



Automotive Dual Remote Antenna Current-Sense LDO/Switch

Features

MAX16948

Automotive Dual Remote Antenna Current-Sense LDO/Switch

ABSOLUTE MAXIMUM RATINGS

(All voltages referenced to GND.)

IN, SHDN1, SHDN2	-0.3V to +45V
ERR1, ERR2, REG	-0.3V to +6.0V
FB1, FB2, SENSE1, SENSE2, LIM1, LIM2	-0.3V to ($V_{REG} + 0.3V$)
OUT1, OUT2	-0.3V to +20V
Continuous Power Dissipation ($T_A = +70^\circ C$)	
TQFN-EP (derate 25.0mW/°C above +70°C)	2000mW
QSOP-EP (derate 22.7mW/°C above +70°C)	1818mW
QSOP (derate 9.5mW/°C above +70°C)	762mW

Operating Temperature Range	-40°C to +105°C
Storage Temperature Range	-65°C to +150°C
Junction Temperature	+150°C
Lead Temperature (soldering, 10s)	+300°C
Soldering Temperature (reflow)	+260°C

PACKAGE THERMAL CHARACTERISTICS (Note 1)

TQFN-EP	16 QSOP
Junction-to-Ambient Thermal Resistance (θ_{JA})40°C/W	Junction-to-Ambient Thermal Resistance (θ_{JA}) 105°C/W
Junction-to-Case Thermal Resistance (θ_{JC})6°C/W	Junction-to-Case Thermal Resistance (θ_{JC})37°C/W
16 QSOP-EP	
Junction-to-Ambient Thermal Resistance (θ_{JA})44°C/W	
Junction-to-Case Thermal Resistance (θ_{JC})6°C/W	

Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to www.maximintegrated.com/thermal-tutorial.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

($V_{IN} = 14V$, $T_A = T_J = -40^\circ C$ to $+105^\circ C$, unless otherwise noted. Typical values are at $T_A = T_J = +25^\circ C$.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
POWER SWITCH/LDO						
IN Operating Supply Range	V_{IN}	Over 19V, output and current sense is switched off (Note 3)	4.5		28	V
		Output switched off, for < 1s (Note 3)			45	
IN Supply Current in Operation	I_{IN}	$V_{SHDN1}/V_{SHDN2} > 2.8V$, $T_A = +25^\circ C$		2.1	3.4	mA
IN Supply Current in Shutdown	I_{SD}	$\overline{SHDN1} = \overline{SHDN2} = GND$, $T_A = +25^\circ C$, $V_{IN} = 12V$			7	μA
IN Undervoltage Lockout	V_{UVL}	V_{IN} rising			4.8	V
		V_{IN} falling		3.8	4.45	
IN Undervoltage Lockout Hysteresis	$V_{UVL-HYS}$			390		mV
Voltage Drop of Internal Switch	$V_{IN} - V_{OUT1/OUT2}$	Measured between IN and OUT1/OUT2, $I_{OUT1/OUT2} = 100mA$, FB1/FB2 = GND, SW operation $V_{IN} \geq 5V$, $T_A = +105^\circ C$ (Note 4)			0.3V	V
		Measured between IN and OUT1/OUT2, $I_{OUT1/OUT2} = 100mA$, FB1/FB2 = GND, SW operation $V_{IN} \geq 5V$, $T_A = +25^\circ C$		0.145	0.220	

MAX16948

Automotive Dual Remote Antenna Current-Sense LDO/Switch

ELECTRICAL CHARACTERISTICS (continued)

($V_{IN} = 14V$, $T_A = T_J = -40^{\circ}C$ to $+105^{\circ}C$, unless otherwise noted. Typical values are at $T_A = T_J = +25^{\circ}C$.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Feedback Voltage	V _{FB}	For LDO mode with external resistive dividers, I _{OUT1/OUT2_} = 5mA to 150mA		0.97	1	1.03	V
Input Bias Current to FB1 and FB2	I _{FB1/FB2}	V _{FB1/FB2} = 1.0V, LDO mode, T _A = +25°C		-0.5		+0.5	μA
Feedback Threshold for Internal Feedback Resistive Divider	V _{FB1/FB2,TH}	Switching to LDO mode with internal 8.5V resistive dividers		V _{REG} - 1.7		V _{REG} - 0.8	V
Adjustable Output Voltage Range	V _{OUT1/OUT2}	LDO mode with external resistive dividers (Notes 3, 5)		1		12.5	V
FB1 and FB2 Load Regulation	ΔV _{FB, LOAD}	V _{IN} - V _{OUT1/OUT2} ≥ 2V, I _{OUT1/OUT2} = 5mA to 250mA, LDO mode (Note 3)			-7		mV
FB1 and FB2 Line Regulation	ΔV _{FB, LINE}	V _{IN} - V _{OUT1/OUT2} ≥ 2V, I _{OUT1/OUT2} = 50mA, LDO mode (Note 3)			100		μV/V
Fixed Output Voltage 8.5V	V _{OUT1/OUT2, 8.5V}	I _{OUT1/OUT2} = 70mA, LDO mode with internal resistive divider, 9V < V _{IN} < 18V		8.33	8.5	8.67	V
Power-Supply Rejection Ratio	PSRR	V _{IN} - V _{OUT1/OUT2} ≥ 2V, I _{OUT1/OUT2} = 10mA at f = 100Hz, LDO mode V _{OUT1/OUT2} = 8.5V, V _{RIPPLE} = 0.5V _{P-P} (Note 3)			73		dB
CURRENT SENSE							
SENSE1 to OUT1 and SENSE2 to OUT2 Current Ratio	A _I	V _{SENSE1/SENSE2} < 3V	I _{OUT1/OUT2} = 100mA, V _{IN} = 5V, 19V	0.485	0.5	0.515	%
			50mA ≤ I _{OUT1/OUT2} ≤ 300mA, V _{IN} ≥ 5V (Note 3)	0.45	0.5	0.55	
			20mA ≤ I _{OUT1/OUT2} ≤ 50mA, V _{IN} ≥ 5V (Note 3)	0.415	0.5	0.585	
			5mA ≤ I _{OUT1/OUT2} ≤ 20mA, V _{IN} ≥ 5V (Note 3)	0.325	0.5	0.675	
			0mA ≤ I _{OUT1/OUT2} ≤ 5mA, V _{IN} ≥ 5V (Notes 3, 6)	0.3	0.5	0.7	
No-Load Current-Sense Offset	I _{CSO}	I _{OUT1} = I _{OUT2} = 0mA (Note 3)			2	10	μA
SENSE1, SENSE2, LIM1, LIM2 Leakage Current	I _{LEAK}	SHDN1 = SHDN2 = GND, T _A = + 25°C				2	μA

MAX16948

Automotive Dual Remote Antenna Current-Sense LDO/Switch

ELECTRICAL CHARACTERISTICS (continued)

($V_{IN} = 14V$, $T_A = T_J = -40^{\circ}C$ to $+105^{\circ}C$, unless otherwise noted. Typical values are at $T_A = T_J = +25^{\circ}C$.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
FAILURE DETECTION						
LIM1 and LIM2 Current-Limit Threshold Voltage	$V_{LIM1/LIM2,TH}$	Voltage on LIM1 and LIM2 for which output current is limited	2.375	2.5	2.625	V
IN Overvoltage Lockout Threshold (Rising)	$V_{OVLO,R}$	V_{IN} rising	19	21	23	V
IN Overvoltage Lockout Threshold (Falling)	$V_{OVLO,F}$	V_{IN} falling	18.5			V
IN Overvoltage Lockout Hysteresis	$V_{OVLO,H}$			600		mV
Short-to-BAT Threshold in Off State	V_{S2B}	$V_{OUT1/OUT2} - V_{IN}$, checked during turn-on sequence	-500	-80	+110	mV
Reverse Current Detection Level	I_{REVCUR}	Power FET on (SW or LDO mode)	-100	-50	-1	mA
Reverse Current (Short-to-BAT) Shutdown Delay	t_{PD-RC}	Delay to shutdown the switch or LDO after drop over R_{ON} becomes negative, $I_{OUT1/OUT2} = -200mA$ (typ), $T_A = +25^{\circ}C$		4.16	20	μs
Reverse Current Blanking Time	t_{BLK-RC}	Blanking time for reverse current and FB1/FB2 out of range after rising edge of $\overline{SHDN1}$, $\overline{SHDN2}$, or current limiting event is over		16		ms
Feedback Voltage Out of Range	V_{FBERR}	LDO mode, internal pulldown resistor switched on, FB rising	1.12	1.15	1.28	V
Feedback Voltage Out-of-Range Hysteresis	$V_{FBERR-HYS}$	LDO mode, internal pulldown resistor switched on (Note 3)		39	100	mV
OVERTEMPERATURE PROTECTION (Note 3)						
Thermal Shutdown	T_{SHDN}			160		$^{\circ}C$
Thermal Shutdown Hysteresis	$T_{SHDN-HYS}$			15		$^{\circ}C$
SUPPORTING CIRCUITS, INTERFACE SIGNALS						
Internal Voltage Regulator	V_{REG}	$T_A = +25^{\circ}C$, $I_{REG} = 0mA$, $V_{IN} = 4.5V$ to $18V$	4	5	5.5	V
		$T_A = +25^{\circ}C$, $I_{REG} = 0mA$, $V_{IN} = 6.5V$ to $18V$	4.5	5	5.5	V
Internal Voltage Regulator Current Limit	I_{REG}	(Note 4)	15			mA
$\overline{ERR1}$ and $\overline{ERR2}$ Output-Voltage Low	V_{OL}	Sinking current = 10mA			0.4	V
$\overline{ERR1}$ and $\overline{ERR2}$ Open-Drain Leakage Current		$\overline{ERR1}$ and $\overline{ERR2}$ not asserted, $V_{ERR1/ERR2} = 5V$, $T_A = +25^{\circ}C$			1	μA
$\overline{SHDN1}$ and $\overline{SHDN2}$ High	$V_{\overline{SHDN1}/\overline{SHDN2},HI}$		2.8			V
$\overline{SHDN1}$ and $\overline{SHDN2}$ Low	$V_{\overline{SHDN1}/\overline{SHDN2},LO}$				0.8	V

MAX16948

Automotive Dual Remote Antenna Current-Sense LDO/Switch

ELECTRICAL CHARACTERISTICS (continued)

($V_{IN} = 14V$, $T_A = T_J = -40^{\circ}C$ to $+105^{\circ}C$, unless otherwise noted. Typical values are at $T_A = T_J = +25^{\circ}C$.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
SHDN1 and SHDN2 Pulldown Resistance	R_{SHDN}	$V_{SHDN1/SHDN2} = 2V$		500		$k\Omega$
Startup Response Time	t_{ST}	SHDN1 and SHDN2 rising to turn on LDO/switch (Note 3)		300		μs
OUT Pulldown Resistor	$R_{OUT1/OUT2-OFF}$	SHDN1 = SHDN2 = GND		56		$k\Omega$

Note 2: Devices are tested at $T_A = +25^{\circ}C$ and guaranteed by design for $T_A = T_{MIN}$ to T_{MAX} .

Note 3: Guaranteed by design; not production tested.

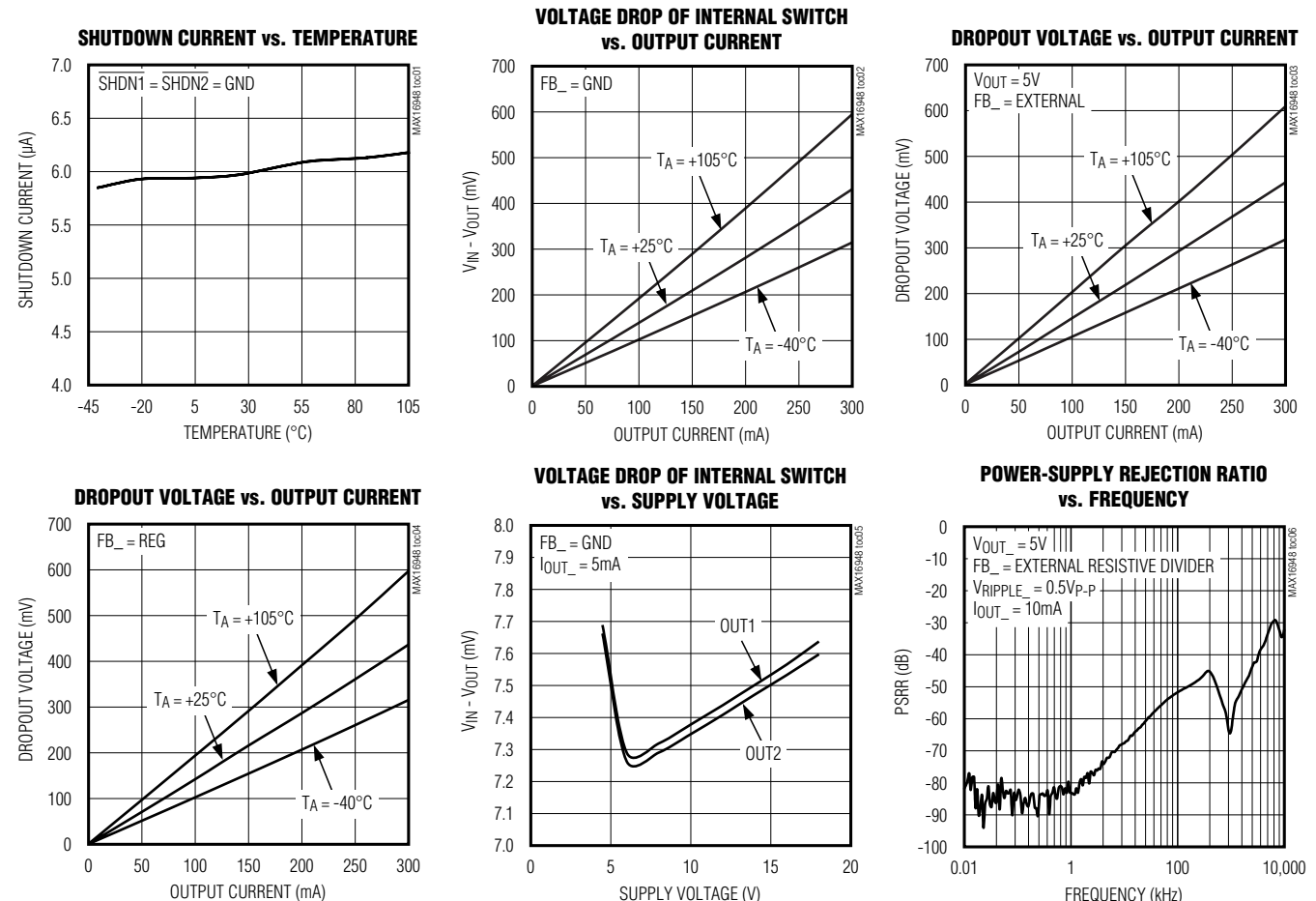
Note 4: Guaranteed by ATE characterization. Limits are not production tested.

Note 5: In case of OUT1/OUT2 shorted to BAT, feedback network must protect FB1/FB2 from violating their absolute maximum ratings. For OUT1/OUT2 set below 3.3V, use an additional 5V clamp on FB1/FB2 with low parasitic capacitance.

Note 6: Limits do not include current-sense offset.

Typical Operating Characteristics

($V_{IN} = 14V$, $T_A = +25^{\circ}C$, see the [Typical Operating Circuits](#), unless otherwise noted.)

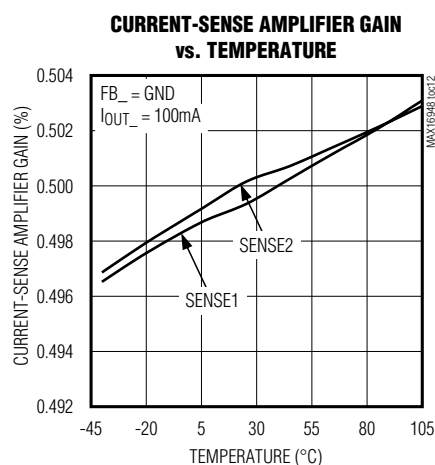
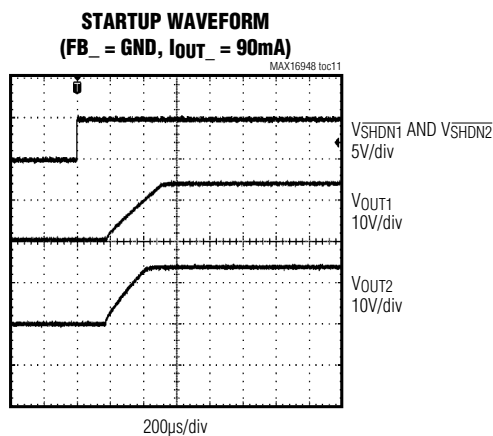
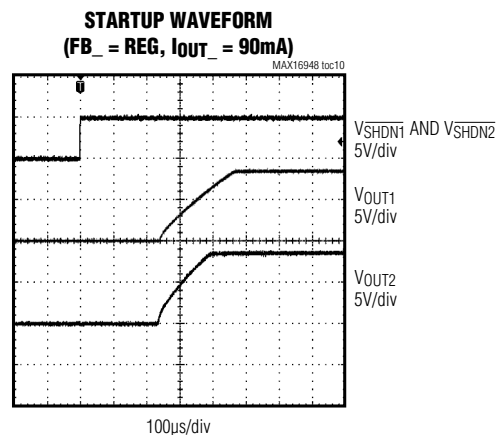
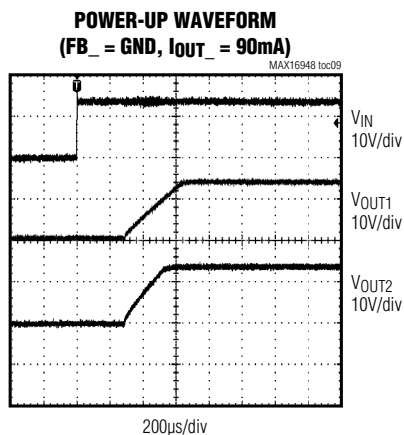
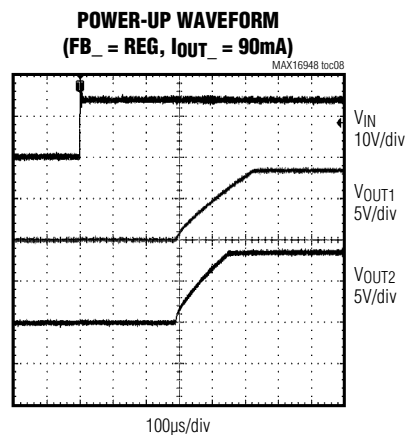
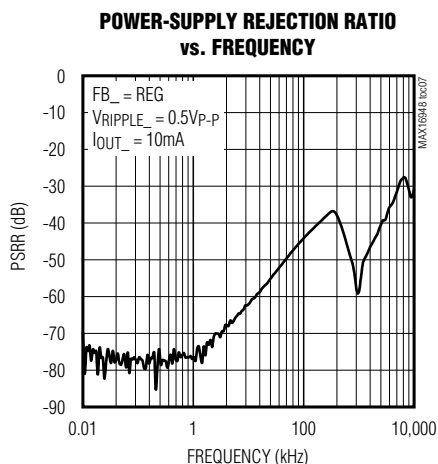


MAX16948

Automotive Dual Remote Antenna Current-Sense LDO/Switch

Typical Operating Characteristics (continued)

($V_{IN} = 14V$, $T_A = +25^\circ C$, see the [Typical Operating Circuits](#), unless otherwise noted.)

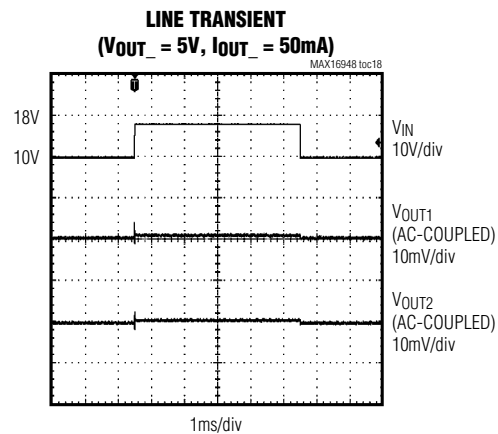
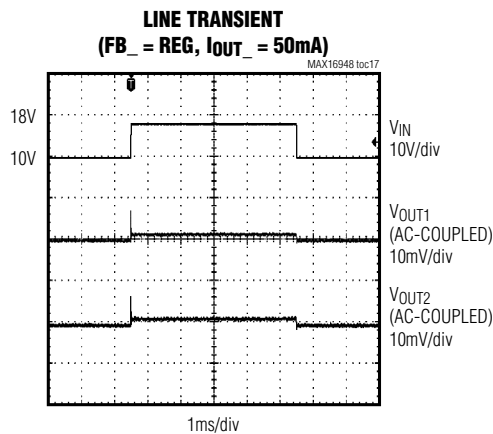
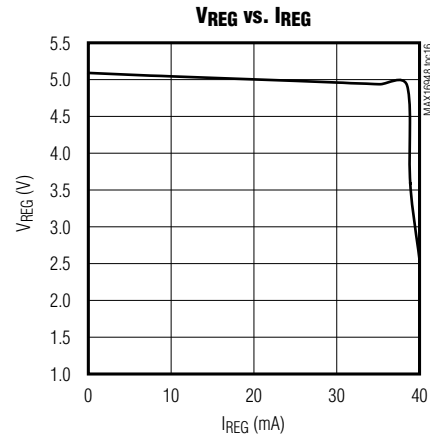
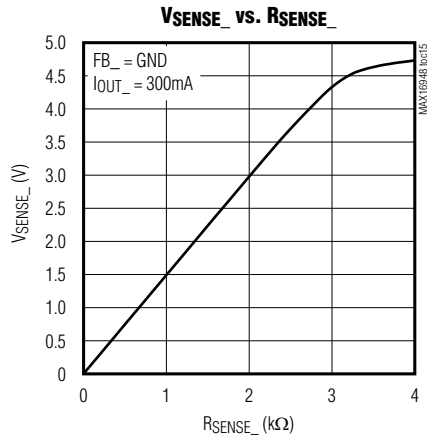
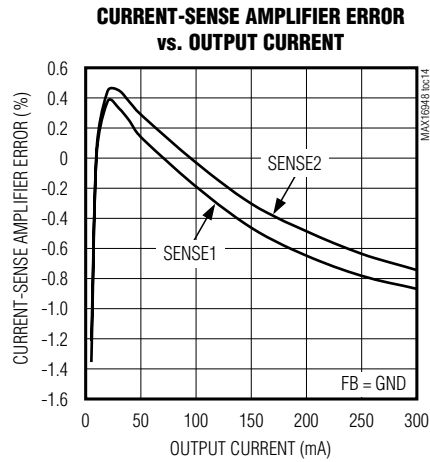
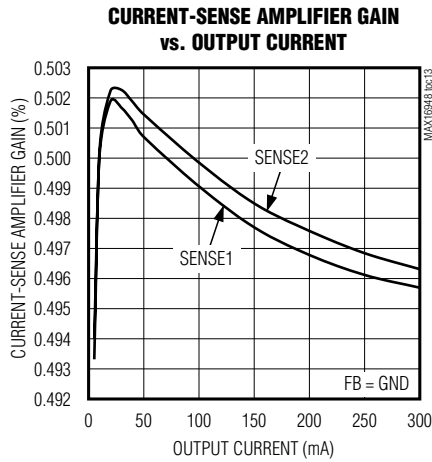


MAX16948

Automotive Dual Remote Antenna Current-Sense LDO/Switch

Typical Operating Characteristics (continued)

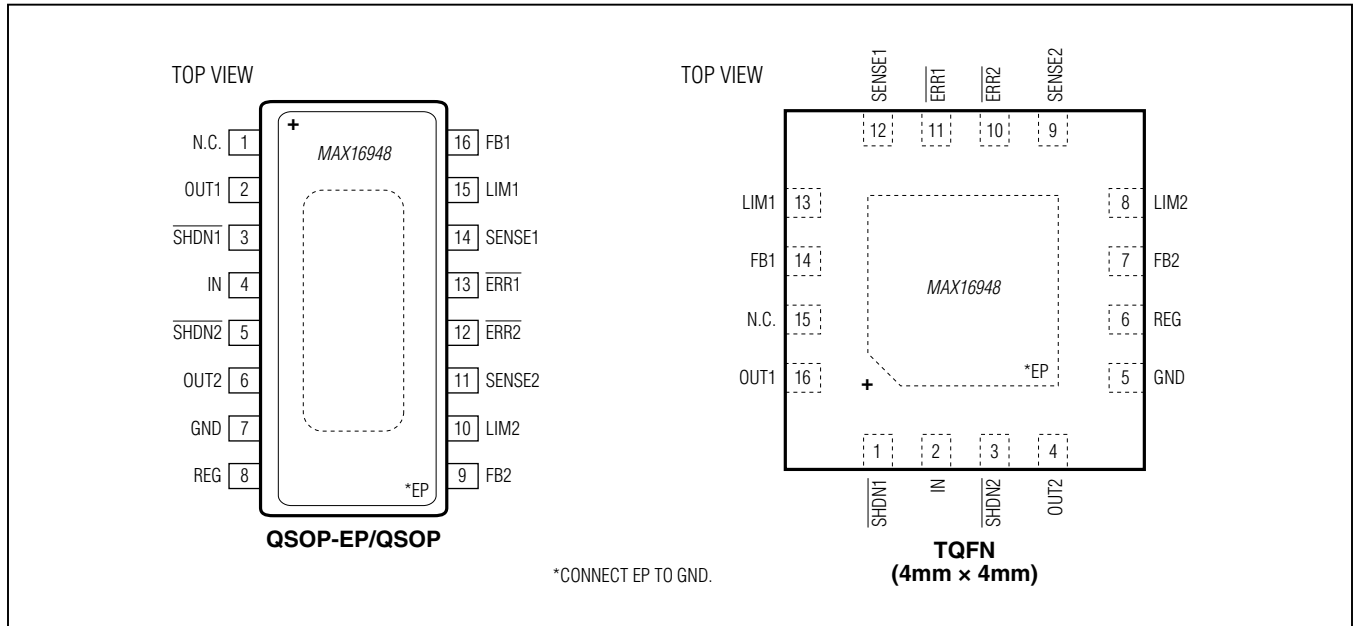
($V_{IN} = 14V$, $T_A = +25^\circ C$, see the [Typical Operating Circuits](#), unless otherwise noted.)



MAX16948

Automotive Dual Remote Antenna Current-Sense LDO/Switch

Pin Configurations



Pin Descriptions

PIN		NAME	FUNCTION
QSOP/ QSOP-EP	TQFN-EP		
1	15	N.C.	No Connection. Not internally connected.
2	16	OUT1	LDO/Switch Output 1 to Remote Load. Bypass OUT1 to GND with >1μF in parallel with 100nF low-ESR ceramic capacitors for regulator stability.
3	1	SHDN1	Active-Low Shutdown Input for OUT1. SHDN1 is pulled to GND with an internal 500kΩ resistor.
4	2	IN	Input Voltage. Bypass IN to GND with an electrolytic capacitor with a minimum value of 10μF and a low-ESR ceramic capacitor with a minimum value of 0.1μF.
5	3	SHDN2	Active-Low Shutdown Input for OUT2. SHDN2 is pulled to GND with an internal 500kΩ resistor.
6	4	OUT2	LDO/Switch Output 2 to remote load. Bypass OUT2 to GND with >1μF in parallel with 100nF low-ESR ceramic capacitors for regulator stability.
7	5	GND	Ground
8	6	REG	Internal 5V Regulator. Provides supply for internal low-voltage blocks, SENSE_, and LIM_ outputs.
9	7	FB2	Feedback Input for Setting the OUT2 Voltage. Connect FB2 to GND to select current-limited switch operation. Connect to an external resistive divider for adjustable output-voltage operation. Connect FB2 to REG to choose the internal resistive divider for the 8.5V regulator option.

MAX16948

Automotive Dual Remote Antenna Current-Sense LDO/Switch

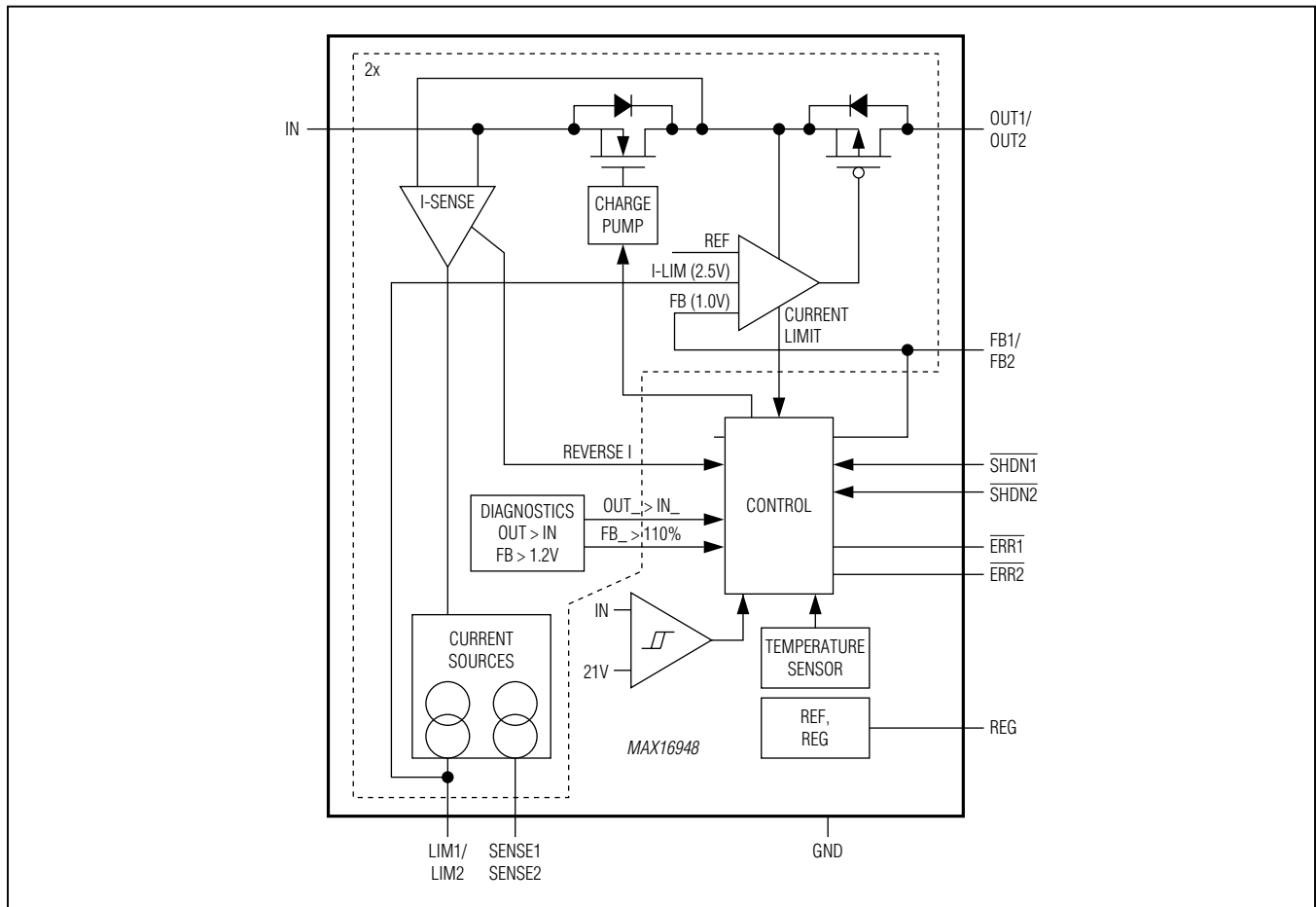
Pin Descriptions (continued)

PIN		NAME	FUNCTION
QSOP/ QSOP-EP	TQFN-EP		
10	8	LIM2	Output of Current Sense for Limiting. This output is proportional to current flowing through OUT2. The internal current-limit amplifier is tripped when the voltage on LIM2 reaches 2.5V. Connect a resistor to GND to set the level for current limitation and a 0.1μF capacitor in parallel for frequency compensation of the current-limit loop.
11	9	SENSE2	Output of Current Sense for Sensing. The current of SENSE2 is proportional to the current flowing through OUT2. Connect a resistor from SENSE2 to GND to set the output voltage level. Additionally, connect a 0.1μF capacitor from SENSE2 to GND.
12	10	$\overline{\text{ERR2}}$	Open-Drain Fault Indicator Output 2
13	11	$\overline{\text{ERR1}}$	Open-Drain Fault Indicator Output 1
14	12	SENSE1	Output of Current Sense for Sensing. The current of SENSE1 is proportional to the current flowing through OUT1. Connect a resistor from SENSE1 to GND to set the output voltage level. Additionally, connect a 0.1μF capacitor from SENSE1 to GND.
15	13	LIM1	Output of Current Sense for Limiting. This output is proportional to current flowing through OUT1. The internal current-limit amplifier is tripped when the voltage on LIM1 reaches 2.5V. Connect a resistor to GND to set the level for current limitation and a 0.1μF capacitor in parallel for frequency compensation of the current-limit loop.
16	14	FB1	Feedback Input for Setting the OUT1 Voltage. Connect FB1 to GND to select current-limited switch operation. Connect to an external resistive divider for adjustable output-voltage operation. Connect FB1 to REG to choose the internal resistive divider for the 8.5V regulator option.
—	—	EP	Exposed Pad (MAX16948A only). Connect EP to the ground plane for optimal heat dissipation. Do not use EP as the primary electrical ground connection.

MAX16948

Automotive Dual Remote Antenna Current-Sense LDO/Switch

Functional Diagram



Detailed Description

The MAX16948 is a dual high-voltage, current-sensing LDO/switch designed to operate with an input voltage range from 4.5V to 28V (45V load dump tolerant). The device provides phantom power over coax cable to remote radio frequency low-noise amplifiers (LNAs) in automotive systems with a maximum current of 300mA per channel. The device also provides a fixed regulated output voltage of 8.5V, an adjustable 1V to 12.5V output voltage, or the device can be configured as a switch. The device is ideal for providing phantom power to remote radio-frequency LNAs in automotive applications.

The device monitors the load current, and its current-sense outputs provide a current proportional to the sensed load current. An accurate programmable current limit protects the input supply against both overcurrent and short-circuit conditions. The device features open-drain fault indicator outputs and overtemperature shutdown.

The device features short-to-battery protection to latch off the internal LDO/switch during a short-to-battery event. During a thermal overload, the device reduces power dissipation by going into thermal shutdown. It includes dual independent, active-low, high-voltage-compatible shutdown inputs to place each channel in a low-power shutdown mode.

MAX16948

Automotive Dual Remote Antenna Current-Sense LDO/Switch

Fault Detection

The device monitors the load current through an internal sense resistor to protect against short-circuit, short-to-battery, and reverse current faults. In addition, the device also detects input and output overvoltage conditions and features thermal shutdown.

Short Circuit and Overcurrent

The current limit of each channel is programmed with an external resistor connected to the output of the current sense amplifiers LIM_ to protect the device during short-circuit or overcurrent conditions. When the voltage at LIM_ reaches the internal 2.5V threshold, the output current of the LDO/switch is limited, and $\overline{\text{ERR}}$ asserts low but the output is not disabled. The $\overline{\text{ERR}}$ and LIM_ outputs should be monitored by the μC , and the channel experiencing the short-circuit or overcurrent condition should be disabled by pulling $\overline{\text{SHDN}}$ low. If this condition persists, thermal shutdown could occur and both outputs could be disabled.

Short-to-Battery and Reverse Current Detection

It is possible for OUT_ to be shorted to the battery due to a fault in the system. Each channel detects this failure by comparing the voltage at OUT_ and IN before the switch turns on. Every time the LDO/switch is enabled on the rising edge of $\overline{\text{SHDN}}$ or during the exiting of thermal shutdown, the short-to-battery detection is performed. At this point, if the device detects the short-to-battery fault, the LDO/switch stays off and $\overline{\text{ERR}}$ is asserted low. The fault is latched, and the startup resumes when the short-to-battery fault is removed and $\overline{\text{SHDN}}$ is toggled.

During normal operation if a short-to-battery fault results in reverse current for more than 5 μs (typ) the LDO/switch is latched off and $\overline{\text{ERR}}$ is asserted low. To remove the

latched condition after a short-to-battery (reverse current) fault, the fault condition must first be removed and $\overline{\text{SHDN}}$ must be toggled.

Series inductance and the output capacitor can produce ringing during large load transients when enabling the LDO/switch, resulting in an output voltage that temporarily exceeds the input voltage. Blanking is implemented during startup. The reverse current blanking time ($t_{\text{REV_BLANK}}$) is 16ms (typ).

When the MAX16948 is operated with input voltage close to the output voltage, as in switch mode operation or LDO mode in dropout, care must be taken to avoid a false reverse current detection in the presence of a short circuit to ground fault. If both channels are enabled and one channel is shorted to ground after startup, the current drawn from C_{IN} may result in a temporary dip in the input voltage, which may trigger the reverse current detection fault. To avoid this false trigger event, use an electrolytic capacitor of at least 100 μF .

Thermal Shutdown

Thermal shutdown circuitry protects the device from overheating. The switch turns off immediately when the junction temperature exceeds +160°C (typ). The switch turns on again after the device temperature drops by approximately 15°C (typ).

FB Inputs (FB1/FB2)

FB1 and FB2 control the output voltage on OUT1 and OUT2. Connect FB_ to GND to select current-limited switch operation. Connect to an external resistive divider for adjustable output voltage operation (feedback voltage 1V (typ)). Connect FB_ to REG to choose the internal resistive divider for the 8.5V regulator option.

Table 1. Fault Response

DEVICE STATUS	V _{LIM} (V)	$\overline{\text{VERR}}$	LDO/SW OUTPUT	LATCHED
Normal Operation	$0 \leq V_{\text{LIM}} < 2.5$	High	Enabled	No
Short Circuit to GND or Overcurrent	2.5	Low	Enabled	No
V _{OUT} > V _{IN} at Startup	0	Low	Disabled	Yes
Reverse Current	0	Low	Disabled	Yes
V _{FB} > 1.2V	0	High	Enabled	No
Input Overvoltage	0	Low	Disabled	No
Thermal Shutdown	0	High	Disabled	No

MAX16948

Automotive Dual Remote Antenna Current-Sense LDO/Switch

Undervoltage and Overvoltage Lockout

The device includes undervoltage lockout circuitry (UVLO) to prevent erroneous switch operation when the input voltage goes below 4.45V (max) during startup and brownout conditions. Input voltages of less than 4.45V inhibit operation of the device by turning off the internal charge pump and the switch.

The device also features an overvoltage lockout (OVLO) threshold of +21V (typ). When V_{IN} is greater than V_{OVLO} , the device immediately turns off the switch and the internal charge pump.

Shutdown ($\overline{SHDN1}$, $\overline{SHDN2}$)

The device features two active-low shutdown inputs ($\overline{SHDN1}$, $\overline{SHDN2}$) to place the device in a low-power shutdown mode. $\overline{SHDN1}$ controls OUT1 and $\overline{SHDN2}$ controls OUT2. The device turns off both channels and consumes a maximum of 7 μ A (at $V_{IN} = 12$ V) of shutdown current when both $\overline{SHDN1}$ and $\overline{SHDN2}$ are low. Driving $\overline{SHDN1}$ and/or $\overline{SHDN2}$ high initiates device turn-on with short-to-battery detection. $\overline{SHDN1}$ and $\overline{SHDN2}$ are pulled to GND with an internal 500k Ω resistor.

Internal Voltage Regulator (REG)

The device features an internal regulator that regulates the input voltage to +5V to power all internal circuitry. Bypass the regulator output (REG) to GND with a 1 μ F ceramic capacitor. The REG output can be used to supply external circuitry up to a maximum 15mA current.

Applications Information/External Component Selection

Fixed/Adjustable Output Voltage

The device is configurable to provide a fixed 8.5V output or as an adjustable LDO with an output between 1V and 12.5V. Connect FB_{-} to REG to configure the device as an 8.5V LDO. Connect a resistive divider between OUT_{-} , FB_{-} , and GND to set the output to the desired voltage. FB_{-} is regulated to 1.0V with $\pm 3\%$ accuracy for a load current between 5mA and 150mA. Select a value for R2 and calculate R1 as follows:

$$R1 = \left(\frac{V_{OUT_{-}}}{V_{FB_{-}}} - 1 \right) R2$$

where R2 must be less than or equal to 1k Ω . Select R1 and R2 such that the maximum input bias current at FB_{-} ($\pm 0.5\mu$ A) is negligible compared to the current flowing through R1 and R2.

Current-Limit Resistor and Capacitor Selection

The current-sense outputs at LIM1/LIM2 are proportional to the load current at OUT1/OUT2 and are internally connected to a current-limit comparator referenced to 2.5V. The desired current limit is set with an external resistor $R_{LIM_{-}}$ using the following equation:

$$R_{LIM_{-}} (\Omega) = \frac{2.5(V)}{0.5\% \times I_{LOAD_MAX}(A)}$$

A 0.1 μ F compensation capacitor $C_{LIM_{-}}$ must be placed in parallel with $R_{LIM_{-}}$ to establish a dominant pole in the current limiting loop to maintain stability and to prevent fast current transients from prematurely triggering the current limit.

Current-Sense Resistor Selection

The current-sense outputs SENSE1/SENSE2 are proportional to the load current at OUT1/OUT2. An output resistor $R_{SENSE_{-}}$ must be connected between SENSE1/SENSE2 and GND to generate a current-sense voltage prior to sampling by an ADC. R_{SENSE} is calculated using the following equation:

$$R_{SENSE_{-}} (\Omega) = \frac{V_{ADC,FS}(V)}{0.5\% \times I_{LOAD_MAX}(A)}$$

where $V_{ADC,FS}$ is the full-scale input voltage of the ADC. A 0.1 μ F capacitor C_{SENSE} should be placed in parallel with R_{SENSE} to hold the voltage during ADC sampling cycles.

In addition to ADC sampling, open-load or overcurrent conditions can be detected by using external comparators and splitting $R_{SENSE_{-}}$ into a resistive divider as shown in [Figure 1](#). In this circuit the outputs of comparators U1 and

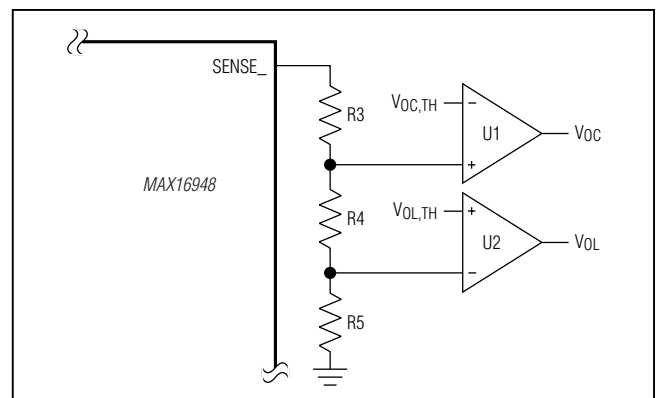


Figure 1. Open-Load and Overcurrent Detection Circuit

MAX16948

Automotive Dual Remote Antenna Current-Sense LDO/Switch

Table 2. Open-Load and Overcurrent-Detection Circuit States

OC	OL	STATE
0	0	Normal Operation
0	1	Open-Load Condition
1	0	Overcurrent Condition
1	1	Invalid State

U2 indicate the operating state of the circuit as shown in [Table 2](#).

The open-load and overcurrent limits are set using the following equations:

$$R5(\Omega) = \frac{V_{OL,TH}(V)}{0.5\% \times I_{OPEN-LOAD}(A)}$$

where $V_{OL,TH}$ is the open-load voltage threshold for comparator U2 and $I_{OPEN-LOAD}$ is the desired open-load threshold for the current flowing through the LDO/switch.

$$R4(\Omega) = \frac{V_{OC,TH}(V)}{0.5\% \times I_{OVERCURRENT}(A)} - R5$$

where $V_{OC,TH}$ is the overcurrent voltage threshold for comparator U1. $I_{OVERCURRENT}$ is the desired overcurrent threshold for the current flowing through the LDO/switch and the internal current limit should be programmed such that $I_{LOAD_MAX} > I_{OVERCURRENT}$.

$$R3(\Omega) = R_{SENSE_} - R4 - R5$$

The open-load voltage threshold $V_{OL,TH}$ and overcurrent threshold $V_{OC,TH}$ can be generated from REG utilizing a resistive divider. The thresholds should be selected such that $V_{OC,TH} > V_{OL,TH}$.

Input Capacitor

Connect a parallel electrolytic capacitor and a low-ESR ceramic capacitor from IN to GND to limit the input-voltage drop during momentary output short-circuit conditions and to protect the device against transients due to inductance on the IN line. For example, use at least a 0.1μF ceramic capacitor in parallel with a 10μF electrolytic capacitor if the input inductance (including any stray inductance) is estimated to be 20μH. Larger capacitor values reduce the voltage undershoot and overshoot in case of reverse current.

Output Capacitor

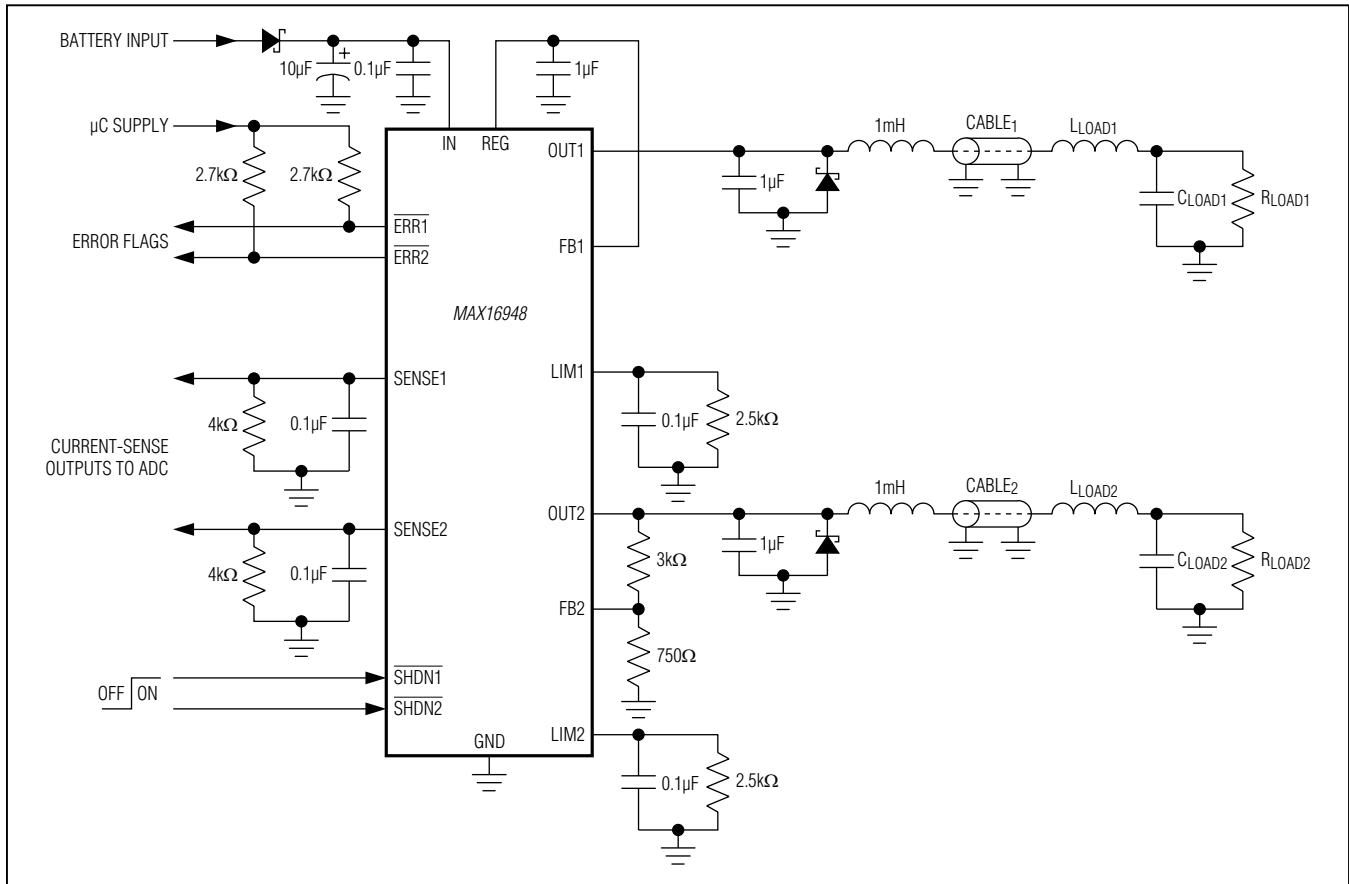
Connect >1μF in parallel with 0.1μF low-ESR ceramic capacitors between OUT1/OUT2 and GND for regulator stability. These capacitors should be placed as close to the device as possible. Utilize capacitors with an X7R dielectric to ensure stability over the operating temperature range of the device.

In an analogous fashion to the input capacitor, an output capacitor protects the device against transients due to any series inductance in the output. Under no conditions should the voltage on OUT1/OUT2 go below -0.3V as specified in the [Absolute Maximum Ratings](#) section. A Schottky diode is required as a clamp if transients are expected to go below ground.

MAX16948

Automotive Dual Remote Antenna Current-Sense LDO/Switch

Typical Operating Circuits (continued)



Ordering Information

PART	TEMP RANGE	PIN-PACKAGE
MAX16948AGTE/V+	-40°C to +105°C	16 TQFN-EP**
MAX16948AGEE/V+	-40°C to +105°C	16 QSOP-EP**
MAX16948BGEE/V+*	-40°C to +105°C	16 QSOP

V denotes an automotive qualified part.

+ Denotes a lead(Pb)-free/RoHS-compliant package.

*Future product—contact factory for availability.

**EP = Exposed pad.

Chip Information

PROCESS: BiCMOS

Package Information

For the latest package outline information and land patterns (footprints), go to www.maximintegrated.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	OUTLINE NO.	LAND PATTERN NO.
16 TQFN-EP	T1644+4	21-0139	90-0070
16 QSOP-EP	E16E+9	21-0112	90-0240
16 QSOP	E16+5	21-0055	90-0167

MAX16948

Automotive Dual Remote Antenna Current-Sense LDO/Switch

Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	9/11	Initial release	—
1	4/12	Updated maximum output voltage	1, 3, 10, 12
2	5/13	Added light-load current-sense specs to <i>Electrical Characteristics</i> table	3



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15

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