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## STEP-UP DC/DC CONVERTER FOR WHITE LED BACK LIGHT

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NO.EA-271-111123

### OUTLINE

The R1203x Series are PWM control type step-up DC/DC converter ICs with low supply current.

The R1203x is fully dedicated to drive White LEDs with constant current. Each of these ICs consists of an NMOS FET, an oscillator, a PWM comparator, a voltage reference unit, an error amplifier, a current limit circuit, an under voltage lockout circuit (UVLO), and an over-voltage protection circuit (OVP).

The R1203x can drive white LEDs in constant current with high efficiency by using an inductor, a diode, a resistor and capacitors as external components.

The LEDs current can be set by an external resistance value and can adjust the dimming of LEDs by CE pin according to the signal of PWM. Feedback voltage is 0.2V, therefore power loss by current setting resistance is small and efficiency is good. Maximum duty cycle is internally fixed, Typ. 91%. LEDs can be driven from low voltage. Protection circuits are the current limit of Lx peak current, the over voltage limit of output, and the under voltage lockout function.

It is controllable the dimming of LEDs quickly when the PWM signal (between 200Hz to 300kHz) input to CE pin. If the CE pin input is "L" in the fixed time (Typ. 0.5ms), the IC becomes the standby mode and turns OFF LEDs.

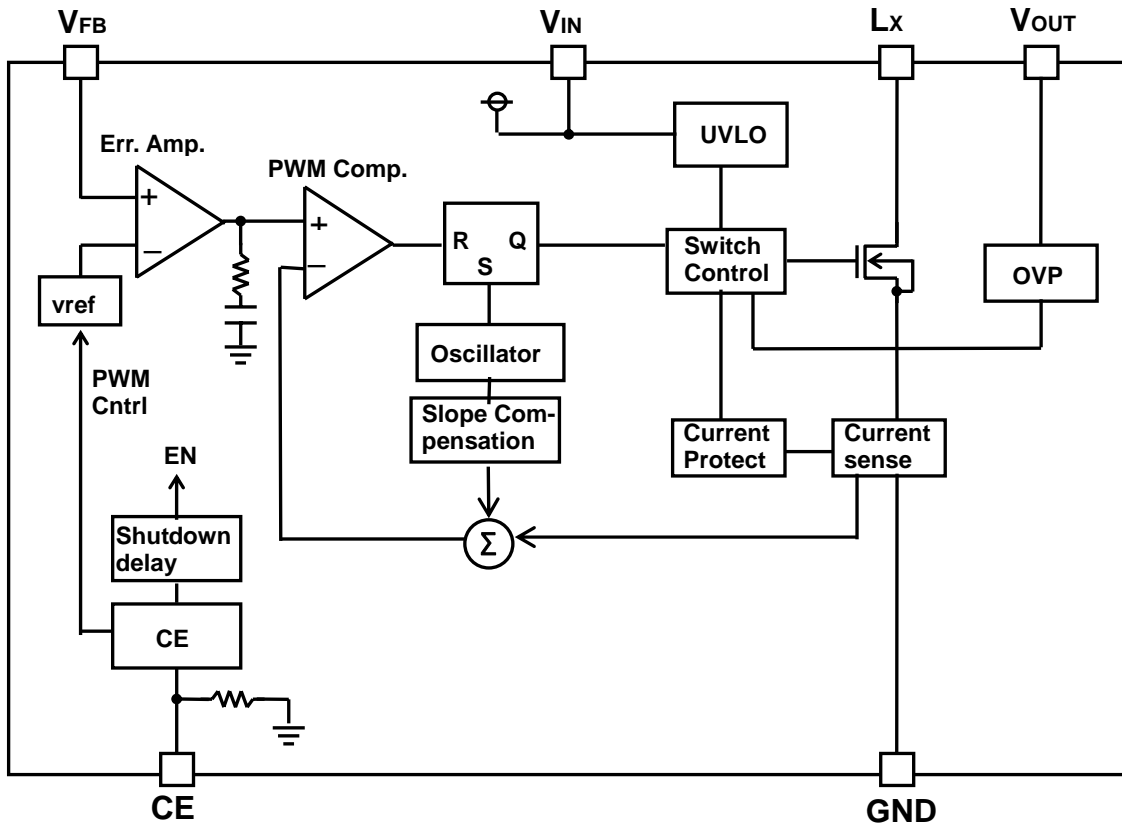
### FEATURES

- Supply Current ..... Typ. 500 $\mu$ A
- Standby Current ..... Max. 5 $\mu$ A
- Input Voltage Range ..... 1.8V to 5.5V
- Feedback Voltage ..... 0.2V
- Feedback Voltage Accuracy .....  $\pm 1.0\%$  ( $\pm 10$ mV)
- Temperature-Drift Coefficient of Feedback Voltage ...  $\pm 150$ ppm/ $^{\circ}$ C
- Oscillator Frequency ..... Typ. 1.2MHz
- Maximum Duty Cycle ..... Typ. 91%
- Switch ON Resistance ..... Typ. 1.35 $\Omega$
- UVLO Detector Threshold ..... Typ. 1.6V
- Lx Current Limit Protection ..... Typ. 700mA
- OVP Detector Threshold ..... Typ. 29.5V
- Switching Control ..... PWM
- LED dimming control ..... by external PWM signal (Frequency 200Hz to 300kHz)
- Packages ..... DFN1616-6B, SOT-23-6
- Ceramic capacitors are recommended ..... 0.22 $\mu$ F

### APPLICATION

- White LED Backlight for portable equipment

## BLOCK DIAGRAMS

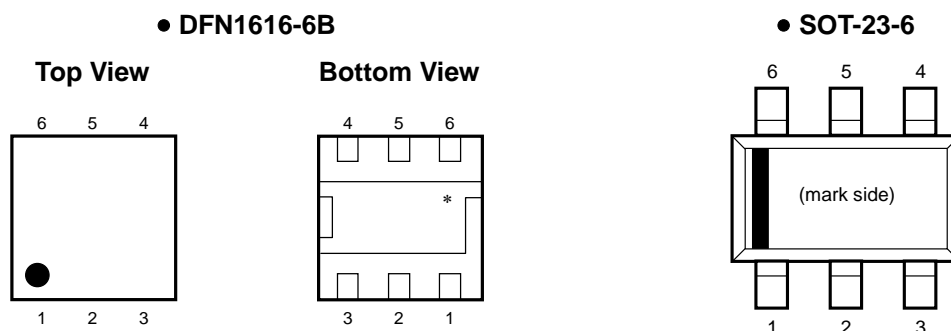


## SELECTION GUIDE

The package for the ICs can be selected at the user's request.

Product Name	Package	Quantity per Reel	Pb Free	Halogen Free
R1203L071B-TR	DFN1616-6B	5,000 pcs	Yes	Yes
R1203N071B-TR-FE	SOT-23-6	3,000 pcs	Yes	Yes

## PIN CONFIGURATIONS



## PIN DESCRIPTIONS

### • DFN1616-6B

Pin No	Symbol	Pin Description
1	CE	Chip Enable Pin ("H" Active)
2	$V_{FB}$	Feedback Pin
3	LX	Switching Pin (Open Drain Output)
4	GND	Ground Pin
5	$V_{IN}$	Input Pin
6	$V_{OUT}$	Output Pin

\*) Tab is GND level. (They are connected to the reverse side of this IC.)

The tab is better to be connected to the GND, but leaving it open is also acceptable.

### • SOT-23-6

Pin No	Symbol	Pin Description
1	CE	Chip Enable Pin ("H" Active)
2	$V_{OUT}$	Output Pin
3	$V_{IN}$	Input Pin
4	LX	Switching Pin (Open Drain Output)
5	GND	Ground Pin
6	$V_{FB}$	Feedback Pin

**ABSOLUTE MAXIMUM RATINGS**

(GND=0V)

Symbol	Item	Rating	Unit
$V_{IN}$	$V_{IN}$ Pin Voltage	-0.3 to 6.5	V
$V_{CE}$	CE Pin Voltage	-0.3 to $V_{IN}+0.3$	V
$V_{FB}$	$V_{FB}$ Pin Voltage	-0.3 to $V_{IN}+0.3$	V
$V_{OUT}$	$V_{OUT}$ Pin Voltage	-0.3 to 32	V
$V_{LX}$	$L_x$ Pin Voltage	-0.3 to 32	V
$I_{LX}$	$L_x$ Pin Current	1000	mA
$P_D$	Power Dissipation (DFN1616-6B)*	640	mW
	Power Dissipation (SOT-23-6)*	420	
$T_a$	Operating Temperature Range	-40 to 85	°C
$T_{stg}$	Storage Temperature Range	-55 to 125	°C

\*) For Power Dissipation, please refer to PACKAGE INFORMATION.

**ABSOLUTE MAXIMUM RATINGS**

Electronic and mechanical stress momentarily exceeded absolute maximum ratings may cause the permanent damages and may degrade the life time and safety for both device and system using the device in the field. The functional operation at or over these absolute maximum ratings is not assured.

## ELECTRICAL CHARACTERISTICS

### • R1203x

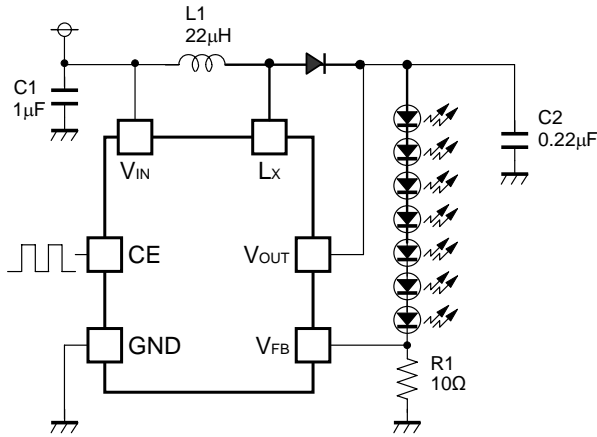
(Ta=25°C)

Symbol	Item	Conditions	Min.	Typ.	Max.	Unit
V <sub>IN</sub>	Operating Input Voltage		1.8		5.5	V
I <sub>DD</sub>	Supply Current	V <sub>IN</sub> =5.5V, V <sub>FB</sub> =0V, Lx at no load		0.5	1.0	mA
I <sub>standby</sub>	Standby Current	V <sub>IN</sub> =5.5V, V <sub>CE</sub> =0V		1.0	5.0	μA
V <sub>UVLO1</sub>	UVLO Detector Threshold	V <sub>IN</sub> falling	1.5	1.6	1.7	V
V <sub>UVLO2</sub>	UVLO Released Voltage	V <sub>IN</sub> rising		V <sub>UVLO1</sub> +0.1	1.8	V
V <sub>CEH</sub>	CE Input Voltage "H"	V <sub>IN</sub> =5.5V	1.5			V
V <sub>CEL</sub>	CE Input Voltage "L"	V <sub>IN</sub> =1.8V			0.5	V
R <sub>CE</sub>	CE Pull Down Resistance	V <sub>IN</sub> =3.6V	600	1200	2200	kΩ
V <sub>FB</sub>	V <sub>FB</sub> Voltage Accuracy	V <sub>IN</sub> =V <sub>CE</sub> =3.6V	0.19	0.20	0.21	V
$\frac{\Delta V_{FB}}{\Delta T_a}$	V <sub>FB</sub> Voltage Temperature Coefficient	V <sub>IN</sub> =V <sub>CE</sub> =3.6V, -40°C ≤ T <sub>a</sub> ≤ 85°C		±150		ppm/°C
I <sub>FB</sub>	V <sub>FB</sub> Input Current	V <sub>IN</sub> =5.5V, V <sub>FB</sub> =0V or V <sub>IN</sub>	-0.1		0.1	μA
R <sub>ON</sub>	Switch ON Resistance	V <sub>IN</sub> =3.6V, I <sub>LX</sub> =100mA		1.35		Ω
I <sub>LXleak</sub>	Switch Leakage Current	V <sub>LX</sub> =30V		0	3.0	μA
I <sub>LXlim</sub>	Switch Current Limit	V <sub>IN</sub> =3.6V	400	700	1000	mA
f <sub>osc</sub>	Oscillator Frequency	V <sub>IN</sub> =3.6V, V <sub>OUT</sub> =V <sub>FB</sub> =0V	1.0	1.2	1.4	MHz
Maxduty	Maximum Duty Cycle	V <sub>IN</sub> =3.6V, V <sub>OUT</sub> =V <sub>FB</sub> =0V	86	91		%
V <sub>OVP1</sub>	OVP Detector Threshold	V <sub>IN</sub> =3.6V, V <sub>OUT</sub> rising	28.7	29.5	30.3	V
$\frac{\Delta V_{OVP1}}{\Delta T_a}$	V <sub>OVP1</sub> Voltage Temperature Coefficient	V <sub>IN</sub> =V <sub>CE</sub> =3.6V, -40°C ≤ T <sub>a</sub> ≤ 85°C		±150		ppm/°C
V <sub>OVP2</sub>	OVP Released Voltage	V <sub>IN</sub> =3.6V, V <sub>OUT</sub> falling		V <sub>OVP1</sub> -1.55		V

### RECOMMENDED OPERATING CONDITIONS (ELECTRICAL CHARACTERISTICS)

All of electronic equipment should be designed that the mounted semiconductor devices operate within the recommended operating conditions. The semiconductor devices cannot operate normally over the recommended operating conditions, even if when they are used over such conditions by momentary electronic noise or surge. And the semiconductor devices may receive serious damage when they continue to operate over the recommended operating conditions.

## TYPICAL APPLICATIONS



C1	CM105B105K06
C2	GRM21BR71H224
L1	LQH32CN220K53

### • LED Current setting

When CE pin input is "H" (Duty=100%), LED current can be set with feedback resistor (R1)

$$I_{LED} = V_{FB} / R1$$

### • LED Dimming Control

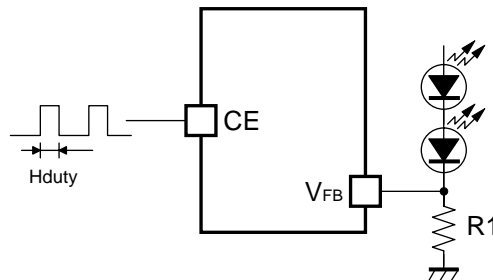
The LED brightness can be controlled by inputting the PWM signal to the CE pin. If the CE pin input is "L" in the fixed time (Typ.0.5ms), the IC becomes the standby mode and turns OFF LEDs.

The current of LEDs when the CE pin is "H" input (Duty=100%) is shown by the above expression. The current of LEDs can be controlled by Duty of the PWM signal of the input CE pin. The current of LEDs when High-Duty of the CE input is Hduty reaches the value as calculatable following formula.

$$I_{LED} = Hduty \times V_{FB} / R1$$

The frequency of the PWM signal is using the range between 200Hz to 300kHz.

When controlling the LED brightness by the PWM signal of 20kHz or less; The increasing or decreasing of the inductor current might be make a sounds in the hearable sound wave area. In that case, please use the PWM signal in the high frequency area.



Dimming control by CE pin input

- **Soft-Start**

The output of the error amplifier starts from 0V and the inrush current is suppressed when starting by the CE pin "H" input.

Moreover, the inrush current can be suppressed by gradually enlarging Duty of the PWM signal to the CE pin.

- **Selection of Inductors**

The peak current of the inductor at normal mode can be calculated as next formula:

$$I_{Lmax}=1.25 \times I_{LED} \times V_{OUT} / V_{IN} + 0.5 \times V_{IN} \times (V_{OUT} - V_{IN}) / (L \times V_{OUT} \times f_{osc})$$

When the start-up or dimming control by CE pin, transient current flows, the peak current must be equal or less than the current limit of the IC. The peak current should not beyond the rating current of the inductor.

When 4-7LEDs are driven with  $V_{IN}=3.6V$ , the recommended inductance value is 10 $\mu$ H -22 $\mu$ H.

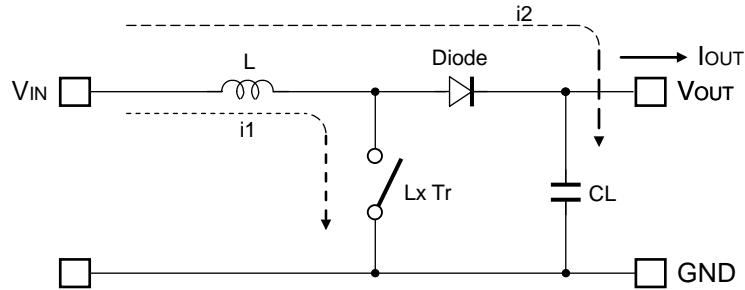
- **Selection of Capacitors**

Set 1 $\mu$ F or more value bypass capacitor C1 between  $V_{IN}$  pin and GND pin as close as possible.

Set 0.22 $\mu$ F or more capacitor C2 between  $V_{OUT}$  pin and GND pin.

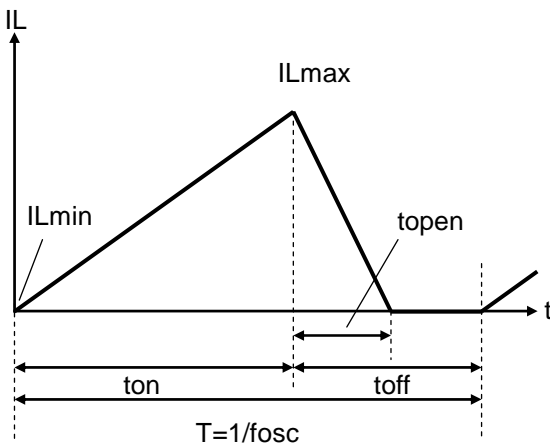
**OPERATION OF STEP-UP DC/DC CONVERTER AND OUTPUT CURRENT**

<Basic Circuit>

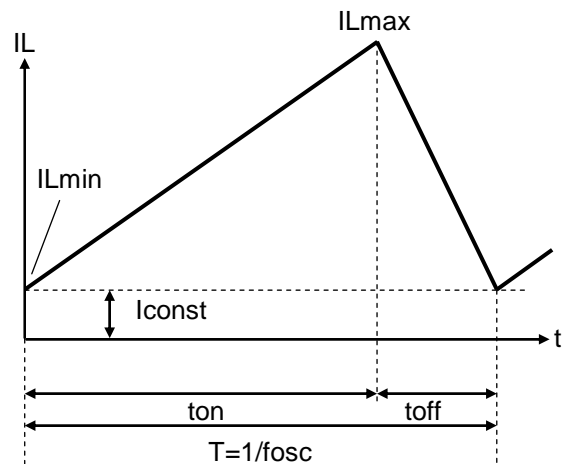


<Current through L>

Discontinuous mode



Continuous mode



There are two operation modes of the step-up PWM control-DC/DC converter. That is the continuous mode and discontinuous mode by the continuousness inductor.

When the transistor turns ON, the voltage of inductor L becomes equal to  $V_{IN}$  voltage. The increase value of inductor current ( $i1$ ) will be

$$\Delta i1 = V_{IN} \times t_{on} / L \dots\dots\dots \text{Formula 1}$$

As the step-up circuit, during the OFF time (when the transistor turns OFF) the voltage is continually supply from the power supply. The decrease value of inductor current ( $i2$ ) will be

$$\Delta i2 = (V_{OUT} - V_{IN}) \times t_{open} / L \dots\dots\dots \text{Formula 2}$$

At the PWM control-method, the inductor current become continuously when  $t_{open} = t_{off}$ , the DC/DC converter operate as the continuous mode.



In the continuous mode, the variation of current of  $i_1$  and  $i_2$  is same at regular condition.

$$V_{IN} \times t_{on} / L = (V_{OUT} - V_{IN}) \times t_{off} / L \dots\dots\dots \text{Formula 3}$$

The duty at continuous mode will be

$$\text{Duty} = t_{on} / (t_{on} + t_{off}) = (V_{OUT} - V_{IN}) / V_{OUT} \dots\dots\dots \text{Formula 4}$$

The average of inductor current at  $t_{open}=t_{off}$  will be

$$I_{L(\text{Ave.})} = V_{IN} \times t_{on} / (2 \times L) \dots\dots\dots \text{Formula 5}$$

If the input power is equal to the output power, the  $I_{OUT}$  will be

$$I_{OUT} = V_{IN}^2 \times t_{on} / (2 \times L \times V_{OUT}) \dots\dots\dots \text{Formula 6}$$

If the  $I_{OUT}$  value is large than above the calculated value (Formula 6), it will become the continuous mode, At this status, the peak current ( $I_{L\text{max}}$ ) of inductor will be

$$I_{L\text{max}} = I_{OUT} \times V_{OUT} / V_{IN} + V_{IN} \times t_{on} / (2 \times L) \dots\dots\dots \text{Formula 7}$$

$$I_{L\text{max}} = I_{OUT} \times V_{OUT} / V_{IN} + V_{IN} \times T \times (V_{OUT} - V_{IN}) / (2 \times L \times V_{OUT}) \dots\dots\dots \text{Formula 8}$$

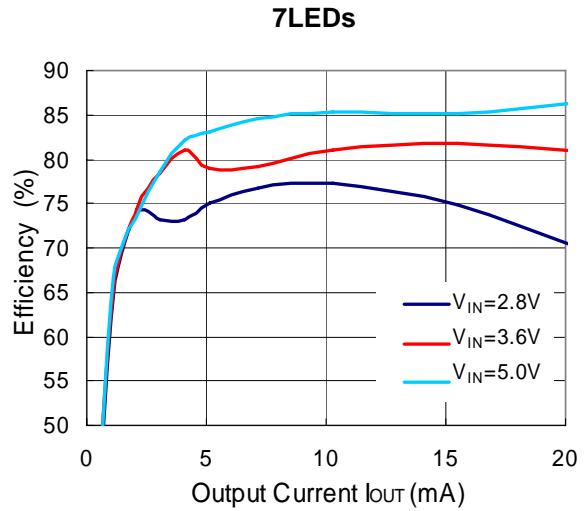
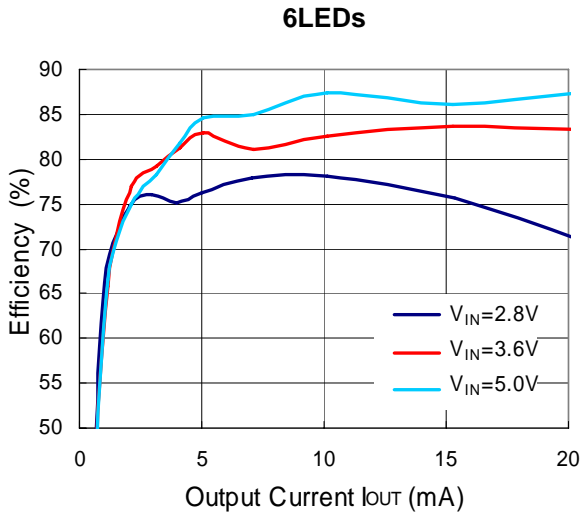
The peak current value is larger than the  $I_{OUT}$  value. In case of this, selecting the condition of the input and the output and the external components by considering of  $I_{L\text{max}}$  value.

The explanation above is based on the ideal calculation, and the loss caused by  $L_x$  switch and the external components are not included.

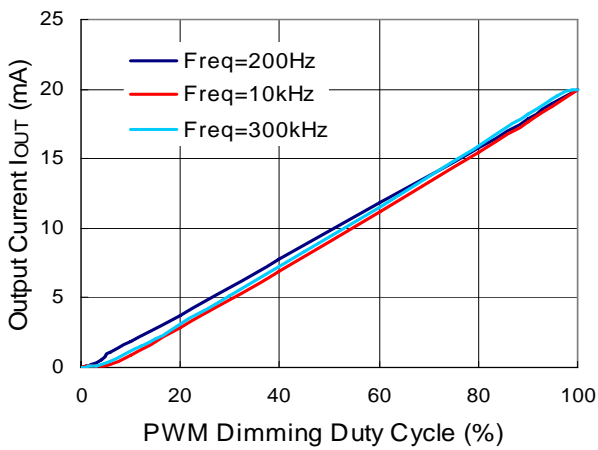
The actual maximum output current will be between 50% and 80% by the above calculations. Especially, when the  $I_L$  is large or  $V_{IN}$  is low, the loss of  $V_{IN}$  is generated with on resistance of the switch. Moreover, it is necessary to consider  $V_F$  of the diode (approximately 0.8V) about  $V_{OUT}$ .

## TYPICAL CHARACTERISTICS

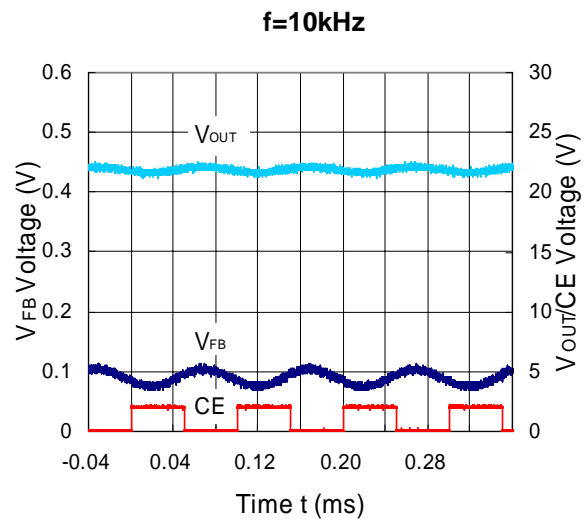
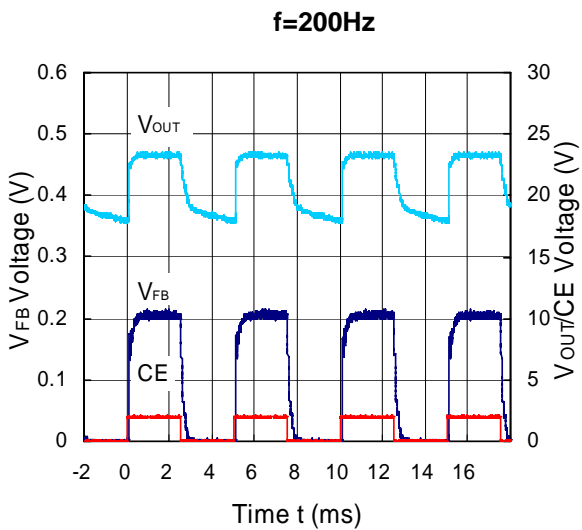
### 1) Efficiency vs. Output Current Characteristics



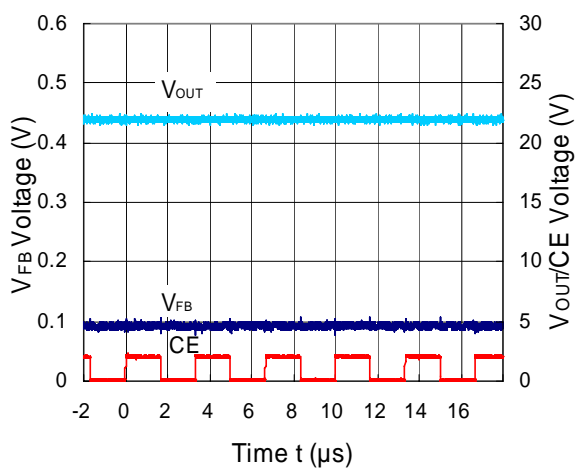
### 2) PWM Dimming Duty Cycle vs. Output Current( $R1=10\Omega$ )



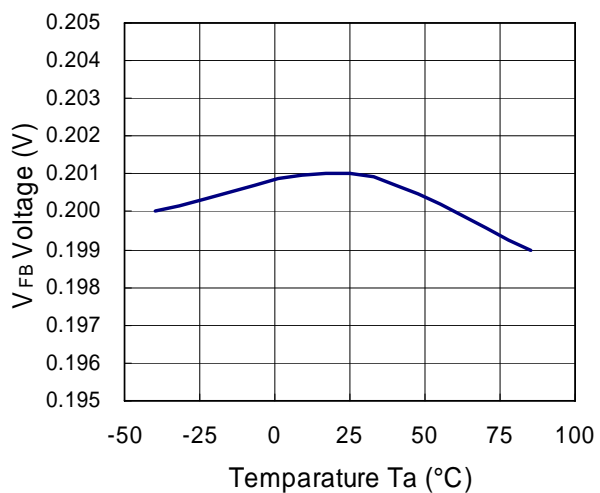
### 3) Output Current Ripple during PWM Dimming



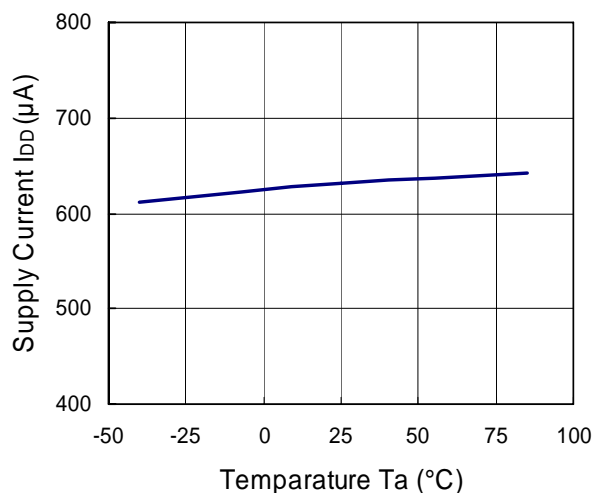
f=300kHz



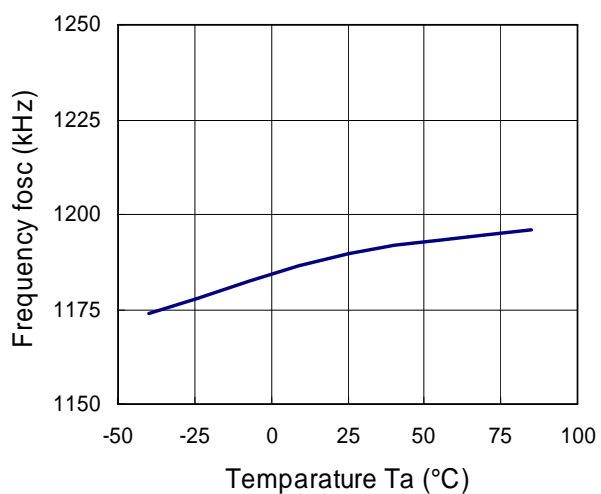
4) VFB Voltage vs. Temperature



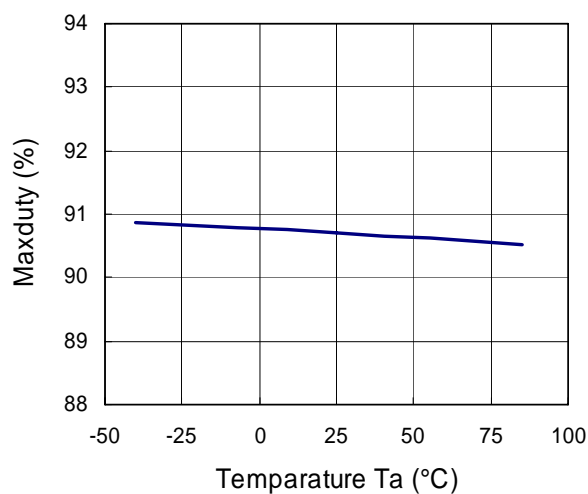
5) Supply Current vs. Temperature



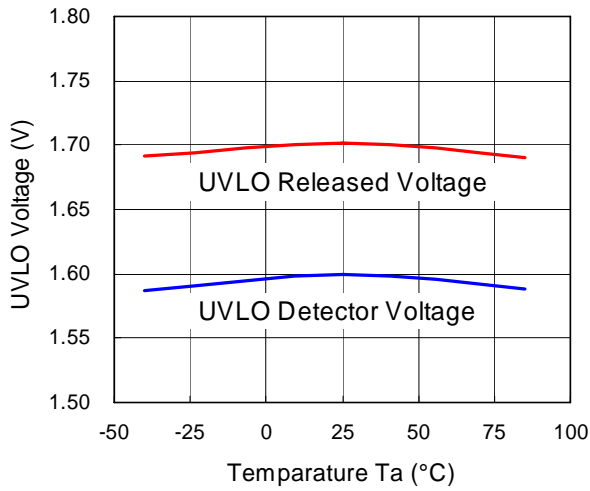
6) Oscillator Frequency vs. Temperature



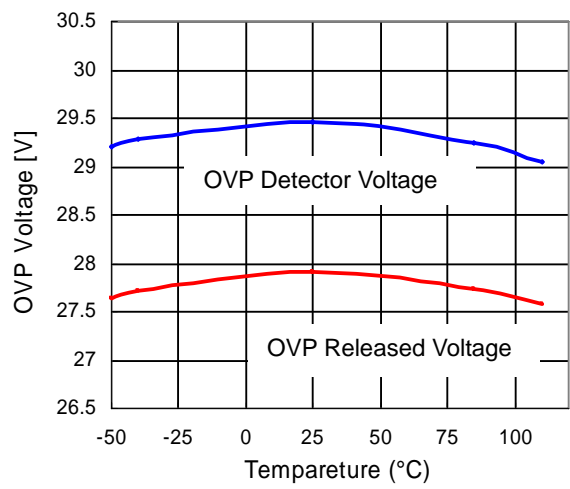
7) Maxduty vs. Temperature



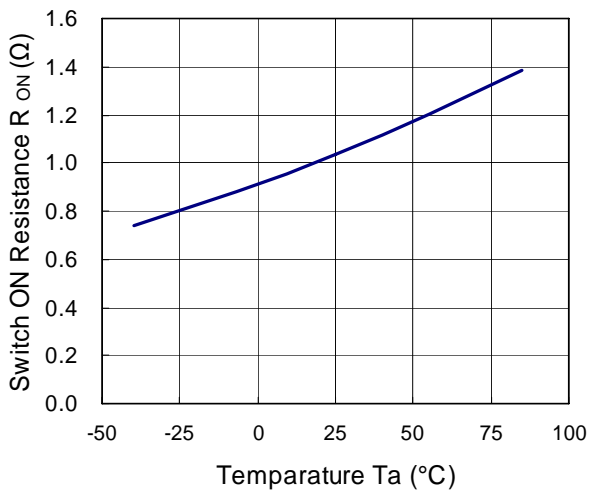
8) UVLO Output Voltage vs. Temperature



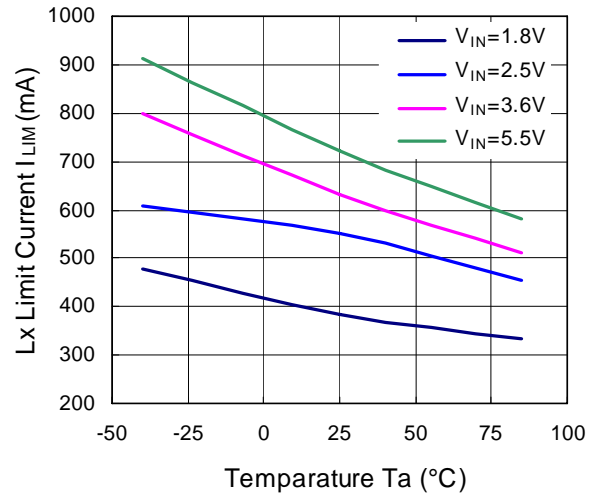
9) OVP Voltage vs. Temperature



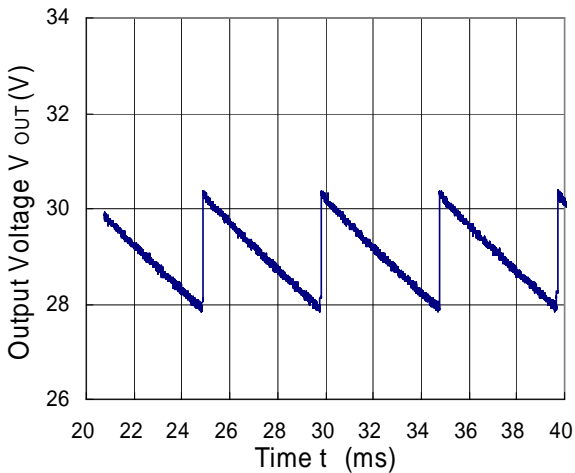
10) Switch ON Resistance vs. Temperature



11) Lx Current Limit vs. Temperature



12) OVP Operating Output Voltage Waveform





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