV[7:0]							
<b>Reset</b>	0x00							
<b>Access Type</b>	Write, Read							

**LED5\_DRV**

These bits set the nominal drive current of LED 5. See *LED1\_DRV* for description.

**LED6 PA (0x28)**

<b>BIT</b>	<b>7</b>	<b>6</b>	<b>5</b>	<b>4</b>	<b>3</b>	<b>2</b>	<b>1</b>	<b>0</b>
<b>Field</b>	LED6_DRV[7:0]							
<b>Reset</b>	0x00							
<b>Access Type</b>	Write, Read							

**LED6\_DRV**

These bits set the nominal drive current of LED 6. See *LED1\_DRV* for description.

**LED PILOT PA (0x29)**

<b>BIT</b>	<b>7</b>	<b>6</b>	<b>5</b>	<b>4</b>	<b>3</b>	<b>2</b>	<b>1</b>	<b>0</b>
<b>Field</b>	PILOT_PA[7:0]							
<b>Reset</b>	0x00							
<b>Access Type</b>	Write, Read							

**PILOT\_PA**

The purpose of PILOT\_PA<7:0> is to set the LED power during the PROX mode, as well as in Multi-LED mode. These bits set the nominal drive current for the pilot mode as shown in the table below.

When LED x is used, the respective LEDx\_RGE is used to control the range of the LED driver in conjunction with PILOT\_PA. For instance, if LED1 is used in the PILOT mode, then, LED1\_RGE together with PILOT\_PA will be used to set the LED1 current.

LEDX_RGE[1:0]	00	01	10	11
PILOT_PA[7:0]	LED Current (mA)	LED Current (mA)	LED Current (mA)	LED Current (mA)
00000000	0.00	0.00	0.00	0.00
00000001	0.12	0.24	0.36	0.48
00000010	0.24	0.48	0.73	0.97
00000011	0.36	0.73	1.09	1.45
.....				
11111100	30.6	61.3	91.9	122.5
11111101	30.8	61.5	92.3	123.0
11111110	30.9	61.8	92.6	123.5
11111111	31.0	62.0	93.0	124.0
LSB	0.12	0.24	0.36	0.48

**LED Range 1 (0x2A)**

BIT	7	6	5	4	3	2	1	0
Field	–	–	LED3_RGE[1:0]		LED2_RGE[1:0]		LED1_RGE[1:0]	
Reset	–	–	0x00		0x00		0x00	
Access Type	–	–	Write, Read		Write, Read		Write, Read	

**LED3\_RGE**

Range selection of the LED current. Refer to *LED1\_PA[7:0]* for more details.

LEDX_RGE[1:0] (X = 1 TO 6)	LED CURRENT(mA)
00	31
01	62
10	93
11	124

**LED2\_RGE**

Range selection of the LED current. Refer to *LED3\_RGE[1:0]* for more details.

**LED1\_RGE**

Range selection of the LED current. Refer to *LED3\_RGE[1:0]* for more details.

**LED Range 2 (0x2B)**

BIT	7	6	5	4	3	2	1	0
Field	–	–	LED6_RGE[1:0]		LED5_RGE[1:0]		LED4_RGE[1:0]	
Reset	–	–	0x00		0x00		0x00	
Access Type	–	–	Write, Read		Write, Read		Write, Read	

**LED6\_RGE**

Range selection of the LED current. Refer to *LED3\_RGE[1:0]* for more details.

**LED5\_RGE**

Range selection of the LED current. Refer to *LED3\_RGE[1:0]* for more details.

**LED4\_RGE**

Range selection of the LED current. Refer to *LED3\_RGE[1:0]* for more details.

**S1 HI RES DAC1 (0x2C)**

BIT	7	6	5	4	3	2	1	0
Field	S1_HRES_DAC1_OVR	–	S1_HRES_DAC1[5:0]					
Reset	0x0	–	0x00					
Access Type	Write, Read	–	Write, Read					

**S1\_HRES\_DAC1\_OVR**

VALUE	ENUMERATION	DECODE
0	OFF	The high resolution DAC for PPG1 is controlled by the chip.
1	ON	This allows the high-resolution DAC for PPG1 used in exposure 1 to be controlled by the software.

**S1\_HRES\_DAC1**

If S1\_HI\_RES\_DAC1\_OVR = 1, then S1\_HRES\_DAC1 sets the high-resolution DAC code used in PPG1 ADC. This allows the algorithm to control ADC subranging.

If S1\_HI\_RES\_DAC1\_OVR = 0, then S1\_HRES\_DAC1 has no effect on the PPG1 ADC.

**S2 HI RES DAC1 (0x2D)**

BIT	7	6	5	4	3	2	1	0
Field	S2_HRES_DAC1_OVR	–	S2_HRES_DAC1[5:0]					
Reset	0x0	–	0x00					
Access Type	Write, Read	–	Write, Read					

**S2\_HRES\_DAC1\_OVR**

VALUE	ENUMERATION	DECODE
0	OFF	The high resolution DAC for PPG1 is controlled by the chip.
1	ON	This allows the high-resolution DAC for PPG1 used in exposure 2 to be controlled by the software.

**S2\_HRES\_DAC1**

If S2\_HI\_RES\_DAC1\_OVR = 1, then S2\_HRES\_DAC1 sets the high-resolution DAC code used in PPG1 ADC. This allows the algorithm to control ADC subranging.

If S2\_HI\_RES\_DAC1\_OVR = 0, then S2\_HRES\_DAC1 has no effect on the PPG1 ADC.

**S3 HI RES DAC1 (0x2E)**

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	S3_HRES_DAC1_OVR	–	S3_HRES_DAC1[5:0]					
<b>Reset</b>	0x0	–	0x0					
<b>Access Type</b>	Write, Read	–	Write, Read					

**S3\_HRES\_DAC1\_OVR**

VALUE	ENUMERATION	DECODE
0	OFF	The high resolution DAC for PPG1 is controlled by the chip
1	ON	This allows the high-resolution DAC for PPG1 used in exposure 3 to be controlled by the software.

**S3\_HRES\_DAC1**

If S3\_HI\_RES\_DAC1\_OVR = 1 then S3\_HRES\_DAC1 sets the high-resolution DAC code used in PPG1 ADC. This allows the algorithm to control ADC subranging.

If S3\_HI\_RES\_DAC1\_OVR = 0 then S3\_HRES\_DAC1 has no effect on the PPG1 ADC.

**S4 HI RES DAC1 (0x2F)**

BIT	7	6	5	4	3	2	1	0
<b>Field</b>	S4_HRES_DAC1_OVR	–	S4_HRES_DAC1[5:0]					
<b>Reset</b>	0b0	–	0x0					
<b>Access Type</b>	Write, Read	–	Write, Read					

**S4\_HRES\_DAC1\_OVR**

VALUE	ENUMERATION	DECODE
0	OFF	The high resolution DAC for PPG1 is controlled by the chip.
1	ON	This allows the high-resolution DAC for PPG1 used in exposure 4 to be controlled by the software.

**S4\_HRES\_DAC1**

If S4\_HI\_RES\_DAC1\_OVR = 1 then S4\_HRES\_DAC1 sets the high-resolution DAC code used in PPG1 ADC. This allows the algorithm to control ADC subranging.

If S4\_HI\_RES\_DAC1\_OVR = 0 then S4\_HRES\_DAC1 has no effect on the PPG1 ADC.



**S5 HI RES DAC1 (0x30)**

BIT	7	6	5	4	3	2	1	0
Field	S5_HRES_DAC1_OVR	–	S5_HRES_DAC1[5:0]					
Reset	0b0	–	0x0					
Access Type	Write, Read	–	Write, Read					

**S5\_HRES\_DAC1\_OVR**

VALUE	ENUMERATION	DECODE
0	OFF	The high-resolution DAC for PPG1 is controlled by the chip.
1	ON	This allows the high-resolution DAC for PPG1 used in exposure 5 to be controlled by the software.

**S5\_HRES\_DAC1**

If S5\_HI\_RES\_DAC1\_OVR = 1, then S5\_HRES\_DAC1 sets the high-resolution DAC code used in PPG1 ADC. This allows the algorithm to control ADC subranging.

If S5\_HI\_RES\_DAC1\_OVR = 0, then S5\_HRES\_DAC1 has no effect on the PPG1 ADC.

**S6 HI RES DAC1 (0x31)**

BIT	7	6	5	4	3	2	1	0
Field	S6_HRES_DAC1_OVR	–	S6_HRES_DAC1[5:0]					
Reset	0b0	–	0x0					
Access Type	Write, Read	–	Write, Read					

**S6\_HRES\_DAC1\_OVR**

VALUE	ENUMERATION	DECODE
0	OFF	The high-resolution DAC for PPG1 is controlled by the chip.
1	ON	This allows the high-resolution DAC for PPG1 used in exposure 6 to be controlled by the software.

**S6\_HRES\_DAC1**

If S6\_HI\_RES\_DAC1\_OVR = 1, then S6\_HRES\_DAC1 sets the high-resolution DAC code used in PPG1 ADC. This allows the algorithm to control ADC subranging.

If S6\_HI\_RES\_DAC1\_OVR = 0, then S6\_HRES\_DAC1 has no effect on the PPG1 ADC.

**S1\_HI RES DAC2 (0x32)**

BIT	7	6	5	4	3	2	1	0
Field	S1_HRES_DAC2_OVR	–	S1_HRES_DAC2[5:0]					
Reset	0x0	–	0x00					
Access Type	Write, Read	–	Write, Read					

**S1\_HRES\_DAC2\_OVR**

VALUE	ENUMERATION	DECODE
0	OFF	The high-resolution DAC for PPG2 is controlled by the chip.
1	ON	This allows the high-resolution DAC for PPG2 used in exposure 1 to be controlled by the software.

**S1\_HRES\_DAC2**

If S1\_HI\_RES\_DAC2\_OVR = 1, then S1\_HRES\_DAC2 sets the high-resolution DAC code used in PPG2 ADC. This allows the algorithm to control ADC subranging.

If S1\_HI\_RES\_DAC2\_OVR = 0, then S1\_HRES\_DAC2 has no effect on the PPG2 ADC.

**S2\_HI RES DAC2 (0x33)**

BIT	7	6	5	4	3	2	1	0
Field	S2_HRES_DAC2_OVR	–	S2_HRES_DAC2[5:0]					
Reset	0x0	–	0x0					
Access Type	Write, Read	–	Write, Read					

**S2\_HRES\_DAC2\_OVR**

VALUE	ENUMERATION	DECODE
0	OFF	The high-resolution DAC for PPG2 is controlled by the chip.
1	ON	This allows the high-resolution DAC for PPG2 used in exposure 2 to be controlled by the software.

**S2\_HRES\_DAC2**

If S2\_HI\_RES\_DAC2\_OVR = 1, then S2\_HRES\_DAC2 sets the high-resolution DAC code used in PPG2 ADC. This allows the algorithm to control ADC subranging.

If S2\_HI\_RES\_DAC2\_OVR = 0, then S2\_HRES\_DAC2 has no effect on the PPG2 ADC.

**S3\_HI RES DAC2 (0x34)**

BIT	7	6	5	4	3	2	1	0
Field	S3_HRES_DAC2_OVR	–	S3_HRES_DAC2[5:0]					
Reset	0b0	–	0x0					
Access Type	Write, Read	–	Write, Read					

**S3\_HRES\_DAC2\_OVR**

VALUE	ENUMERATION	DECODE
0	OFF	The high-resolution DAC for PPG2 is controlled by the chip.
1	ON	This allows the high-resolution DAC for PPG2 used in exposure 3 to be controlled by the software.

**S3\_HRES\_DAC2**

If S3\_HI\_RES\_DAC2\_OVR = 1, then S3\_HRES\_DAC2 sets the high-resolution DAC code used in PPG2 ADC. This allows the algorithm to control ADC subbranging.

If S3\_HI\_RES\_DAC2\_OVR = 0 then S3\_HRES\_DAC2 has no effect on the PPG2 ADC.

**S4 HI RES DAC2 (0x35)**

BIT	7	6	5	4	3	2	1	0
Field	S4_HRES_DAC2_OVR	–	S4_HRES_DAC2[5:0]					
Reset	0b0	–	0x0					
Access Type	Write, Read	–	Write, Read					

**S4\_HRES\_DAC2\_OVR**

VALUE	ENUMERATION	DECODE
0	OFF	The high-resolution DAC for PPG2 is controlled by the chip.
1	ON	This allows the high-resolution DAC for PPG2 used in exposure 4 to be controlled by the software.

**S4\_HRES\_DAC2**

If S4\_HI\_RES\_DAC2\_OVR = 1, then S4\_HRES\_DAC2 sets the high resolution DAC code used in PPG2 ADC. This allows the algorithm to control ADC subbranging.

If S4\_HI\_RES\_DAC2\_OVR = 0, then S4\_HRES\_DAC2 has no effect on the PPG2 ADC.

**S5 HI RES DAC2 (0x36)**

BIT	7	6	5	4	3	2	1	0
Field	S5_HRES_DAC2_OVR	–	S5_HRES_DAC2[5:0]					
Reset	0b0	–	0x0					
Access Type	Write, Read	–	Write, Read					

**S5\_HRES\_DAC2\_OVR**

VALUE	ENUMERATION	DECODE
0	OFF	The high-resolution DAC for PPG2 is controlled by the chip.
1	ON	This allows the high-resolution DAC for PPG2 used in exposure 5 to be controlled by the software.

**S5\_HRES\_DAC2**

If S5\_HI\_RES\_DAC2\_OVR = 1, then S5\_HRES\_DAC2 sets the high-resolution DAC code used in PPG2 ADC. This allows the algorithm to control ADC subbranging.

If S5\_HI\_RES\_DAC2\_OVR = 0, then S5\_HRES\_DAC2 has no effect on the PPG2 ADC.

**S6 HI RES DAC2 (0x37)**

BIT	7	6	5	4	3	2	1	0
Field	S6_HRES_DAC2_OVR	–	S6_HRES_DAC2[5:0]					
Reset	0b0	–	0x0					
Access Type	Write, Read	–	Write, Read					

**S6\_HRES\_DAC2\_OVR**

VALUE	ENUMERATION	DECODE
0	OFF	The high-resolution DAC for PPG2 is controlled by the chip.
1	ON	This allows the high-resolution DAC for PPG2 used in exposure 6 to be controlled by the software.

**S6\_HRES\_DAC2**

If S6\_HI\_RES\_DAC\_2OVR = 1, then S6\_HRES\_DAC2 sets the high-resolution DAC code used in PPG2 ADC. This allows the algorithm to control ADC subranging.

If S6\_HI\_RES\_DAC2\_OVR = 0, then S6\_HRES\_DAC2 has no effect on the PPG2 ADC.

**Die Temperature Configuration (0x40)**

BIT	7	6	5	4	3	2	1	0
Field	–	–	–	–	–	–	–	TEMP_EN
Reset	–	–	–	–	–	–	–	0x0
Access Type	–	–	–	–	–	–	–	Write, Read

**TEMP\_EN**

The bit gets cleared after temperature measurement completes.

VALUE	ENUMERATION	DECODE
0	NORMAL	Idle
1	ASSERTED	Start one temperature measurement

**Die Temperature Integer (0x41)**

BIT	7	6	5	4	3	2	1	0
Field	TEMP_INT[7:0]							
Reset	0x0							
Access Type	Read Only							

**TEMP\_INT**

This register stores the integer temperature data in 2s complimnet form. 0x00 = 0°C, 0x7F = 127°C and 0x80 = -128°C

**Note:** TINT and TFRAC registers should be read through the Serial Interface in burst mode, to ensure that they belong to the same sample.

**Die Temperature Fraction (0x42)**

BIT	7	6	5	4	3	2	1	0
Field	–	–	–	–	TEMP_FRAC[3:0]			
Reset	–	–	–	–	0x0			
Access Type	–	–	–	–	Read Only			

**TEMP\_FRAC**

This register store the fractional temperature data in increments of 0.0625°C. 0x1 = 0.0625°C and 0xF = 0.9375°C.

**Note:** TINT and TFRAC registers should be read through the Serial Interface in burst mode, to ensure that they belong to the same sample.

**SHA Command (0xF0)**

BIT	7	6	5	4	3	2	1	0
Field	SHA_CMD[7:0]							
Reset	0x0							
Access Type	Write, Read							

**SHA\_CMD**

VALUE	DECODE
0x35	MAC WITH ROM ID
0x36	MAC WITHOUT ROM ID
OTHERS	RESERVED

**SHA Configuration (0xF1)**

BIT	7	6	5	4	3	2	1	0
Field	–	–	–	–	–	–	SHA_EN	SHA_START
Reset	–	–	–	–	–	–	0x0	0x0
Access Type	–	–	–	–	–	–	Write, Read	Write, Read

**SHA\_EN**

Authentication is performed using a FIPS 180-3 compliant SHA-256 one-way hash algorithm on a 512-bit message block. The message block consists of a 160-bit secret, a 160-bit challenge and 192 bits of constant data. Optionally, the 64-bit ROM ID replaces 64 of the 192 bits of constant data used in the hash operation. 16 bits out of the 160-bit secret and 16 bits of ROM ID are programmable—8 bits each in metal and 8 bits each in OTP bits.

The host and the MAX86140 both calculate the result based on a mutually known secret. The result of the hash operation is known as the message authentication code (MAC) or message digest. The MAC is returned by the MAX86140 for comparison with the host’s MAC. Note that the secret is never transmitted on the bus and thus cannot be captured by observing bus traffic. Each authentication attempt is initiated by the host system by writing a 160-bit random challenge into the SHA memory address space 0x00h to 0x09h. The host then issues the compute MAC or compute MAC with ROM ID command. The MAC is computed per FIPS 180-3, and stored in address space 0x00h to 0x0Fh overwriting the challenge value.

Note that the results of the authentication attempt are determined by host verification. Operation of the MAX86140 is not affected by authentication success or failure.

Sequence of operation is as follows:

- Enable SHA\_DONE Interrupt.
- Enable SHA\_EN bit.
- Write 160-bit random challenge value to RAM using registers MEM\_IDX and MEM\_DATA.
- Write command, with ROM ID (0x35) or without ROM ID (0x36), to SHA\_CMD register.
- Write 1 to SHA\_START and 1 to SHA\_EN bit.
- Wait for SHA\_DONE interrupt.
- Read 256 MAC value from RAM using registers MEM\_IDX and MEM\_DATA.
- Compare MAC from MAX86140 with Host's precalculated MAC.
- Check PASS or FAIL.
- Disable SHA\_EN bit ( Write 0 to SHA\_EN bit).

VALUE	ENUMERATION	DECODE
0	NORMAL	Authentication is disabled
1	ASSERTED	Authentication is enabled

### SHA\_START

The bit gets cleared after authentication completes. The valid command (0x35 or 0x36) should be written to the SHA\_CMD register and challenge value should be written to the RAM by Host before writing 1 to this bit.

### Memory Control (0xF2)

BIT	7	6	5	4	3	2	1	0
Field	–	–	–	–	–	–	MEM_WR_EN	BANK_SEL
Reset	–	–	–	–	–	–	0x0	0x0
Access Type	–	–	–	–	–	–	Write, Read	Write, Read

### MEM\_WR\_EN

Enable write access to Memory through SPI.

VALUE	ENUMERATION	DECODE
0	DISABLED	Writing to memory through SPI is disabled.
1	ENABLED	Writing to memory through SPI is enabled

### BANK\_SEL

Selects the memory bank for reading and writing.

Burst reading or writing the memory past 0xFF automatically increments BANK\_SEL to 1.

VALUE	DECODE
0	Select Bank 0, address 0x00 to 0xFF
1	Select Bank 1, address 0x100 to 0x17f

**Memory Index (0xF3)**

<b>BIT</b>	<b>7</b>	<b>6</b>	<b>5</b>	<b>4</b>	<b>3</b>	<b>2</b>	<b>1</b>	<b>0</b>
<b>Field</b>	MEM_IDX[7:0]							
<b>Reset</b>	0x0							
<b>Access Type</b>	Write, Read							

**MEM\_IDX**

Index to Memory for reading and writing. The memory is 384 bytes, and is divided into two banks - Bank 0 from 0x00 to 0xFF and Bank 1 is from 0x100 to 0x17F. The bank is selected by the BANK\_SEL register bit. MEM\_IDX is the starting address for burst writing to or reading from memory. Burst accessing the memory past 0xFF accesses Bank 1. The memory address saturates at 0x17F.

**Memory Data (0xF4)**

<b>BIT</b>	<b>7</b>	<b>6</b>	<b>5</b>	<b>4</b>	<b>3</b>	<b>2</b>	<b>1</b>	<b>0</b>
<b>Field</b>	MEM_DATA[7:0]							
<b>Reset</b>	0x0							
<b>Access Type</b>	Write, Read, Dual							

**MEM\_DATA**

Data to be written or data read from Memory

Reading this register does not automatically increment the register address. So burst reading this register read the same register over and over, but the address to the Memory autoincrements until BANK\_SEL becomes 1 and MEM\_IDX becomes 0x7F.

**Part ID (0xFF)**

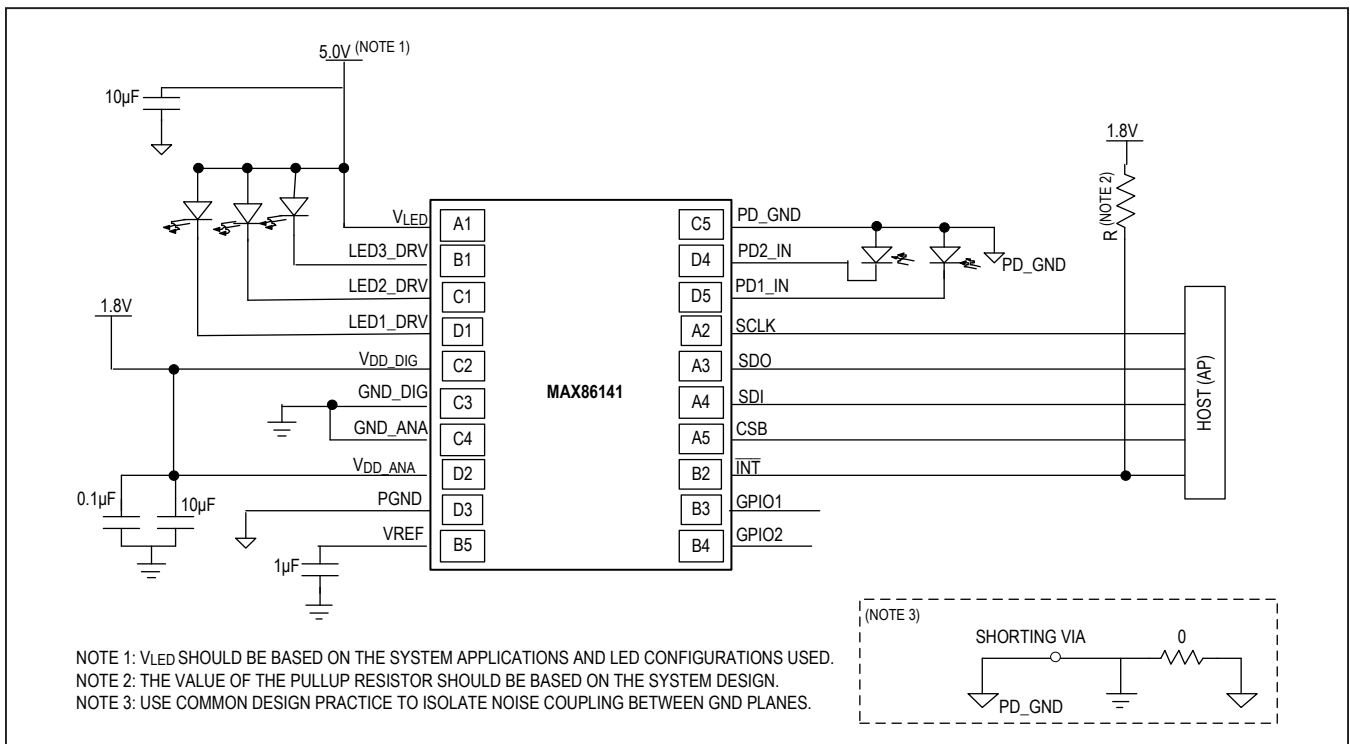
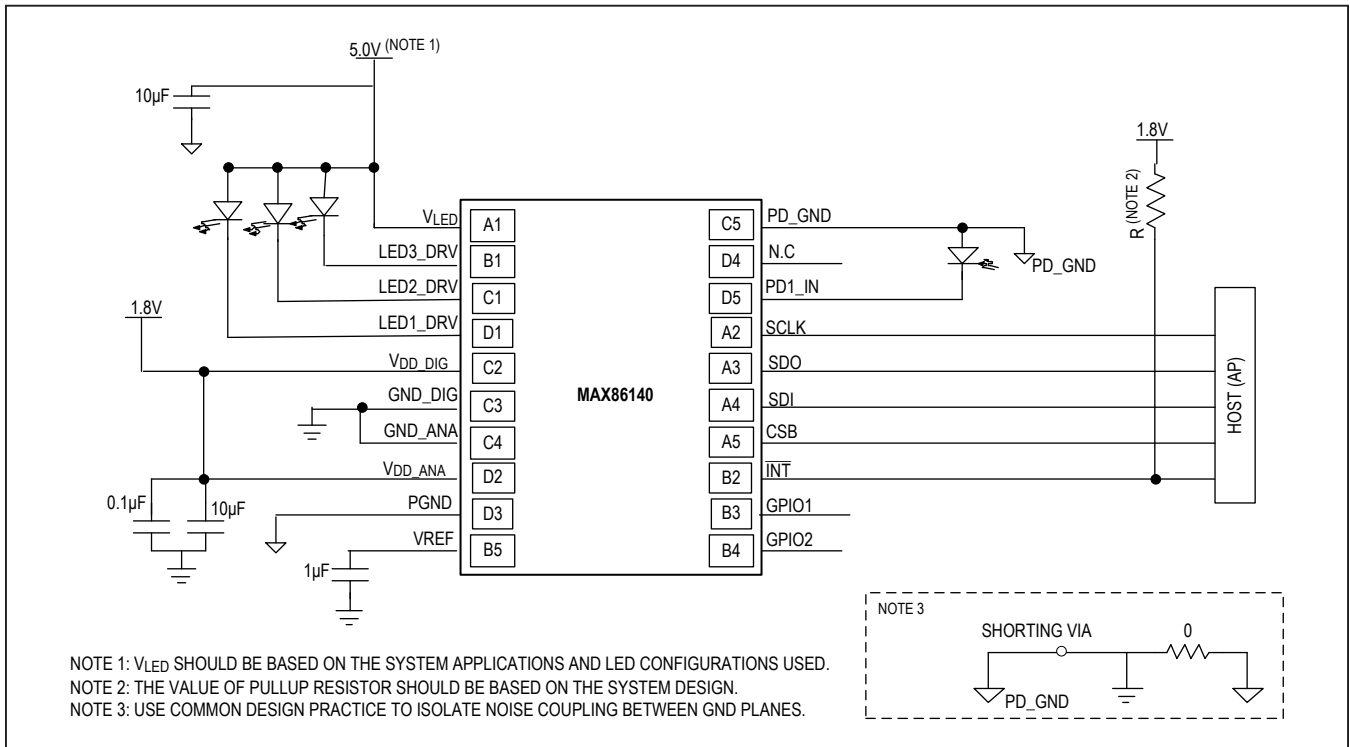
<b>BIT</b>	<b>7</b>	<b>6</b>	<b>5</b>	<b>4</b>	<b>3</b>	<b>2</b>	<b>1</b>	<b>0</b>
<b>Field</b>	PART_ID[7:0]							
<b>Reset</b>	0xFF							
<b>Access Type</b>	Read Only							

**PART\_ID**

This register stores the part identifier for the chip.

<b>PART_ID</b>	<b>MAX #</b>	<b># OF PPG CHANNELS</b>
0x24	MAX86140	1
0x25	MAX86141	2

Typical Application Circuits





MAX86140/  
MAX86141

Best-in-Class Optical Pulse Oximeter and  
Heart-Rate Sensor for Wearable Health

## Ordering Information

PART NUMBER	TEMP RANGE	PIN-PACKAGE	CONFIGURATION
MAX86140ENP+	-40°C to +85°C	20-pin WLP, 2.048mm x 1.848mm, 5 x 4, 0.4mm ball pitch	Single-Channel Optical AFE
MAX86141ENP+	-40°C to +85°C	20-pin WLP, 2.048mm x 1.848mm, 5 x 4, 0.4mm ball pitch	Dual-Channel Optical AFE

+Denotes a lead(Pb)-free/RoHS-compliant package.

T = Tape and reel.

## Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	5/17	Initial release	—
1	8/17	Added MAX86141 part number to data sheet	1–88
2	8/19	Updated the <i>General Description, Benefits and Features, Simplified Block Diagram, Absolute Maximum Ratings, Pin Description, Detailed Description, Optical Subsystem, Table 3, FIFO Data (Register 0x08), FIFO_A_FULL (address 0x09), A_FULL_TYPE (Address 0x0A), GPIO Configuration, Picket Fence Detect-and-Replace Function, Figure 31, Photodiode Biasing, Layout Guidelines, Table 16, SPI FIFO Burst Mode Read Transaction, VDD_OOR Bit, PPG2_ADC_RGE table, PPG1_ADC_RGE table, PPG_TINT table, both PPG_SR tables, SMP_AVE table, LED_SETLNG table, DIG_FILT_SEL table, BURST_RATE table, PDBIAS2 table, PDBIAS1 table, PF_ENABLE table, IIR_TC table, IIR_INIT_VALUE table, THRESHOLD_SIGMA_MULT table, LED1_DRV table, PILOT_PA table, LED3_RGE table, S1_HRES_DAC1 description, S2_HRES_DAC1 description, S3_HRES_DAC1 description, S4_HRES_DAC1 description, S5_HRES_DAC1 description, S6_HRES_DAC1 description, TEMP_EN table, SHA_CMD table, SHA_EN table, MEM_WR_EN table, and BANK_SEL table; replaced the <i>Typical Operating Characteristics, Pin Configurations, Table 6, Pseudo-Code Example of Initializing the Optical AFE, Pseudo-Code for Interrupt Handling with FIFO_A_FULL, Pseudo-Code Example of Reading Data from FIFO, Figure 7, Figure 11, Figure 12, Figure 14, Figure 15, Figure 16, Figure 17, Figure 18, Figure 20, Figure 21, Figure 22, Figure 24, Figure 25, Figure 26, Figure 27, Figure 29, and Typical Application Circuit, adding additional Typical Application Circuits; added a Detailed Block Diagram, new Figures 25-27 (after renumbering), GPIO CTRL[3:0] 1010 Hardware Sync and Table 15 (and renumbered remaining tables); deleted Figure 9, Figure 13, Figure 19, Figure 23, Figure 28, and renumbered existing figures</i></i>	1-15, 17-23, 28-44, 46-49, 53-55, 59, 68-74, 76-86, 88

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(Применяются в телекоммуникациях гражданского и специального назначения, в средствах связи, РЛС, а так же военной, авиационной и аэрокосмической отраслях промышленности).



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