General Description

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The 874328I-01 is a high-performance differential ÷1 and ÷4 clock divider and fanout buffer. The device is designed for the frequency-division and signal fanout of high-frequency, low phase-noise clock signals. The differential input signal is frequency divided by \div 1 and \div 4. Three LVPECL and three LVDS output banks are provided with a total of twenty differential outputs. The 874328I-01 is characterized to operate from a 2.5V power supply. Guaranteed output-to-output and part-to-part skew characteristics make the 874328I-01 ideal for those clock distribution applications demanding well-defined performance and repeatability.

Pin Assignment

Features

- **•** One differential input LVPECL reference clock
- **•** Differential pair can accept the following differential input levels: LVPECL, LVDS, CML, SSTL
- **•** Integrated input termination resistors
- **•** One bank of three LVPECL outputs (÷1 frequency-divided)
- **•** One bank of three LVPECL outputs (÷4 frequency-divided)
- **•** One bank of two LVPECL outputs (÷4 frequency-divided)
- **•** Two banks of three LVDS outputs (÷4 frequency-divided)
- **•** One bank of six LVDS outputs (÷4 frequency-divided)
- **•** Total of twenty differential clock outputs
- **•** Maximum input frequency: 650MHz
- **•** Maximum output frequency: 650MHz (÷1 outputs)
- **•** Maximum output frequency: 162.5MHz (÷4 outputs)
- **•** LVCMOS interface levels for all control inputs
- **•** Output skew: 70ps (maximum)
- **•** Part-to-part skew: 250ps (maximum)
- **•** Full 2.5V supply voltage
- **•** Available in lead-free (RoHS 6) package
- **•** -40°C to 85°C ambient operating temperature

Block Diagram

Table 1. Pin Descriptions

NOTE: *Pullup* refers to internal input resistors. See Table 2, *Pin Characteristics,* for typical values.

Table 2. Pin Characteristics

Function Tables

Table 3A. OEA Configuration Table

NOTE: OEA is an asynchronous control.

Table 3B. OEB Configuration Table

NOTE: OEB is an asynchronous control.

Table 3C. OEC Configuration Table

NOTE: OEC is an asynchronous control.

Table 3D. OED Configuration Table

NOTE: OED is an asynchronous control.

Table 3E. OEE Configuration Table

Table 3F. OEF Configuration Table

NOTE: OEF is an asynchronous control

NOTE: OEE is an asynchronous control

Table 3G. CLK_EN Mode Configuration Table

NOTE: CLK_EN is synchronous to the falling edge of the input clock.

Absolute Maximum Ratings

NOTE: Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics or AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

DC Electrical Characteristics

Table 4A. Power Supply DC Characteristics, $V_{CC} = V_{CCOA} = V_{CCOB} = V_{CCODE} = V_{CCODE} = V_{CCOF} = 2.5V \pm 5\%$ **,** $V_{EE} = 0V$ **,** $T_A = -40^\circ \text{C}$ to 85°C

V_{CCOX} denotes V_{CCOA,} V_{CCOB,} V_{CCOC}, V_{CCODE}, V_{CCOF}. I_{CCOX} denotes I_{CCODE, IccoF}.

Table 4B. LVCMOS/LVTTL Input DC Characteristics, V_{CC} **= 2.5V** \pm **5%,** T_A **= -40°C to 85°C**

Table 4D. Differential LVPECL Input DC Characteristics, $V_{CC} = 2.5V \pm 5\%$, $T_A = -40\degree$ C to 85 \degree C

NOTE 1: V_{IL} should not be less than -0.3V.

NOTE 2: Guaranteed by design.

Table 4E. LVPECL DC Characteristics, $V_{CC} = V_{CCOA} = V_{CCOB} = V_{CCOC} = 2.5V \pm 5\%$ **,** $V_{EE} = 0V$ **,** $T_A = -40^{\circ}C$ **to 85°C**

NOTE 1: Outputs terminated with 50 Ω to $V_{CCO} - 2V$.

Table 4F. LVDS DC Characteristics, $V_{CC} = V_{CCODE} = V_{CCOF} = 2.5V \pm 5\%$ **,** $T_A = -40^{\circ}C$ **to 85°C**

Table 5. AC Electrical Characteristics, $V_{CC} = V_{CCOA} = V_{CCOB} = V_{CCOC} = 0$ **=** $V_{CCODE} = V_{CCOF} = 2.5V \pm 5\%$ **,** $V_{EE} = 0V$ **,** $T_A = -40^\circ \text{C}$ to 85 $^\circ \text{C}$

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE 1: Measured from the differential input crossing point to the differential output crossing point.

NOTE 2: Defined as skew within a bank of outputs at the same supply voltage and with equal load conditions.

NOTE 3: This parameter is defined in accordance with JEDEC Standard 65.

NOTE 4: Defined as skew between outputs at the same supply voltage and with equal load conditions.

Measured at the differential cross points.

NOTE 5: Defined as skew between outputs on different devices operating at the same supply voltage, same temperature, same frequency and with equal load conditions. Using the same type of inputs on each device, the outputs are measured at the differential cross points.

Additive Phase Jitter

The spectral purity in a band at a specific offset from the fundamental compared to the power of the fundamental is called the *dBc Phase Noise.* This value is normally expressed using a Phase noise plot and is most often the specified plot in many applications. Phase noise is defined as the ratio of the noise power present in a 1Hz band at a specified offset from the fundamental frequency to the power value of the fundamental. This ratio is expressed in decibels (dBm) or a ratio of the power in the 1Hz band to the power in the

fundamental. When the required offset is specified, the phase noise is called a *dBc* value, which simply means dBm at a specified offset from the fundamental. By investigating jitter in the frequency domain, we get a better understanding of its effects on the desired application over the entire time record of the signal. It is mathematically possible to calculate an expected bit error rate given a phase noise plot.

As with most timing specifications, phase noise measurements has issues relating to the limitations of the equipment. Often the noise floor of the equipment is higher than the noise floor of the device. This is illustrated above. The device meets the noise floor of what is shown, but can actually be lower. The phase noise is dependent on the input source and measurement equipment.

The source generator "IFR2042 10kHz – 56.4GHz Low Noise Signal Generator as external input to an Agilent 8133A 3GHz Pulse Generator.

Parameter Measurement Information

LVPECL Output Load AC Test Circuit

Differential Input Level

Output Skew

LVDS Output Load AC Test Circuit

Part-to-Part Skew

Parameter Measurement Information, continued

LVPECL Output Rise/Fall Time

Output Duty Cycle/Pulse Width/Period

Offset Voltage Setup

LVDS Output Rise/Fall Time

Differential Output Voltage Setup

Application Information

Recommendations for Unused Input and Output Pins

Inputs:

LVCMOS Control Pins

All control pins have internal pullups; additional resistance is not required but can be added for additional protection. A 1k Ω resistor can be used.

Outputs:

LVPECL Outputs

All unused LVPECL outputs can be left floating. We recommend that there is no trace attached. Both sides of the differential output pair should either be left floating or terminated.

LVDS Outputs

All unused LVDS outputs should be terminated with 100 Ω resistor between the differential pair.

Differential LVPECL Clock Input Interface

The CLK/nCLK accepts LVPECL, LVDS, CML, SSTL and other differential signals. Both V_{SWING} and V_{OH} must meet the V_{PP} and V_{CMR} input requirements. *Figures 2A to 2E* show interface examples for the CLK/nCLK input driven by the most common driver types. The

Figure 2A. CLK/nCLK Input with Built-In 50 Driven by a CML Driver

Figure 2C. CLK/nCLK Input with Built-In 50 Driven by an LVPECL Driver

Figure 2E. CLK/nCLK Input with Built-In 50 Driven by an SSTL Driver

input interfaces suggested here are examples only. If the driver is from another vendor, use their termination recommendation. Please consult with the vendor of the driver component to confirm the driver termination requirements.

Figure 2B. CLK/nCLK Input with Built-In 50 Driven by a CML Driver with Built-In 50 Pullup

Figure 2D. CLK/nCLK Input with Built-In 50 Driven by an LVDS Driver

EPAD Thermal Release Path

In order to maximize both the removal of heat from the package and the electrical performance, a land pattern must be incorporated on the Printed Circuit Board (PCB) within the footprint of the package corresponding to the exposed metal pad or exposed heat slug on the package, as shown in *Figure 3.* The solderable area on the PCB, as defined by the solder mask, should be at least the same size/shape as the exposed pad/slug area on the package to maximize the thermal/electrical performance. Sufficient clearance should be designed on the PCB between the outer edges of the land pattern and the inner edges of pad pattern for the leads to avoid any shorts.

While the land pattern on the PCB provides a means of heat transfer and electrical grounding from the package to the board through a solder joint, thermal vias are necessary to effectively conduct from the surface of the PCB to the ground plane(s). The land pattern must be connected to ground through these vias. The vias act as "heat pipes". The number of vias (i.e. "heat pipes") are application specific and dependent upon the package power dissipation as well as electrical conductivity requirements. Thus, thermal and electrical analysis and/or testing are recommended to determine the minimum number needed. Maximum thermal and electrical performance is achieved when an array of vias is incorporated in the land pattern. It is recommended to use as many vias connected to ground as possible. It is also recommended that the via diameter should be 12 to 13mils (0.30 to 0.33mm) with 1oz copper via barrel plating. This is desirable to avoid any solder wicking inside the via during the soldering process which may result in voids in solder between the exposed pad/slug and the thermal land. Precautions should be taken to eliminate any solder voids between the exposed heat slug and the land pattern. Note: These recommendations are to be used as a guideline only. For further information, refer to the Application Note on the *Surface Mount Assembly* of Amkor's Thermally/Electrically Enhance Leadframe Base Package, Amkor Technology.

Figure 3. Assembly for Exposed Pad Thermal Release Path - Side View (drawing not to scale)

LVDS Driver Termination

A general LVDS interface is shown in *Figure 4.* Standard termination for LVDS type output structure requires both a 100 Ω parallel resistor at the receiver and a 100 Ω differential transmission line environment. In order to avoid any transmission line reflection issues, the 100 Ω resistor must be placed as close to the receiver as possible. IDT offers a full line of LVDS compliant devices with two types of output structures: current source and voltage source. The

standard termination schematic as shown in Figure 4 can be used with either type of output structure. If using a non-standard termination, it is recommended to contact IDT and confirm if the output is a current source or a voltage source type structure. In addition, since these outputs are LVDS compatible, the input receivers amplitude and common mode input range should be verified for compatibility with the output.

Figure 4. Typical LVDS Driver Termination

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Termination for LVPECL Outputs

Figure 5A and *Figure 5B* show examples of termination for 2.5V LVPECL driver. These terminations are equivalent to terminating 50 Ω to V $_{\rm CC}$ – 2V. For V $_{\rm CC}$ = 2.5V, the V $_{\rm CC}$ – 2V is very close to ground

Figure 5A. 2.5V LVPECL Driver Termination Example

Figure 5C. 2.5V LVPECL Driver Termination Example

level. The R3 in Figure 5B can be eliminated and the termination is shown in *Figure 5C.*

Figure 5B. 2.5V LVPECL Driver Termination Example

Power Considerations (typical)

This section provides information on power dissipation and junction temperature for the 874328I-01. Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the 874328I-01 is the sum of the core power plus the power dissipation in the load(s). The following is the power dissipation for $V_{CC} = 2.5V + 5\% = 2.625V$, which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipation in the load.

- Power (core)_{TYP} = $V_{CC~TYP}$ * $I_{CC~TYP}$ = 2.5V * 47mA = 117.5mW
- Power(LVDS)_{TYP} = $V_{\text{CCO_DEF_TYP}}$ * $I_{\text{CCO_DEF_TYP}}$ = 2.5V * 139mA = 347.5mW
- Power (LVPECL) = **29.4mW/Loaded Output pair** If all outputs are loaded, the total power is 8 * 29.4mW = **235.2mW**
- Power Dissipation for internal termination R_T Power $(R_T)_{TYP} = (V_{IN_TYP})^2 / R_{T_TYP} = (0.675V)^2 / 100 \Omega = 4.56$ mW

Total Power_{_TYP} = 117.5mW + 347.5mW + 4.56mW + 235.2mW = **704.76mW**

2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad, and directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature, Tj, to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for Tj is as follows: Tj = θ_{IA} * Pd_total + T_A

 Tj = Junction Temperature

 θ_{JA} = Junction-to-Ambient Thermal Resistance

Pd_total = Total Device Power Dissipation (example calculation is in section 1 above)

 T_A = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance θ_{JA} must be used. Assuming no air flow and a multi-layer board, the appropriate value is 31.8°C/W per Table 6 below.

Therefore, Tj for an ambient temperature of 85°C with all outputs switching is:

 85° C + 0.705W $*$ 31.8°C/W = 107°C. This is below the limit of 125°C.

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

Table 6. Thermal Resistance $\theta_{\rm IA}$ **for 64 Lead TQFP, E-Pad, Forced Convection**

3. Calculations and Equations.

The purpose of this section is to calculate the power dissipation for the LVPECL output pair.

The LVPECL output driver circuit and termination are shown in *Figure 6.*

Figure 6. LVPECL Driver Circuit and Termination

To calculate worst case power dissipation into the load, use the following equations which assume a 50Ω load, and a termination voltage of V_{CC} – 2V.

- For logic high, $V_{\text{OUT}} = V_{\text{OH}}$ MAX = V_{CC} MAX 0.8V $(V_{CC~MAX} - V_{OH~MAX}) = 0.8V$
- \bullet For logic low, $\mathsf{V}_{\mathsf{OUT}}$ = $\mathsf{V}_{\mathsf{OL_MAX}}$ = $\mathsf{V}_{\mathsf{CC_MAX}}$ $-$ **1.7V** (VCC_MAX – VOL_MAX) = **1.7V**

Pd_H is power dissipation when the output drives high.

Pd_L is the power dissipation when the output drives low.

 $Pd_H = [(V_{OH_MAX} - (V_{CC_MAX} - 2V))/R_L] * (V_{CC_MAX} - V_{OH_MAX}) = [(2V - (V_{CC_MAX} - V_{OH_MAX})/(R_L)] * (V_{CC_MAX} - V_{OH_MAX})]$ $[(2V - 0.8V)/50_{\Omega}] * 0.8V = 19.2$ mW

Pd_L = [(V_{OL_MAX} – (V_{CC_MAX} – 2V))/R_L] * (V_{CC_MAX} – V_{OL_MAX}) = [(2V – (V_{CC_MAX} – V_{OL_MAX}))/R_{L]} * (V_{CC_MAX} – V_{OL_MAX}) = [(2V – 1.7V)/50] * 1.7V = **10.2mW**

Total Power Dissipation per output pair = Pd_H + Pd_L = **29.4mW**

Power Considerations (maximum)

This section provides information on power dissipation and junction temperature for the 874328I-01. Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the 874328I-01 is the sum of the core power plus the power dissipation in the load(s). The following is the power dissipation for $V_{CC} = 2.5V + 5\% = 2.625V$, which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipation in the load.

- Power (core)_{MAX} = V_{CC} _{MAX} $*$ I_{CC}_{MAX} = 2.625V $*$ 54mA = **141.75mW**
- Power(LVDS)_{MAX} = $V_{\text{CCO_DEF_MAX}}$ ^{*} $I_{\text{CCO_DEF_MAX}}$ = 2.625V * 160mA = 420mW
- Power (LVPECL) = **29.4mW/Loaded Output pair** If all outputs are loaded, the total power is 8 * 29.4mW = **235.2mW**
- Power Dissipation for internal termination R_T Power $(R_T)_{MAX} = (V_{IN_MAX})^2 / R_{T_MIN} = (1.2V)^2 / 80\Omega = 18mW$

Total Power_MAX = 141.75mW + 420mW + 235.2mW + 18mW = **814.95mW**

2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad, and directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature, Tj, to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for Tj is as follows: Tj = θ_{IA} * Pd_total + T_A

 Tj = Junction Temperature

 θ_{JA} = Junction-to-Ambient Thermal Resistance

Pd_total = Total Device Power Dissipation (example calculation is in section 1 above)

 T_A = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance θ_{JA} must be used. Assuming no air flow and a multi-layer board, the appropriate value is 31.8°C/W per Table 7 below.

Therefore, Tj for an ambient temperature of 85°C with all outputs switching is:

 85° C + 0.815W $*$ 31.8°C/W = 111°C. This is below the limit of 125°C.

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

Table 7. Thermal Resistance $\theta_{\rm IA}$ for 64 Lead TQFP, E-Pad, Forced Convection

3. Calculations and Equations.

The purpose of this section is to calculate the power dissipation for the LVPECL output pair.

The LVPECL output driver circuit and termination are shown in *Figure 7.*

Figure 7. LVPECL Driver Circuit and Termination

To calculate worst case power dissipation into the load, use the following equations which assume a 50Ω load, and a termination voltage of V_{CC} – 2V.

- For logic high, $V_{\text{OUT}} = V_{\text{OH_MAX}} = V_{\text{CC_MAX}} 0.8V$ $(V_{CC_MAX} - V_{OH_MAX}) = 0.8V$
- \bullet For logic low, $\mathsf{V}_{\mathsf{OUT}}$ = $\mathsf{V}_{\mathsf{OL_MAX}}$ = $\mathsf{V}_{\mathsf{CC_MAX}}$ $-$ **1.7V** (VCC_MAX – VOL_MAX) = **1.7V**

Pd_H is power dissipation when the output drives high.

Pd_L is the power dissipation when the output drives low.

Pd_H = [(V_{OH_MAX} – (V_{CC_MAX} – 2V))/R_L] * (V_{CC_MAX} – V_{OH_MAX}) = [(2V - (V_{CC_MAX} – V_{OH_MAX}))/R_L] * (V_{CC_MAX} – V_{OH_MAX}) = $[(2V – 0.8V)/50 Ω] * 0.8V = 19.2mW$

 $Pd_L = [(V_{OL_MAX} - (V_{CC_MAX} - 2V))/R_L] * (V_{CC_MAX} - V_{OL_MAX}) = [(2V - (V_{CC_MAX} - V_{OL_MAX}))/R_L] * (V_{CC_MAX} - V_{OL_MAX}) = [(2V - (V_{CC_MAX} - V_{OL_MAX}))/(R_L)] * (V_{CC_MAX} - V_{OL_MAX})]$ $[(2V – 1.7V)/50 Ω] * 1.7V = 10.2mW$

Total Power Dissipation per output pair = Pd_H + Pd_L = **29.4mW**

Reliability Information

Table 8. θ_{JA} vs. Air Flow Table for a 64 Lead TQFP, E-Pad

Transistor Count

The transistor count for 874328I-01 is: 1453

Package Outline and Package Dimensions

Package Outline - Y Suffix for 64 Lead TQFP, E-Pad

Table 9. Package Dimensions for 64 Lead TQFP, E-Pad

Reference Document: JEDEC Publication 95, MS-026

Table 10. Ordering Information

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Revision History

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