

Near Field Communication (NFC) controller

Rev. 3.6 — 28 November 2017 115436 Product data sheet COMPANY PUBLIC

# 1. General description

The PN532 is a highly integrated transceiver module for contactless communication at 13.56 MHz based on the 80C51 microcontroller core. It supports 6 different operating modes:

- ISO/IEC 14443A/MIFARE Reader/Writer
- FeliCa Reader/Writer
- ISO/IEC 14443B Reader/Writer
- ISO/IEC 14443A/MIFARE Card MIFARE Classic 1K or MIFARE Classic 4K card emulation mode
- FeliCa Card emulation
- ISO/IEC 18092, ECMA 340 Peer-to-Peer

The PN532 implements a demodulator and decoder for signals from ISO/IEC 14443A/MIFARE compatible cards and transponders. The PN532 handles the complete ISO/IEC 14443A framing and error detection (Parity & CRC).

The PN532 supports MIFARE Classic 1K or MIFARE Classic 4K card emulation mode. The PN532 supports contactless communication using MIFARE Higher transfer speeds up to 424 kbit/s in both directions.

The PN532 can demodulate and decode FeliCa coded signals. The PN532 handles the FeliCa framing and error detection. The PN532 supports contactless communication using FeliCa Higher transfer speeds up to 424 kbit/s in both directions.

The PN532 supports layers 2 and 3 of the ISO/IEC 14443 B Reader/Writer communication scheme, except anticollision. This must be implemented in firmware as well as upper layers.

In card emulation mode, the PN532 is able to answer to a Reader/Writer command either according to the FeliCa or ISO/IEC 14443A/MIFARE card interface scheme. The PN532 generates the load modulation signals, either from its transmitter or from the LOADMOD pin driving an external active circuit. A complete secure card functionality is only possible in combination with a secure IC using the NFC-WI/S<sup>2</sup>C interface.

Compliant to ECMA 340 and ISO/IEC 18092 NFCIP-1 Passive and Active communication modes, the PN532 offers the possibility to communicate to another NFCIP-1 compliant device, at transfer speeds up to 424 kbit/s.The PN532 handles the complete NFCIP-1 framing and error detection.

The PN532 transceiver can be connected to an external antenna for Reader/Writer or Card/PICC modes, without any additional active component.



The PN532 supports the following host interfaces:

- SPI
- I<sup>2</sup>C
- High Speed UART (HSU)

An embedded low-dropout voltage regulator allows the device to be connected directly to a battery. In addition, a power switch is included to supply power to a secure IC.

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# 2. Features and benefits

- 80C51 microcontroller core with 40 KB ROM and 1 KB RAM
- Highly integrated demodulator and decoder
- Buffered output drivers to connect an antenna with minimum number of external components
- Integrated RF level detector
- Integrated data mode detector
- Supports ISO/IEC 14443A/MIFARE
- Supports ISO/IEC 14443B (Reader/Writer mode only)
- Typical operating distance in Reader/Writer mode for communication to ISO/IEC 14443A/MIFARE, ISO/IEC 14443B or FeliCa cards up to 50 mm depending on antenna size and tuning
- Typical operating distance in NFCIP-1 mode up to 50 mm depending on antenna size, tuning and power supply
- Typical operating distance in ISO/IEC 14443A/MIFARE or FeliCa card emulation mode of approximately 100 mm depending on antenna size, tuning and external field strength
- Supports MIFARE Classic 1K or MIFARE Classic 4K encryption in Reader/Writer mode and MIFARE higher transfer speed communication at 212 kbit/s and 424 kbit/s
- Supports contactless communication according to the FeliCa protocol at 212 kbit/s and 424 kbit/s
- Integrated RF interface for NFCIP-1 up to 424 kbit/s
- Possibility to communicate on the RF interface above 424 kbit/s using external analog components
- Supported host interfaces
  - SPI interface
  - ◆ I<sup>2</sup>C interface
  - High-speed UART
- Dedicated host interrupts
- Low power modes
  - Hard-Power-Down mode (1 μA typical)
  - Soft-Power-Down mode (22 μA typical)
- Automatic wake-up on I<sup>2</sup>C, HSU and SPI interfaces when device is in Power-down mode
- Programmable timers
- Crystal oscillator
- 2.7 to 5.5 V power supply operating range
- Power switch for external secure companion chip
- Dedicated IO ports for external device control
- Integrated antenna detector for production tests
- ECMA 373 NFC-WI interface to connect an external secure IC

# 3. Applications

- Mobile and portable devices
- Consumer applications

# 4. Quick reference data

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V <sub>BAT</sub>	battery supply voltage			2.7	-	5.5	V
ICV <sub>DD</sub>	LDO output voltage	V <sub>BAT</sub> > 3.4 V V <sub>SS</sub> = 0 V	<u>[1]</u>	2.7	3	3.4	V
$PV_{DD}$	Supply voltage for host interface	$V_{SS} = 0 V$		1.6	-	3.6	V
$SV_{DD}$	Output voltage for secure IC interface	V <sub>SS</sub> = 0 V (SV <sub>DD</sub> Switch Enabled)		DV <sub>DD</sub> -0.5	-	$DV_DD$	V
I <sub>HPD</sub>	Hard-Power-Down current consumption	V <sub>BAT</sub> = 5 V		-	-	2	μΑ
I <sub>SPD</sub>	Soft-Power-Down current consumption	V <sub>BAT</sub> = 5 V, RF level detector on		-	-	45	μA
I <sub>DVDD</sub>	Digital supply current	V <sub>BAT</sub> = 5 V, SV <sub>DD</sub> switch off	<u>[1]</u>	-	25	-	mA
I <sub>SVDD</sub>	$SV_{DD}$ load current	V <sub>BAT</sub> = 5 V, SV <sub>DD</sub> switch on		-	-	30	mA
I <sub>AVDD</sub>	Analog supply current	$V_{BAT} = 5 V$		-	6	-	mA
I <sub>TVDD</sub>	Transmitter supply current	During RF transmission, $V_{BAT} = 5 V$		-	60 <mark>[3]</mark>	150 <mark>[4]</mark>	mA
P <sub>tot</sub>	Continuous total power dissipation	T <sub>amb</sub> = -30 to +85 °C	<u>[2]</u>	-	-	0.5	W
T <sub>amb</sub>	ambient temperature			-30	-	+85	°C

[1]  $DV_{DD}$ ,  $AV_{DD}$  and  $TV_{DD}$  must always be at the same supply voltage.

[2] The total current consumption depends on the firmware version (different internal IC clock speed)

[3] With an antenna tuned at 50  $\Omega$  at 13.56 MHz

[4] The antenna should be tuned not to exceed this current limit (the detuning effect when coupling with another device must be taken into account)

# 5. Ordering information

Type number	Package					
	Name	Description	Version			
PN5321A3HN/C1xx <sup>[1][2][4]</sup>	HVQFN40	Heatsink Very thin Quad Flat package; 40 pins, plastic, body $6 \times 6 \times 0.85$ mm; leadless; MSL level 2 <sup>[3]</sup> .	SOT618-7			

- [1] xx refers to the ROM code version. The ROM code functionalities are described in the User-Manual document. Each ROM code has its own User-Manual.
- [2] This NXP IC is licensed under Innovatron's ISO/IEC 14443 Type B patent license.
- [3] This is tested according the joint IPC/JEDEC standard J-STD-020C of July 2004.
- [4] Purchase of an NXP Semiconductors IC that complies with one of the NFC Standards (ISO/IEC18.092; ISO/IEC21.481) does not convey an implied license under any patent right on that standards.

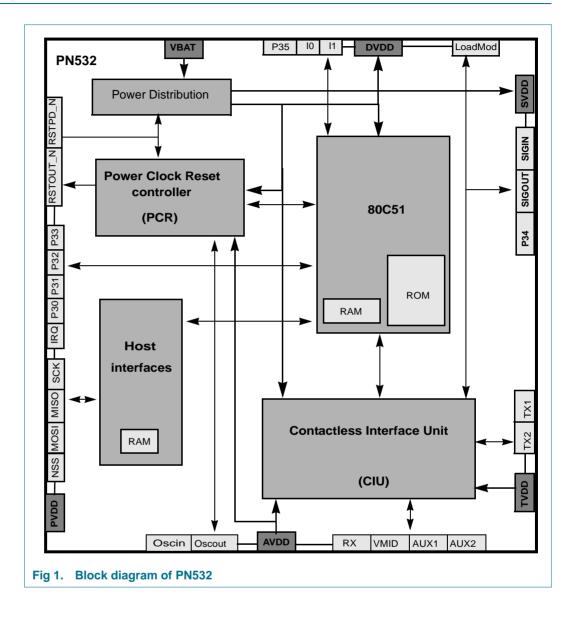
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# 6. Block diagram

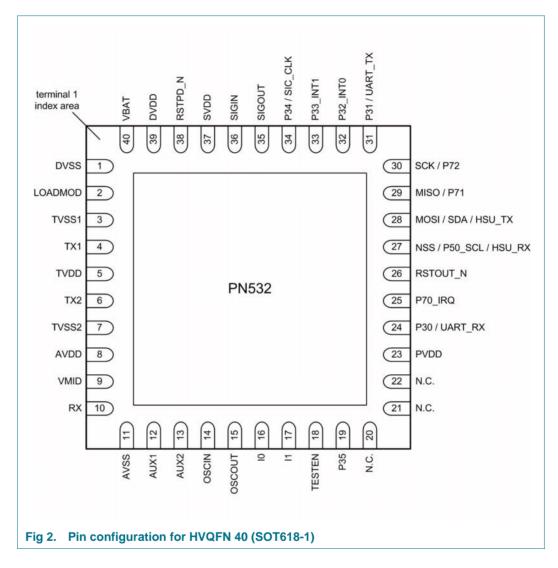


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# 7. Pinning information

# 7.1 Pinning



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# 7.2 Pin description

Symbol	Pin	Туре	Ref Voltage	Description
DVSS	1	PWR	-	Digital ground.
LOADMOD	2	0	DVDD	Load modulation signal.
TVSS1	3	PWR		Transmitter ground.
TX1	4	0	TVDD	Transmitter output 1: transmits modulated 13.56 MHz energy carrier.
TVDD	5	PWR		Transmitter power supply.
TX2	6	0	TVDD	Transmitter output 2: transmits modulated 13.56 MHz energy carrier.
TVSS2	7	PWR		Transmitter ground.
AVDD	8	PWR		Analog power supply.
VMID	9	0	AVDD	Internally generated reference voltage to bias the receiving path
RX	10	I	AVDD	Receiver input.
AVSS	11	PWR		Analog ground.
AUX1	12	0	AVDD	Auxiliary output 1: analog and digital test signals.
AUX2	13	0	AVDD	Auxiliary output 2: analog and digital test signals.
OSCIN	14	I	AVDD	Crystal oscillator input: to oscillator inverting amplifier.
OSCOUT	15	0	AVDD	Crystal oscillator output: from oscillator inverting amplifier.
10	16	I	DVDD	Host interface selector 0.
11	17	I	DVDD	Host interface selector 1.
TESTEN	18	I	DVDD	Reserved for test: connect to ground for normal operation.
P35	19	Ю	DVDD	General purpose IO.
N.C.	20			Not connected.
N.C.	21			Not connected.
N.C.	22			Not connected.
PVDD	23	PWR		Pad power supply.
P30 / UART_RX	24	Ю	PVDD	General purpose IO / Debug UART receive input.
P70_IRQ	25	Ю	PVDD	General purpose IO. Can be used as Interrupt request to host.
RSTOUT_N	26	0	PVDD	Reset indicator: when low, circuit is in reset state.
NSS / P50_SCL / HSU_RX	27	ΙΟ	PVDD	Host interface pin: SPI Not Slave Selected (NSS) or I <sup>2</sup> C clock (SCL) or HSU receive (HSU_RX). Refer to <u>Table 72 on</u> page 48 for details.
MOSI / SDA / HSU_TX	28	ΙΟ	PVDD	Host interface pin: SPI Master Out Slave In (MOSI) or I <sup>2</sup> C data (SDA) or HSU transmit (HSU_TX). Refer to <u>Table 72 on page 48</u> for details.
MISO / P71	29	Ю	PVDD	Host interface pin: SPI Master In Slave Out (MISO). Refer to Table 72 on page 48 for details.
				Con he used as general purpose IO

Can be used as general purpose IO.

Table 3.	PN532	Pin de	escription	continued
Symbol	Pin	Туре	Ref Voltage	Description
SCK / P72	30	Ю	PVDD	Host interface pin: SPI serial clock. Refer to <u>Table 72 on</u> page 48 for details.
				Can be used as general purpose IO.
P31 / UART_TX	31	IO	PVDD	General purpose IO/ Debug UART TX.
P32_INT0	32	IO	PVDD	General purpose IO / Interrupt source INT0.
P33_INT1	33	IO	PVDD	General purpose IO / Interrupt source INT1.
P34 / SIC_CLK	34	Ю	SVDD	General purpose IO / Secure IC clock.
SIGOUT	35	0	SVDD	Contactless communication interface output: delivers a serial data stream according to NFCIP-1 to a secure IC.
SIGIN	36	l	SVDD	Contactless communication interface input: accepts a serial data stream according to NFCIP-1 and from a secure IC.
SVDD	37	0		Switchable output power for secure IC power supply with overload detection. Used as a reference voltage for secure IC communication.
RSTPD_N	38	I	PVDD	Reset and Power-Down: When low, internal current sources are switched off, the oscillator is disabled, and input pads are disconnected from the outside world. The internal reset phase starts on the negative edge on this pin.
DVDD	39	0		Internal digital power supply.
VBAT	40	PWR		Main external power supply.

### Table 3. PN532 Pin description ...continued

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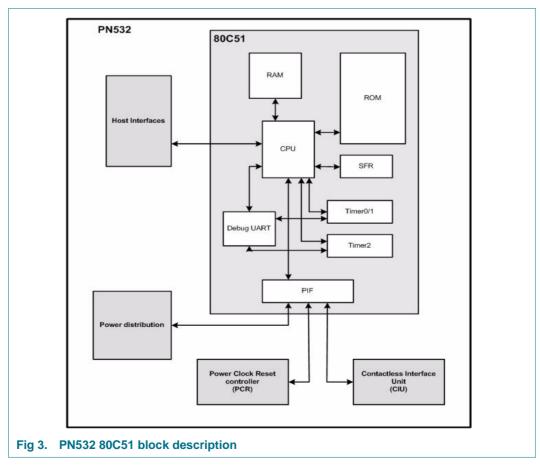
# 8. Functional description

### 8.1 80C51

The PN532 is controlled via an embedded 80C51 microcontroller core (for more details http://www.standardics.nxp.com/support/documents/microcontrollers/?scope=80C51). Its principle features are listed below:

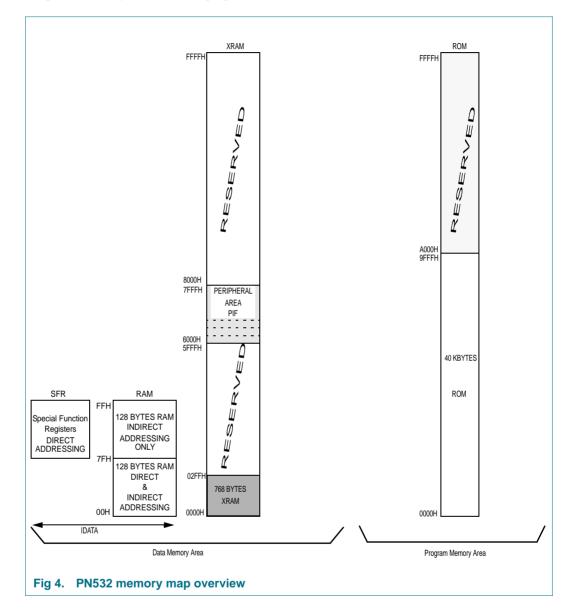
- 6-clock cycle CPU. One machine cycle comprises 6 clock cycles or states (S1 to S6). An instruction needs at least one machine cycle.
- ROM interface
- RAM interface to embedded IDATA and XRAM memories (see Figure 4 on page 11)
- Peripheral interface (PIF)
- Power control module to manage the CPU power consumption
- · Clock module to control CPU clock during Shutdown and Wake-up modes
- · Port module interface to configure I/O pads
- Interrupt controller
- Three timers
- Debug UART

The block diagram describes the main blocks described in this 80C51 section.



### 8.1.1 PN532 memory map

The memory map of PN532 is composed of 2 main memory spaces: data memory and program memory. The following figure illustrates the structure.



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### 8.1.2 Data memory

Data memory is itself divided into 2 spaces:

- 384-byte IDATA with byte-wide addressing
  - 258-byte RAM
  - 128-byte SFR
- 1 bank of 64 KB extended RAM (XRAM) with 2-byte-wide addressing

### 8.1.2.1 IDATA memory

The IDATA memory is mapped into 3 blocks, which are referred as Lower IDATA RAM, Upper IDATA RAM, and SFR. Addresses to these blocks are byte-wide, which implies an address space of only 256 bytes. However, 384 bytes can be addressed within IDATA memory through the use of direct and indirect address mechanisms.

- Direct addressing: the operand is specified by an 8-bit address field in the instruction.
- Indirect addressing: the instruction specifies a register where the address of the operand is stored.

For the range 80h to FFh, direct addressing will access the SFR space; indirect addressing accesses Upper IDATA RAM. For the range 0h0 to 7Fh, Lower IDATA RAM is accessed, regardless of addressing mode. This behavior is summarized in the table below:

### Table 4.IDATA memory addressing

Address	Addressing mode					
	Direct	Indirect				
00h to 7Fh	Lower IDATA RAM	Lower IDATA RAM				
80h to FFh	SFRs	Upper IDATA RAM				

The SFRs and their addresses are described in the Table 5:

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### le 5. SFR map of NFC controller

Address	Bit- addressable		Byte-addressable							
F8h	IP1		XRAMP		P3CFGA	P3CFGB			FFh	
F0h	В				P7CFGA	P7CFGB		P7	F7h	
E8h	IE1	CIU_Status2	CIU_FIFOData	CIU_FIFOLevel	CIU_WaterLevel	CIU_Control	CIU_BitFraming	CIU_Coll	EFh	
E0h	ACC								E7h	
D8h	I <sup>2</sup> CC0N	I <sup>2</sup> CSTA	I <sup>2</sup> CDAT	I <sup>2</sup> CADR				CIU_Status1	DFh	
D0h	PSW	CIU_Command	CIU_CommIEn	CIU_DivIEn	CIU_Commlrq	CIU_DivIrq	CIU_Error		D7h	
C8h	T2CON	T2MOD	RCAP2L	RCAP2H	T2L	T2H			CFh	
C0h									C7h	
B8h	IP0								BFh	
B0h	P3								B7h	
A8h	IE0	SPIcontrol	SPIstatus	HSU_STA	HSU_CTR	HSU_PRE	HSU_CNT		AFh	
A0h		FITEN	FDATA	FSIZE					A7h	
98h	SOCON	SBUF	RWL	TWL	FIFOFS	FIFOFF	SFF	FIT	9Fh	
90h									97h	
88h	T01CON	T01MOD	TOL	T1L	T0H	T1H			8Fh	
80h		SP	DPL	DPH				PCON	87h	

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### 8.1.2.2 XRAM memory

The XRAM memory is divided into 2 memory spaces:

- 0000h to 5FFFh: reserved for addressing embedded RAM. For the PN532, only accesses between 0000h and 02FF are valid.
- 6000h to 7FFFh: reserved for addressing embedded peripherals. This space is divided into 32 regions of 256 bytes each. Addressing can be performed using R0 or R1 and the XRAMP SFR.

The Table 6 depicts the mapping of internal peripherals into XRAM.

Table 6. Peripheral mapping into XRAM memory space

Base Address	End Address	Description
6000h	60FFh	Reserved.
6100h	61FFh	IOs and miscellaneous registers configuration Refer to <u>Section 8.2 "General purpose IOs configurations" on page 38</u>
6200h	62FFh	Power Clock and Reset controller Refer to <u>Section 8.5.7 "PCR extension registers" on page 93</u>
6300h	633Fh	Contactless Unit Interface Refer to Section 8.6 "Contactless Interface Unit (CIU)" on page 99
6340h	FFFFh	Reserved

XRAM is accessed via the dedicated MOVX instructions. There are two access modes:

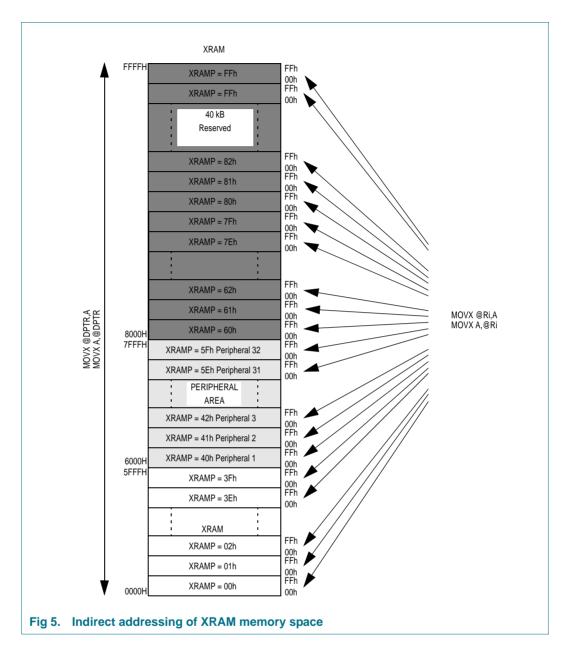
- 16-bit data pointer (DPTR): the full XRAM address space can be accessed.
- paging mechanism: the upper address byte is stored in the SFR register XRAMP; the lower byte is stored in either R1 or R0.

The Figure 5 illustrates both mechanisms.

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### 8.1.3 Program memory

PN532 program memory ranges from 0000h to 9FFFh, which is physically mapped to the 40 KB ROM.

### 8.1.4 PCON module

The Power Control (PCON) module is configured using the PCON SFR register.

 Table 7.
 PCON register (SFR: address 87h) bit allocation

	•							
Bit	7	6	5	4	3	2	1	0
Symbol	SMOD			-			CPU_PD	-
Reset	0	0	0	0	0	0	0	0
Access	R/W	R	R	R	R	R	R/W	R/W

Table 8.	Description of PCON bits					
Bit	Symbol	Description				
7	SMOD	Serial MODe: When set to logic 1, the baud rate of the Debug UART is doubled				
6 to 3	-	Reserved.				
1	CPU_PD	<b>Power-down:</b> When set to logic 1, the microcontroller goes in Power-down mode				
0	Reserved	This bit should only ever contain logic 0.				

### 8.1.5 Interrupt Controller

The interrupt controller has the following features:

- 13 interrupt sources
- Interrupt enable registers IE0 and IE1
- Interrupt priority registers IP0 and IP1
- Wake-up from Power-Down state

### 8.1.5.1 Interrupt vectors

The mapping between interrupt sources and interrupt vectors is shown in Table 9.

### Table 9. Interrupt vector

Interrupt number	Interrupt vector	Interrupt sources	Incremental priority level (conflict resolution level)
0	0003h	External P32_INT0	Highest
1	000Bh	Timer0 interrupt	
2	0013h	External P33_INT1	
3	001Bh	Timer1 interrupt	
4	0023h	Debug UART interrupt	
5	002Bh	Timer2 interrupt	
6	0033h	NFC-WI interrupt	
7	003Bh	LDO overcurrent interrupt	
8	0043h	Reserved	
9	004Bh	CIU interrupt 1	
10	0053h	CIU interrupt 0	
11	005Bh	I <sup>2</sup> C interrupt	
12	0063h	SPI, FIFO, or HSU interrupts	
13	006Bh	Reserved	
14	0073h	General Purpose IRQ	Lowest

### 8.1.5.2 Interrupt enable: IE0 and IE1 registers

Each interrupt source can be individually enabled or disabled by setting a bit in IE0 or IE1. In register IE0, a global interrupt enable bit can be set to logic 0 to disable all interrupts at once.

The 2 following tables describe IE0.

### Table 10. Interrupt controller IE0 register (SFR: address A8h) bit allocation

			0	•	,			
Bit	7	6	5	4	3	2	1	0
Symbol	IE0_7	IE0_6	IE0_5	IE0_4	IE0_3	IE0_2	IE0_1	IE0_0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

### Table 11. Description of IE0 bits

Bit	Symbol	Description
7	IE0_7	<b>Global interrupt enable</b> When set to logic 1, the interrupts can be enabled. When set to logic 0, all the interrupts are disabled.
6	IE0_6	<b>NFC-WI counter interrupt enable</b> When set to logic 1, NFC-WI interrupt is enabled. See <u>Table 164 on page 126</u> .
5	IE0_5	<b>Timer2 interrupt enable</b> When set to logic 1, Timer2 interrupt is enabled. See <u>Table 36 on page 28</u> .
4	IE0_4	<b>Debug UART interrupt enable</b> When set to logic 1, Debug UART interrupt is enabled. See <u>Table 49 on</u> page <u>33</u> .
3	IE0_3	<b>Timer1 interrupt enable</b> When set to logic 1, Timer1 interrupt is enabled. See <u>Table 23 on page 23</u> .
2	IE0_2	<b>P33_INT1 interrupt enable</b> When set to logic 1, P33_INT1 pin interrupt is enabled. See <u>Table 23 on</u> <u>page 23</u> . The polarity of P33_INT1 can be inverted (see <u>Table 73 on page 49</u> ).
1	IE0_1	<b>Timer0 interrupt enable</b> When set to logic 1, Timer0 interrupt is enabled. See <u>Table 23 on page 23</u> .
0	IE0_0	<b>P32_INT0 interrupt enable</b> When set to logic 1, P32_INT0 pin interrupt is enabled. See <u>Table 23 on</u> page 23.

The 2 following tables describe IE1.

### Table 12. Interrupt controller IE1 register (SFR: address E8h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	IE1_7	-	IE1_5	IE1_4	IE1_3	IE1_2	-	IE1_0
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

### Table 13.Description of IE1 bits

Bit	Symbol	Description
7	IE1_7	<b>General purpose IRQ interrupt enable</b> . When set to logic 1, enables interrupt function of P34, P35, P50_SCL and P71 according to their respective enable and level control bits. See <u>Table 19 on</u> page 20, <u>Table 137 on page 95</u> and <u>Table 143 on page 97</u> .
6	-	Reserved. This bit must be set to logic 0
5	IE1_5	<b>FIFO, SPI and HSU interrupt enable.</b> When set to logic 1, enables FIFO interrupts, SPI interrupts, HSU interrupt. In HSU mode, the interrupt is when NSS is at logic 0. For the FIFO interrupts, see <u>Table 112 on page 76</u> . For the SPI interrupts, see <u>Table 122 on page 81</u> .
4	IE1_4	<b>I2C interrupt enable</b> . When set to logic 1, enables I <sup>2</sup> C interrupt. See <u>Table 77 on page 54</u> .
3	IE1_3	CIU interrupt 0 enable. When set to logic 1, enables CIU interrupt 0: CIU_IRQ_0. See <u>Table 190 on</u> page 151.
2	IE1_2	<b>CIU interrupt 1 enable.</b> When set to logic 1, enables the CIU interrupt 1: CIU_IRQ_1. See <u>Table 190</u> on page <u>151</u> .
1	-	Reserved. This bit must be set to logic 0.
0	IE1_0	LDO overcurrent interrupt enable. When set to logic 1, enables the LDO overcurrent detection interrupt. See <u>Table 127 on page 88</u> .

### 8.1.5.3 Interrupt prioritization: IP0 and IP1 registers

Each interrupt source can be individually programmed to be one of two priority levels by setting or clearing a bit in the interrupt priority registers IP0 and IP1. If two interrupt requests of different priority levels are received simultaneously, the request with the high priority is serviced first. On the other hand, if the interrupts are of the same priority, precedence is resolved by comparing their respective conflict resolution levels (see Table 9 on page 16 for details). The processing of a low priority interrupt can be interrupted by one with a high priority.

A RETI (Return From Interrupt) instruction jumps to the address immediately succeeding the point at which the interrupt was serviced. The instruction found at the return address will be executed, prior to servicing any pending interrupts.

The 2 following tables describe IP0.

### Table 14. Interrupt controller IP0 register (SFR: address B8h) bit allocation

			•	•				
Bit	7	6	5	4	3	2	1	0
Symbol	IP0_7	IP0_6	IP0_5	IP0_4	IP0_3	IP0_2	IP0_1	IP0_0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

### Table 15. Description of IP0 bits

Bit	Symbol	Description
7	IP0_7	Reserved
6	IP0_6	When set to logic 1, NFC-WI interrupt is set to high priority.
5	IP0_5	When set to logic 1, Timer2 interrupt is set to high priority.
4	IP0_4	When set to logic 1, Debug UART interrupt is set to high priority.
3	IP0_3	When set to logic 1, Timer1 interrupt is set to high priority.
2	IP0_2	When set to logic 1, external P33_INT1 pin is set to high priority.
1	IP0_1	When set to logic 1, Timer0 interrupt is set to high priority.
0	IP0_0	When set to logic 1, external P32_INT0 pin is set to high priority.

The 2 following tables describe IP1.

### Table 16. Interrupt controller IP1 register (SFR: address F8h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	IP1_7	-	IP1_5	IP1_4	IP1_3	IP1_2	-	
Reset	0	0	0	0	0	0	00	)
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/V	V

### Table 17. Description of IP1 bits

Bit	Symbol	Description
7	IP1_7	When set to logic 1, General Purpose IRQ interrupt is set to high priority.
6	-	Reserved. This bit must be set to logic 0.
5	IP1_5	When set to logic 1, combined SPI, FIFO and HSU interrupt is set to high priority.
4	IP1_4	When set to logic 1, I <sup>2</sup> C interrupt is set to high priority.
3	IP1_3	When set to logic 1, CIU interrupt 0 is set to high priority.
2	IP1_2	When set to logic 1, CIU interrupt 1 is set to high priority.
1	-	Reserved. This bit must be set to logic 0.
0	IP1_0	When set to logic 1, interrupt number 7 is set to high priority.

### 8.1.5.4 General purpose IRQ control

The general purpose interrupts are controlled by register GPIRQ.

NOTE: this is not a standard feature of the 8051.

Table 18.	<b>GPIRQ</b> register	(address 6107h)	bit allocation
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Bit	7	6	5	4	3	2	1	0
Symbol	gpirq_ level_ P71	gpirq_ level_ P50	gpirq_ level_ P35	gpirq_ level_ P34	gpirq_ enable _P71	gpirq_ enable_ P50	gpirq_ enable_ P35	gpirq_ enable_ P34
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

### Table 19. Description of GPIRQ bits

Bit	Symbol	Description
7	gpirq_level_P71	Configures the polarity of signal on P71 to generate a GPIRQ interrupt event (assuming gpirq_enable_P71 is set).
		When set to logic 0, an interrupt will be generated if P71 is at logic 0.
		When set to logic 1, an interrupt will be generated if P71 is at logic 1.
6	gpirq_level_P50	Configures the polarity of signal on P50 to generate a GPIRQ interrupt event (assuming gpirq_enable_P50 is set).
		When set to logic 0, an interrupt will be generated if P50_SCL is at logic 0.
		When set to logic 1, an interrupt will be generated if P50_SCL is at logic 1.
5	gpirq_level_P35	Configures the polarity of signal on P35 to generate a GPIRQ interrupt event (assuming gpirq_enable_P35 is set).
		When set to logic 0, an interrupt will be generated if P35 is at logic 0.
		When set to logic 1, an interrupt will be generated if P35 is at logic 1.
4	gpirq_level_P34	Configures the polarity of signal on P34 to generate a GPIRQ interrupt event (assuming gpirq_enable_P34 is set).
		When set to logic 0, an interrupt will be generated if P34 is at logic 0.
		When set to logic 1, an interrupt will be generated if P34 is at logic 1.
		<b>Remark:</b> If hide_svdd_sig of the register control_rngpower is set and gpirq_enable_P34 is also set then this bit will be asserted independently of the level on the pad P34.
3	gpirq_enable_P71	When set to logic 1, enables pad P71 to generate a GPIRQ interrupt event. <sup>[1]</sup>
2	gpirq_enable_P50	When set to logic 1, enables pad P50_SCL to generate a GPIRQ interrupt event. <sup>[1]</sup>
1	gpirq_enable_P35	When set to logic 1, enables pad P35 to generate a GPIRQ interrupt event. <sup>[1]</sup>
0	gpirq_enable_P34	When set to logic 1, enables pad P34 to generate a GPIRQ interrupt event. <sup>[1]</sup>
[1]	The bit IE1 7 of regist	er IE1 (see Table 13 on page 18) has also to be set to logic 1 to enable the

[1] The bit IE1\_7 of register IE1 (see <u>Table 13 on page 18</u>) has also to be set to logic 1 to enable the corresponding CPU interrupt.

### 8.1.6 Timer0/1 description

Timer0/1 are general purpose timer/counters. Timer0/1 has the following functionality:

- · Configurable edge or level detection interrupts
- Timer or counter operation
- 4 timer/counter modes
- Baud rate generation for Debug UART

Timer0/1 comprises two 16-bit timer/counters: Timer0 and Timer1. Both can be configured as either a timer or an event counter.

Each of the timers can operate in one of four modes:

- Mode 0: 13-bit timer/counter
- Mode 1: 16-bit timer/counter
- Mode 2: 8-bit timer/counter with programmable preload value
- Mode 3: two individual 8-bit timer/counters (Timer0 only)

In the 'timer' function, the timer/counter is incremented every machine cycle. The count rate is 1/6 of the CPU clock frequency (CPU\_CLK).

In the 'counter' function, the timer/counter is incremented in response to a 1-to-0 transition on the input pins P34 / SIC\_CLK (Timer0) or P35 (Timer1). In this mode, the external input is sampled during state S5 of every machine cycle. If the associated pin is at logic 1 for a machine cycle, followed by logic 0 on the next machine cycle, the count is incremented. The new count value appears in the timer/counter in state S3 of the machine cycle following the one in which the transition was detected. The maximum count rate is 1/12 of the CPU\_CLK frequency. There are no restrictions on the duty cycle of the external input signal but to ensure that a given level is sampled at least once before it changes, it should be held for at least one full machine cycle.

The overflow output 't1\_ovf' of Timer1 can be used as a baud rate generator for the Debug UART. The Timer1 interrupt should be disabled in this case. For most applications which drive the Debug UART, Timer1 is configured for 'timer' operation and in auto-reload mode.

### 8.1.6.1 Timer0/1 registers

The Timer0/1 module contains six Special Function Registers (SFRs) which can be accessed by the CPU.

Name	Size [bytes]	Address Offset	Description	Access				
T01CON	1	88h	Timer0/1 control register	R/W				
T01MOD	1	89h	Timer0/1 mode register	R/W				
TOL	1	8Ah	Timer0 timer/counter lower byte	R/W				
T1L	1	8Bh	Timer1 timer/counter lower byte	R/W				
тон	1	8Ch	Timer0 timer/counter upper byte	R/W				
T1H	1	8Dh	Timer1 timer/counter upper byte	R/W				

### Table 20. Timer0/1 Special Function registers list

The firmware performs a register read in state S5 and a register write in state S6. The hardware loads bits TF0 and TF1 of the register T01CON during state S2 and state S4 respectively. The hardware loads bits IE0 and IE1 of the register T01CON during state S1 and reset these bits during state S2. The registers T0L, T0H, T1L, T1H are updated by the hardware during states S1, S2, S3 and S4 respectively. At the end of a machine cycle, the firmware load has overridden the hardware load as the firmware writes in state S6.

CPU STATE										
Register	Bit	S1	S2	S3	S4	S5	S6			
T01CON	TF0		HW read			SW read	SW write			
	TF1				HW read	SW read	SW write			
	IE0 / IE1	HW write	HW reset			SW read	SW write			
TOL		HW write				SW read	SW write			
ТОН			HW write			SW read	SW write			
T1L				HW write		SW read	SW write			
T1H					HW write	SW read	SW write			

### Table 21. Timer0/1 SFR registers CPU state access

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### 8.1.6.2 T01CON register

The register is used to control Timer0/1 and report its status.

### Table 22. Timer0/1 T01CON register (SFR address 88h), bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

### Table 23. Description of Timer0/1 T01CON register bits

Bit	Symbol	Description
7	TF1	<b>Timer1 overflow.</b> Set to logic 1 by hardware on a Timer1 overflow. The flag is set to logic 0 by the CPU after 2 machine cycles. The bit IE0_3 of register IE0 (see <u>Table 11 on page 17</u> ) has to be set to logic 1 to enable the corresponding CPU interrupt.
6	TR1	<b>Timer1 run control</b> . Set by firmware only. When set to logic 1, Timer1 is enabled.
5	TFO	<b>Timer0 overflow.</b> Set by hardware on a Timer0 overflow. The flag is set to logic 0 by the CPU after 2 machine cycles. The bit IE0_1 of register IE0 (see <u>Table 11 on page 17</u> ) has to be set to logic 1 to enable the corresponding CPU interrupt.
4	TR0	<b>Timer0 run control</b> . Set by firmware only. When set to logic 1, Timer0 is enabled.
3	IE1	<b>External Interrupt1 event</b> . Set to logic 1 by hardware when an external interrupt is detected on P33_INT1. The bit IE0_2 of register IE0 (see <u>Table 11 on page 17</u> ) has to be set to logic 1 to enable the corresponding CPU interrupt.
2	IT1	<b>External Interrupt1 control</b> . Set by firmware only. When set to logic 1, Interrupt1 triggers on a falling edge of P33_INT1. When set to logic 0, Interrupt1 triggers on a low level of P33_INT1.
1	IEO	<b>External Interrupt0 event</b> . Set to logic 1 by hardware when an external interrupt is detected on P32_INT0. The bit IE0_0 of register IE0 (see <u>Table 11 on page 17</u> ) has to be set to logic 1 to enable the corresponding CPU interrupt.
0	IT0	<b>External Interrupt0 control</b> . Set by firmware only. When set to logic 1, Interrupt0 triggered by a falling edge on P32_INT0. When set to logic 0, Interrupt0 triggered by a low level on P32_INT0.

### 8.1.6.3 T01MOD register

This register is used to configure Timer0/1.

### Table 24. Timer 0/1 T01MOD register (SFR address 89h), bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	GATE1	C/T1	M11	M10	GATE0	C/T0	M01	M00
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Bit	Symbol	Description
7	GATE1	Timer1 gate control. Set by firmware only.
		When set to logic 1, Timer1 is enabled only when P33_INT1 is high and bir TR1 of register T01CON is set.
		When set to logic 0, Timer1 is enabled.
6	C/T1	Timer1 timer/counter selector. Set by firmware only.
		When set to logic 1, Timer1 is set to counter operation.
		When set to logic 0, Timer1 is set to timer operation.
5 to 4	M[11:10]	Timer1 mode. Set by firmware only.
		<ul> <li>Mode 0: M11 = 0 and M10 = 0</li> </ul>
		– 8192 counter
		<ul> <li>T1L serves as a 5-bit prescaler</li> </ul>
		<ul> <li>Mode 1: M11 = 0 and M10 = 1</li> </ul>
		<ul> <li>16-bit timer/counter</li> </ul>
		<ul> <li>T1H and T1L are cascaded</li> </ul>
		<ul> <li>Mode 2: M11 = 1 and M10 = 0</li> </ul>
		<ul> <li>8-bit auto-reload timer/counter.</li> </ul>
		<ul> <li>T1H stores value to be reloaded into T1L each time T1L overflows</li> </ul>
		• Mode 3: M11 = 1 and M10 = 1
		<ul> <li>Timer1 is stopped (count frozen).</li> </ul>

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Bit	Symbol	Description
3	GATE0	Timer0 gate control. Set by firmware only.
		When set to logic 1, Timer0 is enabled only when P32_INT0 is high and bit TR0 of register T01CON is set.
		When set to logic 0, Timer0 is enabled.
2	C/T0	Timer0 timer/counter selector. Set by firmware only.
		When set to logic 1, Timer0 is set to counter operation.
		When set to logic 0, Timer0 is set to timer operation.
1 to 0	M[01:00]	Timer0 mode. Set by firmware only.
		<ul> <li>Mode 0: M01 = 0 and M00 = 0</li> </ul>
		– 8192 timer
		<ul> <li>T0L acts as a 5-bit prescaler.</li> </ul>
		<ul> <li>Mode 1: M01 = 0 and M00 = 1</li> </ul>
		<ul> <li>16-bit timer/counter</li> </ul>
		<ul> <li>T0H and T0L are cascaded.</li> </ul>
		<ul> <li>Mode 2: M01 = 1 and M00 = 0</li> </ul>
		<ul> <li>8-bit auto-reload timer/counter</li> </ul>
		<ul> <li>T0H stores value to be reloaded into T0L each time T0L overflows</li> </ul>
		<ul> <li>Mode 3: M01 = 1 and M00 = 1</li> </ul>
		<ul> <li>Timer0 split into two 8-bit timer/counters T0H and T0L</li> </ul>
		<ul> <li>T0H is controlled by the control bit of Timer1: bit TR1 of register T01CON</li> </ul>
		<ul> <li>T0L is controlled by standard Timer0 control: "{P32_INT0 OR</li> </ul>

### Table 25. Description of T01MOD bits ...continued

### 8.1.6.4 TOL and TOH registers

These are the actual timer/counter bytes for Timer0: T0L is the lower byte; T0H is the upper byte.

(NOT GATE0)} AND bit TR0".

Bit	7	6	5	4	3	2	1	0
Symbol	T0L.7	T0L.6	T0L.5	T0L.4	T0L.3	T0L.2	T0L.1	T0L.0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

### Table 26. Timer0/1 T0L register (SFR address 8Ah), bit allocation

### Table 27. Description of T0L bits

Bit	Symbol	Description
7:0	T0L.7 to T0L.0	Timer0 timer/counter lower byte

### Table 28. Timer0/1 T0H register (SFR address 8Ch), bit allocation

Bit	7	6	5	4	3	2	1	0
	•	Ū	J	-	J	-		U
Symbol	T0H.7	T0H.6	T0H.5	T0H.4	T0H.3	T0H.2	T0H.1	T0H.0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

### Table 29.Description of T0H bits

Bit	Symbol	Description
7 to 0	T0H.7 to T0H.0	Timer0 timer/counter upper byte

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### 8.1.6.5 T1L and T1H registers

These are the actual timer/counter bytes for Timer1. T1L is the lower byte, T1H is the upper byte.

### Table 30. Timer0/1 T1L register (SFR address 8Bh), bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	T1L.7	T1L.6	T1L.5	T1L.4	T1L.3	T1L.2	T1L.1	T1L.0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

### Table 31.Description of T1L bits

Bit	Symbol	Description
7 to 0	T1L.7 to T1L.0	Timer1 timer/counter lower byte

### Table 32. Timer0/1 T1H register (SFR address 8Dh), bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	T1H.7	T1H.6	T1H.5	T1H.4	T1H.3	T1H.2	T1H.1	T1H.0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

### Table 33. Description of T1H bits

Bit	Symbol	Description
7 to 0	T1H.7 to T1H.0	Timer1 timer/counter upper byte

### 8.1.6.6 Incrementer

The two 16-bit timer/counters are built around an 8-bit incrementer. The Timer0/1 are incremented in the CPU states S1 to S4; the overflow flags are set in CPU states S2 and S4.

- CPU state S1: TOL is incremented if Timer0 is set to:
  - timer operation
  - counter operation and when a 1-to-0 transition is detected on P34 / SIC\_CLK input.
- CPU state S2: TOH is incremented if:
  - T0L overflows. The overflow flag TF0 in register T01CON is updated.
- CPU state S3: T1L is incremented if Timer1 is set to:
  - timer operation or
  - counter operation and when a 1-to-0 transition is detected on P35 input.
- CPU state S4: T1H is incremented if:
  - T1L overflows. The overflow flag TF1 in register T01CON is updated.

### 8.1.6.7 Overflow detection

For both the upper and lower bytes of the Timer0/1, an overflow is detected by comparing the incremented value of the most significant bit with its previous value. An overflow occurs when this bit changes from logic 1 to logic 0. An overflow event in the lower byte is clocked into a flip-flop and is used in the next state as the increment enable for the upper byte. An overflow event in the upper byte will set the corresponding overflow bit in the T01CON register to logic 1. The upper byte overflow is also clocked into a flip-flop to generate the output signals 't0\_ovf' and 't1\_ovf'.

The overflow flags TF0 and TF1, found in register T01CON, are loaded during states S2 and S4 respectively. The interrupt controller of the 80C51 scans all requests at state S2. Thus, an overflow of Timer0 or Timer1 is detected one machine cycle after it occurred. When the request is serviced, the interrupt routine sets the overflow flag to logic 0.

Execution of the interrupt routine starts on the fourth machine cycles following the timer overflow. When Timer0/1 receives the acknowledge from the CPU:

- the overflow flag TF0 in register T01CON is set to logic 0
- two machine cycles later, the overflow flag TF1 in register T01CON is set to logic 0

If during the same machine cycle, an overflow flag is set to logic 0 due to a CPU acknowledge and set to logic 1 due to an overflow, the set to logic 1 is the strongest.

### 8.1.7 Timer2 description

Timer2 supports a subset of the standard Timer2 found in the 8052 microcontroller. Timer2 can be configured into 2 functional modes via the T2CON and T2MOD registers:

- Mode1: Auto-reload up/down counting
- Mode2: Baud rate generation for Debug UART

Timer2 can operate either as a timer or as an event counter.

### 8.1.7.1 Timer2 registers

Timer2 contains six Special Function Registers (SFRs) which can be accessed by the CPU.

Name	Size [bytes]	SFR address	Description	Access
T2CON	1	C8h	Timer2 control register	R/W
T2MOD	1	C9h	Timer2 mode register	R/W
RCAP2L	1	CAh	Timer2 reload lower byte	R/W
RCAP2H	1	CBh	Timer2 reload upper byte	R/W
T2L	1	CCh	Timer2 timer/counter lower byte	R/W
T2H	1	CDh	Timer2 timer/counter upper byte	R/W

### Table 34. Timer2 SFR register List

Timer2 registers can be written to by either hardware or firmware. If both the hardware and firmware attempt to update the registers T2H, T2L, RCAP2H or RCAP2L during the same machine cycle, the firmware write takes precedence. A firmware write occurs in state S6 of the machine cycle.

Each increment or decrement of Timer2 occurs in state S1 except when in baud rate generation mode and configured as a counter. In this mode, Timer2 increments on each clock cycle. When configured as a timer, Timer2 is incremented every machine cycle. Since a machine cycle consists of 6 clock periods, the count rate is 1/6 of the CPU clock frequency.

### 8.1.7.2 T2CON register

The register is used to control Timer2 and report its status.

Table 35.	Timer2 T2CON registe	r (SFR address	C8h) bit allocation
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					,			
Bit	7	6	5	4	3	2	1	0
Symbol	TF2	-	RCLK0	TCLK0	-	TR2	C/T2	-
Reset	0	0	0	0	0	0	0	0
Access	R	R/W	R/W	R/W	R/W	R/W	R/W	R/W

### Table 36.Description of T2CON bits

Bit	Symbol	Description
7	TF2	Timer2 overflow
		Set to logic 1 by a Timer2 overflow.
		Set to logic 0 by firmware. TF2 is not set when in baud rate generation mode.
		The bit IE0_5 of register IE0 (see <u>Table 11 on page 17</u> ) has to be set to logic 1 to enable the corresponding CPU interrupt.
6	-	Reserved.
5	RCLK0	Timer2 Debug UART Receive Clock selector. Set by firmware only.
		When set to logic 1, Debug UART uses Timer2 overflow pulses.
		When set to logic 0, Debug UART uses overflow pulses from another source (e.g. Timer1 in a standard configuration).
4	TCLK0	Timer2 Debug UART Transmit Clock selector. Set by firmware only.
		When set to logic 1, Debug UART uses Timer2 overflow pulses.
		When set to logic 0, Debug UART uses overflow pulses from another source (e.g. Timer1 in a standard configuration).
3	-	Reserved.
2	TR2	Timer2 Run control. Set by firmware only.
		When set to logic 1, Timer2 is started.
		When set to logic 0, Timer2 is stopped.
1	C/T2	Timer2 Counter/Timer selector. Set by firmware only.
		When set to logic 1, Timer2 is set to counter operation.
		When set to logic0, Timer2 is set to timer operation.
0	-	Reserved. This bit must be set to logic 0 by firmware.

### 8.1.7.3 T2MOD register

This Special Function Register is used to configure Timer2.

### Table 37. Timer2 T2MOD register (SFR address C9h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	-	T2RD	-	DCEN
Reset	0	0	0	0	0	0	0	0
Access	R	R	R	R	R	R	R/W	R/W

### Table 38. Description of TMOD bits

Bit	Symbol	Description
7 to 3	-	Reserved.
2	T2RD	Timer2 ReaD flag. Set by hardware and firmware.
		This bit is set to logic 1 by hardware, if T2H is incremented between reading T2L and reading T2H. This bit is set to logic 0, on the trailing edge of next T2L read.
		This bit is used to indicate that the16 bit Timer2 register is not read properly since the T2H part was incremented by hardware before it was read.
1	-	Reserved
0	DCEN	Timer2 Down Count ENable. Set by firmware only.
		When this bit is set, Timer2 can be configured (in auto_reload mode) as an up-counter.
		When this bit is reset, Timer2 can be configured (in auto-reload mode) as a down-counter.

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### 8.1.7.4 T2L, T2H registers

These are the actual timer/counter bytes. T2L is the lower byte, T2H the upper byte.

On the fly reading can give a wrong value since T2H can be changed after T2L is read and before T2H is read. This situation is indicated by flag T2RD in T2MOD.

These two 8-bit registers are always combined to operate as one 16-bit timer/counter.

TILLO	T' 0 TO	1		001	and the second second second
Table 39.	Timer2 12L	. register	(SFR address	CCh	) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	T2L.7	T2L.6	T2L.5	T2L.4	T2L.3	T2L.2	T2L.1	T2L.0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

### Table 40. Description of T2L bits

Bit	Symbol	Description
7 to 0	T2L.7 to T2L.0	Timer2 timer/counter lower byte

### Table 41. Timer2 T2H register (SFR address CDh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	T2H.7	T2H.6	T2H.5	T2H.4	T2H.3	T2H.2	T2H.1	T2H.0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

### Table 42. Description of T2H bits

Bit	Symbol	Description
7 to 0	T2H.7 to T2H.0	Timer2 timer/counter upper byte

### 8.1.7.5 RCAP2L, RCAP2H registers

These are the reload bytes. In the reload mode the T2H/T2L counters are loaded with the values found in the RCAP2H/RCAP2L registers respectively.

Table 43.	Timer2 RCAP2L	register (SFR	address CAh)	bit allocation
-----------	---------------	---------------	--------------	----------------

Bit	7	6	5	4	3	2	1	0
Symbol	R2L.7	R2L.6	R2L.5	R2L.4	R2L.3	R2L.2	R2L.1	R2L.0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

### Table 44. Description of RCAP2L bits

Bit	Symbol	Description
7 to 0	R2L.7 to R2L.0	Timer2 lower rele

```
Timer2 lower reload byte
R2L.7 to R2L.0
```

### Table 45. Timer2 RCAP2H register (SFR address CBh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	R2H.7	R2H.6	R2H.5	R2H.4	R2H.3	R2H.2	R2H.1	R2H.0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

### Table 46. Description of RCAP2H bits

Bit	Symbol	Description
7 to 0	R2H.7 to R2H.0	Timer2 upper reload byte

### 8.1.8 Debug UART

The Debug UART is implemented to assist debug using UART\_RX and UART\_TX pins.

### 8.1.8.1 Feature list

The Debug UART has the following characteristics:

- Full duplex serial port
- Receive buffer to allow reception of a second byte while the first byte is being read out by the CPU
- Four modes of operation which support 8-bit and 9-bit data transfer at various baud rates
- Supports multi-processor communication
- Baud rate can be controlled through Timer1 or Timer2 baud rate generator

### 8.1.8.2 Debug UART functional description

The serial port has a receive buffer: a second byte can be stored while the previous one is read out of the buffer by the CPU. However, if the first byte has still not been read by the time reception of the second byte is complete, one of the bytes will be lost.

The receive and transmit data registers of the serial port are both accessed by firmware via the Special Function Register S0BUF. Writing to S0BUF loads the transmit register; reading from S0BUF accesses a physically separate receive register.

The serial port can operate in 4 modes. These modes are selected by programming bits SM0 and SM1 in S0CON:

- Mode 0:
  - Serial data are received and transmitted through UART\_RX. UART\_TX outputs the shift clock. 8 bits are transmitted/received (LSB first)

Baud rate: fixed at 1/6 of the frequency of the CPU clock

- Mode 1:
  - 10 bits are transmitted through UART\_TX or received through UART\_RX: a start bit (0), 8 data bits (LSB first), and a stop bit (1)
  - Receive: The received stop bit is stored into bit RB8 of register S0CON
  - Baud rate: variable (depends on overflow of Timer1 or Timer2)
- Mode 2:
  - 11 bits are transmitted through UART\_TX or received through UART\_RX: start bit (0), 8 data bits (LSB first), a 9th data bit, and a stop bit (1)
  - Transmit: the 9th data bit is taken from bit TB8 of S0CON. For example, the parity bit could be loaded into TB8.
  - Receive: the 9th data bit is stored into RB8 of S0CON, while the stop bit is ignored
  - Baud rate: programmable to either 1/16 or 1/32 the frequency of the CPU clock

- Mode 3:
  - 11 bits are transmitted through UART\_TX or received through UART\_RX: a start bit (0), 8 data bits (LSB first), a 9th data bit, and a stop bit (1). In fact, mode 3 is the same as mode 2 in all aspects except the baud rate
  - Transmit: as mode 2, the 9th data bit is taken from TB8 of S0CON
  - Receive: as mode 2, the 9th data bit is stored into RB8 of S0CON
  - Baud rate: depends on overflows of Timer1 or Timer2

The Debug UART initiates transmission and/or reception as follows.

- Transmission is initiated, in modes 0, 1, 2, 3, by any instruction that uses S0BUF as destination
- Reception is initiated, in mode 0, if RI and REN in S0CON are set to logic 0 and 1
  respectively
- Reception is initiated in modes 1, 2, 3 by the incoming start bit if REN in S0CON is set to a logic 1

The Debug UART contains 2 SFRs:

### Table 47. Debug UART SFR register list

Name	Size [bytes]	SFR address	Description	Access
S0CON	1	0098h	Control and status register	R/W
S0BUF	1	0099h	Transmit and receive buffer	R/W

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### 8.1.8.3 S0CON register

The Special Function Register S0CON is the control and status register of the Debug UART. This register contains the mode selection bits (SM2, SM1, SM0), the 9th data bit for transmit and receive (TB8 and RB8), and the serial port interrupt bits (TI and RI).

### Table 48. Debug UART S0CON register (SFR: address 98h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	SM0	SM1	SM2	REN	TB8	RB8	TI	RI
Reset	0	0	0	0	0	0	0	0
Access	R/W							

### Table 49.Description of S0CON bits

		•
Bit	Symbol	Description
7 to 6	SM (0:1)	<b>Mode selection bit 0 and 1.</b> Set by firmware only. The Debug UART has 4 modes (Table 50 "Debug UART modes" on page 34).
5	SM2	Multi-processor communication enable.
		Enables the multi-processor communication feature. Set by firmware only.
		In mode 2 and 3:
		if SM2 is set to logic 1, then RI will not be activated and RB8 and S0BUF will not be loaded if the 9th data bit received is a logic 0
		if SM2 is set to logic 0, it has no influence on the activation of RI and RB8
		In mode 1:
		if SM2 is set to logic 1, then RI will not be activated and RB8 and S0BUF will not be loaded if no valid stop bit was received
		if SM2 is set to logic 0, it has no influence on the activation of RI and RB8
		In mode 0, SM2 has no influence
4	REN	Serial reception enable. Set by firmware only.
		When set to logic 1, enables reception.
3	TB8	Transmit data bit. Set by firmware only.
		In modes 2 and 3, the value of TB8 is transmitted as the 9th data bit
		In modes 0 and 1, the TB8 bit is not used

Bit	Symbol	Description
2 RB8		Receive data bit. Set by hardware and by firmware.[1]
	When set to logic 1:	
		In modes 2 or 3, the hardware stores the 9th data bit that was received in RB
		In mode 1, the hardware stores the stop bit that was received in RB8
		In mode 0, the hardware does not change RB8.
1	TI	Transmit interrupt flag [3]. TI must be set to logic 0 by firmware.
		In modes 2 or 3, when transmitting, the hardware sets to logic 1 the transmit interrupt flag TI at the end of the 9th bit time
	In modes 0 or 1, when transmitting, the hardware sets to logic 1 the transmit interrupt flag TI at the end of the 8th bit time.	
0	RI	Receive interrupt flag 3. RI must be set to logic 0 by firmware.
		In modes 2 or 3, when receiving, the hardware sets to logic 1 the receive interrupt flag 1 clock period after sampling the 9th data bit (if SM2=1 setting R can be blocked, see bit description of SM2 above)
		In mode 1, when receiving, the hardware sets to logic 1 the receive interrupt flag 1 clock period after sampling the stop bit [2]
		In mode 0, when receiving, the hardware sets to logic 1 RI at the end of the CPU state 1 of the 9th machine cycle after the machine cycle where the data reception started by a write to S0CON.

### Table 49. Description of S0CON bits ... continued

- [1] If SM2 is set to logic 1, loading RB8 can be blocked, see bit description of SM2 above.
- [2] If SM2 is set to logic 1, setting RI can be blocked, see bit description of SM2 above.
- [3] The bit IE0\_4 of register IE0 (see <u>Table 13 on page 18</u>) has to be set to logic 1 to enable the corresponding CPU interrupt.

**Remark:** The SOCON register supports a locking mechanism to prevent firmware read-modify-write instructions to overwrite the contents while hardware is modifying the contents of the register.

### Table 50. Debug UART modes

Mode	SM0	SM1	Description	Baud rate
0	0	0	Shift register	f <sub>clk</sub> /6
1	0	1	8 bits Debug UART	Variable
2	1	0	9 bits Debug UART	$f_{clk}/64$ or $f_{clk}/32$
3	1	1	9 bits Debug UART	Variable

### 8.1.8.4 SOBUF register

This register is implemented twice. Writing to S0BUF writes to the transmit buffer. Reading from S0BUF reads from the receive buffer. Only hardware can read from the transmit buffer and write to the receive buffer.

### Table 51. Debug UART SOBUF Register (SFR: address 99h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol		S0BUF[7:0]						
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 52.	Description of S0BUF bits
-----------	---------------------------

Bit	Symbol	Description
7 to 0	S0BUF[7:0]	Writing to S0BUF writes to the transmit buffer.
		Reading from S0BUF reads from the receive buffer.

### 8.1.8.5 Mode 0 baud rate

In mode 0, the baud rate is derived from the CPU states signals and thus:

Baud	rate	in ı	mode	<u>ə 0</u>
------	------	------	------	------------

 $\frac{clk}{6}$ 

The next table lists the baud rates in Debug UART mode 0.

### Table 53. Baud rates in mode 0

Conditions	Min	Тур	Max	Unit
f <sub>CLK</sub>	6.78	13.56	27.12	MHz
Baud rate	1.13	2.26	4.52	Mb/s

### 8.1.8.6 Mode 2 baud rate

In mode 2, the baud rate depends on the value of bit SMOD from the SFR register PCON.

### Baud Rate using mode 2

$$\frac{2^{SMOD}}{32} \times f_{clk}$$

The next table lists the baud rates in Debug UART mode 2.

### Table 54.Baud rates in mode 2

Conditions	Min	Тур	Max	Unit
f <sub>CLK</sub>	6.78	13.56	27.12	MHz
Baud rate (SMOD=0)	212	424	847.5	kb/s
Baud rate (SMOD=1)	424	847.5	1695	kb/s

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(1)

(2)

### 8.1.8.7 Mode 1 and 3 baud rates

In modes 1 and 3, the baud rates are determined by the rate of timer1 and timer2 overflow bits: 't1\_ovf' and 't2\_ovf'. The register bit TCLK0 from the register T2CON selects if 't1\_ovf' or 't2\_ovf' should be used as a source when transmitting. The register bit RCLK0 from the register T2CON selects if 't1\_ovf' or 't2\_ovf' should be used as a source when receiving. The timers interrupt should be disabled when used to define the Debug UART baud rates.

The data rate is also dependant on the value of the bit SMOD from the SFR register PCON.

If over1rate is the equivalent 't1\_ovf' frequency and over2rate is the equivalent 't2\_ovf' frequency then:

Baud rate in mode 1 and 3 when related to timer1 overflow

(4)

$$\frac{2^{SMOD}}{32} \cdot over1rate$$

See also Section 8.1.8.8 "Baud rates using Timer1 (Debug UART mode 1 and 3)"

Baud rate in mode 1 and 3 when related to timer2 overflow

 $\frac{1}{16}$  · over2rate

See also Section 8.1.8.9 "Baud rates using Timer2 (Debug UART mode 1 and 3)"

The next table shows the trigger select:

Table 55.	Trigger select			
RCLK0	TCLK0	SMOD	receive trigger rate	transmit trigger rate
0	-	0	over1rate/32	-
0	-	1	over1rate/16	-
1	-	-	over2rate/16	-
-	0	0	-	over1rate/32
-	0	1	-	over1rate/16
-	1	-	-	over2rate/16

### 8.1.8.8 Baud rates using Timer1 (Debug UART mode 1 and 3)

The Timer1 interrupt should be disabled in this application. The Timer1 itself can be configured for either 'timer' or 'counter' operation, and in any of its 3 running modes. In the most typical applications, it is configured for 'timer' operation, in the auto-reload mode (Timer1 mode 2: high nibble of T01MOD = 0010b). In that case the baud rate is given by the formula:

Baud rate

$$\frac{2^{SMOD}}{32} \times \frac{f_{clk}}{6 \times (256 - T1H)}$$

<u>(5)</u>

<u>(6)</u>

When rewriting this formula, the value for the Timer1 reload value T1H is calculated from the desired baud rate as follows:

Timer1 reload value T1H

$$256 - \frac{2^{SMOD} \times f_{clk}}{32 \times 6 \times Baudrate}$$

One can achieve very low baud rates with Timer1 by leaving the Timer1 interrupt enabled, and configuring the timer to run as a 16-bit timer (high nibble of T01MOD = 0001b), and using the Timer1 interrupt to do a 16-bit firmware reload. Note: the frequency  $f_{clk}$  is the internal microcontroller frequency. If there is no clock divider then  $f_{clk} = f_{osc}$ .

For details on programming Timer1 to function as baud rate generator for the Debug UART see <u>Section 8.1.6 "Timer0/1 description" on page 21</u>.

The next table lists the maximum baud rates for using mode 2 of Timer1.

#### Table 56. Maximum baud rates using mode 2 of Timer1

Reload value	f <sub>CLK</sub> divided by	SMOD	Baud rate at f <sub>CLK</sub>			Unit
			6.78	13.56	27.12	MHz
FF	96	1	70.6	141.2	282.5	kb/s

The next table shows commonly used baud rates using mode 2 of Timer1 and a CLK frequency of 27.12 MHz.

#### Table 57. Baud rates using mode 2 of Timer1 with f<sub>CLK</sub> = 27.12 MHz

Reload value	$f_{CLK}$ divided by	SMOD	Baud rate at f <sub>CLK</sub>	Unit
FC	706	0	38.4	kb/s
F9	1412	0	19.2	kb/s
F1	2825	0	9.6	kb/s
E3	5650	0	4.8	kb/s
C5	11300	0	2.4	kb/s
8A	22600	0	1.2	kb/s

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## 8.1.8.9 Baud rates using Timer2 (Debug UART mode 1 and 3)

Timer2 has a programming mode to function as baud rate generator for the Debug UART. In this mode the baud rate is given by formula:

Baud rate using Timer2

 $\frac{f_{clk}}{16 \times [65536 - (T2RCH, T2RCL)]}$ 

When rewriting this formula, the value for the Timer2 reload values T2RCH/L is calculated from the desired baud rate as follows:

Reload value T2RCH/L

$$65536 - \frac{f_{clk}}{16 \times Baudrate}$$

For details on programming Timer2 to function as baud rate generator for the Debug UART (see <u>Section 8.1.7 "Timer2 description" on page 27</u>).

Note: the frequency  $f_{clk}$  is the internal microcontroller frequency. If there is no clock divider then  $f_{clk} = f_{osc}$ .

The next table lists the maximum baud rates when using Timer2.

Table 58.	Maximum	baud rates	using	Timer2
	maximum	bada rates	using	

Reload value T2RCH/L	f <sub>CLK</sub> divided by	Baud rate			Unit
		6.78	13.56	27.12	MHz
FFFF	16	424	847.5	1695	kb/s

## 8.2 General purpose IOs configurations

This chapter describes the different configurations for the IO pads:

P72, alternate function SCK

P71, alternate function MISO

P70\_IRQ

P35

P34, alternate function SIC\_CLK

P33\_INT1

P32\_INT0

P31, alternate function UART\_TX

P30, alternate function UART\_RX

Note that in Hard Power Down mode, these ports are disconnected from their supply rail.

For a given port x, there are three configuration registers:

PxCFGA[n] PxCFGB[n] Px[n]

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<u>(7)</u>

(8)

where x is 3 or 7 and n is the bit index.

At maximum 4 different controllable modes can be supported. These modes are defined with the following bits:

- PxCFGA[n]=0 and PxCFGB[n]=0: Open drain
- PxCFGA[n]=1 and PxCFGB[n]=0: Quasi Bidirectional (Reset mode)
- PxCFGA[n]=0 and PxCFGB[n]=1: input (High Impedance)
- PxCFGA[n]=1 and PxCFGB[n]=1: Push/pull output

Px[n] is used to write or read the port value.

Here is the list of the registers used for these GPIO configuration

Name	Size [bytes]	SFR address	Description	Access
P3CFGA	1	FCh	Port 3 configuration	R/W
P3CFGB	1	FDh	Port 3 configuration	R/W
P3	1	B0h	Port 3 value	R/W
P7CFGA	1	F4h	Port 7configuration	R/W
P7CFGB	1	F5h	Port 7 configuration	R/W
P7	1	F7h	Port 7 value	R/W

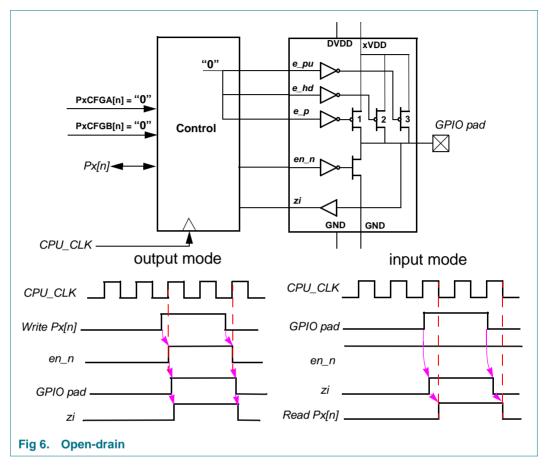
Table 59. Timer0/1 Special Function registers List

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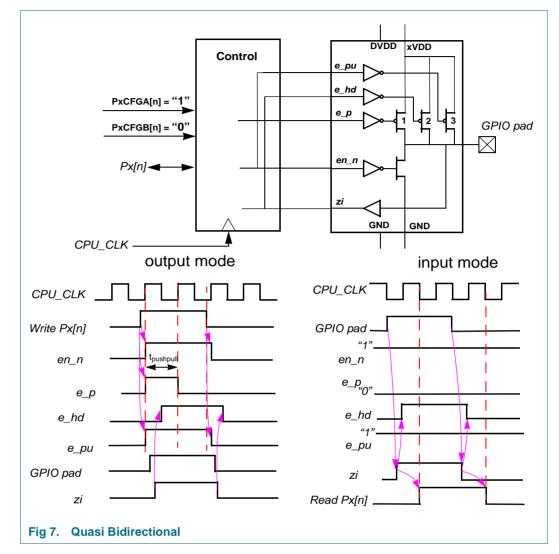
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## 8.2.1 Pad configurations description

## 8.2.1.1 Open-drain



In open drain configuration, an external pull-up resistor is required to output or read a logic 1. When writing polarity Px[n] to logic 0, the GPIO pad is pulled down to logic 0. When writing polarity Px[n] to logic 1 the GPIO pad is in High Impedance.



## 8.2.1.2 Quasi Bidirectional

In Quasi Bidirectional configuration, e\_p is driven to logic 1 for only one CPU\_CLK period when writing Px[n]. During the t<sub>pushpull</sub> time the pad drives a strong logic 1 at its output.

While zi (GPIO) is logic 1, the weak hold transistor (e\_hd) is ON, which implements a latch function. Because of the weaker nature of this hold transistor, the pad cell can now act as an input as well.

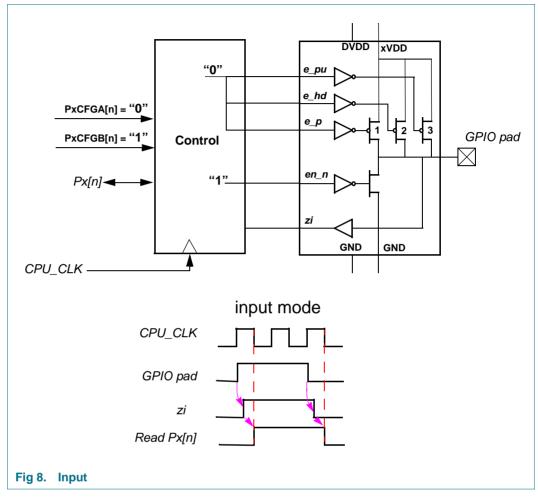
A third very weak pull-up transistor (e\_pu) ensures that an high impedance input is read as logic 1. e\_pu is clocked and is at logic 1 while Px[n] is at logic 1.

On a transition from logic 0 to logic 1 externally driven on GPIO pad, when the voltage on the pad is at the supply voltage divided by 2, zi goes to logic 1, the pull-up e\_hd is ON. e\_hd is an asynchronous signal.

The maximum currents that can be sourced by the e\_pu transistor is 80 mA and 500 mA by e\_hd transistor.

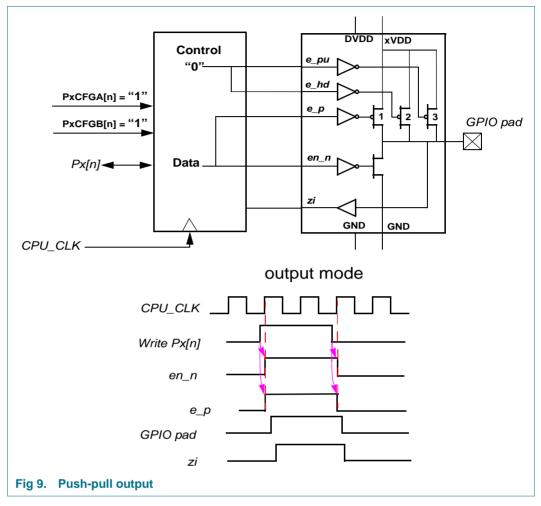
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In input configuration, no pull up or hold resistor are internally connected to the pad.

## 8.2.1.4 Push-pull output



In push-pull output, the output pin drives a strong logic 0 or a logic 1 continuously. It is possible to read back the pin output value.

## 8.2.2 GPIO registers description

#### 8.2.2.1 P7CFGA register

#### Table 60. P7CFGA register (SFR: address F4h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	-	P7CFGA[2]	P7CFGA[1]	P7CFGA[0]
Reset	1	1	1	1	1	1	1	1
Access	R	R	R	R	R	R/W	R/W	R/W

#### Table 61. Description of P7CFGA bits

Bit	Symbol	Description
7 to 3	-	Reserved.
2	P7CFGA[2]	Out of SPI mode, and in conjuction with P7CFGB[2], it configures the functional mode of the P72 pin.
1	P7CFGA[1]	Out of SPI mode, and in conjuction with P7CFGB[1], it configures the functional mode of the P71 pin.
0	P7CFGA[0]	In conjuction with P7CFGB[0], it configures the functional mode of P70_IRQ pin.

**Remark:** When in Hard power down mode, the P72 to P70\_IRQ pins are forced in quasi bidirectional mode. Referring to Figure 7, en\_n = e\_pu = "1", e\_p = "0". And e\_hd = "1" if P7x pin value is "1" and e\_hd = "0" if P7x pin value is "0".

#### 8.2.2.2 P7CFGB register

#### Table 62. P7CFGB register (SFR: address F5h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	-	P7CFGB[2]	P7CFGB[1]	P7CFGB[0]
Reset	0	0	0	0	0	0	0	0
Access	R	R	R	R	R	R/W	R/W	R/W

#### Table 63. Description of P7CFGB bits

Bit	Symbol	Description
7 to 3	-	Reserved.
2	P7CFGB[2]	Out of SPI mode, and in conjuction with P7CFGA[2], it configures the functional mode of the P72 pin.
1	P7CFGB[1]	Out of SPI mode, and in conjuction with P7CFGA[1], it configures the functional mode of the P71 pin.
0	P7CFGB[0]	In conjuction with P7CFGA[0], it configures the functional mode of P70_ IRQ pin.

**Remark:** When in Hard power down mode, the P72 to P70\_IRQ pins are forced in quasi bidirectional mode. Referring to Figure 7, en\_n = e\_pu = "1", e\_p = "0". And e\_hd = "1" if P7x pin value is "1" and e\_hd = "0" if P7x pin value is "0".

#### 8.2.2.3 P7 register

#### Table 64. P7 register (SFR: address F7h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	-	P7[2]	P7[1]	P7[0]
Reset	1	1	1	1	1	1	1	1
Access	R	R	R	R	R	R/W	R/W	R/W

#### Table 65.Description of P7 bits

Bit	Symbol	Description
	Cymson	beeenpiten
7 to 3	-	Reserved.
2	P7[2]	<b>Out of SPI mode:</b> Writing to P7[2] writes the corresponding value to the P72 pin according to the configuration mode defined by P7CFGA[2] and P7CFGB[2].
		Reading from P7[2] reads the state of P72 pin.
1	P7[1]	<b>Out of SPI mode:</b> Writing to P7[1] writes the corresponding value to the P71 pin according to the configuration mode defined by P7CFGA[1] and P7CFGB[1].
		Reading from P7[1] reads the state of P71 pin.
0	P7[0]	Writing to P7[0] writes the corresponding value to the P70_IRQ pin according to the configuration mode defined by P7CFGA[0] and P7CFGB[0].
		Reading from P7[0] reads the state of P70_IRQ pin.

#### 8.2.2.4 P3CFGA register

#### Table 66. P3CFGA register (SFR: address FCh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	P3CFGA[ 5]	P3CFGA[ 4]	P3CFGA[ 3]	P3CFGA[ 2]	P3CFGA[ 1]	P3CFGA[ 0]
Reset	1	1	1	1	1	1	1	1
Access	R	R	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 67. Description of P3CFGA bits

Bit	Symbol	Description
7 to	o 6	Reserved.
5	P3CFGA[5]	In conjuction with P3CFGB[5], it configures the functional mode of P35.
4	P3CFGA[4]	In conjuction with P3CFGB[4], it configures the functional mode of P34.
3	P3CFGA[3]	In conjuction with P3CFGB[3], it configures the functional mode of P33_INT1.
2	P3CFGA[2]	In conjuction with P3CFGB[2], it configures the functional mode of P32_INT0 <sup>[1]</sup>
1	P3CFGA[1]	In conjuction with P3CFGB[1], it configures the functional mode of P31
0	P3CFGA[0]	In conjuction with P3CFGB[0], it configures the functional mode of P30
[1]	When CPU PD is	set to logic 1(see Table 7 on page 16), for P32 INTO and referring to Section 8.2.1, e hd

When CPU\_PD is set to logic 1(see <u>Table 7 on page 16</u>), for P32\_INT0 and referring to <u>Section 8.2.1</u>, e\_hd is forced to logic 1.

**Remark:** When in Hard power down mode, the P35 to P30 pins are forced in quasi bidirectional mode. Referring to Figure 7, en\_n = e\_pu = "1", e\_p = "0". And e\_hd = "1" if P3x pin value is "1" and  $e_hd = "0"$  if P3x pin value is "0".

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#### 8.2.2.5 P3CFGB register

#### Table 68. P3CFGB register (SFR: address FDh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	P3CFGB[ 5]	P3CFGB[ 4]	P3CFGB[ 3]	P3CFGB[ 2]	P3CFGB[ 1]	P3CFGB[ 0]
Reset	0	0	0	0	0	0	0	0
Access	R	R	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 69. Description of P3CFGB bits

Bit	Symbol	Description
7 to 6		Reserved.
5	P3CFGB[5]	In conjuction with P3CFGA[5], it configures the functional mode of P35.
4	P3CFGB[4]	In conjuction with P3CFGA[4], it configures the functional mode of P34.
3	P3CFGB[3]	In conjuction with P3CFGA[3], it configures the functional mode of P33_INT1.
2	P3CFGB[2]	In conjuction with P3CFGA[2], it configures the functional mode of P32_INT0.[1]
1	P3CFGB[1]	In conjuction with P3CFGA[1], it configures the functional mode of P31.
0	P3CFGB[0]	In conjuction with P3CFGA[0], it configures the functional mode of P30.

[1] When CPU\_PD is set to logic 1(see <u>Table 7 on page 16</u>), for P32\_INT0 and referring to <u>Section 8.2.1</u>, e\_hd is forced to logic 1.

**Remark:** When in Hard power down mode, the P35 to P30 pins are forced in quasi bidirectional mode. Referring to Figure 7, en\_n = e\_pu = "1", e\_p = "0". And e\_hd = "1" if P3x pin value is "1" and e\_hd = "0" if P3x pin value is "0".

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## 8.2.2.6 P3 register

#### Table 70. P3 register (SFR: address B0h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	P3[5]	P3[4]	P3[3]	P3[2]	P3[1]	P3[0]
Reset	1	1	1	1	1	1	1	1
Access	R	R	R/W	R/W	R/W	R/W	R/W	R/W

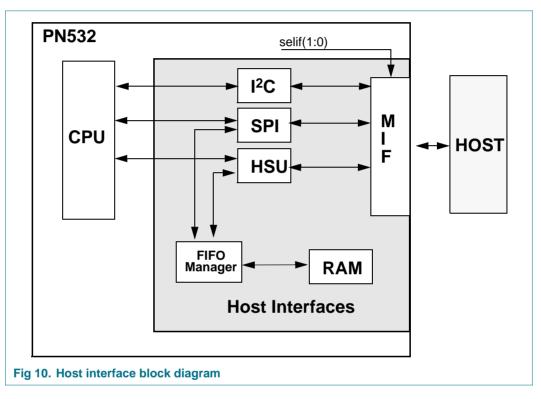
#### Table 71.Description of P3 bits

Bit	Symbol	Description
7 to 6	-	Reserved.
5	P3[5]	Writing to P3[5] writes the corresponding value to P35 pin according to the configuration mode defined by P3CFGA[5] and P3CFGB[5].
		Reading from P3[5] reads the state of P35 pin.
4	P3[4]	When P34 alternate function SIC_CLK is not used, writing to P3[4] writes the corresponding value to P34 pin according to the configuration mode defined by P3CFGA[4] and P3CFGB[4].
		Reading from P3[4] reads the state of P34 pin.
3	P3[3]	Writing to P3[3] writes the corresponding value to P33_INT1 pin according to the configuration mode defined by P3CFGA[3] and P3CFGB[3].
		Reading from P3[3] reads the state of P33_INT1 pin.
2	P3[2]	Writing to P3[2] writes the corresponding value to P32_INT0 pin according to the configuration mode defined by P3CFGA[2] and P3CFGB[2].
		Reading from P3[2] reads the state of P32_INT0 pin.
1	P3[1]	When the P31 pin alternate function UART_TX is not used, writing to P3[1] writes the corresponding value to P31 pin according to the configuration mode defined by P3CFGA[1] and P3CFGB[1].
		Reading from P3[1] reads the state of P31 pin.
0	P3[0]	When the P30 pin alternate function UART_RX is not used, writing to P3[0] writes the corresponding value to P30 pin according to the configuration mode defined by P3CFGA[0] and P3CFGB[0].
		Reading from P3[0] reads the state of P30 pin.

## 8.3 Host interfaces

PN532 must be able to support different kind of interfaces to communicate with the HOST. All the interfaces that have to be supported are exclusive.

- SPI interface
- I<sup>2</sup>C interface: Standard and Fast modes
- High Speed UART (HSU): supporting specific high baud rates



## 8.3.1 Multi-InterFace (MIF) description

The Multi-InterFace (MIF) manages the configuration of the host interface pins, supplied by PVDD, according to the selected links with the bits selif[1:0] of register Config\_I0\_I1 (see <u>Table 74 on page 49</u>):

The firmware must copy the value of the pads I0 and I1 to respectively selif[0] and selif[1].

Selif [1:0]	00	01	10	11
Host interface selected	HSU	SPI	l <sup>2</sup> C	Reserved
Pin number				
27	HSU_RX	NSS	P50_SCL	-
28	HSU_TX	MOSI	SDA	-
29	P71	MISO	P71	-
30	P72	SCK	P72	-

#### Table 72. HOST interface selection

#### 8.3.1.1 MIF register

The Config I0\_I1 register is used to select the host interface. It manages also the polarity of P33\_INT1.

Table 73.	Config I0_I1	register (addres	ss 6103h) bit allocation
-----------	--------------	------------------	--------------------------

Bit	7	6	5	4	3	2	1	0
Symbol	int1_pol	-	pad_l1	-	pad_l0	enselif	Selif	[1:0]
Reset	0	0	Х	0	Х	0	0	0
Access	R/W	R	R/W	R	R/W	R/W	R/W	R/W

## Table 74. Description of Config I0\_I1 bits

Bit	Symbol	Description
7	int1_pol	When set to logic 1, the value of the P33_INT1 pin is inverted.
6	-	Reserved.
5	pad_l1	When read this bit gives the state of the I1 pin.
4	-	Reserved.
3	pad_I0	When read this bit gives the state of the I0 pin.
2	enselif	When set to logic 1, this bit indicates that the selif bits are valid and that the selected interface on the MIF can drive the pins. The firmware must copy the value of the pads I0 and I1 to respectively selif[0] and selif[1]
		When set to logic 0, the MIF cannot drive the IO lines.
1:0	Selif[1:0]	These bits are used by the firmware to select the host interface communication link, see Table 72 on page 48.

#### 8.3.1.2 Configuration modes of the host interface pins.

In I<sup>2</sup>C mode, P50\_SCL and SDA are configured in Open Drain mode.

In HSU mode, HSU\_RX is in input mode and HSU\_TX is in push-pull mode.

In SPI mode, NSS, MOSI and SCK are in inputs mode. MISO is in push-pull mode.

## 8.3.2 I<sup>2</sup>C interface

It is recommended to refer the I<sup>2</sup>C standard for more information.

The I<sup>2</sup>C interface implements a Master/Slave I<sup>2</sup>C bus interface with integrated shift register, shift timing generation and Slave address recognition. I<sup>2</sup>C Standard mode (100 kHz SCLK) and Fast mode (400 kHz SCLK) are supported.

General Call +W is supported, not hardware General Call (GC +R).

The mains characteristics of the I<sup>2</sup>C module are:

- Support Master/Slave I<sup>2</sup>C bus
- Standard and Fast mode supported
- Wake-up of the PN532 on its own address
- Wake-up on General Call +W (GC +W)

The I<sup>2</sup>C module is control through 5 registers:

#### Table 75. I<sup>2</sup>C register list

Name	Size [bytes]	Address	Description	Access
I <sup>2</sup> CCON	1	D8h (SFR)	Control register	R/W
I <sup>2</sup> CSTA	1	D9h (SFR)	Status register	R/W
I <sup>2</sup> CDAT	1	DAh (SFR)	Data register	R/W
I <sup>2</sup> CADR	1	DBh (SFR)	Slave Address register	R/W
i <sup>2</sup> c_wu_control	1	610Ah	Control register for the $I^2C$ wake-up conditions	R/W

#### 8.3.2.1 I<sup>2</sup>C functional description

The I<sup>2</sup>C interface may operate in any of the following four modes:

- Master Transmitter
- Master Receiver
- Slave Receiver
- Slave Transmitter

Two types of data transfers are possible on the I<sup>2</sup>C bus:

- Data transfer from a Master transmitter to a Slave receiver. The first byte transmitted by the Master is the Slave address. Next follows a number of data bytes. The Slave returns an acknowledge bit after each received byte.
- Data transfer from a Slave transmitter to a Master receiver. The first byte (the Slave address) is transmitted by the Master. The Slave then returns an acknowledge bit. Next follows the data bytes transmitted by the Slave to the Master. The Master returns an acknowledge bit after each received byte except the last byte. At the end of the last received byte, a "not acknowledge" is returned.

In a given application, the I<sup>2</sup>C interface may operate as a Master or as a Slave.

In the PN532, the I<sup>2</sup>C is typically configured as a Slave, because the host is Master.

In the Slave mode, the I<sup>2</sup>C interface hardware looks for its own Slave address and the general call address. If one of these addresses is detected, an interrupt is requested. When the PN532 microcontroller wishes to become the bus Master, the hardware waits until the bus is free before the Master mode is entered so that a possible Slave action is not interrupted. If bus arbitration is lost in the Master mode, the I<sup>2</sup>C interface switches to the Slave mode immediately and can detect its own Slave address in the same serial transfer.

## 8.3.2.2 Master transmitter mode

As a Master, the  $I^2C$  logic will generate all of the serial clock pulses and the START and STOP conditions. A transfer is ended with a STOP condition or with a repeated START condition. Since a repeated START condition is also the beginning of the next serial transfer, the  $I^2C$  bus will not be released.

I<sup>2</sup>C data are output through SDA while P50\_SCL outputs the serial clock. The first byte transmitted contains the Slave address of the receiving device (7-bit SLA) and the data direction bit. In this case the data direction bit (R/W) will be a logic '0' (W). I<sup>2</sup>C data are transmitted 8 bits at a time. After each byte is transmitted, an acknowledge bit is received. START and STOP conditions are output to indicate the beginning and the end of a serial transfer.

In the Master transmitter mode, a number of data bytes can be transmitted to the Slave receiver. Before the Master transmitter mode can be entered, I<sup>2</sup>CCON must be initialized with the ENS1 bit set to logic 1 and the STA, STO and SI bits set to logic 0. ENS1 must be set to logic 1 to enable the I<sup>2</sup>C interface. If the AA bit is set to logic 0, the I<sup>2</sup>C interface will not acknowledge its own Slave address or the general call address if they are present on the bus. This will prevent the I<sup>2</sup>C interface from entering a Slave mode.

The Master transmitter mode may now be entered by setting the STA bit. The  $I^2C$  interface logic will then test the  $I^2C$  bus and generate a start condition as soon as the bus becomes free. When a START condition is transmitted, the serial interrupt flag (SI) is set to logic 1, and the status code in the status register ( $I^2CSTA$ ) will be 08h. This status code must be used to vector to an interrupt service routine that loads  $I^2CDAT$  with the Slave address and the data direction bit (SLA+W). The SI bit in  $I^2CCON$  must then be set to logic 0 before the serial transfer can continue.

When the Slave address and the direction bit have been transmitted and an acknowledgment bit has been received, the serial interrupt flag (SI) is set to logic 1 again, and a number of status codes in I<sup>2</sup>CSTA are possible. The appropriate action to be taken for any of the status codes is detailed in <u>Table 80 on page 58</u>. After a repeated start condition (state 10h), the I<sup>2</sup>C interface may switch to the Master receiver mode by loading I<sup>2</sup>CDAT with SLA+R.

## 8.3.2.3 Master receiver mode

As a Master, the  $I^2C$  logic will generate all of the serial clock pulses and the START and STOP conditions. A transfer is ended with a STOP condition or with a repeated START condition. Since a repeated START condition is also the beginning of the next serial transfer, the  $I^2C$  bus will not be released.

The first byte transmitted contains the Slave address of the transmitting device (7-bit SLA) and the data direction bit. In this case the data direction bit (R/W) will be logic 1 (R). I<sup>2</sup>C data are received via SDA while P50\_SCL outputs the serial clock. I<sup>2</sup>C data are received 8 bits at a time. After each byte is received, an acknowledge bit is transmitted. START and STOP conditions are output to indicate the beginning and end of a serial transfer.

In the Master receiver mode, a number of data bytes are received from a Slave transmitter. The transfer is initialized as in the Master transmitter mode. When the START condition has been transmitted, the interrupt service routine must load I<sup>2</sup>CDAT with the 7-bit Slave address and the data direction bit (SLA+R). The SI bit in I<sup>2</sup>CCON must then be set to logic 0 before the serial transfer can continue.

When the Slave address and the data direction bit have been transmitted and an acknowledgment bit has been received, the serial interrupt flag (SI) is set to logic 1 again, and a number of status codes are possible in I<sup>2</sup>CSTA. The appropriate action to be taken for each of the status codes is detailed in <u>Table 81 on page 59</u>. After a repeated start condition (state 10h), the I<sup>2</sup>C interface may switch to the Master transmitter mode by loading I<sup>2</sup>CDAT with SLA+W.

## 8.3.2.4 Slave receiver mode

I<sup>2</sup>C data and the serial clock are received through SDA and P50\_SCL. After each byte is received, an acknowledge bit is transmitted. START and STOP conditions are recognized as the beginning and end of a serial transfer. Address recognition is performed by hardware after reception of the Slave address and direction bit.

In the Slave receiver mode, a number of data bytes are received from a Master transmitter. To initiate the Slave receiver mode, I<sup>2</sup>CADR must be loaded with the 7-bit Slave address to which the I<sup>2</sup>C interface will respond when addressed by a Master. Also the least significant bit of I<sup>2</sup>CADR should be set to logic 1 if the interface should respond to the general call address (00h). The control register, I<sup>2</sup>CCON, should be initialized with ENS1 and AA set to logic 1 and STA, STO, and SI set to logic 0 in order to enter the Slave receiver mode. Setting the AA bit will enable the logic to acknowledge its own Slave address or the general call address and ENS1 will enable the interface.

When I<sup>2</sup>CADR and I<sup>2</sup>CCON have been initialized, the I<sup>2</sup>C interface waits until it is addressed by its own Slave address followed by the data direction bit which must be '0' (W) for the I<sup>2</sup>C interface to operate in the Slave receiver mode. After its own Slave address and the W bit have been received, the serial interrupt flag (SI) is set to logic 1 and a valid status code can be read from I<sup>2</sup>CDAT. This status code should be used to vector to an interrupt service routine, and the appropriate action to be taken for each of the status codes is detailed in <u>Table 82 on page 60</u>. The Slave receiver mode may also be entered if arbitration is lost while the I<sup>2</sup>C interface is in the Master mode.

If the AA bit is set to logic 0 during a transfer, the I<sup>2</sup>C interface will return a not acknowledge (logic 1) to SDA after the next received data byte. While AA is set to logic 0, the I<sup>2</sup>C interface does not respond to its own Slave address or a general call address. However, the I<sup>2</sup>C bus is still monitored and address recognition may be resumed at any time by setting AA. This means that the AA bit may be used to temporarily isolate the I<sup>2</sup>C interface from the I<sup>2</sup>C bus.

#### 8.3.2.5 Slave transmitter mode

The first byte is received and handled as in the Slave receiver mode. However, in this mode, the direction bit will indicate that the transfer direction is reversed. I<sup>2</sup>C data are transmitted via SDA while the serial clock is input through P50\_SCL. START and STOP conditions are recognized as the beginning and end of a serial transfer.

In the Slave transmitter mode, a number of data bytes are transmitted to a Master receiver. Data transfer is initialized as in the Slave receiver mode. When I<sup>2</sup>CADR and I<sup>2</sup>CCON have been initialized, the I<sup>2</sup>C interface waits until it is addressed by its own Slave address followed by the data direction bit which must be '1' (R) for the I<sup>2</sup>C interface to operate in the Slave transmitter mode. After its own Slave address and the R bit have been received, the serial interrupt flag (SI) is set to logic 1 and a valid status code can be read from I<sup>2</sup>CSTA. This status code is used to vector to an interrupt service routine, and the appropriate action to be taken for each of these status codes is detailed in <u>Table 83 on page 62</u>. The Slave transmitter mode.

If the AA bit is set to logic 0 during a transfer, the I<sup>2</sup>C interface will transmit the last byte of the transfer and enter state C0h or C8h. the I<sup>2</sup>C interface is switched to the not addressed Slave mode and will ignore the Master receiver if it continues the transfer. Thus the Master receiver receives all '1's as I<sup>2</sup>C data. While AA is set to logic 0, the I<sup>2</sup>C interface does not respond to its own Slave address or a general call address. However, the I<sup>2</sup>C bus is still monitored, and address recognition may be resumed at any time by setting AA. This means that the AA bit may be used to temporarily isolate the I<sup>2</sup>C interface from the I<sup>2</sup>C bus.

#### 8.3.2.6 I<sup>2</sup>C wake-up mode

The wake up block can only be used when I<sup>2</sup>C is configured as a Slave.

It is a dedicated circuitry, separated from the main I<sup>2</sup>C peripheral which functionality is to wake-up the PN532 from Soft-Power-Down mode.

Before entering the Soft-Power-Down mode, the following actions must be taken:

- Enable the block and select the wake-up conditions (see Table 90 on page 65).
- Enable the I<sup>2</sup>C wake-up event in the PCR (see <u>Table 143 on page 97</u>)

Once in Soft-Power-Down mode, the wake up block will monitor the I<sup>2</sup>C bus. If it recognizes its own address and the command type is valid (read only, write only, or both depending of settings in register i<sup>2</sup>c\_wu\_control, see <u>Table 90 on page 65</u>), the wake up block will generate an acknowledge, stretch P50\_SCL, configure the I<sup>2</sup>C interface in Slave Transmitter or Slave Receiver mode depending on the command. Finally, i<sup>2</sup>c\_on is set to logic 1, which initiates the wake-up sequence (see <u>Section 8.5 "Power clock and reset</u> controller" on page 90).

When the microcontroller has been woken up, the firmware must identify the wake up source and must disable the wake up block (see <u>Table 90 on page 65</u>) to use  $I^2C$ . It is now the  $I^2C$  peripheral which stretches P50\_SCL.

To enable wake up on GC +W, the LSB bit of  $I^2CADR$  should be set to logic 1 (see <u>Table 88 on page 65</u>). The wake-up block and the wake-up on a write command should be enabled before entering in Soft-Power-Down mode. When the wake up on GC +W condition is recognized, the behavior is the same as described above.

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## 8.3.2.7 I<sup>2</sup>CCON register

The CPU can read from and write to this 8-bit SFR. Two bits are affected by the Serial IO (the  $I^2C$  interface) hardware: the SI bit is set to logic 1 when a serial interrupt is requested, and the STO bit is set to logic 0 when a STOP condition is present on the  $I^2C$  bus. The STO bit is also set to logic 0 when ENS1 = '0'.

#### Table 76. I<sup>2</sup>CCON register (SFR: address D8h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	CR[2]	ENS1	STA	STO	SI	AA	CR[	1:0]
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 77. Description of I<sup>2</sup>CCON bits

Bit	Symbol	Description
7	CR[2]	Serial clock frequency selection in Master mode. Together with CR[1:0], this bit determines the clock rate (serial clock frequency) when the $I^2C$ interface is in a Master mode. Special attention has to be made on the $I^2C$ bit frequency in case of dynamic switching of the CPU clock frequency.
6	ENS1	<b>Serial IO enable.</b> When ENS1 bit is to logic 0, SDA and P50_SCL are in high impedance. The state of SDA and P50_SCL is ignored, the I <sup>2</sup> C interface is in the "not addressed" Slave state, and the STO bit in I <sup>2</sup> CCON is forced to logic 0. No other bits are affected.
		When ENS1 is logic 1, the I <sup>2</sup> C interface is enabled, assuming selif[1:0] bits are 10b (see <u>Table 72 on page 48</u> ).
		ENS1 should not be used to temporarily release the $I^2C$ interface from the $I^2C$ bus since, when ENS1 is set to logic 0, the $I^2C$ bus status is lost. The AA flag should be used instead.
5	STA	<b>START control.</b> When the STA bit is set to logic 1 to enter Master mode, the $I^2C$ interface hardware checks the status of the $I^2C$ bus and generates a START condition if the bus is free. If the bus is not free, then the $I^2C$ interface waits for a STOP condition (which will free the bus) and generates a START condition after a delay of a half clock period of the internal serial clock generator.
		If STA is set to logic 1, while the I <sup>2</sup> C interface is already in a Master mode and one or more bytes are transmitted or received, the I <sup>2</sup> C interface transmits a repeated START condition.
		STA may be set to logic 1 at any time. This includes the case when the I <sup>2</sup> C interface is the addressed Slave.
		When the STA bit is set to logic 0, no START condition or repeated START condition will be generated.

#### Table 77. Description of I<sup>2</sup>CCON bits ...continued

Bit Symbol Des
----------------

STO STOP control. When the STO bit is set to logic 1, while the l<sup>2</sup>C interface is in Master mode, a STOP condition is transmitted to the l<sup>2</sup>C bus. When the STOP condition is detected on the bus, the l<sup>2</sup>C interface hardware automatically sets STO to logic 0.
 In Slave mode, STO may be set to logic 1 to recover from an error condition. In this case, no STOP condition is transmitted to the l<sup>2</sup>C bus. However, the l<sup>2</sup>C interface hardware behaves as if a STOP condition has been received and switches to the defined "not addressed" Slave Receiver mode.
 If the STA and STO bits are both set to logic 1, the STOP condition is transmitted to the l<sup>2</sup>C bus if the l<sup>2</sup>C interface is in Master mode (in Slave mode, the l<sup>2</sup>C interface generates an internal STOP condition which is not transmitted). The l<sup>2</sup>C interface then transmits a START condition.

When the STO bit is set to logic 0, no STOP condition will be generated.

3 SI Serial interrupt flag. When SI is set to logic 1, then if the serial interrupt from the I<sup>2</sup>C interface port is enabled, the CPU will receive an interrupt. SI is set by hardware when any one of 25 of the possible 26 states of the I<sup>2</sup>C interface are entered. The only state that does not cause SI to be set to logic 1 is state F8h. which indicates that no relevant state information is available. While SI is set by hardware to logic 1, P50\_SCL is held in logic 0 when the SCL line is logic 0, and P50\_SCL is held in high impedance when the SCL line is logic 1. SI must be set to logic 0 by firmware. When the SI flag is set to logic 0, no serial interrupt is requested, and there is no stretching of the SCL line via P50\_SCL. The bit IE1\_4 of register IE1 (see Table 13 on page 18) has also to be set to logic 1 to enable the corresponding I<sup>2</sup>C interrupt to the CPU. 2 AA Assert Acknowledge flag. If AA is set to logic 1, an acknowledge (low level to SDA) will be returned during the acknowledge clock pulse on the P50 SCL line when: The "own Slave address" has been received. • • The general call address has been received while the general call bit (GC) in I<sup>2</sup>CADR is set. A data byte has been received while the I<sup>2</sup>C interface is in Master Receiver mode A data byte has been received while the I<sup>2</sup>C interface is in the addressed Slave Receiver mode. When the I<sup>2</sup>C interface is in the addressed Slave Transmitter mode, state C8h will be entered after the last serial bit is transmitted. When SI is set to logic 0, the I<sup>2</sup>C interface leaves state C8h, enters the Not-addressed Slave Receiver mode. and the SDA line remains at logic 1. In state C8h, AA can be set to logic 1 again for future address recognition. When the I<sup>2</sup>C interface is in the Not-addressed Slave mode, its own Slave address and the general call address are ignored. Consequently, no acknowledge is returned, and a serial interrupt is not requested. Thus, the I<sup>2</sup>C interface can be temporarily released from the I<sup>2</sup>C bus while the bus status is monitored. While the I<sup>2</sup>C interface is released from the bus, START and STOP

> conditions are detected, and I<sup>2</sup>C data are shifted in. Address recognition can be resumed at any time by setting AA to logic 1. If AA is set to logic 1 when the I<sup>2</sup>C own Slave address or the general call address has been partly received, the address will be recognized at the end of the byte

Bit	Symbol	Desc	Description												
1 to 0	CR[1:0]	Seria	l cloc	k freq	uency selection in Master m	ode.									
		CR2	CR1	CR0	CPU_CLK division factor	2C bit frequency									
		0	0	0	10	CPU_CLK/10									
		0	0	1	20	CPU_CLK/20									
		0	1	0	30	CPU_CLK/30									
		0	1	1	40	CPU_CLK/40									
		1	0	0	80	CPU_CLK/80									
		1	0	1	120	CPU_CLK/120									
		1	1	0	160	CPU_CLK/160									
		1	1	1	(256-T1 reload value)*12 243072	CPU_CLK/3072CPU_CLK/24									

## Table 77. Description of I<sup>2</sup>CCON bits ...continued

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## 8.3.2.8 I<sup>2</sup>CSTA register

I<sup>2</sup>CSTA is an 8-bit read-only special function register. The three least significant bits are always at logic 0. The five most significant bits contain the status code. There are 26 possible status codes. When I<sup>2</sup>CSTA contains F8h, no relevant state information is available and no serial interrupt is requested. Reset initializes I<sup>2</sup>CSTA to F8h. All other I<sup>2</sup>CSTA values correspond to defined I<sup>2</sup>C interface states. When each of these states is entered, a serial interrupt is requested (SI = '1'), this can happen in any CPU cycle, and a valid status code will be present in I<sup>2</sup>CSTA. This status code will remain present in I<sup>2</sup>CSTA until SI is set to logic 0 by firmware.

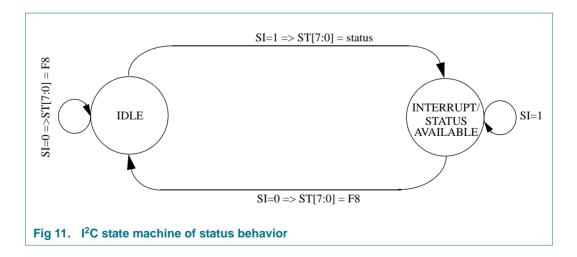
Note that I<sup>2</sup>CSTA changes one CPU\_CLK clock cycle after SI changes, so the new status can be visible in the same machine cycle SI changes or possibly (in one out of six CPU states) the machine cycle after that. This should not be a problem since you should not read I<sup>2</sup>CSTA before either polling SI or entry of the interrupt handler (which in itself takes several machine cycles).

Bit	7	6	5	4	3	2	1	0				
Symbol	ST[7:0]											
Reset	1	1	1	1	1	0	0	0				
Access	R	R	R	R	R	R	R	R				

#### Table 78. I<sup>2</sup>CSTA register (SFR: address D9h) bit allocation

#### Table 79. Description of I<sup>2</sup>CSTA bits

Bit	Symbol	Description
7 to 0	ST[7:0]	<b>Encoded status bit for the different functional mode.</b> Several Status codes are returned in a certain mode (Master Transmitter, Master Receiver, Slave Transmitter, Slave Receiver) plus some miscellaneous status codes that can be returned at any time.



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## Table 80. I<sup>2</sup>C Master Transmitter Mode status code

	Status of the I <sup>2</sup> C Bus	Application firmw	are R	espon	se		Next Action Taken By the I <sup>2</sup> C interface Hardware
Code	and of the	To/from I <sup>2</sup> CDAT	TO I <sup>2</sup>	CCON	I		
51[7:0]	I <sup>2</sup> C interface Hardware		STA	STO	SI	AA	
08h	A START condition has been transmitted	Load SLA+W	Х	0	0	Х	SLA+W will be transmitted ACK will be received
10h	A repeated START condition has been transmitted	Load SLA+W	Х	0	0	Х	As above
		Load SLA+R	Х	0	0	Х	SLA+W will be transmitted; the I <sup>2</sup> C interface will be switched to MST/(TRX or REC) mode
18h	SLA+W has been transmitted;	Load data byte	0	0	0	Х	Data byte will be transmitted; ACK bit will be received
	ACK has been received	No I <sup>2</sup> CDAT action	1	0	0	Х	Repeated START will be transmitted
		No I <sup>2</sup> CDAT action	0	1	0	Х	STOP condition will be transmitted STO flag will be set to logic 0
		No I <sup>2</sup> CDAT action	1	1	0	Х	STOP condition followed by a START condition will be transmitted STO flag will be set to logic 0
20h	SLA+W has been transmitted; NOT ACK has been received	Load data byte	0	0	0	Х	Data byte will be transmitted ACK bit will be received
		No I <sup>2</sup> CDAT action	1	0	0	Х	Repeated START will be transmitted
		No I <sup>2</sup> CDAT action	0	1	0	Х	STOP condition will be transmitted STO flag will be set to logic 0
		No I <sup>2</sup> CDAT action	1	1	0	Х	STOP condition followed by a START condition will be transmitted STO flag will be set to logic 0
28h	Write data byte in I <sup>2</sup> CDAT has	Load data byte	0	0	0	Х	Data byte will be transmitted; ACK bit will be received
	been transmitted;	No I <sup>2</sup> CDAT action	1	0	0	Х	Repeated START will be transmitted
	ACK has been received	No I <sup>2</sup> CDAT action	0	1	0	Х	STOP condition will be transmitted STO flag will be set to logic 0
		No I <sup>2</sup> CDAT action	1	1	0	Х	STOP condition followed by a START condition will be transmitted STO flag will be set to logic 0

#### Table 80. I<sup>2</sup>C Master Transmitter Mode status code ...continued PZ

Status	Status of the I <sup>2</sup> C Bus	Application firmw	are R	espon	se		Next Action Taken By the I <sup>2</sup> C interface Hardware
Code	and of the I <sup>2</sup> C interface Hardware		TO I <sup>2</sup>	CCON	l		
51[7.0]	I C Internace Haidware		STA	STO	SI	AA	
30h	Write data byte in I <sup>2</sup> CDAT has been transmitted; NOT ACK has been received	Load data byte	0	0	0	Х	Data byte will be transmitted; ACK bit will be received
		No I <sup>2</sup> CDAT action	1	0	0	Х	Repeated START will be transmitted
		No I <sup>2</sup> CDAT action	0	1	0	Х	STOP condition will be transmitted STO flag will be set to logic 0
		No I <sup>2</sup> CDAT action	1	1	0	Х	STOP condition followed by a START condition will be transmitted STO flag will be set to logic 0
38h	Arbitration lost in SLA+R/W or	No I <sup>2</sup> CDAT action	0	0	0	Х	I <sup>2</sup> C bus will be released; a Slave mode will be entered
	Data bytes	No I <sup>2</sup> CDAT action	1	0	0	Х	A START condition will be transmitted when the bus becomes free

## Table 81. I<sup>2</sup>C Master Receiver Mode status codes

mation pro	Status	Status of the I <sup>2</sup> C Bus and	Application firmw	are R	espon	se		Next Action Taken By the I <sup>2</sup> C interface Hardware
be	Code	the I <sup>2</sup> C interface Hardware	To /from I <sup>2</sup> CDAT	TO I <sup>2</sup>	CCON	I		
n this c	ST[7:0]			STA	STO	SI	AA	
nent is	08h	A START condition has been transmitted	Load SLA+W	Х	0	0	Х	SLA+W will be transmitted, ACK will be received
subject to	10h	A repeated START condition	Load SLA+W	Х	0	0	Х	As above
t to legal disc	h	has been transmitted	Load SL+R	Х	0	0	Х	SLA+W will be transmitted; the I <sup>2</sup> C interface will be switched to MST/(TRX or REC) mode
	38h	Arbitration lost in SLA+R/W or Data bytes	No I <sup>2</sup> CDAT action	0	0	0	Х	I <sup>2</sup> C bus will be released; a Slave mode will be entered
S.			No I <sup>2</sup> CDAT action	1	0	0	Х	A START condition will be transmitted when the bus becomes free
	40h	SLA+R has been transmitted; ACK has been received	No I <sup>2</sup> CDAT action	0	0	0	0	Data byte will be received; NOT ACK bit will be returned
			No I <sup>2</sup> CDAT action	0	0	0	1	Data byte will be received; ACK bit will be returned
	48h	SLA+R has been transmitted;	No I <sup>2</sup> CDAT action	1	0	0	Х	Repeated START condition will be transmitted
© NXP		NOT ACK has been received	No I <sup>2</sup> CDAT action	0	1	0	Х	STOP condition will be transmitted; STO flag will be set to logic 0
B.V. 2017. All rights reserved			No I <sup>2</sup> CDAT action	1	1	0	Х	STOP condition followed by a START condition will be transmitted; STO flag will be set to logic 0

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## Table 81. I<sup>2</sup>C Master Receiver Mode status codes ...continued

	Status of the I <sup>2</sup> C Bus and	Application firmw	are R	espon	se		Next Action Taken By the I <sup>2</sup> C interface Hardware
Code	the I <sup>2</sup> C interface Hardware	To /from I <sup>2</sup> CDAT	TO I <sup>2</sup>	CCON	I		
ST[7:0]			STA	STO	SI	AA	
50h	Read data byte has been received; ACK has been returned	Read data byte or	0	0	0	0	Data byte will be received; NOT ACK bit will be returned
		Read data byte	0	0	0	1	Data byte will be received; ACK bit will be returned
58h	Read data byte has been	Read data byte	1	0	0	Х	Repeated START condition will be transmitted
	received; NOT ACK has been returned	Read data byte	0	1	0	Х	STOP condition will be transmitted; STO flag will be set to logic 0
		Read data byte	1	1	0	Х	STOP condition followed by a START condition will be transmitted; STO flag will be set to logic 0

## Table 82. I<sup>2</sup>C Slave Receiver Mode status codes

1	Status of the I <sup>2</sup> C Bus and	Application firmw	are R	espor	ise		Next Action Taken By the I <sup>2</sup> C interface Hardware		
Code	the I <sup>2</sup> C interface Hardware	To /from I <sup>2</sup> CDAT	TO I <sup>2</sup>		1				
Code ST[7:0] 60h			STA	STO	SI	AA			
60h	Own SLA+W has been	No I <sup>2</sup> CDAT action	Х	0	0	0	Data byte will be received an NOT ACK will be returned		
68h	received; ACK has been returned	No I <sup>2</sup> CDAT action	Х	0	0	1	Data bye will be received and ACK will be returned		
	Arbitration lost in SLA+R/W as Master; Own SLA+W has been received, ACK returned	No I <sup>2</sup> CDAT action	Х	0	0	0	Data byte will be received an NOT ACK will be returned		
		No I <sup>2</sup> CDAT action	x	0	0	1	Data byte will be received and ACK will be returned		
70h	0h General call address (00h) has been received; ACK has been returned	No I <sup>2</sup> CDAT action	Х	0	0	0	Data byte will be received and NOT ACK will be returned		
		No I <sup>2</sup> CDAT action	Х	0	0	1	Data byte will be received and ACK will be returned		
78h	Maatari Canaral aall addraaa	No I <sup>2</sup> CDAT action	Х	0	0	0	Data byte will be received an NOT ACK will be returned		
		No I <sup>2</sup> CDAT action	Х	0	0	1	Data byte will be received and ACK will be returned		
80h	Previously addressed with own	Read data byte	Х	0	0	0	Data byte will be received an NOT ACK will be returned		
NYD D V 2017 All risks soorood	SLA; Write data byte has been received; ACK has been returned	Read data byte	Х	0	0	1	Data byte will be received and ACK will be returned		

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	Code ST[7:0]	the I <sup>2</sup> C interface Hardware	To /from I <sup>2</sup> CDAT	TO I <sup>2</sup>	CCON	I		
	01[7.0]			STA	STO	SI	AA	
	88h	Previously addressed with own SLA; Write data byte has been received; NOT ACK has been returned	Read data byte	0	0	0	0	Switched to not addressed SLV mode; No recognition of own SLA or General call address
			Read data byte	0	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if I <sup>2</sup> CADR[0] is set to logic 1.
			Read data byte	1	0	0	0	Switched to not addressed SLV mode; No recognition of own SLA or General call address. A START condition will be transmitted when the bus becomes free
			Read data byte	1	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if I <sup>2</sup> CADR[0] is set to logic 1. A START condition will be transmitted when the bus becomes free
	90h	Previously addressed with General Call; Write data byte has been received; ACK has been returned	Read data byte	Х	0	0	0	Data byte will be received and NOT ACK will be returned
			Read data byte	Х	0	0	1	Data byte will be received and ACK will be returned
	98h	Previously addressed with General Call; Write data byte	Read data byte	0	0	0	0	Switched to not addressed SLV mode; no recognition of own SLA or General call address
		has been received; NOT ACK has been returned	Read data byte	0	0	0	1	Switched to not addressed SLV mode; own SLA will be recognized; General call address will be recognized if I <sup>2</sup> CADR[0] is set to logic 1.
to logal disclaimore			Read data byte	1	0	0	0	Switched to not addressed SLV mode; no recognition of own SLA or General call address. A START condition will be transmitted when the bus becomes free
5			Read data byte	1	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if I <sup>2</sup> CADR[0] is set to logic 1. A START condition will be transmitted when the bus becomes free

## Table 82. I<sup>2</sup>C Slave Receiver Mode status codes ...continued

	Status of the I <sup>2</sup> C Bus and	Application firmw	are R	espon	se		Next Action Taken By the I <sup>2</sup> C interface Hardware
	the I <sup>2</sup> C interface Hardware	To /from I <sup>2</sup> CDAT	TO I <sup>2</sup>	CCON	l		
T[7:0]			STA	STO	SI	AA	
\0h	A STOP condition or repeated START condition has been received while still addressed as SLV/(REC or TRX)	Read data byte	0	0	0	0	Switched to not addressed SLV mode; no recognition of own SLA or General call address
		Read data byte	0	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if I <sup>2</sup> CADR[0] is set to logic 1
		Read data byte	1	0	0	0	Switched to not addressed SLV mode; No recognition of own SLA or General call address. A START condition will be transmitted when the bus becomes free
		Read data byte	1	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if I <sup>2</sup> CADR[0] is set to logic 1 A START condition will be transmitted when the bus becomes free

## Table 83. I<sup>2</sup>C Slave Transmitter Mode status codes

Status	Status of the I <sup>2</sup> C Bus and	Application firmw	are R	espon	nse		Next Action Taken By the I <sup>2</sup> C interface Hardware
Code	the I <sup>2</sup> C interface Hardware	To /from I <sup>2</sup> CDAT	TO I <sup>2</sup>	CCON	1		
ST[7:0]			STA	STO	SI	AA	
A8h	8h Own SLA+R has been received; ACK has been returned	Load data byte	Х	0	0	0	Last data byte will be transmitted and ACK bit will be received
		Load data byte	Х	0	0	1	Data byte will be transmitted; ACK will be received
B0h	Arbitration lost in SLA+R/W as	Load data byte	Х	0	0	0	Last data byte will be transmitted and ACK bit will be received
	Master; Own SLA+R has been received. ACK has been returned	Load data byte	Х	0	0	1	Data byte will be transmitted; ACK will be received
B8h	Read data bye in I <sup>2</sup> CDAT has	Load data byte	Х	0	0	0	Last data byte will be transmitted and ACK bit will be received
	been transmitted; ACK has been received	Load data byte	Х	0	0	1	Data byte will be transmitted; ACK will be received

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		Status of the I <sup>2</sup> C Bus and	Application firmw	are R	espon	se		Next Action Taken By the I <sup>2</sup> C interface Hardware
	Code ST[7:0]	the I <sup>2</sup> C interface Hardware	To /from I <sup>2</sup> CDAT	TO I <sup>2</sup> CCON				
	51[7.0]			STA	STA STO SI AA		AA	
	C0h	Read data byte in I <sup>2</sup> CDAT has been transmitted; NOT ACK has	No I <sup>2</sup> CDAT action	0	0	0	0	Switched to not addressed SLV mode; no recognition of own SLA or General call address
		been received	No I <sup>2</sup> CDAT action	0	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if I <sup>2</sup> CADR[0] is set to logic 1.
			No I <sup>2</sup> CDAT action	1	0	0	0	Switched to not addressed SLV mode; no recognition of own SLA or General call address. A START condition will be transmitted when the bus becomes free
			No I <sup>2</sup> CDAT action	1	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if I <sup>2</sup> CADR[0] is set to logic1. A START condition will be transmitted when the bus becomes free
	C8h	Last read data byte in I <sup>2</sup> CDAT has been transmitted (AA is set	No I <sup>2</sup> CDAT action	0	0	0	0	Switched to not addressed SLV mode; no recognition of own SLA or General call address
		to logic 0); ACK has been received	No I <sup>2</sup> CDAT action	0	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if I <sup>2</sup> CADR[0] is set to logic 1.
			No I <sup>2</sup> CDAT action	1	0	0	0	Switched to not addressed SLV mode; no recognition of own SLA or General call address. A START condition will be transmitted when the bus becomes free
			No I <sup>2</sup> CDAT action	1	0	0	1	Switched to not addressed SLV mode; Own SLA will be recognized; General call address will be recognized if I <sup>2</sup> CADR[0] is set to logic 1. A START condition will be transmitted when the bus becomes free

## Table 84. I<sup>2</sup>C Miscellaneous status codes

	Status of the I <sup>2</sup> C Bus and	Application firmware Response					Next Action Taken By the I <sup>2</sup> C interface Hardware
Code I <sup>2</sup> CSTA	the I <sup>2</sup> C interface Hardware	To /from I <sup>2</sup> CDAT	TO I <sup>2</sup> CCON				
I COIA			STA	STO	SI	AA	
00h	Bus error	No I <sup>2</sup> CDAT action	Х	1	0	Х	Hardware will enter the "not addressed" Slave mode
F8h	No information available	formation available No I <sup>2</sup> CDAT action					

## 8.3.2.9 I<sup>2</sup>CDAT register

I<sup>2</sup>CDAT contains a byte of I<sup>2</sup>C data to be transmitted or a byte which has just been received. The CPU can read from and write to this 8-bit SFR while it is not in the process of shifting a byte. This occurs when the I<sup>2</sup>C interface is in a defined state and the serial interrupt flag SI is set to logic 1. Data in I<sup>2</sup>CDAT remains stable as long as SI is set to logic 1. The first bit to be transmitted is the MSB (bit 7), and, after a byte has been received, the first bit of received data is located at the MSB of I<sup>2</sup>CDAT. While data is being shifted out, data on the bus is simultaneously being shifted in; I<sup>2</sup>CDAT always contains the last data byte present on the bus. Thus, in the event of lost arbitration, the transition from Master Transmitter to Slave Receiver is made with the correct data in I<sup>2</sup>CDAT.

				,				
Bit	7	6	5	4	3	2	1	0
Symbol	Symbol I <sup>2</sup> CDAT[7:0]							
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 85. I<sup>2</sup>CDAT register (SFR: address DAh) bit allocation

#### Table 86. Description of I<sup>2</sup>CDAT bits

	-	
Bit	Symbol	Description
7 to 0	I <sup>2</sup> CDAT[7:0]	<b>I2C data.</b> Eight bits to be transmitted or just received. A logic 1 in I <sup>2</sup> CDAT corresponds to a logic 1 on the I <sup>2</sup> C bus, and a logic 0 corresponds to a logic 0 on the bus. I <sup>2</sup> C data shift through I <sup>2</sup> CDAT from right to left.

I<sup>2</sup>CDAT[7:0] and the ACK flag form a 9-bit shift register which shifts in or shifts out an 8-bit byte, followed by an acknowledge bit. The ACK flag is controlled by the I<sup>2</sup>C interface hardware and cannot be accessed by the CPU. I<sup>2</sup>C data are shifted through the ACK flag into I<sup>2</sup>CDAT on the rising edges of clock pulses on P50\_SCL. When a byte has been shifted into I<sup>2</sup>CDAT, the I<sup>2</sup>C data are available in I<sup>2</sup>CDAT, and the acknowledge bit is returned by the control logic during the ninth clock pulse. I<sup>2</sup>C data are shifted out from I<sup>2</sup>CDAT via a buffer on the falling edges of clock pulses on P50\_SCL.

When the CPU writes to  $I^2$ CDAT, the buffer is loaded with the contents of  $I^2$ CDAT[7] which is the first bit to be transmitted to the SDA line. After nine serial clock pulses, the eight bits in  $I^2$ CDAT will have been transmitted to the SDA line, and the acknowledge bit will be present in ACK. Note that the eight transmitted bits are shifted back into  $I^2$ CDAT.

## 8.3.2.10 I<sup>2</sup>CADR register

The CPU can read from and write to this 8-bit SFR. I<sup>2</sup>CADR is not affected by the I<sup>2</sup>C interface hardware. The content of this register is irrelevant when the I<sup>2</sup>C interface is in a Master mode. In the Slave modes, the seven most significant bits must be loaded with the microcontroller's own Slave address, and, if the least significant bit is set to logic 1, the general call address (00h) is recognized; otherwise it is ignored.

#### Table 87. I<sup>2</sup>CADR register (SFR: address DBh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol				SA[6:0]				GC
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

## Table 88. Description of I<sup>2</sup>CADR bits

Bit	Symbol	Description
7 to 1	SA[6:0]	<b>Slave address.</b> These bits correspond to the 7-bit Slave address which will be recognized on the incoming data stream from the I <sup>2</sup> C bus. When the Slave address is detected and the interface is enabled, a serial interrupt SI will be generated to the CPU.
0	GC	<b>General call.</b> When set to logic 1, will cause the $I^2C$ logic to watch for the general call address to be transmitted on the $I^2C$ bus. If a general call address is detected and this bit is set to logic 1, SI will be set to logic 1.

## 8.3.2.11 I<sup>2</sup>C\_wu\_control register

The wake up block has to be enabled before the whole chip enters in Soft-Power-Down mode. The choice of the wake-up conditions is made within the register I<sup>2</sup>C\_wu\_control. Read and Write conditions can be set together.

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	-	i <sup>2</sup> c_wu_en_wr	i <sup>2</sup> c_wu_en_rd	i²c_wu_en
Reset	0	0	0	0	0	0	0	0
Access	R	R	R	R	R	R/W	R/W	R/W

#### Table 89. I<sup>2</sup>C\_wu\_control register (address 610Ah) bit allocation

## Table 90. Description of I<sup>2</sup>C\_wu\_control bits

Bit	Symbol	Description
7 to 3	-	Reserved.
2	i <sup>2</sup> c_wu_en_wr	When set to logic 1, the wake-up is valid for write commands
1	i <sup>2</sup> c_wu_en_rd	When set to logic 1, the wake-up is valid for read commands
0	i <sup>2</sup> c_wu_en	When set to logic 1, enable the I <sup>2</sup> C wake-up conditions. The bit i <sup>2</sup> c_wu_en of register PCR Wakeupen (see <u>Table 144 on</u> page 97) has also to be set to logic 1 to enable the corresponding PN532 wake-up.

## 8.3.3 FIFO manager

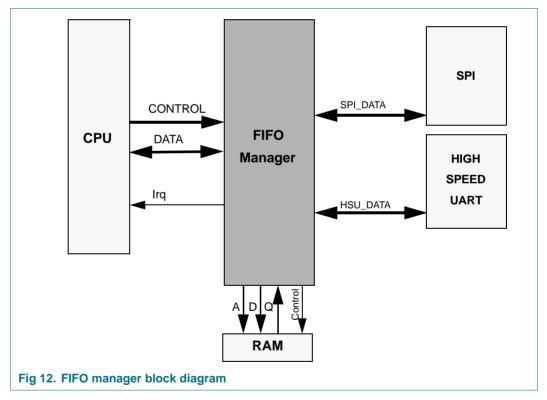
This block is designed to manage a RAM as a FIFO in order to optimize the data exchange between the CPU and the HOST.

#### 8.3.3.1 FIFO manager functional description

The RAM used for the FIFO is shared between the SPI and HSU interfaces. Indeed, these interfaces cannot be used simultaneously. The selection of the interface used is done by firmware. The FIFO manager block is the common part between the SPI and the HSU interfaces. It consists of a Data register, a Status register and also some registers to define the characteristics of the FIFO. These registers are addressed by the CPU as SFRs.

The RAM used as a FIFO is divided into two part: a receive part and a transmit part.

This block also manages the possible conflicts existing around the FIFO between the CPU and the interfaces. Indeed, a request coming from the interface (TR\_req or RCV\_req) can be simultaneous with a request to access to the data register coming from the CPU.



9 SFR registers are needed to manage the FIFO manager.

Name	Size [bytes]	SFR Address	Description	Access
RWL	1	9Ah	FIFO Receive Waterlevel: Controls the threshold of the FIFO in reception	R/W
TWL	1	9Bh	FIFO Transmit Waterlevel: Controls the threshold of the FIFO in transmission	R/W
FIFOFS	1	9Ch	FIFO Transmit FreeSpace: Status of the number of characters which can still be loaded in the FIFO	R/W
FIFOFF	1	9Dh	FIFO Receive Fullness: Status of the number of received characters in the FIFO	R/W
SFF	1	9Eh	Global Status/Error messages	R
FIT	1	9Fh	Interrupt Source	R/W
FITEN	1	A1h	Interrupt Enable and Reset FIFO	R
FDATA	1	A2h	Data reception/transmission buffer	R/W
FSIZE	1	A3h	Control the size of the FIFO in Reception	R/W

#### Table 91. Fifo manager SFR register list

## 8.3.3.2 RWL register

This register defines the warning level of the Receive FIFO for the CPU. It implies a FIFO buffer overflow.

## Table 92. RWL register (SFR: address 9Ah) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol RWaterlevel[7:0]								
Reset	0	0	0	0	0	0	0	0
Access	R/W							

#### Table 93. Description of RWL bits

Bit	Symbol	Description
7 to 0	RWaterlevel[7:0]	Overflow threshold of the Receive FIFO to set a warning

#### 8.3.3.3 TWL register

This register defines the warning level of the Transmit FIFO for the CPU. It implies a FIFO buffer underflow.

#### Table 94. TWL register (SFR: address 9Bh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol				TWaterl	evel[7:0]			
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 95. Description of TWL bits

Bit	Symbol	Description
7 to 0	TWaterlevel[7:0]	Underflow threshold of the Transmit FIFO to set a warning

## 8.3.3.4 FIFOFS register

This register indicates the number of bytes that the CPU can still load into the FIFO until the Transmit FIFO is full.

#### Table 96. FIFOFS register (SFR: address 9Ch) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol			-	TransmitFre	espace[7:0	)]		
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 97. Description of FIFOFS register bits

Bit	Symbol	Description
7 to 0	TransmitFreespace[7:0]	Freespace into the FIFO

## 8.3.3.5 FIFOFF register

This register indicates the number of bytes already received and loaded into the Receive FIFO.

#### Table 98. FIFOFF register (SFR: address 9Dh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol		ReceiveFullness[7:0]						
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 99. Description of FIFOFF bits

Bit	Symbol	Description
7 to 0	ReceiveFullness[7:0]	Number of bytes received in the FIFO

## 8.3.3.6 SFF register

The register bits are used to allow the CPU to monitor the status of the FIFO. The primary purpose is to detect completion of data transfers.

Bit	7	6	5	4	3	2	1	0
Symbol	FIFO_EN	-	TWLL	TFF	TFE	RWLH	RFF	RFE
Reset	0	0	1	0	1	0	0	1
Access	R/W	R	R	R	R	R	R	R

#### Table 101. Description of SFF bits

Bit	Symbol	Description
7	FIFO_EN	Fifo Enable: Set to logic 1 this bit enables the FIFO manager clock (CPU_CLK).
		Set to logic 0 the clock remains low.
6	-	Reserved.
5	TWLL	<b>Transmit WaterILevelLow:</b> This bit is set to logic 1 when the number of bytes stored into the Transmit FIFO is equal or smaller than the threshold TWaterlevel.
4	TFF	<b>Transmit FIFO Full:</b> This is set to logic 1 if the transmit part of the FIFO is full. It is set to logic 0 when a transfer is completed.
3	TFE	<b>Transmit FIFO Empty:</b> This bit indicates when the transmit part of the FIFO is empty.
		It is set to logic 0 when the CPU writes a character in the data register.
2	RWLH	<b>Receive WaterLevel High</b> : This bit is set to logic 1 when the number of bytes stored into the Receive FIFO is greater or equal to the threshold RWaterlevel.
1	RFF	<b>Receive FIFO Full:</b> This bit is set to logic 1 if the receive part of the FIFO is full. It is set to logic 0 by reading the FDATA register.
0	RFE	<b>Receive FIFO Empty:</b> This bit indicates when the receive part of the FIFO is empty.
		Set to logic 1, when the Receive FIFO is empty.
		Set to logic 0, when the Receive FIFO contains at least 1 byte.

# 8.3.3.7 FIT register

The FIT register contains 6 read-write bits which are logically OR-ed to generate an interrupt going to the CPU.

Bit	7	6	5	4	3	2	1	0
Symbol	Reset	-	WCOL_ IRQ	TWLL_ IRQ	TFF_ IRQ	RWLH_ IRQ	ROVR_ IRQ	RFF_ IRQ
Reset	0	0	0	0	0	0	0	0
Access	W	R	R/W	R/W	R/W	R/W	R/W	R/W

## Table 103. Description of FIT bits

Bit	Symbol	Description
7	Reset	<b>Reset:</b> Set to logic 1, Reset defines that the bits set to logic 1 in the write command are set to logic 0 in the register.
6	-	Reserved
5	WCOL_IRQ	Write COLlision IRQ: This bit is set to logic 1 when the transmitted part of the FIFO is already full (TFF is set to logic 1) and a new character is written by the CPU in the data register.
4	TWLL_IRQ	<b>Transmit WaterILevelLow IRQ:</b> This bit is set to logic 1 when the number of bytes stored into the Transmit FIFO is equal or smaller than the threshold TWaterlevel.
3	TFF_IRQ	<b>Transmit FIFO Full IRQ:</b> This is set to logic 1 if the transmitted part of the FIFO is full.
2	RWLH_IRQ	<b>Receive WaterLevel High IRQ:</b> This bit is set to logic 1 when the number of bytes stored into the Receive FIFO is greater or equal to the threshold RWaterlevel.
1	ROVR_IRQ	<b>Read OVeRrun IRQ:</b> This bit indicates that a read overrun has occured.It occurs when the receiver part of the FIFO is full and a new data transfer is completed. Then the new received data is lost and ROVR_IRQ is set.
0	RFF_IRQ	<b>Receive FIFO Full IRQ:</b> This bit is set to logic 1 if the received part of the FIFO is full.

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## 8.3.3.8 FITEN register

The FITEN register enables or disables the interrupt requests to the CPU. It is also used to reset the content of the Receive and Transmit FIFO.

Bit	7	6	5	4	3	2	1	0
Symbol	TFLUSH	RFLUSH	EN_ WCOL_ IRQ	EN_ TWLL_ IRQ	EN_ TFF_ IRQ	EN_ RWLH_ IRQ	EN_ ROVR_ IRQ	EN_ RFF_ IRQ
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

## Table 105. Description of FITEN bits

au	e 105. Descriptio	
Bit	Symbol	Description
7	TFLUSH	When set to logic level 1, the pointer of the Transmit FIFO is reset. This bit and RFLUSH must not be set at the same time.
6	RFLUSH	When set to logic level 1, the pointer of the Receive FIFO is reset. This bit and TFLUSH must not be set at the same time but one after the other.
5	EN_WCOL_IRQ	ENable Write COLlision IRQ: When set to logic 1, the WCOL_IRQ is enabled.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 18</u> ) has also to be set to logic 1 to enable the corresponding CPU interrupt.
4	EN_TWLL_IRQ	ENable Transmit WaterILevelLow IRQ: When set to logic 1, the TWLL_IRQ is enabled.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 18</u> ) has also to be set to logic 1 to enable the corresponding CPU interrupt.
3	EN_TFF_IRQ	<b>ENable Transmit FIFO Full IRQ:</b> When set to logic level 1, the TFF_IRQ is enabled.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 18</u> ) has also to be set to logic 1 to enable the corresponding CPU interrupt.
2	EN_RWLH_IRQ	ENable Receive WaterLevel High IRQ: When set to logic 1, the RWLH_IRQ is enabled.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 18</u> ) has also to be set to logic 1 to enable the corresponding CPU interrupt.
1	EN_ROVR_IRQ	<b>ENable Read OVeRrun IRQ:</b> When set to logic 1, the ROVR_IRQ is enabled.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 18</u> ) has also to be set to logic 1 to enable the corresponding CPU interrupt.
0	EN_RFF_IRQ	ENable Receive FIFO Full IRQ: When set to logic 1, the RFF_IRQ is enabled.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 18</u> ) has also to be set to logic 1 to enable the corresponding CPU interrupt.

## 8.3.3.9 FDATA register

The FDATA register is used to provide the transmitted and received data bytes. Each data written in the data register is pushed into the Transmit FIFO. Each data read from the data register is popped from the Receive FIFO.

#### Table 106. FDATA register (SFR: address A2h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol		FDATA[7:0]						
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 107. Description of FDATA bits

Bit	Symbol	Description
7 to 0	FDATA[7:0]	Writing to FDATA writes to the transmit buffer.
		Reading from FDATA reads from the receive buffer.

#### 8.3.3.10 FSIZE register

This register defines the size of the Receive FIFO. The maximum size is 182 bytes. The free space not used by the Receive FIFO in the RAM will be allocated to Transmit FIFO.

#### Table 108. FSIZE register (SFR: address A3h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	ReceiveSize[7:0]							
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 109. Description of FSIZE bits

Bit	Symbol	Description
7 to 0	ReceiveSize[7:0]	Size of the Receive FIFO

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# 8.3.4 HIGH SPEED UART (HSU)

The High Speed UART (HSU) provides a high speed link to the host (up to 1.288 Mbit/s).

The HSU is a full duplex serial port. The serial port has a Receive-buffer: in conjunction with the FIFO manager, the reception of several bytes can be performed without strong CPU real time constraints. However, if the Receive FIFO still has not been read by the CPU, and the number of receive bytes is greater than the Receive FIFO size then the new incoming bytes will be lost.

The HSU receive and transmit data registers are both accessed by firmware in the FIFO manager FDATA register. Writing to FDATA loads the transmit register, reading from FDATA accesses the separate receive register.

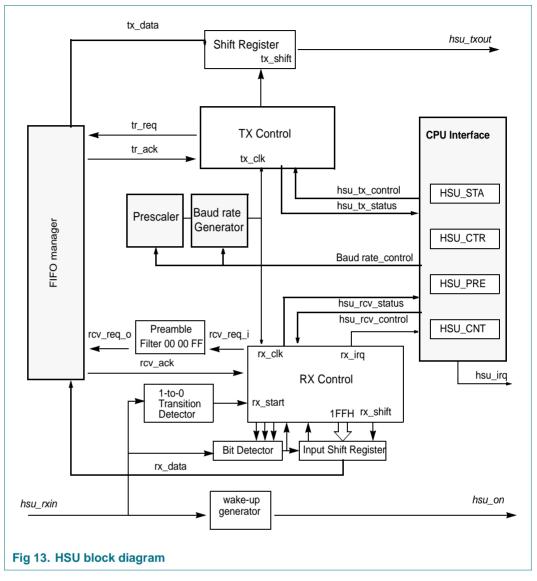
The characteristics of the UART are the following:

- Full duplex serial port
- Receive buffer to allow reception of byte while the previous bytes are stored into the FIFO manager
- 8-bit data transfers
- Programmable baud rate generator using prescaler for transmission and reception
- Based on 27.12 MHz clock frequency
- Dedicated protocol preamble filter
- Wake-up generator

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The HSU contains 4 SFRs:

#### Table 110. HSU SFR register list

	•			
Name	Size [bytes]	SFR Address	Description	Access
HSU_STA	1	ABh	HSU STAtus register	R/W
HSU_CTR	1	ACh	HSU ConTRol register	R/W
HSU_PRE	1	ADh	HSU PREscaler for baud rate generator	R/W
HSU_CNT	1	AEh	HSU CouNTer for baud rate generator	R/W

## 8.3.4.1 Mode of operation

The HSU supports only one operational mode, which has the following characteristics:

- Start bit:
  - Start bit is detected when a logic 0 is asserted on the HSU\_RX line.
- 8 data bits:
  - The data bits are sent or received LSB first.
- Stop bit:
  - During reception, the Stop bit(s) is detected when all the data bits are received and when Stop bit(s) is sampled to logic 1. The number of Stop bits is programmable. It can be 1 or 2.
  - During Transmission, after the complete data bit transmission, a variable number of Stop bit(s) is transmitted. This number is programmable from 1 to 4.

### 8.3.4.2 HSU Baud rate generator

To reach the high speed transfer rate, the HSU has it own baud rate generator. The baud rate generator comprises a prescaler and a counter. The prescaler is located before the counter. The purpose of the prescaler is to divide the frequency of the count signal to enlarge the range of the counter (at the cost of a lower resolution). The division factor of the prescaler is equal to 2 to the power HSU\_PRE[8:0] (Table 113 on page 77), resulting in division factors ranging from 1 (20) to 256 (28). The combination of these 2 blocks defines the bit duration and the bit sampling.

### 8.3.4.3 HSU preamble filter

Received characters are sent to the FIFO manager after three consecutive characters have been received: 00 00 FF. When the frame is finished, and before a new frame arrives, firmware shall write a logic 1 in the start\_frame bit of the HSU\_CTR register to re-activate the preamble filter. If firmware does not write a logic 1 then all characters of the frame are sent to the FIFO manager (including the preamble).

## 8.3.4.4 HSU wake-up generator

The wake-up generator is a 3-bit counter which counts on every rising edge of the HSU\_RX pin. When the counter reaches 5, the hsu\_on signal is set to logic 1 in order to wake up the PN532. This block is useful in Soft-Power-Down mode. The firmware shall reset this counter just before going in Soft-Power-Down by writing a logic 1 in the hsu\_wu\_en bit into the HSU\_CTR register.

## 8.3.4.5 HSU\_STA register

The SFR HSU\_STA is the status register of the HSU.

Bit	7	6	5	4	3	2	1	0
Symbol	set_bit	-	-	disable_ preamb	irq_rx_ over_en	irq_rx_fer _ en	irq_rx_ over	irq_rx_fer
Reset	0	0	0	0	0	0	0	0
Access	R/W	R	R	R/W	R/W	R/W	R/W	R/W

#### Table 111. HSU\_STA register (SFR: address ABh) bit allocation

#### Table 112. Description of HSU\_STA bits

Bit	Symbol	Description
7	set_bit	When set to logic 0 during write operation, the bits set to logic 1 in the write command are written to logic 0 in the register.
		When set to logic 1 during write operation, the bits set to logic 1 in the write command are written to logic 1 in the register.
6 to 5	-	Reserved
4	disable_preamb	<b>Preamble filter disable.</b> When set to logic 1, this bit disables the preamble filtering, it means that HSU_RX line transmit any received bytes to the FIFO manager.
3	irq_rx_over_en	<b>FIFO overflow interrupt enable.</b> When set to logic 1, this bit enables the interrupt generation when the bit irq_rx_over is set to logic 1.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 18</u> ) has also to be set to logic 1 to enable the corresponding CPU interrupt.
2	irq_rx_fer_en	Framing error interrupt enable. When set to logic 1, this bit enables the interrupt generation when the bit irq_rx_fer is set to logic 1.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 18</u> ) has also to be set to logic 1 to enable the corresponding CPU interrupt.
1	irq_rx_over	<b>Receive FIFO overflow interrupt.</b> Set to logic 1 when the FIFO manager is full (rcv_ack is set to logic 0) and when HSU shift register is ready to send another byte to the FIFO manager.
0	irq_rx_fer	Framing error interrupt. Set to logic 1 when a framing error has been detected. Framing error detection is based on Stop bit sampling.
		When Stop bit is expected at logic 1 but is sampled at logic 0, this bit is set to logic 1.

## 8.3.4.6 HSU\_CTR register

This register controls the configuration of the HSU.

### Table 113. HSU\_CTR register (SFR: address ACh) bit allocation

		•		-				
Bit	7	6	5	4	3	2	1	0
Symbol	hsu_wu_ en	start_frame	tx_stop	bit[1:0]	rx_stopbit	tx_en	rx_en	soft_reset_n
Reset	0	0	0	0	0	0	0	1
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

## Table 114. Description of HSU\_CTR bits

Bit	Symbol	Description
7	hsu_wu_en	<b>HSU wake-up enable.</b> When set to logic 1 this bit re-activates the NSS / SCL / HSU_RX rising-edge counter. When the counter is 5 then a signal hsu_on is activated. This signal is one of the possible wake-up events from Soft-Power-Down mode in the PCR block.
		The firmware shall set this bit to logic 1 just before requesting a Soft-Power-Down mode.
		The bit HSU_on_en of register PCR Wakeupen (see <u>Table 144 on page 97</u> ) has also to be set to logic 1 to enable the corresponding PN532 wake-up.
6	start_frame	<b>Enables the preamble filter for next frame.</b> When set to logic 1 this bit indicates that a new frame is coming. This re-activates the preamble filter (when enabled), meaning that the first "00 00 FF" characters will not be sent to the FIFO manager.
5:4	tx_stopbit[1:0]	<b>Defines the number of stop bit during transmission.</b> These 2 bits define the number of Stop bit(s) inserted at the end of the transmitted frame.
		The number of Stop bit(s) transmitted is equal to tx_stopbit +1.
3	rx_stopbit	<b>Defines the number of stop bit during reception.</b> This bit defines the number of Stop bit(s) inserted at the end of the received frame.
		The number of Stop bit(s) expected in reception is equal to rx_stopbit +1.
2	tx_en	<b>Enables the transmission of HSU.</b> When set to logic 1 this bit enables the transmission of characters.
		When set to logic 0, the transmission is disabled only after the completion of the current transmission.
1	rx_en	<b>Enables the reception of the HSU.</b> When set to logic 1 this bit enables the reception of characters.
		When set to logic 0, the reception is disabled only after the completion of the current reception.
0	soft_reset_n	<b>HSU Reset.</b> When set to logic 0, this bit disables the clock of the HSU_RX control, HSU_TX control and baud rate generator modules.

## 8.3.4.7 HSU\_PRE register

This register is used to configure the baud rate generator prescaler. The prescaler enlarges the range of the counter (at the cost of a lower resolution). The division factor of the prescaler ranges from 1 (20) to 256 (28).

#### Table 115. HSU\_PRE register (SFR: address ADh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol		hsu_prescaler[7:0]						
Reset	0	0	0	1	1	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 116. Description of HSU\_PRE bits

Bit	Symbol	Description
7 to 0	hsu_prescaler[7:0]	In conjunction with HSU_CNT, defines the HSU baud rate.
		Baud rate = f <sub>clk</sub> / ((hsu_prescaler +1) * hsu_counter)

### 8.3.4.8 HSU\_CNT register

This register is used to configure the baud rate generator counter.

	H30_CN	register	(SFR. auu	less ALII)	bit anocati			
Bit	7	6	5	4	3	2	1	0
Symbol		hsu_counter[7:0]						
Reset	0	1	1	1	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

## Table 117. HSU\_CNT register (SFR: address AEh) bit allocation

#### Table 118. Description of HSU\_CNT bits

Bit	Symbol	Description
7 to 0	hsu_counter[7:0]	In conjunction with HSU_PRE, defines the HSU baud rate.
		Baud rate = f <sub>clk</sub> / ((hsu_prescaler +1) * hsu_counter)

Here is a table of recommendation for some data rates:

#### Table 119. Recommendation for HSU data rates

Targeted data rate	HSU_CNT value	HSU_PRE value	Real HSU freq	Min recommended Host HSU freq	Max recommended Host HSU freq
9 600	0x71	0x18	9 516	9 326	9 706
19 200	0x9D	0x08	19 193	18 810	19 576
38 400	0x65	0x06	38 359	37 592	39 126
57 600	0x9D	0x02	57 579	56 428	58 730
115 200	0xEB	0x00	115 404	113 096	117 712
230 400	0x76	0x00	229 831	225 234	234 427
460 800	0x3B	0x00	459 661	450 467	468 854
921 600	0x1D	0x00	935 172	916 468	953 875
1 288 000	0x15	0x00	1 291 429	1 265 600	1 317 257

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# 8.3.5 Serial Parallel Interface (SPI)

The SPI has the following features:

- Compliant with Motorola de-facto Serial Peripheral Interface (SPI) standard
- Synchronous, Serial, Half-Duplex communication, 5 MHz max
- Slave configuration
- 8 bits bus interface

Through the SPI interface, the host can either access the FIFO manager (acting as data buffer) or the SPI status register. This selection is made through the hereafter described protocol.

The SPI interface is managed by 2 SFRs.

#### Table 120. SPI SFR register list

Name	Size [bytes]	SFR address	Description	R/W
SPIcontrol	1	A9h	SPI control bits	R/W
SPIstatus	1	AAh	SPI Status/Error bits	R

## 8.3.5.1 Shift register pointer

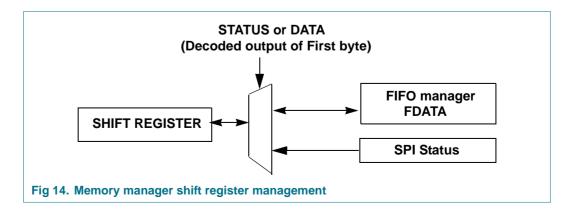
Table 121 CDI energian

A shift register is used to address the SPI interface. The value loaded in this register is either the first byte of the FIFO manager or the SPI status register.

The first byte received from the host will contain the address of the register to access (SPI status or FIFO manager FDATA) and also whether it is a SPI write or read. This character is managed by hardware.

The bits used to define these operations are the 2 LSBs of the first byte.

Table 121	. SPI operati	on
Bit 1	Bit 0	Operation
0	0	No effect
0	1	FIFO manager write access
1	0	SPI Status register read access
1	1	FIFO manager read access



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### 8.3.5.2 Protocol

Once the FIFO is full enough (see FIFO manager thresholds in <u>Table 91 on page 67</u>), the CPU sets bit READY in the SPI Status register to logic 1. Polling the SPI Status register, the host is informed of the READY flag and can start the data transfer.

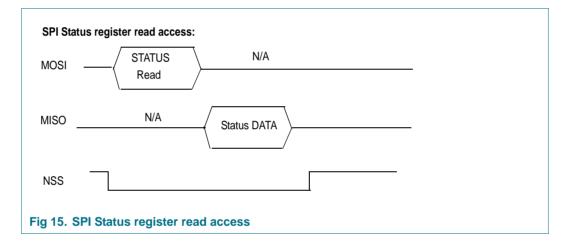
The protocol used is based on:

- ADDRESS / DATA protocol for status data exchanges
- ADDRESS / DATA / DATA / DATA... for data transfers

An exchange starts on the falling edge of NSS and follows the diagram described below.

## 8.3.5.3 SPI status register read

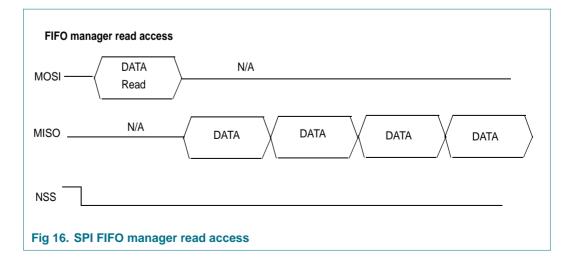
There is in that case no read request going to the FIFO manager. The content of the status register is loaded in the SPI shift register.



### 8.3.5.4 FIFO manager read access

Bytes are loaded from the FIFO manager into the SPI shift register and sent back to the host.

Remark: for proper operation, the firmware should write an additional byte in the FIFO manager (FDATA). This byte will not be transmitted.

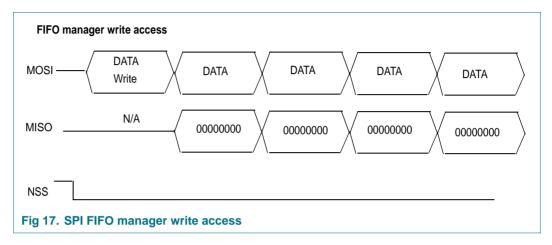


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## 8.3.5.5 FIFO manager write access

MISO is maintained at logic 0. Once a byte is received, a write request is sent to the FIFO manager and the byte is loaded from SPI shift register into Receive FIFO of the FIFO manager.



### 8.3.5.6 SPIcontrol register

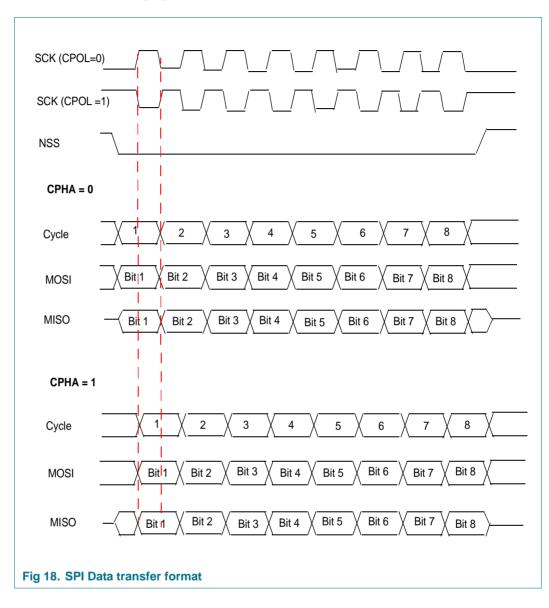
SPIcontrol register contains programmable bits used to control the function of the SPI block. This register has to be set prior to any data transfer.

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	Enable	-	CPHA	CPOL	IE1	IE0
Reset	0	0	0	0	0	0	0	0
Access	R	R	R/W	R	R/W	R/W	R/W	R/W

### Table 123. Description of SPIcontrol bits

Bit	Symbol	Description
7 to 6	-	Reserved.
5	Enable	<b>SPI enable:</b> When set to logic 1, enables the SPI interface assuming that selif[1:0] are set to 01b.
4	-	Reserved.
3	CPHA	<b>Clock PHAse:</b> This bit controls the relationship between the data and the clock on SPI transfers.
		When set to logic 0: Data is always sampled on the first clock edge of SCK.
		When set to logic 1: Data is always sampled on the second clock edge of SCK.
2	CPOL	<b>Clock POLarity:</b> This bit controls the polarity of SCK clock. When set to logic 1, SCK starts from logic 0 else starts from logic 1.
1	IE1	<b>Interrupt Enable 1:</b> When set to logic 1, the hardware interrupt generated by TR_FE in SPIstatus register is enabled.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 18</u> ) has also to be set to logic 1 to enable the corresponding CPU interrupt.
0	IE0	<b>Interrupt Enable 0:</b> When set to logic 1, the hardware interrupt generated by RCV_OVR in SPIstatus register is enabled.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 18</u> ) has also to be set to logic 1 to enable the corresponding CPU interrupt.

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**Remark:** The following figure explains how bits CPOL and CPHA can be used.

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## 8.3.5.7 SPIstatus register

The SPIstatus register is byte addressable. It contains bits which are used to monitor the status of the SPI interface, including normal functions, and exception conditions. The primary purpose of this register is to detect completion of a data transfer. The remaining bits in this register are exception condition indicators.

#### Table 124. SPIstatus register (SFR: address AAh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	TR_FE	RCV_OVR	-	READY
Reset	0	0	0	0	0	0	0	0
Access	R	R	R	R	R/W	R/W	R/W	R/W

#### Table 125. Description of SPIstatus bits

Bit	Symbol	Description
7 to 4	-	Reserved.
3	TR_FE	<b>Transmit FIFO Empty:</b> Set to logic 1 when the host attempts to read a new byte and FIFO manager is empty. An interrupt can be generated if enabled (see IE1 bit in register SPIcontrol). It is set to logic 0 by firmware.
2	RCV_OVR	<b>Receive Overrun:</b> Set to logic 1 when the host attempts to write a new byte and FIFO manager is full, or has not yet processed the previous byte. An interrupt can be generated if enabled (see IE0 bit in register SPIcontrol) It is set to logic 0 by firmware.
1	-	Reserved. This bit must be set to logic 0.
0	READY	Ready flag. The firmware set READY to logic 1 to inform the host when PN532 is ready to send data.

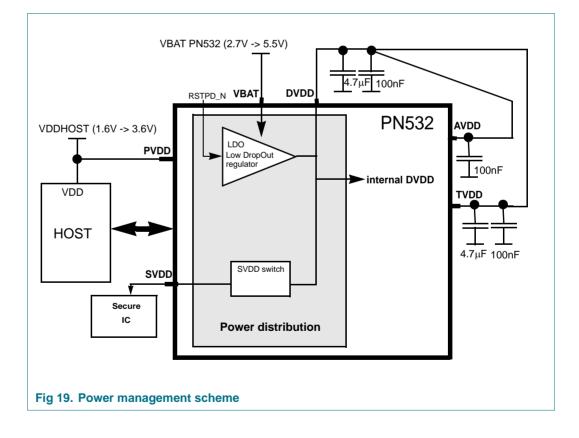
# 8.4 Power management

Figure 19 "Power management scheme" depicts the internal and external power distribution management. Power is supplied to the PN532 via pins VBAT and PVDD. VBAT is driven by the battery and is used to supply the all blocks excluding the host interface. PVDD is connected to the host's power supply and powers the PN532's host interface. No specific sequencing is required between the two supply rails: VBAT can be present without PVDD and vice versa.

An internal low drop-out (LDO) voltage regulator generates DVDD and SVDD, which are used to supply the internal digital logic and the secure IC respectively. DVDD is also routed externally to supply AVDD (analog power) and TVDD (transmit power). DVDD, AVDD and TVDD must be separately decoupled.

When another host interface than SPI is used, the PN532 can be used with reduced functionalities; all functionalities, except those related to the PVDD supplied pins (like host interfaces) when:

PVDD < 0.4V 5.5V > VBAT > 2.7V 3.6V > RSTPD\_N > VBAT \* 0.65

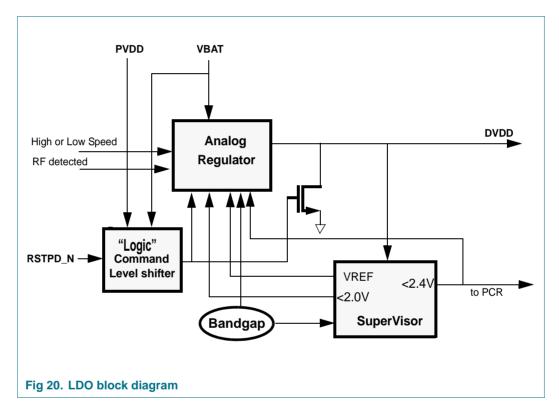


# 8.4.1 Low drop-out voltage regulator

## 8.4.1.1 LDO block diagram

The regulator is used to reduce the VBAT voltage to the typical voltage rating of the PN532. It acts as a 3.0 V linear regulator with resistive feed-back, as long as the VBAT voltage is above 3.4 V. It is designed to cope with a maximum fluctuation of 400 mV on the VBAT line (due to voltage bursts exhibited by the battery).

If VBAT falls below 3.4 V, the output of the regulator tracks VBAT with a variable delta. It continues to reject any noise on the VBAT line via the use of an internal band-gap reference.

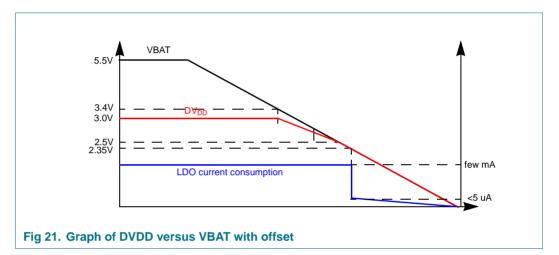


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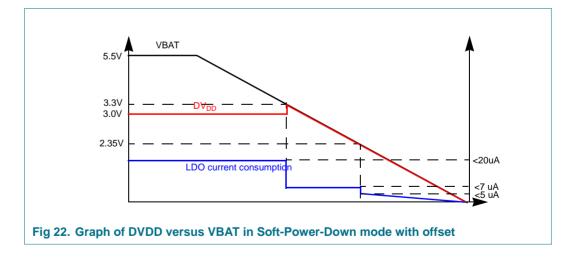
## 8.4.1.2 LDO with offset

The LDO generates DVDD. When RSTPD\_N is high, and PVDD is above 1.6 V, this voltage is defined by:

- VBAT > 3.4V: DVDD is fixed at 3V and bursts on VBAT up to 400 mV are suppressed.
- 3.4V > VBAT > 2.5V: DVDD follows VBAT with an offset, which decreases with VBAT from 400mV at 3.4V to 0mV at 2.5V.
- 2.5V > VBAT > 2.35V: DVDD=VBAT.
- 2.35V > VBAT=DVDD and the PN532 is in reset.



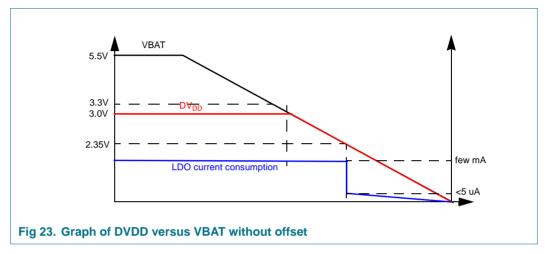
When the PN532 is in Soft-Power-Down mode, bursts rejection is no longer present and the behavior then becomes:



## 8.4.1.3 LDO without offset

The LDO generates DVDD but any voltage fluctuation on VBAT is not compensated for. When RSTPD\_N is high and PVDD is above 1.6 V, this voltage is defined by:

- VBAT > 3.0V: DVDD = 3 V.
- 3.0V > VBAT > 2.35V: DVDD = VBAT.
- 2.35V > VBAT=DVDD and the PN532 is in reset.



When in Soft-Power-Down mode, the behavior is the same as that with offset. See Figure 22 on page 86.

### 8.4.1.4 LDO overcurrent detection

The LDO integrates an overcurrent detector. When the current on VBAT exceeds a programmable threshold, an error bit is set. See <u>Table 126 on page 88</u>. If IE1\_0 is set to logic 1 (see <u>Table 13 on page 18</u>), an 80C51 interrupt will be asserted when an overcurrent is detected.

## 8.4.1.5 LDO register

#### Table 126. LDO register- (address 6109h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	overcurrent_ status		urrent[1:0] -	enoffset	soft_ highspeedreg	control_ highspeedreg
Reset	0	0	0	0	0	1	0	0
Access	R	R	R	R/W	R/W	R/W	R/W	R/W

#### Table 127. Description of LDO bits

Bit	Symbol	Description
7 to 6	-	Reserved
5	overcurrent_status	Set to logic 1 by PN532 when overcurrent is detected.
		The bit IE1_0 of register IE1 (see <u>Table 13 on page 18</u> ) has also to be set to logic 1 to enable the corresponding CPU interrupt.
4 to 3	sel_overcurrent[1:0]	Select overcurrent threshold.
		00: 300 mA
		01: 210 mA
		10: 180 mA
		11: 150 mA
2	enoffset	Enable of the LDO offset. When set to logic 1, offset is present.
1	soft_highspeedreg	<b>Control the LDO regulation speed.</b> When set to logic 0, the bandwidth of LDO is reduced to filter bursts on VBAT.
		When set to logic 1, the bandwidth is increased to establish DVDD supply quickly.
0	control_highspeedreg	<b>Select the control source of the LDO regulation speed.</b> When set to logic 1, LDO bandwidth controlled by soft_highspeedreg.
		When set to logic 0, LDO bandwidth controlled by output of RF level detector. When RF is detected, bandwidth is reduced.

# 8.4.2 SVDD switch

The SVDD switch is used to control power to the secure IC. The switch is controlled by register Control\_switch\_rng (address 6106h). The switch is enabled with bit sic\_switch\_en. When disabled, the SVDD pin is tied to ground. A current limiter is incorporated into the switch. Current consumption exceeding 40 mA triggers the limiter and the status bit sic\_switch\_overload is set.

Register Control\_switch\_rng also controls the random generator within the Contactless Interface Unit (CIU).

Bit	7	6	5	4	3	2	1	0
Symbol	-	hide_svdd_ sig	sic_switch_ overload	sic_switch_ en	-	cpu_need_ rng	random_ dataready	-
Reset	0	1	0	0	0	0	0	1
Access	R	R/W	R	R/W	R	R/W	R/W	R

#### Table 128. Control\_switch\_rng register (address 6106h) bit allocation

#### Table 129. Description of Control\_switch\_rng bits

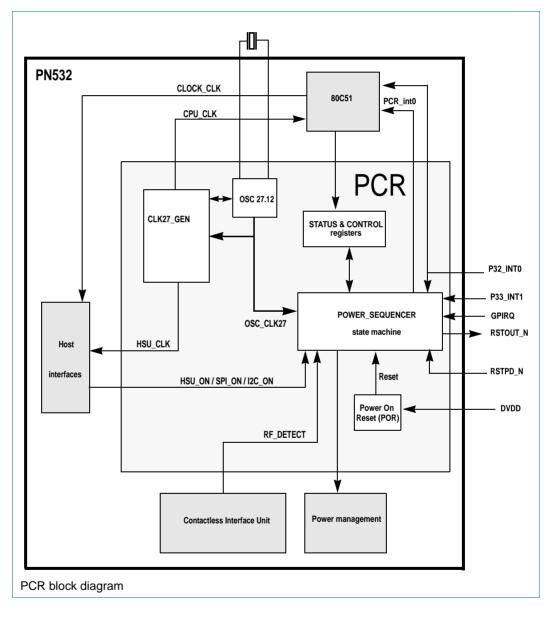
Bit	Symbol	Description
7	-	Reserved.
6	hide_svdd_sig	<b>Configures internal state of input signals SIGIN and P34 when idle.</b> This bit can be used to avoid spikes on SIGIN and P34 when the SVDD switch is enabled or disabled.
		When set to logic 0, internal state of SIGIN and P34 are driven by pads SIGIN and P34 respectively.
		When set to logic 1, internal state of SIGIN is set to logic 0 and internal state of P34 is set to logic 1.
5	sic_switch_overload	<b>Indicates state of SVDD switch current limiter.</b> When set to logic 0, indicates that current consumption through SVDD switch does not exceed limit (40mA).
		When set to logic 1, the SVDD switch current limiter is activated.
4	sic_switch_en	<b>Enables or disables power to SVDD switch.</b> When set to logic 0, SVDD switch is disabled and SVDD output is tied to the ground.
		When set to logic 1, the SVDD switch is enabled and the SVDD output delivers power to secure IC and internal pads (SIGIN, SIGOUT and P34).
3	-	Reserved.
2	cpu_need_rng	Forces random number generator into running mode. When set to logic 0, random number generator is under control of Contactless Interface Unit.
		When set to logic 1, random number generator is forced to run.
1	random_dataready	Indicates availability of random number. When set to logic 1, a new random number is available.
		Automatically set to logic 0 when register data_rng (address 6105h) is read.
0	-	Reserved.

# 8.5 Power clock and reset controller

The PCR controller is responsible for the clock generation, power management and reset mechanism within the PN532.

## 8.5.1 PCR block diagram

The block diagram shows the relationship between the PCR, other embedded blocks and external signals.



#### Table 130. PN532 clock source characteristics

Clock name	Frequency MHz	Tolerance	Clock source	Comments
OSC_CLK27	27.12	$\pm$ 14 kHz	OSC 27.12	Output of OSC 27
CPU_CLK	27.12/13.56/6.78	$\pm$ 500 ppm	OSC 27.12	Default is 6.78 MHz
HSU_CLK	27.12	$\pm$ 14 kHz	OSC 27.12	

PN532\_C1

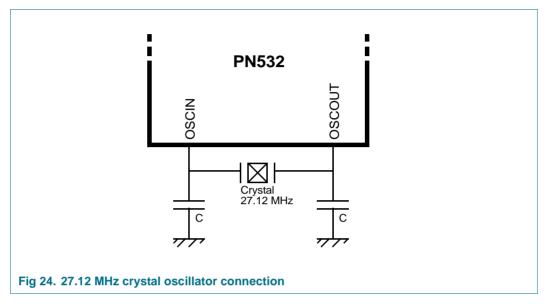
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# 8.5.2 27.12 MHz crystal oscillator

The 27.12 MHz clock applied to the PN532 is the time reference for the embedded microcontroller. Therefore stability of the clock frequency is an important factor for reliable operation. It is recommended to adopt the circuit shown in Figure 24.



## 8.5.3 Reset modes

The possible reset mechanisms are listed below:

• Supply rail variation

When DVDD falls below 2.4 V, the POR (Power-On-Reset) asserts an internal reset signal. The Power Sequencer disables all clocks. When DVDD rises above 2.4V, the POR deasserts the internal reset signal and the Power Sequencer starts the power-up sequence. Once the PN532 is out of reset, the RSTOUT\_N pin is driven high.

Glitch on DVDD

When DVDD falls below 2.35 V for more than 1 ms, the POR asserts an internal reset signal. The power sequencer starts the Power-down sequence. The PN532 goes into reset and the RSTOUT\_N signal is driven low.

• Hard Power Down mode (HPD)

When RSTPD\_N is set to logic 0, the PN532 goes into Hard Power Down (HPD) mode. The PN532 goes into reset and the RSTOUT\_N signal is driven low. The power consumption is at the minimum. DVDD is tied to ground and ports are disconnected from their supply rails.

When in Hard Power Down mode, the GPIO pins are forced in quasi bidirectional mode. Referring to Figure 7 on page 41,  $en_n = e_pu = "1"$ ,  $e_p = "0"$ .  $e_hd = "1"$  if GPIO pin value is "1" and  $e_hd = "0"$  if GPIO pin value is "0".

## 8.5.4 Soft-Power-Down mode (SPD)

In order to initiate the Soft-Power-Down mode with minimal power consumption, the firmware should:

- Configure I/Os to minimize power consumption. Be careful that for P32\_INT0, referring to <u>Section 8.2.1 "Pad configurations description" on page 40</u>, e\_hd is forced to logic 1.
- Shut down unused functions
  - Contactless Interface Unit with bit Power-down of SFR register D1h, see <u>Table 179</u> on page 146.
  - Disable the SVDD switch, see <u>Table 129 on page 89</u>
  - Power down the RF level detector if RF wake up is not enabled, see <u>Table 287 on</u> page 188.
- Enable relevant wake-up sources
- Disable unwanted interrupts
- Assert bit CPU\_PD in register PCON, see <u>Table 7 on page 16</u>

When bit CPU\_PD is set, all clocks are stopped and the LDO is put into Soft-Power-Down mode. Finally, the Power Sequencer goes into Stopped state.

### 8.5.5 Low power modes

There are 2 different low power modes.

- Hard-Power-Down mode (HPD): controlled by the pin RSTPD\_N. The PN532 goes into reset and power consumption is at a minimum, see <u>Section 8.5.3 "Reset modes"</u>.
- Soft-Power-Down mode (SPD): controlled by firmware. See <u>Section 8.5.4</u>
   <u>"Soft-Power-Down mode (SPD)</u>" to optimize the power consumption in this mode.

#### Table 131. Current consumption in low power modes

Mode	Conditions	Maximum current consumption
Hard-Power-Down	RSTPD_N is set to logic 0	2 μΑ
Soft-Power-Down with no RF detector	Sequence of <u>Section 8.5.4</u> is applied	40 μΑ
Soft-Power-Down with RF detector active	Sequence of <u>Section 8.5.4</u> is applied	45 μΑ

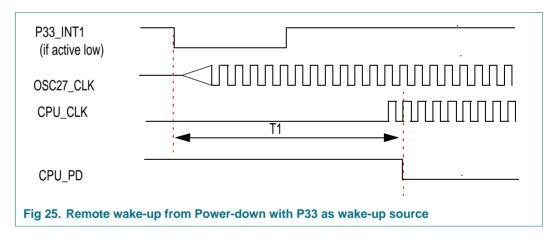
## 8.5.6 Remote wake-up from SPD

The PN532 can be woken up from a Soft-Power-Down mode when an event occurs on one of the wake up sources, which has been enabled. There are eight wake-up sources:

- P32\_INT0
- P33\_INT1
- RF field detected (RF\_DETECT)
- HSU wake-up (HSU\_ON)
- I<sup>2</sup>C wake-up (I<sup>2</sup>C\_ON)
- SPI wake-up (SPI\_ON)
- NFC\_WI counters
- GPIRQ: P34, P35, P50\_SCL, P71.

When one of these signals is asserted, if its corresponding enable bit is set (see <u>Table 144</u> on page 97), the Power Sequencer starts the wake-up sequence. The wake up event can only be serviced if the Power Sequencer is in the Stopped state, which means the PN532 is fully entered in Soft-Power-Down mode.

<u>Figure 25</u> illustrates the wake-up mechanism, using an event on P33\_INT1 as an example. CPU\_CLK is active T1 after the falling edge of P33\_INT1 and the PN532 is ready. T1 depends on the choice of crystal oscillator and its layout. For devices such as TAS-3225A, TAS-7 or KSS2F, T1 is a maximum of 2ms. Exit from the Power-down mode is signaled by CPU\_PD going low one clock cycle later.



### 8.5.7 PCR extension registers

The PCR is controlled via several registers given in Table 132:

### Table 132. PCR registers

Name	Size [bytes]	Address offset	Description	Reset	R/W
CFR	1	6200h	Clock Frequency Register	02	R/W
CER	1	6201h	Clock Enable Register	0E	R/W
ILR	1	6202h	Interrupt Level Register	40	R/W
Control	1	6203h	Control	C0	R/W
Status	1	6204h	Status	00	R
Wakeupen	1	6205h	Wake-up Enable	00	R/W

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## 8.5.8 PCR register description

### 8.5.8.1 CFR register

The Clock Frequency Register is used to select the frequency of the CPU and its associated peripherals. The clock frequency can be changed dynamically by writing to this register at any time.

#### Table 133. PCR CFR register- (address 6200h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	-	-	cpu_fr	eq[1:0]
Reset	0	0	0	0	0	0	1	0
Access	R	R	R	R	R	R	R/W	R/W

### Table 134. Description of PCR CFR bits

Bit	Symbol	Description	
7 to 2	-	Reserved	
1 to 0	cpu_frq[1:0]	Select CPU cloc	k frequency.
		cpu_frq[1:0]	CPU clock frequency
		00	27.12 MHz
		01	13.56 MHz
		10	6.78 MHz
		11	27.12 MHz

### 8.5.8.2 CER register

The Clock Enable Register is used to enable or disable the clock of the HSU (frequency is fixed at 27.12 MHz). The clock can be switched on or off at any time.

#### Table 135. PCR CER register (address 6201h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	hsu_enable	-	-	-
Reset	0	0	0	0	1	1	1	0
Access	R	R	R	R	R/W	R	R	R

### Table 136. Description of PCR CER bits

Bit	Symbol	Description					
7 to 4	-	Reserved.					
3	hsu_enable	Enable HSU clock. When 1, HSU is enabled. When 0, HSU is disabled.					
2 to 0	-	Reserved.					

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# 8.5.8.3 ILR register

The Interrupt Level Register is used to program the level of the external interrupts. Firmware can write to this register at any time.

Table 137.	PCR ILR re	egister (a	ddress 6	6202h) bit	allocation
------------	------------	------------	----------	------------	------------

Bit	7	6	5	4	3	2	1	0
Symbol	-	porpulse_ latched	-	enable_pdselif	-	gpirq_level	int1_level	int0_level
Reset	0	1	0	0	0	0	0	0
Access	R	R/W	R	R/W	R	R/W	R/W	R/W

#### Table 138. Description of PCR ILR bits

Bit	Symbol	Description							
7	-	Reserved	Reserved						
6	porpulse_latched	indicates that the	Indicates that a reset has been generated. When set to logic 1, indicates that the system has been reset. The firmware can write a "0" during the firmware reset sequence.						
5	-	Reserved							
4	enable_pdselif		Indicates that a reset has been generated. When set to logic 1, P33_INT1 directly controls state of host interface pins:						
			<ul> <li>If P33_INT1 is set to logic 1, host interface output pins are driven according to selected interface protocol</li> </ul>						
		<ul> <li>If P33_INT1 is set to logic 0, host interface output pins are set into high-impedance state</li> </ul>							
		When set to logic 0, P33_INT1 does not control host interface pins. Their state is determined by selected interface protocol.							
		enable_pdselif	P33_INT1	State of host interface pins					
		0	x	Active					
		1	0	High Impedance					
		1	1	Active					
3	-	Reserved.							
2	gpirq_level	Selects gpirq in true when gpirq i		en set to logic 1, wake-up condition is					
		When set to logic	c 0, wake-up cond	lition is true when gpirq is low.					
1	int1_level	Selects P33_INT is true when P33	•	. When set to logic 1, wake-up condition					
		When set to logic	c 0, wake-up conc	lition is true when P33_INT1 is high.					
0	int0_level	Selects P32_IN is true when P32		. When set to logic 1, wake-up condition					
		When set to logi	c 0, wake-up cond	lition is true when P32_INT0 is low.					
-									

## 8.5.8.4 PCR Control register

The Control register is used to perform a firmware reset and clear wake-up conditions in the Status register.

#### Table 139. PCR Control register (address 6203h) bit allocation

		-	-					
Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	-	-	clear_wakeup_cond	soft_reset
Reset	1	1	0	0	0	0	0	0
Access	R	R	R	R	R	R	R/W	R/W

#### Table 140. Description of PCR Control bits

Bit	Symbol	Description
7 to 2	-	Reserved.
1	clear_wakeup_cond	<b>Clears value of wakeupcond in Status register.</b> When set to logic 1, wake-up conditions stored in PCR Status register are set to logic 0. Bit is set to logic 0 automatically by hardware.
0	soft_reset	Initiates a firmware reset. When set to logic 1, system goes into firmware reset mode. Bit is set to logic 0 automatically by hardware after performing firmware reset sequence.

### 8.5.8.5 PCR Status register

The PCR Status register stores the state of the 8 wake-up events, reported within 7 flags.

**Remark:** The following status bits are not masked by the corresponding enable bit of the PCR Wakeupen register (see <u>Table 143</u>). But if not enabled, the event does not wake-up the PN532.

**Remark:** Be careful when handling the status register, not all the status events are latched. Therefore it be possible that the status register does not indicate any wake-up event when reading this register after wake-up.

**Remark:** There is no priority management. More than one wake-up event may be signalled in the register. Therefore it may not be possible to detect the source of the wake-up event by reading this register.

Bit	7	6	5	4	3	2	1	0
Symbol	i²c_wu	gpirq_wu	SPI_wu	HSU_wu	CIU_wu	-	int1_wu	int0_wu
Reset	0	0	0	0	0	0	0	0
Access	R	R	R	R	R	R	R	R

#### Table 141. PCR Status register (address 6204h) bit allocation

An event on a given wake-up condition is flagged by a logic 1 in the associated bit field.

#### Table 142. Description of PCR Status bits

Symbol	Description
i²c_wu	<b>I2C wake-up event (on its own address).</b> Set to logic 1, when PN532 woke up due to recognition of its own I <sup>2</sup> C address appearing on I <sup>2</sup> C interface <sup>[1]</sup> .
gpirq_wu	<b>gpirq wake-up event (or function of P34, P35, P50_SCL and P71 signals when enabled and level-controlled).</b> Set to logic 1, when PN532 woke up from a GIRQ event (GPIRQ at logic 0) <sup>[2]</sup> .
SPI_wu	<b>SPI wake-up event (spi_on signal).</b> Set to logic 1, when PN532 woke up from a SPI event (NSS at logic 0) <sup>[2]</sup> .
HSU_wu	HSU wake-up event (hsu_on signal). Set to logic 1, when PN532 woke up from a HSU event (5 rising edges on HSU_RX) [1].
CIU_wu	<b>Contactless wake-up event.</b> RF detected signal <sup>[2]</sup> or NFC-WI event <sup>[1]</sup> ). Set to logic 1, when PN532 woke up from a Contactless interrupt.
-	Reserved.
int1_wu	P33_INT1 wake-up event. Set to logic 1, when the system woke up from a P33_INT1 interrupt <sup>[2]</sup> .
int0_wu	<b>P32_INT0 wake-up event.</b> Set to logic 1, when the system woke up from a P32_INT0 interrupt. <sup>[2]</sup> .
	i <sup>2</sup> c_wu gpirq_wu SPI_wu HSU_wu CIU_wu - int1_wu

[1] This wake-up event is latched. The firmware must set the status byte to logic 0 after reading it (by writing a logic 1 to bit clear\_wakeup\_cond in register PCR Control)

[2] If this wake-up event does not last up to the CPU clock is available, it will not be available within the status register; it is not latched when no CPU clock is available and it directly reflects the state of the event.

### 8.5.8.6 PCR Wakeupen register

Register Wakeupen allows the selection of different wake-up events.

#### Table 143. PCR Wakeupen register (address 6205h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	i <sup>2</sup> c_wu_ en	GPIRQ_ wu_en	SPI_on_ en	HSU_on_ en	CIU_wu_ en	-	int1_en	int0_en
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R	R/W	R/W

#### Table 144. Description of PCR Wakeupen bits

Bit	Symbol	Description
7	i <sup>2</sup> c_wu_en	I2C wake-up source enable. When set to logic 1, I <sup>2</sup> C event (recognition of its own address) can wake up PN532.
		See Table 90 on page 65 to enable the corresponding event.
6	GPIRQ_wu_en	<b>General Purpose IRQ wake-up source enable.</b> When set to logic 1, a GPIRQ event can wake up PN532.
5	SPI_on_en	<b>SPI wake-up source enable.</b> When set to logic 1, a SPI event can wake up PN532.
4	HSU_on_en	<b>HSU wake-up source enable.</b> When set to logic 1, an HSU event can wake up PN532.
		See Table 114 on page 77 to enable the corresponding event.
3	CIU_wu_en	<b>Contactless Interface Unit wake-up source enable.</b> When set to logic 1, a CIU event (RF detected or NFC-WI event) can wake up PN532.

Bit	Symbol	Description			
2	-	Reserved.			
1	int1_en	<b>P33_INT1 wake-up source enable.</b> When set to logic 1, a P33_INT1 event can wake up PN532.			
0	int0_en	<b>P32_INT0 wake-up source enable.</b> When set to logic 1, a P32_INT0 event can wake up PN532.			

#### Table 144. Description of PCR Wakeupen bits ...continued

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# 8.6 Contactless Interface Unit (CIU)

The PN532 CIU is a modem for contactless communication at 13.56 MHz. It supports 6 different operating modes

- ISO/IEC 14443A/MIFARE Reader/Writer.
- FeliCa Reader/Writer.
- ISO/IEC 14443B Reader/Writer
- ISO/IEC 14443A/MIFARE Card 1K or MIFARE 4K card emulation mode
- FeliCa Card emulation
- ISO/IEC 18092, ECMA 340 NFCIP-1 Peer-to-Peer

The CIU implements a demodulator and decoder for signals from ISO/IEC 14443A/MIFARE compatible cards and transponders. The CIU handles the complete ISO/IEC 14443A framing and error detection (Parity & CRC).

The CIU supports MIFARE Classic 1K or MIFARE Classic 4K card emulation mode. The CIU supports contactless communication using MIFARE Higher transfer speeds up to 424 kbit/s in both directions.

The CIU can demodulate and decode FeliCa coded signals. The CIU digital part handles the FeliCa framing and error detection. The CIU supports contactless communication using FeliCa Higher transfer speeds up to 424 kbit/s in both directions.

The CIU supports layers 2 and 3 of the ISO/IEC 14443 B Reader/Writer communication scheme, except anticollision which must be implemented in firmware as well as upper layers.

In card emulation mode, the CIU is able to answer to a Reader/Writer command either according to the FeliCa or ISO/IEC 14443A/MIFARE card interface scheme. The CIU generates the load modulation signals, either from its transmitter or from the LOADMOD pin driving an external active circuit. A complete secure card functionality is only possible in combination with a secure IC using the NFC-WI/S<sup>2</sup>C interface.

Compliant to ECMA 340 and ISO/IEC 18092 NFCIP-1 Passive and Active communication modes, the CIU offers the possibility to communicate to another NFCIP-1 compliant device, at transfer speeds up to 424 kbit/s.The CIU handles the complete NFCIP-1 framing and error detection.

The CIU transceiver can be connected to an external antenna for Reader/Writer or Card/PICC modes, without any additional active component.

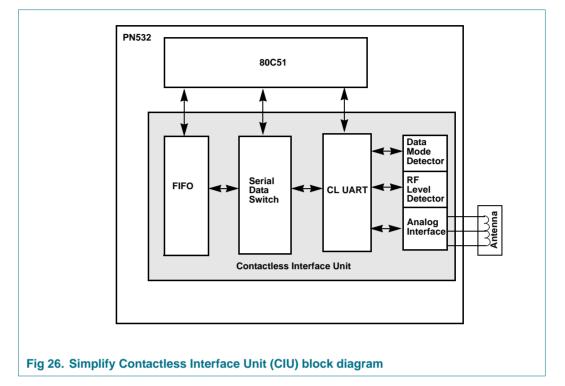
# 8.6.1 Feature list

- Frequently accessed registers placed in SFR space
- Highly integrated analog circuitry to demodulate and decode received data
- Buffered transmitter drivers to minimize external components to connect an antenna.
- Integrated RF level detector
- Integrated data mode detector
- Typical operating distance of 50 mm in ISO/IEC 14443A/MIFARE or FeliCa in Reader/Writer mode depending on the antenna size, tuning and power supply
- Typical operating distance of 50 mm in NFCIP-1 mode depending on the antenna size, tuning and power supply
- Typical operating distance in ISO/IEC 14443A/MIFARE card or FeliCa card operation mode of about 100 mm depending on the antenna size, tuning and the external field strength
- Supports MIFARE Classic 1K or MIFARE Classic 4K encryption in Reader/Writer mode
- Supports MIFARE higher data rate at 212 kbit/s and 424 kbit/s
- Supports contactless communication according to the FeliCa scheme at 212 kbit/s and 424 kbit/s
- Support of the NFC-WI/S<sup>2</sup>C interface
- 64 byte send and receive FIFO-buffer
- Programmable timer
- CRC Co-processor
- Internal self test and antenna presence detector
- 2 interrupt sources
- Adjustable parameters to optimize the transceiver performance according to the antenna characteristics

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# 8.6.2 Simplified block diagram

The Analog Interface handles the modulation and demodulation of the analog signals according to the Card emulation mode, Reader/Writer mode and NFCIP-1 mode communication scheme.

The RF level detector detects the presence of an external RF-field delivered by the antenna to the RX pin.

The data mode detector detects a ISO/IEC 14443-A MIFARE, FeliCa or NFCIP-1 mode in order to prepare the internal receiver to demodulate signals, which are sent to the PN532.

The NFC-WI/S<sup>2</sup>C interface supports communication to secure IC. It also supports digital signals for transfer speeds above 424 kbit/s.

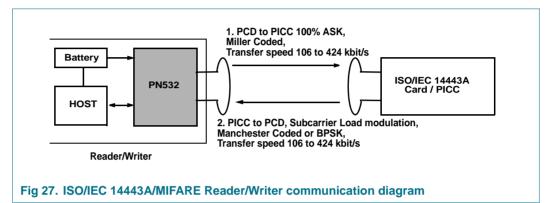
The CL UART handles the protocol requirements for the communication schemes in co-operation with the appropriate firmware. The FIFO buffer allows a convenient data transfer from the 80C51 to the CIU and vice versa.

## 8.6.3 Reader/Writer modes

All indicated modulation indices and modes in this chapter are system parameters. This means that beside the IC settings a suitable antenna tuning is required to achieve the optimal performance.

## 8.6.3.1 ISO/IEC 14443A Reader/Writer

The following diagram describes the communication on a physical level, the communication overview in the <u>Table 145</u> describes the physical parameters.



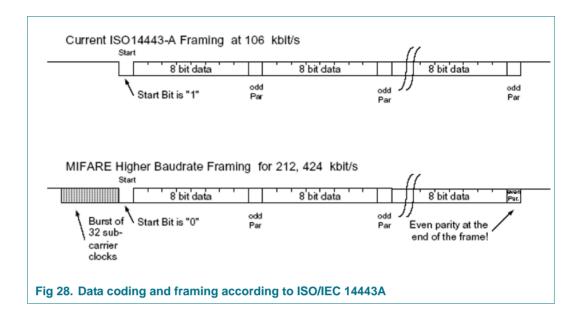
Communicatio	on scheme	ISO/IEC 14443A MIFARE	MIFARE Higher Baud Rate				
Baud rate		106 kbit/s	212 kbit/s	424 kbit/s			
Bit length		$\frac{128}{13,56MHz} \approx 9,44 \mu s$	$\frac{64}{13,56MHz} \approx 4,72 \mu s$	$\frac{32}{13,56MHz} \approx 2,36 \mu s$			
PN532 to	Modulation	100% ASK	100% ASK	100% ASK			
PICC/Card	Bit coding	Modified Miller coding	Modified Miller coding	Modified Miller coding			
PICC/Card to PN532	Modulation	Subcarrier load modulation	Subcarrier load modulation	Subcarrier load modulation			
	Subcarrier frequency	13.56 MHz/16	13.56 MHz/16	13.56 MHz/ <sub>16</sub>			
	Bit coding	Manchester coding	BPSK	BPSK			

#### Table 145. Communication overview for ISO/IEC 14443A/MIFARE Reader/Writer

The internal CRC co-processor calculates the CRC value according the data coding and framing defined in the ISO/IEC 14443A part 3, and handles parity generation internally according to the transfer speed.

With appropriate firmware, the PN532 can handle the complete ISO/IEC 14443A/MIFARE protocol.

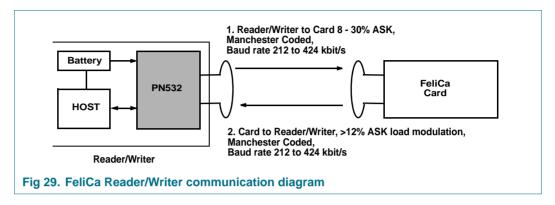
PN532/C1



PN532\_C1

## 8.6.3.2 FeliCa Reader/Writer

The following diagram describes the communication at the physical level. <u>Table 146</u> describes the physical parameters.



#### Table 146. Communication overview for FeliCa Reader/Writer

Communication scheme		FeliCa	FeliCa higher baud rate
Baud rate		212 kbit/s	424 kbit/s
Bit length		$\frac{64}{13,56MHz} \approx 4,72 \mu s$	$\frac{32}{13,56MHz} \approx 2,36\mu s$
PN532 to	Modulation	8 - 30% ASK	8 - 30% ASK
PICC/Card	Bit coding	Manchester coding	Manchester coding
PICC/Card to	Modulation	>12% ASK	>12% ASK
PN532	Bit coding	Manchester coding	Manchester coding

With appropriate firmware, the PN532 can handle the FeliCa protocol.

The FeliCa Framing and coding must comply with the following table:

#### Table 147. FeliCa Framing and Coding

Preamble					SYNC	;	LEN	n-Data	a		CRC		
00h	00h	00h	00h	00h	00h	B2h	4Dh						

To enable the FeliCa communication a 6-byte preamble (00h, 00h, 00h, 00h, 00h, 00h) and 2-byte SYNC bytes (B2h, 4Dh) are sent to synchronize the receiver.

The following LEN byte indicates the length of the sent data bytes plus the LEN byte itself. The CRC calculation is done according to the FeliCa definitions with the MSB first.

To transmit data on the RF interface, the 80C51 has to send the LEN and data bytes to the CIU. The Preamble and SYNC bytes are generated by the CIU automatically and must not be written to the FIFO. The CIU performs internally the CRC calculation and adds the result to the frame.

The starting value for the CRC Polynomial is 2 null bytes: (00h), (00h)

### Example of frame:

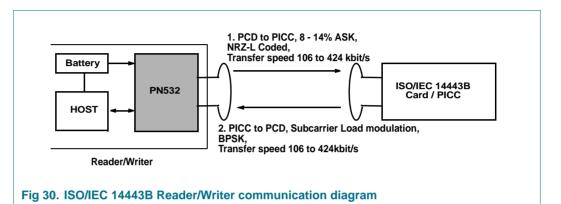
#### Table 148. FeliCa framing and coding

Prea	Preamble				SYN	C	LEN	2 Data	a Bytes	CRC		
00	00	00	00	00	00	B2	4D	03	AB	CD	90	35

## 8.6.3.3 ISO/IEC 14443B Reader/Writer

The CIU supports layers 2 and 3 of the ISO/IEC 14443 B Reader/Writer communication scheme, except anticollision which must be implemented in firmware as well as upper layers.

The following diagram describes the communication at the physical level. <u>Table 149</u> describes the physical parameters.



With appropriate firmware, the PN532 can handle the ISO/IEC 14443B protocol.

Communication	n scheme	ISO/IEC 14443B	Type B higher baud rate				
Baud rate		106 kbit/s	212 kbit/s	424 kbit/s			
Bit length		$\frac{128}{13,56MHz} \approx 9,44 \mu s$	$\frac{64}{13,56MHz} \approx 4,72\mu s$	$\frac{32}{13,56MHz} \approx 2,36 \mu s$			
PN532 to	Modulation	8 -14% ASK	8 -14% ASK	8 -14% ASK			
PICC/Card	Bit coding	NRZ-L	NRZ-L	NRZ-L			
PICC/Card to PN532	Modulation	Subcarrier load modulation	Subcarrier load modulation	Subcarrier load modulation			
	Subcarrier frequency	13.56 MHz/16	13.56 MHz/16	13.56 MHz/16			
	Bit coding	BPSK	BPSK	BPSK			

#### Table 149. Communication overview for ISO/IEC 14443B Reader/Writer

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## 8.6.4 ISO/IEC 18092, ECMA 340 NFCIP-1 operating mode

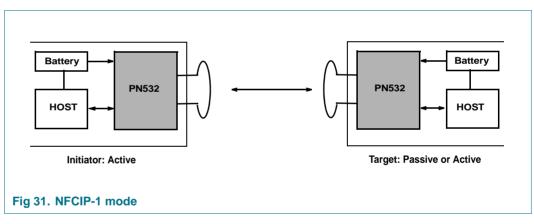
A NFCIP-1 communication takes place between 2 devices:

- Initiator: generates RF field at 13.56 MHz and starts the NFCIP-1 communication.
- Target: responds to initiator command either in a load modulation scheme in Passive Communication mode or using a self generated and self modulated RF field for Active Communication mode.

The NFCIP-1 communication differentiates between Active and Passive communication modes.

- Active Communication mode means both the initiator and the target are using their own RF field to transmit data
- Passive Communication mode means that the Target answers to an Initiator command in a load modulation scheme. The Initiator is active in terms of generating the RF field.

In order to fully support the NFCIP-1 standard the PN532 supports the Active and Passive Communications mode at the transfer speeds 106 kbit/s, 212 kbit/s and 424 kbit/s as defined in the NFCIP-1 standard

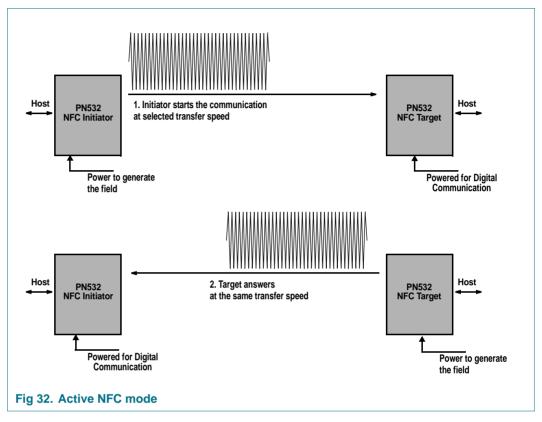


With appropriate firmware, the PN532 can handle the NFCIP-1 protocol, for all communication modes and data rates, for both Initiator and Target.

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## 8.6.4.1 ACTIVE Communication mode

Active Communication Mode means both the Initiator and the Target are using their own RF field to transmit data.



The following table gives an overview of the active communication modes:

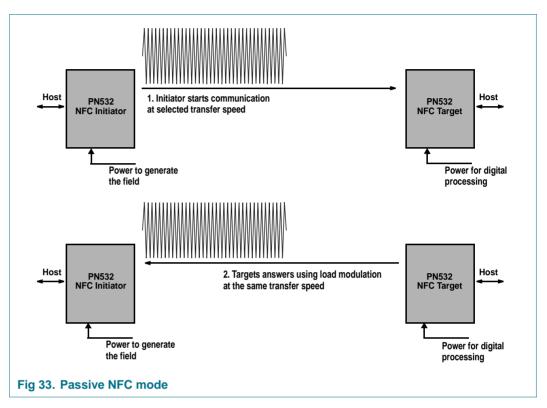
Communication	scheme	ISO/IEC 18092, ECMA 340, NFCIP-1					
Baud rate		106 kbit/s	212 kbit/s	424 kbit/s			
Bit length		$\frac{128}{13,56MHz} \approx 9,44 \mu s$	$\frac{64}{13,56MHz} \approx 4,72\mu s$	$\frac{32}{13,56MHz} \approx 2,36 \mu s$			
Initiator to Target	Modulation	100% ASK	8-30%ASK	8-30%ASK			
	Bit coding	Miller Coded	Manchester Coded	Manchester Coded			
Target to Initiator	Modulation	100% ASK	8-30%ASK	8-30%ASK			
	Bit coding	Miller Coded	Manchester Coded	Manchester Coded			

#### Table 150. Communication overview for NFC Active Communication mode

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## 8.6.4.2 PASSIVE Communication mode

Passive Communication Mode means that the target answers to an Initiator command in a load modulation scheme.



The following table gives an overview of the active communication modes:

Communication	n scheme	ISO/IEC 18092, ECMA 340, NFCIP-1					
Baud rate		106 kbit/s	212 kbit/s	424 kbit/s			
Bit length		$\frac{128}{13,56MHz} \approx 9,44 \mu s$	$\frac{64}{13,56MHz} \approx 4,72 \mu s$	$\frac{32}{13,56MHz} \approx 2,36\mu s$			
PN532 to	Modulation	100% ASK	100% ASK	100% ASK			
PICC/Card	Bit coding	Modified Miller coding	Modified Miller coding	Modified Miller coding			
PICC/Card to PN532	Modulation	Subcarrier load modulation	>12% ASK	>12% ASK			
	Subcarrier frequency	13.56 MHz/16	No subcarrier	No subcarrier			
	Bit coding	Manchester coding	Manchester coding	Manchester coding			

Table 151	Communication	overview for NE	FC Passive	Communication mo	de
	Communication		C F assive	Communication mo	ue

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# 8.6.4.3 NFCIP-1 framing and coding

The NFCIP-1 framing and coding in Active and Passive communication modes are defined in the NFCIP-1 standard: ISO/IEC 18092 or ECMA 340.

## 8.6.4.4 NFCIP-1 protocol support

The NFCIP-1 protocol is not completely described in this document. For detailed explanation of the protocol refer to the ISO/IEC 18092 / ECMA340 NFCIP-1 standard. However the datalink layer is according to the following policy:

- Transaction includes initialization, anticollision methods and data transfer. This sequence must not be interrupted by another transaction.
- · Speed should not be changed during a data transfer

In order not to disturb current infrastructure based on 13.56 MHz general rules to start NFC communication are defined in the following way:

- Per default NFCIP-1 device is in target mode, meaning its RF field is switched off.
- The RF level detector is active.
- Only if application requires the NFCIP-1 device shall switch to Initiator mode.
- Initiator shall only switch on its RF field if no external RF field is detected by RF Level detector during a time of TIDT.
- The initiator performs initialization according to the selected mode.

# 8.6.5 Card operating modes

The PN532 can be addressed like a FeliCa or ISO/IEC 14443A/MIFARE card. This means that the PN532 can generate an answer in a load modulation scheme according to the ISO/IEC 14443A/MIFARE or FeliCa interface description.

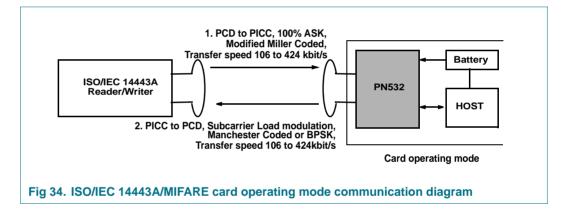
**Remark:** The PN532 does not support a secure storage of data. This has to be handled by a dedicated secure IC or a host. The secure IC is optional.

**Remark:** The PN532 can not be powered by the field in this mode and needs a power supply.

## 8.6.5.1 ISO/IEC 14443A/MIFARE card operating mode

With appropriate firmware, the PN532 can handle the ISO/IEC 14443A including the level 4, and the MIFARE protocols.

The following diagram describes the communication at the physical level. <u>Table 152</u> describes the physical parameters.



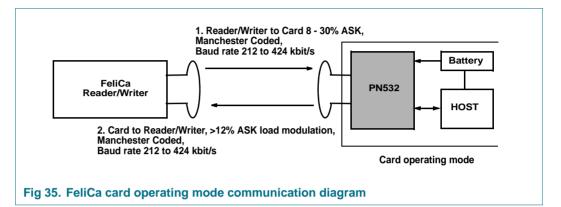
### Table 152. Communication overview for ISO/IEC 14443A/MIFARE Card operating mode

Communicatio	n scheme	ISO/IEC 14443A MIFARE	MIFARE higher baud rate			
Baud rate		106 kbit/s	212 kbit/s	424 kbit/s		
Bit length		$\frac{128}{13,56MHz} \approx 9,44 \mu s$	$\frac{64}{13,56MHz} \approx 4,72\mu s$	$\frac{32}{13,56MHz} \approx 2,36 \mu s$		
Reader/Writer	Modulation	100% ASK	100% ASK	100% ASK		
to PN532	Bit coding	Modified Miller coding	Modified Miller coding	Modified Miller coding		
PN532 to Reader/Writer	Modulation	Subcarrier load modulation	Subcarrier load modulation	Subcarrier load modulation		
	Subcarrier frequency	13.56 MHz/16	13.56 MHz/16	13.56 MHz/16		
	Bit coding	Manchester coding	BPSK	BPSK		

# 8.6.5.2 FeliCa Card operating mode

With appropriate firmware, the PN532 can handle the FeliCa protocol.

The following diagram describes the communication at the physical level. <u>Table 153</u> describes the physical parameters.



Communication s	scheme	FeliCa	FeliCa higher baud rate
Baud rate		212 kbit/s	424 kbit/s
Bit length		$\frac{64}{13,56MHz} \approx 4,72\mu s$	$\frac{32}{13,56MHz} \approx 2,36 \mu s$
Reader/Writer to	Modulation	8 - 30% ASK	8 - 30% ASK
PN532	Bit coding	Manchester coding	Manchester coding
PN532 to	Modulation	>12% ASK	>12% ASK
Reader/Writer	Bit coding	Manchester coding	Manchester coding

#### Table 153. Communication overview for FeliCa Card operating mode

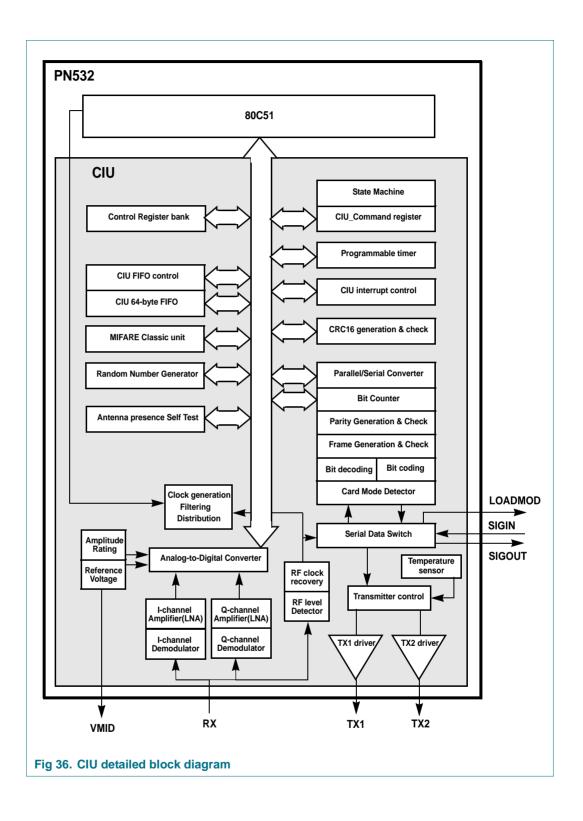
## 8.6.6 Overall CIU block diagram

The PN532 supports different contactless communication modes. The CIU supports the internal 80C51 for the different selected communication schemes such as Card Operation mode, Reader/Writer Operating mode or NFCIP-1 mode up to 424 kbit/s. The CIU generates bit- and byte-oriented framing and handles error detection according to these different contactless protocols.

Higher transfer speeds up to 3.39 Mbit/s can be handled by the digital part of the CIU. To modulate and demodulate the data an external circuit has to be connected to the communication interface pins SIGIN/SIGOUT.

Remark: The size and tuning of the antenna have an important impact on the achievable operating distance.

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# 8.6.7 Transmitter control

The signals delivered by the transmitter are on pins TX1 and pin TX2. The supply and grounds of the transmitter drivers are TVDD, TVSS1 and TVSS2.

The signals delivered are the 13.56 MHz energy carrier modulated by an envelope signal. It can be used to drive an antenna directly, using a few passive components for matching and filtering, see <u>Section 13 "Application information" on page 212</u>. The signals on TX1 and TX2 can be configured by the register CIU\_TxControl, see <u>Table 212 on page 160</u>.

The modulation index can be set by adjusting the impedance of the drivers. The impedance of the p-driver can be configured by the registers CIU\_CWGsP and CIU\_ModGsP. The impedance of the n-driver can be configured by the registers CIU\_GsNOn and CIU\_GsNOff. Furthermore, the modulation index depends on the antenna design and tuning.

Remark: It is recommended to use a modulation index in the range of 8% for the FeliCa and NFCIP-1 communication scheme at 212 and 424 kbit/s.

The registers CIU\_TxMode and CIU\_TxAuto control the data rate and framing during the transmission and the setting of the antenna driver to support the different requirements at the different modes and transfer speeds.

In the following tables, these abbreviations are used:

- RF: 13. 56 MHz clock derived from 27.12 MHz quartz divided by 2
- RF\_n: inverted 13.56 MHz clock
- GsPMos: Conductance of the transmitter PMOS
- GsNMos: Conductance of the transmitter NMOS
- CWGsP: PMOS conductance value for Continuous Wave (see <u>Table 249 on</u> page 177)
- ModGsP: refers to ModGsP[5:0], PMOS conductance value for Modulation (see <u>Table 250 on page 177</u>)
- CWGsNOn: refers to CWGsP[5:0], NMOS conductance value for Continuous Wave (see <u>Table 247 on page 176</u>)
- ModGsNOn: NMOS conductance value for Modulation when generating RF field (see <u>Table 247 on page 176</u>)
- CWGsNOff: NMOS conductance value for Continuous Wave when no RF is generated by the PN532 itself (see <u>Table 239 on page 172</u>)
- ModGsNOff: NMOS conductance value for modulation when load Modulation (see Table 239 on page 172)

Remark: If only 1 driver is switched on, the values for ModGsNOn and CWGsNOn are used for both drivers.

# Table 154. Settings for TX1

PN532	Table 1	54. Settin	igs for T	<b>K1</b>							
2	TX1 RFEn	Force 100ASK	InvTx1 RFON	InvTx1 RFOFF	Envelope	TX1	GsPMos	GsNMos	Remarks		
	0	Х	Х	0	0	0		ModGsNOff	If TX1RFEN is set to logic 0, the pin TX1 is set to logic 0 or 1 depending on		
			1 0		0		CWGsNOff	InvTx1RFOFF. The bit Force 100ASK has no effect. Envelope modulates the transconductance value.			
				1	0	1	ModGsP				
					1	1	CWGsP				
	1	0	0	Х	0	RF	ModGsP	ModGsNON	If TX1RFEN is set to logic 1, the RF phase of TX1 is depending on		
					1	RF	CWGsP	CWGsNON	InvTx1RFON. The bit Force100ASK has effect; when Envelope is set to logic 0, TX1 is pulled to ground.		
		0	1	1	1	Х	0	RF_n	ModGsP	ModGsNON	
					1	RF_n	CWGsP	CWGsNON			
All ir		1	0	Х	0	0		ModGsNON			
format					1	RF	CWGsP	CWGsNON			
All information provided		1	1	Х	0	0		ModGsNON			
ovided i					1	RF_n	CWGsP	CWGsNON			

# Table 155. Settings for TX2

TX2 RFEn	Force 100ASK	TX2CW	InVTx2 RFON		Envelope	TX2	GsPMos	GsNMos	Remarks
0	Х	0	Х	0	0 0 ModGsNOff		ModGsNOff	If Tx2RFEn is set to logic 0, the pin TX2 is forced to 0 or 1 depending	
					1	0		CWGsNOff	the InvTx2RFOFF bit. The bit ForceASK100 has no effect. The signal
				1	0	1	ModGsP		<ul> <li>Envelope modulates the transconductance value</li> </ul>
					1	1	CWGsP		
		1	Х	0	0	0		CWGsNOff	When Tx2CW bit is set, the transconductance values are always
					1	0		CWGsNOff	CWGsP or CWGsNOff
				1	0	1	CWGsP		
					1	1	CWGsP		

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## Table 155. Settings for TX2 ...continued

<u></u>	X2 RFEn	Force 100ASK	TX2CW	InVTx2 RFON	InvTx2 RFOFF	Envelope	TX2	GsPMos	GsNMos	Remarks		
1		0	0	0	Х	0	RF	ModGsP	ModGsNOn	5		
						1	RF	CWGsP	CWGsNOn	phase of TX2 is depending on InvTx2RFON. If Tx2CW bit is set to logic 1, the transconductance values are always CWGsP or		
				1	Х	0	RF_n	ModGsP	ModGsNOn	CWGsNOn, independent of Envelope.		
						1	RF_n	CWGsP	CWGsNOn			
			1	0	Х	Х	RF	CWGsP	CWGsNOn			
				1	Х	Х	RF_n	CWGsP	CWGsNOn			
		1	0	0	Х	0	0		ModGsNOn	If TX2RFEn is set to logic 1 and TX2CW to logic 0, the bit Force100ASK		
						1	RF	CWGsP	CWGsNOn	has effect; when Envelope is set to logic 0, TX2 is pulled to ground.		
Þ				1	Х	0	0		ModGsNOn			
ll infor						1	RF_n	CWGsP	CWGsNOn			
nation			1	0	Х	Х	RF	CWGsP	CWGsNOn			
provide				1	Х	Х	RF_n	CWGsP	CWGsNOn			

# 8.6.8 RF level detector

The RF level detector is integrated to fulfill NFCIP-1 protocol requirements (e.g. RF collision avoidance).

Furthermore the RF level detector can be used to wake up the PN532 and to generate an interrupt.

The sensitivity of the RF level detector is adjustable in a 4-bit range using the bits RFLevel in register CIU\_RFCfg (see <u>Table 245 on page 175</u>). The sensitivity itself depends on the antenna configuration and tuning.

Possible sensitivity levels at the RX pin are listed below:

VRx typica CIU Power-Down		CIU_RFCfg setting	CIU_RFCfg setting with additional amplifier see Remark		
1	0				
2	1.9	1111b			
1.35	1.3	1110b			
0.95	0.9	1101b			
0.6	0.57	1100b			
0.41	0.40	1011b			
0.28	0.27	1010b			
0.17	0.17	1001b			
0.12	0.12	1000b	1xxx1111b		
0.085	-	0111b <sup>[1]</sup>	1xxx1110b		
0.055	-	0110b <sup>[1]</sup>	1xxx1101b		
0.040	-	0101b <sup>[1]</sup>	1xxx1100b		
-	-	0100b <sup>[1]</sup>	1xxx1011b <sup>[1]</sup>		
-	-	0011b <sup>[1]</sup>	1xxx1010b <sup>[1]</sup>		
-	-	0010b <sup>[1]</sup>	1xxx1001b <sup>[1]</sup>		
-	-	0001b <sup>[1]</sup>	1xxx1000b <sup>[1]</sup>		
-	-	0000b[1]	1xxx0111b <sup>[1]</sup>		

#### Table 156. Setting of the RF level detector

[1] Due to noise, it is recommended not to use this setting to avoid misleading results.

To increase the sensitivity of the RF level detector an amplifier can be activated by setting the bit RFLevelAmp in register CIU\_RFCfg to logic 1 (see <u>Table 245 on page 175</u>).

**Remark:** With typical antenna, lower sensitivity levels without the additional amplifier set (below 1000b) can provoke misleading results because of intrinsic noise in the environment.

**Remark:** For the same reasons than above, it is recommended to use the RFLevelAmp only with upper RF level settings (above 1001b).

**Remark:** During the CIU Power-down mode the additional amplifier of the RF level detector is automatically switched off to ensure that the power consumption is minimal.

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# 8.6.9 Antenna presence self test

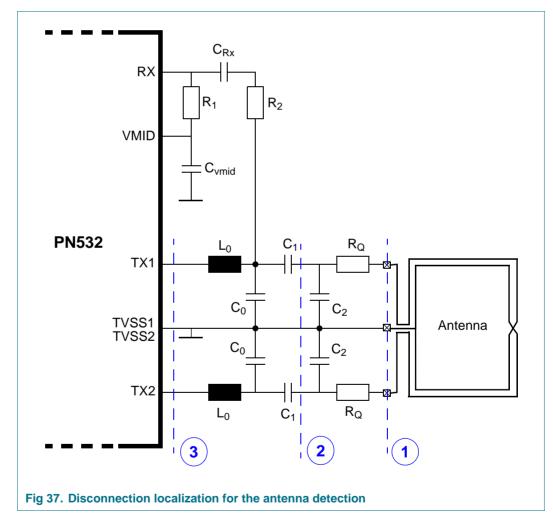
The goal of the Antenna Presence Self Test is to facilitate at assembly phase the detection of the absence of the antenna and/or antenna matching components. Such a detection is done by mean of measuring the current consumption.

Therefore the functionality is guaranteed within a restricted temperature and supply voltage range:

- VBAT voltage is above 5 V
- Ambient temperature is between 0 and 40 °C

## 8.6.9.1 Principle

The principle is explained with typical antenna tuning and matching components.



The testing operation can be managed via a dedicated register <u>Table 158 on page 118</u> and requires the transmitter to be activated. When activated by asserting bit 0, the detector will monitor the current consumption through the internal low dropout voltage regulator. Any violation to the current limits will be reported via bits 7 and 6 of the register.

Several levels of detection can be programmed through the register to offer a large panel of compatibility to different type of antennas. The high current threshold can be programmed from 40 mA to 150 mA with 15 mA steps (total current consumption of the IC). The low current threshold can be programmed from 5mA to 35 mA with 10 mA step (total current consumption of the IC).

There is no dedicated pin for the output of the detector. The result of the detection is to be read out from the antenna test register.

- Cases 1 and 2: If the antenna and/or the tuning network are not connected, the TVDD current is higher than the nominal one. The antenna detector detects this higher consumption and the andet\_up bit in andet\_control register is set to high
- Case 3: If the EMC filter is not correctly connected, the current within TVDD is lower than the nominal one. The antenna detector detects this lower consumption and the andet\_bot bit in andet\_control register is set to high.

To have this functionality working properly it is needed to have the transmitter generating some RF in the antenna.

## 8.6.9.2 Antenna presence detector register

### Table 157. andet\_control register (address 610Ch) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	andet_bot	andet_up	andet_ithl[1:0]		an	andet_en		
Reset	0	0	0	0	0	0	0	0
Access	R	R	R/W	R/W	R/W	R/W	R/W	R/W

### Table 158. Description of andet\_control bits

Bit	Symbol	Description				
7	andet_bot	A too low power consumption has been detected				
6	andet_up	too high power consumption has been detected				
5 to 4	andet_ithl[1:0	Set the low current consumption threshold to be detected				
		Define the overcurrent threshold				
		00: do not use				
		01: do not use				
		10: 25 mA				
		11: 35 mA				
3 to 1	andet_ithh[2:0]	Set the high current consumption threshold to be detected				
		000: 45 mA				
		001: 60 mA				
		010: 75 mA				
		011: 90 mA				
		100: 105 mA				
		101: 120 mA				
		110: 130 mA				
		111: 150 mA				
0	andet_en	Enable the detection of the antenna presence detector functionality.				

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# 8.6.10 Random generator

The random generator is used to generate various random number needed for the NFCIP-1 protocol, as well as for MIFARE security.

It can also be used for test purpose, by generating random data through the field.

 Table 159.
 Data\_rng register (address 6105h) bit allocation

				· · · · · ·				
Bit	7	6	5	4	3	2	1	0
Symbol		data_rng						
Reset	Х	Х	Х	Х	Х	Х	Х	Х
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

### Table 160. Description of Data\_rng bits

Bit	Symbol	Description
7 to 0	data_rng	Random number data register.

The Control\_switch\_rng register can also be used to control the behavior of the SVDD switch.

Bit	7	6	5	4	3	2	1	0
Symbol	-	hide_svdd_ sig	sic_switch_ overload	sic_switch_ en	-	cpu_need_ rng	random_ dataready	-
Reset	0	1	0	0	0	0	0	1
Access	R	R/W	R	R/W	R	R/W	R/W	R

## Table 162. Description of Control\_switch\_rng bits

Bit	Symbol	Description
7	-	Reserved.
6	hide_svdd_sig	<b>Configure the internal state of SIGIN and P34 in an idle state.</b> This bit can be used to avoid spikes on SIGIN and P34 when the SVDD switch becomes enabled or disabled.
		When set to logic 0, the internal state of SIGIN and P34 signals are driven by respectively the pads SIGIN and P34.
		When set to logic 1, the internal state of SIGIN is fixed to 0 and the internal state of P34 is fixed set to logic 1.
5	sic_switch_overload	State of the current limitation of the SVDD switch. When set to logic 0, it indicates that the current consumption into the SVDD switch does no exceed the limit.
		When set to logic 1, the current limitation of the SVDD switch is activated by the switch.
4	sic_switch_en	<b>Enable of the SVDD switch.</b> When set to logic 0, the SVDD switch is disabled and the SVDD output power is tied to the ground.
		When set to logic 1, the SVDD switch is enabled and the SVDD output deliver power to the secure IC and to the internal pads (SIGIN, SIGOUT and P34).
3	-	Reserved

Bit	Symbol	Description
2	cpu_need_rng	Force the random number generator in running mode. When set to logic 0, the random number generator is under control of the CIU.
		When set to logic 1, the random number generator is forced to run.
1	random_dataready	Indicates availability of random number. When set to logic 1, it indicates that a new random number is available.
		It is automatically set to logic 0 when the register data_rng is read.
0	-	Reserved.

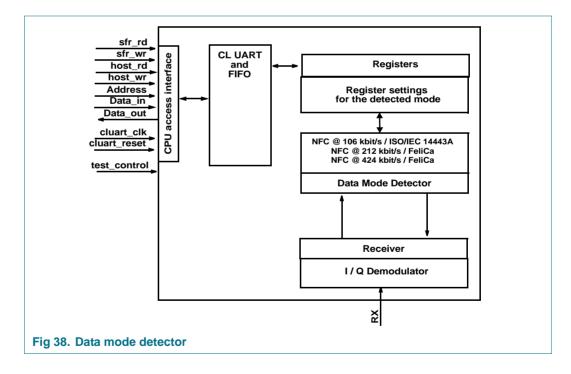
### Table 162. Description of Control\_switch\_rng bits ...continued

# 8.6.11 Data mode detector

The data mode detector is able to detect received signals according to the ISO/IEC 14443A/MIFARE, FeliCa or NFCIP-1 schemes and the standard baud rates for 106 kbit/s, 212 kbit/s and 424 kbit/s in order to prepare the internal receiver in a fast and convenient way for further data processing.

The data mode detector can only be activated by the AutoColl command (see <u>Section</u> <u>8.6.20.12 "AutoColl command" on page 137</u>). The mode detector is reset, when no external RF field is detected by the RF level detector.

The data mode detector could be switched off during the Autocoll command by setting the bit ModeDetOff in the register Mode to logic 1 (see Table 207 on page 157).



# 8.6.12 Serial data switch

Two main blocks are implemented in the CIU. A digital block comprising state machines, coder and decoder logic and an analog block with the modulator and antenna drivers, receiver and amplifier. The Serial Data Switch is the interface between these two blocks.

The Serial Data Switch can route the interfacing signals to the pins SIGIN and SIGOUT.

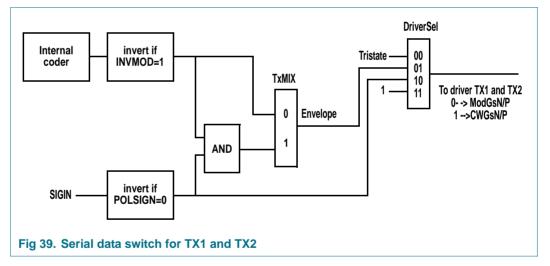
SIGOUT and SIGIN are mainly used to enable the NFC-WI/S<sup>2</sup>C interface in the secure IC to emulate card functionality with the PN532. SIGIN is capable of processing a digital signal on transfer speeds above 424 kbit/s. SIGOUT pin can also provide a digital signal that can be used with an additional external circuit to generate transfer speeds at 106 kbit/s, 212 kbit/s, 424 kbit/s and above.

Load modulation is usually performed internally by the CIU, via TX1 and TX2. However, it is possible to use LOADMOD to drive an external circuitry performing load modulation at the antenna (see optional circuitry of Figure 51 on page 212).

The Serial Data Switch is controlled by the registers CIU\_TxSel (see <u>Table 217 on</u> page 162) and CIU\_RxSel (see <u>Table 219 on page 163</u>).

## 8.6.12.1 Serial data switch for driver and loadmod

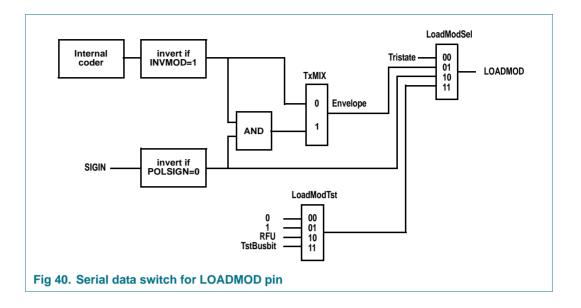
The following figure shows the serial data switch for pins TX1 and TX2.



SIGIN is in general only used for secure IC communication. If TxMix is set to logic 1 (see <u>Table 217 on page 162</u>), the driver pins are simultaneously controlled by SIGIN and the internal coder.

The following figure shows the serial data switch for the LOADMOD pin.

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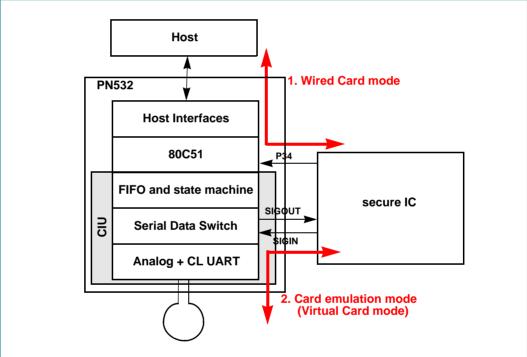
PN532\_C1

PN532/C1

# 8.6.13 NFC-WI/S<sup>2</sup>C interface support

The NFC-WI/S<sup>2</sup>C provides the possibility to directly connect a secure IC to the PN532 in order to act as a contactless smart card IC via the PN532. The interfacing signals can be routed to the pins SIGIN and SIGOUT. SIGIN can receive either a digital FeliCa or digital ISO/IEC 14443A signal sent by the secure IC. The SIGOUT pin can provide a digital signal and a clock to communicate to the secure IC. A secure IC can be a smart card IC provided by NXP Semiconductors.

The PN532 generates the supply SVDD to the secure IC. The pins SIGIN and SIGOUT are referred to this supply, as well as pin P34 / SIC\_CLK, which can be used as an extra pin for the connection to a secure IC.



The following figure outlines the supported communication flows via the PN532 to the secure core IC.

Fig 41. Communication flows supported by the NFC-WI interface

Configured in the Wired Card mode the host controller can directly communicate to the secure IC via SIGIN/SIGOUT. In this mode the PN532 generates the RF clock and performs the communication on the SIGOUT line. To enable the Wired Card mode the clock has to be derived by the internal oscillator of the PN532 (see bits sic\_clock\_sel in Table 265 on page 181.)

Configured in Card emulation mode the secure IC can act as contactless smart card IC via the PN532. In this mode the signal on the SIGOUT line is provided by the RF field of the external Reader/Writer. To enable the Virtual Card mode the clock derived by the external RF field has to be used.

The configuration of the NFC-WI/S<sup>2</sup>C interface differs for the FeliCa and MIFARE scheme as outlined in the following chapters.

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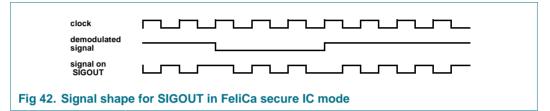
# 8.6.13.1 Signal shape for FeliCa NFC-WI/S<sup>2</sup>C interface support

The FeliCa secure IC is connected to the PN532 via the pins SIGOUT and SIGIN.

The signal at SIGOUT contains the information of the 13.56 MHz clock and the digitized demodulated signal. The clock and the demodulated signal are combined by using the logical function exclusive OR; XOR.

To ensure that this signal is free of spikes, the demodulated signal is digitally filtered first. The time delay for the digital filtering is in the range of one bit length. The demodulated signal changes only at a positive edge of the clock.

The register CIU\_TxSel (see Table 217 on page 162) controls the setting at SIGOUT



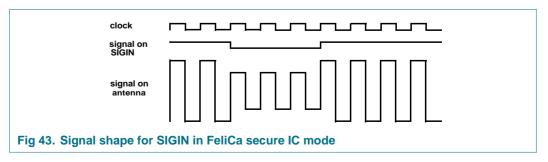
**Remark:** The PN532 differs from the ECMA 373 specification, by the fact that when in FeliCa card emulation mode, the PN532 does send preamble bytes at 212kbps on SIGOUT as soon as the PN532 detects RF field.

**Remark:** In FeliCa card emulation mode, when the PN532 mode detector is activated, the data sent on SIGOUT are clocked at the received data rate only after the SYNC bytes are received. If per default the FeliCa card emulation mode is expected at 212kpbs, the 424kbps may need specific implementation at application level: the PN532 will sent beginning of first received frame (preamble+SYNC bytes) at 212kbps.

**Remark:** To properly work in FeliCa wired card mode, the SIGIN signal generated by the FeliCa secure element must be synchronous with the received SIGOUT bit clock, and the bit RCVOFF in the register 6331h (or SFR register D1h) must be set to logic level 1. The phase relationship of the SIGIN and SIGOUT bit clocks must respect a modulo[4] 13.56MHz clock cycles.

The response from the FeliCa secure IC is transferred from SIGIN directly to the antenna driver. The modulation is done according to the register setting of the antenna drivers.

The 13.56MHz clock can be switched to P34 / SIC\_CLK (see sic\_clk\_p34\_en bit in Table 177 on page 145).



**Remark:** The signal on antenna is shown in principle only. This signal is sinusoidal. The clock for SIGIN is the same as the clock for SIGOUT.

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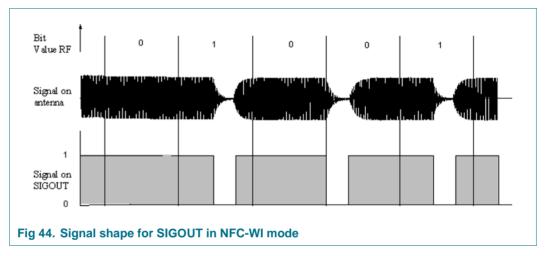
## 8.6.13.2 Signal shape for ISO/IEC14443A and MIFARE NFC-WI/S<sup>2</sup>C support

The secure IC, e.g. the SmartMX is connected to the PN532 via the pins SIGOUT, SIGIN and P34 / SIC\_CLK.

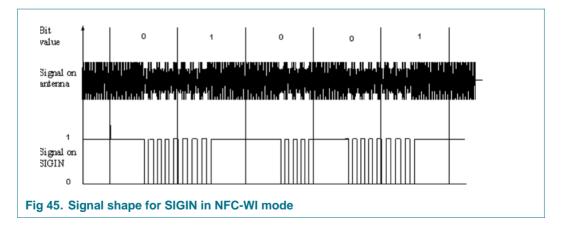
The signal at SIGOUT is a digital 13.56 MHz Miller coded signal between PVSS and SVDD. It is either derived from the external 13.56 MHz carrier signal when in Virtual Card Mode or internally generated when in Wired Card mode.

The register CIU\_TxSel controls the setting at SIGOUT.

Note: The clock settings for the Wired Card mode and the Virtual Card mode differ. Refer to the description of the bit SicClockSel in register CIU\_TestSel1.



The signal at SIGIN is a digital Manchester coded signal compliant with ISO/IEC 14443A with a subcarrier frequency of 847.5 kHz generated by the secure IC.



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# 8.6.13.3 NFC-WI/S<sup>2</sup>C initiator mode

The PN532 includes 2 counters of 127 and 31, with digital filtering, to enable activation from the secure IC (ACT\_REQ\_Si), or the command to go from data to command mode (ESC\_REQ).

## Table 163. NFC\_WI\_control register (address 610Eh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	nfc_wi_status	-	nfc_wi_en_ act_req_im	nfc_wi_en_ clk
Reset	0	0	0	0	0	0	0	0
Access	R	R	R	R	R	R/W	R/W	R/W

## Table 164. Description of NFC\_WI\_control bits

Bit	Symbol	Description
7 to 4	-	Reserved.
3	nfc_wi_status	Indicates a NFC-WI counter has reached its limit. Set to logic 1, when the counter has reached its limit. It can also be used as an interrupt for the 80C51 if the IE0_6 bit is set to logic 1 (see Table 10 on page 17).
2	-	Reserved.
1	nfc_wi_en_act_req_im	Selection of the NFC-WI counter. This bit is used to select the 31 or 127 counter.
		When set to logic 0, the 31 counter is selected.
		When set to logic 1, the 127 counter is selected.
0	nfc_wi_en_clk	Enable the NFC-WI counters on SIGIN. When set to logic 1, the counters can run and count the clock cycles within 2 and 12 MHz.

# 8.6.14 Hardware support for FeliCa and NFC polling

## 8.6.14.1 Polling sequence functionality for initiator

- 1. Timer: The CIU has a timer, which can be programmed to generate an interrupt at the end of each timeslot, or if required at the end of the last timeslot only.
- 2. The receiver can be configured to receive frames continuously. The receiver is ready to receive immediately after the last frame has been transmitted. This mode is activated by setting to logic 1 the bit RxMultiple in the register CIU\_RxMode. It has to be set to logic 0 by firmware.
- 3. The CIU adds one byte at the end of every received frame, before it is transferred into the FIFO buffer. This byte indicates whether the received frame is correct (see register Err). The first byte of each frame contains the length byte of the frame.
- 4. The length of one frame is 18 or 20 bytes (+1 byte error Info). The size of the FIFO is 64 bytes. This means 3 frames can be stored in the FIFO at the same time. If more than 3 frames are expected, the 80C51 has to read out data from the FIFO, before the FIFO is filled completely. In the case that the FIFO overflows, data is lost. (See error flag BufferOvfl).

# 8.6.14.2 Polling sequence functionality for target

- 1. The 80C51 has to configure the CIU with the correct polling response parameters for the Polling command.
- 2. To activate the automatic polling in target mode, the AutoColl Command has to be activated.
- 3. The CIU receives the polling command send out by an initiator and answers with the polling response. The timeslot is selected automatically (The timeslot itself is randomly generated, but in the range 0 to TSN, which is defined by the polling command). The CIU compares the system code, stored in byte 17 and 18 of the Config Command with the system code received with the polling command by an initiator. If the system code is equal, the CIU answers according to the configured polling response. The system code FF(hex) acts as a wildcard for the system code bytes (i.e. a target of a system code 1234(hex), 12FF(hex), FF34 (hex) or FFFF(hex)). If the system code does not match no answer is sent back by the PN532. If a valid command, which is not a Polling command, is received by the CIU, no answer is sent back and the command AutoColl is stopped. The received frame is stored in the FIFO.

## 8.6.14.3 Additional hardware support for FeliCa and NFC

Additionally to the polling sequence support for the FeliCa mode, the PN532 supports the check of the LEN-byte.

The received LEN-byte is checked by the registers CIU\_FeINFC1 and CIU\_FeINFC2:

DataLenMin in register CIU\_FelNFC1 defines the minimum length of the accepted frame length. This register is 6 bits long. Each value represents a length of 4.

DataLenMax in register CIU\_FeINFC2 defines the maximum length of the accepted frame. This register is 6 bits long. Each value represents a length of 4. If set to logic 0 this limit is switched off. If the length is not in the supposed area, the packed is not transferred to the FIFO and receiving is kept active.

## Example 1:

- DataLenMin = 4
  - The length shall be greater or equal 16.
- DataLenMax = 5
  - The length shall be smaller than 20. Valid area: 16, 17, 18, 19

### Example 2:

- DataLenMin = 9
  - The length shall be greater or equal 36.
- DataLenMax = 0
  - The length shall be smaller than 256. Valid area: 36 to 255

# 8.6.15 CRC co-processor

The CRC preset value of the CRC co-processor can be configured to 0000h, 6363h, A671h or FFFFh depending of the bits CRCPreset in the register Mode. This is only valid when using CalcCRC command (see <u>Section 8.6.20.7 "CalcCRC command" on page</u> 135)

During a communication, the preset value of the CRC coprocessor is set according to the bits CIU\_RxMode and CIU\_TxMode.

The CRC polynomial for the 16-bit CRC is fixed to  $x^{16} + x^{12} + x^5 + 1$ .

The CRC co-processor is configurable to handle the different MSB and LSB requirements for the different protocols. The bit MSBFirst in the register CIU\_Mode indicates that the data will be loaded with MSB first

The registers CRCResult-Hi and CRCResult-Lo indicate the result of the CRC calculation.

# 8.6.16 FIFO buffer

An 64\*8 bits FIFO buffer is implemented in the CIU. It buffers the input and output data stream between the 80C51 and the internal state machine of the CIU. Thus, it is possible to handle data streams with lengths of up to 64 bytes without taking timing constraints into account.

## 8.6.16.1 Accessing the FIFO buffer

The FIFO-buffer input and output data bus is connected to the register CIU\_FIFOData. Writing to this register stores one byte in the FIFO-buffer and increments the internal FIFO-buffer write-pointer. Reading from this register shows the FIFO-buffer contents stored at the FIFO-buffer read-pointer and decrements the FIFO-buffer read-pointer. The distance between the write- and read-pointer can be obtained by reading the register CIU\_FIFOLevel.

When the 80C51 starts a command, the CIU may, while the command is in progress, access the FIFO-buffer according to that command. Physically only one FIFO-buffer is implemented, which can be used in input- and output direction. Therefore the 80C51 has to take care, not to access the FIFO-buffer in an unintended way.

## 8.6.16.2 Controlling the FIFO buffer

Besides writing to and reading from the FIFO-buffer, the FIFO-buffer pointers might be reset by setting the bit FlushBuffer in the register CIU\_FIFOLevel. Consequently, the FIFOLevel[6:0] bits are set to logic 0, the bit BufferOvfl in the register CIU\_Error is set to logic 0, the actually stored bytes are not accessible anymore and the FIFO-buffer can be filled with another 64 bytes again.

## 8.6.16.3 Status information about the FIFO buffer

The 80C51may obtain the following data about the FIFO-buffers status:

- Number of bytes already stored in the FIFO-buffer: FIFOLevel[6:0] in register CIU\_FIFOLevel
- Warning, that the FIFO-buffer is quite full: HiAlert in register CIU\_Status1
- · Warning, that the FIFO-buffer is quite empty: LoAlert in register CIU\_Status1
- Indication, that bytes were written to the FIFO-buffer although it was already full: BufferOvfl in register CIU\_Error.

BufferOvfl can be set to logic 0 only by setting to logic 1 bit FlushBuffer in the register CIU\_FIFOLevel.

The CIU can generate an interrupt signal

- If LoAlertIEn in register CIU\_CommIEn is set to logic 1, it will set to logic 1 CIU\_IRQ\_0 in the register CIU\_Status1, when LoAlert in the same register changes to logic 1.
- If HiAlertIEN in register CIU\_CommIEn is set to logic 1, it will set to logic 1 CIU\_IRQ\_0 in the register CIU\_Status1, when HiAlert in the same register changes to logic 1.

The flag HiAlert is set to logic 1 if only WaterLevel[5:0] bits (as set in register CIU\_WaterLevel) or less can be stored in the FIFO-buffer. It is generated by the following equation:

 $HiAlert = (64 - FIFOLenght) \le WaterLevel$ 

The flag LoAlert is set to logic 1 if WaterLevel[5:0] bits (as set in register CIU\_WaterLevel) or less are actually stored in the FIFO-buffer. It is generated by the following equation:

$$LoAlert = FIFOLenght \leq WaterLevel$$

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# 8.6.17 CIU\_timer

A timer unit is implemented in the CIU: CIU\_timer. The 80C51 use CIU\_timer to manage timing relevant tasks for contactless communication. CIU\_timer may be used in one of the following configurations:

- Timeout-Counter
- Watch-Dog Counter
- Stop Watch
- Programmable One-Shot
- Periodical Trigger

CIU\_timer can be used to measure the time interval between two events or to indicate that a specific event occurred after a specific time. CIU\_timer can be triggered by events which will be explained in the following, but it does not itself influence any internal event (e.g. A timeout during data reception does not influence the reception process automatically). Furthermore, several timer related bits are set and these bits can be used to generate an interrupt.

CIU\_timer has a input clock of 6.78 MHz (derived from the 27.12 MHz quartz). CIU\_timer consists of 2 stages: 1 prescaler and 1 counter.

The prescaler is a 12 bits counter. The reload value for the prescaler can be defined between 0 and 4095 in register CIU\_TMode and CIU\_TPrescaler. This decimal value is called TPrescaler.

The reload value TReloadVal for the counter is defined with 16 bits in a range from 0 to 65535 in the registers CIU\_TReloadVal\_Lo and CIU\_TReloadVal\_Hi.

The current value of CIU\_timer is indicated by the registers CIU\_TCounterVal\_lo and CIU\_TCounterVal\_hi.

If the counter reaches 0 an interrupt will be generated automatically indicated by setting the TimerIRq flag in the register CommonIRq. If enabled, it will set to logic 1 CIU\_IRQ\_1 in the register CIU\_Status1. TimerIRq flag can be set to logic 1 or to logic 0 by the 80C51. Depending on the configuration, CIU\_timer will stop at 0 or restart with the value of the registers CIU\_TReloadVal\_Lo and CIU\_TReloadVal\_Hi.

Status of CIU\_timer is indicated by the bit TRunning in the register CIU\_Status1.

CIU\_timer can be manually started by TStartNow in register Control or manually stopped by TStopNow in register Control.

Furthermore CIU\_timer can be activated automatically by setting the bit TAuto in the register CIU\_TMode to fulfill dedicated protocol requirements automatically.

The time delay of a timer stage is the reload value +1.

Maximum time:

TPrescaler = 4095, TReloadVal = 65535 => 4096\*65536/6.78 MHz = 39.59 s

Example:

To indicate 100 ms it is required to count 678 clock cycles. This means the value for TPrescaler has to be set to TPrescaler = 677. The timer has now an input clock of 100 us. The timer can count up to 65535 timeslots of 100 ms.

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# 8.6.18 Interrupt request system

The CIU indicates certain events by setting interrupt bits in the register CIU\_Status1 and, in addition it will set to logic 1 CIU\_IRQ\_1 or CIU\_IRQ\_0. If this interrupt is enabled (see <u>Table 12 on page 18</u>) the 80C51 will be interrupted. This allows the implementation of efficient interrupt-driven firmware.

## 8.6.18.1 Interrupt sources

The following table shows the integrated interrupt flags, the corresponding source and the condition for its activation.

The interrupt flag TimerIRq in the register CIU\_CommIrq indicates an interrupt set by the timer unit. The setting is done when the timer decrements from logic 1 down to logic 0.

The TxIRq bit in the register CIU\_CommIrq indicates that the transmitter has finished. If the state changes from sending data to transmitting the end of frame pattern, the transmitter unit sets automatically the interrupt bit to logic 1.

The CRC coprocessor sets the flag CRCIRq in the register CIU\_DivIrq after having processed all data from the FIFO buffer. This is indicated by the flag CRCReady set to logic 1.

The RxIRq flag in the register CIU\_CommIrq indicates an interrupt when the end of the received data is detected.

The flag IdleIRq in the register CIU\_CommIrq is set to logic 1 if a command finishes and the content of the CIU\_Command register changes to idle.

The flag HiAlertIRq in the register CIU\_CommIrq is set to logic 1 if the HiAlert bit is set to logic 1, that means the Contactless FIFO buffer has reached the level indicated by the bits WaterLevel[5:0].

The flag LoAlertIRq in the register CIU\_CommIrq is set to logic 1 if the LoAlert bit is set to logic 1, that means the Contactless FIFO buffer has reached the level indicated by the bits WaterLevel[5:0].

The flag RFOnIRq in the register CIU\_DivIrq is set to logic 1, when the RF level detector detects an external RF field.

The flag RFOffIRq in the register CIU\_DivIrq is set to logic 1, when a present external RF field is switched off.

The flag ErrIRq in the register CIU\_CommIrq indicates an error detected by the CIU during sending or receiving. This is indicated by any bit set to logic 1 in register CIU\_Error.

The flag ModelRq in the register CIU\_DivIrq indicates that the data mode detector has detected the current mode.

These flags are summarized with 2 interrupt bits within the register CIU\_Status1:

- the high priority interrupt sources are summarized with CIU\_IRQ\_0.
- the low priority interrupt sources are summarized with CIU\_IRQ\_1.

See the register Table 190 on page 151.

Table 105. Tight	priority interrupt sou	
Interrupt Flag	Interrupt source	Set automatically, WHEN
TxIRq	Transmitter	a transmitted data stream ends
RxIRq	Receiver	a received data stream ends
HiAlertIRq	FIFO-buffer	the FIFO-buffer is getting full
LoAlertIRa	FIFO-buffer	the FIFO-buffer is aetting empty

### Table 165. High priority interrupt sources (CIU\_IRQ\_0)

### Table 166. Low priority interrupt sources (CIU\_IRQ\_1)

Interrupt Flag	Interrupt source	Set automatically, WHEN
TimerIRq	Timer Unit	the timer counts from 1 to 0
CRCIRq	CRC-Coprocessor	all data from the FIFO buffer have been processed
IdleIRq	CIU_Command Register	a command execution finishes
RFOnIRq	<b>RF</b> Level Detector	an external RF field is detected
RFOffIRq	RF Level Detector	a present external RF field is switched off
ErrlRq	CIU	an error is detected
ModelRq	data mode detector	the mode has been detected

### 8.6.19 CIU Power Reduction Modes

### 8.6.19.1 Hard-Power-Down

A Hard-Power-Down is enabled when RSTPD\_N is low. None of the CIU blocks are running, even the RF level detector.

### 8.6.19.2 CIU Power-down

The CIU Power-down mode is entered immediately by setting the Power-down bit in the register CIU\_Command. All CIU blocks are switched off, except the 27.12 MHz oscillator and the RF level detector.

All registers and the FIFO will keep the content during CIU Power-down.

If the bit AutoWakeUp in the register CIU\_TxAuto is set and an external RF field is detected, the CIU Power-down mode is left automatically.

After setting bit Power-down to logic 0 in the register CIU\_Command, it needs 1024 clocks cycle until the CIU Power-down mode is left indicated by the Power-down bit itself. Setting it to logic 0 does not immediately set it to logic 0. It is automatically set to logic 0 by the CIU when the CIU Power-down mode is left.

When in CIU Power-down mode and DriverSel[1:0] is no set to 00b (see <u>Table 217 on</u> page 162), to ensure a minimum impedance at the transmitter outputs, the CWGsNOn[3], CWGsNOff[3], ModGsNOn[3], ModGsNOff[3], CWGsP[5], ModGsP[5] bits are set to logic 1, but it is not readable in the registers.

## 8.6.19.3 Transmitter Power-down

The Transmitter Power-down mode switches off the internal antenna drivers to turn off the RF field by setting the bits Tx1RFEn and Tx2RFEn in the register CIU\_TxControl to logic 0. The receiver is still switched on, meaning the CIU can be accessed by a second NFC device as a NFCIP-1 target.

Note: In case the bit InitialRFOn has been set to logic 1, when the drivers were already switched on, it is needed either to set InitialRFOn to logic 0, before setting the bits Tx1RFEn and Tx2RFEn in the register CIU\_TxControl to logic 0, or to set also the bits Tx1RFAutoEn and Tx2RFAutoEn in the register CIU\_TxAuto to logic 0.

## 8.6.20 CIU command set

## 8.6.20.1 General description

The CIU behavior is determined by an internal state machine capable to perform a certain set of commands. Writing the according command code to the CIU\_Command register starts the commands.

Arguments and/or data necessary to process a command are mainly exchanged via the FIFO buffer.

## 8.6.20.2 General behavior

- Each command, that needs a data stream (or data byte stream) as input will immediately process the data it finds in the FIFO buffer. An exception to this rule is the Transceive command. Using this command the transmission is started with the StartSend bit in CIU\_BitFraming register.
- Each command that needs a certain number of arguments will start processing only when it has received the correct number of arguments via the FIFO buffer.
- The FIFO buffer is not cleared automatically at command start. Therefore, it is also possible to write the command arguments and/or the data bytes into the FIFO buffer and start the command afterwards.
- Each command may be interrupted by the 80C51 by writing a new command code into the CIU\_Command register e.g.: the Idle command.

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### 8.6.20.3 Commands overview

Command	Command code	Action
Idle	0000	No action; cancels current command execution.
Config	0001	Configures the CIU for FeliCa, MIFARE and NFCIP-1 communication.
Generate RandomID	0010	Generates 10-byte random ID number
CalcCRC	0011	Activates the CRC co-processor or perform self-test.
Transmit	0100	Transmits data from the FIFO buffer.
NoCmdChange	0111	No command change. This command can be used to modify different bits in the CIU_Command register without touching the command. e.g. Power-down bit.
Receive	1000	Activates the receiver circuitry.
SelfTest	1001	Activates the self-test. Not described in this chapter.
Transceive	1100	If bit Initiator in the register CIU_Control is set to logic 1:
		Transmits data from FIFO buffer to the antenna and activates automatically the receiver after transmission is finished.
		If bit Initiator in the register CIU_Control is set to logic 0:
		Receives data from antenna and activates automatically the transmitter after reception.
AutoColl	1101	Handles FeliCa polling (Card operating mode only) and MIFARE anticollision (Card operating mode only)
MFAuthent	1110	Performs the MIFARE Classic 1K or MIFARE Classic 4K authentication in MIFARE Reader/Writer mode only.
Soft Reset	1111	Resets the CIU.

### 8.6.20.4 Idle command

The CIU is in idle mode. This command is also used to terminate the actual command.

### 8.6.20.5 Config command

To configure the automatic MIFARE Anticollision, FeliCa Polling and NFCID3, the data used for these transactions have to be stored internally. All the following data have to be written to the FIFO in this order:

- SENS\_RES (2 bytes): in order byte0, byte1
- NFCID1 (3 Bytes): in order byte0, byte1, byte 2; the first NFCID1 byte if fixed to 08h and the check byte is calculated automatically
- SEL\_RES (1 byte)
- Polling response (2 bytes (shall be 01h, FEh)+ 6 bytes NFCID2 + 8 bytes Pad + 2 bytes system code)
- NFCID3 (1 byte)

In total 25 bytes which are transferred into an internal buffer with the Config command.

The complete NFCID3 is 10 bytes long and consist of the 3 NFCID1 bytes, the 6 NFCID2 bytes and the NFCID3 byte listed above.

To read out this configuration (after it has been loaded), the command Config with an empty FIFO buffer has to be started. In this case the 25 bytes are transferred from the internal buffer to the FIFO.

The CIU has to be configured after each power up, before using the automatic Anticollision/Polling function (AutoColl command). During a Hard-Power-Down (RSTPD\_N set to logic 0) this configuration remains unchanged.

This command terminates automatically when finished and the active command is Idle.

## 8.6.20.6 Generate RandomID command

This command generates a 10-byte random number stored in the internal 25 bytes buffer and overwrites the 10 NFCID3 bytes. This random number might be used for fast generation of all necessary ID bytes for the automatic Anticollision / Polling function.

Note: To configure the CIU, Config command has to be used first.

This command terminates automatically when finished and the active command is Idle

## 8.6.20.7 CalcCRC command

The content of the FIFO is transferred to the CRC co-processor and a CRC calculation is started. The result is stored in the CRCResult register. The CRC calculation is not limited to a dedicated number of bytes. The calculation is not stopped when the FIFO gets empty during the data stream. The next byte written to the FIFO is added to the calculation.

The preset value of the CRC is defined by the CRCPreset bits of the register CIU\_Mode, and the chosen value is loaded to the CRC co-processor when the command is started.

This command has to be terminated by firmware by writing any command to the CIU\_Command register e.g. the Idle command.

If SelfTest in register CIU\_AutoTest is set to logic 1, the CRC co-processor is in Self Test mode and performs a digital self-test. The result of the self-test is written in the FIFO.

## 8.6.20.8 Transmit command

The content of the FIFO is transmitted immediately after starting the command. Before transmitting FIFO content, all relevant register settings have to be set to transmit data in the selected mode.

This command terminates automatically when the FIFO gets empty and the active command is Idle. It can be terminated by any other command written to the CIU\_Command register.

## 8.6.20.9 NoCmdChange command

This command does not influence any ongoing command in the CIU\_Command register. It can be used to manipulate any bit except the command bits in the CIU\_Command register, e.g. the bits RcvOff or Power-down.

## 8.6.20.10 Receive command

The CIU activates the receiver path and waits for any data stream to be received. The correct settings for the expected mode have to be set before starting this command.

This command terminates automatically when the reception ends and the active command is Idle.

In case of data rates at 212 kbps or 424 kbps with NFC or FeliCa framing, the reception ends when the number of bytes indicated by the LEN byte (received) are received. If less bytes than indicated by LEN are received from the RF, the CIU will sample noise to add the missing bytes. If more bytes than indicated by LEN are received from the RF, the last bytes will be ignored. When LEN is 0, the frame is ignored, and the PN532 waits for a new frame.

In case of data rate at 106 kbps in NFC communication mode, the reception ends when the CIU detects an end of frame, except if the CIU detects more bytes than indicated by the received LEN byte. In that case, after receiving LEN bytes, a new reception restarts and the CIU timer starts (if Tauto in register CIU\_TMode <u>Table 252</u> is set to logic 1). The end of reception is then seen if a new valid frame is received or the firmware has to end the reception phase on time-out.

In all other cases, the end of the reception is detected by the end of frame.

In case no frame is received and Tauto in register CIU\_TMode <u>Table 252</u> is set to logic 1, then TimerIRQ in register CommIRQ is set to logic 1. The firmware has to end the reception phase.

Note: If the bit RxMultiple in the register CIU\_RxMode is set to logic 1, the Receive command does not terminate automatically. It has to be terminated by setting any other command in the CIU\_Command register.

## 8.6.20.11 Transceive command

This circular command repeats transmitting data from the FIFO and receiving data from the RF field continuously. If the bit Initiator in the register CIU\_Control is set to logic 1, it indicates that the first action is transmitting and after having finished transmission the receiver is activated to receive data. If the bit Initiator in the CIU\_Control register is set to logic 0, the first action is receiving and after having received a data stream, the transmitter is activated to transmit data. In the second configuration the PN532 first acts as a receiver and if a data stream is received it switches to the Transmit mode.

The end of the reception phase is detected in the same way than for the receive command and also when the RF field is cut.

The transceive command always take into account the presence or absence of the RF field. No transmission or reception can be done when no RF field.

Communication step	Initiator =1	Initiator=0
1	Send	Receive
2	Receive	Send
3	Send	Receive
4	Receive	Send

### Table 168. Transceive command scenario

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Each transmission process has to be started with setting bit StartSend in the register CIU\_BitFraming. This command has to be cleared by firmware by writing any command to the CIU\_Command register e.g. the command idle.

Note: If the bit RxMultiple in register CIU\_RxMode is set, this command will never leave the receiving state, because the receiving will not be cancelled automatically.

## 8.6.20.12 AutoColl command

This command automatically handles the MIFARE activation and the FeliCa polling in the Card Operation mode. The bit Initiator in the CIU\_Control register has to be set to logic 0 for correct operation. During this command, Mode Detector is active if not deactivated by setting the bit ModeDetOff in the CIU\_Mode register. After Mode Detector detects a mode, the mode dependent registers are set according to the received data. In case of no external RF field this command resets the internal state machine and returns to the initial state but it will not be terminated.

When the Autocoll command terminates the Transceive command gets active.

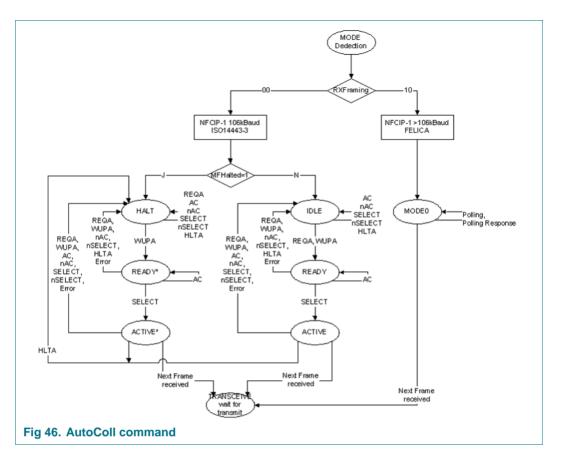
During Autocoll command:

- The CIU interrupt bits, except RfOnIRq, RfOffIRq and SIGINActIRq (see <u>Table 187 on page 149</u>), are not supported. Only the last received frame will serve the CIU interrupts.
- During ISO/IEC 14443A activation, TxCRCEn and RxCRCEn bits are defined by the AutoColl command. The changes cannot be observed at the CIU\_TxMode and CIU\_RxMode registers. When the Transceive command is active, the value of the bits is relevant.
- During Felica activation (polling), TxCRCEn and RxCRCEn bits are always relevant and are not overruled by the Autocoll command. Their value must be set to logic 1 according the FeliCa protocol.

Note: Pay attention, that the FIFO will also receive the two CRC check bytes of the last command, even if they are already checked and correct, and if the state machine (Anticollision and Select routine) has not been executed, and 106 kbit is detected.

This command can be cleared by firmware by writing any other command to the CIU\_Command register, e.g. the Idle command. Writing the same content again to the CIU\_Command register resets the state machine.

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• NFCIP-1 106 kbps passive communication mode:

The MIFARE anticollision is finished and the command changes automatically to Transceive. The FIFO contains the ATR\_REQ frame including the start byte F0h. The bit TargetActivated in the register CIU\_Status2 is set to logic 1

• NFCIP-1 212 and 424 kbps passive communication mode:

The FeliCa polling command is finished and the command has automatically changed to Transceive. The FIFO contains the ATR\_REQ frame. The bit TargetActivated in the register CIU\_Status2 is set to logic 1.

• NFCIP-1 106, 212 and 424 kbps active communication mode:

This command is changing automatically to Transceive. The FIFO contains the ATR\_REQ frame. The bit TargetActivated in the register CIU\_Status2 is set to logic 0. For 106 kbps only, the first byte in the FIFO indicates the start byte F0h and the CRC is added into the FIFO.

• ISO/IEC 14443A/MIFARE (Card Operating mode):

The MIFARE anticollision is finished and the command has automatically changed to Transceive. The FIFO contains the first command after the Select. The bit TargetActivated in the register CIU\_Status2 is set to logic 1.

• FeliCa (Card Operating mode):

The FeliCa polling command is finished and the command has automatically changed to Transceive. The FIFO contains the command after the Polling in the FeliCa protocol. The bit TargetActivated in the register CIU\_Status2 is set to logic 1.

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## 8.6.20.13 MFAuthent command

This command handles the MIFARE authentication in Reader/Writer mode to enable a secure communication to any MIFARE Clasic 1K and MIFARE Classic 4K emulation card. The following data shall be written to the FIFO before the command can be activated:

- Authentication command code (60h for key A, 61h for key B)
- Block address
- Sector key byte 0
- Sector key byte 1
- Sector key byte 2
- Sector key byte 3
- Sector key byte 4
- Sector key byte 5
- Card serial number byte 0
- Card serial number byte 1
- Card serial number byte 2
- Card serial number byte 3

In total 12 bytes shall be written to the FIFO.

Note: When the MFAuthent command is active, any FIFO access is blocked. Anyhow if there is an access to the FIFO, the bit WrErr in the register CIU\_Error is set to logic 1.

This command terminates automatically when the MIFARE Classic 1K or MIFARE Classic 4K emulation card is authenticated. The bit MFCrypto1On in the register CIU\_Status2 is set to logic 1.

This command does not terminate automatically when the card does not answer, therefore CIU timer should be initialized to automatic mode. In this case, beside the bit IdleIRq, the bit TimerIRq can be used as termination criteria. During authentication processing, the bits RxIRq and TxIRq of CIU\_CommIrq register are blocked.

The Crypto1On bit is only valid after termination of the MFAuthent command (either after processing the authentication or after writing the Idle command in the register CIU\_Command).

In case there is an error during the MIFARE authentication, the ProtocolErr bit in the CIU\_Error register is set to logic 1 and the Crypto1On bit in CIU\_Status2 register is set to logic 0.

# 8.6.20.14 SoftReset command

This command performs a reset of the CIU. The configuration data of the internal buffer remains unchanged. All registers are set to the reset values.

When SoftReset is finished, the active command switches to Idle.

# 8.6.21 CIU tests signals

## 8.6.21.1 CIU self-test

The CIU has the capability to perform a self-test. To start the self-test the following procedure has to be performed:

- 1. Perform a SoftReset.
- 2. Clear the internal buffer by writing 25 bytes of 00h and perform the Config command.
- 3. Enable the self-test by writing the value 09h to the register CIU\_AutoTest.
- 4. Write 00h to the FIFO.
- 5. Start the self-test with the CalcCRC command.
- 6. The self-test will be performed.
- 7. When the self-test is finished, the FIFO is contains the following bytes:

Correct answer for VersionReg equal to 80h: 0x00, 0xaa, 0xe3, 0x29, 0x0c, 0x10, 0x29, 0x6b 0x76, 0x8d, 0xaf, 0x4b, 0xa2, 0xda, 0x76, 0x99 0xc7, 0x5e, 0x24, 0x69, 0xd2, 0xba, 0xfa, 0xbc 0x3e, 0xda, 0x96, 0xb5, 0xf5, 0x94, 0xb0, 0x3a 0x4e, 0xc3, 0x9d, 0x94, 0x76, 0x4c, 0xea, 0x5e 0x38, 0x10, 0x8f, 0x2d, 0x21, 0x4b, 0x52, 0xbf 0xfb, 0xf4, 0x19, 0x94, 0x82, 0x5a, 0x72, 0x9d 0xba, 0x0d, 0x1f, 0x17, 0x56, 0x22, 0xb9, 0x08

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## 8.6.21.2 CIU test bus

The test bus is implemented for production test purposes. The following configuration can be used to improve the design of a system using the PN532. The test bus allows to route internal signals to output pins.

The Observe\_testbus register is used to enable this functionality.

	Table 103. Observe_testbus register (address 01041) bit anocation							
Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	-	-	-	observe_ciu
Reset	0	0	0	0	0	0	0	0
Access	R	R	R	R	R	R	R	R/W

Table 169. Observe\_testbus register (address 6104h) bit allocation

### Table 170. Description of Observe\_testbus bits

Bit	Symbol	Description
7 to 1	-	Reserved.
0	observe_ciu	Configure the pads P3x (P30 to P35), RSTOUT_N and P70_IRQ to observe internal CIU data bus.
		When set to logic 1, the pads are configured in output mode and show the internal data bus D0 to D6 of the CIU. P70_IRQ is the 13.56 MHz digital clock of CIU (generated from field or crystal).

The test bus signals are selected by accessing TestBusSel in register CIU\_TestSel2.

### Table 171. TstBusBitSel set to 07h

Test bus bit	Test signal	Comments
D6	sdata	shows the actual received data value.
D5	scoll	shows if in the actual bit a collision has been detected (106 kbit/s only)
D4	svalid	shows if sdata and scoll are valid
D3	sover	shows that the receiver has detected a stop bit (ISO/IEC 14443A/MIFARE mode only)
D2	RCV_reset	shows if the receiver is reset
D1	RFon filtered	shows the value of the internal RF level detector
D0	Envelope	shows the output of the internal coder

### Table 172. TstBusBitSel set to 0Dh

Test bus bit	Test signal	Comments
D6	clkstable	shows if the oscillator delivers a stable signal
D5	clk27/8	shows the output signal of the oscillator divided by 8
D4	clk27rf/8	shows the clk27rf signal divided by 8
D3	clk13/4	shows the clk13rf divided by 4
D2	clk27	shows the output signal of the oscillator
D1	clk27rf	shows the RF clock multiplied by 2
D0	clk13rf	shows the RF clock of 13.56 MHz

# 8.6.21.3 Test signals at pin AUX

Each signal can be switched to pin AUX1 or AUX2 by setting SelAux1 or SelAux2 in the register CIU\_AnalogTest. See <u>Table 279 on page 185</u>

## 8.6.21.4 PRBS

Enables the Pseudo Random Bit Stream of 9-bit or 15-bit length sequence, PRBS9 or PRBS15, according to ITU-TO150. To start the transmission of the defined datastream, Transmit command has to be activated. The preamble/Sync byte/start bit/parity bit are generated automatically depending on the selected mode.

Note: All relevant registers to transmit data have to be configured before entering PRBS mode according ITU-TO150.

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# 8.6.22 CIU memory map

The registers of the CIU are either map into the SFR or into the XRAM memory space.

#### Table 173. Contactless Interface Unit SFR memory map

ADDR	Byte size	Register name	Description
D1h	1	CIU_Command	Starts and stops the command execution
D2h	1	CIU_CommlEn	Control bits to enable and disable the passing of interrupt requests
D3h	1	CIU_DivlEn	Control bits to enable and disable the passing of interrupt requests
D4h	1	CIU_CommIrq	Contains common Interrupt Request flags
D5h	1	CIU_DivIrq	Contains diverse Interrupt Request flags
D6h	1	CIU_Error	Error flags showing the error status of the last command executed
DFh	1	CIU_Status1	Contains status flags of the CRC, Interrupt Request System and FIFO buffer
E9h	1	CIU_Status2	Contain status flags of the Receiver, Transmitter and Data Mode Detector
EAh	1	CIU_FIFOData	in- and output of 64 bytes FIFO buffer
EBh	1	CIU_FIFOLevel	Indicates the number of bytes stored in the FIFO
ECh	1	CIU_WaterLevel	Defines the thresholds for FIFO under- and overflow warning
EDh	1	CIU_Control	Contains miscellaneous Control bits
EEh	1	CIU_BitFraming	Adjustments for bit oriented frames
EFh	1	CIU_Coll	Bit position of the first bit collision detected on the RF-interface
-			

### Table 174. Contactless Interface Unit extension memory map

ADDR	Byte size	Register name	Description
6301h	1	CIU_Mode	Defines general modes for transmitting and receiving
6302h	1	CIU_TxMode	Defines the transmission data rate and framing during transmission
6303h	1	CIU_RxMode	Defines the transmission data rate and framing during receiving
6304h	1	CIU_TxControl	Controls the logical behavior of the antenna driver pins TX1 and TX2 $% \left( {T_{\rm T}} \right) = 0$
6305h	1	CIU_TxAuto	Controls the settings of the antenna driver
6306h	1	CIU_TxSel	Selects the internal sources for the antenna driver
	1	CIU_RxSel	Selects internal receiver settings
6308h	1	CIU_RxThreshold	Selects thresholds for the bit decoder
6309h	1	CIU_Demod	Defines demodulator settings
630Ah	1	CIU_FelNFC1	Defines the length of the valid range for the received frame
630Bh	1	CIU_FelNFC2	Defines the length of the valid range for the received frame
630Ch	1	CIU_MifNFC	Controls the communication in ISO/IEC 14443/MIFARE and NFC target mode at 106 kbit/s
630Dh	1	CIU_ManualRCV	Allows manual fine tuning of the internal receiver

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ADDR	Byte size	Register name	Description
630Eh	1	CIU_TypeB	Configure the ISO/IEC 14443 type B
630Fh	1	-	Reserved
6310h	1	-	Reserved
6311h	1	CIU_CRCResultMSB	Shows the actual MSB values of the CRC calculation
6312h	1	CIU_CRCResultLSB	Shows the actual LSB values of the CRC calculation
6313h	1	CIU_GsNOff	Selects the conductance of the antenna driver pins TX1 and TX2 for load modulation when own RF field is switched OF
6314h	1	CIU_ModWidth	Controls the setting of the width of the Miller pause
6315h	1	CIU_TxBitPhase	Bit synchronization at 106 kbit/s
6316h	1	CIU_RFCfg	Configures the receiver gain and RF level
6317h	1	CIU_GsNOn	Selects the conductance of the antenna driver pins TX1 and TX2 for modulation, when own RF field is switched ON
6318h	1	CIU_CWGsP	Selects the conductance of the antenna driver pins TX1 an TX2 when not in modulation phase
6319h	1	CIU_ModGsP	Selects the conductance of the antenna driver pins TX1 an TX2 when in modulation phase
631Ah	1	CIU_TMode	Defines settings for the internal timer
631Bh	1	CIU_TPrescaler	
631Ch	1	CIU_TReloadVal_hi	Describes the 16-bit long timer reload value (Higher 8 bits)
631Dh	1	CIU_TReloadVal_lo	Describes the 16-bit long timer reload value (Lower 8 bits)
631Eh	1	CIU_TCounterVal_hi	Describes the 16-bit long timer actual value (Higher 8 bits)
631Fh	1	CIU_TCounterVal_lo	Describes the 16-bit long timer actual value (Lower 8 bits)
6320h	1	-	Reserved
6321h	1	CIU_TestSel1	General test signals configuration
6322h	1	CIU_TestSel2	General test signals configuration and PRBS control
6323h	1	CIU_TestPinEn	Enables test signals output on pins.
6324h	1	CIU_TestPinValue	Defines the values for the 8-bit parallel bus when it is used as I/O bus
6325h	1	CIU_TestBus	Shows the status of the internal test bus
6326h	1	CIU_AutoTest	Controls the digital self-test
6327h	1	CIU_Version	Shows the CIU version
6328h	1	CIU_AnalogTest	Controls the pins AUX1 and AUX2
6329h	1	CIU_TestDAC1	Defines the test value for the TestDAC1
632Ah	1	CIU_TestDAC2	Defines the test value for the TestDAC2
632Bh	1	CIU_TestADC	Show the actual value of ADC I and Q
632Ch	1	-	Reserved for tests
632Dh	1	-	Reserved for tests
632Eh	1	-	Reserved for tests
632Fh	1	CIU_RFlevelDet	Power down of the RF level detector
		CIU_SIC_CLK_en	Enables the use of secure IC clock on P34 / SIC CLK.

### Table 174. Contactless Interface Unit extension memory map ...continued

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# 8.6.23 CIU register description

# 8.6.23.1 CIU register bit behavior

Depending of the functionality of a register, the access condition to the bits can vary. The following table describes the access conditions:

A11	B.I. I.I.	-
Abbreviation	Behavior	Description
R/W	Read and Write	These bits can be written and read by the 80C51. Since they are used only for control means, there content is not influenced by internal state machines, e.g. CIU_CommIEn may be written and read by the CPU. It will also be read by internal state machines, but never changed by them.
DY	DYnamic	These bits can be written and read by the 80C51. Nevertheless, they may also be written automatically by CIU internal state machines, e.g. the commands in the CIU_Command register change their values automatically after their execution.
R	Read only	These registers hold flags, which value is determined by CIU internal states only, e.g. the CRCReady register can not be written from external but shows CIU internal states.
W	Write only	These registers are used for control means only. They may be written by the 80C51 but can not be read. Reading these registers returns always logic 0.
Reserved		These registers are not implemented or reserved for NXP testing use.

# 8.6.23.2 CIU\_SIC\_CLK\_en register (6330h)

Enables the use of P34 / SIC\_CLK as secure IC clock.

#### Table 176. CIU\_SIC\_CLK\_en register (address 6330h) bit allocation

		•	•		· · · · · · · · · · · · · · · · · · ·			
Bit	7	6	5	4	3	2	1	0
Symbol	sic_clk_p34_en	-	-	-	Errorbusbitenable	Error	busbitse	l[2:0]
Reset	0	0	0	0	0	0	0	0
Access	R/W	R	R	R	R/W	R/W	R/W	R/W

### Table 177. Description of CIU\_SIC\_CLK\_en bits

Bit	Symbol	Description						
7	sic_clk_p34_en	Set to logic 1, IC clock: SIC_	this bit configures P34 / SIC_CLK to be used as secure CLK.					
		Set to logic 0,	P34 / SIC_CLK is in normal mode: P34.					
6 to 4	-	Reserved	eserved					
3	Errorbusbitenable	Set to logic 1, enable the error source selected by Errorbusbitsel on AUX pads according to SelAux1 and SelAux2 bits (code 1010b).						
2 to 0	Errorbusbitsel[2:0]	Define the error	or source on ErrorBusBit:					
		Value	Description					
		000	selects ProtocollErr on test bus					
		001	selects ParityErr on test bus					
		010	selects CRCErr on test bus					
		011	selects CollErr on test bus					
		100	selects BufferOvfl on test bus					
		101	selects RFErr on test bus					
		110	selects TempErr on test bus					
		111	selects WrErr on test bus					

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# 8.6.23.3 CIU\_Command register (D1h or 6331h)

Starts and stops the command execution.

# Table 178. CIU\_Command register (address D1h or 6331h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	RcvOff	Power-down	wn Command			
Reset	0	0	1	0	0	0	0	0
Access	R	R	R/W	DY	DY	DY	DY	DY

# Table 179. Description of CIU\_Command bits

Bit	Symbol	Description
7 to 6	-	Reserved
5	RcvOff	Set to logic 1, the analog part of the receiver is switched off.
4	Power-down	Set to logic 1, the CIU Power-down mode is entered. This means, internal current consuming blocks of the contactless analog module are switched off, except for the RF level detector.
		Set to logic 0, the PN532 starts the wake up procedure. During this procedure this bit still shows a logic 1. A logic 0 indicates that the PN532 is ready for operations; see <u>Section 8.6.19.2</u> "CIU Power-down" on page 132.
		<b>Note:</b> The Power-down bit can not be set, when the SoftReset command has been activated.
3 to 0	Command	Activates a command according the Command Code.
		Reading this register shows, which command is actually executed. See
		Section 8.6.20 "CIU command set" on page 133

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# 8.6.23.4 CIU\_CommlEn register (D2h or 6332h)

Control bits to enable and disable the passing of interrupt requests.

# Table 180. CIU\_CommIEn register (address D2h or 6332h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	TxlEn	RXIEn	IdleIEn	HiAlertIEn	LoAlertIEn	ErrlEn	TimerlEn
Reset	0	0	0	0	0	0	0	0
Access	R	R/W	R/W	R/W	R/W	R/W	R/W	R/W

### Table 181. Description of CIU\_CommIEn bits

Bit	Symbol	Description
7	-	Reserved.
6	TxlEn	When set to logic 1, allows the transmitter interrupt request (indicated by bit TxIRq) to be propagated to CIU_IRQ_1.
5	RxIEn	When set to logic 1, allows the receiver interrupt request (indicated by bit RxIRq) to be propagated to CIU_IRQ_1.
4	IdleIEn	When set to logic 1, allows the idle interrupt request (indicated by bit IdleIRq) to be propagated to CIU_IRQ_0.
3	HiAlertIEn	When set to logic 1, allows the high alert interrupt request (indicated by bit HiAlertIRq) to be propagated to CIU_IRQ_1.
2	LoAlertIEn	When set to logic 1, allows the low alert interrupt request (indicated by bit LoAlertIRq) to be propagated to CIU_IRQ_1.
1	ErrlEn	When set to logic 1, allows the error interrupt request (indicated by bit ErrIRq) to be propagated to CIU_IRQ_0.
0	TimerIEn	When set to logic 1, allows the timer interrupt request (indicated by bit TimerIRq) to be propagated to CIU_IRQ_0.

# 8.6.23.5 CIU\_DivIEn register (D3h or 6333h)

Controls bits to enable and disable the passing of interrupt requests.

# Table 182. CIU\_DivIEn register (address D3h or 6333h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	SiginAct IEn	ModelEn	CRCIEn	RfOnIEn	RfOffIEn
Reset	0	0	0	0	0	0	0	0
Access	R	R	R	R/W	R/W	R/W	R/W	R/W

#### Table 183. Description of CIU\_DivIEn bits

Bit	Symbol	Description
7 to 5	-	Reserved.
4	SiginAct IEn	Allows the SIGIN active interrupt request to be propagated to CIU_IRQ_0.
3	ModelEn	When set to logic 1, allows the mode interrupt request (indicated by bit ModelRq) to be propagated to CIU_IRQ_0.
2	CRCIEn	When set to logic 1, allows the CRC interrupt request (indicated by bit CRCIRq) to be propagated to CIU_IRQ_0.
1	RfOnIEn	When set to logic 1, allows the RF field on interrupt request (indicated by bit RfOnIRq) to be propagated to CIU_IRQ_0.
0	RfOfflEn	When set to logic 1, allows the RF field off interrupt request (indicated by bit RfOffIRq) to be propagated to CIU_IRQ_0.

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# 8.6.23.6 CIU\_Commlrq register (D4h or 6334h)

Contains common CIU interrupt request flags.

# Table 184. CIU\_Commlrq register (address D4h or 6334h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	Set1	TxlRq	RxIRq	IdleIrq	HiAltertIRq	LoAlertIRq	ErrlRq	TimerIRq
Reset	0	0	0	1	0	1	0	0
Access	W	DY	DY	DY	DY	DY	DY	DY

### Table 185. Description of CIU\_CommIRQ bits

Bit	Symbol	Description
7	Set1	When set to logic 0 during write operation, the bit set to logic 1 in the write command are written to logic 0 in the register.
		When set to logic 1 during write operation, the bit set to logic 1 in the write command are written to logic 1 in the register.
6	TxIRq	Set to logic 1, immediately after the last bit of the transmitted data was sent out.
5	RxIRq	Set to logic 1 when the receiver detects the end of a valid datastream.
		If the RxNoErr bit in CIU_RxMode register is set to logic 1, RxIRQ is only set to logic 1 when data bytes are available in the FIFO.
4	IdleIrq	Set to logic 1, when a command terminates by itself e.g. when the CIU_Command register changes its value from any command to the Idle command.
		If an unknown command is started, the CIU_Command register changes its value to the Idle command and the IdleIRq bit is set.
		Starting the Idle Command by the 80C51 does not set IdleIRq bit.
3	HiAltertIRq	Set to logic 1, when HiAlert bit in CIU_Status1 register is set to logic 1. In opposition to HiAlert, HiAlertIRq stores this event and can only be reset by Set1 bit.
2	LoAlertIRq	Set to logic 1, when LoAlert bit in CIU_Status1 register is set. In opposition to LoAlert, LoAlertIRq stores this event and can only be reset by Set1 bit.
1	ErrlRq	Set to logic 1, if any error flag in the CIU_Error register is set.
0	TimerIRq	Set to logic 1, when the timer decrements the TimerValue register to zero.

[1] Remark: All bits in the register CIU\_CommIrq shall be set to logic 0 by firmware.

# 8.6.23.7 CIU\_DivIrq register (D5h or 6335h)

Contains miscellaneous interrupt request flags. These bits are latched.

Table 186. CIU\_DivIrq register (address D5h or 6335h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	Set2	-	-	SiginActIrq	ModelRq	CRCIRq	RfOnIRq	RfOffIRq
Reset	0	0	0	Х	0	0	Х	Х
Access	W	R	R	DY	DY	DY	DY	DY

#### Table 187. Description of CIU\_DivIrq bits

Bit	Symbol	Description
7	Set2	When set to logic level 0 during write operation, the bit set to logic 1 in the write command are written to logic 0 in the register.
		When set to logic level 1 during write operation, the bit set to logic 1 in the write command are written to logic 1 in the register.
6 to 5	-	Reserved.
4	SiginActIrq	Set to logic 1, when SIGIN is active. See <u>Section 8.6.13 "NFC-WI/S<sup>2</sup>C</u> <u>interface support" on page 123</u> . This interrupt is set to logic 1 when either a rising or falling edge is detected on SIGIN. <sup>[1]</sup>
3	ModelRq	Set to logic level 1, when the mode has been detected by the Data Mode Detector.
		<b>Note:</b> The Data Mode Detector can only be activated by the AutoColl command and is terminated automatically having the detected the communication mode.
		<b>Note:</b> The Data Mode Detector is automatically restarted after each RF reset.
2	CRCIRq	Set to logic level 1, when the CRC command is active and all data are processed.
1	RfOnIRq	Set to logic level 1, when an external RF field is detected.[1][2]
0	RfOffIRq	Set to logic level 1, when an present external RF field is switched off.[1][2]

[1] At power-up, after reset modes (including Hard Power Down), the logical value of this bit is undefined.

[2] After Power-Down bit of <u>Table 181 on page 147</u> goes from logic 1 to logic 0, after pd\_rfleveldet bit of Table 286 on page 188 goes from logic 1 to logic 0, the logical value of this bit is undefined.

# 8.6.23.8 CIU\_Error register (D6h or 6336h)

Error flags showing the error status of the last command executed.

Table 188.	CIU	_Error	register	(address	D6h d	or 6336h)	bit allocation
------------	-----	--------	----------	----------	-------	-----------	----------------

Bit	7	6	5	4	3	2	1	0
Symbol	WrErr	TempErr	RFErr	BufferOvfl	CollErr	CRCErr	ParityErr	ProtocollErr
Reset	0	0	0	0	0	0	0	0
Access	R	R	R	R	R	R	R	R

### Table 189. Description of CIU\_Error bits

Bit	Symbol	Description
7	WrErr	Set to logic 1, when data is written into the FIFO by the 80C51 during the AutoColl command or MFAuthent command or if data is written into the FIFO by the 80C51 during the time between sending the last bit on the RF interface and receiving the last bit on the RF interface.
6	TempErr <sup>[1]</sup>	Set to logic 1, if the internal temperature sensor detects overheating. In this case the antenna drivers are switched off automatically.
5	RFErr	Set to logic 1, if in active communication mode the counterpart does not switch on the RF field in time as defined in NFCIP-1 standard.
		<b>Note:</b> RFErr is only used in active communication mode. The bit RxFraming or the bit TxFraming has to be set to 01h to enable this functionality.
4	BufferOvfl	Set to logic 1, if the 80C51 or if the internal state machine (e.g. receiver) tries to write data into the FIFO buffer although the FIFO buffer is already full.
3	CollErr	Set to logic 1, if a bit-collision is detected. It is set to logic 0 automatically at receiver start phase. This flag is only valid during the bitwise anticollision at 106 kbit/s. During communication schemes at 212 and 424 kbit/s this flag is always set to logic 0.
2	CRCErr	Set to logic 1, if RxCRCEn in CIU_RxMode register is set to logic 1 and the CRC calculation fails. It is set to logic 0 automatically at receiver start-up phase.
1	ParityErr	Set to logic 1, if the parity check has failed. It is set to logic 0 automatically at receiver start-up phase. Only valid for ISO/IEC 14443A/MIFARE or NFCIP-1 communication at 106 kbit/s.
0	ProtocollErr	Set to logic 1, if one out of the following cases occurs:
		<ul> <li>Set to logic 1 if the SOF is incorrect. It is set to logic 0 automatically at receiver start-up phase. The bit is only valid for 106 kbit in Active and Passive Communication mode.</li> </ul>
		<ul> <li>If bit DetectSync in CIU_Mode register is set to logic 1 during FeliCa communication or Active Communication with transfer speeds higher than 106 kbit, ProtocolErr is set to logic 1 in case of a byte length violation.</li> </ul>
		<ul> <li>During the AutoColl command, ProtocolErr is set to logic 1, if the Initiator bit in CIU_Control register is set to logic 1.</li> </ul>
		<ul> <li>During the MFAuthent Command, ProtocolErr is set to logic 1, if the number of bytes received in one data stream is incorrect.</li> </ul>
		<ul> <li>Set to logic 1, if the Miller Decoder detects 2 pauses below the minimum time according to the ISO/IEC 14443A definitions.</li> </ul>

[1] Command execution will clear all error flags except for bit TempErr. A setting by firmware is impossible.

# 8.6.23.9 CIU\_Status1 register (DFh or 6337h)

Contains status flags of the CRC, Interrupt Request System and FIFO buffer.

			<b>(</b>					
Bit	7	6	5	4	3	2	1	0
Symbol	CIU_IRQ_ 1	CRCOk	CRCReady	CIU_IRQ _0	TRunning	RFOn	HiAlert	LoAlert
Reset	0	0	1	0	0	Х	0	1
Access	R	R	R	R	R	R	R	R

# Table 190. CIU\_Status1 register (address DFh or 6337h) bit allocation

#### Table 191. Description of CIU\_Status1 bits

Bit	Symbol	Description						
7	CIU_IRQ_1	This bit shows, if any CIU_IRQ_1 source requests attention (with respect to the setting of the interrupt enable flags, see CIU_CommIEn and CIU_DivIEn registers).						
		The bit IE1_2 of register IE1 (see <u>Table 13 on page 18</u> ) has to be set to logic 1 to enable the corresponding CPU interrupt.						
6	CRCOk	Set to logic 1, the CRC result is zero. For data transmission and reception the bit CRCOk is undefined (use CRCErr in CIU_Error register). CRCok indicates the status of the CRC coprocessor, during calculation the value changes to logic 0, when the calculation is done correctly, the value changes to logic 1.						
5	CRCReady	Set to logic 1, when the CRC calculation has finished. This bit is only valid for the CRC co-processor calculation using the CalcCRC command.						
4	CIU_IRQ_0	This bit shows, if any CIU_IRQ_0 source requests attention (with respect to the setting of the interrupt enable flags, see CIU_CommIEn and CIU_DivIEn registers).						
		The bit IE1_3 of register IE1 (see <u>Table 13 on page 18</u> ) has to be set to logic 1 to enable the corresponding CPU interrupt.						
3	TRunning	Set to logic 1, the CIU_timer is running, e.g. the CIU_timer will decrement the CIU_TCounterVal_lo with the next timer clock.						
		<b>Note:</b> In the gated mode TRunning is set to logic 1, when the CIU_timer is enabled by the register bits. This bit is not influenced by the gated signal.						
2	RFOn	Set to logic 1, if an external RF field is detected. This bit does not store the state of the RF field.						
1	HiAlert	Set to logic 1, when the number of bytes stored in the FIFO buffer fulfils the following equation:						
		$HiAlert = (64 - FIFOLength) \le WaterLevel$						
		$FIFOLenght = 60, WaterLevel = 4 \rightarrow HiAlert = 1$						
		Example: $FIFOLenght = 59, WaterLevel = 4 \rightarrow HiAlert = 0$						
0	LoAlert	Set to logic 1, when the number of bytes stored in the FIFO buffer fulfills the following equation:						
		$LoAlert = FIFOLength \leq WaterLevel$						
		Example:						
		$FIFOLenght = 4$ , $WaterLevel = 4 \rightarrow LoAlert = 1$						
		$FIFOLenght = 5$ , $WaterLevel = 4 \rightarrow LoAlert = 0$						

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R

# **Near Field Communication (NFC) controller**

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# 8.6.23.10 CIU\_Status2 register (E9h or 6338h)

Access

Contain status flags of the receiver, transmitter and Data Mode Detector.

R

Table 192.	CIU_Status2 reg	ste	r (address E9	h or 6338h) b	it allocation			
Bit	7	6	5	4	3	2	1	0
Symbol	TempSensClear	-	RFFreqOK	TgActivated	MFCrypto1On	Mode	emStat	e[2:0]
Reset	0	0	0	0	0	0	0	0

DY

#### Table 102 CILL Status 2 register (addr C220b) bit allegati ----

R

### Table 193. Description of CIU\_Status2 bits

R/W

		-						
Bit	Symbol	Descri	Description					
7	TempSensClear		ogic 1, this bit clears the temperature error, if the temperature is the alarm limit of 125 °C.					
6	-	Reserv	Reserved					
5	RFFreqOK	Indicat 13.56 I	es if the frequency detected at the RX pin is in the range of MHz.					
			logic 1, if the frequency at the RX pin is in the range 12 MHz < frequency < 15 MHz.					
			The value of RFFreqOK is not defined if the external RF ncy is in the range of 9 to 12 MHz or in the range of 15 to z.					
4	TgActivated		logic 1 if the Select command is received correctly or if the command was answered.					
			This bit can only be set during the AutoColl command in Passive unication mode or Card operating modes.					
		Note: field.	This bit is set to logic 0 automatically by switching off the RF					
3	MFCrypto1On		logic 1, MIFARE Crypto1 unit is switched on and therefore all ommunication with the card is encrypted.					
		This bit can only be set to logic 1 by a successful execution of the MFAuthent command. This is only valid in Reader/Writer mode for MIFARE Classic 1K or MIFARE Classic 4K emulation cards.						
		This bi	t shall be set to logic 0 by firmware.					
2 to 0	ModemState[2:0]	Moden machir	nState shows the state of the transmitter and receiver state nes.					
		Value	Description					
		000	Idle					
		001	Wait for StartSend in CIU_BitFraming register					
		010	TxWait: Wait until RF field is present, if TxWaitRF is set to logic 1. The minimum time for TxWait is defined by the TxWait register.					
		011	Transmitting					
		100	RxWait: Wait until RF field is present, if the bit RxWaitRF is set to logic 1. The minimum time for RxWait is defined by the RxWait register					
		101	Wait for data					
		110	Receiving					

# 8.6.23.11 CIU\_FIFOData register (EAh or 6339h)

In- and output of 64 byte FIFO buffer.

### Table 194. CIU\_FIFOData register (address EAh or 6339h) bit allocation

Bit	7	6	5	4	3	2	1	0		
Symbol		FIFOData[7:0]								
Reset	Х	Х	Х	Х	Х	Х	Х	Х		
Access	DY	DY	DY	DY	DY	DY	DY	DY		

#### Table 195. Description of CIU\_FIFOData bits

Bit	Symbol	Description
7 to 0	FIFOData[7:0]	Data input and output port for the internal 64 bytes FIFO buffer. The FIFO buffer acts as parallel in/parallel out converter for all data stream in- and outputs

# 8.6.23.12 CIU\_FIFOLevel register (EBh or 633Ah)

Indicates the number of bytes stored in the FIFO.

#### Table 196. CIU\_FIFOLevel register (address EBh or 633Ah) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	FlushBuffer			FI	FOLevel[6	:0]		
Reset	0	0	0	0	0	0	0	0
Access	W	R	R	R	R	R	R	R

# Table 197. Description of CIU\_FIFOLevel bits

Bit	Symbol	Description
7	FlushBuffer	Set to logic 1, this bit clears the internal FIFO-buffer's read- and write-pointer and the bit BufferOvfl in the CIU_Error register immediately. Reading this bit will always return logic 0.
6 to 0	FIFOLevel[6:0]	Indicates the number of bytes stored in the FIFO buffer. Writing to the CIU_FIFOData Register increments, reading decrements FIFOLevel.

# 8.6.23.13 CIU\_WaterLevel register (ECh or 633Bh)

Defines the thresholds for FIFO under- and overflow warning.

### Table 198. CIU\_WaterLevel register (address ECh or 633Bh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	WaterLevel[6:0]					
Reset	0	0	0	0	1	0	0	0
Access	R	R	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 199. Description of CIU\_WaterLevel bits

Bit	Symbol	Description
7 to 6	-	Reserved.
5 to 0	WaterLevel[5:0]	This register defines a threshold to indicate a FIFO buffer over- or underflow to the 80C51:
		The HiAlert bit in CIU_Status1 register is set to logic 1, if the remaining number of bytes in the FIFO-buffer space is equal or less than the defined WaterLevel[5:0] bits.
		The LoAlert bit in CIU_Status1 register is set to logic 1, if equal or less than WaterLevel[5:0] bits are in the FIFO.
		<b>Remark:</b> For the calculation of the HiAlert and LoAlert see <u>Table 191</u> on page 151.

# 8.6.23.14 CIU\_Control register (EDh or 633Ch)

Contains miscellaneous control bits.

#### Table 200. CIU\_Control register (address EDh or 633Ch) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	TStopNow	TStartNow	WrNFCIP-1IDtoFIFO	Initiator	-	RxL	.astBits[	2:0]
Reset	0	0	0	0	0	0	0	0
Access	W	W	DY	R/W	R	R	R	R

#### Table 201. Description of CIU\_Control bits

Bit	Symbol	Description
7	TStopNow	Set to logic 1, the timer stops immediately.
		Reading this bit will always return logic 0.
6	TStartNow	Set to logic 1, the timer starts immediately.
		Reading this bit will always return logic 0.
5	WrNFCIP-1IDtoFIFO	Set to logic 1, the internal stored NFCID3 (10 bytes) is copied into the FIFO.
		Afterwards the bit is set to logic 0 automatically
4	Initiator	Set to logic 1, the PN532 acts as Initiator or Reader/Writer, otherwise it acts as Target. or a Card.
3	-	Reserved.
2 to 0	RxLastBits[2:0]	Shows the number of valid bits in the last received byte. If set to 000b, the whole byte is valid.

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# 8.6.23.15 CIU\_BitFraming register (EEh or 633Dh)

Adjustments for bit oriented frames.

# Table 202. CIU\_BitFraming register (address EEh or 633Dh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	StartSend	F	RxAlign[2:0]			TxLastBits[2:0]		
Reset	0	0	0	0	0	0	0	0
Access	W	R/W	R/W	R/W	R	R/W	R/W	R/W

#### Table 203. Description of CIU\_BitFraming bits

Bit	Symbol	Description			
7	StartSend	Set to logic 1, the transmission of data starts.			
		This bit is only valid in combination with the Transceive command.			
6 to 4 RxAlign[2:0]		Used for reception of bit oriented frames: RxAlign[2:0] defines the bit position for the first received bit to be stored in the FIFO. Further received bits are stored in the following bit positions.			
		Example:			
		RxAlign[2:0] = 0: The LSB of the received bit is stored at bit 0, the second received bit is stored at bit position 1.			
		RxAlign[2:0] = 1: The LSB of the received bit is stored at bit 1, the second received bit is stored at bit position 2			
		RxAlign[2:0] = 7: The LSB of the received bit is stored at bit 7, the second received bit is stored in the following byte at bit position 0.			
		These bits shall only be used for bitwise anticollision at 106 kbit/s in Passive Communication or Reader/Writer mode. In all other modes it shall be set to logic 0.			
3	-	Reserved.			
2 to 0	TxLastBits[2:0]	Used for transmission of bit oriented frames: TxLastBits defines the number of bits of the last byte that shall be transmitted. A 000b indicates that all bits of the last byte shall be transmitted.			

# 8.6.23.16 CIU\_Coll register (EFh or 633Eh)

Defines the first bit collision detected on the RF interface.

### Table 204. CIU\_Coll register (address EFh or 633Eh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	ValuesAfterColl	-	CollPosNotValid			CollPos		
Reset	1	0	1	Х	Х	Х	Х	Х
Access	R/W	R	R	R	R	R	R	R

# Table 205. Description of CIU\_Coll bits

Bit	Symbol	Description
7	ValuesAfterColl	If this bit is set to logic 0, all receiving bits will be cleared after a collision.
		This bit shall only be used during bitwise anticollision at 106 kbit/s, otherwise it shall be set to logic 1.
6	-	Reserved
5	CollPosNotValid	Set to logic 1, if no Collision is detected or the Position of the collision is out of range of the CollPos[4:0] bits.
		This bit shall only be interpreted in Passive Communication mode at 106 kbit/s or ISO/IEC 14443A/MIFARE Reader/Writer mode.
4 to 0	CollPos	These bits show the bit position of the first detected collision in a received frame, only data bits are interpreted.
		Example:
		00h indicates a bit collision in the 32 <sup>nd</sup> bit.
		01h indicates a bit collision in the 1 <sup>st</sup> bit
		08h indicates a bit collision in the 8 <sup>th</sup> bit
		This bit shall only be interpreted in Passive Communication mode at 106 kbit/s or ISO/IEC 14443A/MIFARE Reader/Writer mode if CollPosNotValid is set to logic 0.

# 8.6.23.17 CIU\_Mode register (6301h)

Defines general modes for transmitting and receiving.

Table 206. CIU_Mode register (address 6301	ו) bi	oit allocation
--	-------	----------------

		<u> </u>		<i>'</i>				
Bit	7	6	5	4	3	2	1	0
Symbol	MSBFirst	DetectSync	TXWaitRF	RxWaitRF	PolSigin	ModeDet Off	CRCF [1:	Preset :0]
Reset	0	0	1	1	1	0	1	1
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

### Table 207. Description of CIU\_Mode bits

Bit	Symbol	Description				
7	MSBFirst	The bit order i	the CRC co-processor calculates the CRC with MSB first. n the registers CRCResultMSB and the ultLSB is reversed.			
		Note: During	RF communication this bit is ignored.			
6	DetectSync	activated and	, the CIU waits for the F0h byte before the receiver is F0h byte is added as a Sync-byte for transmission. valid for 106 kbit/s during NFCIP-1 data exchange			
		protocol.	value for the total during for on a data excitatinge			
		•	des it shall be set to logic 0.			
5	TXWaitRF	NFCIP-1 can	he transmitter in Reader/Writer or Initiator mode for only be started, if an own RF field is generated (i.e. /or Tx2RFen is set to logic 1).			
4	RxWaitRF	Set to logic 1, the counter for RxWait starts only, if an external RF field is detected in Target mode for NFCIP-1 or in Card Operating mode				
3	PolSigin	PolSigin defin	es the polarity of the SIGIN pin.			
		Set to logic 1,	the polarity of SIGIN pin is active high.			
		Set to logic 0 t	he polarity of SIGIN pin is active low.			
		Note: The inte	rnal envelope signal is coded active low.			
		Note: Changir	ng this bit will generate a SiginActIrq event.			
2	ModeDetOff	Set to logic 1,	the internal Data Mode Detector is switched off.			
		Note: The Dat command.	a Mode Detector is only active during the AutoColl			
1 to 0	CRCPreset[1:0]	Defines the pr command.	eset value for the CRC co-processor for the CalCRC			
			any communication, the preset values is selected according to the mode definition in the CIU_RxMode and registers.			
		Value	Description			
		00	00 00			
		01	63 63			
		10	A6 71			
		11	FF FF			

# 8.6.23.18 CIU\_TxMode register (6302h)

Defines the transmission data rate and framing during transmission.

### Table 208. CIU\_TxMode register (address 6302h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	TxCRCEn	Тx	Speed[2	:0]	InvMod	TxMix	TxFram	ing[1:0]
Reset	0	0	0	0	0	0	0	0
Access	R/W	DY	DY	DY	R/W	R/W	DY	DY

# Table 209. Description of CIU\_TxMode bits

Bit	Symbol	Description	
7	TxCRCEn	Set to logic 1, transmission.	this bit enables the CRC generation during data
		Note: This bit	shall only set to logic 0 at 106 kbit/s.
6 to 4	TxSpeed[2:0]	Defines bit ra	te while data transmission.
		Value	Description
		000	106 kbit/s
		001	212 kbit/s
		010	424 kbit/s
		011	848 kbit/s
		100	1696 kbit/s
		101	3392 kbit/s
		110 - 111	Reserved
		to the bit codi	coding for transfer speeds above 424 kbit/s is equivalent ng of the Active Communication mode of the 424 kbit/s of 8092 / ECMA340.
3	InvMod	Set to logic 1,	the modulation for transmitting data is inverted.
2	TxMix	<b>U</b> ,	the signal at SIGIN is mixed with the internal coder. See 2 "Serial data switch" on page 121.
1 to 0	TxFraming[1:0]	Defines the fr	aming used for data transmission.
		Value	Description
		00	ISO/IEC 14443A/MIFARE and Passive Communication mode 106 kbit/s
		01	Active Communication mode
		10	FeliCa and Passive Communication mode at 212 kbit/s and 424 kbit/s
		11	ISO/IEC 14443B

# 8.6.23.19 CIU\_RxMode register (6303h)

Defines the reception data rate and framing during receiving.

### Table 210. CIU\_RxMode register (address 6303h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	RXCRCEn	R۶	RxSpeed[2:0]		RxNoErr	RxMultiple	RxFram	ning[1:0]
Reset	0	0	0	0	0	0	0	0
Access	R/W	DY	DY	DY	R/W	R/W	DY	DY

Bit	Symbol	Description					
7	RxCRCEn	Set to logic 1,	this bit enables the CRC calculation during reception. The II not be written within the CIU FIFO.				
		Note: This bit	shall only set to logic 0 at 106 kbit/s.				
6 to 4	RxSpeed[2:0]	Defines the bit rate while data receiving.					
			art of the CIU handles only transfer speeds up to 424 kbit/s digital part of the CIU handles the higher transfer speeds as				
		Value	Description				
		000	106 kbit/s				
		001	212 kbit/s				
		010	424 kbit/s				
		011	848 kbit/s				
		100	1696 kbit/s				
		101	3392 kbit/s				
		110 - 111	Reserved				
		the bit coding	coding for transfer speeds above 424 kbit/s is equivalent to of the active communication mode of the 424 kbit/s of the of 8092 / ECMA340.				
3	RxNoErr		, a not valid received data stream (less than 4 bits received) d. The receiver will remain active.				
2	RxMultiple	Set to logic 0,	the receiver is deactivated after receiving a data frame.				
		is only valid fo set this bit, the automatically.	it is possible to receive more than one data frame. This bit r 212 and 424 kbit/s to handle the Polling command. Having e receive and transceive commands will not end In this case the multiple receiving can only be deactivated Idle command to the CIU_Command register or clearing this 51.				
			1, at the end of a received data stream an error byte is FIFO. The error byte is a copy of the CIU_Error register.				
1 to 0	RxFraming[1:0]	Defines the ex	spected framing for data reception.				
		Value	Description				
		00	ISO/IEC 14443A/MIFARE and Passive Communication mode 106 kbit/s				
		01	Active communication mode				
		10	FeliCa and Passive Communication mode at 212 kbit/s and 424 kbit/s				
		11	ISO/IEC 14443B				

# Table 211. Description of CIU\_RxMode bits

# 8.6.23.20 CIU\_TxControl register (6304h)

Controls the logical behavior of the antenna driver pins TX1 and TX2. See also<u>Table 154</u> on page 114 and <u>Table 155 on page 114</u>.

Table 212. Cll	J_TxControl	register	(address	6304h) bit allocation
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Bit	7	6	5	4	3	2	1	0
Symbol	InvTx2 RFon	InvTx1 RFon	InvTx2 RFoff	InvTx1 RFoff	Tx2 CW	CheckRF	Tx2 RFEn	Tx1 RFEn
Reset	1	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	W	R/W	R/W

# Table 213. Description of CIU\_TxControl bits

# Bit Symbol Description

7 InvTx2RFon Set to logic1 and Tx2RFEn set to logic 1, TX2 output signal is inverted.

- 6 InvTx1RFon Set to logic1 and Tx1RFEn set to logic 1, TX1 output signal is inverted.
- 5 InvTx2RFoff Set to logic1 and Tx2RFEn set to logic 0, TX2 output signal is inverted.
- 4 InvTx1RFoff Set to logic1 and Tx1RFEn set to logic 0, TX1 output signal is inverted.

3	Tx2CW	Set to logic 1, the output signal on pin TX2 will deliver continuously the un-modulated 13.56 MHz energy carrier.
		Set to logic 0, Tx2CW is enabled to modulate of the 13.56 MHz energy carrier.
2	CheckRF	Set to logic 1, Tx2RFEn and Tx1RFEn can not be set if an external RF field is detected.
		Only valid when using in combination with Tx2RFAutoEn and TX1RFAutoEn bits in CIU_TxAuto register.
1	Tx2RFEn	Set to logic 1, the output signal on pin TX2 will deliver the 13.56 MHz energy carrier modulated by the transmission data.
0	Tx1RFEn	Set to logic 1, the output signal on pin TX1 will deliver the 13.56 MHz energy carrier modulated by the transmission data.

# 8.6.23.21 CIU\_TxAuto register (6305h)

Controls the setting of the antenna driver.

Table 214.	CIU	TxAuto	register	(address	6305h)	bit allocation
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Bit	7	6	5	4	3	2	1	0
Symbol	Auto RFOFF	Force 100ASK	AutoWakeUp	-	CAOn	InitialRFOn	Tx2 RFAutoEn	Tx1 RFAutoEn
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R	R/W	W	R/W	R/W

# Table 215. Description of CIU\_TxAuto bits

Bit	Symbol	Description
7	AutoRFOff	Set to logic 1, own RF field is switched off after the last data bit has been transmitted as defined in the NFCIP-1 standard.
6	Force100ASK	Set to logic 1, Force100ASK forces a 100% ASK modulation independent of the setting in CIU_ModGsP register.
5	AutoWakeUp	Set to logic 1, the PN532 in CIU Power-down mode can be woken up by the RF level detector.
4		Reserved
3	CAOn	Set to logic 1, the collision avoidance is activated and internally the value n is set in accordance to the ISO/IEC 18092 / ECMA340 NFCIP-1 standards.
2	InitialRFOn	Set to logic 1, the initial RF collision avoidance is performed and the bit InitialRFOn is set to logic 0 automatically, if the RF is switched ON.
		<b>Note:</b> The driver(s) which should be switched on, have to enabled by Tx2RFAutoEn and/or Tx1RFAutoEn bits.
		<b>Note:</b> If the own RF field is already ON when the bit InitialRFOn is set, it is not set to logic 0.
1	Tx2RFAutoEn	Set to logic 1, RF is switched on at TX2 (i.e. Tx2RFEn is set to logic 1) after the external RF field is switched off according to the time TADT. If the InitialRFOn and Tx2RFAutoEn bits are set to logic 1, RF is switched on at TX2 if no external RF field is detected during the time TIDT.
		<b>Note:</b> The times TADT and TIDT are in accordance to the ISO/IEC 18092 / ECMA340 NFCIP-1 standards.
0	Tx1RFAutoEn	Set to logic 1, RF is switched on at TX1(i.e. Tx1RFEn is set to logic 1) after the external RF field is switched off according to the time TADT. If the InitialRFOn and Tx1RFAutoEn bits are set to logic 1, RF is switched on at TX1 if no external RF field is detected during the time TIDT.
		<b>Note:</b> The times TADT and TIDT are in accordance to the ISO/IEC 18092 / ECMA340 NFCIP-1 standards.

# 8.6.23.22 CIU\_TxSel register (6306h)

Selects the sources for the analogue transmitter part

### Table 216. CIU\_TxSel register (address 6306h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	LoadMo	LoadModSel[1:0]		DriverSel[1:0]		SigOutSel[3:0]		
Reset	0	0	0	1	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Bit	Symbol	Description	
7 to 6	LoadModSel[1:0]	Selects the s	signal to be output on LOADMOD
		Value	Description
		00	Tristate
		01	Modulation signal (envelope) from the internal coder
		10	Modulation signal (envelope) from SIGIN
		11	Test signal defined by LoadModtest in register CIU_TestSel1
5 to 4	DriverSel[1:0]	Selects the s	signals to be output on Tx1 and Tx2.
		Value	Description
		00	Tristate
		01	Modulation signal (envelope) from the internal coder
		10	Modulation signal (envelope) from SIGIN
		11	HIGH
			HGH level depends on the setting of N/InvTx1RFOff and InvTx2RFON/InvTx2RFOff.
3 to 0	SigOutSel[3:0]	Select the si	ignal to be output on SIGOUT
		0000	Tristate
		0001	Low
		0010	High
		0011	Test bus signal as defined by TestBusBitSel in CIU_TestSel1.
		0100	Modulation signal (envelope) from the internal coder
		0101	Serial data stream to be transmitted
		0110	Output signal of the receiver circuit (card modulation sign regenerated and delayed). This signal is used as data output signal for secure IC interface connection using 3 lines.
			<b>Note:</b> To have a valid signal the CIU has to be set to the receiving mode by either the Transceive or Receive command. The RxMultiple bit can be used to keep the Cl in receiving mode.
			<b>Note:</b> Do not use this setting in ISO/IEC 14443A/MIFAR mode. Data collisions will not be transmitted on SIGOUT when using Manchester coding.
		0111	Serial data stream received.
			<b>Note:</b> Do not use this setting in ISO/IEC 14443A/MIFAR mode. Miller coding parameters as the bitlength can vary

#### Table 217. Description of CIU\_TxSel bits

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Bit	Symbol	Description	n
		1000-1011	FeliCa secure IC modulation
			1000 RX*
			1001 TX
			1010 Demodulator comparator output
			1011 Reserved
			<b>Note:</b> * To have a valid signal the CIU has to be set to the receiving mode by either the Transceive or Receive commands. The bit RxMultiple can be used to keep the CIU in receiving mode
		1000-1011	MIFARE secure IC modulation
			1100 RX* with RF carrier
			1101 TX with RF carrier
			1110 RX with RF carrier unfiltered
			1111 RX envelope unfiltered
			<b>Note:</b> * To have a valid signal the CIU has to be set to the receiving mode by either the Transceive or Receive commands. The bit RxMultiple can be used to keep the CIU in receiving mode

#### Table 217. Description of CIU\_TxSel bits ... continued

### 8.6.23.23 CIU\_RxSel register (6307h)

Selects internal receiver settings.

# Table 218. CIU\_RxSel register (address 6307h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	UartSel[1:0]				RxWa	it[5:0]		
Reset	1	0	0	0	0	1	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

### Table 219. Description of CIU\_RxSel bits

Bit	Symbol	Description			
7 to 6	UartSel[1:0]	Selects the input of the digital part (CL UART) of the CIU			
		Value	Description		
		00	Constant Low		
		01	Envelope signal at SIGIN		
		10	Modulation signal from the internal analog part		
		11	Modulation signal from SIGIN pin. Only valid for transfer speeds above 424 kbit/s		
5 to 0	RxWait[5:0]		ssion, the activation of the receiver is delayed for RxWait this 'frame guard time' any signal at pin Rx is ignored.		
		(e.g. Transceive, <i>J</i> the mode of the C Communication m transmitted data s	ignored by the Receive command. All other commands Autocoll, MFAuthent) use this parameter. Depending on IU, the counter starts differently. In Passive node the counters starts with the last modulation of the tream. In Active Communication mode the counter starts the external RF field is switched on.		

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# 8.6.23.24 CIU\_RxThreshold register (6308h)

Selects thresholds for the bit decoder.

# Table 220. CIU\_RxThreshold register (address 6308h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol		MinLev	vel[3:0]		-	(	Collevel[2:0	)]
Reset	1	0	0	0	0	1	0	0
Access	R/W	R/W	R/W	R/W	R	R/W	R/W	R/W

### Table 221. Description of CIU\_RxThreshold bits

Bit	Symbol	Description
7 to 4	MinLevel[3:0]	Defines the minimum signal strength at the decoder input that shall be accepted. If the signal strength is below this level, it is not evaluated.
3	-	Reserved
2 to 0	Collevel[2:0]	Defines the minimum signal strength at the decoder input that has to be reached by the weaker half-bit of the Manchester-coded signal to generate a bit-collision relatively to the amplitude of the stronger half-bit.

# 8.6.23.25 CIU\_Demod register (6309h)

Defines demodulator settings.

#### Table 222. CIU\_Demod register (address 6309h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	AddIQ[1:0]		FixIQ	-	TauR	cv[1:0]	TauSy	nc[1:0]
Reset	0	1	0	0	1	1	0	1
Access	R/W	R/W	R/W	R	R/W	R/W	R/W	R/W

### Table 223. Description of CIU\_Demod bits

Bit	Symbol	Description
7 to 6	AddIQ[1:0]	Defines the use of I and Q channel during reception.
		<b>Note:</b> FixIQ has to be set to logic 0 to enable the following settings. ValueDescription
		00Select the stronger channel
		01Select the stronger and freeze the selected during communication
		10Combines the I and Q channel
		11RFU
5	FixIQ	If set to logic 1 and AddIQ[0] is set to logic 0, the reception is fixed to I channel.
		If set to logic 1 and AddIQ[0] is set to logic 1, the reception is fixed to Q channel.
4	-	Reserved
3 to 2	TauRcv[1:0]	Changes time-constant of internal PLL during data receiving.
		Note: If set to 00h, the PLL is frozen during data receiving.
1 to 0	TauSync[1:0]	Changes time-constant of internal PLL during burst (out of data reception)

# 8.6.23.26 CIU\_FeINFC1 register (630Ah)

Defines the length of the FeliCa Sync bytes and the minimum length of the received frame.

<b>Table 224.</b>	CIU	_FeINFC1	register	(address	630Ah)	bit allocation
-------------------	-----	----------	----------	----------	--------	----------------

Bit	7	6	5	4	3	2	1	0
Symbol	FelSync	Len[1:0]			DataLen	Min[5:0]		
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 225. Description of CIU\_FeINFC1 bits

Bit	Symbol	Description				
7 to 6	FelSyncLen[1:0]]	Defines the length of the Sync bytes.				
		Value	Description			
		00	B2 4D			
		01	00 B2 4D			
		10	00 00 B2 4D			
		11	00 00 00 B2 4D			
5 to 0	DataLenMin[5:0]	These bits def	ine the minimum length of the accepted frame length.			
		Data	$LenMin \times 4 \leq DataPacketLenght$			
		This parameter is ignored at 106 kbit/s if the DetectSync bit in CIU_Mode register is set to logic 0. If a received frame is shorter a defined DataLenMin value, the frame will be ignored.				

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# 8.6.23.27 CIU\_FeINFC2 register (630Bh)

Defines the maximum length of the received frame.

### Table 226. CIU\_FeINFC2 register (address 630Bh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	WaitForSelected	ShortTimeSlot			DataLen	Max[5:0]		
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 227. Description of CIU\_FelNFC2 bits

Bit	Symbol	Description
7	WaitForSelected	Set to logic 1, the AutoColl command is automatically ended only when:
		<ol> <li>A valid command has been received after performing a valid Select procedure according to ISO/IEC 14443A.</li> </ol>
		<ol><li>A valid command has been received after performing a valid Polling procedure according to the FeliCa specification.</li></ol>
		Note: If this bit is set, no Active Communication is possible.
		<b>Note:</b> Setting this bit reduces the 80C51 interaction in case of a communication to another device in the same RF field during Passive Communication mode.
6	ShortTimeSlot	Defines the time slot length for Active Communication mode at 424 kbit/s.
		Set to logic 1 a short time slot is used (half of the timeslot at 212 kbit/s).
		Set to logic 0 a long timeslot is used (equal to the timeslot for 212 kbit/s).
5 to 0	DataLenMax[5:0]	These bits define the maximum length of the accepted frame length:
		$DataLenMax \times 4 \ge DataPacketLenght$
		Note: If set to logic 0 the maximum data length is 256 bytes.
		This parameter is ignored at 106 kbit/s if the bit DetectSync in register CIU_Mode is set to logic 0.
		If a received frame is larger as the defined DataLenMax value, the frame will be ignored.

# 8.6.23.28 CIU\_MifNFC register (630Ch)

Defines ISO/IEC 14443A/MIFARE/NFC specific settings in target or card operating mode.

			(	,				
Bit	7	6	5	4	3	2	1	0
Symbol	SensMiller[2:0]			TauMil	ler[1:0]	MFHalted	TxWa	uit[1:0]
Reset	0	1	1	0	0	0	1	0
Access	R/W	R/W	R/W	R/W	R/W	DY	R/W	R/W

# Table 228. CIU\_MifNFC register (address 630Ch) bit allocation

# Table 229. Description of CIU\_MifNFC bits

Bit	Symbol	Description
7 to 5	SensMiller[2:0]	This bit defines the sensitivity of the Miller decoder.
4 to 3	TauMiller[1:0]	This bit defines the time constant of the Miller decoder.
2	MFHalted	Set to logic 1, this bit indicates that the CIU is set to HALT mode in Card Operating mode at 106 kbit/s. This bit is either set by the 80C51 or by the internal state machine and indicates that only the code 52h is accepted as a Request command.
		This bit is automatically set to logic 0 by RF reset.
1 to 0	TxWait[1:0]	In combination with TxBitPhase[6:0] in CIU_TxBitPhase register, defines the additional response time for the target at 106 kbit/s in Passive Communication mode and during the AutoColl command. See CIU_TxBitPhase register.

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# 8.6.23.29 CIU\_ManualRCV register (630Dh)

Allows manual fine tuning of the internal receiver.

**IMPORTANT NOTE:** For standard application it is not recommended to change this register settings.

Bit	7	6	5	4	3	2	1	0
Symbol	-	FastFilt MF_SO	Delay MF_SO	ParityDisable	LargeBWPLL	ManualHPCF	HPCI	-[1:0]
Reset	0	0	0	0	0	0	0	0
Access	R	R/W	R/W	R/W	R/W	R/W	R/W	R/W

# Table 231. Description of CIU\_ManualRCV bits

Bit	Symbol	Description
7	-	Reserved
6	FastFiltMF_SO	If this bit is set to logic 1, the internal filter for the Miller-Delay circuit is set to Fast-Mode
		<b>Note:</b> This bit should only be set to logic 1, if the Miller pauses length expected are less than 400 ns. At 106 kbit/s, the Miller pauses duration is around 3 $\mu$ s.
5	DelayMF_SO	If this bit is set to logic 1, when SigoutSel=1100b (register 6306h), the Signal at SIGOUT-pin is delayed according the delay defined by TxBitPhase[6:0] (register 6315h) and TxWait bits (register 630Ch).
		<b>Note:</b> In ISO/IEC 14443A/MIFARE Card MIFARE Classic 1K or MIFARE Clasic 4K card emulation (Virtual Card) mode (DriverSel = 10b and SigoutSel=1110b), the Signal at SIGIN must then be 128 /fc faster compared to the ISO/IEC 14443A restrictions on the RF-Field for the Frame Delay Time.
		<b>Note:</b> This delay shall only be activated for setting bits SigOutSel to (1110b) or (1111b) in register CIU_TxSel.
		If this bit is set to logic 0, the SIGOUT-pin delay is not adjustable.
		<b>Note:</b> In ISO/IEC 14443A/MIFARE Card MIFARE Classic 1K or MIFARE Clasic 4K card emulation (Virtual Card) mode (DriverSel = 10b and SigoutSel=1110b), the ISO/IEC 14443A restrictions on the RF-Field for the Frame Delay Time should be adjusted on the secure IC side
4	ParityDisable	If this bit is set to logic 1, the generation of the Parity bit for transmission and the parity check for receiving is switched off. The received parity bit is handled like a data bit.

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Bit	Symbol	Description	n					
3	LargeBWPLL	Set to logic extended.	Set to logic 1, the bandwidth of the internal PLL for clock recovery is extended.					
		<b>Note:</b> As the bandwidth is extended, the PLL filtering effect is weaker and the performance of the communication may be affected.						
2	ManualHPCF	•	Set to logic 0, the HPCF[1:0] bits are ignored and the HPCF[1:0] settings are adapted automatically to the receiving mode.					
1 to 0	HPCF[1:0]	Selects the internal rec	High Pass Corner Frequency (HPCF) of the filter in the eiver chain					
		Value	Description					
		00	For signals with frequency spectrum down to 106 kHz					
		01	For signals with frequency spectrum down to 212 kHz					
		10	For signals with frequency spectrum down to 424 kHz					
		11	For signals with frequency spectrum down to 848 kHz					

#### Table 231. Description of CIU\_ManualRCV bits ...continued

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# 8.6.23.30 CIU\_TypeB register (630Eh)

Selects the specific settings for the ISO/IEC 14443B

### Table 232. CIU\_TypeB register (address 630Eh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	Rx SOFReq	Rx EOFReq	-	EOFSOF Width	NoTx SOF	NoTx EOF	TxEG	T[1:0]
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R	R/W	R/W	R/W	R/W	R/W

### Table 233. Description of CIU\_TypeB bits

Bit	Symbol	Description							
7	RxSOFReq		If this bit is set to logic 1, the SOF is required. A datastream starting without SOF is ignored.						
			If this bit is set to logic 0, a datastream with and without SOF is accepted. The SOF will be removed and not written into the FIFO.						
6	RxEOFReq	without EOF v	t to logic 1, the EOF is required. A datastream ending vill generate a protocol error: ProtocollErr in the CIU_Error e set to logic 1.						
		If this bit is set to logic 0, a datastream with and without EOF is accepted. The EOF will be removed and not written into the FIFO.							
5	-	Reserved.	Reserved.						
4	EOFSOFWidth	If this bit is set to logic 1, the SOF and EOF will have the maximum length defined in the ISO/IEC 14443B.							
		If this bit is set to logic 0, the SOF and EOF will have the minimum length defined in the ISO/IEC 14443B.							
3	NoTxSOF	If this bit is set	t to logic 1, the generation of the SOF is suppressed.						
2	NoTxEOF	If this bit is set	t to logic 1, the generation of the EOF is suppressed.						
1 to 0	TxEGT[1:0]	These bits def ISO/IEC 1444	ine the length of the EGT, as defined in the 3B						
		Value	Description						
		00	0 bit						
		01	1 bit						
		10	2 bits						
		11	3 bits						

# 8.6.23.31 CIU\_CRCResultMSB register (6311h)

Shows the actual MSB values of the CRC calculation.

Note: The CRC is split into two 8-bit registers. See also the CIU\_CRCResultLSB register.

Note: Setting the bit MSBFirst in CIU\_Mode register reverses the bit order, the byte order is not changed

Table 234. CIU\_CRCResultMSB register (address 6311h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol		CRCResultMSB[7:0]						
Reset	1	1	1	1	1	1	1	1
Access	R	R	R	R	R	R	R	R

### Table 235. Description of CIU\_CRCResultMSB bits

Bit	Symbol	Description
7 to 0	CRCResultMSB[7:0]	This register shows the actual value of the most significant byte of the CRC calculation. It is valid only if CRCReady bit in CIU_Status1 register is set to logic 1.

# 8.6.23.32 CIU\_CRCResultLSB register (6312h)

Shows the actual LSB values of the CRC calculation.

Note: The CRC is split into two 8-bit registers. See also the CIU\_CRCResultMSB register.

Note: Setting the bit MSBFirst in CIU\_Mode register reverses the bit order, the byte order is not changed

# Table 236. CIU\_CRCResultLSB register (address 6312h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	CRCResultLSB[7:0]							
Reset	1	1	1	1	1	1	1	1
Access	R	R	R	R	R	R	R	R

#### Table 237. Description of CIU\_CRCResultLSB bits

Bit	Symbol	Description
7 to 0	CRCResultLSB[7:0]	This register shows the actual value of the most significant byte of the CRC register. It is valid only if CRCReady bit in CIU_Status1 register is set to logic 1.

# 8.6.23.33 CIU\_GsNOff register (6313h)

Selects the conductance for the N-driver of the antenna driver pins TX1 and TX2 when there is no RF generated by the PN532.

Table 238.	CIU	_GsNOff	register	(address	6313h) bit a	allocation
------------	-----	---------	----------	----------	--------------	------------

Bit	7	6	5	4	3	2	1	0	
Symbol	CWGsNOff[3:0]				ModGsNOff[3:0]				
Reset	1	0	0	0	1	0	0	0	
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

### Table 239. Description of CIU\_GsNOff bits

Bit	Symbol	Description
7 to 4	CWGsNOff[3:0]	The value of this register defines the conductance of the output N-driver during the time of no modulation and when there is no RF generated by the PN532 (neither Tx1RFEn nor Tx2RFEn is set to logic 1).
		Note: The conductance value is binary weighted.
		<b>Note:</b> During CIU Power-down mode, if DriverSel[1:0] is not equal to 01b, CWGsNOff[3] is set to logic 1. This is not readable in the register.
		<b>Note:</b> The value of the register is only used if no RF is generated by the driver, otherwise the value CWGsNOn in the CIU_GsNOn register is used.
3 to 0	ModGsNOff[3:0]	The value of this register defines the conductance of the output N-driver for the time of modulation and when there is no RF generated by the PN532 (neither Tx1RFEn nor Tx2RFEn is set to logic 1).
		This may be used to regulate the modulation index when doing load modulation.
		Note: The conductance value is binary weighted.
		<b>Note:</b> During CIU Power-down, if DriverSel[1:0] is not equal to 01b, ModGsNOff[3] is set to logic 1. This is not readable in the register.
		<b>Note:</b> The value of the register is only used if no RF is generated by the driver, otherwise the value ModGsNOn in the CIU_GsNOn register is used.

# 8.6.23.34 CIU\_ModWidth register (6314h)

Controls the setting of the modulation width.

### Table 240. CIU\_ModWidth register (address 6314h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol		ModWidth[7:0]						
Reset	0	0	1	0	0	1	1	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

### Table 241. Description of CIU\_ModWidth bits

Bit	Symbol	Description
7 to 0	ModWidth[7:0]	These bits define the width of the Miller modulation as initiator in Active and Passive Communication mode as multiples of the carrier frequency (ModWidth+1 / fc). The maximum value is half the bit period.
		Acting as a target in Passive Communication mode at 106 kbit/s or in Card Operating mode for ISO/IEC 14443A/MIFARE these bits are used to change the duty cycle of the subcarrier frequency.
		Number of cycles with low value: NCLV = (Modwidth modulo 8)+1
		Number of cycles with high value: NCHV = 16 - NCLV

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# 8.6.23.35 CIU\_TxBitPhase register (6315h)

Adjust the bit phase at 106 kbit/s during transmission.

# Table 242. CIU\_TxBitPhase register (address 6315h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	RcvClkChange			Txl	BitPhase[6	6:0]		
Reset	1	0	0	0	0	1	1	1
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

### Table 243. Description of CIU\_TxBitPhase bits

Bit	Symbol	Description
7	RcvClkChange	Set to logic 1, the demodulator's clock is derived from the external RF field.
6 to 0	TxBitPhase[6:0]	TXBitPhase[6:0] in addition with TxWait bits (register 63 0Ch), define a delay to adjust the bit synchronization during Passive Communication mode at 106 kbit/s and in ISO/IEC 14443A/MIFARE Reader/Writer mode. TxBitphase[6:0] are representing a delay in number of carrier frequency clock cycles.
		<b>Note:</b> The ranges to be used for TxWait[1:0] and TxBitPhase[6:0] are between:
		TXWait=01b and TxBitPhase = 1Bh (equivalent to an added delay of 20 clock cycles) and TXWait=01b and TxBitPhase = 7Fh (equivalent to an added delay of 120 clock cycles)
		TxWait=10b and TxBitPhase = 00h (equivalent to an added delay of 121 clock cycles) and TxWait=10b and TxBitPhase = 0Fh (equivalent to an added delay of 136 clock cycles)
		Note: The delay can vary depending of antenna circuits.
		<b>Note:</b> When DriverSel = 01b (the transmitter modulation input is coming from the internal coder), this delay is added to the waiting period before transmitting data in all communication modes.
		<b>Note:</b> When SigoutSel=1110b (CIU_TxSel register), and DelayMF_SC =1b (CIU_ManualRCV register), this delay is added on SIGOUT.
		Note: II the Signal at SIGIN is 128/fc faster compared to the ISO/IEC 14443A restrictions on the RF-Field for the Frame Delay Time, this delay is made so that if the FDT is correct when DriverSel = 01b, the same values of TxWait[1:0] and TxBitPhase[6:0] are also correct for this configuration when DriverSel = 10b (the transmitter modulation input is coming from SIGIN).

# 8.6.23.36 CIU\_RFCfg register (6316h)

Configures the receiver gain and RF level detector sensitivity.

### Table 244. CIU\_RFCfg register (address 6316h) bit allocation

Bit	7	6	5	4	3	2	1	0		
Symbol	RFLevelAmp	F	RxGain[2:0]			RFLevel[3:0]				
Reset	0	1	0	0	1	0	0	0		
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W		

### Table 245. Description of CIU\_RFCfg bits

Bit	Symbol	Descrip tion				
7	RFLevelAmp	0	Set to logic 1, this bit activates the RF level detector's amplifier, see Section 8.6.8 "RF level detector" on page 116.			
6 to 4	RxGain[2:0]	This registe	This register defines the receivers signal voltage gain factor:			
		Value	Description			
		000	18 dB			
		001	23 dB			
		010	18 dB			
		011	23 dB			
		100	33 dB			
		101	38 dB			
		110	43 dB			
		111	48 dB			
3 to 0	RFLevel[3:0]		e sensitivity of the RF level detector, for description see <u>Section</u> evel detector" on page 116.			

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# 8.6.23.37 CIU\_GsNOn register (6317h)

Selects the conductance for the N-driver of the antenna driver pins TX1 and TX2 when generating RF.

Bit	7	6	5	4	3	2	1	0		
Symbol		CWGsNOn[3:0]				ModGsNOn[3:0]				
Reset	1	0	0	0	1	0	0	0		
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W		

### Table 247. Description of CIU\_GsNOn bits

Bit	Symbol	Description
7 to 4	CWGsNOn[3:0]	The value of this register defines the conductance of the output N-driver. during times of no modulation and when the PN532 generates the RF field.
		This may be used to regulate the output power and subsequently current consumption and operating distance.
		Note: The conductance value is binary weighted.
		<b>Note:</b> During CIU Power-down mode, if DriverSel[1:0] is not equal to 01b, CWGsNOn[3] is set to logic 1. This is not readable in the register.
		<b>Note:</b> The value of the register is only used if RF is generated by the driver (either Tx1RFEn or Tx2RFEn is set to logic 1), otherwise the value CWGsNOff in the register CIU_GsNOff is used.
3 to 0	ModGsNOn[3:0]	The value of this register defines the conductance of the output N-driver for the time of modulation and when the PN532 generates the RF field.
		This may be used to regulate the modulation index.
		Note: The conductance value is binary weighted.
		<b>Note:</b> During CIU Power-down mode, if DriverSel[1:0] is not equal to 01b, ModGsNOn[3] is set to logic 1. This is not readable in the register.
		<b>Note:</b> The value of the register is only used if RF is generated by the driver (either Tx1RFEn or Tx2RFEn is set to logic 1), otherwise the value ModGsNOff in the register CIU_GsNOff is used.

# 8.6.23.38 CIU\_CWGsP register (6318h)

Defines the conductance of the P-driver.

### Table 248. CIU\_CWGsP register (address 6318h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-			CWGs	P[5:0]		
Reset	0	0	1	0	0	0	0	0
Access	R	R	R/W	R/W	R/W	R/W	R/W	R/W

### Table 249. Description of CIU\_CWGsP bits

Symbol	Description
-	Reserved.
CWGsP[5:0]	The value of this register defines the conductance of the output P-driver, during times of no modulation.
	This may be used to regulate the output power and subsequently current consumption and operating distance.
	Note: The conductance value is binary weighted.
	<b>Note:</b> During CIU Power-down mode, if DriverSel[1:0] is not equal to 01b, CWGsP[5] is set to logic 1. This is not readable in the register.
	-

### 8.6.23.39 CIU\_ModGsP register (6319h)

Defines the driver P-output conductance for the time of modulation.

#### Table 250. CIU\_ModGsP register (address 6319h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-			ModG	sP[5:0]		
Reset	0	0	1	0	0	0	0	0
Access	R	R	R/W	R/W	R/W	R/W	R/W	R/W

# Table 251. Description of CIU\_ModGsP bits

Bit	Symbol	Description	
7 to 6	-	Reserved.	
5 to 0	ModGsP[5:0]	The value of this register defines the conductance of the output P-driv the time of modulation.	
		This may be used to regulate the modulation index.	
		Note: The conductance value is binary weighted.	
		<b>Note:</b> During CIU Power-down mode, if DriverSel[1:0] is not equal to 01b, ModGsP[5] is set to logic 1. This is not readable in the register.	
		<b>Note:</b> If Force100ASK in CIU_TxAuto register is set to logic 1, the ModGsP[5:0] setting has no effect.	

# 8.6.23.40 CIU\_TMode register (631Ah)

Defines settings for the internal timer.

### Table 252. CIU\_TMode register (address 631Ah) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	TAuto	TGate	ed[1:0]	TAutoRestart		TPrescal	er_Hi[3:0]	
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

# Table 253. Description of CIU\_TMode bits

Bit	Symbol	Description				
7	TAuto	Set to logic 1, the timer starts automatically at the end of the transmission in all communication modes at all speed, or when InitialRFOn (in CIU_TxAuto) is set to logic 1 and the external RF field is switched on. The timer stops immediately after receiving the first data bit if RxMultiple in the CIU_RxMode register is set to logic 0.				
			is set to logic 1, the timer never stops. In this case the stopped by setting the bit TStopNow in register to 1.			
		Set to logic 0 protocol.	indicates, that the timer is not influenced by the			
6 to 5	TGated[1:0]	The internal t	timer is running in gated mode.			
			gated mode, the bit TRunning is logic 1 when the timer is ne register bits. This bit does not influence the gating			
		Value	Description			
		00	No gated mode			
		01	Gated by SIGIN			
		10	Gated by AUX1			
		11	Reserved			
4	TAutoRestart	Ų	the timer automatically restart its count-down from e defined within when reaches zero.			
		Set to logic 0 the timer decrements to zero and the bit TimerIRq is set to logic 1.				
3 to 0	TPrescaler_Hi[3:0]	Defines highe	er 4 bits for the TPrescaler.			
		The following formula is used to calculate f <sub>Timer</sub> :				
			$f_{Timer} = 6,78MHz/T_{PreScaler}$			
		For detailed	description see Section 8.6.17 "CIU_timer" on page 130.			
			caler is defined with TPreScaler_Hi[3:0] in this register ler_LO[7:0] in CIU_TPrescaler.			

# 8.6.23.41 CIU\_TPrescaler register (631Bh)

Define the LSB of the Timer-Prescaler.

#### Table 254. CIU\_TPrescaler register (address 631Bh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	TPrescaler_LO[7:0]							
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 255. Description of CIU\_TPrescaler bits

Bit	Symbol	Description
7 to 0	TPrescaler_LO[7:0]	Defines lower 8 bits for TPrescaler.
		The following formula is used to calculate $f_{\ensuremath{Timer}}$
		$f_{Timer} = 6,78MHz/T_{PreScaler}$
		For detailed description see Section 8.6.17 "CIU_timer" on page

<u>130</u>. Note: The TPreScaler time is defined with TPreScaler\_Hi[3:0] in

CIU\_TMode and TPreScaler\_LO[7:0] in this register.

# 8.6.23.42 CIU\_TReload\_hi register (631Ch)

Defines the MSB of the 16-bit long timer reload value.

#### Table 256. CIU\_TReloadVal\_hi register (address 631Ch) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol		TReloadVal_Hi[7:0]						
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 257. Description of CIU\_TReloadVal\_hi bits

Bit	Symbol	Description
7 to 0	TReloadVal_Hi[7:0]	Defines the higher 8 bits for the TReloadValue.
		With a start event the timer loads with the TReloadValue. Changing this register affects the timer only with the next start event.
		<b>Note:</b> The reload value is defined with TReloadVal_Hi[7:0] in this register and TReloadVal_Lo[7:0] in CIU_TReloadVal_lo

# 8.6.23.43 CIU\_TReloadVal\_lo register (631Dh)

Defines the LSB of the 16 bit long timer reload value.

#### Table 258. CIU\_TReload\_lo register (address 631Dh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol		TReloadVal_Lo[7:0]						
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 259. Description of CIU\_TReload\_lo bits

Bit	Symbol	Description
7 to 0	TReloadVal_Lo[7:0]	Defines the lower 8 bits for the TReloadValue.
		With a start event the timer loads with the TReloadValue. Changing this register affects the timer only with the next start event.
		<b>Note:</b> The reload value is defined with TReloadVal_Lo[7:0] in this register and TReloadVal_Hi[7:0] in CIU_TReload_Hi.

# 8.6.23.44 CIU\_TCounterVal\_hi register (631Eh)

Defines the MSB byte of the current value of the timer.

#### Table 260. CIU\_TCounterVal\_hi register (address 631Eh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	TCounterVal_Hi[7:0]							
Reset	х	x	х	х	x	х	x	х
Access	R	R	R	R	R	R	R	R

### Table 261. Description of CIU\_TCounterVal\_hi bits

Bit	Symbol	Description
7 to 0	TCounterVal_Hi[7:0]	MSB of the current value of the timer (Higher 8 bits).

# 8.6.23.45 Register CIU\_TCounterVal\_lo (631Fh)

Defines the LSB byte of the current value of the timer.

### Table 262. CIU\_TCounterVal\_lo register (address 631Fh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	TCounterVal_LO[7:0]							
Reset	х	x	х	x	x	x	x	х
Access	R	R	R	R	R	R	R	R

# Table 263. Description of CIU\_TCounterVal\_lo bits

Bit	Symbol	Description
7 to 0	TCounterVal_LO[7:0]	LSB of the current value of the timer (Lower 8 bits).

## 8.6.23.46 CIU\_TestSel1 register (6321h)

General test signal configuration.

## Table 264. CIU\_TestSel1 register (address 6321h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	LoadMo	dTst[1:0]	SICclk	sel[1:0]	SICCIkD1	TstB	susBitSel	[2:0]
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Bit	Symbol	Description					
7 to 6	LoadModTst[1:0]	Defines the test signal for the LOADMOD pin					
			its LoadModSel in register CIU_TxSel has to be set to able LoadModTst:				
		Value	Description				
		00	Low				
		01	High				
		10	RFU				
		11	TstBusBit as defined by the TestBusBitSel bit of this register				
5 to 4	SICclksel[1:0]	Defines the	source for the 13.56 MHz secure IC clock				
		Value	Description				
		00	GND - secure IC clock is switched off				
		01	Clock derivated by the internal oscillator				
		10	Internal CIU clock				
		11	Clock derivated from the RF Field				
3	SICClkD1	-	I, the secure IC clock is delivered to P31 / UART_TX if the bit is set to logic 1.				
2 to 0	TstBusBitSel(2:0]	Select the Ts	stBusBit from the test bus.				

### Table 265. Description of CIU\_TestSel1 bits

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## 8.6.23.47 CIU\_TestSel2 register (6322h)

General test signal configuration and PRBS control.

## Table 266. CIU\_TestSel2 register (address 6322h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	TstBusFlip	PRBS9	PRBS15		Tst	BusSel[4	4:0]	
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

### Table 267. Description of CIU\_TestSel2 bits

Bit	Symbol	Description
7	TstBusFlip	If set to logic 1, the internal test bus(D6-D0) is mapped to the external test bus pins by the following order: D4,D3, D2,D6,D5, D0, D1. See Section 8.6.21.2 "CIU test bus" on page 141.
6	PRBS9	Starts and enables the PRBS9 sequence according ITU-TO150.
		<b>Note:</b> All relevant register to transmit data have to be configured before entering PRBS9 mode.
		<b>Note:</b> The data transmission of the defined sequence is started by the Transmit command.
5	PRBS15	Starts and enables the PRBS15 sequence according ITU-TO150.
		<b>Note:</b> All relevant register to transmit data have to be configured before entering PRBS15 mode.
		<b>Note:</b> The data transmission of the defined sequence is started by the Transmit command.
4 to 0	TstBusSel[4:0]	Selects the test bus source. See <u>Section 8.6.21.2 "CIU test bus" on</u> page 141.

## 8.6.23.48 CIU\_TestPinEn register (6323h)

Enable the output drivers for the test pins.

### Table 268. CIU\_TestPinEn register (address 6323h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol				TestPin	En[7:0]			
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

## Table 269. Description of CIU\_TestPinEn bits

Bit	Symbol	Description
7 to 0	TestPinEn[7:0]	Each of the bit enables the output driver for an internal test pin: P70_IRQ (MSB), RSTOUT_N, P35, P34 / SIC_CLK, P33_INT1, P32_INT0, P31 / UART_TX, P30 / UART_RX (LSB). DataEn[7] enables P70_IRQ, DataEn[0] enables P30 / UART_RX.
		<b>Note:</b> The data transmission of the defined sequence is started by the Transmit command.

## 8.6.23.49 CIU\_TestPinValue register (6324h)

Defines the values for the 7 bit test bus signals to be I/O on P70\_IRQ, RSTOUT\_N, P35, P34 / SIC\_CLK, P33\_INT1, P32\_INT0, P31 / UART\_TX and P30 / UART\_RX pins.

### Table 270. CIU\_TestPinValue register (address 6324h) bit allocation

			<u> </u>		-			
Bit	7	6	5	4	3	2	1	0
Symbol	useio		TestPinValue[6:0]					
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

## Table 271. Description of CIU\_TestPinValue bits

Bit	Symbol	Description
7	useio	Set to logic 1, this bit enables the I/O functionality for the internal test bus on the pins P70_IRQ (MSB), RSTOUT, P35, P34 / SIC_CLK, P33_INT1, P32_INT0, P31 / UART_TX, P30 / UART_RX (LSB)
		<b>Note:</b> Before using P34 / SIC_CLK as a test output, the SVDD switch should be closed. See register address 6106h.
6 to 0	TestPinValue[6:0]	UseIO set to logic 1, Read or write the value of the test bus. UseIO set to logic 0, Read 000_0000. No write.

## 8.6.23.50 CIU\_TestBus register (6325h)

Shows the status of the internal test bus.

### Table 272. CIU\_TestBus register (address 6325h) bit allocation

	_	•						
Bit	7	6	5	4	3	2	1	0
Symbol		TestBus[7:0]						
Reset	х	х	x	х	х	х	х	х
Access	R	R	R	R	R	R	R	R

### Table 273. Description of CIU\_TestBus bits

Bit	Symbol	Description
7 to 0	TestBus[7:0]	Shows the status of the internal test bus. The test bus is selected by the register CIU_TestSel2.
		See Section 8.6.21.2 "CIU test bus" on page 141.

## 8.6.23.51 CIU\_AutoTest register (6326h)

Controls the digital self-test.

### Table 274. CIU\_AutoTest register (address 6326h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	AmpRcv	-	-		SelfTes	st[3:0]	
Reset	0	1	0	0	0	0	0	0
Access	R	R	R	R	R/W	R/W	R/W	R/W

## Table 275. Description of CIU\_AutoTest bits

Bit	Symbol	Description
7	-	Reserved.
6	AmpRcv	Set to logic 1 the internal signal processing in the receiver chain is performed non-linear. This increases the operating distance in communication modes at 106 kbit/s.
		<b>Note:</b> Due to non linearity the effects of MinLevel and CollLevel in CIU_RxThreshold register are as well non linear.
5 to 4	-	Reserved
3 to 0	SelfTest[3:0]	Enables the digital Self Test. The self-test can be started by the Selftest command in the CIU_Command register. The self-test is enabled by 1001.
		<b>Note:</b> For default operation the self-test has to be disabled (0000).

### 8.6.23.52 CIU\_Version register (6327h)

Shows the version of the CIU.

### Table 276. CIU\_Version register (address 6327h) bit allocation

Bit	7	6	5	4	3	2	1	0		
Symbol		Product[3:0]				Version[3:0]				
Reset	1	0	0	0	0	0	0	0		
Access	R	R	R	R	R	R	R	R		

## Table 277. Description of CIU\_Version bits

Bit	Symbol	Description
7 to 4	Product	Product 1000 (PN532)
3 to 0	Version	Version 0000

## 8.6.23.53 CIU\_AnalogTest register (6328h)

Bit 7 to

Controls the pins AUX1 and AUX2.

## Table 278. CIU\_AnalogTest register (address 6328h) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol	AnalogSelAux1[3:0]				AnalogSelAux2[3:0]				
Reset	0	0	0	0	0	0	0	0	
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

## Table 279. Description of CIU\_AnalogTest bits

			-
	Symbol	Desc	ription
4	AnalogSelAux1[3:0]		ols the AUX1 pin. Note: All test signals are described in on 8.6.21.3 "Test signals at pin AUX" on page 142.
		0000	Tristate
		0001	DAC output: register CIU_TestDAC1 <sup>[1]</sup>
		0010	DAC output: test signal corr1 <sup>[1]</sup>
		0011	DAC output: test signal corr2 <sup>[1]</sup>
		0100	DAC output: test signal MinLevel <sup>[1]</sup>
		0101	DAC output: ADC_I <sup>[1]</sup>
		0110	DAC output: ADC_Q <sup>[1]</sup>
		0111	DAC output: ADC_I combined with ADC_Q <sup>[1]</sup>
		1000	Test signal for production test
		1001	secure IC clock
		1010	ErrorBusBit as described in Table 177 on page 145
		1011	Low
		1100	TxActive
			At 106 kbit/s: High during Start bit, Data bits, Parity and CRC At 212 kbit/s and 424 kbit/s: High during Preamble, Sync, Data bits and CRC
		1101	RxActive
			At 106 kbit/s: High during Data bits, Parity and CRC At 212 kbit/s and 424 kbit/s: High during Data bits and CRC
		1110	Subcarrier detected
			At 106 kbit/s: not applicable At 212 kbit/s and 424 kbit/s: High during last part of preamble, Sync, Data bits and CRC.
		1111	Test bus bit as defined by the TstBusBitSel in <u>Table 265 on</u> page 181

Bit Symbol		Description					
3 to 0	AnalogSelAux2[3:0]	Controls the AUX2 pin. Note: All test signals are described in Section 8.6.21.3 "Test signals at pin AUX" on page 142.					
		0000 Tristate					
		0001 DAC output: register CIU_TestDAC2 <sup>[2]</sup>					
		0010 DAC output: test signal corr1[2]					
		0011 DAC output: test signal corr2 <sup>[2]</sup>					
		0100 DAC output: test signal MinLevel <sup>[2]</sup>					
		0101 DAC output: ADC_I <sup>[2]</sup>					
		0110 DAC output: ADC_Q <sup>[2]</sup>					
		0111 DAC output: ADC_I combined with ADC_Q <sup>[2]</sup>					
		1000 Test signal for production test					
		1001 secure IC clock					
		1010 ErrorBusBit as described in Table 177 on page 145					
		1011 Low					
		1100 TxActive					
		At 106 kbit/s: High during Start bit, Data bits, Parity and CRC At 212 kbit/s and 424 kbit/s: High during Preamble, Sync, Data bits and CRC					
		1101 RxActive					
		At 106 kbit/s: High during Data bits, Parity and CRC At 212 kbit/s and 424 kbit/s: High during Data bits and CR(					
		1110 Subcarrier detected					
		At 106 kbit/s: not applicable At 212 kbit/s and 424 kbit/s: High during last part of preamble, Sync, Data bits and CRC.					
		1111 Test bus bit as defined by the TstBusBitSel in <u>Table 264 on</u> page 181					

Table 279.	Description	of CIU	AnalogTest	bits continued
	Description		_Analog icst	

[1] Current output. The use of 1  $\,k\Omega$  pull down resistor on AUX1 is recommended.

[2] Current output. The use of 1  $k\Omega$  pull down resistor on AUX2 is recommended.

## 8.6.23.54 CIU\_TestDAC1 register (6329h)

Defines the test value for TestDAC1.

### Table 280. CIU\_TestDAC1 register (address 6329h) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol	-	-		TestDAC1[5:0]					
Reset	0	0	Х	Х	Х	Х	Х	Х	
Access	R	R	R/W	R/W	R/W	R/W	R/W	R/W	

### Table 281. Description of CIU\_TestDAC1 bits

Bit	Symbol	Description
7 to 6	-	Reserved.
5 to 0	TestDAC1[5:0]	Defines the test value for TestDAC1. The output of the DAC1 can be switched to AUX1 by setting AnalogSelAux1 to 0001 in the CIU_AnalogTest register.

## 8.6.23.55 CIU\_TestDAC2 register (632Ah)

Defines the test value for TestDAC2.

### Table 282. CIU\_TestDAC2 register (address 632Ah) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	TestDAC2[6:0]					
Reset	0	0	Х	Х	Х	Х	Х	Х
Access	R	R	R/W	R/W	R/W	R/W	R/W	R/W

### Table 283. Description of CIU\_TestDAC2 bits

Bit	Symbol	Description
7 to 6	-	Reserved.
5 to 0	TestDAC2[6:0]	Defines the test value for TestDAC2. The output of the DAC2 can be switched to AUX2 by setting AnalogSelAux2 to 0001 in the CIU_AnalogTest register.

## 8.6.23.56 CIU\_TestADC register (632Bh)

Shows the actual value of ADC I and Q channel.

### Table 284. CIU\_TestADC register (address 632Bh) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol	ADC_I[3:0]			ADC_Q[3:0]					
Reset	Х	Х	Х	Х	Х	Х	Х	Х	
Access	R	R	R	R	R	R	R	R	

### Table 285. Description of CIU\_TestADC bits

Bit	Symbol	Description
7 to 4	ADC_I[3:0]	Shows the actual value of ADC I channel.
3 to 0	ADC_Q[3:0]	Shows the actual value of ADC Q channel.

## 8.6.23.57 CIU\_RFlevelDet register (632Fh)

Power down of the RF level detector.

### Table 286. CIU\_RFlevelDet register (address 632Fh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	pd_rflvldet	-	-	-	-
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

## Table 287. Description of CIU\_RFlevelDet bits

Bit	Symbol	Description
7 to 5	-	Reserved. These bits must be set to logic 0.
4	pd_rfleveldet	Power down of the RF level detector.
		When set to logic 1, the RF level detector is in power down mode.
3 to 0	-	Reserved. These bits must be set to logic 0.

## 8.7 Registers map

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## 8.7.1 Standard registers

## Table 288. Standard registers mapping

	Register address	Register name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
	6000h to 6102h	Reserved			1	1	-	1		-
e	6103h	Config I0_I1	int1_pol	Reserved	pad_l1	Reserved	pad_l0	enselif	Selif	[1:0]
e	6104h	Observe_testbus				Reserved	1			observe_ciu
e	6105h	Data_rng					data_rng			
e	6106h	Control_switch_rng	Reserved	hide_svdd_sig	sic_switch_overload	sic_switch_en	Reserved	cpu_need_rng	random_dataready	Reserved
e	6107h	GPIRQ	gpirq_level_P71	gpirq_level_P50	gpirq_level_P35	gpirq_level_P34	gpirq_enable_P71	gpirq_enable_P50	gpirq_enable_P35	gpirq_enable_P34
e	6108h	Reserved			"					
e	6109h	LDO	Res	erved	overcurrent_status	sel_over	current[1:0]	enoffset	soft_highspeedreg	control_highspeedre
e	610Ah	i <sup>2</sup> c_wu_control			Reserved			i <sup>2</sup> c_wu_en_wr	i <sup>2</sup> c_wu_en_rd	i <sup>2</sup> c_wu_en
e	610Bh	Reserved								
e	610Ch	Andet_control	andet_bot	andet_up	andet_ith	l[1:0]		andet_ithh[2:0]		andet_en
e	610Dh	Reserved			"					
е	610Eh	NFC_WI_control		Re	served		nfc_wi_status	Reserved	nfc_wi_en_act_req_im	nfc_wi_en_clk
	610Fh to 61FFh	Reserved				1				
e	6200h	PCR CFR			Rese	rved			cpu_fre	eq[1:0]
e	6201h	PCR CER		Re	served		hsu_enable	Reserved		
e	6202h	PCR ILR	Reserved	porpulse_latched	Reserved	enable_pdselif	Reserved	gpirq_level	int1_level	int0_level
e	6203h	PCR Control			Rese	rved			clear_wakeup_cond	soft_reset
e	6204h	PCR Status	i <sup>2</sup> c_wu	gpirq_wu	SPI_wu	HSU_wu	CIU_wu	Reserved	int1_wu	int0_wu
e	6205h	PCR Wakeupen	i <sup>2</sup> c_wu_en	GPIRQ_wu_en	SPI_on_en	HSU_on_en	CIU_wu_en	Reserved	int1_en	int0_en
	6206h to 6300h	Reserved								
e	6301h	CIU_Mode	MSBFirst	DetectSync	TXWaitRF	RxWaitRF	PolSigin	ModeDetOff	CRCPre	eset[1:0]
e	6302h	CIU_TxMode	TxCRCEn		TxSpeed[2:0]		InvMod	TxMix	TxFram	ing[1:0]
е	6303h	CIU_RxMode	RXCRCEn		RxSpeed[2:0]		RxNoErr	RxMultiple	RxFram	ing[1:0]
е	6304h	CIU_TxControl	InvTx2RFon	InvTx1RFon	InvTx2RFoff	InvTx1RFoff	Tx2CW	CheckRF	Tx2RFEn	Tx1RFEn
e	6305h	CIU_TxAuto	AutoRFOFF	Force100ASK	AutoWakeUp	Reserved	CAOn	InitialRFOn	Tx2RFAutoEn	Tx1RFAutoEn
e	6306h	CIU_TxSel	LoadMo	odSel[1:0]	DriverSe	I[1:0]		Sig	OutSel[3:0]	
e	6307h	CIU_RxSel	UartS	Sel[1:0]			R	xWait[5:0]		
e	6308h	CIU_RxThreshold		MinL	.evel[3:0]		Reserved		Collevel[2:0]	
e	6309h	CIU_Demod	Addl	Q[1:0]	FixIQ	Reserved	TauR	cv[1:0]	TauSyı	nc[1:0]
F	630Ah	CIU_FelNFC1	FelSynd	cLen[1:0]			Data	aLenMin[5:0]		

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## Table 288. Standard registers mapping ... continued

5		o. otanuaru regi	otoro mapping							
V532_C1	Register address	Register name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
	630Bh	CIU_FelNFC2	WaitForSelected	ShortTimeSlot			Data	aLenMax[5:0]		
	630Ch	CIU_MifNFC		SensMiller[2:0]		TauN	liller[1:0]	MFHalted	TxWait[         HPCF[1         TxEGT[         GsNOff[3:0]         *Level[3:0]         GsNOn[3:0]         scaler_Hi[3:0]         scaler_Hi[3:0]         scaler_Hi[3:0]         scaler_Hi[3:0]         version         gSelAux2[3:0]	it[1:0]
	630Dh	CIU_ManualRCV	Reserved	FastFiltMF_SO	DelayMF_SO	ParityDisable	LargeBWPLL	ManualHPCF	HPC	-[1:0]
	630Eh	CIU_TypeB	RxSOFReq	RxEOFReq	Reserved	EOFSOFWidth	NoTxSOF	NoTxEOF	TxEG	T[1:0]
	630Fh to 6310h	Reserved		·	·		,		,	
	6311h	CIU_CRCResultMSB				CRO	ResultMSB[7:0]			
	6312h	CIU_CRCResultLSB				CRO	CResultLSB[7:0]			
	6313h	CIU_GsNOff		CWGs	NOff[3:0]			Мос	lGsNOff[3:0]	
	6314h	CIU_ModWidth				N	lodWidth[7:0]			
	6315h	CIU_TxBitPhase	RcvClkChange				TxBitPhase[6	§:0]		
All inf	6316h	CIU_RFCfg	RFLevelAmp		RxGain[2:0]			Rf	FLevel[3:0]	
ormati	6317h	CIU_GsNOn		CWGs	NOn[3:0]			Мос	lGsNOn[3:0]	
on pro	6318h	CIU_CWGsP	Res	erved		CWGsP[5:0]				
vided	6319h	CIU_ModGsP	Res	erved			М	odGsP[5:0]		
in thi	631Ah	CIU_TMode	TAuto	TGa	ited[1:0]	TAutoRestart		TPres	scaler_Hi[3:0]	
s docu	631Bh	CIU_TPrescaler		1		TPr	escaler_LO[7:0]			
Iment	631Ch	CIU_TReloadVal_Hi				TRe	eloadVal_Hi[7:0]			
All information provided in this document is subject to legal disclaimers	631Dh	CIU_TReloadVal_Lo				TRe	loadVal_Lo[7:0]			
ject to	631Eh	CIU_TCounterVal_hi				TCo	unterVal_Hi[7:0]			
legal	631Fh	CIU_TCounterVal_lo				TCo	unterVal_LO[7:0]			
discla	6320h	Reserved								
imers.	6321h	CIU_TestSel1	LoadMo	dTst[1:0]	SICclkse	el[1:0]	SICCIkD1		TstBusBitSel[2:0]	
	6322h	CIU_TestSel2	TstBusFlip	PRBS9	PRBS15			TstBusSel[4	:0]	
	6323h	CIU_TestPinEn				т	estPinEn[7:0]			
	6324h	CIU_TestPinValue	useio				TestPinValue[	6:0]		
	6325h	CIU_TestBus					TestBus[7:0]			
	6326h	CIU_AutoTest	Reserved	AmpRcv	Reserv	/ed		Se	elfTest[3:0]	
	6327h	CIU_Version		Pr	oduct				Version	
0	6328h	CIU_AnalogTest		AnalogS	elAux1[3:0]			Analo	gSelAux2[3:0]	
NXP	6329h	CIU_TestDAC1	Res	erved			Те	stDAC1[5:0]		
B.V. 2	632Ah	CIU_TestDAC2	Res	erved			Те	stDAC2[5:0]		
017. A	632Bh	CIU_TestADC		ADC_I[3:0]				ADC_Q[3:0	)]	
© NXP B.V. 2017. All rights reserved.	632Ch to 632Eh	Reserved								
eserved.	632Fh	CIU_RFlevelDet	Reserved			pd_rfleveldet	Reserved			

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## PN532\_C1 Table 288. Standard registers mapping ... continued

Register address	Register name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
6330h	SIC_CLK	sic_clk_p34_en		Reserved	1	Errorbusbitenable		Errorbusbitsel[2:0]	
6331h	CIU_Command			RcvOff	Power-down		C	Command	
6332h	CIU_CommIEn	Reserved	TxIEn	RXIEn	IdleIEn	HiAlertIEn	LoAlertIEn	ErrlEn	TimerIEn
6333h	CIU_DivIEn		Reserved		SiginAct IEn	ModelEn	CRCIEn	RfOnIEn	RfOffIEn
6334h	CIU_CommIrq	Set1	TxIRq	RxIRq	IdleIrq	HiAltertIRq	LoAlertIRq	ErrlRq	TimerIRq
6335h	CIU_DivIrq	Set2	Re	served	SiginActIrq	ModelRq	CRCIRq	RfOnIRq	RfOffIRq
6336h	CIU_Error	WrErr	TempErr	RFErr	BufferOvfl	CollErr	CRCErr	ParityErr	ProtocollErr
6337h	CIU_Status1	CIU_IRQ_1	CRCOk	CRCReady	CIU_IRQ_0	TRunning	RFOn	HiAlert	LoAlert
6338h	CIU_Status2	TempSensClear	Reserved	RFFreqOK	TgActivated	MFCrypto1On		ModemState[2:0]	1
6339h	CIU_FIFOData				F	IFOData[7:0]	1		
<u>≥</u> 633Ah	CIU_FIFOLevel	FlushBuffer				FIFOLevel[6	:0]		
og 633Bh	CIU_WaterLevel	Res	erved			Wa	terLevel[5:0]		
633Ch	CIU_Control	TStopNow	TStartNow	WrNFCIP-1IDtoFIFO	Initiator	Reserved		RxLastBits[2:0]	
633Dh	CIU_BitFraming	StartSend		RxAlign[2:0]		Reserved		TxLastBits[2:0]	
633Eh	CIU_Coll	ValuesAfterColl	Reserved	CollPosNotValid			CollPos		
All information provided in this docu	Reserved	·							

## 8.7.2 SFR registers

## Table 289. SFR registers mapping

SFR address	Register name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
80h					Reserved				
81h <mark>[1]</mark>	SP Stack Pointer					SP[7:0]			
82h <mark>[1]</mark>	DPL Data Pointer Low					DPL[7:0]			
83h <mark>[1]</mark>	DPH Data Pointer High					DPLH7:0]			
84h to 86h	Reserved								
87h	PCON	SMOD			Reserve	d		CPU_PD	Reserved
88h	T01CON	TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0
89h	T01MOD	GATE1	C/T1	M11	M10	GATE0	C/T0	M01	M00
BAh	TOL	T0L.7	T0L.6	T0L.5	T0L.4	T0L.3	T0L.2	T0L.1	T0L.0
8Bh	T1L	T1L.7	T1L.6	T1L.5	T1L.4	T1L.3	T1L.2	T1L.1	T1L.0
BCh	тон	T0H.7	T0H.6	T0H.5	T0H.4	T0H.3	T0H.2	T0H.1	T0H.0
3Dh	T1H	T1H.7	T1H.6	T1H.5	T1H.4	T1H.3	T1H.2	T1H.1	T1H.0
BEh to 97h	Reserved	l		-1	I	L			
98h	SOCON	SM0	SM1	SM2	REN	TB8	RB8	TI	RI

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## Table 289. SFR registers mapping ... continued

SFI	R	Register name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0		
add	Idress											
99ŀ		SOBUF					BUF[7:0]					
9Ał		RWL					terlevel[7:0]					
9Bł		TWL					terlevel[7:0]					
9CI	Ch	FIFOFS					Freespace[7:0]					
9DI	Dh	FIFOFF				Receiv	eFullness[7:0]					
9Eł	h	SFF	FIFO_EN	Reserved	TWLL	TFF	TFE	RWLH	RFF	RFE		
9Fł	ħ	FIT	Reset	Reserved	WCOL_IRQ	TWLL_IRQ	TFF_IRQ	RWLH_IRQ	ROVR_IRQ	RFF_IRQ		
A0ł	)h	Reserved										
A1h	lh	FITEN	TFLUSH	RFLUSH	EN_WCOL_IRQ	EN_TWLL_IRQ	EN_TFF_ IRQ	EN_RWLH_IRQ	EN_ROVR_IRQ	EN_RFF_ IRG		
A2ł	2h	FDATA				FD	DATA[7:0]					
⊵ A3ł	3h	FSIZE				Rece	eiveSize[7:0]					
A4	h to A7h	Reserved										
A8h	3h	IE0	IE0_7	IE0_6	IE0_5	IE0_4	IE0_3	IE0_2	IE0_1	IE0_0		
A9ł	9h	SPIcontrol	Reser	rved	Enable	Reserved	CPHA	CPOL	IE1	IEO		
AA	٨h	SPIstatus		Re	served	I	TR_FE	RCV_OVR	Reserved	READY		
ABI	3h	HSU_STA	set_bit	R	Reserved		irq_rx_over_en	irq_rx_fer_en	irq_rx_over	irq_rx_fer		
AC	Ch	HSU_CTR	hsu_wu_en	start_frame	tx_stop	bit[1:0]	rx_stopbit	tx_en	rx_en	soft_reset_n		
AD	Dh	HSU_PRE				hsu_prescaler[7:0]						
AEI	Eh	HSU_CNT				hsu_	counter[7:0]					
AFI	ħ	Reserved	1									
BO	)h	P3	Reser	rved	P3[5]	P3[4]	P3[3]	P3[2]	P3[1]	P3[0]		
Allinking and the document is exhibited to be a difference of the document is	lh to B7h		1			Reserved						
B8h	Bh	IP0	IP0_7	IP0_6	IP0_5	IP0_4	IP0_3	IP0_2	IP0_1	IP0_0		
B9ł	h to C7h		<u> </u>			Reserved						
C8	3h	T2CON	TF2	Reserved	RCLK0	TCLK0	Reserved	TR2	C/T2	CP/RL2		
C9	9h	T2MOD			Reserved			T2RD		DCEN		
CA	۹h	RCAP2L	R2L.7	R2L.6	R2L.5	R2L.4	R2L.3	R2L.2	R2L.1	R2L.0		
СВ	3h	RCAP2H	R2H.7	R2H.6	R2H.5	R2H.4	R2H.3	R2H.2	R2H.1	R2H.0		
СС	Ch	T2L	T2L.7	T2L.6	T2L.5	T2L.4	T2L.3	T2L.2	T2L.1	T2L.0		
D CD	Dh	T2H	T2H.7	T2H.6	T2H.5	T2H.4	T2H.3	T2H.2	T2H.1	T2H.0		
CE	Eh to CFh					Reserved						
<u> </u>	)h <mark>[1]</mark>	PSW Program Status Word										
5 ▶ D1I		CIU_Command	Reser	rved	RcvOff	Power-down		Com	imand			
D11 D21 D31		CIU_CommIEn		TxlEn	RXIEn	IdleIEn	HiAlertIEn	LoAlertIEn	ErrlEn	TimerIEn		
d	 3h	CIU_DivIEn		Reserved		SiginAct IEn	ModelEn	CRCIEn	RfOnIEn	RfOfflEn		

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## Table 289. SFR registers mapping ... continued

SFR address	Register name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0			
D4h	CIU_CommIrq	Set1	TxlRq	RxIRq	IdleIrq	HiAltertIRq	LoAlertIRq	ErrlRq	TimerIRq			
D5h	CIU_DivIrq	Set2	Re	served	SiginActIrq	ModelRq	CRCIRq	RfOnIRq	RfOffIRq			
D6h	CIU_Error	WrErr	TempErr	RFErr	BufferOvfl	CollErr	CRCErr	ParityErr	ProtocollEr			
D7h	Reserved		_	-				I				
D8h	I <sup>2</sup> CCON	CR[2]	ENS1	STA	STO	SI	AA	CR	[1:0]			
D9h	I <sup>2</sup> CSTA				1	ST[7:0]						
DAh	I <sup>2</sup> CDAT				2	CDAT[7:0]						
DBh	I <sup>2</sup> CADR				SA[6:0]				GC			
DCh to DEh					Reserved							
DFh	CIU_Status1	CIU_IRQ_1	CRCOk	CRCReady	CIU_IRQ_0	TRunning	RFOn	HiAlert	LoAlert			
E0h <mark>[1]</mark>	ACC Accumulator	ACC[7:0]	_					I				
E1h to E7h	Reserved											
E8h	IE1	IE1_7	Reserved	IE1_5	IE1_4	IE1_3	IE1_2	Reserved	IE1_0			
E9h	CIU_Status2	TempSensClear	Reserved	RFFreqOK	TgActivated	MFCrypto1On		ModemState[2:0]				
EAh	CIU_FIFOData				FIF	OData[7:0]						
EBh	CIU_FIFOLevel	FlushBuffer				FIFOLevel[6	5:0]					
ECh	CIU_WaterLevel		_			Wate	aterLevel[5:0]					
EDh	CIU_Control	TStopNow	TStartNow	WrNFCIP-1ID to FIFO	Initiator	Reserved		RxLastBits[2:0]				
EEh	CIU_BitFraming	StartSend		RxAlign[2:0]		Reserved		TxLastBits[2:0]				
EFh	CIU_Coll	ValuesAfterColl		CollPosNotValid			CollPos					
F0h <sup>[1]</sup>	B register					B[7:0]						
F1h to F3h					Reserved							
F4h	P7FGA						P7CFGA[2]	P7CFGA[1]	P7CFGA[0			
F5h	P7FGB						P7CFGB[2]	P7CFGB[1]	P7CFGB[0			
F6h	Reserved											
F7h	P7		_				P7[2]	P7[1]	P7[0]			
F8h	IP1	IP1_7		IP1_5	IP1_4	IP1_3	IP1_2					
F9h					Reserved							
FAh <sup>[1]</sup>	XRAMP						XRAMP[4:0]					
FBh	Reserved											
FCh	P3FGA			P3CFGA[5]	P3CFGA[4]	P3CFGA[3]	P3CFGA[2]	P3CFGA[1]	P3CFGA[0			
FDh	P3FGB			P3CFGB[5]	P3CFGB[4]	P3CFGB[3]	P3CFGB[2]	P3CFGB[1]	P3CFGB[0			
FEh to FFh				4	Reserved			1				

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## 9. Limiting values

#### Table 290. Limiting values In accordance with the Absolute Maximum Rating System (IEC 60134). Symbol Parameter Conditions Min Max Unit **PV**<sub>DD</sub> Supply Voltage -0.5 4 V -0.5 6.0 V Power Supply Voltage VBAT Total power dissipation P<sub>tot</sub> 500 mW [1] ITVDD Maximum current in TVDD 150 mΑ Maximum current in SVDD switch 30 ISVDD mΑ Electrostatic discharge voltage VESD VESDH ESD Susceptibility (Human Body model) 1500 Ω, 100pF; $\pm 2.0 \text{ kV}$ EIA/JESD22-A114-D ESD Susceptibility (Machine model) V VESDM 0.75 mH, 200 pF; 200 EIA/JESD22-A115-A VESDC ESD Susceptibility (Charge Device model) Field induced model; $\pm 1.0 \text{ kV}$ EIA/JESC22-C101-C T<sub>sta</sub> Storage temperature -55 150 °C Junction temperature -40 125 °C Ti

[1] The antenna should be tuned not to exceed this current limit (the detuning effect when coupling with another device must be taken into account)

## 10. Recommended operating conditions

## Table 291. Operating conditions

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
T <sub>amb</sub>	Ambient Temperature		-30	+25	+85	°C
VBAT	Power Supply Voltage	$V_{SS} = 0 V$	[1] 2.7 [2]		5.5	V
$PV_DD$	Supply Voltage for host interface	$V_{SS} = 0 V$	<u>3</u> 1.6	1.8 to 3.3	3.6	V

[1]  $V_{SS}$  represents  $DV_{SS}$ ,  $TV_{SS1}$ ,  $TV_{SS2}$ ,  $AV_{SS}$ .

[2] Supply voltage of VBAT below 3.3 V reduces the performance (e.g. the achievable operating distance).

[3] It is possible to supply PVDD=0V and to use the PN532 with reduced functionality (see Section 8.4 "Power management" on page 84)

## **11. Thermal characteristics**

## Table 292. Thermal characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
R <sub>thj-a</sub>	thermal resistance from junction to ambient	in free air with exposed pad soldered on a 4 layer Jedec PCB-0.5		37	41.1	K/W

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## **12. Characteristics**

Unless otherwise specified, the limits are given for the full operating conditions. The typical value is given for  $25^{\circ}$ C, VBAT = 3.4 V and PVDD = 3 V.

Timings are only given from characterization results.

## **12.1** Power management characteristics

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
VBAT	Battery power supply range	$V_{SS} = 0 V$		2.7		5.5	V
V <sub>DVDD</sub>	LDO output	VBAT > 3.4 V I <sub>DVDD</sub> = 150 mA		2.7	3	3.4	V
LDO <sub>OFF</sub>	LDO offset	VBAT = $3.4 \text{ V}$ I <sub>DVDD</sub> =150 mA Offset enabled		300	400	700	mV
V <sub>th1</sub>	Reset threshold on $DV_DD$ falling			2.04	2.32	2.6	V
Vhys1	Vth1 hysteresis			40		90	mV
Cdec	DVDD decoupling capacitor	Low ESR, like X7R or X5R ceramic capacitor	<u>[1]</u>	10			μF
PSR	Power supply rejection on $DV_DD$	LDO offset enabled. F <10 kHz. Cdec = 10 $\mu$ F			-46		dB
F <sub>cPSR</sub>	PSR -3dB cut-off frequency	Standard LDO mode. Cdec = 10 $\mu$ F			10		kHz
$SV_{DD}$	$SV_{DD}$ output voltage	VBAT > 3.4 V I <sub>SVDD</sub> = 30 mA		2.7		3.3	VV

Table 293. Power management characteristics

[1] Decreasing the decoupling capacitance can decrease the power supply bursts rejection.

[2] The capacitance should be placed closed to the pins, and can be splitted (see Figure 51 on page 212)

## 12.2 Overcurrent detection

The following values are guaranteed by design. Only functional testing is done in production for case Sel\_overcurrent1 = Sel\_overcurrent0 = 1.

Table 294. Overcurrent detection characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
I <sub>Overcurrent</sub>	$I_{DVDD}$ threshold for overcurrent detection	Sel_overcurrent1=0 Sel_overcurrent0=0		300		mA
		Sel_overcurrent1=0 Sel_overcurrent0=1		Typ         Max           300         -           210         -           180         -           150         -	mA	
		Sel_overcurrent1=1 Sel_overcurrent0=0		180		mA
		Sel_overcurrent1=1 Sel_overcurrent0=1		150		mA

## 12.3 Current consumption characteristics

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
I <sub>HPD</sub>	Hard-Power-Down current	VBAT = 3.4 V PVDD = 3 V	<u>[5]</u>			2	μΑ
I <sub>SPD</sub>	Soft-Power-Down current	VBAT = 3.6 V PVDD = 3 V RF level detector ON	<u>[5][7]</u>		25	45	μΑ
I <sub>SPD</sub>	Soft-Power-Down current	VBAT = 3.6 V PVDD = 3 V RF level detector OFF	<u>[5][7]</u>		18	40	μΑ
I <sub>AVDD</sub>	Analog supply current	VBAT = 3.4 V PVDD = 3 V			3	10	mA
I <sub>PVDD</sub>	Pad supply current		[2]		0.5	45	mA
I <sub>SVDD</sub>	Secure IC supply current	Switch closed	[3]		3	30	mA
I <sub>TVDD</sub>	Transmitter supply current	Continuous wave, VBAT = 3.4 V	<u>[1]</u>		60	100	mA
$I_{VBAT}$	Total supply current	Continuous wave, VBAT = 3.4 V	<u>[4][6]</u>			150	mA

### Table 295. Current consumption characteristics

[1] Typical value using a complementary driver configuration and an antenna matched to 40  $\Omega$  between TX1 and TX2 at 13.56 MHz.

[2] IPV<sub>DD</sub> depends on the overall load at the pins. The maximum is given assuming 4mA output current for the I/O or output pads.

- [3]  $ISV_{DD}$  depends on the overall load on  $SV_{DD}$  pad.
- [4] During operation with recommended antenna tuning circuitry the overall current is below 100 mA.
- [5] ISPD and IHPD are the total currents across all supplies with the PN532xA3HN/C104 and PN532xA3HN/C105.
- [6] The antenna should be tuned not to exceed this current limit (the detuning effect when coupling with another device must be taken into account)
- [7] These values are valid when applied the Soft-Power-Down sequence described in <u>Section 8.5.4 on</u> page 92, and with TESTEN pin connected to DVSS.

## **12.4** Antenna presence self test thresholds

The following values are guaranteed by design. Testing is done in production for cases andet\_ithl[1:0]=10b and for andet\_ithh[2:0]=011b.

The operating range is:

- VBAT voltage above 5V
- Ambient temperature between 0 and 40°C

### Table 296. Antenna presence detection lower levels characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
I <sub>AndetH</sub>	$I_{\text{DVDD}}$ lower current threshold for	andet_ithl[1:0] = 10b	20		37	mA
	antenna presence detection	andet_ithl[1:0] = 11b	27		49	mA

### Table 297. Antenna Presence Detection Upper Levels characteristics

Symbol	Parameter	Conditions	Min	Тур	Мах	Unit
I <sub>AndetH</sub>	$I_{\text{DVDD}}$ upper current threshold for	andet_ithh[2:0] = 000b	31		56	mA
	antenna presence detection	andet_ithh[2:0] = 001b	42		74	mA
		andet_ithh[2:0] = 010b	52		92	mA
		andet_ithh[2:0] = 011b	63		110	mA
		andet_ithh[2:0] = 100b	73		128	mA
		andet_ithh[2:0] = 101b	84		146	mA
		andet_ithh[2:0] = 110b	94		164	mA
		andet_ithh[2:0] = 111b	105		182	mA

## 12.5 Typical 27.12 MHz Crystal requirements

### Table 298. Crystal requirements

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>XTAL</sub>	XTAL frequency		27.107	27.12	27.133	MHz
ESR	Equivalent series resistance			50		Ω
$C_{LOAD}$	Load capacitance			10		pF
$P_{XTAL}$	Drive level		100			μW

## 12.6 Pin characteristics for 27.12 MHz XTAL Oscillator (OSCIN, OSCOUT)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
C <sub>inOSCIN</sub>	OSCIN Input Capacitance	AV <sub>DD</sub> = 2.8 V, VDC = 0.65 V, VAC = 0.9 Vpp		2		pF
VOHOSCOUT	High level output voltage			1.1		V
VOLOSCOUT	Low level output voltage			0.2		V
CinOSCOUT	OSCOUT Input Capacitance			2		pF
foscin	Clock Frequency	with appropriate quartz and capacitances values	[1]	27.12		MHz
D <sub>FEC</sub>	Duty Cycle of Clock Frequency	with appropriate quartz and capacitances values	<u>[1]</u> 45	50	55	%
t <sub>jitter</sub>	Jitter of Clock Edges	with appropriate quartz and capacitances values	<u>[1]</u>		10	ps RMS

## Table 299. Pin characteristics for 27.12 MHz XTAL Oscillator (OSCIN, OSCOUT)

[1] See the Figure 51 on page 212 for example of appropriate connected components. The layout should ensure minimum distance between the pins and the components

## 12.7 RSTPD\_N input pin characteristics

## Table 300. RSTPD\_N input pin characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
Symbol	Falameter	Conditions		тур	Wax	Unit
V <sub>IH</sub>	High level input voltage	PVDD > 1.6 V	$0.9\times PV_{DD}$		3.6	V
V <sub>IH</sub>	High level input voltage	PVDD < 0.4 V (see <u>Section 8.4 on</u> page 84)	$0.65 \times VBAT$		3.6	V
V <sub>IL</sub>	Low level input voltage	PVDD > 1.6 V	0		0.4	V
V <sub>IL</sub>	Low level input voltage	PVDD < 0.4 V (see <u>Section 8.4 on</u> page 84)	0		0.4	V
I <sub>IH</sub>	High level input current	$V_I = PV_{DD}$	-1		1	μA
IIL	Low level input current	$V_{I} = 0 V$	-1		1	μA
C <sub>in</sub>	Input capacitance			2.5		pF

## 12.8 Input pin characteristics for I0 and I1

### Table 301. Input pin characteristics for I0, I1 and TESTEN

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>IH</sub>	High level input voltage		$[1]  0.7 \times DV_{DD}$		$DV_DD$	V
V <sub>IL</sub>	Low level input voltage		<u>[2]</u> 0		$0.3\times DV_{DD}$	V
IIH	High level input current I0 and I1	$V_I = DV_{DD}$	-1		1	μΑ
I <sub>IL</sub>	Low level input current	$V_I = 0 V$	-1		1	μΑ
C <sub>in</sub>	Input capacitance			2.5		pF

[1] To minimize power consumption when in Soft-Power-Down mode, the limit is  $DV_{DD}$  - 0.4 V.

[2] To minimize power consumption when in Soft-Power-Down mode, the limit is 0.4 V.

## 12.9 RSTOUT\_N output pin characteristics

### Table 302. RSTOUT\_N output pin characteristics

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V <sub>OH</sub>	High level output voltage	$PV_{DD} = 3 V,$ $I_{OH} = -4 mA$		$0.7\times PV_{DD}$		$PV_{DD}$	V
		PV <sub>DD</sub> = 1.8 V, I <sub>OH</sub> = -2 mA	[2]	$0.7\times PV_{DD}$		$PV_{DD}$	V
V <sub>OL</sub> Low level of	Low level output voltage	$PV_{DD} = 3 V,$ $I_{OL} = 4 mA$		0		$0.3\times PV_{DD}$	V
		PV <sub>DD</sub> = 1.8 V, I <sub>OL</sub> = 2 mA	[2]	0		$0.3\times PV_{DD}$	V
Cout	Load capacitance					30	pF
t <sub>rise,fall</sub>	Rise and fall times	$\begin{split} PV_{DD} &= 3 \text{ V}, \\ V_{OH} &= 0.8 \times PV_{DD}, \\ V_{OL} &= 0.2 \times PV_{DD}, \\ C_{out} &= 30 \text{ pF} \end{split}$	[1]		13.5		ns
		$\label{eq:VDD} \begin{split} & PV_{DD} = 1.8 \ V, \\ & V_{OH} = 0.7 \times PV_{DD}, \\ & V_{OL} = 0.3 \times PV_{DD}, \\ & C_{out} = 30 \ pF \end{split}$			10.8		ns

[1]  $I_{OH}$  and  $I_{OL}$  give the output drive capability from which the rise and fall times may be calculated as a function of the load capacitance.

[2] Data at PVDD= 1.8V are only given from characterization results.

## 12.10 Input/output characteristics for pin P70\_IRQ

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V <sub>IH</sub>	High level input voltage		[1]	$0.7\times \text{PV}_{\text{DD}}$		$PV_{DD}$	V
V <sub>IL</sub>	Low level input voltage		[2]	0		$0.3\times PV_{DD}$	V
V <sub>OH</sub>	Push-pull mode high level output voltage	PV <sub>DD</sub> = 3 V, I <sub>OH</sub> = -4 mA		$0.7\times PV_{DD}$		$PV_{DD}$	V
		PV <sub>DD</sub> = 1.8 V, I <sub>OH</sub> = -2 mA	[3]	$0.7\times PV_{DD}$		$PV_{DD}$	V
V <sub>OL</sub>	Push-pull mode low level output voltage	PV <sub>DD</sub> = 3 V, I <sub>OL</sub> = 4 mA		0		$0.3\times PV_{DD}$	V
		PV <sub>DD</sub> = 1.8 V, I <sub>OL</sub> = 2 mA	<u>[3]</u>	0		$0.3\times PV_{DD}$	V
I <sub>IH</sub>	Input mode high level input current	$V_{I} = DV_{DD}$		-1		1	μΑ
IIL	Input mode low level input current	V <sub>1</sub> = 0 V		-1		1	μΑ
I <sub>Leak</sub>	Input leakage current	$RSTPD_N = 0.4 V$		-1		1	μA
Cin	Input capacitance				2.5		pF
Cout	Load capacitance					30	pF
t <sub>rise,fall</sub>	Rise and fall times	$\begin{array}{l} PV_{DD} = 3 \ V, \\ V_{OH} = 0.8 \times PV_{DD}, \\ V_{OL} = 0.2 \times PV_{DD}, \\ C_{out} = 30 \ pF \end{array}$			13.5		ns
		$\begin{array}{l} PV_{DD} = 1.8 \ V, \\ V_{OH} = 0.7 \times PV_{DD}, \\ V_{OL} = 0.3 \times PV_{DD}, \\ C_{out} = 30 \ pF \end{array}$			10.8		ns

### Table 303. Input/output pin characteristics for pin P70\_IRQ

[1] To minimize power consumption when in Soft-Power-Down mode, the limit is  $PV_{DD}$  - 0.4 V.

[2] To minimize power consumption when in Soft-Power-Down mode, the limit is 0.4 V

[3] Data at PVDD= 1.8V are only given from characterization results.

## 12.11 Input/output pin characteristics for P30 / UART\_RX, P31 / UART\_TX, P32\_INT0, P33\_INT1

	P33_INT1						
Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V <sub>IH</sub>	High level input voltage		[1]	$0.7\times PV_{DD}$		$PV_{DD}$	V
V <sub>IL</sub>	Low level input voltage		[2]	0		$0.3\times PV_{DD}$	V
V <sub>OH</sub>	Push-pull mode high level output voltage	PV <sub>DD</sub> = 3 V, I <sub>OH</sub> = -4 mA		PV <sub>DD</sub> - 0.4		$PV_{DD}$	V
	Duck and as a large land	PV <sub>DD</sub> = 1.8 V, I <sub>OH</sub> = -2 mA	[3]	PV <sub>DD</sub> - 0.4		$PV_{DD}$	V
V <sub>OL</sub>	Push-pull mode low level output voltage	$PV_{DD} = 3 V,$ $I_{OL} = 4 mA$		0		0.4	V
		$PV_{DD} = 1.8 V,$ $I_{OL} = 2 mA$	[3]	0		0.4	V
I <sub>IH</sub>	Input mode high level input current	$V_I = PV_{DD}$		-1		1	μA
I <sub>IL</sub>	Input mode low level input current	$V_1 = 0 V$		-1		1	μA
I <sub>Leak</sub>	Input leakage current	$RSTPD_N = 0.4 V$		-1		1	μΑ
C <sub>in</sub>	Input capacitance				2.5		pF
Cout	Load capacitance					30	pF
t <sub>rise,fal</sub> l	Rise and fall times	$\begin{split} PV_{DD} &= 3 \text{ V}, \\ V_{OH} &= 0.8 \times PV_{DD}, \\ V_{OL} &= 0.2 \times PV_{DD}, \\ C_{out} &= 30 \text{ pF} \end{split}$			13.5		ns
		$\begin{array}{l} PV_{DD} = 1.8 \ V, \\ V_{OH} = 0.7 \ PV_{DD}, \\ V_{OL} = 0.3 \times PV_{DD}, \\ C_{out} = 30 \ pF \end{array}$			10.8		ns

Table 304. Input/output pin characteristics for P30 / UART\_RX, P31 / UART\_TX, P32\_INT0, P33\_INT1

[1] To minimize power consumption when in Soft-Power-Down mode, the limit is  $PV_{DD}$  - 0.4 V.

[2] To minimize power consumption when in Soft-Power-Down mode, the limit is 0.4 V

[3] Data at PVDD= 1.8V are only given from characterization results.

## 12.12 Input/output pin characteristics for P34 / SIC\_CLK

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V <sub>IH</sub>	High level input voltage		[1]	$0.7\times SV_{DD}$		SV <sub>DD</sub>	V
V <sub>IL</sub>	Low level input voltage		[2]	0		$0.3 \times \mathrm{SV}_{\mathrm{DD}}$	V
V <sub>OH</sub>	Push-pull mode high level output voltage	VBAT = 3.4 V, I <sub>OH</sub> = -4 mA		SV <sub>DD</sub> - 0.4		SV <sub>DD</sub>	V
V <sub>OL</sub>	Push_pull mode low level output voltage	VBAT = 3.4 V, I <sub>OL</sub> = 4 mA		0		0.4	V
I <sub>IH</sub>	Input mode high level input current	$V_{I} = SV_{DD}$		-1		1	μΑ
I <sub>IL</sub>	Input mode low level input current	$V_I = 0 V$		-1		1	μΑ
I <sub>Leak</sub>	Input leakage current	$RSTPD_N = 0.4 V$		-1		1	μA
C <sub>in</sub>	Input Capacitance				2.5		pF
C <sub>out</sub>	Load Capacitance					30	pF
t <sub>rise,fall</sub>	Rise and fall times	$ \begin{array}{l} V_{BAT} = 3.4 \ V, \\ V_{OH} = 0.8 \times SV_{DD}, \\ V_{OL} = 0.2 \times SV_{DD}, \\ C_{out} = 30 \ pF \end{array} $			13.5		ns

### Table 305. Input/output pin characteristics for P34 / SIC\_CLK

[1] To minimize power consumption when in Soft-Power-Down mode, the limit is  $SV_{DD}$  - 0.4 V.

[2] To minimize power consumption when in Soft-Power-Down mode, the limit is 0.4 V.

## 12.13 Input/output pin characteristics for P35

### Table 306. Input/output pin characteristics for P35

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V <sub>IH</sub>	High level Input voltage		[1]	$0.7\times DV_{DD}$		$DV_DD$	V
V <sub>IL</sub>	Low level Input voltage		[2]	0		$0.3\times DV_{DD}$	V
V <sub>OH</sub>	High level output voltage	VBAT = 3.4 V, I <sub>OH</sub> = -4 mA		DV <sub>DD</sub> - 0.4		$DV_DD$	V
V <sub>OL</sub>	Low level output voltage	VBAT = 3.4 V, I <sub>OL</sub> = 4 mA		0		0.4	V
I <sub>IH</sub>	High level input current	$V_{I} = DV_{DD}$		-1		1	μA
I <sub>IL</sub>	Low level input current	$V_{I} = 0 V$		-1		1	μA
I <sub>Leak</sub>	Input leakage current	$RSTPD_N = 0.4 V$		-1		1	μA
C <sub>in</sub>	Input Capacitance				2.5		pF
C <sub>out</sub>	Load Capacitance					30	pF
t <sub>rise,fall</sub>	Rise and fall times	$\begin{split} V_{BAT} &= 3.4 \text{ V}, \\ V_{OH} &= DV_{DD} \text{ - } 0.4, \\ V_{OL} &= 0.4, \\ C_{out} &= 30 \text{ pF} \end{split}$			16.5		ns

[1] To minimize power consumption when in Soft-Power-Down mode, the limit is  $DV_{DD}$  - 0.4 V.

[2] To minimize power consumption when in Soft-Power-Down mode, the limit is 0.4 V.

## 12.14 Input pin characteristics for NSS / P50\_SCL / HSU\_RX

### Table 307. Input pin characteristics for NSS / HSU\_RX for HSU / SPI interfaces

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V <sub>IH</sub>	High level Input voltage	PV <sub>DD</sub> > 1.6V	[1] [3]	$0.7\times PV_{DD}$		$PV_{DD}$	V
V <sub>IL</sub>	Low level Input voltage	PV <sub>DD</sub> > 1.6V	[2]	0		$0.3\times PV_{DD}$	V
I <sub>IH</sub>	High level input current	$V_{I} = DV_{DD}$		-1		1	μΑ
IIL	Low level input current	$V_{I} = 0 V$		-1		1	μΑ
I <sub>Leak</sub>	Input leakage current	$RSTPD_N = 0.4 V$		-1		1	μΑ
C <sub>in</sub>	Input Capacitance				2.5		pF
TSP	Width of suppressed spikes				20		ns

[1] To minimize power consumption when in Soft-Power-Down mode, the limit is  $PV_{DD}$  - 0.4 V.

[2] To minimize power consumption when in Soft-Power-Down mode, the limit is 0.4 V.

[3] When PVDD is not present, it is not possible to define a high level on NSS. When using SPI host interface, a wake-up condition can not be avoided if PVDD is absent.

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V <sub>IH</sub>	High level Input voltage		[1]	$0.7\times \text{PV}_{\text{DD}}$		$PV_{DD}$	V
V <sub>IL</sub>	Low level Input voltage		[2]	0		$0.3\times PV_{DD}$	V
V <sub>OL</sub>	Low level output voltage	$PV_{DD} = 3 V,$ $I_{OL} = 4 mA$		0		0.3	V
		PV <sub>DD</sub> = 1.8 V, I <sub>OL</sub> = 2 mA	<u>[3]</u>	0		0.3	V
I <sub>IH</sub>	High level input current	$V_{I} = DV_{DD}$		-1		1	μA
IIL	Low level input current	$V_{I} = 0 V$		-1		1	μΑ
I <sub>Leak</sub>	Input leakage current	$RSTPD_N = 0.4 V$		-1		1	μΑ
C <sub>in</sub>	Input Capacitance				2.5		pF
C <sub>out</sub>	Load Capacitance					30	pF
T <sub>SP</sub>	Width of suppressed spikes	Out of SPD mode			20		ns
T <sub>SP</sub>	Width of suppressed spikes	In SPD mode			120		ns

### Table 308. Input/open drain output pin characteristics for P50\_SCL for I<sup>2</sup>C interface

[1] To minimize power consumption when in Soft-Power-Down mode, the limit is  $PV_{DD}$  - 0.4 V.

[2] To minimize power consumption when in Soft-Power-Down mode, the limit is 0.4 V.

[3] Data at PVDD= 1.8V are only given from characterization results.

## 12.15 Input/output pin characteristics for MOSI / SDA / HSU\_TX

## Table 309. Input/output pin characteristics for MOSI / HSU\_TX for HSU and SPI Interfaces

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V <sub>IH</sub>	High level Input voltage		[1]	$0.7 \times PV_D$		$PV_{DD}$	V
V <sub>IL</sub>	Low level Input voltage		[2]	0		$0.3 \times PV_D$	V
V <sub>OH</sub>	HSU_TX high level output voltage	PV <sub>DD</sub> = 3 V, I <sub>OH</sub> = -4 mA		PV <sub>DD</sub> - 0.4		$PV_{DD}$	V
		PV <sub>DD</sub> =1.8 V, I <sub>OH</sub> = -2 mA	<u>[3]</u>	PV <sub>DD</sub> - 0.4		$PV_{DD}$	V
V <sub>OL</sub>	HSU_TX low level output voltage	PV <sub>DD</sub> = 3 V, I <sub>OL</sub> = 3 mA		0		0.3	V
IIH	MOSI high level input current	$V_{I} = DV_{DD}$		-1		1	μA
IIL	MOSI low level input current	V <sub>1</sub> = 0 V		-1		1	μA
I <sub>Leak</sub>	Input leakage current	$RSTPD_N = 0.4 V$		-1		1	μΑ
C <sub>in</sub>	Input Capacitance				2.5		pF
C <sub>out</sub>	Load Capacitance					30	pF
T <sub>SP</sub>	Width of suppressed spikes				20		ns

[1] To minimize power consumption when in Soft-Power-Down mode, the limit is PV<sub>DD</sub> - 0.4 V.

[2] To minimize power consumption when in Soft-Power-Down mode, the limit is 0.4 V.

[3] Data at PVDD= 1.8V are only given from characterization results.

### Table 310. Input/open drain output pin characteristics for SDA for I<sup>2</sup>C interface

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V <sub>IH</sub>	High level Input voltage		[1]	$0.7\times \text{PV}_{\text{DD}}$		$PV_{DD}$	V
V <sub>IL</sub>	Low level Input voltage		[2]	0		$0.3\times PV_{DD}$	V
V <sub>OL</sub>	Low level output voltage	PV <sub>DD</sub> = 3 V, I <sub>OL</sub> = -4 mA		0		0.3	V
		PV <sub>DD</sub> =1.8 V, I <sub>OL</sub> = -2 mA	[3]	0		0.3	V
I <sub>IH</sub>	High level input current	$V_{I} = DV_{DD}$		-1		1	μΑ
I <sub>IL</sub>	Low level input current	$V_{I} = 0 V$		-1		1	μΑ
I <sub>Leak</sub>	Input leakage current	$RSTPD_N = 0.4 V$		-1		1	μΑ
C <sub>in</sub>	Input Capacitance				2.5		pF
C <sub>out</sub>	Load Capacitance					30	pF
T <sub>SP</sub>	Width of suppressed spikes	Out of SPD mode			20		ns
		In SPD mode			150		ns

[1] To minimize power consumption when in Soft-Power-Down mode, the limit is  $PV_{DD}$  - 0.4 V.

[2] To minimize power consumption when in Soft-Power-Down mode, the limit is 0.4 V.

[3] Data at PVDD= 1.8V are only given from characterization results.

## 12.16 Input/output pin characteristics for MISO / P71 and SCK / P72

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V <sub>IH</sub>	High level Input voltage		[1]	$0.7\times PV_{DD}$		$PV_{DD}$	V
V <sub>IL</sub>	Low level Input voltage		[2]	0		$0.3\times PV_{DD}$	V
V <sub>OH</sub>	Push_pull / MISO mode high level output voltage	PV <sub>DD</sub> = 3 V, I <sub>OH</sub> = -4 mA		PV <sub>DD</sub> - 0.4		$PV_{DD}$	V
		PV <sub>DD</sub> =1.8 V, I <sub>OH</sub> = -2 mA	[3]	PV <sub>DD</sub> - 0.4		$PV_{DD}$	V
V <sub>OL</sub>	Push_pull / MISO mode low level output voltage	PV <sub>DD</sub> = 3 V, I <sub>OL</sub> = -4 mA		0		0.4	V
		PV <sub>DD</sub> =1.8 V, I <sub>OL</sub> = -2 mA	[3]	0		0.4	V
I <sub>IH</sub>	Input mode high level input current	$V_{I} = DV_{DD}$		-1		1	μA
I <sub>IL</sub>	Input mode low level input current	V <sub>1</sub> = 0 V		-1		1	μA
I <sub>Leak</sub>	Input leakage current	$RSTPD_N = 0.4 V$		-1		1	μA
C <sub>in</sub>	Input Capacitance				2.5		pF
Cout	Load Capacitance					30	pF
T <sub>SP</sub>	Width of suppressed spikes	Only valid for SCK			15		ns

### Table 311. Input/output pin characteristics for MISO / P71 and SCK / P72

[1] To minimize power consumption when in Soft-Power-Down mode, the limit is  $PV_{DD}$  - 0.4 V.

[2] To minimize power consumption when in Soft-Power-Down mode, the limit is 0.4 V.

[3] Data at PVDD= 1.8 V are only given from characterization results.

## 12.17 Input pin characteristics for SIGIN

## Table 312. Input/output pin characteristics for SIGIN

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V <sub>IH</sub>	High level Input voltage		[1]	$0.7\times SV_{DD}$		$SV_{DD}$	V
V <sub>IL</sub>	Low level Input voltage		[2]	0		$0.3\times SV_{DD}$	V
I <sub>IH</sub>	High level input current	$V_I = SV_{DD}$		-1		1	μA
I <sub>IL</sub>	Low level input current	$V_I = 0 V$		-1		1	μA
I <sub>Leak</sub>	Input leakage current	$RSTPD_N = 0.4 V$		-1		1	μA
C <sub>in</sub>	Input Capacitance				2.5		pF

[1] To minimize power consumption when in Soft-Power-Down mode, the limit is  $SV_{DD}$  - 0.4 V.

[2] To minimize power consumption when in Soft-Power-Down mode, the limit is 0.4 V.

## 12.18 Output pin characteristics for SIGOUT

### Table 313. Output pin characteristics for SIGOUT

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>OH</sub>	High level output voltage	$DV_{DD} - 0.1 < SV_{DD} < DV_{DD}$ $I_{OH} = -4 \text{ mA}$	SV <sub>DD</sub> - 0.4		$SV_{DD}$	V
V <sub>OL</sub>	Low level output voltage	$DV_{DD}$ - 0.1 < $SV_{DD}$ < $DV_{DD}$ $I_{OL}$ = +4 mA	0		0.4	V
I <sub>Leak</sub>	Input leakage current	$RSTPD_N = 0.4 V$	-1		1	μΑ
C <sub>in</sub>	Input Capacitance			2.5		pF
Cout	Load Capacitance				30	pF
t <sub>rise,fall</sub>	Rise and fall times	$\label{eq:VBAT} \begin{array}{l} VBAT = 3.4 \ V, \\ V_{OH} = 0.8 \times SV_{DD}, \\ V_{OL} = 0.2 \times SV_{DD}, \\ C_{out} = 30 \ pF \end{array}$		9		ns

## 12.19 Output pin characteristics for LOADMOD

## Table 314. Output pin characteristics for LOADMOD

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>OH</sub>	High level output voltage	VBAT = 3.4 V, I <sub>OH</sub> = -4 mA	DV <sub>DD -</sub> 0.4		$DV_DD$	V
V <sub>OL</sub>	Low level output voltage	VBAT = 3.4 V, I <sub>OL</sub> = 4 mA	0		0.4	V
C <sub>out</sub>	Load Capacitance				10	pF
t <sub>rise,fall</sub>	Rise and fall times	$\label{eq:VBAT} \begin{array}{l} VBAT = 3.4 \ V, \\ V_{OH} = 0.8 \times DV_{DD}, \\ V_{OL} \times = 0.2 \times DV_{DD}, \\ C_{out} = 10 \ pF \end{array}$		4.5		ns

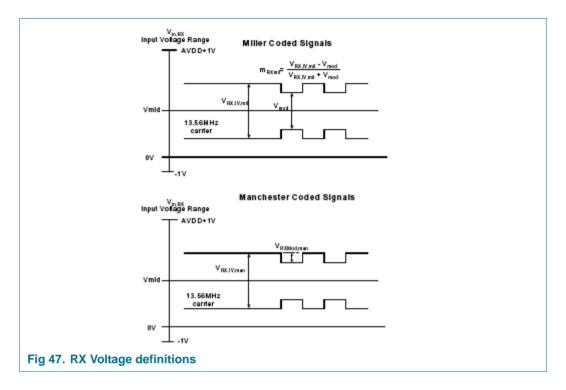
## 12.20 Input pin characteristics for RX

### Table 315. Input pin characteristics for RX

Parameter	Conditions	Min	Тур	Max	Unit
Dynamic Input voltage Range	VBAT = 3.4 V	-1		AV <sub>DD</sub> +1	V
RX Input Capacitance			10		pF
RX Input Series resistance	VBAT = 3.4 V, Receiver active, VRX = 1 Vpp, 1.5 V DC offset		350		Ω
Minimum Dynamic Input voltage, Miller coded	VBAT = 3.4 V, 106 kbit/s		150	500	mVpp
Minimum Dynamic Input voltage, Manchester Coded	VBAT = 3.4 V, 212 and 424 kbit/s		100	200	mVpp
Maximum Dynamic Input voltage, Miller coded	VBAT = 3.4 V, 106 kbit/s	AV <sub>DD</sub> +1			Vpp
Maximum Dynamic Input voltage, Manchester Coded	VBAT = 3.4 V, 212 and 424 kbit/s	AV <sub>DD</sub> +1			Vpp
Minimum Modulation index, Miller coded	VBAT = 3.4 V, 106 kbit/s VRX = 1.5 Vpp, SensMiller = 3		33		%
Minimum modulation voltage	VBAT = 3.4 V, RxGain = 6 and 7	<u>[1]</u>		6	mV
Minimum modulation voltage	VBAT = 3.4 V, RxGain = 4 and 5	[1]		18	mV
Minimum modulation voltage	VBAT = 3.4 V, RxGain = 0 to 3	<u>[1]</u>		120	mV
	Dynamic Input voltage RangeRX Input CapacitanceRX Input Series resistanceMinimum Dynamic Input voltage, Miller codedMinimum Dynamic Input voltage, Manchester CodedMaximum Dynamic Input voltage, Manchester CodedMaximum Dynamic Input voltage, Manchester CodedMinimum Dynamic Input voltage, Manchester CodedMinimum Dynamic Input voltage, Manchester CodedMinimum Modulation index, Miller codedMinimum modulation voltageMinimum modulation voltageMinimum modulation voltageMinimum modulation voltageMinimum modulation voltageMinimum modulation voltage	Dynamic Input voltage RangeVBAT = 3.4 VRX Input CapacitanceRX Input Series resistanceVBAT = 3.4 V, Receiver active, VRX = 1 Vpp, 1.5 V DC offsetMinimum Dynamic Input voltage, Miller codedVBAT = 3.4 V, 106 kbit/sMinimum Dynamic Input voltage, Manchester CodedVBAT = 3.4 V, 212 and 424 kbit/sMaximum Dynamic Input voltage, Manchester CodedVBAT = 3.4 V, 212 and 424 kbit/sMaximum Dynamic Input voltage, Manchester CodedVBAT = 3.4 V, 212 and 424 kbit/sMaximum Dynamic Input voltage, Manchester CodedVBAT = 3.4 V, 106 kbit/sMaximum Dynamic Input voltage, Manchester CodedVBAT = 3.4 V, 212 and 424 kbit/sMinimum Modulation index, Miller codedVBAT = 3.4 V, 106 kbit/sMinimum modulation voltageVBAT = 3.4 V, RxGain = 6 and 7Minimum modulation voltageVBAT = 3.4 V, RxGain = 4 and 5Minimum modulationVBAT = 3.4 V, RxGain = 4 and 5	Dynamic Input voltage RangeVBAT = $3.4 \text{ V}$ -1RX Input CapacitanceRX Input Series resistanceVBAT = $3.4 \text{ V}$ , Receiver active, VRX = $1 \text{ Vpp}$ , $1.5 \text{ V DC offset}-1Minimum Dynamic Inputvoltage, Miller codedVBAT = 3.4 \text{ V},106 \text{ kbit/s}-1Minimum Dynamic Inputvoltage, ManchesterCodedVBAT = 3.4 \text{ V},212 \text{ and } 424 \text{ kbit/s}-1Maximum Dynamic Inputvoltage, ManchesterCodedVBAT = 3.4 \text{ V},212 \text{ and } 424 \text{ kbit/s}AV_DD +1Maximum Dynamic Inputvoltage, ManchesterCodedVBAT = 3.4 \text{ V},212 \text{ and } 424 \text{ kbit/s}AV_DD +1Minimum Modulationindex,Miller codedVBAT = 3.4 \text{ V},212 \text{ and } 424 \text{ kbit/s}AV_DD +1Minimum modulationvoltageVBAT = 3.4 \text{ V},106 \text{ kbit/s}I1Minimum modulationvoltageVBAT = 3.4 \text{ V},11 \text{ RxGain = 6 and 7}I1Minimum modulationvoltageVBAT = 3.4 \text{ V},11 \text{ minimum modulation}I1VBAT = 3.4 \text{ V},11 \text{ minimum modulation}VBAT = 3.4 \text{ V},11 \text{ minimum modulation}I1Minimum modulationvoltageVBAT = 3.4 \text{ V},11 \text{ minimum modulation}I1Minimum modulationvoltageVBAT = 3.4 \text{ V},11 \text{ minimum modulation}I1VBAT = 3.4 \text{ V},11 \text{ minimum modulation}I1Minimum modulationVBAT = 3.4 \text{ V},11 \text{ minimum modulation}I1Minimum modulationVBAT = 3.4 \text{ V},11 \text{ minimum modulation}I1Minimum modulation<$	Dynamic Input voltage RangeVBAT = 3.4 V-1RX Input Capacitance10RX Input Series resistanceVBAT = 3.4 V, Receiver active, VRX = 1 Vpp, 1.5 V DC offset350Minimum Dynamic Input voltage, Miller codedVBAT = 3.4 V, 106 kbit/s150Minimum Dynamic Input voltage, Manchester CodedVBAT = 3.4 V, 212 and 424 kbit/s100Maximum Dynamic Input voltage, Miller codedVBAT = 3.4 V, 212 and 424 kbit/s100Maximum Dynamic Input voltage, Manchester CodedVBAT = 3.4 V, 212 and 424 kbit/sAV_DD +1Maximum Dynamic Input voltage, Manchester CodedVBAT = 3.4 V, 212 and 424 kbit/sAV_DD +1Minimum Modulation index, Miller codedVBAT = 3.4 V, 212 and 424 kbit/s33Minimum modulation voltageVBAT = 3.4 V, RxGain = 6 and 711Minimum modulation voltageVBAT = 3.4 V, RxGain = 4 and 511Minimum modulation voltageVBAT = 3.4 V, RxGain = 4 and 511	Dynamic Input voltage RangeVBAT = $3.4 \text{ V}$ -1 $AV_{DD}$ +1RX Input Capacitance10RX Input Series resistanceVBAT = $3.4 \text{ V}$ , Receiver active, VRX = $1 \text{ Vpp}$ , $1.5 \text{ V DC offset}350Minimum Dynamic Inputvoltage, Miller codedVBAT = 3.4 \text{ V},106 \text{ kbit/s}150500Minimum Dynamic Inputvoltage, ManchesterCodedVBAT = 3.4 \text{ V},212 \text{ and } 424 \text{ kbit/s}100200Maximum Dynamic Inputvoltage, ManchesterCodedVBAT = 3.4 \text{ V},212 \text{ and } 424 \text{ kbit/s}AV_{DD} +1-1Maximum Dynamic Inputvoltage, ManchesterCodedVBAT = 3.4 \text{ V},212 \text{ and } 424 \text{ kbit/s}AV_{DD} +1-1Maximum Dynamic Inputvoltage, ManchesterCodedVBAT = 3.4 \text{ V},212 \text{ and } 424 \text{ kbit/s}AV_{DD} +1-1Minimum Modulationindex,Miller codedVBAT = 3.4 \text{ V},106 \text{ kbit/s}AV_{DD} +1-1Minimum modulationvoltageVBAT = 3.4 \text{ V},106 \text{ kbit/s}33-1Minimum modulationvoltageVBAT = 3.4 \text{ V},RxGain = 6 \text{ and }71118Minimum modulationvoltageVBAT = 3.4 \text{ V},RxGain = 4 \text{ and }511120$

[1] The minimum modulation voltage is valid for all modulation schemes except Miller coded signals.

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## 12.21 Output pin characteristics for AUX1/AUX2

## Table 316. Output pin characteristics for AUX1/AUX2

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>OH</sub>	High level output voltage	VBAT = 3.4 V, I <sub>OH</sub> = -4 mA	DV <sub>DD</sub> - 0.4		$DV_DD$	V
V <sub>OL</sub>	Low level output voltage	VBAT = 3.4 V, I <sub>OL</sub> = 4 mA	$DV_{SS}$		DV <sub>SS</sub> +0.4	V
I <sub>Leak</sub>	Input leakage current	$RSTPD_N = 0 V$	-1		1	μΑ
Cin	Input Capacitance			2.5		pF
C <sub>out</sub>	Load Capacitance				15	pF

## 12.22 Output pin characteristics for TX1/TX2

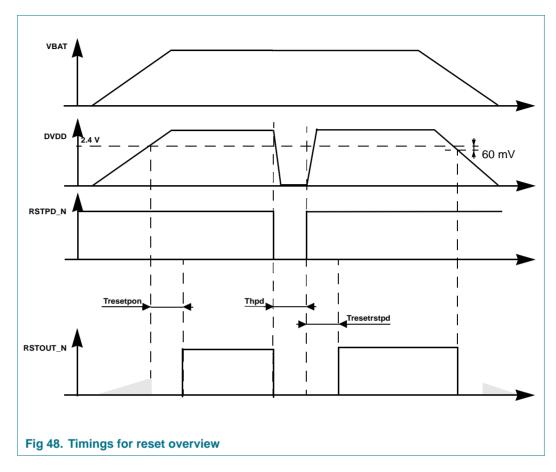
### Table 317. Output pin characteristics for TX1/TX2

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>OH</sub> , C32, 3 V	High level output voltage	VBAT = 3.4 V, I <sub>TX1/2</sub> = 32 mA, CWGsP[5:0] = 3Fh	TV <sub>DD</sub> - 150			mV
V <sub>OH</sub> , C80, 3 V	High level output voltage	VBAT =3.4 V, I <sub>TX1/2</sub> = 80 mA, CWGsP[5:0] = 3Fh	TV <sub>DD</sub> - 400			mV
V <sub>OL</sub> , C32, 3 V	Low level output voltage	VBAT = 3.4 V, $I_{TX1/2} = 32 mA,$ CWGsN[3:0] = Fh			150	mV
V <sub>OL</sub> , C80, 3 V	Low level output voltage	VBAT = 3.4 V, I <sub>TX1/2</sub> = 80 mA, CWGsN[3:0] = Fh			400	mV

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## 12.23 Timing for Reset and Hard-Power-Down

## Table 318. Reset duration time

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
T <sub>resetpon</sub>	Reset time at power on		[1]	0.1	0.4	2	ms
T <sub>HPD</sub>	Hard Power-Down time	User dependent		20			ns
T <sub>resetRSTPD_N</sub>	Reset time when RSTPD_N is released		<u>[1]</u>	0.1	0.4	2	ms

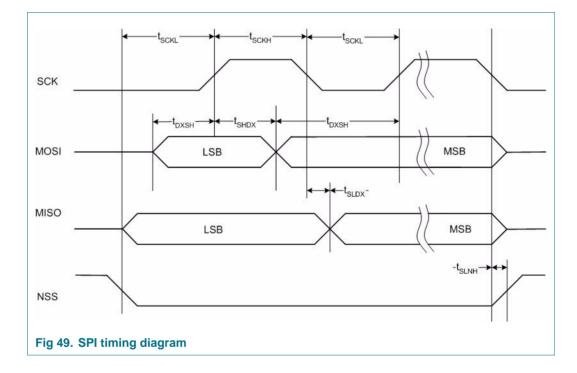
 27.12 MHz quartz starts in less than 800 μs. For example, quartz like TAS-3225A, TAS-7 or KSS2F with appropriate layout.

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## 12.24 Timing for the SPI compatible interface

## Table 319. SPI timing specification

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
t <sub>SCKL</sub>	SCK low pulse width		50			ns
t <sub>SCKH</sub>	SCK high pulse width		50			ns
t <sub>SHDX</sub>	SCK high to data changes		25			ns
t <sub>DXSH</sub>	data changes to SCK high		25			ns
t <sub>SLDX</sub>	SCK low to data changes				25	ns
t <sub>SLNH</sub>	SCK low to NSS high		0			ns



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## 12.25 Timing for the I<sup>2</sup>C interface

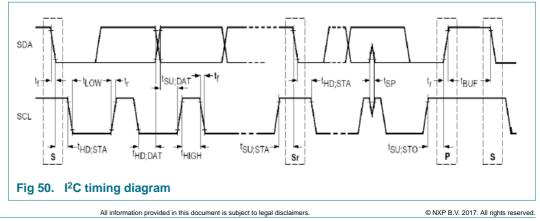
### Table 320. I<sup>2</sup>C timing specification

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
f <sub>SCL</sub>	SCL clock frequency			0		400	kHz
t <sub>HD; STA</sub>	Hold time (repeated) START condition.	After this period, the first clock pulse is generated		600			ns
$t_{SU; STA}$	Set-up time for a repeated START condition			600			ns
t <sub>SU; STO</sub>	Set-up time for STOP condition			600			ns
t <sub>LOW</sub>	LOW period of the P50_SCL clock			1300			ns
t <sub>HIGH</sub>	HIGH period of the P50_SCL clock			600			ns
t <sub>HD; DAT</sub>	Data hold time			0		900	ns
$t_{SU; DAT}$	Data set-up time			100			ns
t <sub>r</sub>	Rise time P50_SCL and SDA		<u>[1]</u>	20		1000	ns
t <sub>f</sub>	Fall time P50_SCL and SDA		<u>[1]</u>	20		300	ns
t <sub>BUF</sub>	Bus free time between a STOP and START condition			1.3			ms
t <sub>StrWuSpd</sub>	Stretching time on P50_SCL when woken-up on its own address		[2]			1	ms
t <sub>HDSDA</sub>	Internal hold time for SDA			330		590	ns
t <sub>HDSDA</sub>	Internal hold time for SDA in SPD mode		<u>[3]</u>		270		ns

[1] The PN532 has a slope control according to the I<sup>2</sup>C specification for the Fast mode. The slope control is always present and not dependant of the I<sup>2</sup>C speed.

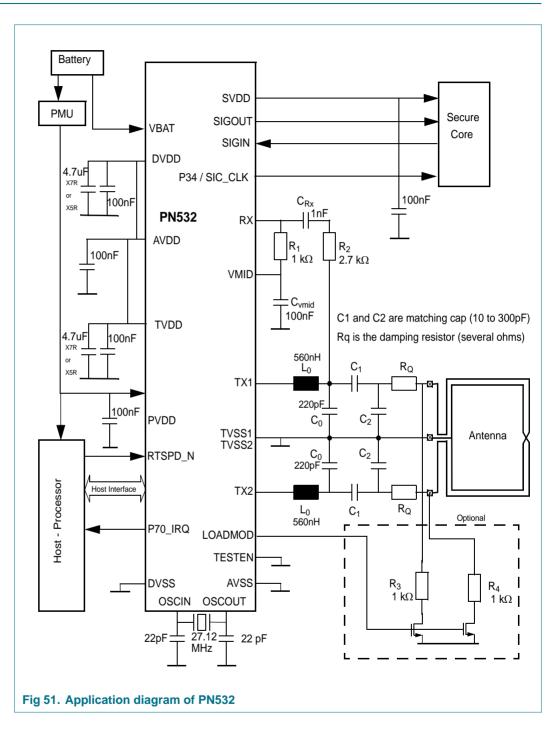
[2] 27.12 MHz quartz starts in less than 800 μs. For example, quartz like TAS-3225A, TAS-7 or KSS2F with appropriate layout.

[3] The PN532 has an internal hold time of around 270ns for the SDA signal to bridge the undefined region of the falling edge of P50\_SCL.



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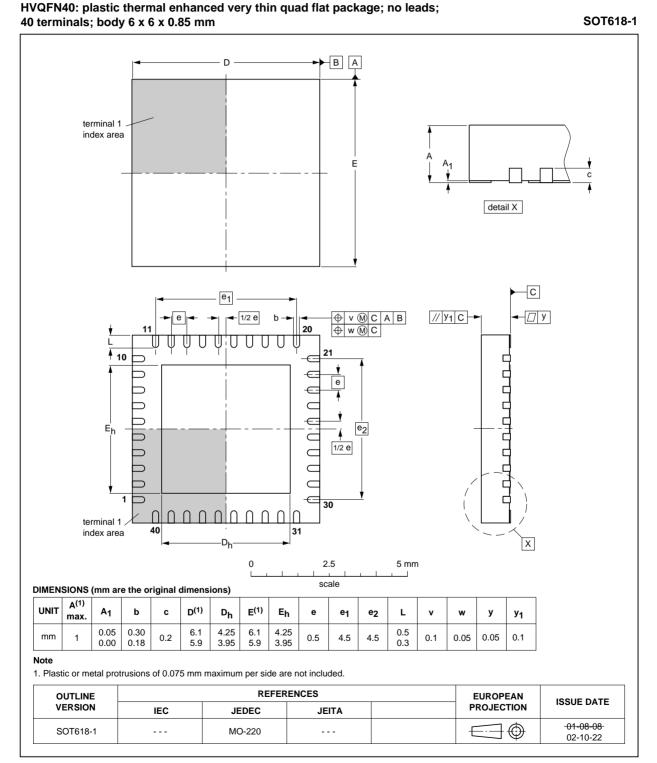
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## 14. Package outline



## Fig 52. Package outline HVQFN40 (SOT618-1)

This package is MSL level 2.

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## **15. Abbreviations**

Acronym	Description
ASK	Amplitude Shift keying
BPSK	Bit Phase Shift Keying
CIU	Contactless Interface Unit
CRC	Cyclic Redundancy Check
ECMA	European Computer Manufacturers Association organization
GPIO	General Purpose Input Output
GPIRQ	General Purpose Interrupt ReQuest
HPD	Hard Power Down (see Section 8.5.3 on page 91)
HSU	High Speed UART
Initiator	Generates RF field at 13.56 MHz and starts the NFCIP-1 communication.
LDO	Low Drop-Out regulator
Load modulation Index	The load modulation index is defined as the card's voltage ratio (Vmax - Vmin) / (Vmax + Vmin) measured at the card's coil.
MIF	Multi InterFace block
Modulation Index	The modulation index is defined as the voltage ratio (Vmax - Vmin) / (Vmax + Vmin).
MSL	Moisture Sensitivity Level
NFCIP	NFC Interface and Protocol
NFC-WI	NFC Wired Interface to connect NFC front end to a SIC
PCD	Proximity Coupling Device. Definition for a Card Reader/ Writer according to the ISO/IEC 14443 Specification
PCR	Power Clock Reset controller
PICC	Proximity Cards. Definition for a contactless Smart Card according to the ISO/IEC 14443 specification
SIC	Secure Integrated Circuit (can be a Smart Card IC, a Secure Access Module (SAM),)
SPI	Serial Parallel Interface
SPD	Soft Power Down mode (see Section 8.5.4 on page 92)
Target	Responds to initiator command either using load modulation scheme (RF field generated by Initiator) or using modulation of self-generated RF field (no RF field generated by initiator during target answer).
UART	Universal Asynchronous Receive Transmit

## **16. Revision history**

Document ID	on history Release date	Data sheet status	Change notice	Supersedes
PN532_C1 v. 3.6	20171128	Product data sheet	-	PN532_C1 v. 3.5
Modifications:			Public, no content change	11032_01 0.3.0
PN532_C1 v. 3.5	20120920	Product data sheet	Tublic, no content change	PN532_115434
			- -	FIN002_110404
Modifications:		dering information": updated	eu	
PN532_115434	20111201	Product data sheet	-	PN532_115433
Modifications:	Section 5 "Ore	dering information": update	ed	
PN532_115433	20111110	Product data sheet	-	115432
Modifications:	NXP Semicor	nductors.	redesigned to comply with the new ider	ntity guidelines of
	<ul> <li>Table 2 "Orde</li> </ul>	ring information": Table no	te [2] updated.	
	<ul> <li>Section 17 "Le</li> </ul>	egal information": updated		
115432	20071203	Product data sheet		Revision 3.1
Modifications:	Apart from typo co	prrections:		
	<ul> <li>General rewo</li> </ul>	rding of Mifare designation	n (e.g. "MIFARE")	
	removed. It is 18.1 "Tradema	replaced with the Table 2 arks" on page 216.	re for appropriate third party patent lice "Ordering information" on page 5 notes 5, notes have been modified.	
		description in Table 3 on p		
	-		I mode added in Section 8.4 on page 8	3
		page 96 Table note 2 has		
			a added in Section 8.6.13.1 on page 12	3
	<ul> <li>VIH of RSTPE page 198</li> </ul>	D_N when PVDD<0.4V ali	gned between Section 8.4 on page 83 a	and Section 12.7 o
	• Table 156 on	page 115 for RF level dete	ector has been modified	
	<ul> <li>Functional con page 197.</li> </ul>	nditions have been added	in Section 8.6.9 on page 116 and in Se	ection 12.4 on
	<ul> <li>Table 158 on aligned</li> </ul>	page 117, Table 296 on pa	ge 197 and Table 297 on page 197 hav	e been modified an
	<ul> <li>2 remarks have</li> </ul>	ve been added in Section	8.6.13.1 on page 123	
	<ul> <li>Table 293 on</li> </ul>	page 195 modified the VB	AT condition for VDVDD to 3.4V instea	d of 3.7V.
	<ul> <li>Table 307 on</li> </ul>	page 203 has been modifi	ed and a note has been added.	
	Table 315 on	page 207 for RX characte	ristics has been modified.	
115431	3 April 2007	Product data sheet		Revision 3.0
	17 October 2006	Product data sheet	complete review of the specification	Revision 2.3
115430			•	1(60131011 2.5
115430 115423	17 August 2006	Preliminary data sheet	update for rev1.3 after spec review	Revision 2.2

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Document ID	Release date	Data sheet status	Change notice	Supersedes
115421	2 June 2006	Preliminary data sheet		Revision 2.0
115420	11 May 2006	Preliminary data sheet		Revision 1.2
115412	21 December 2005	Objective data sheet		Revision 1.1
115411	24 October 2005	Objective data sheet		Revision 1.0
115410	17 October 2005	Objective data sheet	Initial version	-

### Table 322. Revision history ... continued

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## **17. Legal information**

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Document status[1][2]	Product status <sup>[3]</sup>	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
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[2] The term 'short data sheet' is explained in section "Definitions"

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