

1A Step-down DC/DC Converters with Synchronous Rectifier

NO.EA-273-150317

OUTLINE

The RP505K Series are low supply current CMOS-based 1A*1 step-down DC/DC Converters with synchronous rectifier. Each of these ICs consists of an oscillator, a reference voltage unit, an error amplifier, a switching control circuit, a mode control circuit, a soft-start circuit, a latch type protection circuit, an under voltage lock out (UVLO) circuit, a thermal shutdown circuit, and a switching transistors. A low ripple, high efficiency synchronous rectifier step-down DC/DC converter can be easily composed of this IC with only an inductor and capacitors. Since the package is DFN(PLP)2020-8, high density mounting on boards is possible.

In the RP505K series, as for the A version and B version, since feedback resistors are built-in, the voltage is fixed internally. 0.1V step output can be set by laser-trim and $\pm 1.5\%$ ($V_{OUT} \geq 1.2V$) or 18mV ($V_{OUT} < 1.2V$) tolerance is guaranteed. As for the C version, output voltage is adjustable with external divider resistors.

By inputting a signal to MODE pin, the RP505K Series can choose PWM/VFM alternative mode or forced PWM mode. In low output current, PWM/VFM alternative mode automatically switches from PWM to VFM in order to achieve high efficiency. Likewise, in low output current, Forced PWM mode switches at fixed frequency in order to reduce noise.

As protection circuits, the RP505K Series contain a current limit circuit which limits the Lx peak current in each clock cycle, and a latch type protection circuit which latches the built-in driver to the OFF state if the load current exceeds the limit value or the output short continues for a specified time (the protection delay time). The latch protective circuit can be released by once putting the IC into the standby mode with the CE pin and then into the active mode, or, by turning the power off and back on. Setting the supply voltage lower than the UVLO detector threshold can also release the latch protective circuit. The RP505K Series also contain a thermal shutdown circuit which detects the overheating and resets the IC when the junction temperature of the RP505K Series exceeds the specified temperature.

*1 This is an approximate value, because output current depends on conditions and external parts.

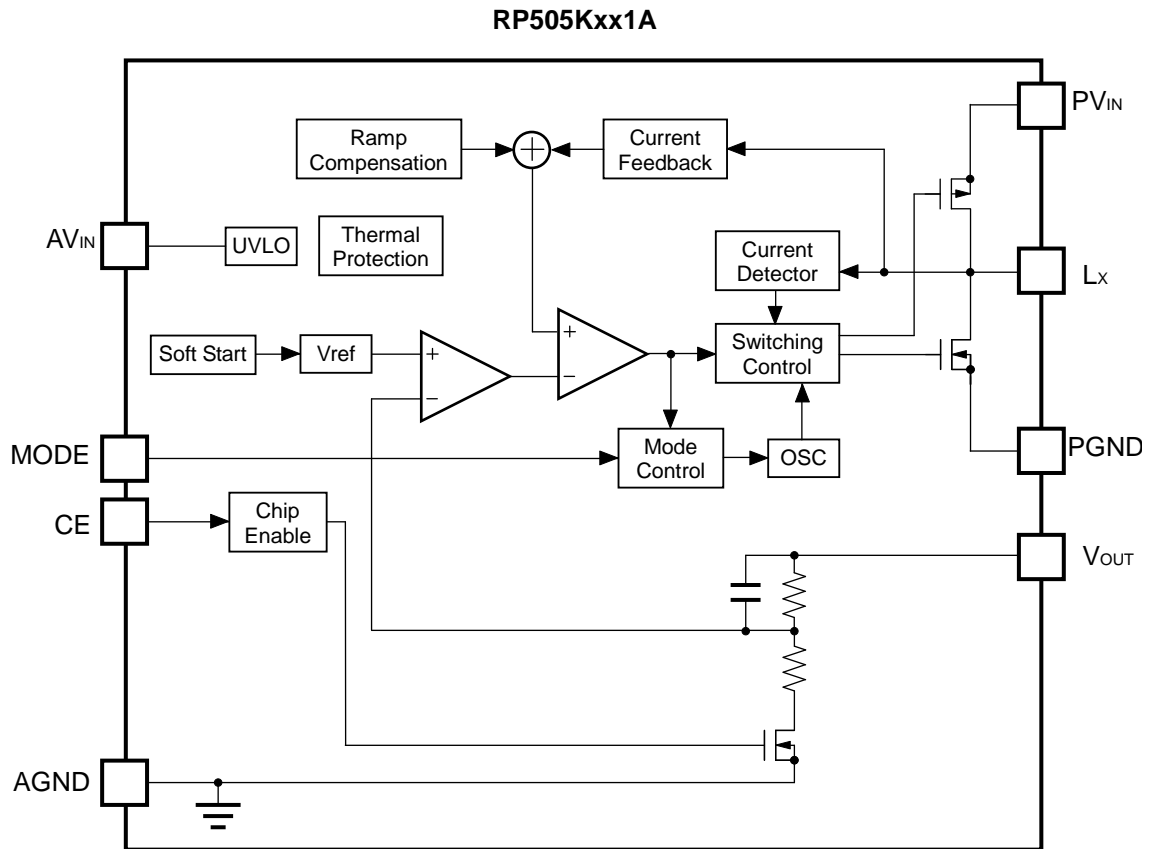
FEATURES

- Supply Current Typ. 40 μ A (VFM mode with no-load)
- Standby Current Max. 5 μ A
- Input Voltage Range 2.3V to 5.5V (Absolute maximum rating: 6.5V)
- Output Voltage Range (Ver.A,B) 0.6V to 3.3V (Adjustable in 0.1V steps.
Note: As for 0.8V or less, input voltage range is limited.)
(Ver.C) 0.8V to 3.3V
- Output Voltage Accuracy (Ver.A,B) $\pm 1.5\%$ ($V_{OUT} \geq 1.2V$), $\pm 18mV$ ($V_{OUT} < 1.2V$)
- Feedback Voltage Accuracy (Ver.C) $\pm 9mV$ ($V_{FB} = 0.6V$)
- Temperature-Drift Coefficient of Output Voltage/ Feedback
..... Typ. $\pm 100ppm/^\circ C$
- Oscillator Frequency Typ. 2.25MHz
- Oscillator Maximum Duty Cycle Min. 100%
- Built-in Driver ON Resistance Typ. Pch. 0.23 Ω , Nch. 0.20 Ω ($V_{IN} = 3.6V$)
- UVLO Detector Threshold Typ. 2.0V
- Soft-start Time Typ. 0.15ms
- Lx Current Limit Circuit Typ. 1.7A
- Latch Type Protection Circuit Typ. 1.5ms
- Package DFN(PLP)2020-8

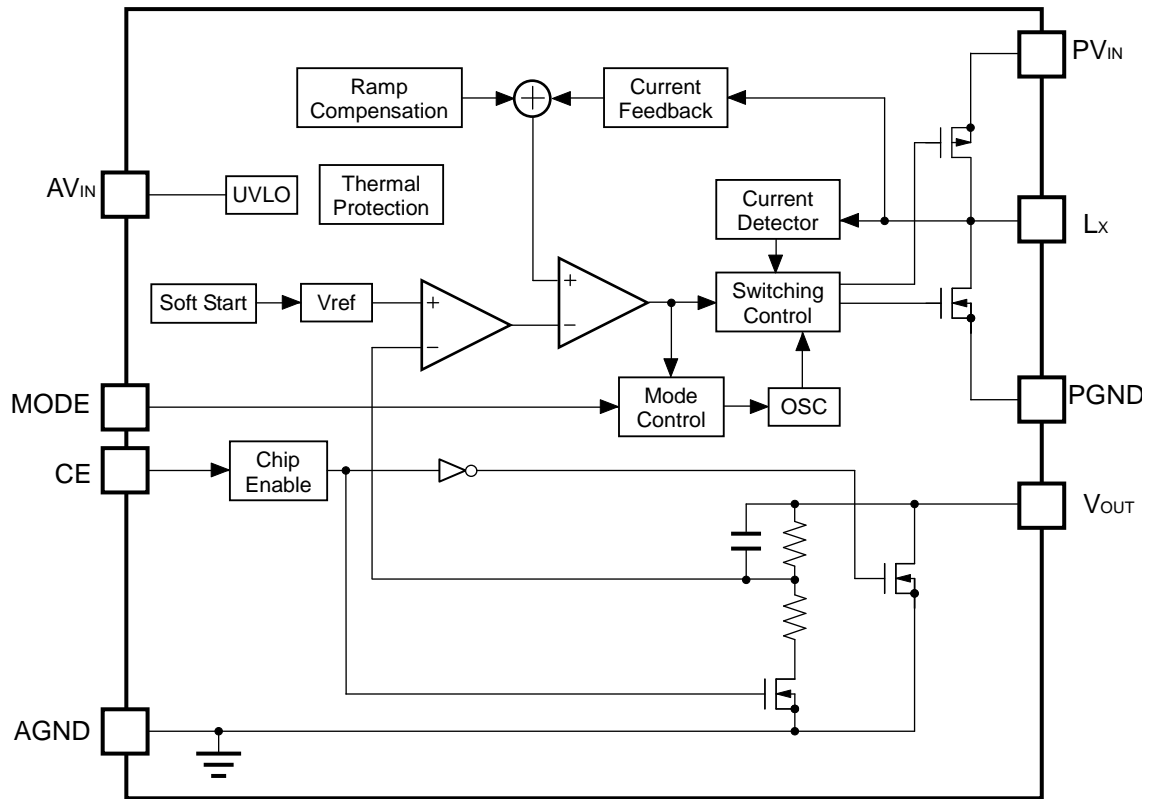
APPLICATIONS

- Power source for portable equipment such as cellular, PDA, DSC, Notebook PC
- Power source for HDD, WLAN.
- Power source for Li-ion battery-used equipment

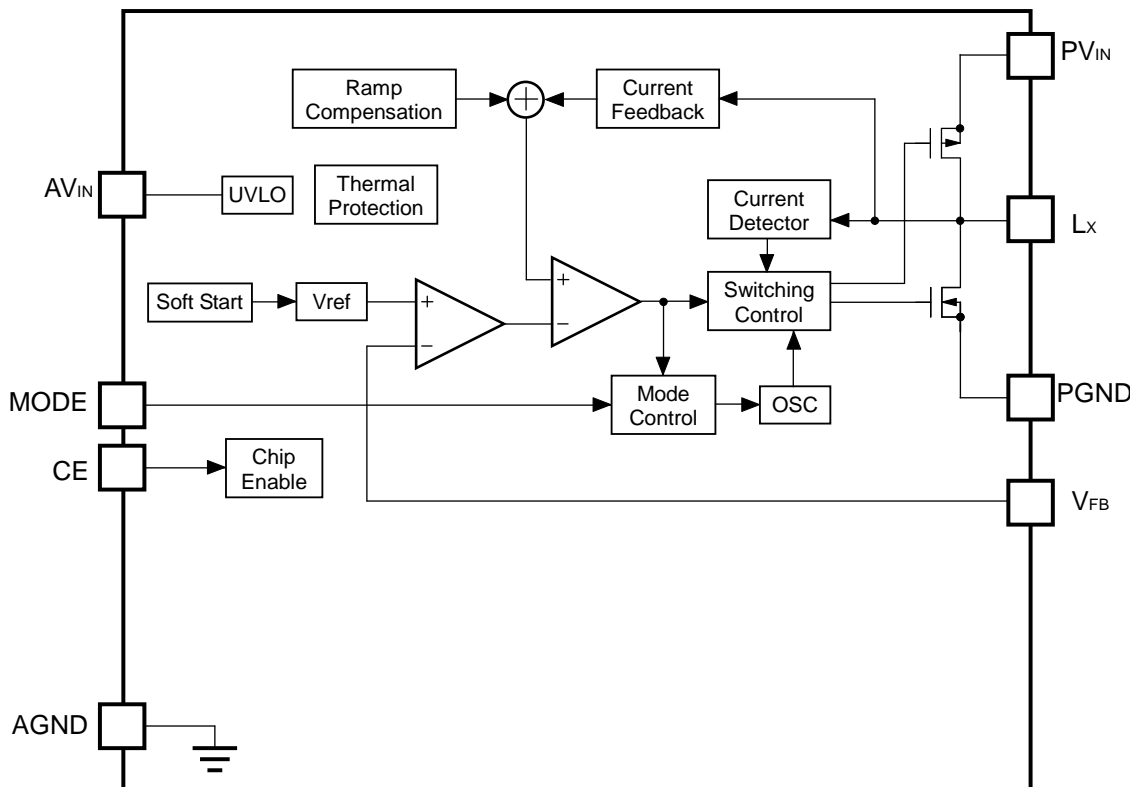
BLOCK DIAGRAMS



RP505Kxx1B



RP505K001C



SELECTION GUIDE

In the RP505K Series, output voltage, and auto discharge function for the IC are selectable at the user's request.

Product Name	Package	Quantity per Reel	Pb Free	Halogen Free
RP505Kxx1\$-TR	DFN(PLP)2020-8	5,000pcs	Yes	Yes

xx : The output voltage can be designated in the range from 0.6V(06) to 3.3V(33) in 0.1V^{*1} steps.

The output voltage adjustable type: xx=00

(For other voltages, please refer to MARK INFORMATION.)

\$: Designation of Mask Option

A) Fixed output voltage type, without auto-discharge function at off state

B) Fixed output voltage type, with auto-discharge function at off state

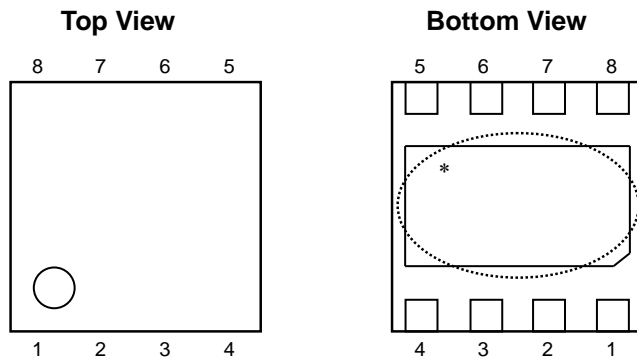
C) Adjustable output voltage type, without auto-discharge function at off state

Auto-discharge function quickly lowers the output voltage to 0V, when the chip enable signal is switched from the active mode to the standby mode, by releasing the electrical charge accumulated in the external capacitor.

*1 0.05V step is also available as a custom code.

PIN CONFIGURATIONS

• DFN(PLP)2020-8



PIN DESCRIPTIONS

• RP505K: DFN(PLP)2020-8

Pin No.	Symbol	Description
1	MODE	Mode Control Pin ("H" Forced PWM Mode, "L" PWM/VFM Alternative Mode)
2	CE	Chip Enable Pin ("H" Active)
3	AV _{IN}	Input Pin* ¹
4	PV _{IN}	Input Pin* ¹
5	L _X	L _X Switching Pin
6	PGND	Ground Pin* ¹
7	AGND	Ground Pin* ¹
8	V _{OUT} /V _{FB}	Output Pin / Feedback Pin

Tab is GND level. (They are connected to the reverse side of this IC.) The tab is better to be connected to the GND, but leaving it open is also acceptable.

*¹ No.3 pin and No.4 pin, and also No.6 pin and No.7 pin must be wired each other when mounted on boards.

ABSOLUTE MAXIMUM RATINGS

(AGND=PGND=0V)

Symbol	Item	Rating	Unit
A/PV _{IN}	AV _{IN} /PV _{IN} Input Voltage	-0.3 to 6.5	V
V _{LX}	L _X Pin Voltage	-0.3 to A/PV _{IN} + 0.3	V
V _{CE}	CE Pin Input Voltage	-0.3 to 6.5	V
V _{OUT} /V _{FB}	V _{OUT} /V _{FB} Pin Voltage	-0.3 to 6.5	V
V _{MODE}	MODE Pin Voltage	-0.3 to 6.5	V
I _{LX}	L _X Pin Output Current	1.7	A
P _D	Power Dissipation (Standard Test Land Pattern)* ¹	880	mW
T _a	Operating Temperature Range	-40 to 85	°C
T _{stg}	Storage Temperature Range	-55 to 125	°C

*¹ For Power Dissipation and Standard Test Land Pattern, please refer to PACKAGE INFORMATION.

ABSOLUTE MAXIMUM RATINGS

Electronic and mechanical stress momentarily exceeded absolute maximum ratings may cause the permanent damages and may degrade the life time and safety for both device and system using the device in the field. The functional operation at or over these absolute maximum ratings is not assured.

RECOMMENDED OPERATING CONDITIONS (ELECTRICAL CHARACTERISTICS)

All of electronic equipment should be designed that the mounted semiconductor devices operate within the recommended operating conditions. The semiconductor devices cannot operate normally over the recommended operating conditions, even if when they are used over such conditions by momentary electronic noise or surge. And the semiconductor devices may receive serious damage when they continue to operate over the recommended operating conditions.

ELECTRICAL CHARACTERISTICS

● RP505Kxx1A/B

(Ta=25°C)

Symbol	Item	Conditions	Min.	Typ.	Max.	Unit
A/PVIN	Operating Input Voltage	$0.8V \leq V_{OUT} \leq 3.3V$	2.3		5.5	V
		$0.7V \leq V_{OUT} < 0.8V$	2.3		4.4	
		$0.6V \leq V_{OUT} < 0.7V$, MODE="L" ^{*1}	2.3		4.2	
VOUT	Output Voltage	Refer to the conditions below. ^{*2}	$V_{OUT} \geq 1.2V$	-1.5%	+1.5%	V
			$V_{OUT} < 1.2V$	-0.018	+0.018	
$\Delta V_{OUT}/\Delta T$	Output Voltage Temperature Coefficient	$-40^{\circ}C \leq T_a \leq 85^{\circ}C$		±100		ppm/°C
fosc	Oscillator Frequency	Refer to the conditions below. ^{*2}	2.00	2.25	2.50	MHz
I _{DD1}	Supply Current 1	A/PVIN=VCE=5.5V, VOUT=VSET×0.8		500	840	μA
I _{DD2}	Supply Current 2	A/PVIN=VCE=VOUT=5.5V	V _{MODE} =0V	40	60	μA
			V _{MODE} =5.5V	500	840	
I _{standby}	Standby Current	A/PVIN=5.5V, VCE=0V		0	5	μA
I _{CEH}	CE "H" Input Voltage	A/PVIN=VCE=5.5V	-1	0	1	μA
I _{CEL}	CE "L" Input Voltage	A/PVIN=5.5V, VCE=0V	-1	0	1	μA
I _{MODEH}	Mode "H" Input Current	A/PVIN=V _{MODE} =5.5V, VCE=0V	-1	0	1	μA
I _{MODEL}	Mode "L" Input Current	A/PVIN=5.5V, VCE=V _{MODE} =0V	-1	0	1	μA
I _{VOUTH}	VOUT "H" Input Current ^{*3}	A/PVIN=VOUT=5.5V, VCE=0V	-1	0	1	μA
I _{VOUTL}	VOUT "L" Input Current	A/PVIN=5.5V, VCE=VOUT=0V	-1	0	1	μA
R _{LOW}	Nch On Resistance for Auto Discharge ^{*4}	A/PVIN=3.6V, VCE=0V		30		Ω
I _{LXLEAKH}	LX Leakage Current "H"	A/PVIN=V _{LX} =5.5V, VCE=0V	-1	0	5	μA
I _{LXLEAKL}	LX Leakage Current "L"	A/PVIN=5.5V, VCE=V _{LX} =0V	-5	0	1	μA
V _{CEH}	CE "H" Input Voltage	A/PVIN=5.5V	1.0			V
V _{CEL}	CE "L" Input Voltage	A/PVIN=2.3V			0.4	V
V _{MODEH}	Mode "H" Input Voltage	A/PVIN=VCE=5.5V	1.0			V
V _{MODEL}	Mode "L" Input Voltage	A/PVIN=VCE=2.3V			0.4	V
R _{ONP}	On Resistance of Pch Tr.	A/PVIN=3.6V, I _{LX} =-100mA		0.23		Ω
R _{ONN}	On Resistance of Nch Tr.	A/PVIN=3.6V, I _{LX} =-100mA		0.20		Ω
Maxduty	Oscillator Maximum Duty Cycle		100			%
t _{start}	Soft-start Time	Refer to the conditions below. ^{*2}		150	300	μs
I _{LXlim}	Lx Current Limit	Refer to the conditions below. ^{*2}	1400	1700		mA
t _{prot}	Protection Delay Time	Refer to the conditions below. ^{*2}	0.5	1.5	5	ms
V _{UVLO1}	UVLO Detector Threshold	A/PVIN=VCE	1.9	2.0	2.1	V
V _{UVLO2}	UVLO Released Voltage	A/PVIN=VCE	2.0	2.1	2.2	V
T _{TSD}	Thermal Shutdown Temperature	Junction Temperature		140		°C
T _{TSR}	Thermal Shutdown Released Temperature	Junction Temperature		100		°C

Note: Test circuit is "OPEN LOOP" and AGND=PGND=0V unless otherwise specified.

^{*1} $0.6V \leq V_{OUT} < 0.7V$: MODE="L" PWM/VFM Alternative Mode.

^{*2} A/PVIN=VCE=3.6V (V_{SET} ≤ 2.6V), A/PVIN=VCE=V_{SET}+1V (V_{SET}>2.6V)

^{*3} with no auto discharge version only

^{*4} with auto discharge version only

● RP505K001C

(Ta=25°C)

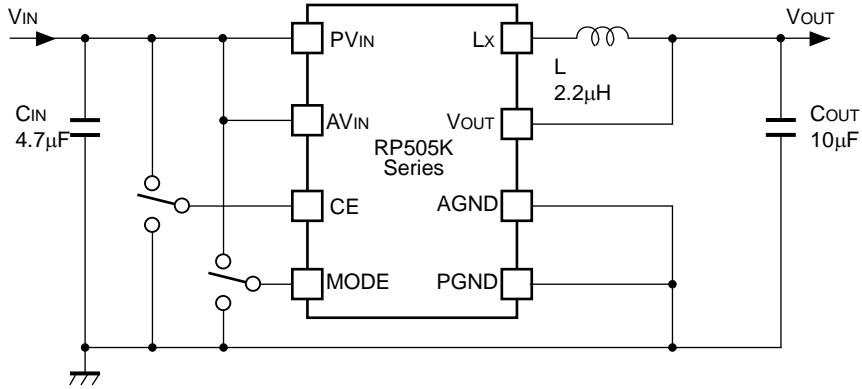
Symbol	Item	Conditions	Min.	Typ.	Max.	Unit
A/PVIN	Operating Input Voltage		2.3		5.5	V
VFB	Feedback Output Voltage	A/PVIN=VCE=3.6V	0.591	0.600	0.609	V
$\Delta V_{FB}/\Delta T$	Feedback Output Voltage Temperature Coefficient	-40°C ≤ Ta ≤ 85°C		±100		ppm/°C
fosc	Oscillator Frequency	Refer to the conditions below. *1	2.00	2.25	2.50	MHz
IDD1	Supply Current 1	A/PVIN=VCE=5.5V, VFB=0.48V		500	840	μA
IDD2	Supply Current 2	A/PVIN=VCE=VOUT=5.5V		40	60	μA
		VMODE=0V		500	840	
Istandby	Standby Current	A/PVIN=5.5V, VCE=0V		0	5	μA
ICEH	CE "H" Input Voltage	A/PVIN=VCE=5.5V	-1	0	1	μA
ICEL	CE "L" Input Voltage	A/PVIN=5.5V, VCE=0V	-1	0	1	μA
IMODEH	Mode "H" Input Current	A/PVIN=VMODE=5.5V, VCE=0V	-1	0	1	μA
IMODEL	Mode "L" Input Current	A/PVIN=5.5V, VCE=VMODE=0V	-1	0	1	μA
IVFBH	VFB "H" Input Current	A/PVIN=VOUT=5.5V, VCE=0V	-1	0	1	μA
IVFBL	VFB "L" Input Current	A/PVIN=5.5V, VCE=VOUT=0V	-1	0	1	μA
ILXLEAKH	LX Leakage Current "H"	A/PVIN=VLX=5.5V, VCE=0V	-1	0	5	μA
ILXLEAKL	LX Leakage Current "L"	A/PVIN=5.5V, VCE=VLX=0V	-5	0	1	μA
VCEH	CE "H" Input Voltage	A/PVIN=5.5V	1.0			V
VCEL	CE "L" Input Voltage	A/PVIN=2.3V			0.4	V
VMODEH	Mode "H" Input Voltage	A/PVIN=VCE=5.5V	1.0			V
VMODEL	Mode "L" Input Voltage	A/PVIN=VCE=2.3V			0.4	V
RONP	On Resistance of Pch Tr.	A/PVIN=3.6V, ILX=-100mA		0.23		Ω
RONN	On Resistance of Nch Tr.	A/PVIN=3.6V, ILX=-100mA		0.20		Ω
Maxduty	Oscillator Maximum Duty Cycle		100			%
tstart	Soft-start Time	Refer to the conditions below. *1		150	300	μs
ILXlim	LX Current Limit	Refer to the conditions below. *1	1400	1700		mA
tprot	Protection Delay Time	Refer to the conditions below. *1	0.5	1.5	5	ms
VUVLO1	UVLO Detector Threshold	A/PVIN=VCE	1.9	2.0	2.1	V
VUVLO2	UVLO Released Voltage	A/PVIN=VCE	2.0	2.1	2.2	V
TSD	Thermal Shutdown Temperature	Junction Temperature		140		°C
TTSR	Thermal Shutdown Released Temperature	Junction Temperature		100		°C

Test circuit is "OPEN LOOP" and AGND=PGND=0V unless otherwise specified.

*1 A/PVIN=VCE=3.6V (VSET ≤ 2.6V), A/PVIN=VCE=VSET+1V (VSET>2.6V)

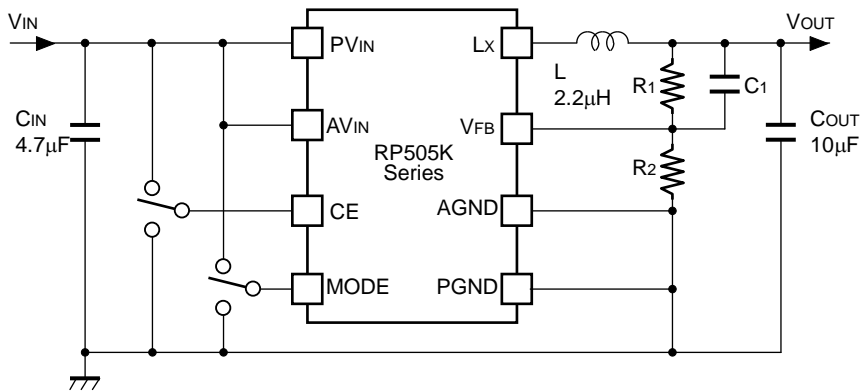
TYPICAL APPLICATION

(Fixed Output Voltage Type)



*) MODE="H" Forced PWM Mode
 MODE="L" PWM/VFM Alternative Mode

(Adjustable Output Voltage Type)



*) MODE="H" Forced PWM Mode
 MODE="L" PWM/VFM Alternative Mode

Symbol	Recommendation components		
C _{IN}	4.7µF	Ceramic	C1608JB0J475K(TDK)
C _{OUT}	10µF	Ceramic	C1608JB0J106M(TDK)
L	2.2µH	Inductor	MIPSA2520D2R2(FDK)

TECHNICAL NOTES

When you use these ICs, consider the following issues:

- Set the same level as AGND and PGND.
- Set the same level as AV_{IN} and PV_{IN}.
- Place the external parts as close as possible to the IC by using a short as possible wiring. Especially, place the capacitor as close as possible to the PV_{IN} and PGND pins. Ensure the V_{DD} and GND lines are sufficiently robust. If their impedances are too high, the electrical potential of the inside of the IC could be fluctuated by switching current, and noise pickup or unstable operation could be the results. Please note that the large switching current flows through the V_{DD} line, the GND line, an inductor, the L_X, and the V_{OUT} line. Separate the line between the V_{OUT} pin and an inductor (A and B versions), and the line between a resistor for setting output voltage (R1) and an inductor (C version), from the line connected to the load. Use a ceramic capacitor with the small ESR value.
- The recommended capacitance value for the C_{IN} capacitor connected between the PV_{IN} and PGND pins is 4.7μF or more. Also, the recommended capacitance value for the C_{OUT} capacitor is 10μF.
- The Inductance value should be set within the range of 1.0 to 2.2μH. However, the inductance value is limited by output voltage, so please refer to the table below. For stable operation, the phase compensation is set according to the specified inductance value and the specified C_{OUT} capacitance value. Select the inductor with low DC resistance, with large permissive current, with high resistant to magnetic saturation. Select the inductance value considering the load current by the conditions of use. If the inductance value is small, the L_X peak current may increase along with the increase of load current. When the L_X peak current reaches to the “L_X limit current”, the current limit circuit may be activated.

[Inductance Range vs. Output Voltage Range]

V _{OUT} [V]	L=1.0μH	L=1.5μH	L=2.2μH
0.6~1.55	○	○	○
1.6~2.3	×	○	○
2.35~3.3	×	×	○

- Please note that Current limit circuit and Latch type protection circuit could be affected by self-heating or heat dissipation environment.
- For adjustable output voltage type (C version), the output voltage (V_{OUT}) is adjustable by changing the R₁ and R₂ values as follows.

$$V_{OUT} = V_{FB} \times (R_1 + R_2) / R_2 \quad (0.8V \leq V_{OUT} \leq 3.3V)$$

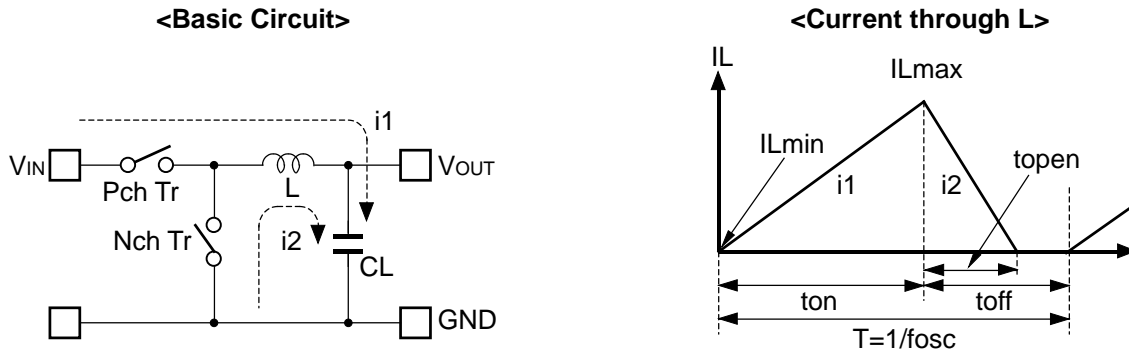
If the R₁ and R₂ values are large, the impedances of the V_{FB} pin become large and could be easily affected by noise. Therefore, set the R₂ value to 220kΩ or less. If the operation becomes unstable due to the high impedance, the impedance should be decreased. The C₁ value can be calculated by the following formula. Please use the value close to the calculation result.

$$C_1 = 4.84 \times 10^{-6} / R_2 \text{ [F]}$$

* The performance of power supply circuits using this IC largely depends on the peripheral circuits. Please be very careful when setting the peripheral parts. When designing the peripheral circuits of each part, PCB patterns, and this IC, please do not exceed the rated values (Voltage, Current, Power).

Operation of Step-down DC/DC Converter and Output Current

The DC/DC converter charges energy in the inductor when Lx transistor is ON, and discharges the energy from the inductor when Lx transistor is OFF and controls with less energy loss, so that a lower output voltage than the input voltage is obtained. The operation will be explained with reference to the following diagrams:



- Step 1: Pch Tr. turns on and current IL (=i1) flows, and energy is charged into CL. At this moment, IL increases from ILmin (=0) to reach ILmax in proportion to the on-time period (ton) of Pch Tr.
- Step 2: When Pch Tr. turns off, Synchronous rectifier Nch Tr. turns on in order that L maintains IL at ILmax, and current IL (=i2) flows.
- Step 3: IL (=i2) decreases gradually and reaches IL=ILmin=0 after a time period of toff, and Nch Tr. turns off. Provided that in the continuous mode, next cycle starts before IL becomes to 0 because toff time is not enough. In this case, IL value increases from this ILmin (>0).

In the case of PWM control system, the output voltage is maintained by controlling the on-time period (ton), with the oscillator frequency (fosc) being maintained constant.

The maximum value (ILmax) and the minimum value (ILmin) of the current flowing through the inductor are the same as those when Pch Tr. turns on and off.

The difference between ILmax and ILmin, which is represented by ΔI:

$$\Delta I = I_{Lmax} - I_{Lmin} = V_{OUT} \times t_{open} / L = (V_{IN} - V_{OUT}) \times t_{on} / L \dots\dots\dots \text{Equation 1}$$

Wherein,

$$T = 1 / f_{osc} = t_{on} + t_{off}$$

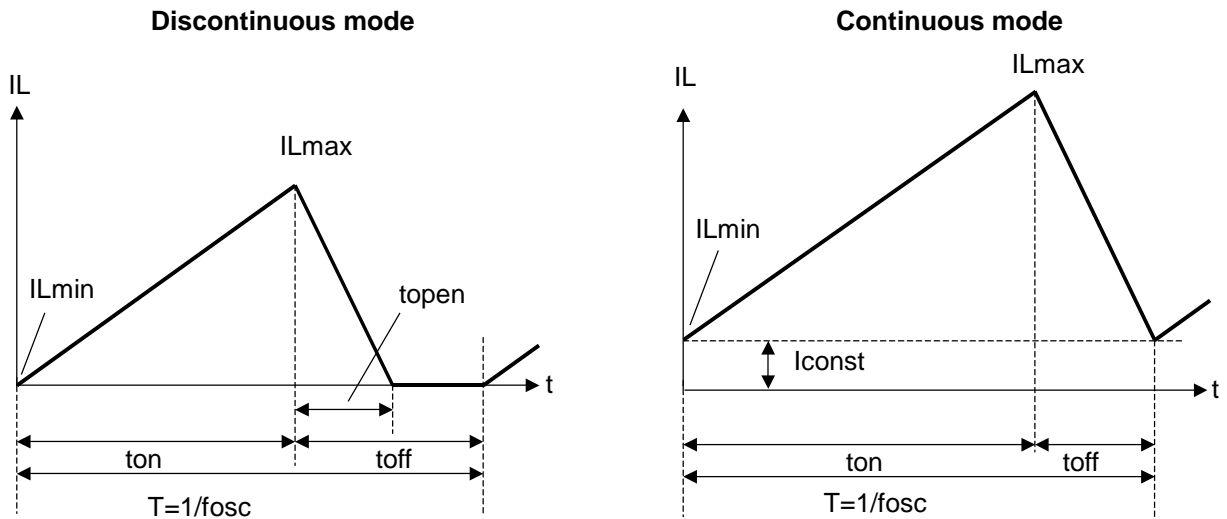
$$\text{duty (\%)} = t_{on} / T \times 100 = t_{on} \times f_{osc} \times 100$$

$$t_{open} \leq t_{off}$$

In Equation 1, $V_{OUT} \times t_{open} / L$ and $(V_{IN} - V_{OUT}) \times t_{on} / L$ respectively show the change of the current at "ON", and the change of the current at "OFF".

Discontinuous mode and Continuous mode

When the output current (I_{OUT}) is relatively small, $t_{open} < t_{off}$ as illustrated in the above diagram. In this case, the energy is charged in the inductor during the time period of t_{on} and is discharged in its entirety during the time period of t_{off} , therefore I_{Lmin} becomes to zero ($I_{Lmin}=0$). When I_{OUT} is gradually increased, eventually, t_{open} becomes to t_{off} ($t_{open}=t_{off}$), and when I_{OUT} is further increased, I_{Lmin} becomes larger than zero ($I_{Lmin}>0$). The former mode is referred to as the discontinuous mode and the latter mode is referred to as continuous mode.



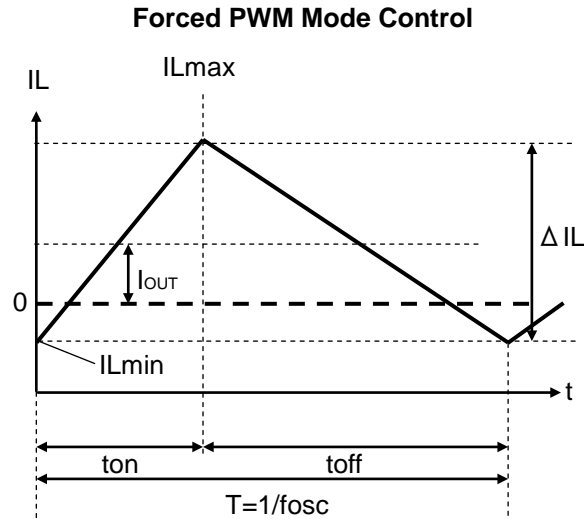
In the continuous mode, when Equation 1 is solved for t_{on} and assumed that the solution is t_{onc} ,

$$t_{onc} = T \times V_{OUT} / V_{IN} \dots \dots \dots \text{Equation 2}$$

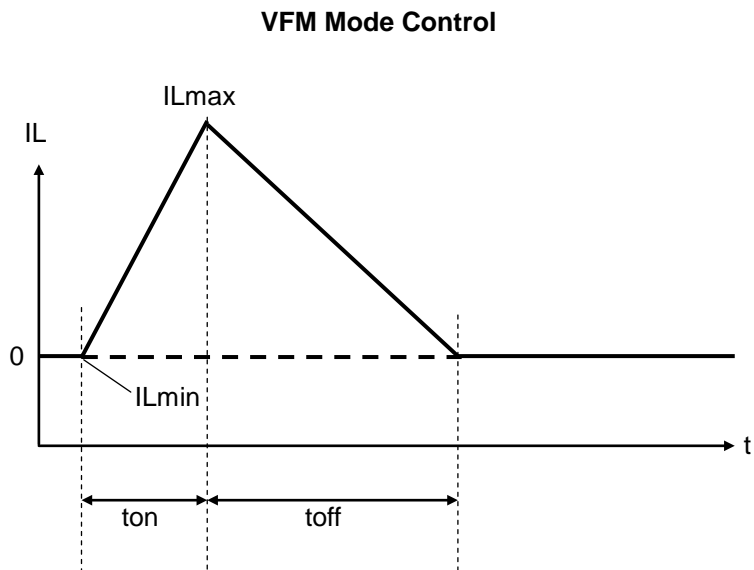
When $t_{on} < t_{onc}$, the mode is the discontinuous mode, and when $t_{on} = t_{onc}$, the mode is the continuous mode.

Forced PWM Mode Control and VFM Mode Control

By setting the Mode pin to H, the IC switches the frequency at the fixed rate to reduce noise even when output load is light. Therefore, when I_{OUT} is $\Delta I_L/2$ or less, I_{Lmin} becomes less than 0. That is, the IC discharges the electrical charge in CL to the IC side until the IL changes from I_{Lmin} to 0 during t_{on} time, and the IL changes from 0 to I_{Lmin} during t_{off} time.



By setting the Mode pin to L, the IC automatically switches into VFM mode for high efficiency when output load is light. Under VFM mode, t_{on} indicates the time until the IC reaches to the pre-set I_{Lmax} . With the RP505K Series, I_{Lmax} during VFM control is pre-set to 280mA or so. However, even if the IC is not reached to I_{Lmax} yet, t_{on} turns off when it becomes around 1.5 times of $T=1/f_{osc}$.



Output Current and Selection of External Components

The relation between the output current and external components is as follows:

(Wherein, Ripple Current p-p value is described as I_{RP} , ON resistance of Pch Tr. and Nch Tr. of L_x are respectively described as R_{ONP} and R_{ONN} , and the DC resistor of the inductor is described as R_L .)

When Pch Tr. of L_x is ON:

$$V_{IN} = V_{OUT} + (R_{ONP} + R_L) \times I_{OUT} + L \times I_{RP} / t_{on} \dots\dots\dots \text{Equation 3}$$

When Pch Tr. of L_x is "OFF" (Nch Tr. is "ON"):

$$L \times I_{RP} / t_{off} = R_{ONN} \times I_{OUT} + V_{OUT} + R_L \times I_{OUT} \dots\dots\dots \text{Equation 4}$$

Put Equation 4 to Equation 3 and solve for ON duty of Pch transistor, $D_{ON} = t_{on} / (t_{off} + t_{on})$,

$$D_{ON} = (V_{OUT} + R_{ONN} \times I_{OUT} + R_L \times I_{OUT}) / (V_{IN} + R_{ONN} \times I_{OUT} - R_{ONP} \times I_{OUT}) \dots\dots\dots \text{Equation 5}$$

Ripple Current is as follows:

$$I_{RP} = (V_{IN} - V_{OUT} - R_{ONP} \times I_{OUT} - R_L \times I_{OUT}) \times D_{ON} / f_{osc} / L \dots\dots\dots \text{Equation 6}$$

wherein, peak current that flows through L , and L_x Tr. is as follows:

$$I_{Lxmax} = I_{OUT} + I_{RP} / 2 \dots\dots\dots \text{Equation 7}$$

*Consider I_{Lxmax} , condition of input and output and select external components.

*The above explanation is directed to the calculation in an ideal case in continuous mode.

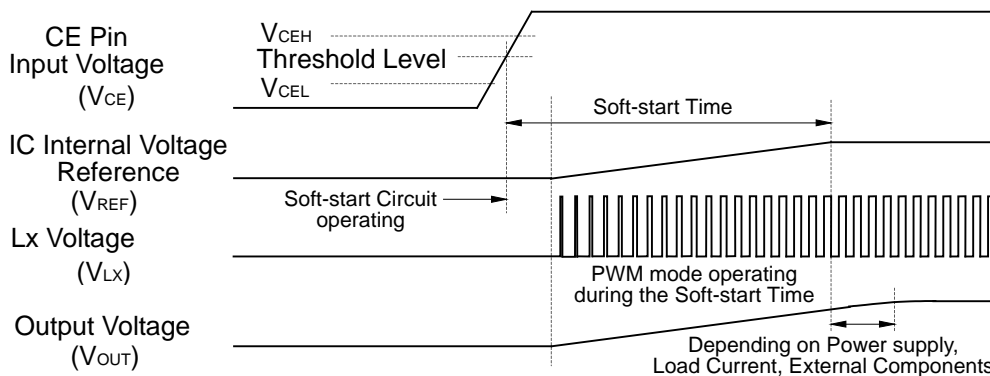
TIMING CHART

(1) Soft-start Time

- In the case of starting this IC with CE

In the case of starting this IC with CE, the operation can be as in the timing chart below.

When the voltage of CE pin (V_{CE}) is beyond the threshold level, the operation of the IC starts. The threshold voltage of CE pin is in between CE "H" input voltage (V_{CEH}) and CE "L" input voltage (V_{CEL}) described in the electrical characteristics table. Soft-start circuit operates, and after the certain time, the reference voltage inside the IC (V_{REF}) is rising gradually up to the constant value.



Soft-start time is the time interval from soft-start circuit starting point to the reference voltage level reaching point up to this constant level.

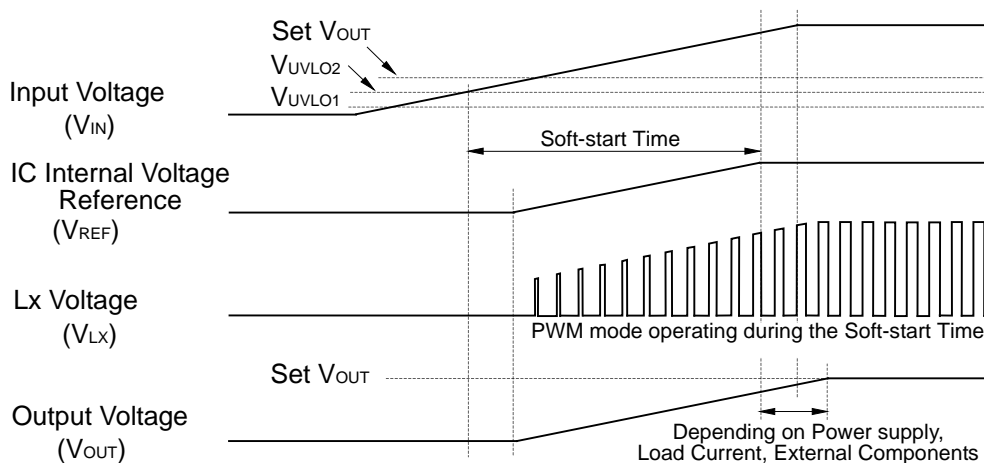
*Soft-start time is not always equal to the turn-on speed of DC/DC converter.

The power supply capacity for this IC, load current, inductance and capacitance values affect the turn-on speed.

- In the case of starting with power supply

In the case of starting with power supply, when the input voltage (V_{IN}) is larger than UVLO released voltage (V_{UVLO2}), soft-start circuit operates, and after that, the same explanation above is applied to the operation.

Soft-start time is the time interval from soft-start circuit starting point to the reference voltage level reaching point up to this constant level.



*Turn-on speed is affected by next conditions;

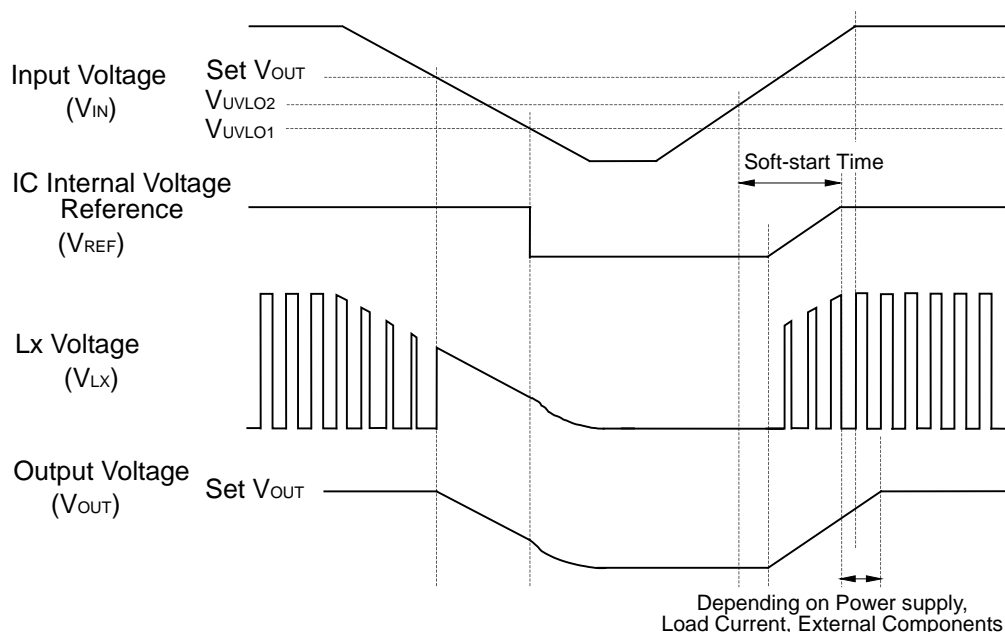
(a) Input Voltage (V_{IN}) rising speed depending on the power supplier to the IC and input capacitor C_{IN} .

(b) Output Capacitor C_{OUT} value and load current value.

(2) Under Voltage Lockout (UVLO) Circuit

The step-down DC/DC converter stops and ON duty becomes 100%, if input voltage (V_{IN}) becomes less than the set output voltage (Set V_{OUT}), the output voltage (V_{OUT}) gradually drops according to the input voltage (V_{IN}). If the input voltage drops more and becomes less than UVLO detector threshold (V_{UVLO1}), the under voltage lockout circuit (UVLO) operates, the IC internal reference voltage (V_{REF}) stops, switching transistors turn off and the output voltage drops according to the load and output capacitor C_{OUT} value.

To restart the normal operation, the input voltage (V_{IN}) must be more than the UVLO released voltage (V_{UVLO2}). The timing chart below describes the operation with varying the input voltage (V_{IN}).



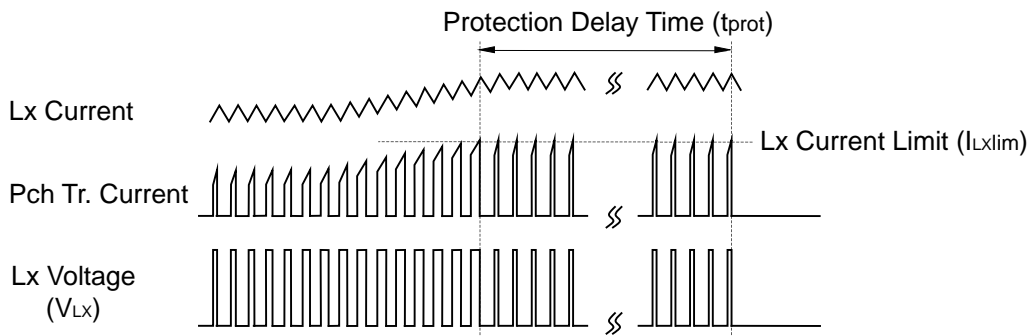
*Actually, the waveform of V_{OUT} at UVLO working and releasing varies depending on the initial voltage of C_{OUT} and load current situation.

(3) Over Current Protection Circuit, Latch Type Protection Circuit

Over current protection circuit supervises the inductor peak current (the current flowing through Pch transistor) in each switching cycle, and if the current exceeds the L_x current limit (I_{Lxlim}), turns off Pch transistor. The L_x current limit of RP505K is Typ.1700mA.

Latch type protection circuit latches the built-in driver to the OFF state and stops the operation of DC/DC converter if the over current status continues or the output voltage continues being the half of the setting voltage for equal or longer than protection delay time (t_{prot}).

*L_x current limit (I_{Lxlim}) and protection delay time (t_{prot}) could be easily affected by self-heating or ambient environment. If the input voltage (V_{IN}) drops drastically or becomes unstable due to short-circuit, the protection operation and protection delay time may be affected.

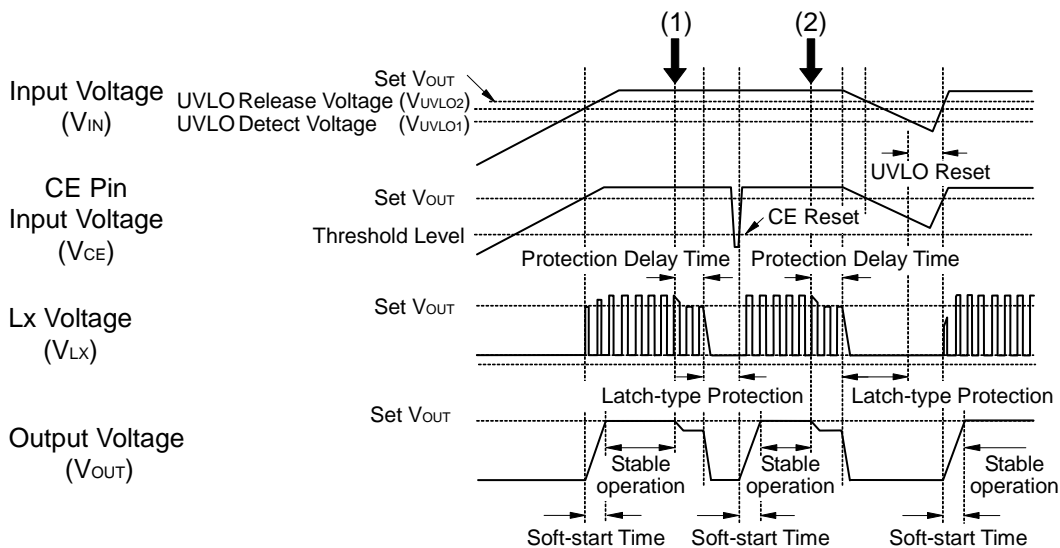


To release the condition of latch type protection, restart this IC by inputting "L" signal to CE pin, or restart this IC with power-on or make the supply voltage lower than UVLO detector threshold (V_{UVLO1}) level.

The timing chart shown below describes the changing process of input voltage rising, stable operating, operating with large current, reset with CE pin, stable operating, input voltage falling, input voltage recovering, and stable operating.

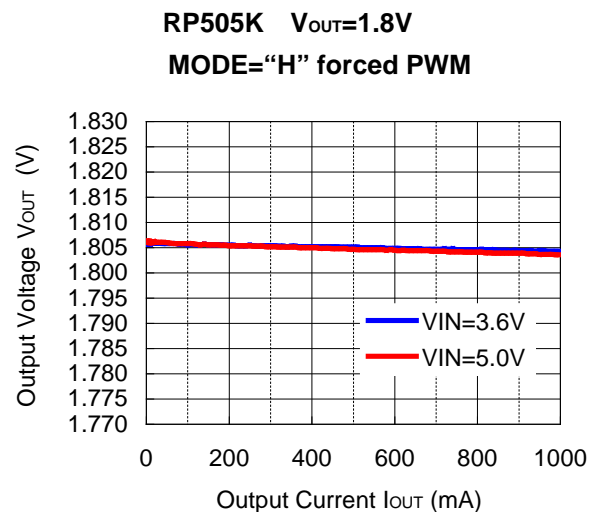
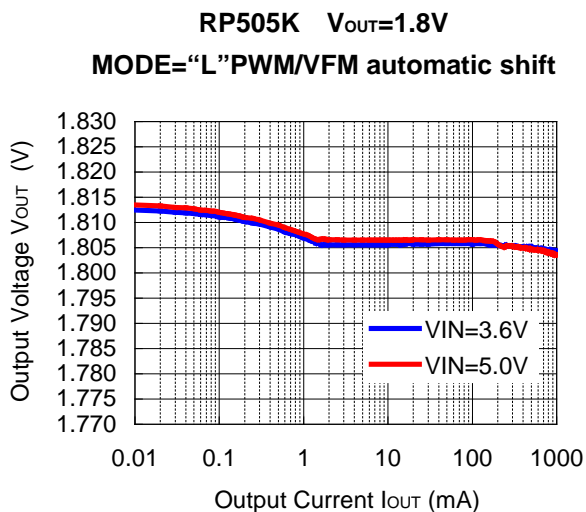
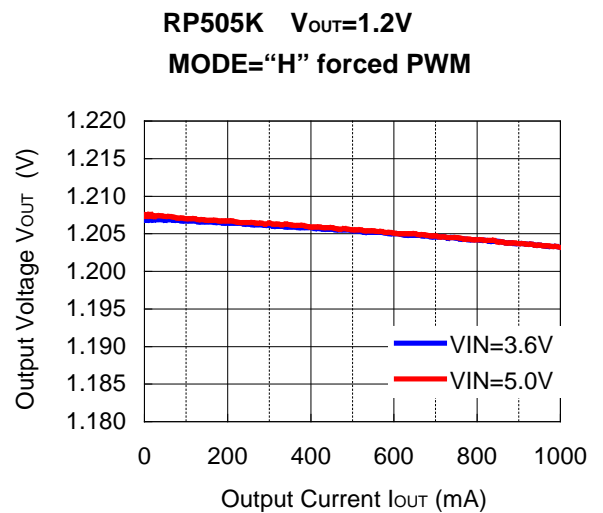
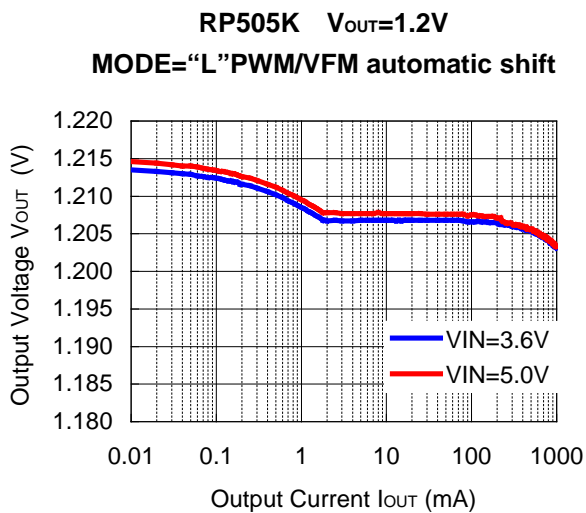
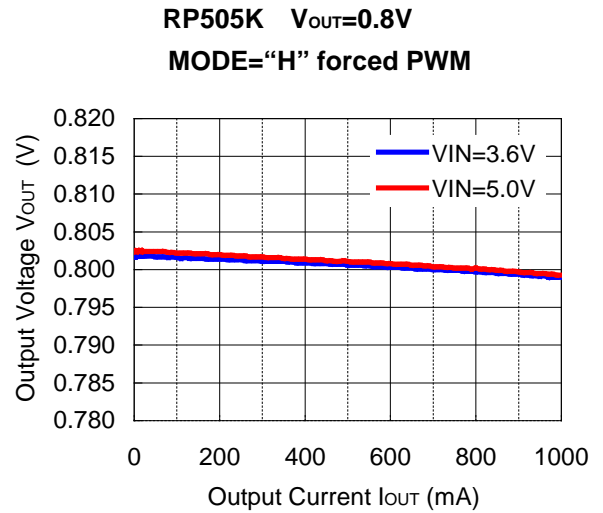
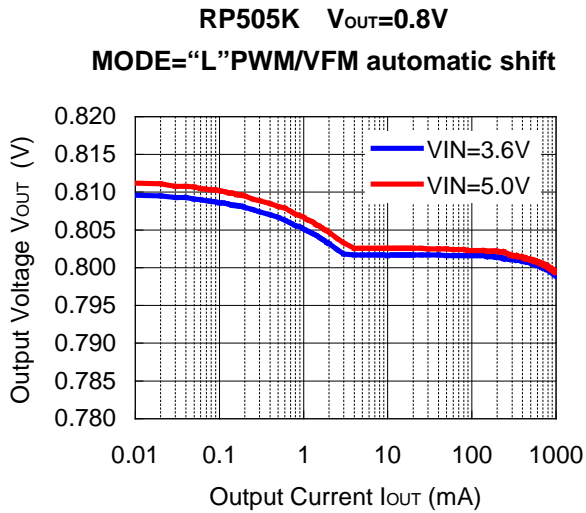
Point(1) : If the large current flows through the circuit or the IC goes into low output voltage condition due to short-circuit or other reasons, the latch type protection circuit latches the built-in driver to OFF state after the protection delay time (t_{prot}). Then, V_{Lx} becomes "L" and the output voltage turns OFF. In this timing chart below, the latch protective circuit can be released by once putting the IC into "L" with the CE pin and then into "H" again.

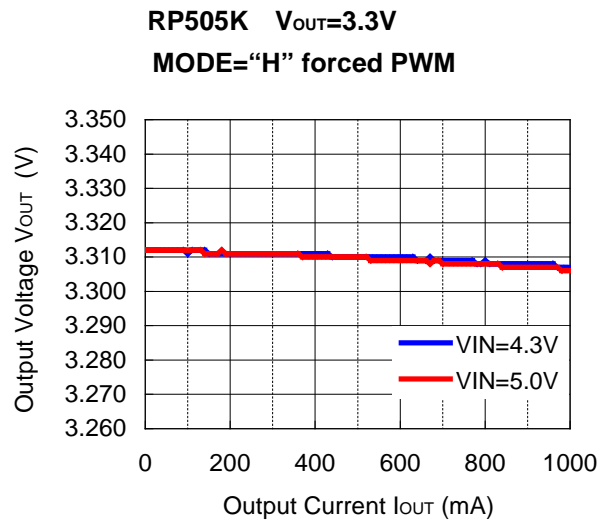
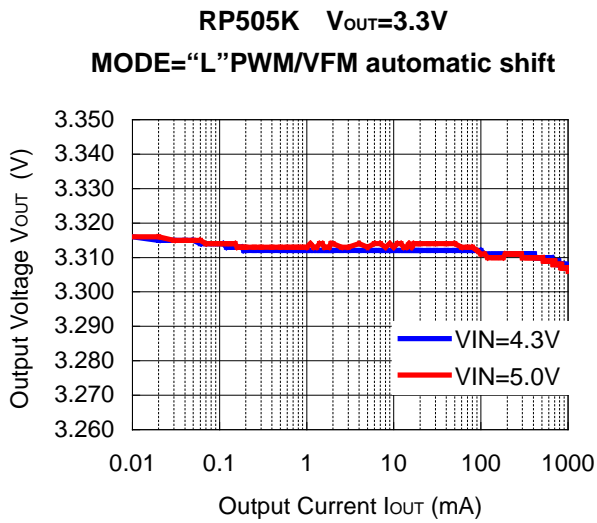
Point(2) : The latch type protection can be released by UVLO reset by making the input voltage lower than the UVLO detector threshold(V_{UVLO1})



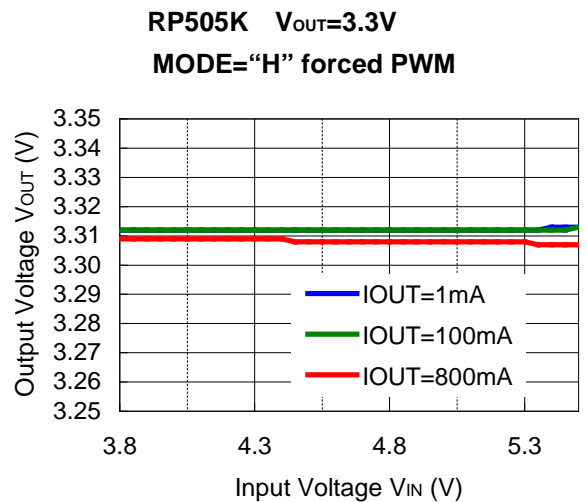
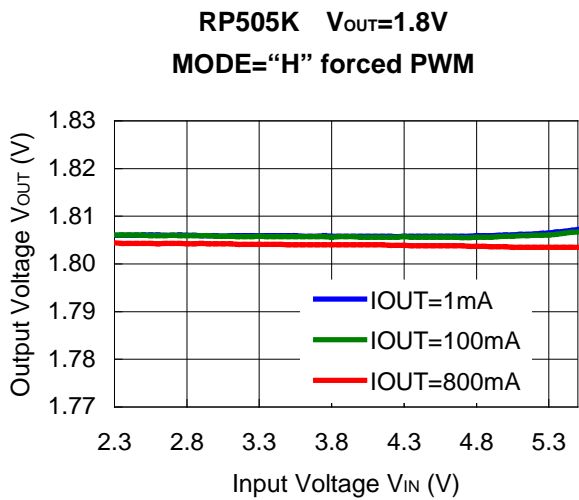
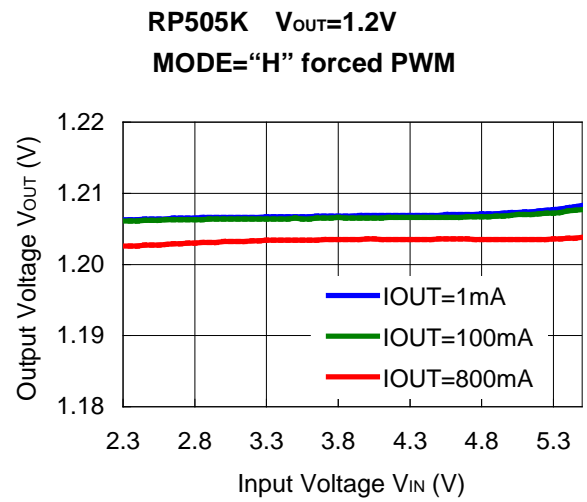
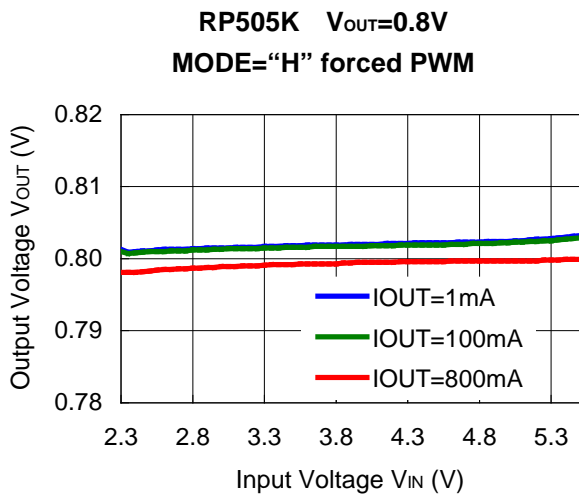
TYPICAL CHARACTERISTICS

1) Output Voltage vs. Output Current



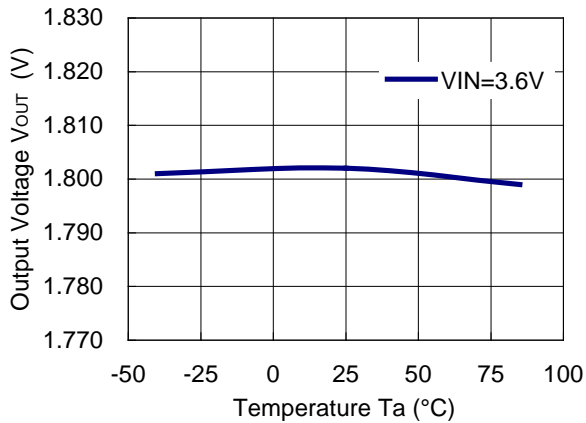


2) Output Voltage vs. Input Voltage



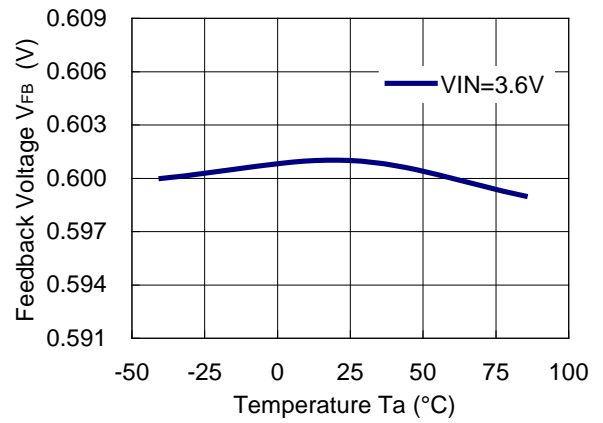
3) Output Voltage vs. Temperature

RP505K181A/B



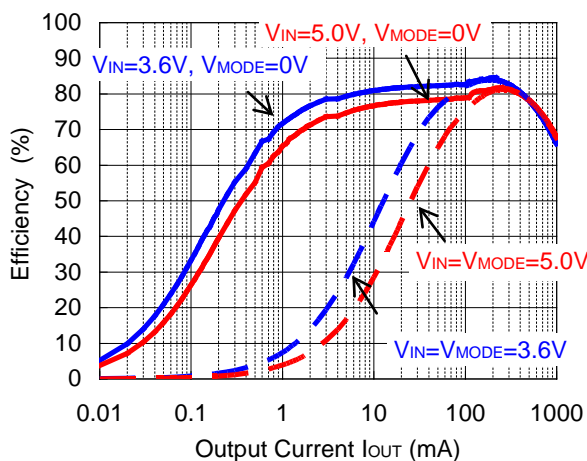
4) Feedback Voltage vs. Temperature

RP505K001C

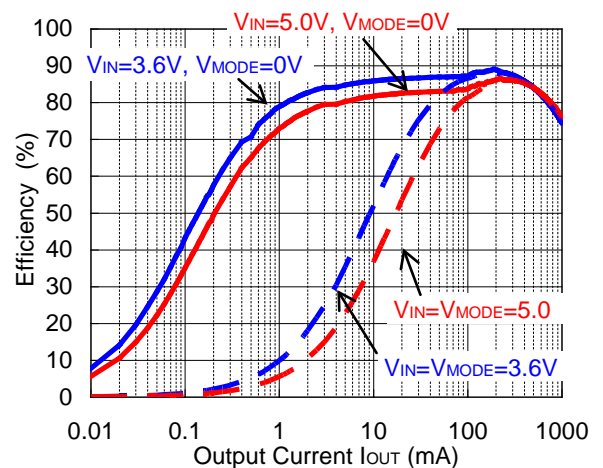


5) Efficiency vs. Output Current

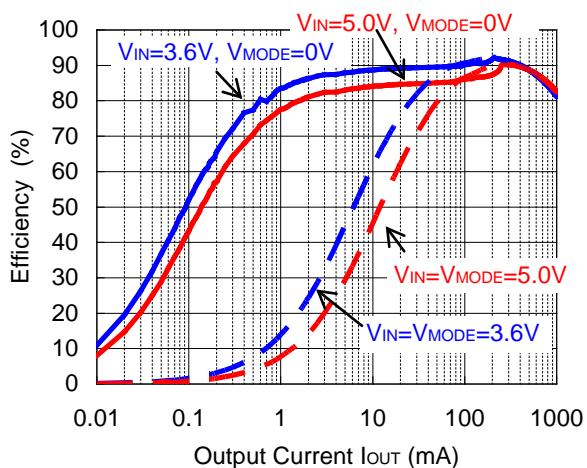
RP505K V_{OUT}=0.8V



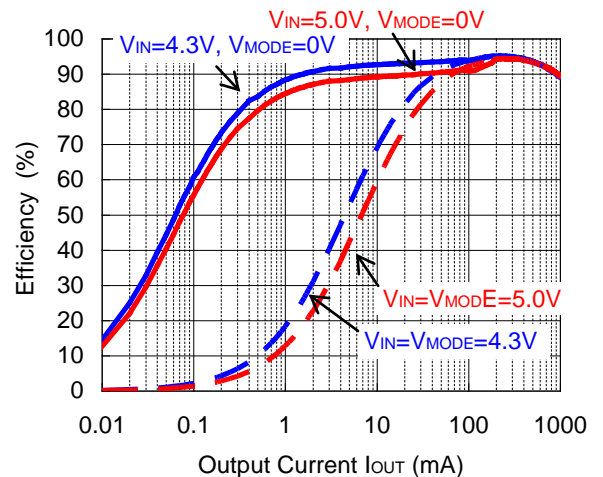
RP505K V_{OUT}=1.2V



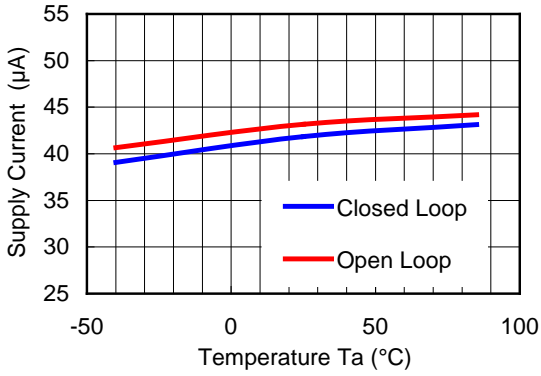
RP505K V_{OUT}=1.8V



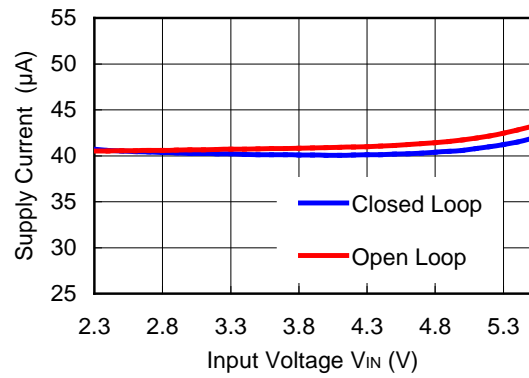
RP505K V_{OUT}=3.3V



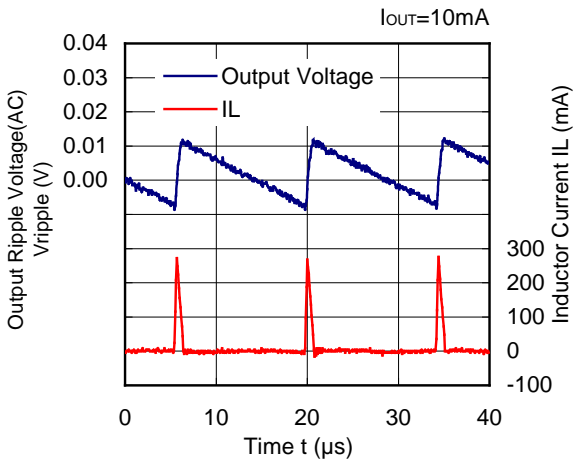
6) Supply Current vs. Temperature
 RP505K $V_{OUT}=1.8V(V_{IN}=5.5V)$
 MODE="L" PWM/VFM automatic shift



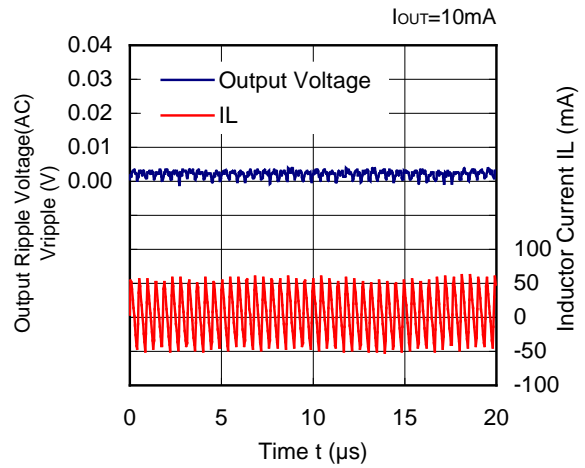
7) Supply Current vs. Input Voltage
 RP505K $V_{OUT}=1.8V$
 MODE="L" PWM/VFM automatic shift



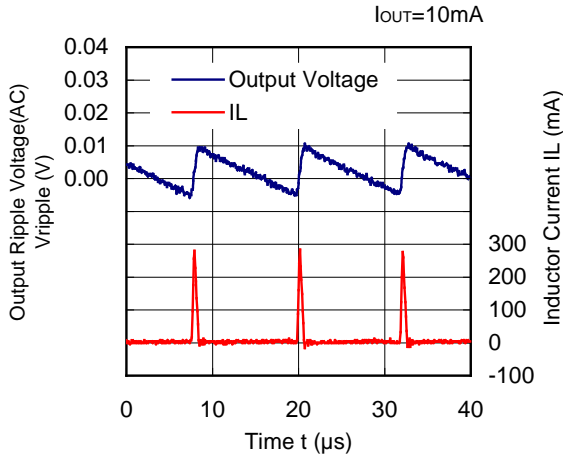
8) DC/DC Output Waveform
 RP505K $V_{OUT}=0.8V(V_{IN}=3.6V)$
 MODE="L" PWM/VFM automatic shift



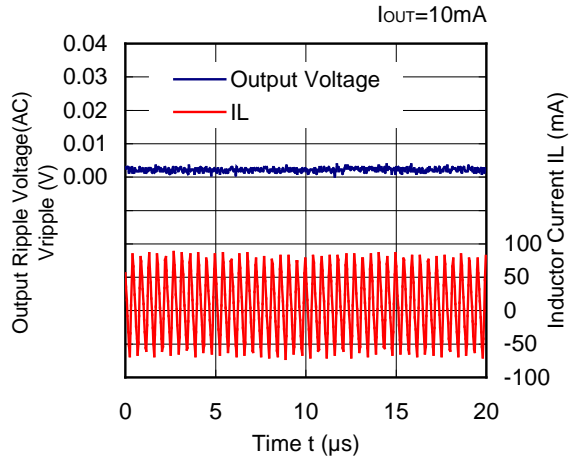
RP505K $V_{OUT}=0.8V(V_{IN}=3.6V)$
 MODE="H" forced PWM



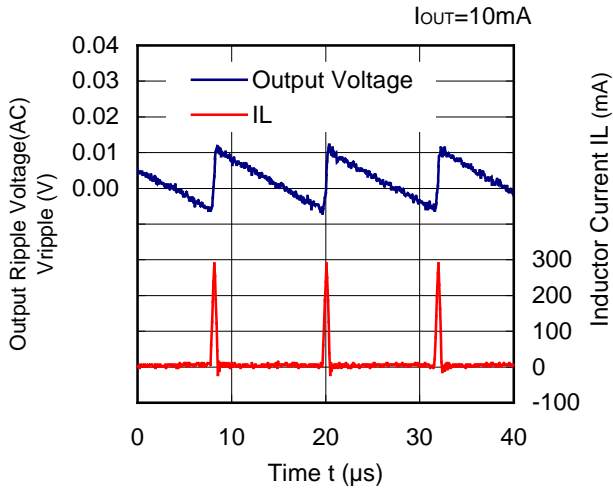
RP505K $V_{OUT}=1.2V(V_{IN}=3.6V)$
 MODE="L" PWM/VFM automatic shift



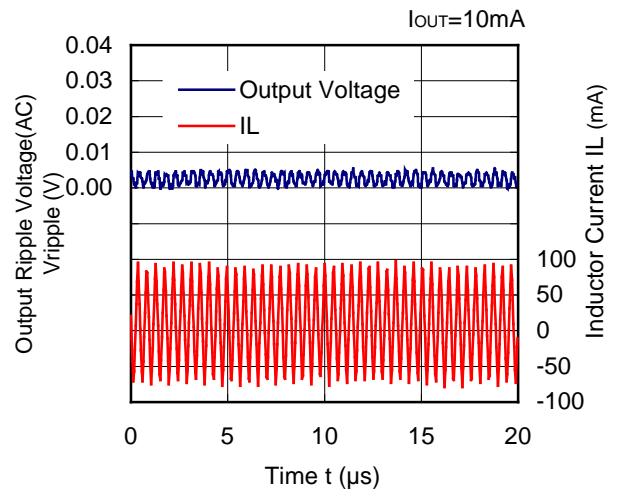
RP505K $V_{OUT}=1.2V(V_{IN}=3.6V)$
 MODE="H" forced PWM



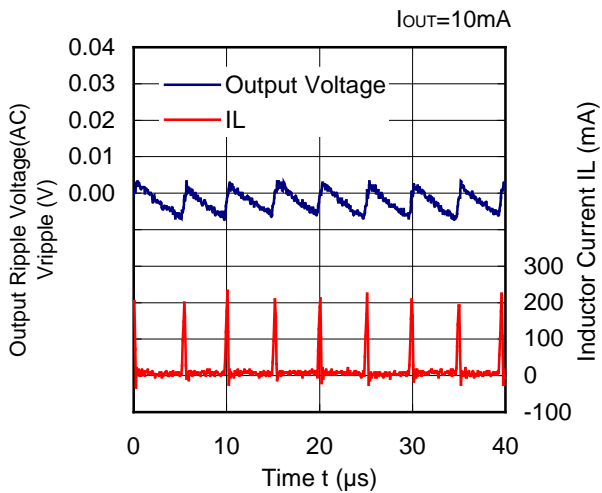
**RP505K $V_{OUT}=1.8V(V_{IN}=3.6V)$
MODE="L" PWM/VFM automatic shift**



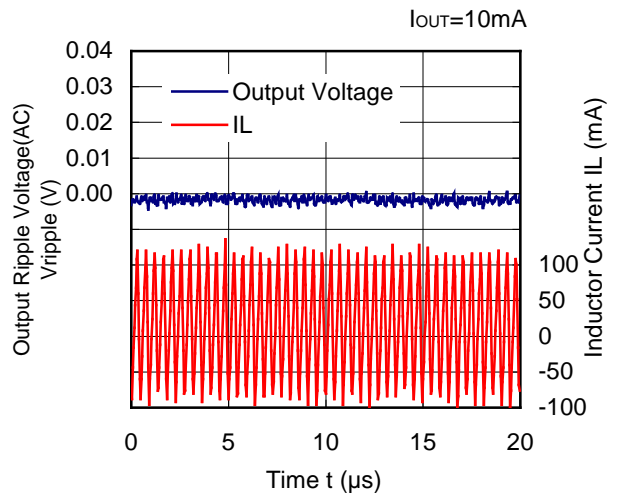
**RP505K $V_{OUT}=1.8V(V_{IN}=3.6V)$
MODE="H" forced PWM**



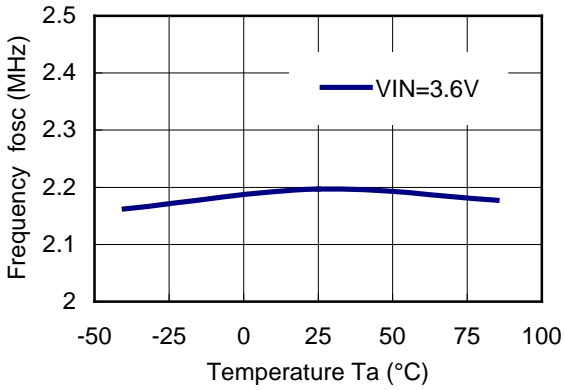
**RP505K $V_{OUT}=3.3V(V_{IN}=5.0V)$
MODE="L" PWM/VFM automatic shift**



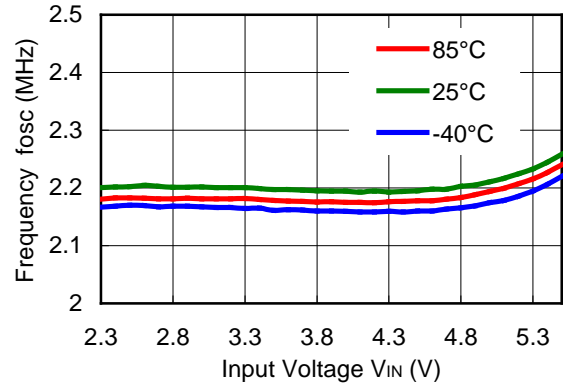
**RP505K $V_{OUT}=3.3V(V_{IN}=5.0V)$
MODE="H" forced PWM**



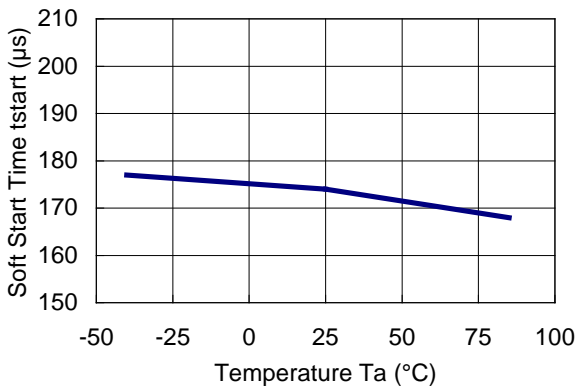
9) Oscillator Frequency vs. Temperature



10) Oscillator Frequency vs. Input Voltage

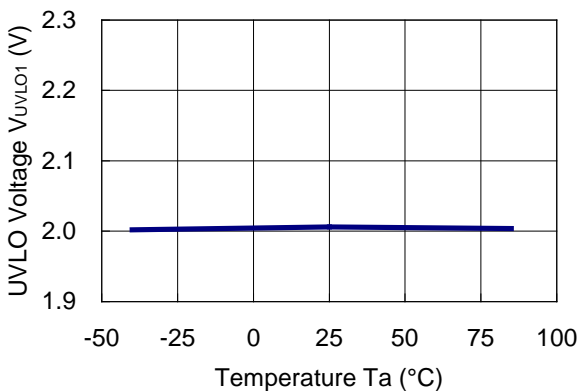


11) Soft-start Time vs. Temperature

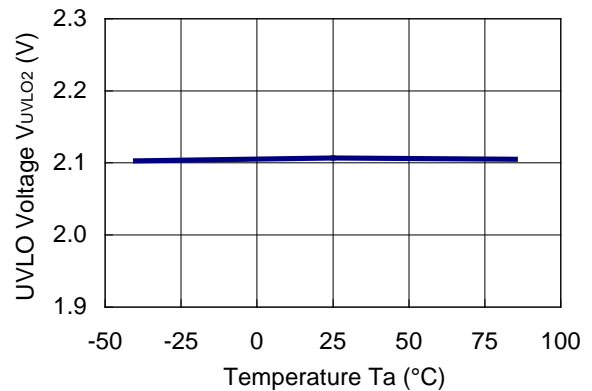


12) UVLO Detector Threshold / Released Voltage vs. Temperature

UVLO Detector Threshold

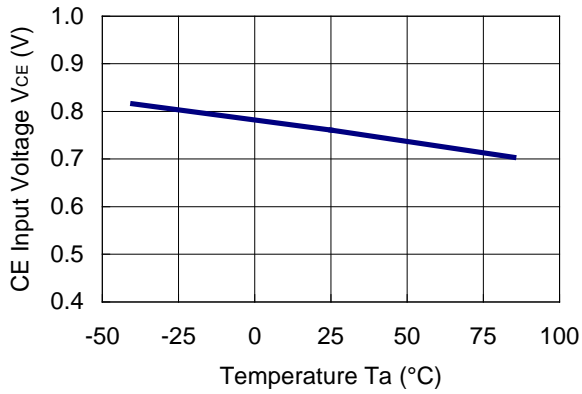


UVLO Released Voltage

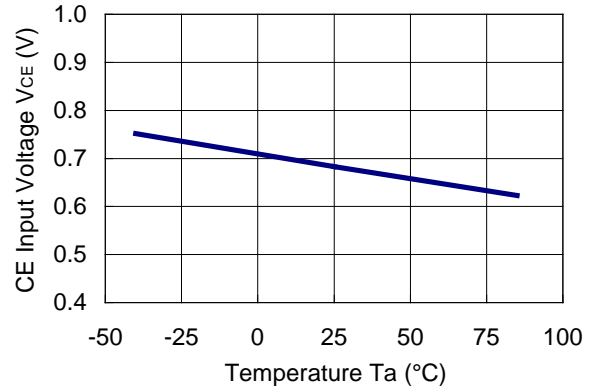


13) CE Input Voltage vs. Temperature

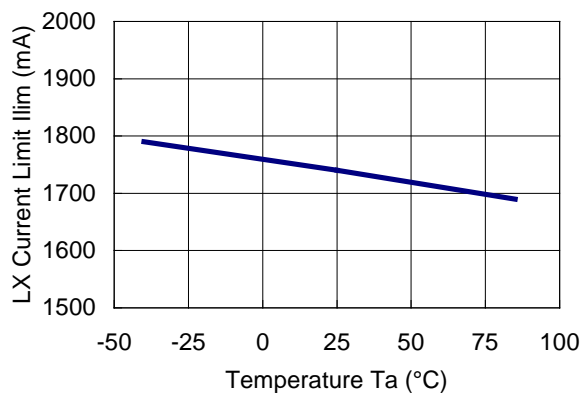
CE“H” Input Voltage ($V_{IN}=5.5V$)



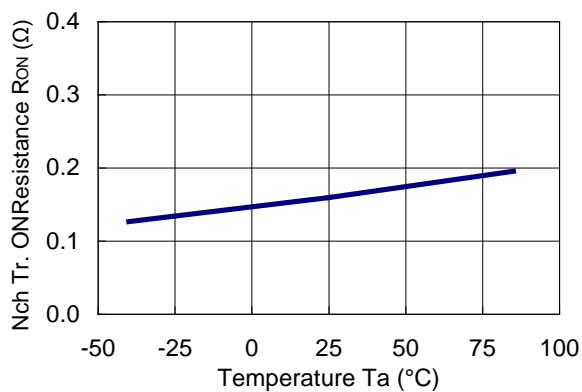
CE“L” Input Voltage ($V_{IN}=2.3V$)



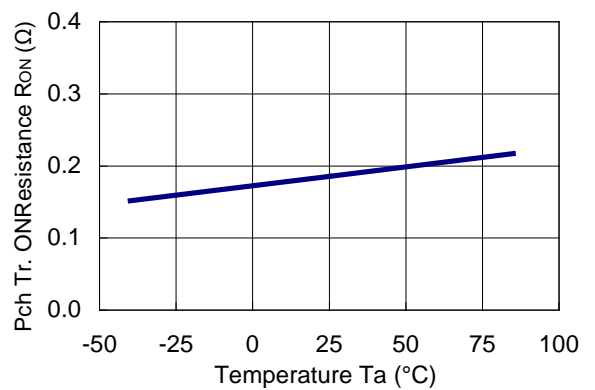
14) Lx Current Limit vs. Temperature



15) Nch Tr. ON Resistance vs. Temperature

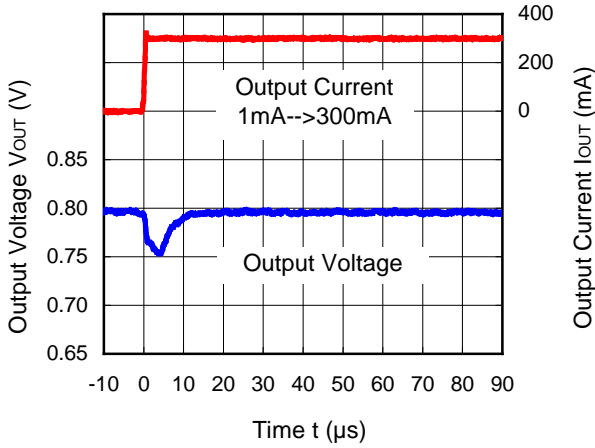


16) Pch Tr. ON Resistance vs. Temperature

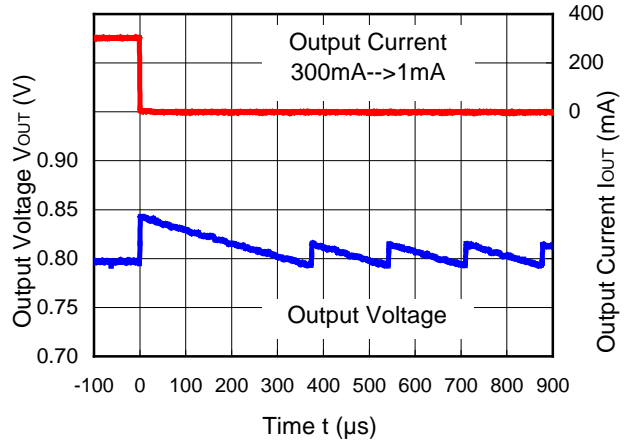


17) Load Transient Response

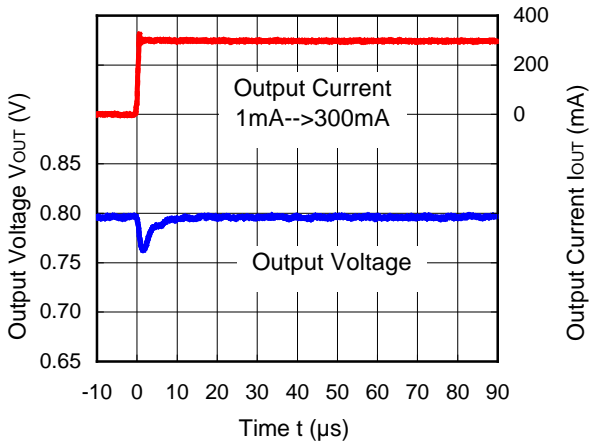
RP505K081A/B ($V_{IN}=3.6V$)
MODE="L" PWM/VFM automatic shift



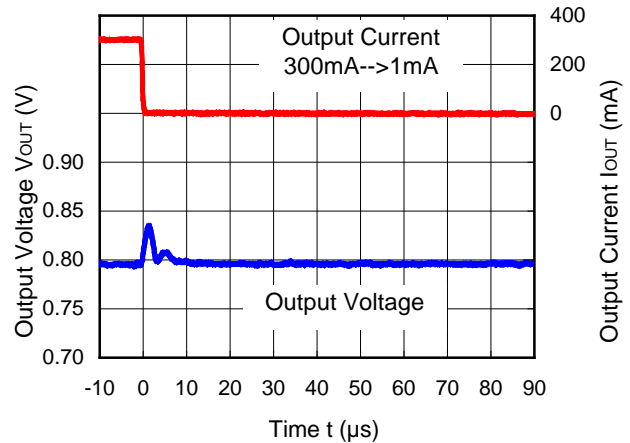
RP505K081A/B ($V_{IN}=3.6V$)
MODE="L" PWM/VFM automatic shift



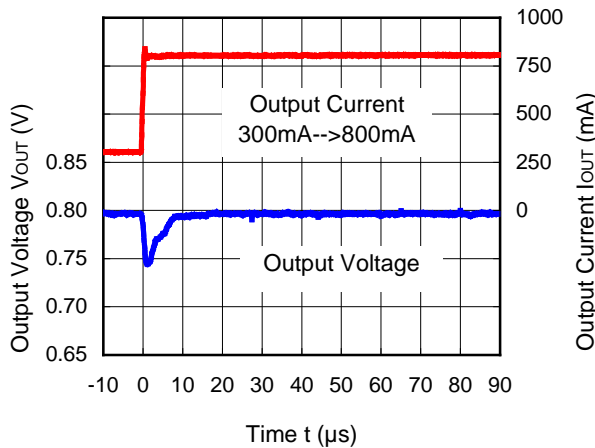
RP505K081A/B ($V_{IN}=3.6V$)
MODE="H" forced PWM



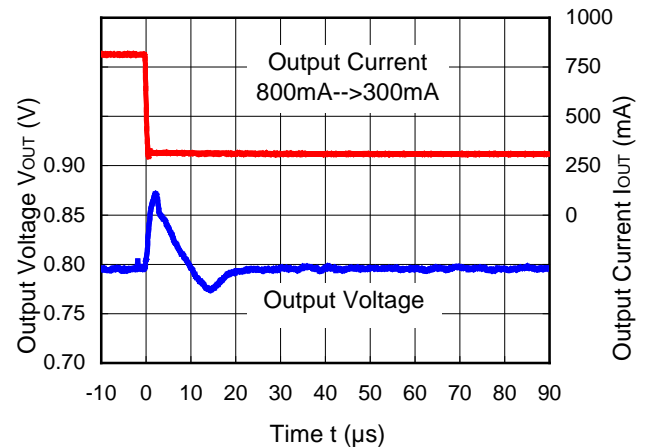
RP505K081A/B ($V_{IN}=3.6V$)
MODE="H" forced PWM



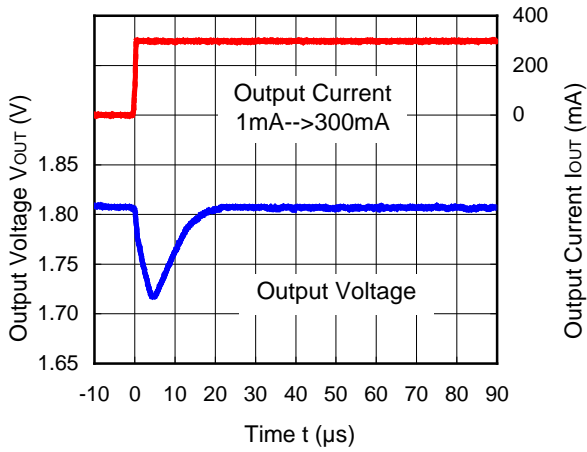
RP505K081A/B ($V_{IN}=3.6V$)



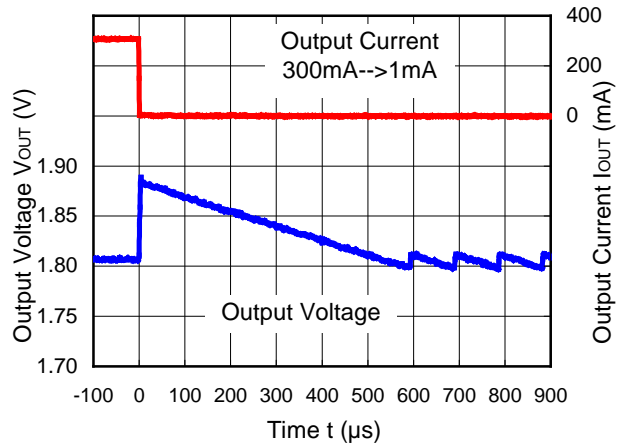
RP505K081A/B ($V_{IN}=3.6V$)



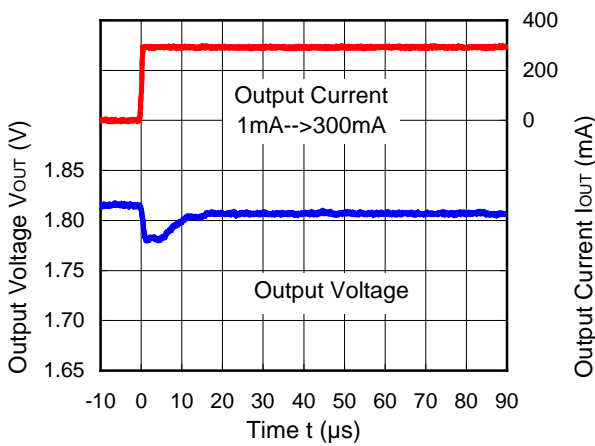
RP505K181A/B (V_{IN}=3.6V)
MODE="L" PWM/VFM automatic shift



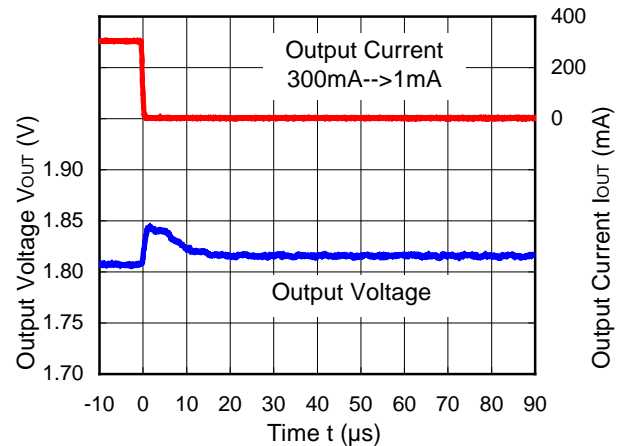
RP505K181A/B (V_{IN}=3.6V)
MODE="L" PWM/VFM automatic shift



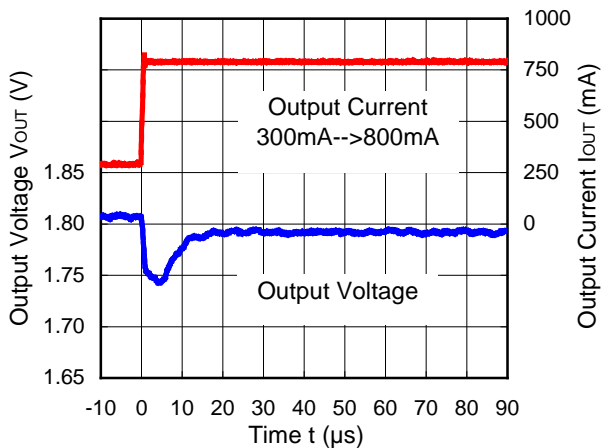
RP505K181A/B (V_{IN}=3.6V)
MODE="H" forced PWM



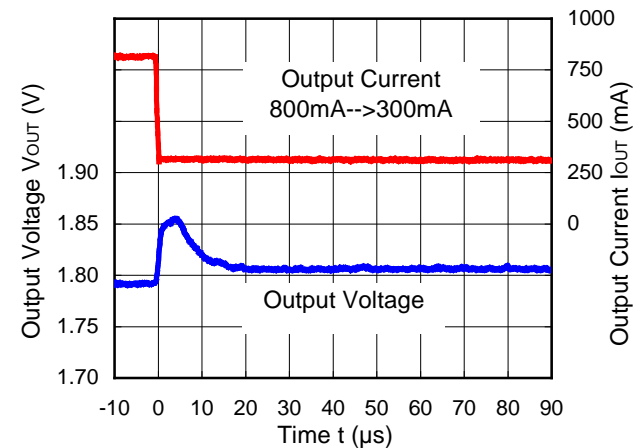
RP505K181A/B (V_{IN}=3.6V)
MODE="H" forced PWM



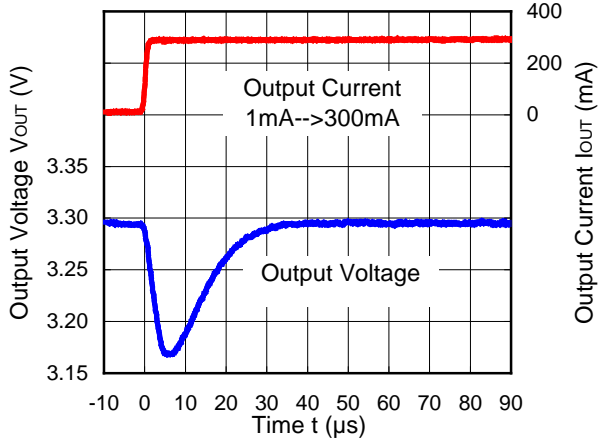
RP505K181A/B (V_{IN}=3.6V)



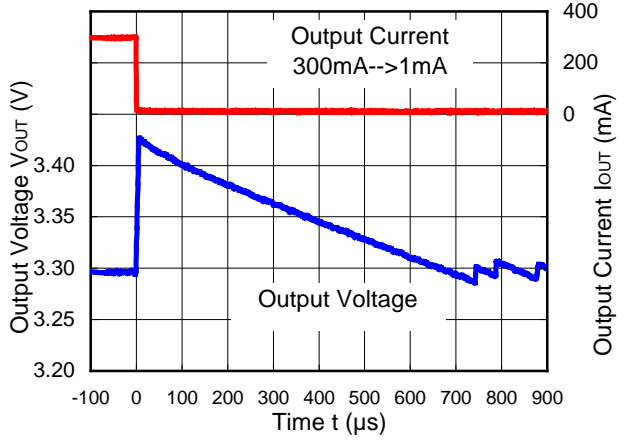
RP505K181A/B (V_{IN}=3.6V)



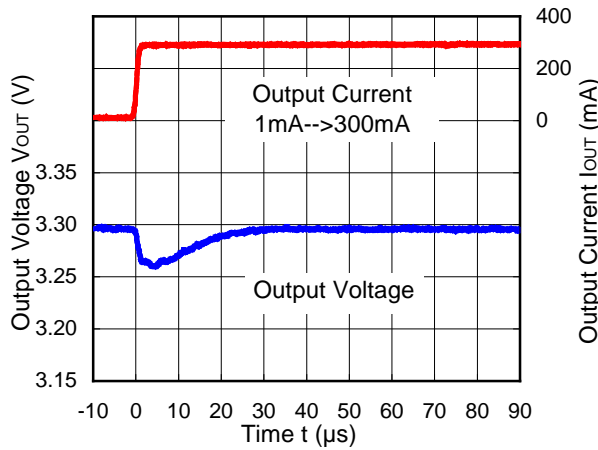
RP505K331A/B (V_{IN}=5.0V)
MODE="L" PWM/VFM automatic shift



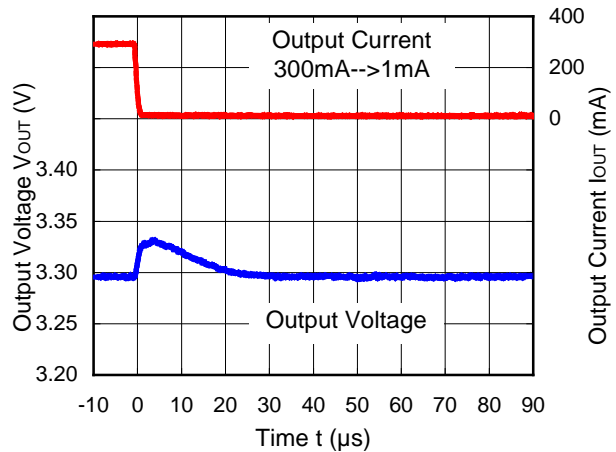
RP505K331A/B (V_{IN}=5.0V)
MODE="L" PWM/VFM automatic shift



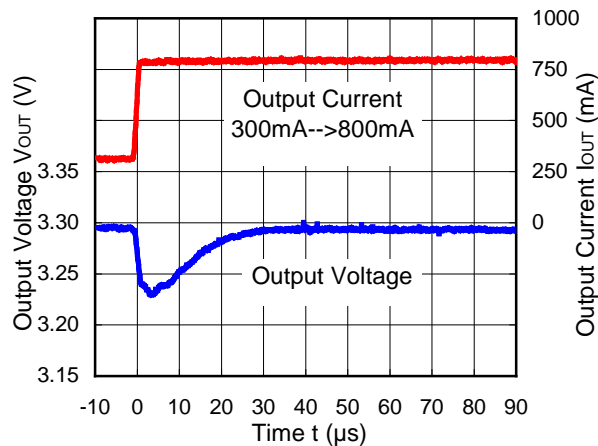
RP505K331A/B (V_{IN}=5.0V)
MODE="H" forced PWM



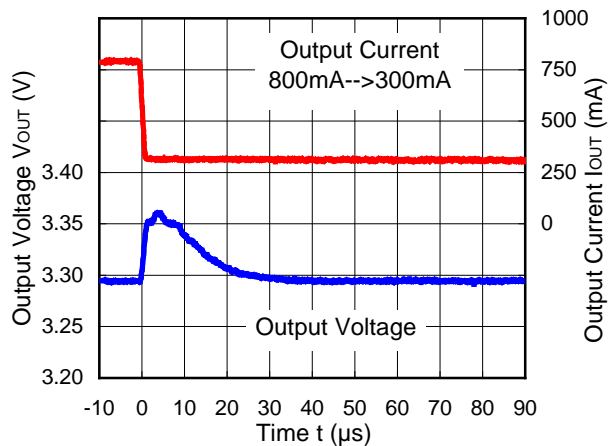
RP505K331A/B (V_{IN}=5.0V)
MODE="H" forced PWM



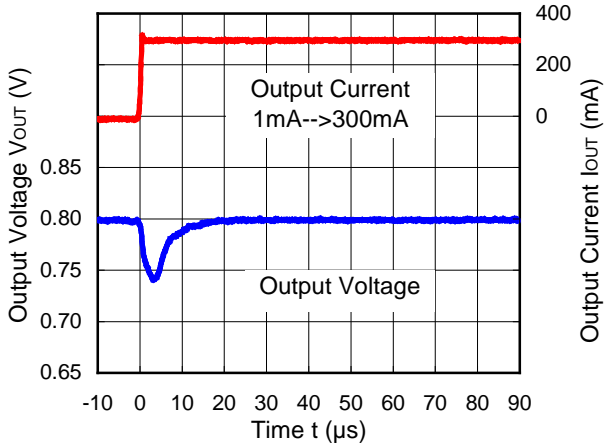
RP505K331A/B (V_{IN}=5.0V)



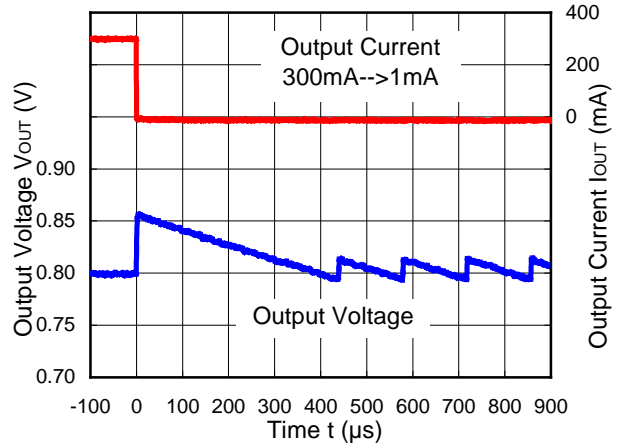
RP505K331A/B (V_{IN}=5.0V)



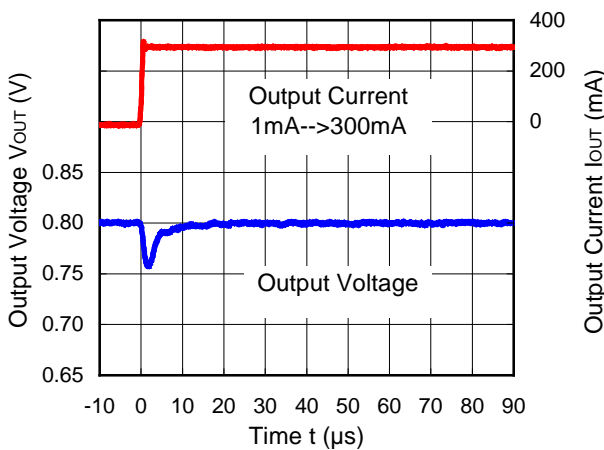
RP505K001C ($V_{IN}=3.6V$, $V_{OUT}=0.8V$)
 MODE="L" PWM/VFM automatic shift



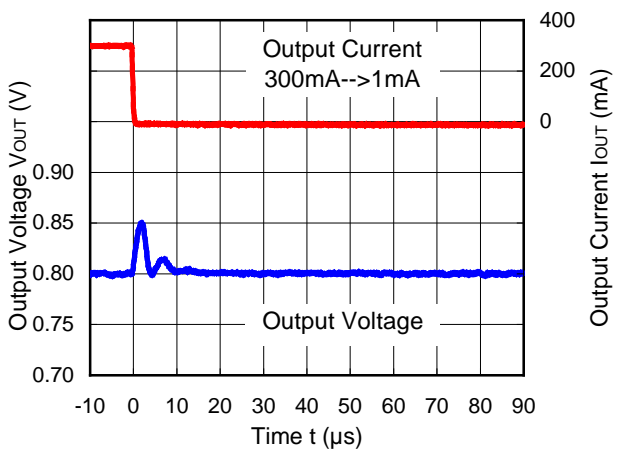
RP505K001C ($V_{IN}=3.6V$, $V_{OUT}=0.8V$)
 MODE="L" PWM/VFM モード自動切替え



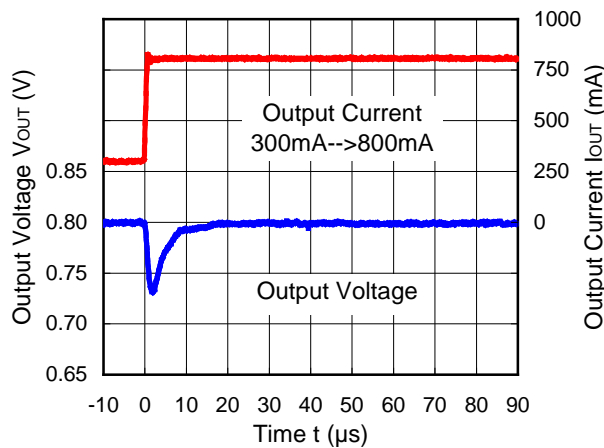
RP505K001C ($V_{IN}=3.6V$, $V_{OUT}=0.8V$)
 MODE="H" forced PWM



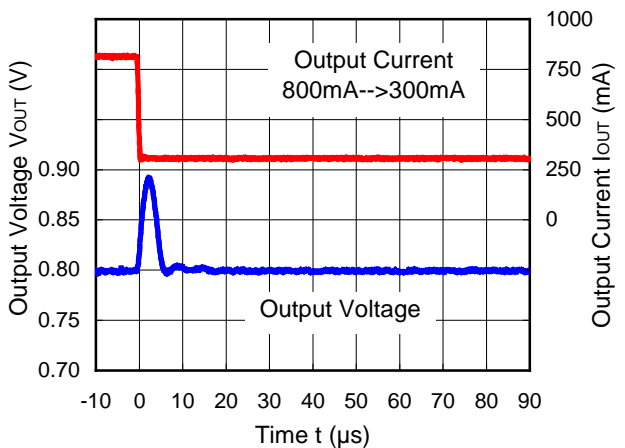
RP505K001C ($V_{IN}=3.6V$, $V_{OUT}=0.8V$)
 MODE="H" forced PWM



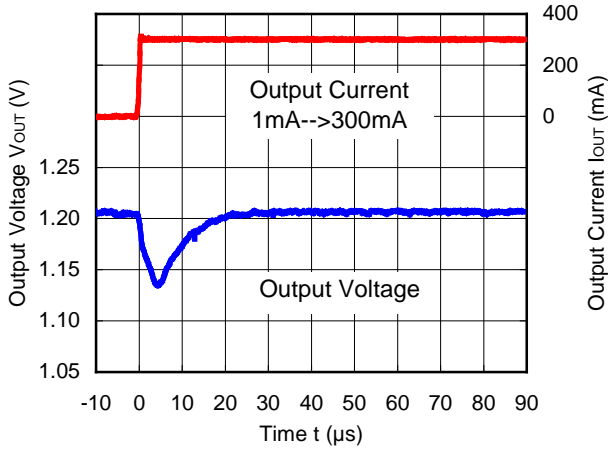
RP505K001C ($V_{IN}=3.6V$, $V_{OUT}=0.8V$)



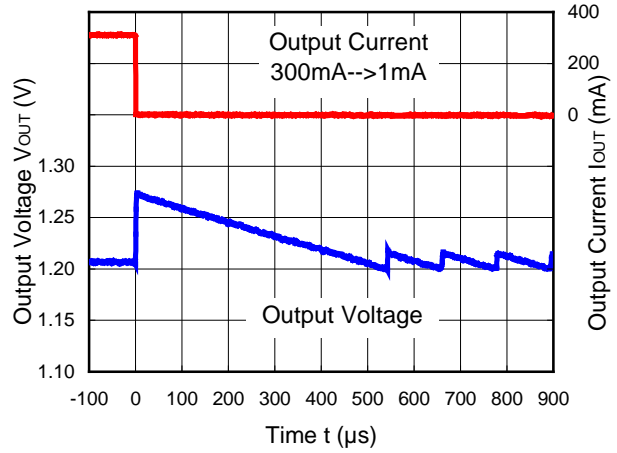
RP505K001C ($V_{IN}=3.6V$, $V_{OUT}=0.8V$)



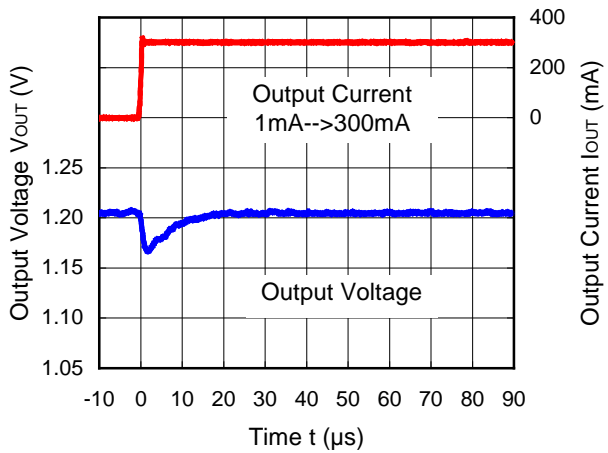
RP505K001C ($V_{IN}=3.6V$, $V_{OUT}=1.2V$)
 MODE="L" PWM/VFM automatic shift



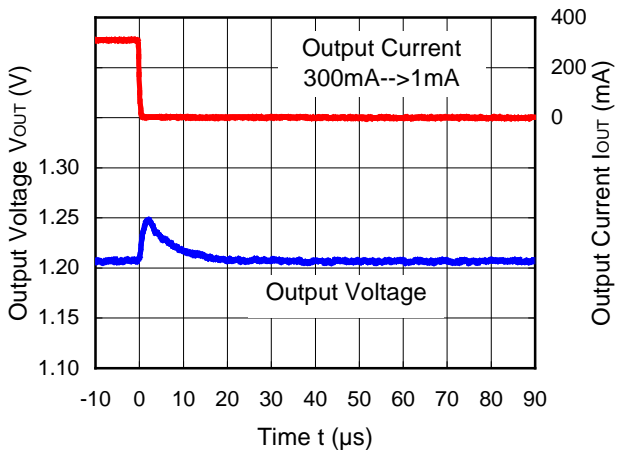
RP505K001C ($V_{IN}=3.6V$, $V_{OUT}=1.2V$)
 MODE="L" PWM/VFM automatic shift



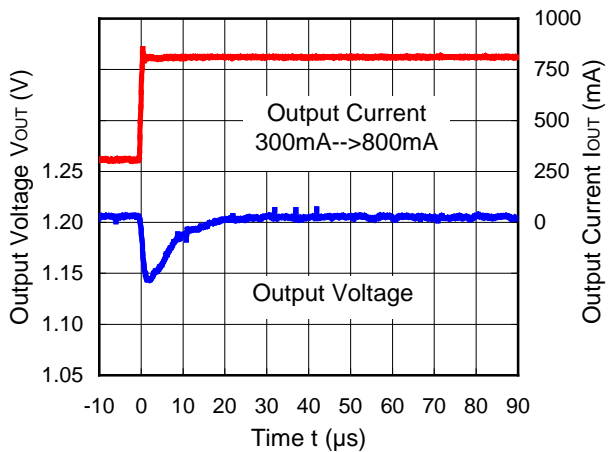
RP505K001C ($V_{IN}=3.6V$, $V_{OUT}=1.2V$)
 MODE="H" forced PWM



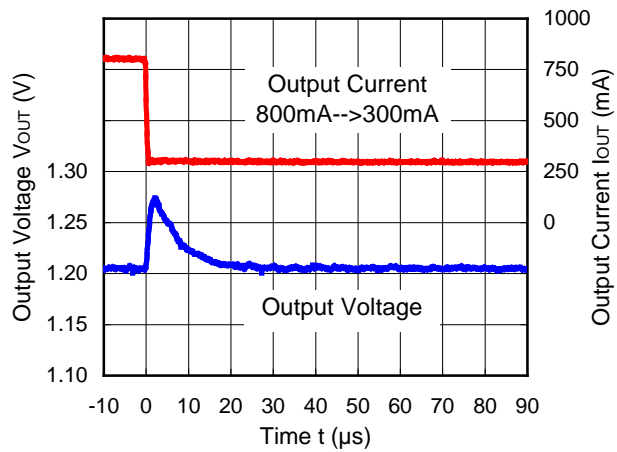
RP505K001C ($V_{IN}=3.6V$, $V_{OUT}=1.2V$)
 MODE="H" forced PWM



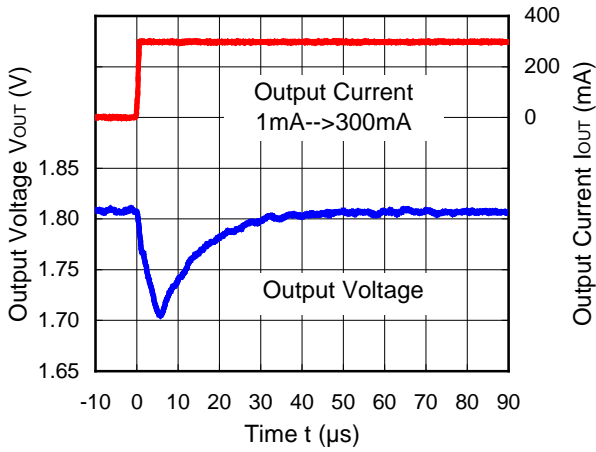
RP505K001C ($V_{IN}=3.6V$, $V_{OUT}=1.2V$)



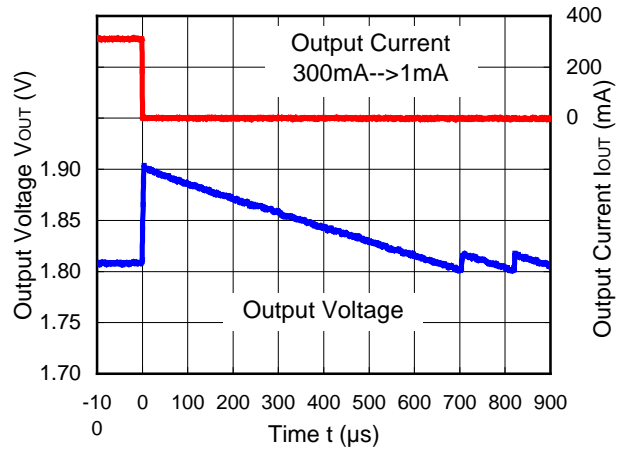
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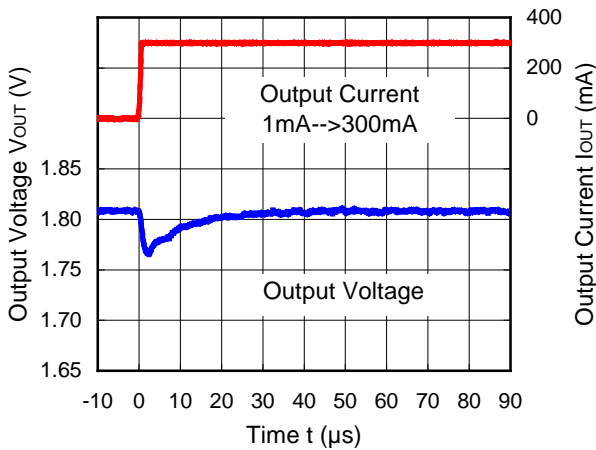
RP505K001C (V_{IN}=3.6V, V_{OUT}=1.8V)
MODE="L" PWM/VFM automatic shift



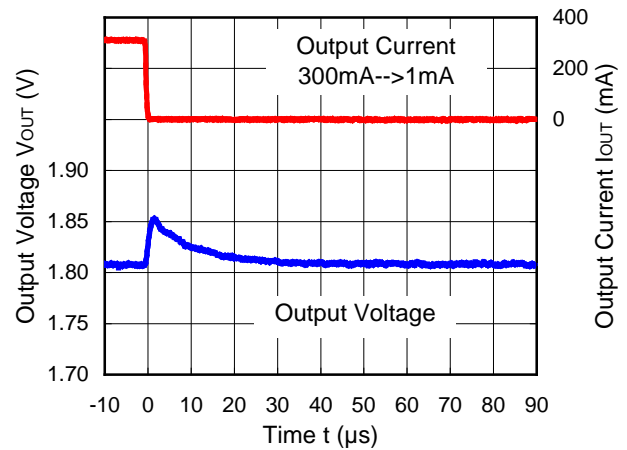
RP505K001C (V_{IN}=3.6V, V_{OUT}=1.8V)
MODE="L" PWM/VFM automatic shift



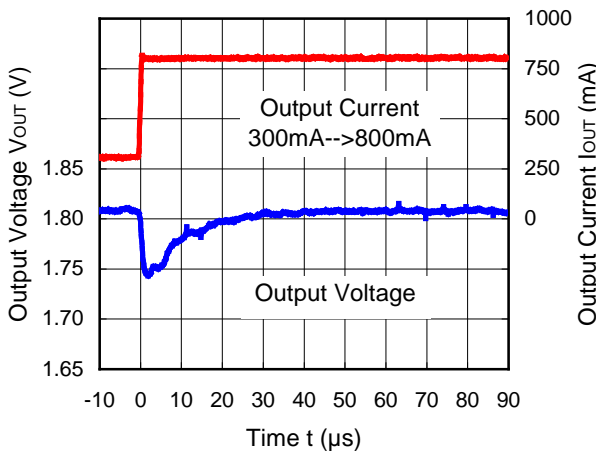
RP505K001C (V_{IN}=3.6V, V_{OUT}=1.8V)
MODE="H" forced PWM



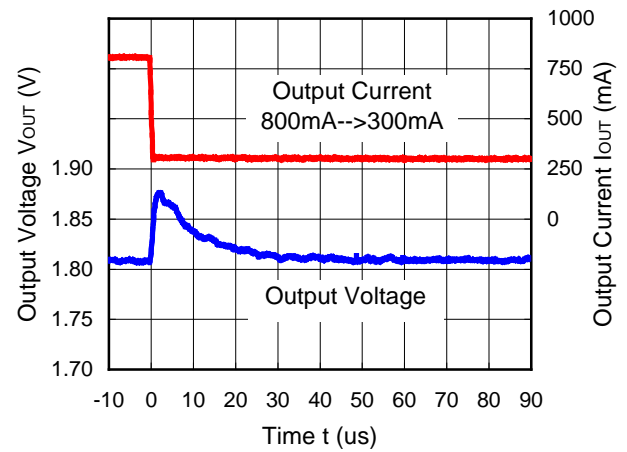
RP505K001C (V_{IN}=3.6V, V_{OUT}=1.8V)
MODE="H" forced PWM



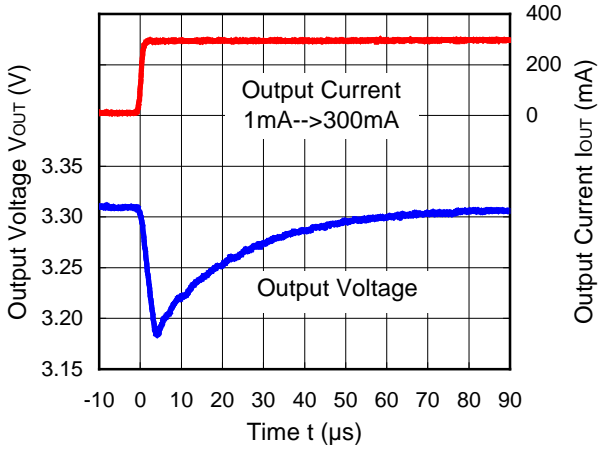
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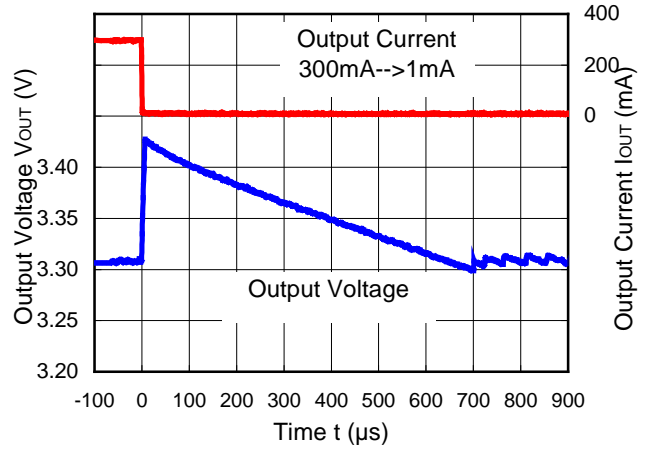
RP505K001C (V_{IN}=3.6V, V_{OUT}=1.8V)



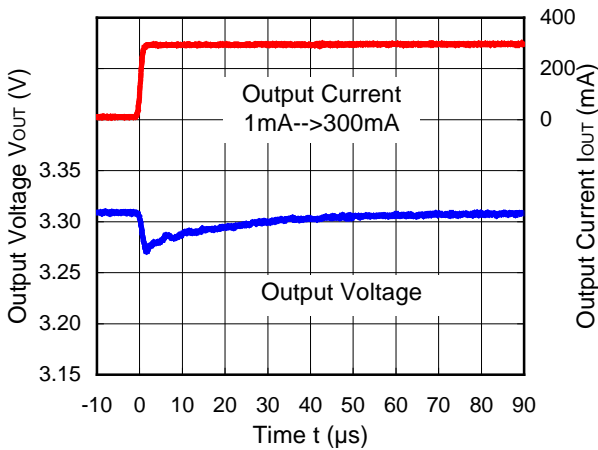
RP505K001C (V_{IN}=5.0V, V_{OUT}=3.3V)
MODE="L" PWM/VFM automatic shift



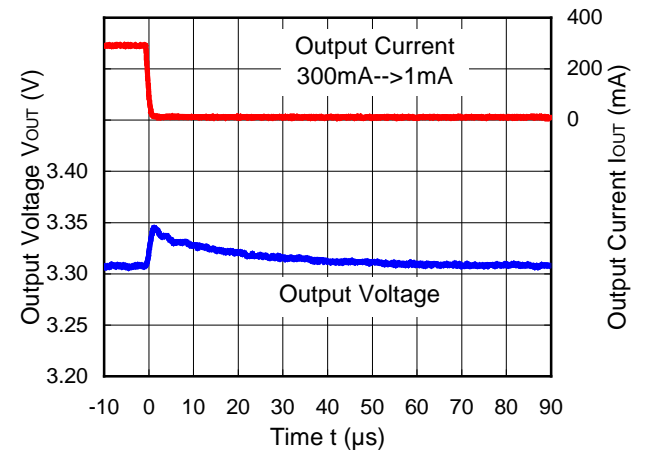
RP505K001C (V_{IN}=5.0V, V_{OUT}=3.3V)
MODE="L" PWM/VFM automatic shift



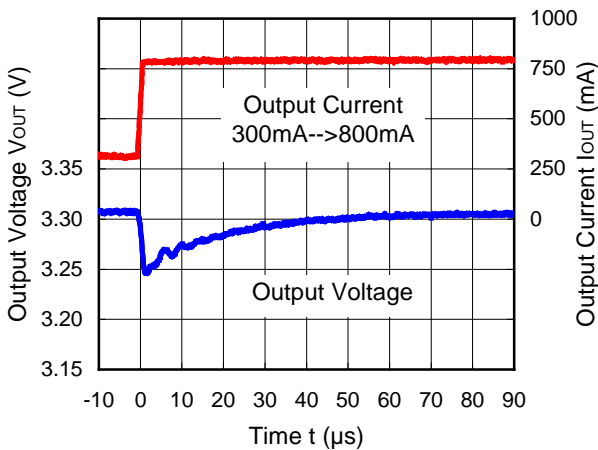
RP505K001C (V_{IN}=5.0V, V_{OUT}=3.3V)
MODE="H" forced PWM



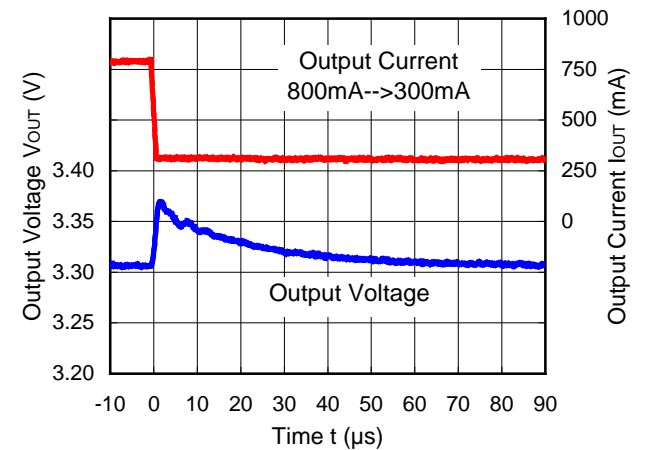
RP505K001C (V_{IN}=5.0V, V_{OUT}=3.3V)
MODE="H" forced PWM



RP505K001C (V_{IN}=5.0V, V_{OUT}=3.3V)

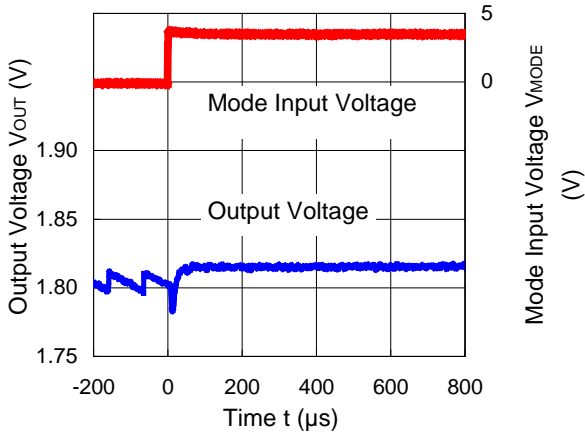


RP505K001C (V_{IN}=5.0V, V_{OUT}=3.3V)

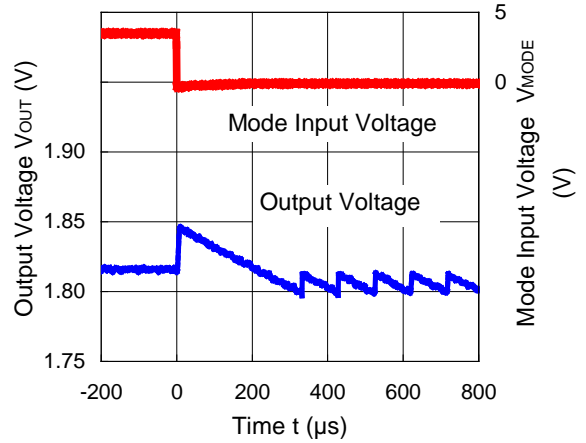


18) Mode Switching Waveform

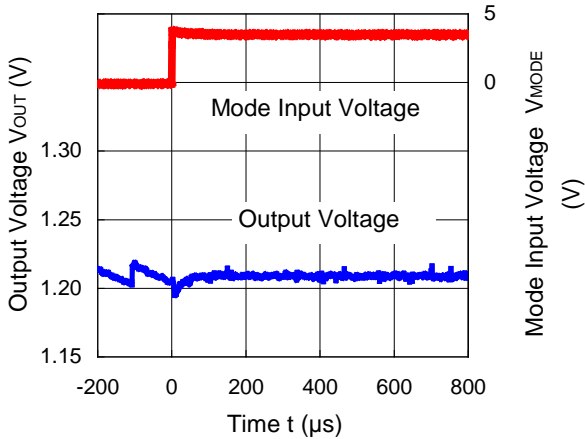
RP505K181A/B ($V_{IN}=3.6V$, $I_{OUT}=1mA$)
MODE="L" --> MODE="H"



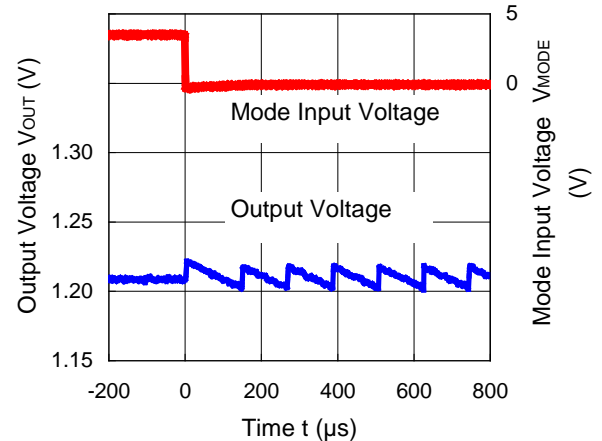
RP505K181A/B ($V_{IN}=3.6V$, $I_{OUT}=1mA$)
MODE="H" --> MODE="L"



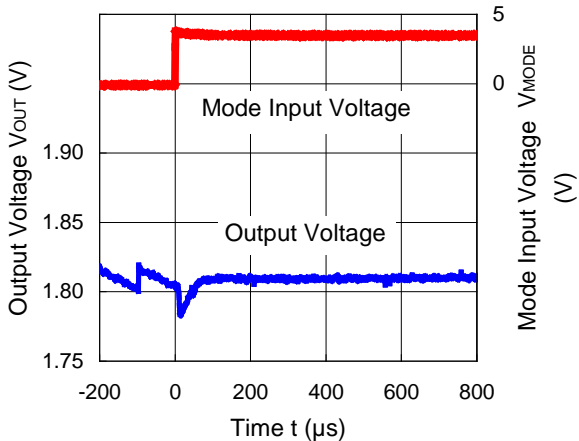
RP505K001C ($V_{IN}=3.6V$, $V_{OUT}=1.2V$, $I_{OUT}=1mA$)
MODE="L" --> MODE="H"



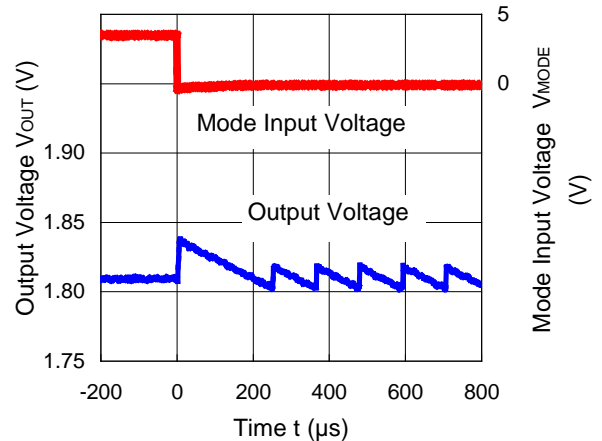
RP505K001C ($V_{IN}=3.6V$, $V_{OUT}=1.2V$, $I_{OUT}=1mA$)
MODE="H" --> MODE="L"



RP505K001C ($V_{IN}=3.6V$, $V_{OUT}=1.8V$, $I_{OUT}=1mA$)
MODE="L" --> MODE="H"



RP505K001C ($V_{IN}=3.6V$, $V_{OUT}=1.8V$, $I_{OUT}=1mA$)
MODE="H" --> MODE="L"





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